

MODULE-1 : DISCRETE FOURIER TRANSFORMS (DFT)

To perform frequency analysis on a discrete time sequence $x(n)$, we convert time-domain sequence to an equivalent frequency domain representation.

Applying Fourier Transform on $x(n)$, we get $X(\omega)$, which is continuous and periodic function of frequency. It is not a computationally convenient representation of the sequence $x(n)$.

Representation of a sequence $x(n)$ by samples of its spectrum $X(\omega)$ is known as the Discrete Fourier Transform (DFT).

FREQUENCY DOMAIN SAMPLING :-

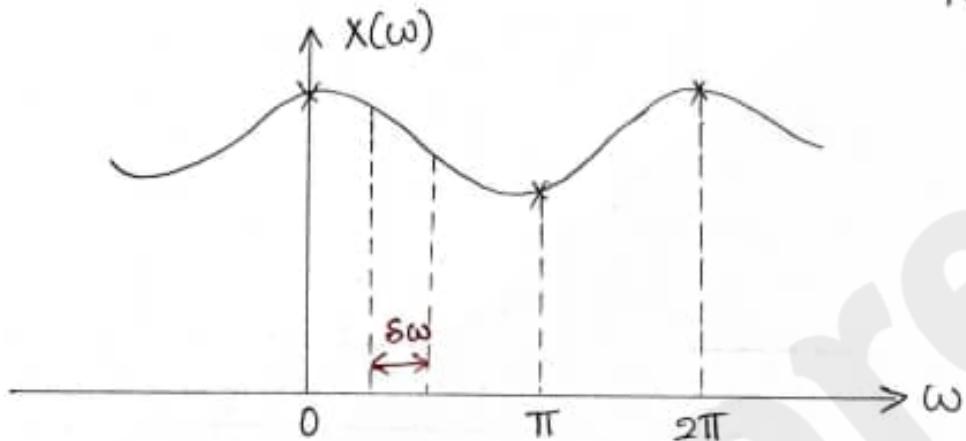
W.K.T aperiodic finite energy signals have continuous spectra.

Fourier Transform(FT) of aperiodic DTS $x(n)$ is given by

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n} \quad \text{--- (1)}$$

Since $X(\omega)$ is periodic, with period 2π , only samples in the fundamental frequency range are obtained by sampling periodically in frequency at

a spacing of $\delta\omega$ radians between successive samples. We take N equidistant samples in the interval $0 \leq \omega \leq 2\pi$ with spacing $\delta\omega = \frac{2\pi}{N}$.



Substituting $\omega = \frac{2\pi}{N}k$ in eqn ①,

$$X\left(\frac{2\pi}{N}k\right) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\frac{2\pi}{N}kn} \quad \text{--- ②}$$

where $k = 0, 1, 2, \dots, N-1$

Summation in eqn ② can be subdivided into infinite number of summations, where each sum contains N terms.

$$\begin{aligned} \therefore X\left(\frac{2\pi}{N}k\right) &= \dots + \sum_{n=-N}^{-1} x(n) e^{-j\frac{2\pi}{N}kn} + \\ &\quad \sum_{n=0}^{N-1} x(n) e^{-j\frac{2\pi}{N}kn} + \sum_{n=N}^{2N-1} x(n) e^{-j\frac{2\pi}{N}kn} + \dots \end{aligned}$$

$$= \sum_{l=-\infty}^{+\infty} \sum_{n=lN}^{lN+N-1} x(n) e^{-j\frac{2\pi}{N} km}$$

If we change the index in the inner summation from n to $n-lN$ and interchange the order of summation, we get,

$$X\left(\frac{2\pi}{N}k\right) = \sum_{n=0}^{N-1} \left[\sum_{l=-\infty}^{\infty} x(n-lN) \right] e^{-j\frac{2\pi}{N} k(m-lN)}$$

for $k=0, 1, 2, \dots, N-1$.

$$\text{But } e^{-j\frac{2\pi}{N} k(m-lN)} = e^{-j\frac{2\pi}{N} km} \cdot e^{+j\frac{2\pi}{N} klN}$$

$$\text{where } e^{j\frac{2\pi}{N} klN} = e^{j2\pi kl} = 1$$

$$\therefore X\left(\frac{2\pi}{N}k\right) = \sum_{n=0}^{N-1} \left[\sum_{l=-\infty}^{+\infty} x(n-lN) \right] e^{-j\frac{2\pi}{N} km} \quad \boxed{3}$$

$$\text{Let } x_p(n) = \sum_{l=-\infty}^{\infty} x(n-lN) \quad \boxed{4}$$

$x_p(n)$ is obtained by the periodic repetition of $x(n)$ every N samples. It is periodic with fundamental period N .

$x_p(n)$ can hence be expanded in a Fourier Series as

$$x_p(n) = \sum_{k=0}^{N-1} c_k e^{+j\frac{2\pi}{N} kn} \quad \boxed{5}$$

$n=0, 1, 2, \dots, N-1$

with Fourier coefficients,

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x_p(n) e^{-j\frac{2\pi}{N} kn} \quad \text{--- (6)}$$

$k = 0, 1, \dots, N-1$

Comparing eqns (3) & (6),

$$c_k = \frac{1}{N} X\left(\frac{2\pi}{N} k\right) ; k = 0, 1, \dots, N-1$$

(7)

