
Beamforming, Beam Rotation and Angle of Arrival

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DFT Properties

$$z_m = \sum_{n=0}^{N-1} x_n e^{\frac{-j2\pi \cdot m \cdot n}{N}}$$

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$$

Linearity property: $x_1(t) \leftrightarrow X_1(f)$, $x_2(t) \leftrightarrow X_2(f) \equiv a_1 x_1(t) + a_2 x_2(t) \leftrightarrow a_1 X_1(f) + a_2 X_2(f)$

$$\begin{aligned} a_1 x_1(t) + a_2 x_2(t) &\leftrightarrow \int_{-\infty}^{\infty} \{a_1 x_1(t) + a_2 x_2(t)\} e^{-j2\pi f t} dt \\ &= \int_{-\infty}^{\infty} a_1 x_1(t) e^{-j2\pi f t} dt + \int_{-\infty}^{\infty} a_2 x_2(t) e^{-j2\pi f t} dt = a_1 X_1(f) + a_2 X_2(f) \end{aligned}$$

DFT Properties

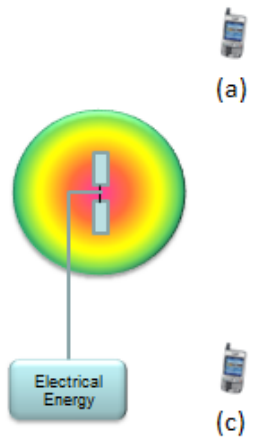
$$z_m = \sum_{n=0}^{N-1} x_n e^{\frac{-j2\pi \cdot m \cdot n}{N}}$$

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$$

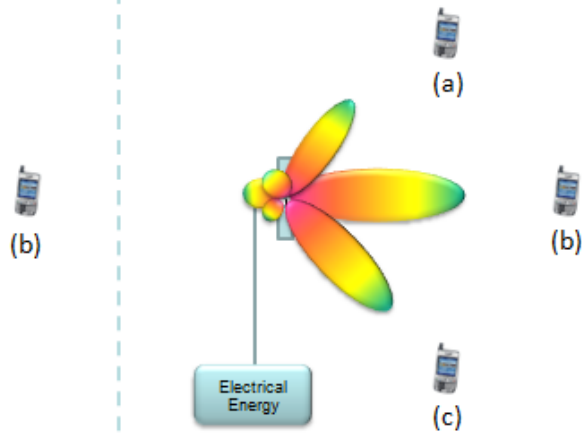
Time shifting property: $x(t) \leftrightarrow X(f) \equiv x(t - t_0) \leftrightarrow e^{-j2\pi f t_0} X(f)$

$$\begin{aligned} x(t - t_0) &\leftrightarrow \int_{-\infty}^{\infty} x(t - t_0) e^{-j2\pi f t} dt && t - t_0 = \tau \quad \therefore t = \tau + t_0 \quad \therefore dt = d\tau \\ &= \int_{-\infty}^{\infty} x(\tau) e^{-j2\pi f (\tau + t_0)} d\tau = e^{-j2\pi f t_0} \int_{-\infty}^{\infty} x(\tau) e^{-j2\pi f \tau} d\tau = e^{-j2\pi f t_0} X(f) \end{aligned}$$

