

transformation given by

$$s = \frac{2}{T} \frac{z - 1}{z + 1}. \quad (5.26)$$

Tables 5.2–5.5 contain the coefficients of the second-order transfer function

$$H(z) = \frac{a_0 + a_1 z^{-1} + a_2 z^{-2}}{1 + b_1 z^{-1} + b_2 z^{-2}}, \quad (5.27)$$

which are determined by the bilinear transformation and the auxiliary variable  $K = \tan(\omega_c T/2)$  for all audio filter types discussed. Further filter designs of peak and shelving filters are discussed in [Moo83, Whi86, Sha92, Bri94, Orf96a, Dat97, Cla00]. A method for reducing the warping effect of the bilinear transform is proposed in [Orf96b]. Strategies for time-variant switching of audio filters can be found in [Rab88, Mou90, Zöl93, Din95, Vål98].

**Table 5.2** Low-pass/high-pass/band-pass filter design.

| Low-pass (second order)                       |                                    |  |   |   |
|---|------------------------------------|--|---|---|
| $a_0$   | $a_1$                              | $a_2$  | $b_1$                                       | $b_2$   |
| $\frac{K^2}{1 + \sqrt{2}K + K^2}$             | $\frac{2K^2}{1 + \sqrt{2}K + K^2}$ | $\frac{K^2}{1 + \sqrt{2}K + K^2}$              | $\frac{2(K^2 - 1)}{1 + \sqrt{2}K + K^2}$    | $\frac{1 - \sqrt{2}K + K^2}{1 + \sqrt{2}K + K^2}$       |
| High-pass (second order)                      |                                    |  |   |   |
| $a_0$   | $a_1$                              | $a_2$  | $b_1$                                       | $b_2$   |
| $\frac{1}{1 + \sqrt{2}K + K^2}$               | $\frac{-2}{1 + \sqrt{2}K + K^2}$   | $\frac{1}{1 + \sqrt{2}K + K^2}$                | $\frac{2(K^2 - 1)}{1 + \sqrt{2}K + K^2}$    | $\frac{1 - \sqrt{2}K + K^2}{1 + \sqrt{2}K + K^2}$       |
| Band-pass (second order)                      |                                    |  |   |   |
| $a_0$   | $a_1$                              | $a_2$  | $b_1$                                       | $b_2$   |
| $\frac{\frac{1}{Q}K}{1 + \frac{1}{Q}K + K^2}$ | 0                                  | $-\frac{\frac{1}{Q}K}{1 + \frac{1}{Q}K + K^2}$ | $\frac{2(K^2 - 1)}{1 + \frac{1}{Q}K + K^2}$ | $\frac{1 - \frac{1}{Q}K + K^2}{1 + \frac{1}{Q}K + K^2}$ |

### 5.2.2 Parametric Filter Structures

Parametric filter structures allow direct access to the parameters of the transfer function, like center/cutoff frequency, bandwidth and gain, via control of associated coefficients. To modify one of these parameters, it is therefore not necessary to compute a complete set of coefficients for a second-order transfer function, but instead only one coefficient in the filter structure is calculated.

Independent control of gain, cutoff/center frequency and bandwidth for shelving and peak filters is achieved by a feed forward (FF) structure for *boost* and a feed backward