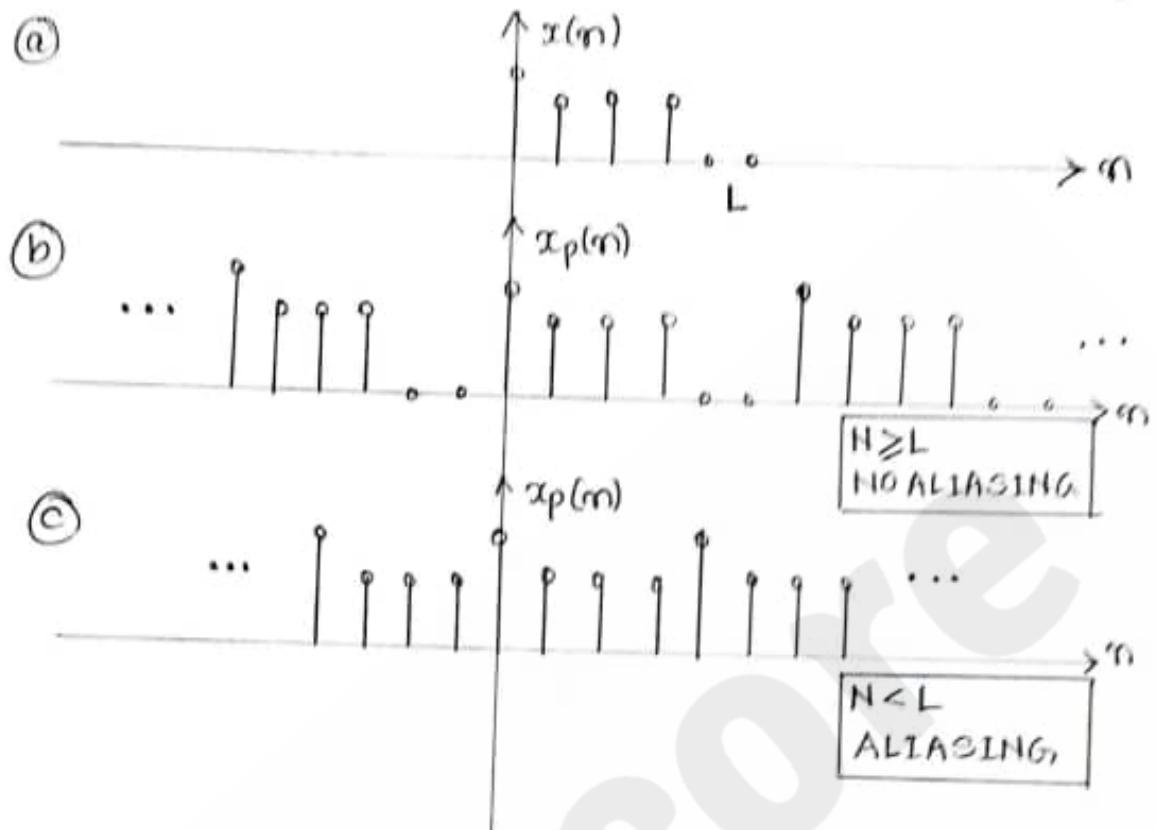
Substituting eqn (7) in eqn (5),

$$x_p(n) = \frac{1}{N} \sum_{k=0}^{N-1} X\left(\frac{2\pi}{N} k\right) e^{j\frac{2\pi}{N} kn} \quad \text{--- (8)}$$

Equation (8) provides the reconstruction of the periodic signal $x_p(n)$ from the samples of the spectrum $X(\omega)$.

However, to recover $X(\omega)$ or $x(n)$ from the samples certain condition needs to be satisfied.

Since $x_p(n)$ is the periodic extension of $x(n)$, it is clear that $x(n)$ can be recovered from $x_p(n)$ if there is no aliasing in the time domain i.e, if $x(n)$ is time limited to less than the period N of $x_p(n)$.



Consider fig @, $x(n)$ is a sequence nonzero in the interval $0 \leq n \leq L-1$.

When $N \geq L$, $x(n) = x_p(n)$ for $0 \leq n \leq N-1$ [fig (b)]

Hence $x(n)$ can be recovered from $x_p(n)$.

When $N < L$, it cannot be recovered due to time-domain aliasing.

THE DISCRETE FOURIER TRANSFORM (DFT)

In general, the equally spaced frequency samples $X\left(\frac{2\pi}{N}k\right)$, $k=0, 1, \dots, N-1$, do not uniquely represent the original sequence $x(n)$ when $x(n)$ has infinite duration.

Instead, the frequency samples $X\left(\frac{2\pi k}{N}\right)$ correspond to a periodic sequence $x_p(n)$ of period N , where $x_p(n)$ is an aliased version of $x(n)$ given by

$$x_p(n) = \sum_{l=-\infty}^{\infty} x(n-lN)$$

When the sequence $x(n)$ has a finite duration of length $L \leq N$, then $x_p(n)$ is simply a periodic repetition of $x(n)$, where $x_p(n)$ over a single period is given by

$$x_p(n) = \begin{cases} x(n), & 0 \leq n \leq L-1 \\ 0, & L \leq n \leq N-1 \end{cases}$$

Hence, the frequency samples $X\left(\frac{2\pi k}{N}\right)$, $k=0, 1, \dots, N$ uniquely represent the finite duration sequence $x(n)$. Since $x(n) \equiv x_p(n)$ over a single period (padded by $N-L$ zeros), the original finite duration sequence $x(n)$ can be obtained from the frequency samples $X\left(\frac{2\pi k}{N}\right)$ from Eqn ⑧.

\Rightarrow Summarising: A finite-duration sequence $x(n)$ of length L has a Fourier transform

$$X(\omega) = \sum_{m=0}^{L-1} x(m) e^{-j\omega m}, \quad 0 \leq \omega \leq 2\pi$$

Limits indicate $x(n)$ is zero outside the range $0 \leq n \leq L-1$.