Beamforming Examples



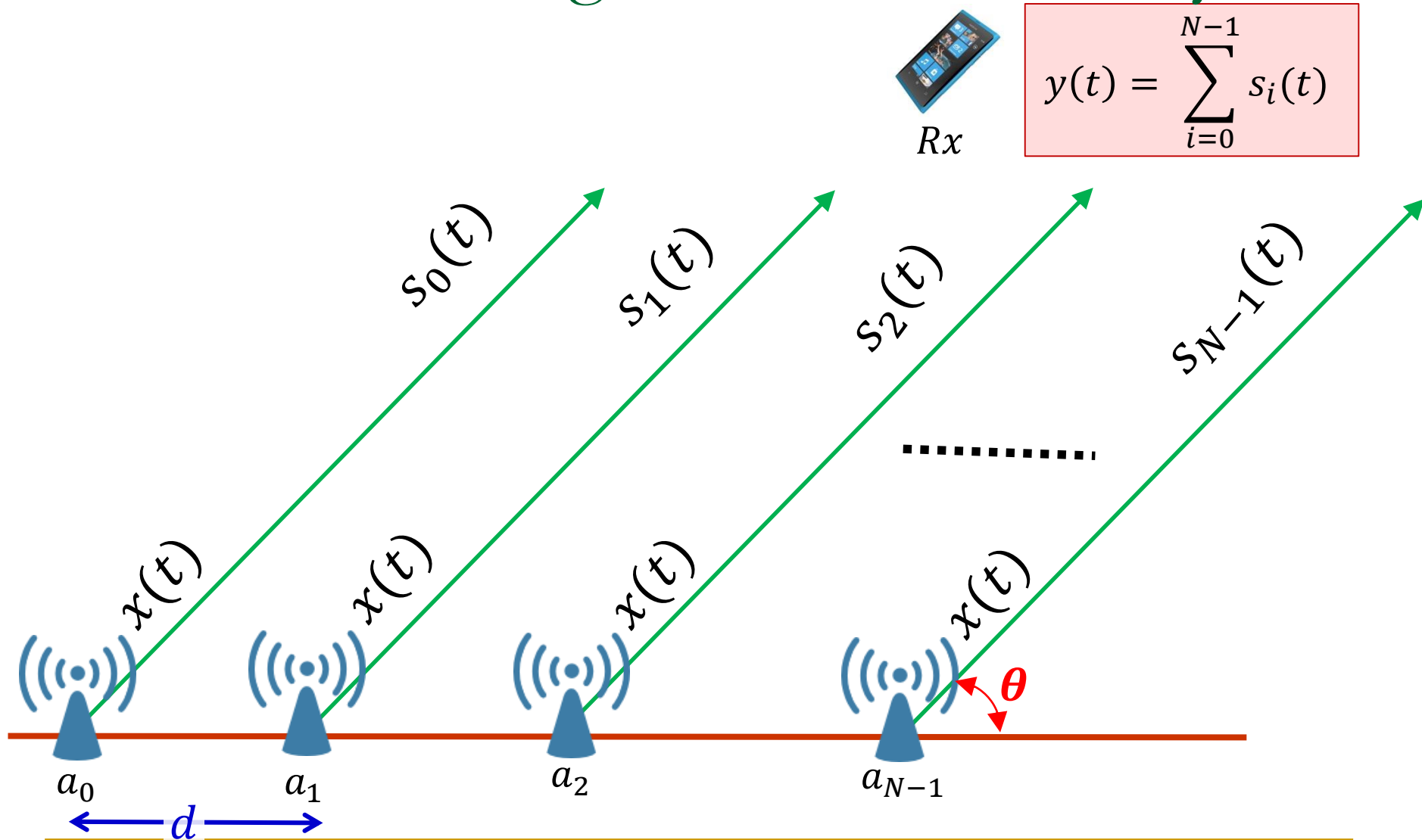
< Case 1 >



< Case 2 >



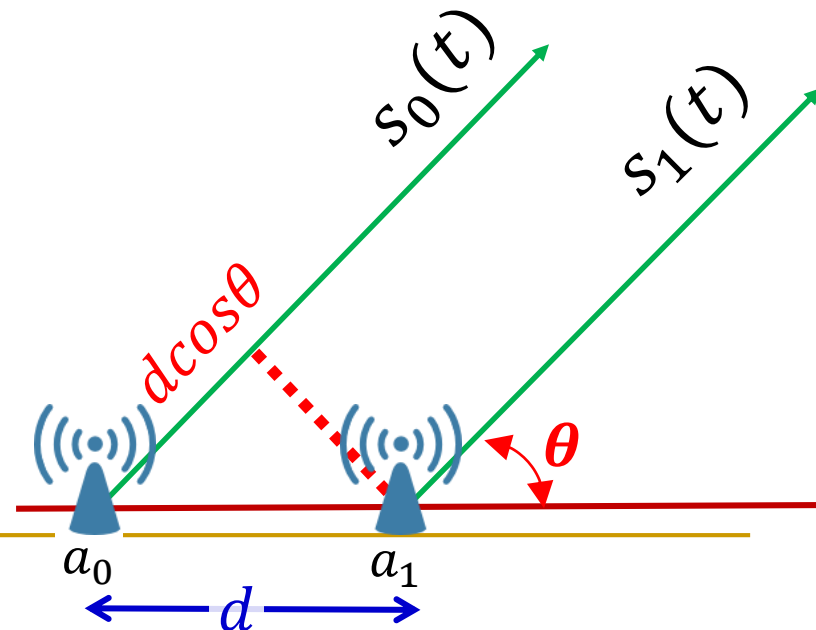
Beamforming and Antenna Array



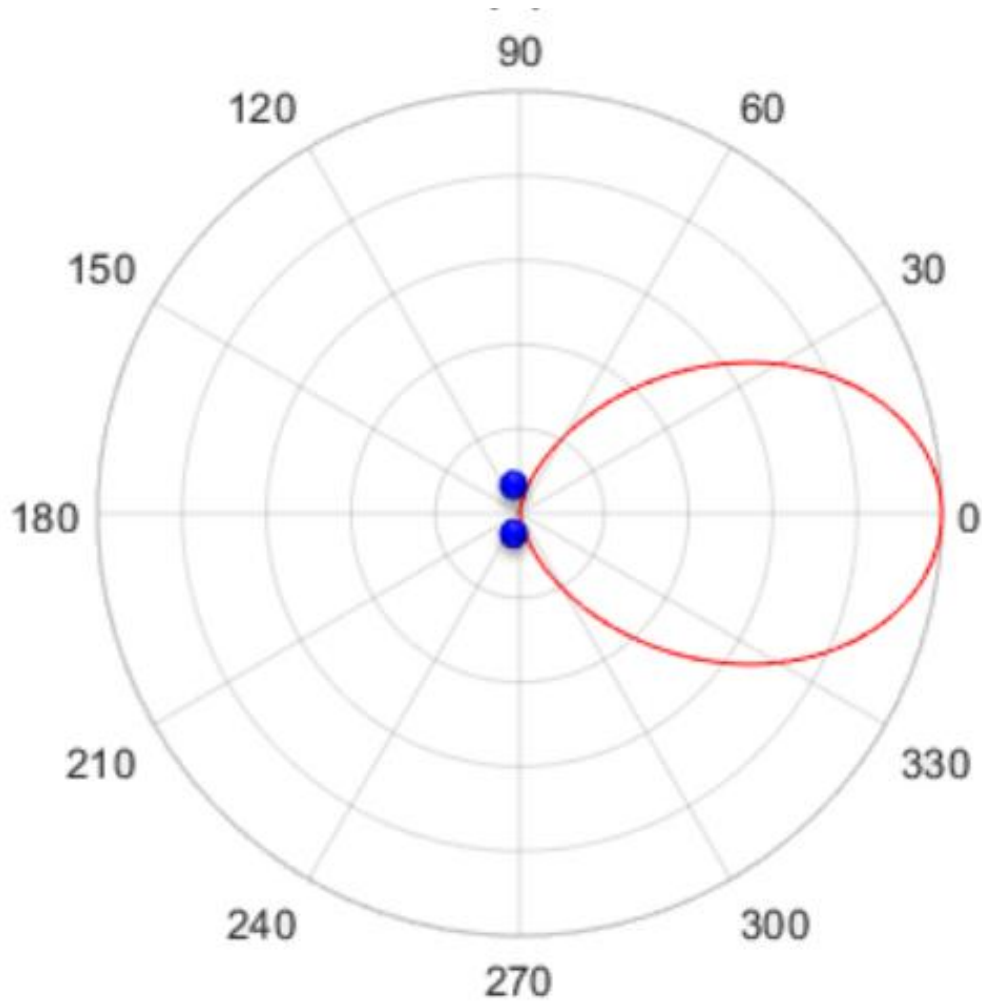
Beamforming and Antenna Array

- $s_1(t)$ travels $d\cos\theta$ less than $s_0(t)$
 - How much phase shift ϕ does it cause?
- λ distance causes a phase shift of 2π
 - $d\cos\theta$ distance causes a phase shift ϕ of $\frac{2\pi}{\lambda} d\cos\theta$
- Assume $s_0(t) = \cos(2\pi f_0 t)$, then $s_1(t) = \cos(2\pi f_0 t + \phi)$
 - $S_1(f) = S_0(f)e^{j\phi}$
 - $S_2(f) = S_0(f)e^{2j\phi}$
 - $S_3(f) = S_0(f)e^{3j\phi}$
 - $S_{N-1}(f) = S_0(f)e^{(N-1)j\phi}$

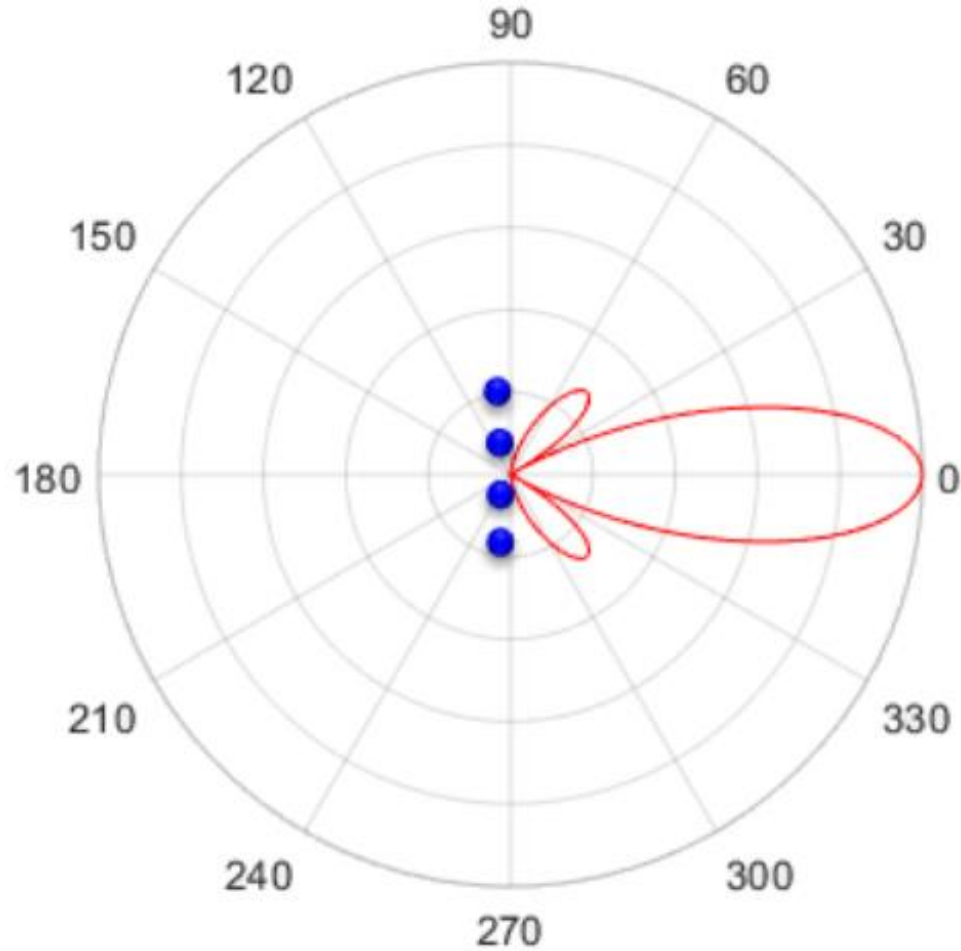
$$Y(f) = \sum_{k=0}^{N-1} S_0(f)e^{kj\phi} = S_0(f) \left(\frac{1 - e^{jN\phi}}{1 - e^{j\phi}} \right)$$