When we sample $X(\omega)$ at equally spaced frequencies $\omega_k = \frac{2\pi k}{N}$, $k=0, 1, \dots, N-1$ where $N \geq L$, the samples are

$$X(k) \equiv X\left(\frac{2\pi k}{N}\right) = \sum_{m=0}^{L-1} x(m) e^{-j\frac{2\pi}{N} km}$$

or
$$X(k) = \sum_{m=0}^{N-1} x(m) e^{-j\frac{2\pi}{N} km}, \quad k=0, 1, 2, \dots, N-1$$

Upper index in the sum has been increased from $L-1$ to $N-1$ since $x(m) = 0$ for $m \geq L$.

The above relation is a formula for transforming a sequence $\{x(m)\}$ of length $L \leq N$ into a sequence of frequency samples $\{X(k)\}$ of length N .

Since, the frequency samples are obtained by evaluating the Fourier transform $X(\omega)$ at a set of N (equally spaced) discrete frequencies, the above relation is called discrete Fourier transform (DFT) of $x(m)$.

Let $W_N = e^{-j\frac{2\pi}{N}}$, then

$\therefore X(k) = \sum_{m=0}^{N-1} x(m) W_N^{km}, \quad k=0, 1, 2, \dots, N-1$

The relation which allows us to recover the sequence $x(n)$ from the frequency samples

$$\boxed{x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{\frac{j2\pi}{N} kn}, \quad n=0, 1, 2, \dots, N-1}$$

or $x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}$

is called the Inverse DFT (IDFT).

DFT AS A LINEAR TRANSFORMATION

WKT,

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$k=0, 1, 2, \dots, N-1$$

$$\text{IDFT } \{X(k)\} = x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}$$

$$n=0, 1, 2, \dots, N-1$$

where $W_N = e^{-j\frac{2\pi}{N}}$ Twiddle Factor

From the above equations we observe that the computation of each point of the DFT requires N -complex multiplications and $(N-1)$ complex additions.

Hence, N -point DFT requires N^2 complex multiplications and $N(N-1)$ complex additions.

DFT and IDFT can be viewed as linear transformations on sequences $\{x(n)\}$ and $\{X(k)\}$, respectively.

Let us define a N -point vector x_N of the signal sequence $x(n)$, an N -point vector X_N of frequency samples and a $N \times N$ matrix W_N as

$$x_N = \begin{bmatrix} x(0) \\ x(1) \\ \vdots \\ x(N-1) \end{bmatrix}$$

$$X_N = \begin{bmatrix} X(0) \\ X(1) \\ \vdots \\ X(N-1) \end{bmatrix}$$

$$W_N = \begin{bmatrix} W_N^{0,0} & W_N^{0,1} & \dots & W_N^{0,(N-1)} \\ W_N^{1,0} & W_N^{1,1} & \dots & W_N^{1,(N-1)} \\ \vdots & \vdots & & \vdots \\ W_N^{(N-1),0} & W_N^{(N-1),1} & \dots & W_N^{(N-1)(N-1)} \end{bmatrix}$$

or $W_N = \begin{bmatrix} 1 & 1 & 1 & \dots & 1 \\ 1 & W_N & W_N^2 & \dots & W_N^{N-1} \\ 1 & W_N^2 & W_N^4 & \dots & W_N^{2(N-1)} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W_N^{N-1} & W_N^{2(N-1)} & \dots & W_N^{(N-1)(N-1)} \end{bmatrix}$

From the above definitions, the N -point DFT can be expressed in matrix form as

$$\boxed{X_N = W_N x_N}$$

Where W_N is the matrix of the linear transformation.

We observe that W_N is a symmetric matrix. If we assume that inverse of W_N exists then IDFT can be expressed in matrix form as

$$x_N = W_N^{-1} X_N$$

$$\text{or } \boxed{x_N = \frac{1}{N} W_N^* X_N}$$

where W_N^* denotes the complex conjugate of W_N .

PROBLEMS

1) Compute the 4-point DFT of a sequence

$$x(n) = \{1, 2, 3, 4\}.$$

Sol'n:— I-METHOD :

From the def'n of DFT wkt,

$$\text{DFT}\{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}; k=0, 1, \dots, N-1$$

Given $N=4$,

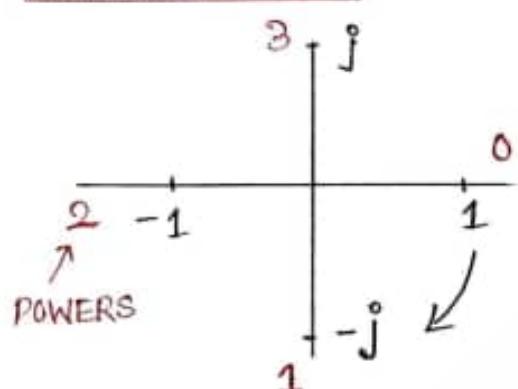
$$\therefore X(k) = \sum_{n=0}^3 x(n) W_4^{kn}$$

$$= x(0) + x(1)w_4^k + x(2)w_4^{2k} + x(3)w_4^{3k}$$

$$X(k) = 1 + 2w_4^k + 3w_4^{2k} + 4w_4^{3k} ; k=0,1,2,3$$

Twiddle Factors:

For N=4, First four twiddle factors.



$$w_4^0 = 1 \quad w_4^2 = -1$$

$$w_4^1 = -j \quad w_4^3 = j$$

Alternatively, use $w_N = e^{-j \frac{2\pi}{N}}$ and solve.

$$\text{Let } k=0, X(0) = 1 + 2 + 3 + 4 = \underline{\underline{10}}$$

$$\begin{aligned} \text{Let } k=1, X(1) &= 1 + 2w_4^1 + 3w_4^2 + 4w_4^3 \\ &= 1 + 2(-j) + 3(-1) + 4(j) \\ &= \underline{\underline{-2+2j}} \end{aligned}$$

$$\begin{aligned} \text{Let } k=2, X(2) &= 1 + 2w_4^2 + 3w_4^4 + 4w_4^6 \\ &= 1 + 2(-1) + 3(1) + 4(-1) \\ &= \underline{\underline{-2}} \end{aligned}$$

$$\begin{aligned} \text{Let } k=3, X(3) &= 1 + 2w_4^3 + 3w_4^6 + 4w_4^9 \\ &= 1 + 2(j) + 3(-1) + 4(-j) \\ &= \underline{\underline{-2-2j}} \end{aligned}$$

$$\therefore X(k) = \{ 10, -2+2j, -2, -2-2j \}$$

II - METHOD :-

$$X_N = x_N W_N \quad X_N \& x_N \rightarrow \text{Row vectors}$$

$$\text{For } N=4, \quad X_4 = x_4 W_4$$

$$= [1 \ 2 \ 3 \ 4] \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^1 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^4 & W_4^6 \\ 1 & W_4^3 & W_4^6 & W_4^9 \end{bmatrix}$$

From periodicity property of W_N : $W_N^{k+N} = W_N^k$

and symmetry property : $W_N^{k+\frac{N}{2}} = -W_N^k$

$$X_4 = [1 \ 2 \ 3 \ 4] \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$\therefore X(k) = \{10, -2+2j, -2, -2-2j\}$$

Additional problems on DFT :-

2) $x(n) = \{0, 1, 2, 3\}$ for $N=4$.

ANS: $X(k) = \{6, -2+2j, -2, -2-2j\}$

3) $x(n) = \{1, 2, 2, 1\}$ for $N=4$.

ANS: $X(k) = \{6, -1-j, 0, -1+j\}$

$$4) \quad x(n) = s(n) + s(n-1) - s(n-2) - s(n-3) \\ = \{ 1, 1, -1, -1 \}$$

$$\text{ANS : } X(k) = \{ 0, 2-2j, 0, 2+2j \}$$

5) Find the 4-point IDFT of $X(k) = \{ 10, -2+2j, -2, -2-2j \}$.

Soln:- From the defn of IDFT,

$$\text{IDFT } \{ X(k) \} = x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}$$

$$n = 0, 1, 2, \dots, N-1$$

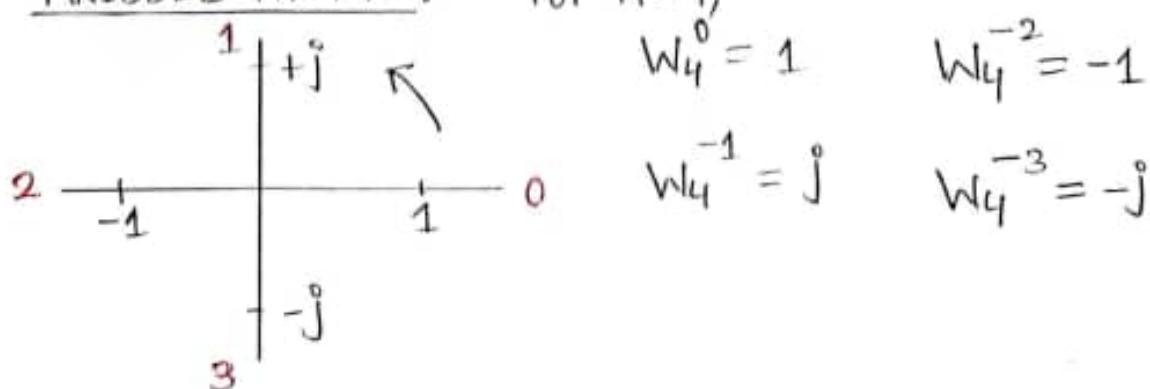
For $N=4$,

$$x(n) = \frac{1}{4} \sum_{k=0}^3 X(k) W_4^{-kn}$$

$$x(n) = \frac{1}{4} \left[X(0) + X(1) W_4^{-n} + X(2) W_4^{-2n} + X(3) W_4^{-3n} \right]$$

$$x(n) = \frac{1}{4} \left[10 + (-2+2j) W_4^{-n} + (-2) W_4^{-2n} + (-2-2j) W_4^{-3n} \right], n=0, 1, 2, 3.$$

TWIDDLE FACTORS:



Let $n=0$,

$$x(0) = \frac{1}{4} \left[10 - 2+2j - 2 - 2-2j \right]$$

$$x(0) = \frac{4}{4} = \underline{\underline{1}}$$

Let $n=1$,

$$\begin{aligned} x(1) &= \frac{1}{4} \left[10 + (-2+2j) w_4^{-1} + (-2) w_4^{-2} \right. \\ &\quad \left. + (-2-2j) w_4^{-3} \right] \end{aligned}$$

$$\begin{aligned} &= \frac{1}{4} \left[10 + (-2+2j)(j) + (-2)(-1) \right. \\ &\quad \left. + (-2-2j)(-j) \right] \end{aligned}$$

$$x(1) = \frac{8}{4} = \underline{\underline{2}}$$

Let $n=2$,

$$\begin{aligned} x(2) &= \frac{1}{4} \left[10 + (-2+2j) w_4^{-2} + (-2) w_4^{-4} \right. \\ &\quad \left. + (-2-2j) w_4^{-6} \right] \end{aligned}$$

$$\begin{aligned} &= \frac{1}{4} \left[10 + (-2+2j)(-1) + (-2)(1) \right. \\ &\quad \left. + (-2-2j)(-1) \right] \end{aligned}$$

$$x(2) = \frac{12}{4} = \underline{\underline{3}}$$

$$\text{Let } n=3, \quad x(3) = \frac{1}{4} \left[10 + (-2+2j) w_4^{-3} + (-2) w_4^{-6} \right. \\ \left. + (-2-2j) w_4^{-9} \right]$$

$$x(3) = \frac{16}{4} = \underline{\underline{4}}$$

$$\therefore \boxed{x(n) = \{1, 2, 3, 4\}}$$

II - METHOD :-

$$\text{WKT, } x(n) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) w_N^{-kn}$$