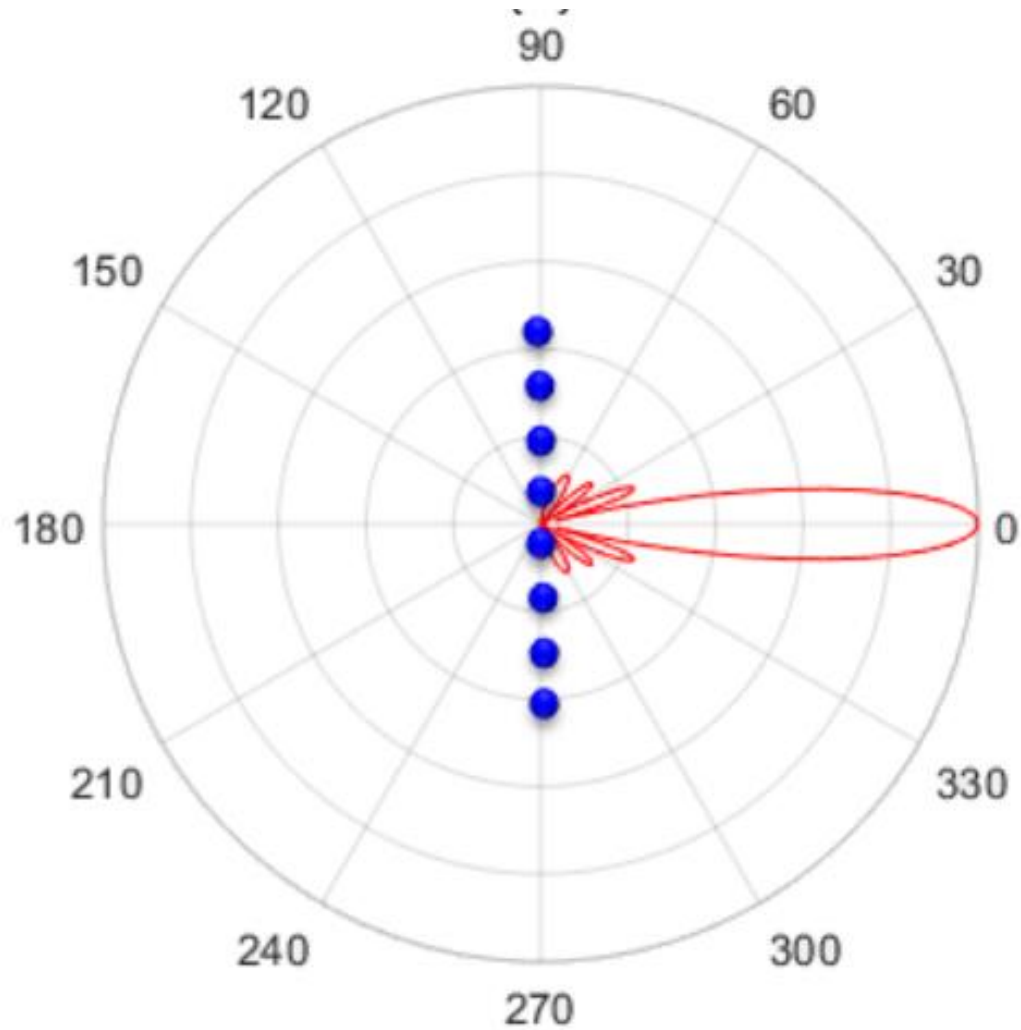
Beamforming ($N=2$)



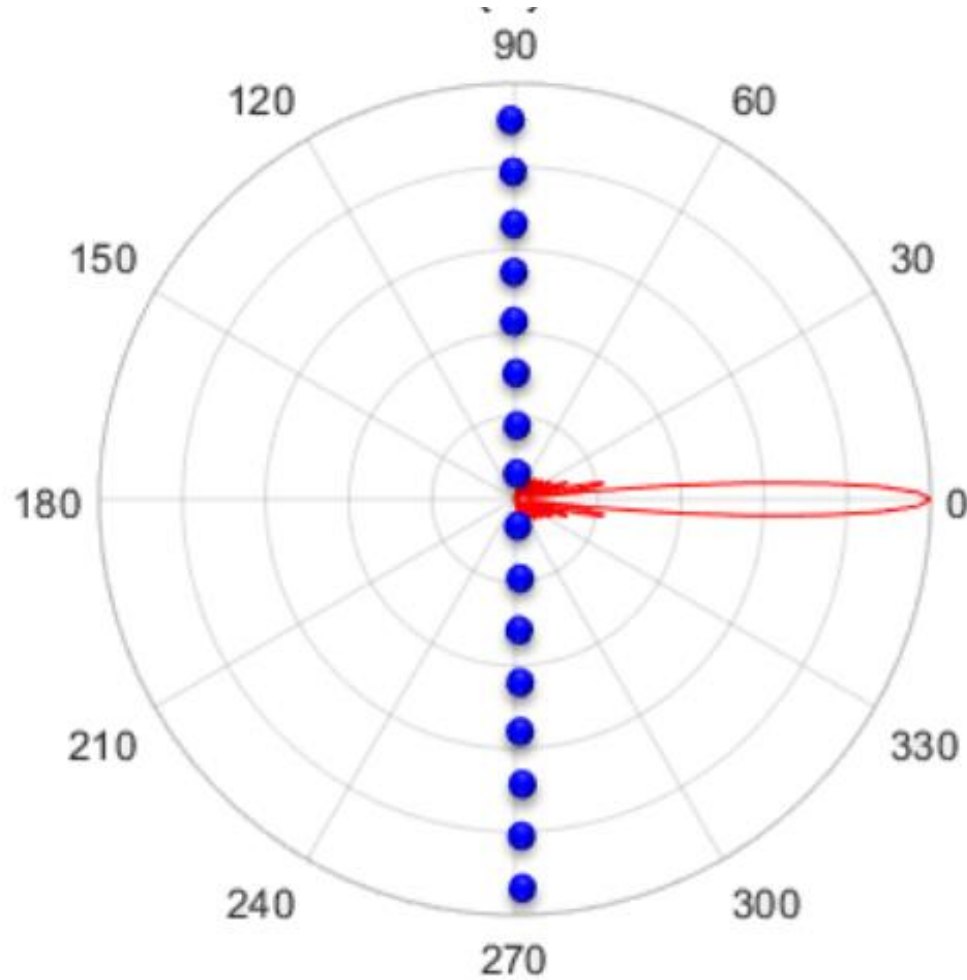
Beamforming ($N=4$)



Beamforming ($N=8$)



Beamforming ($N=16$)



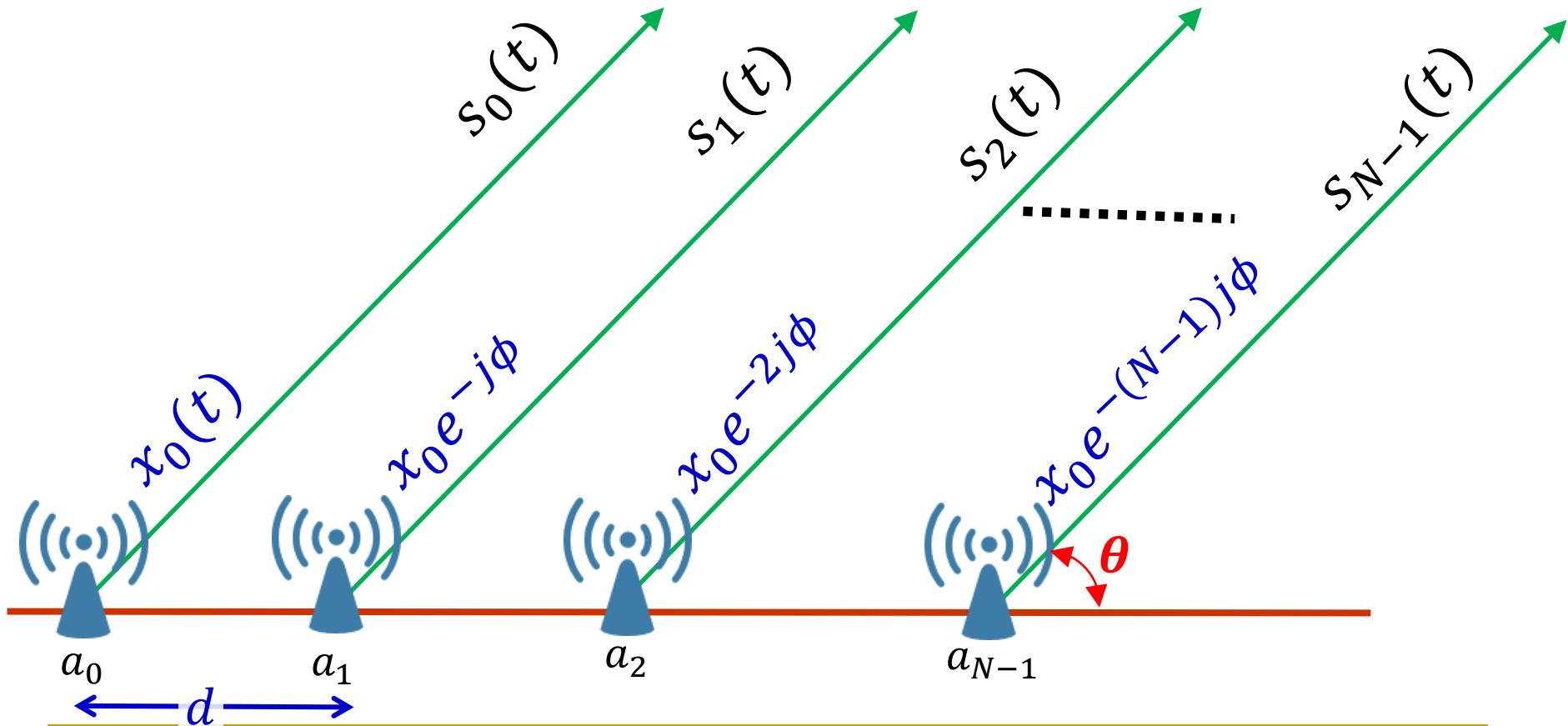
Beam Rotation



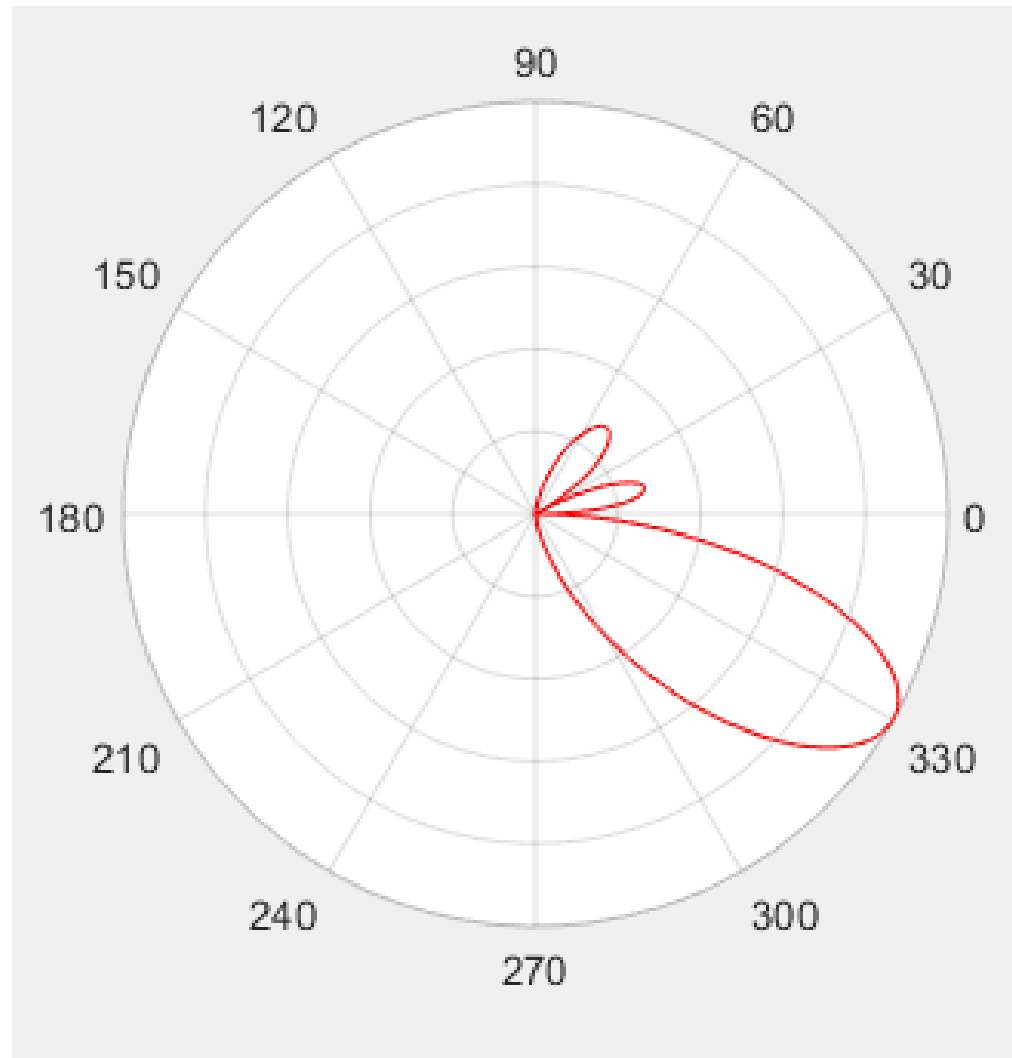
Rx