Above equation can also be written as,

$$x_N = \frac{1}{N} X_N W_N^* \quad X_N, X_N \rightarrow \text{Row vectors.}$$

Given $N=4$,

$$x_4 = \frac{1}{4} X_4 W_4^*$$

$$X_4 = \frac{1}{4} \begin{bmatrix} 10 & (-2+2j) & -2(-2-2j) \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix}$$

$$= \frac{1}{4} \begin{bmatrix} 10 - 2 + 2j - 2 - 2 - 2j \\ 10 - 2j - 2 + 2 + 2j - 2 \\ 10 + 2 - 2j - 2 + 2 + 2j \\ 10 + 2j + 2 + 2 - 2j + 2 \end{bmatrix} = \frac{1}{4} \begin{bmatrix} 4 \\ 8 \\ 12 \\ 16 \end{bmatrix}$$

∴ $x(n) = \{1, 2, 3, 4\}$

NOTE :- ANSWER
IS IN ROW,
REPRESENTED AS
COLUMN ONLY FOR
CONVENIENCE.

⑥ Find the DFT of $x(n) = \delta(n)$.

Soln:- From def'n of DFT,

$$\text{DFT} \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$\text{Hence, DFT} \{s(n)\} = \sum_{n=0}^{N-1} s(n) W_N^{kn}$$

$$\text{W.K.T, } s(n) = \begin{cases} 1, & \text{at } n=0 \\ 0, & \text{for } n \neq 0 \end{cases}$$

$$\therefore \text{DFT} \{s(n)\} = s(0) W_N^0 = 1 //$$

or
$$\boxed{s(n) \xleftrightarrow[\text{DFT}]{\text{N-PT}} 1}$$

⑦ Find the DFT of $x(n) = \delta(n - m_0)$.

Soln:- From the def'n of DFT,

$$\text{DFT} \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$\text{Hence, DFT} \{s(n - m_0)\} = \sum_{n=0}^{N-1} s(n - m_0) W_N^{kn}$$

$$\text{W.K.T, } s(n - m_0) = \begin{cases} 1, & \text{at } n = m_0 \\ 0, & \text{for } n \neq m_0 \end{cases}$$

$$\therefore \text{DFT} \{s(n - m_0)\} = s(0) W_N^{k m_0} = W_N^{k m_0}$$

$$\boxed{s(n - m_0) \xleftrightarrow[\text{DFT}]{\text{N-PT}} W_N^{k m_0}}$$

⑧ Find N-point DFT of $x(n) = e^{\frac{j2\pi}{N} k_0 n}$.

Soln:- From definition of DFT

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$\text{Hence, DFT } \{e^{\frac{j2\pi}{N} k_0 n}\} = \sum_{n=0}^{N-1} e^{\frac{j2\pi}{N} k_0 n} \cdot e^{-\frac{j2\pi}{N} kn}$$

$$= \sum_{n=0}^{N-1} \left[e^{-\frac{j2\pi}{N} (k-k_0)} \right]^n$$

$$X(k) = \frac{1 - \left[e^{-\frac{j2\pi}{N} (k-k_0)} \right]^N}{1 - e^{-\frac{j2\pi}{N} (k-k_0)}} \quad \because \sum_{n=0}^N a^n = \frac{1-a^{N+1}}{1-a}$$

at $k=k_0$, $X(k) = \frac{0}{0}$, Using L-Hospital's Method,

$$= \frac{0 - \left(-j2\pi e^{-j2\pi(k-k_0)} \right)}{0 - \left(-j2\pi \frac{e^{-j2\pi(k-k_0)}}{N} \right)} \Bigg|_{k=k_0}$$

At $k=k_0$,

$$= \frac{j2\pi \cdot e^0}{j2\pi \frac{e^0}{N}} = N$$

$\therefore X(k) = N$ at $k=k_0$

at $k \neq k_0$ $X(k) = 0$

$$\therefore X(k) = N \delta(k-k_0)$$

$$X(k) = \begin{cases} N & \text{at } k=k_0 \\ 0 & \text{for } k \neq k_0 \end{cases}$$

⑨ Find the DFT of $x(n) = e^{-j\frac{2\pi}{N}k_0 n}$.

Soln:- From the def'n of DFT,

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$\text{Hence, DFT } \left\{ e^{-j\frac{2\pi}{N}k_0 n} \right\} = \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N}k_0 n} \cdot e^{-j\frac{2\pi}{N}kn}$$

$$= \sum_{n=0}^{N-1} \left[e^{-j\frac{2\pi}{N}(k+k_0)} \right]^n$$

$$= 1 - \left[e^{-j\frac{2\pi}{N}(k+k_0)} \right]^N$$

$$\frac{1 - e^{-j\frac{2\pi}{N}(k+k_0)}}{1 - e^{-j\frac{2\pi}{N}(k+k_0)}}$$

$$\therefore \sum_{n=0}^N a^n = \frac{1 - a^{N+1}}{1 - a}$$

$$\text{at } k = -k_0, \quad X(k) = \frac{0}{0}.$$

Using L-Hospital's Rule,

$$X(k) = \frac{0 - (-j\frac{2\pi}{N}) e^{-j\frac{2\pi}{N}(k+k_0)}}{0 - (-j\frac{2\pi}{N}) e^{-j\frac{2\pi}{N}(k+k_0)}} \Big|_{k=-k_0}$$

$$\text{at } k = -k_0, \quad X(k) = \frac{j\frac{2\pi}{N}}{j\frac{2\pi}{N}/N} = N$$

$$\text{at } k \neq -k_0, \quad X(k) = 0$$

$$\therefore X(k) = N S(k+k_0)$$

$$\text{or } X(k) = N S(k-N+k_0)$$

$$X(k) = \begin{cases} N & \text{at } k=k_0 \\ 0 & \text{at } k \neq k_0 \end{cases}$$