$$y(t) = \sum_{i=0}^{N-1} s_i(t)$$

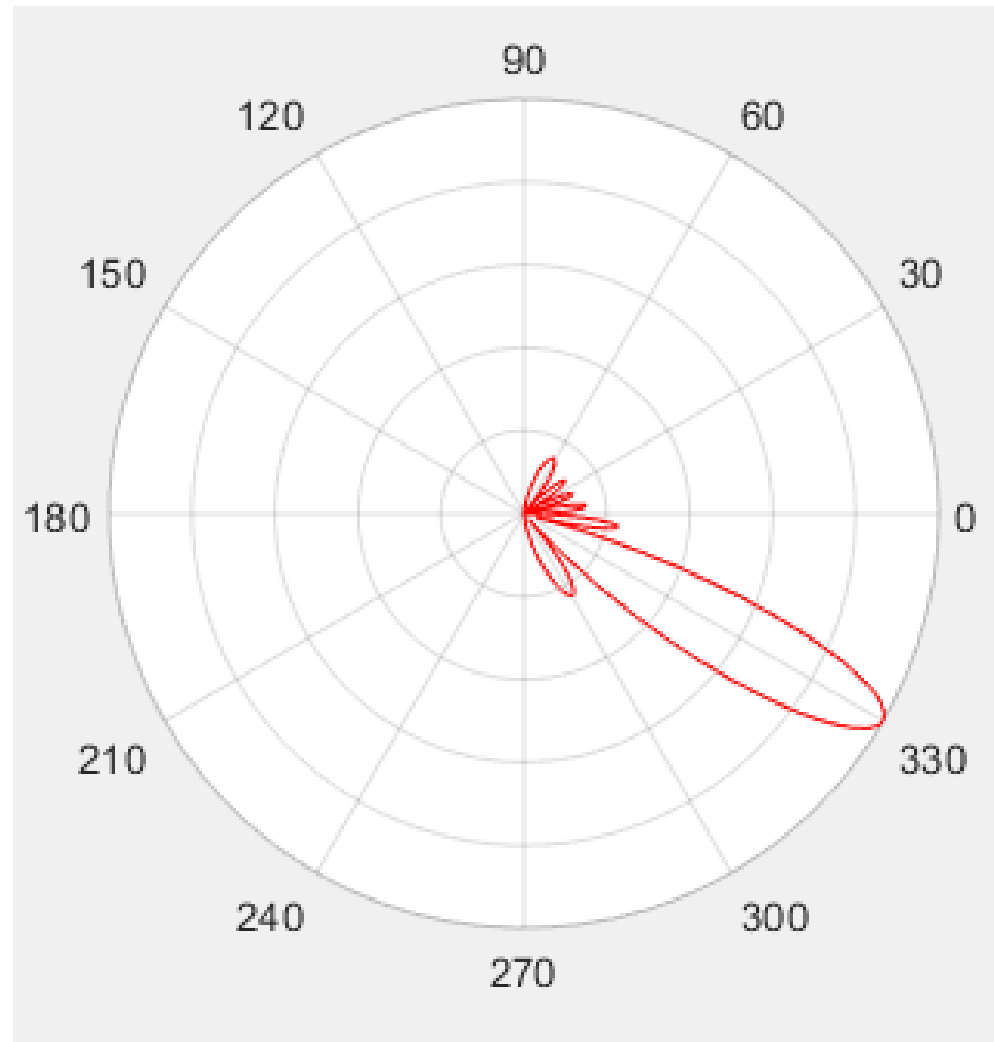
$$Y(f) = \sum_{k=0}^{N-1} (S_0(f)e^{-kj\phi})e^{kj\phi} = \sum_{k=0}^{N-1} S_0(f) = NS_0(f)$$