- (10) Find the N-point DFT of $x(n) = \cos \frac{2\pi}{N} k_0 n$

Sol'n:- From def'n of DFT,

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$\text{DFT } \left\{ \cos \left(\frac{2\pi}{N} k_0 n \right) \right\} = \sum_{n=0}^{N-1} \cos \left(\frac{2\pi}{N} k_0 n \right) W_N^{kn}$$

$$= \sum_{n=0}^{N-1} \left[\frac{e^{j \frac{2\pi}{N} k_0 n} + e^{-j \frac{2\pi}{N} k_0 n}}{2} \right] e^{-j \frac{2\pi}{N} kn}$$

$$= \frac{1}{2} \left\{ \sum_{n=0}^{N-1} e^{-j \frac{2\pi}{N} (k - k_0) n} + \sum_{n=0}^{N-1} e^{-j \frac{2\pi}{N} (k + k_0) n} \right\}$$

$$\stackrel{\because \cos \theta}{=} \frac{e^{j\theta} + e^{-j\theta}}{2}$$

$$= \frac{1}{2} \left\{ N \delta(k - k_0) + N \delta(k + k_0) \right\}$$

$$\therefore \boxed{\cos \left(\frac{2\pi}{N} k_0 n \right) \xleftrightarrow{\text{DFT}} \frac{N}{2} [\delta(k - k_0) + \delta(k + k_0)]}$$

- (11) Similar problem, $x(n) = \sin \frac{2\pi}{N} k_0 n$

$$\underline{\text{Ans:}} \quad \sin \left(\frac{2\pi}{N} k_0 n \right) \xleftrightarrow{\text{DFT}} \frac{N}{2j} [\delta(k - k_0) - \delta(k + k_0)]$$

(12) Find the 5-point DFT of $x(n) = \{1, 1, 1\}$.

Sol'n: From the def'm of DFT, WKT,

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad k=0, 1, \dots, N-1$$

Given $N = 5$,

$$X(k) = \sum_{n=0}^4 x(n) W_5^{kn} \quad ; \quad k=0, 1, 2, 3, 4$$

$$X(k) = x(0) + x(1) W_5^k + x(2) W_5^{2k} + x(3) W_5^{3k} \\ + x(4) W_5^{4k}$$

$X(k) = 1 + W_5^k + W_5^{2k}$	$k=0, 1, 2, 3, 4$
-------------------------------	-------------------

Twiddle Factors: $N=5$

$$W_5^0 = 1$$

$$W_5^1 = e^{-j \frac{2\pi}{5}} = \cos\left(\frac{2\pi}{5}\right) - j \sin\left(\frac{2\pi}{5}\right) \\ = 0.3090 - j 0.95$$

$$W_5^2 = e^{-j \frac{4\pi}{5}} = \cos\left(\frac{4\pi}{5}\right) - j \sin\left(\frac{4\pi}{5}\right) \\ = -0.309 - j 0.8278$$

$$W_5^3 = e^{-j \frac{6\pi}{5}} = \cos\left(\frac{6\pi}{5}\right) - j \sin\left(\frac{6\pi}{5}\right) \\ = -0.309 + j 0.587$$

$$\begin{aligned} w_5^4 &= e^{-j \frac{2\pi}{5} \cdot 4} = \cos\left(\frac{8\pi}{5}\right) - j \sin\left(\frac{8\pi}{5}\right) \\ &= 0.309 + j 0.95 \end{aligned}$$

Evaluate $X(k)$,

$$\text{at } k=0, X(0) = 1 + 1 + 1 = \underline{\underline{3}}$$

$$\text{at } k=1, X(1) = 1 + w_5^1 + w_5^2 = \underline{\underline{0.5 - j 1.5388}}$$

$$\text{at } k=2, X(2) = 1 + w_5^2 + w_5^4 = \underline{\underline{0.5 + j 0.3633}}$$

$$\text{at } k=3, X(3) = 1 + w_5^3 + w_5^6 = \underline{\underline{0.5 - j 0.3633}}$$

$$\text{at } k=4, X(4) = 1 + w_5^4 + w_5^8 = \underline{\underline{0.5 + j 1.5388}}$$

$$X(k) = \left\{ \begin{array}{l} 3, 0.5 - j 1.5388, 0.5 + j 0.3633, \\ 0.5 - j 0.3633, 0.5 + j 1.5388 \end{array} \right\}$$

(B) Find IDFT for the following sequence

$$X(k) = \left\{ 5, 0, (1-j), 0, 1, 0, (1+j), 0 \right\}$$

Soln:- Length of the sequence is 8.

$$\therefore N = 8$$

From the def'n of IDFT,

$$\text{IDFT } \{ X(k) \} = x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) w_N^{-kn}$$

where $n = 0, 1, 2, \dots, N-1$.