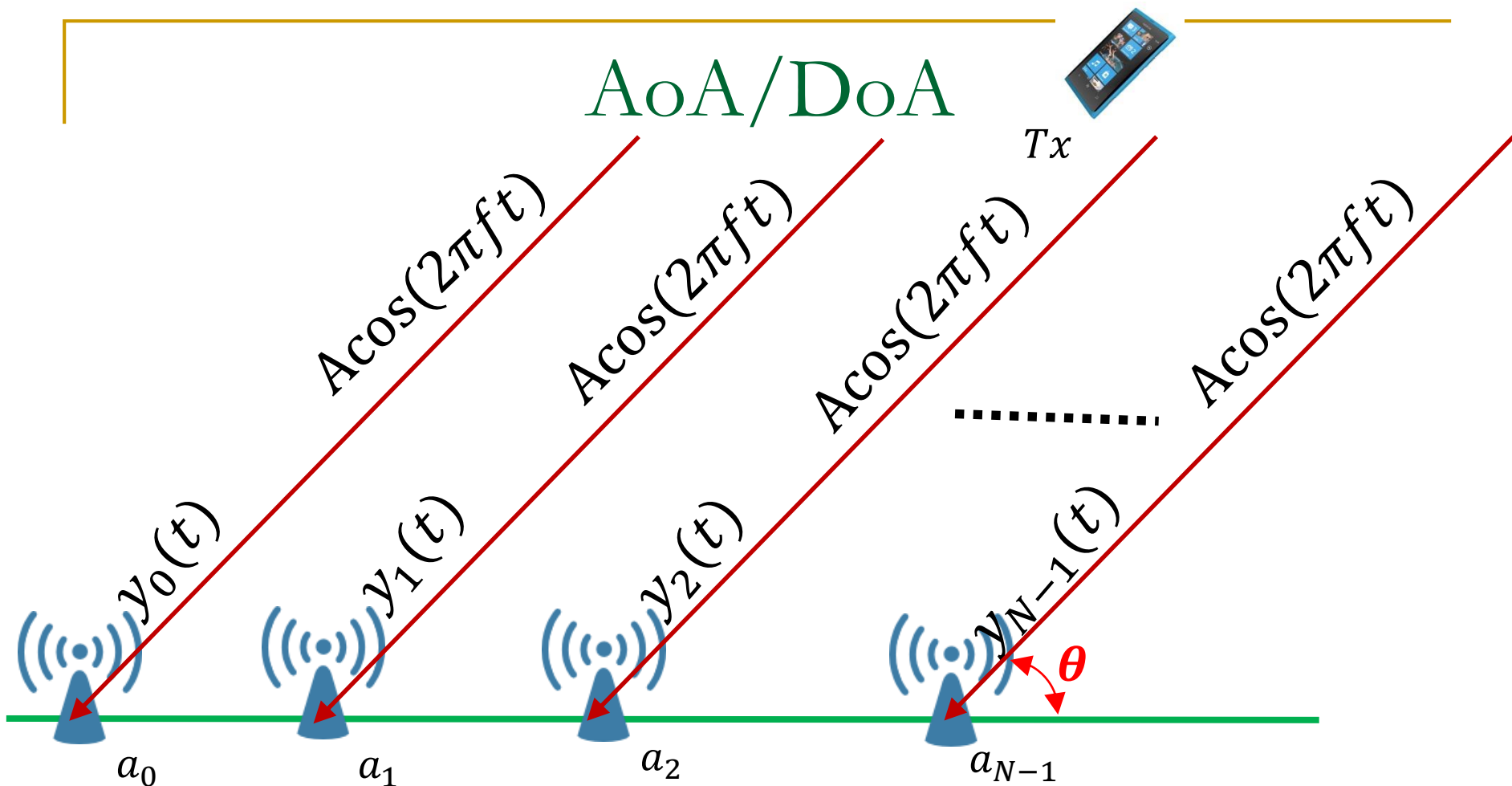
Beam Rotation ($N=4$)



Beam Rotation ($N=8$)



AoA/DoA



$$y(t) = \begin{bmatrix} y_0 \\ y_1 \\ y_2 \\ \vdots \\ y_{N-1} \end{bmatrix} = \begin{bmatrix} B\cos(2\pi ft) \\ B\cos(2\pi ft + \phi) \\ B\cos(2\pi ft + 2\phi) \\ \vdots \\ B\cos(2\pi ft + (N-1)\phi) \end{bmatrix} \leftrightarrow \begin{bmatrix} S(f)e^{j0} \\ S(f)e^{j\phi} \\ S(f)e^{j2\phi} \\ \vdots \\ S(f)e^{j(N-1)\phi} \end{bmatrix}$$

AoA/DoA

For $\theta_i = -\pi$ to π {

$$\alpha_i = \frac{2\pi}{\lambda} d \cos \theta_i;$$

$$C_{\theta_i} = [e^{-j0} \quad e^{-j\alpha_i} \quad e^{-j2\alpha_i} \quad \dots \quad e^{-j(N-1)\alpha_i}]. \begin{bmatrix} y_0 \\ y_1 \\ y_2 \\ \dots \\ y_{N-1} \end{bmatrix}$$