For $N=8$,

$$x(n) = \frac{1}{8} \sum_{k=0}^7 X(k) W_8^{-kn}$$

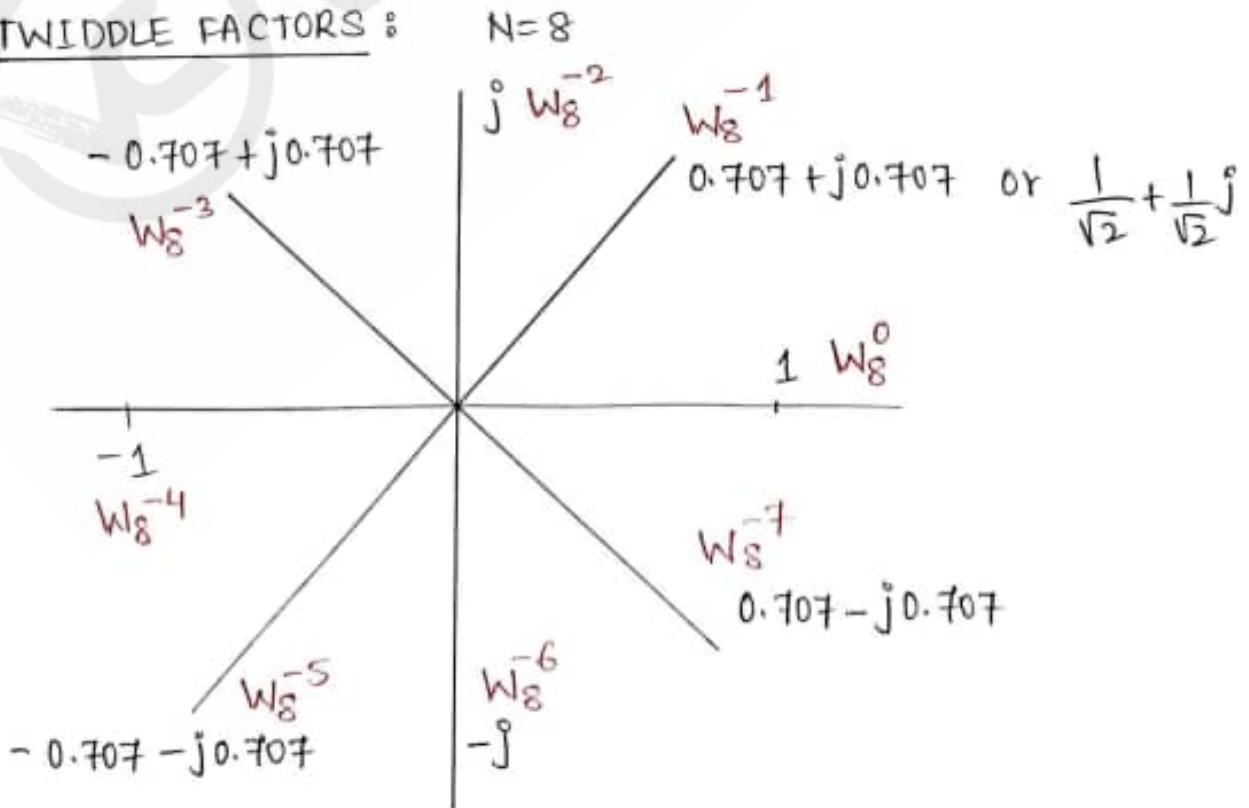
$$\begin{aligned} x(n) = \frac{1}{8} & \left[X(0) W_8^0 + X(1) W_8^{-n} + X(2) W_8^{-2n} \right. \\ & + X(3) W_8^{-3n} + X(4) W_8^{-4n} + X(5) W_8^{-5n} \\ & \left. + X(6) W_8^{-6n} + X(7) W_8^{-7n} \right] \end{aligned}$$

$$\begin{aligned} x(n) = \frac{1}{8} & \left[5 + 0 + (1-j) W_8^{-2n} + 0 + W_8^{-4n} \right. \\ & \left. + 0 + (1+j) W_8^{-6n} + 0 \right] \end{aligned}$$

where $n = 0, 1, 2, \dots, 7$

or
$$x(n) = \frac{1}{8} \left[5 + (1-j) W_8^{-2n} + W_8^{-4n} + (1+j) W_8^{-6n} \right]$$

TWIDDLE FACTORS :



$$\text{For } m=0, \quad x(0) = \frac{1}{8} \left\{ s + 1-j + 1 + 1+j \right\}$$

$$= \frac{8}{8} = \underline{\underline{1}}$$

$$\text{For } m=1, \quad x(1) = \frac{1}{8} \left\{ s + (1-j) w_8^{-2} + w_8^{-4} \right.$$

$$\quad \quad \quad \left. + (1+j) w_8^{-6} \right\}$$

$$= \frac{1}{8} \left\{ s + (1-j)j + (-1) + (1+j)(-j) \right\}$$

$$= \frac{1}{8} \left\{ s + j - j^2 - 1 - j - j^2 \right\}$$

$$= \frac{6}{8} = \frac{3}{4} \underline{\underline{}}$$

$$\text{For } m=2, \quad x(2) = \frac{1}{8} \left\{ s + (1-j) w_8^{-4} + w_8^{-8} \right.$$

$$\quad \quad \quad \left. + (1+j) w_8^{-12} \right\}$$

$$= \frac{1}{8} \left\{ s + (1-j)(-1) + 1 + (1+j)(-1) \right\}$$

$$= \frac{1}{8} \left\{ s - 1 + j + 1 - 1 - j \right\}$$

$$= \frac{4}{8} = \frac{1}{2} \underline{\underline{}}$$

$$\text{For } m=3, \quad x(3) = \frac{1}{8} \left\{ s + (1-j) w_8^{-6} + w_8^{-12} + \right.$$