}

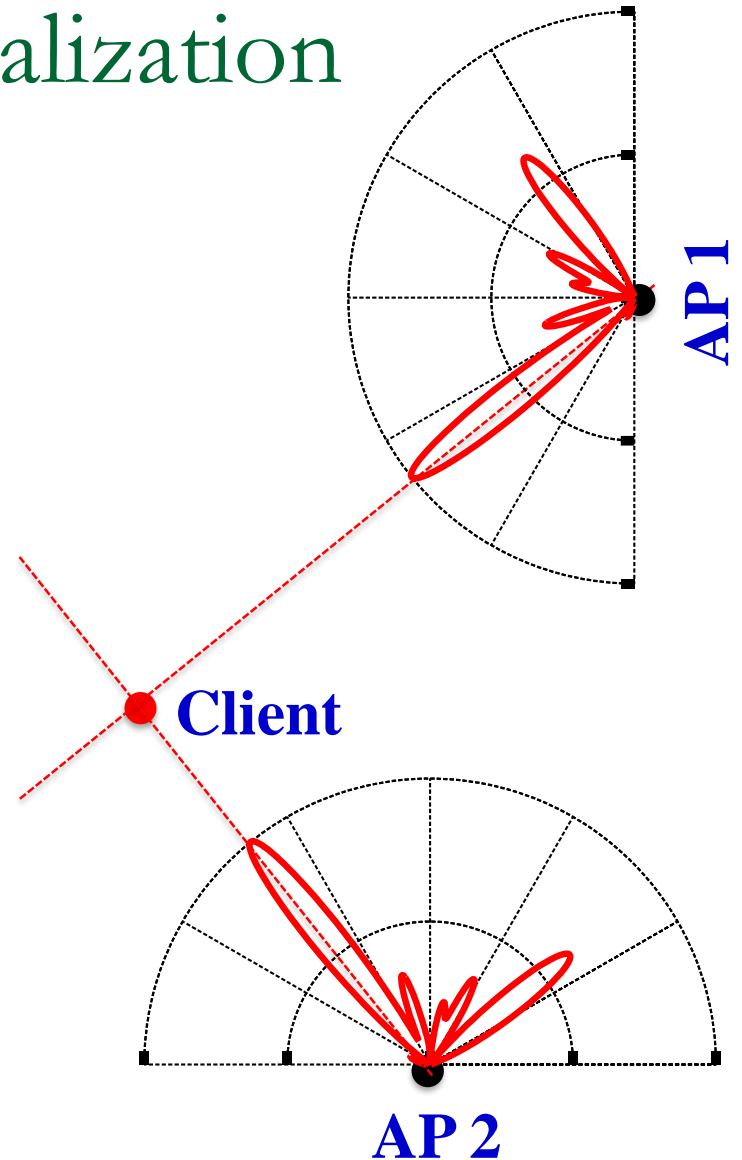
Return $\operatorname{argmax}_{\theta_i} C_{\theta_i};$

Beamforming and Antenna Array



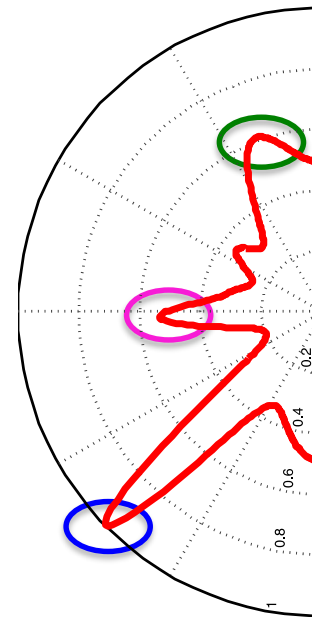
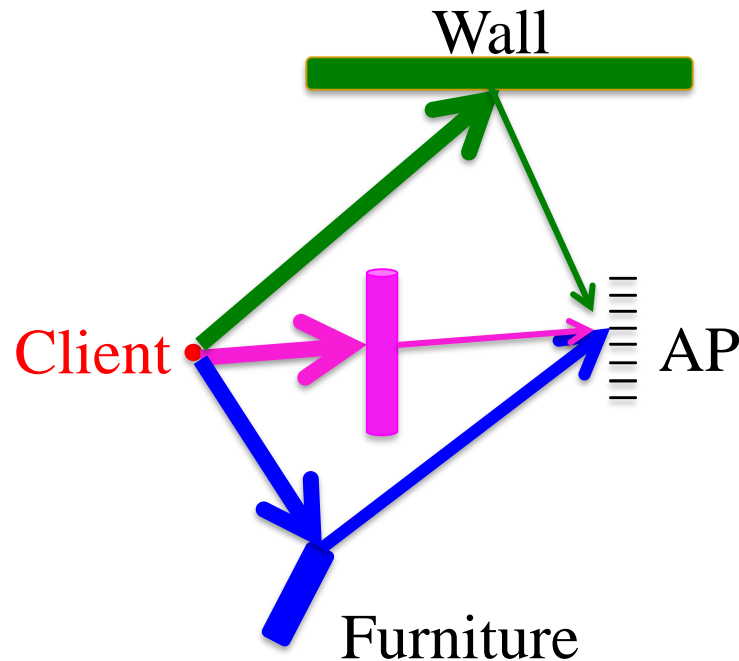
AoA Based Localization

- AP overhears a client's transmission
- APs **leverage multiple antennas** to generate angles of arrival of a client's signals:
- With multiple APs, central server **synthesizes** AoA spectra to obtain a location estimate for the client



AoA Based Localization

- **Problem #1:** Strong multipath reflections indoors
- **Problem #2:** **Direct path** attenuated or completely blocked
 - Direct path signal may not be the strongest



AoA Based Localization

- Direct path is more stable than reflection paths when client moves slightly

**Median: 23 cm
(with 8 APs)**