$$\quad \quad \quad \left. (1+j) w_8^{-18} \right\}$$

$$= \frac{1}{8} \left\{ s + (1-j)(-j) + (-1) + (1+j)(j) \right\}$$

$$\begin{aligned}x(3) &= \frac{1}{8} \left\{ s - j + j^2 - 1 + j + j^2 \right\} \\&= \frac{1}{8} \left\{ s - j - 1 - 1 + j - 1 \right\} = \frac{2}{8} = \underline{\underline{\frac{1}{4}}}\end{aligned}$$

for $m=4$, $x(4) = \frac{1}{8} \left\{ s + (1-j) w_8^{-8} + w_8^{-16} + (1+j) w_8^{-24} \right\}$

$$\begin{aligned}&= \frac{1}{8} \left\{ s + (1-j) + 1 + (1+j) \right\} \\&= \underline{\underline{\frac{8}{8}}} = \underline{\underline{\frac{1}{1}}}\end{aligned}$$

for $m=5$, $x(5) = \frac{1}{8} \left\{ s + (1-j) w_8^{-10} + w_8^{-20} + (1+j) w_8^{-30} \right\}$

$$\begin{aligned}&= \frac{1}{8} \left\{ s + (1-j)j + (-1) + (1+j)(-j) \right\} \\&= \frac{1}{8} \left\{ s + j - j^2 - 1 - j - j^2 \right\} \\&= \underline{\underline{\frac{6}{8}}} = \underline{\underline{\frac{3}{4}}}\end{aligned}$$

for $m=6$, $x(6) = \frac{1}{8} \left\{ s + (1-j) w_8^{-12} + w_8^{-24} + (1+j) w_8^{-36} \right\}$

$$\begin{aligned}&= \frac{1}{8} \left\{ s + (1-j)(-1) + 1 + (1+j)(-1) \right\} \\&= \frac{1}{8} \left\{ s - 1 + j + 1 - 1 - j \right\} = \underline{\underline{\frac{4}{8}}} = \underline{\underline{\frac{1}{2}}}\end{aligned}$$

$$\begin{aligned}
 \text{for } n=7, \quad x(7) &= \frac{1}{8} \left\{ 5 + (1-j) w_8^{-14} + w_8^{-28} \right. \\
 &\quad \left. + (1+j) w_8^{-42} \right\} \\
 &= \frac{1}{8} \left\{ 5 + (1-j)(-j) + (-1) + (1+j)(j) \right\} \\
 &= \frac{1}{8} \left\{ 5 - j + j^2 - 1 + j + j^2 \right\} \\
 &= \frac{2}{8} = \frac{1}{4} \\
 \therefore x(n) &= \boxed{\left\{ 1, 1, \frac{3}{4}, \frac{1}{2}, \frac{1}{4}, 1, \frac{3}{4}, \frac{1}{2}, \frac{1}{4} \right\}}
 \end{aligned}$$

(14) Find DFT of the sequence $x(n) = \begin{cases} 1, & 0 \leq n \leq 2 \\ 0, & \text{otherwise} \end{cases}$
 for $N=8$. Plot $|X(k)|$ & $\angle X(k)$.

Soln:- From the defn of DFT,

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) w_N^{kn}$$

for $N=8$,

$$X(k) = \sum_{n=0}^7 x(n) w_8^{kn}, \quad 0 \leq k \leq 7$$

$$\text{Given } x(n) = \{1, 1, 1\}$$

$$\therefore X(k) = 1 + W_8^k + W_8^{2k} + 0 + 0 + 0 + 0 + 0$$

$$\boxed{X(k) = 1 + W_8^k + W_8^{2k}}, k=0,1,2,\dots,7$$

Twiddle Factors for N=8

$$W_8^0 = 1$$

$$W_8^4 = -1$$

$$W_8^1 = 0.707 - j0.707$$

$$W_8^5 = -0.707 + j0.707$$

$$W_8^2 = -j$$

$$W_8^6 = j$$

$$W_8^3 = -0.707 - j0.707 \quad W_8^7 = 0.707 + j0.707$$

(Locate points on the unit circle in complex plane
direction is clockwise for DFT)

$$\text{for } k=0, \quad X(0) = 1 + 1 + 1 = \underline{\underline{3}}$$

$$\begin{aligned} \text{for } k=1, \quad X(1) &= 1 + W_8^1 + W_8^2 \\ &= 1 + (0.707 - j0.707) + (-j) \\ &= \underline{\underline{1.707 - j1.707}} \end{aligned}$$

$$\begin{aligned} \text{for } k=2, \quad X(2) &= 1 + W_8^2 + W_8^4 \\ &= 1 + (-j) + (-1) = \underline{\underline{-j}} \end{aligned}$$

$$\begin{aligned} \text{for } k=3, \quad X(3) &= 1 + W_8^3 + W_8^6 \\ &= 1 + (-0.707 - j0.707) + j \\ &= \underline{\underline{0.293 + j0.293}} \end{aligned}$$

$$\text{for } k=4, \quad X(4) = 1 + w_8^4 + w_8^8 \\ = 1 + (-1) + 1 = \underline{\underline{1}}$$

$$\text{for } k=5, \quad X(5) = 1 + w_8^5 + w_8^{10} \\ = 1 + (-0.707 + j0.707) + (-j) \\ = \underline{\underline{0.293 - j0.293}}$$

$$\text{for } k=6, \quad X(6) = 1 + w_8^6 + w_8^{12} \\ = 1 + j + (-1) = \underline{\underline{j}}$$

$$\text{for } k=7, \quad X(7) = 1 + w_8^7 + w_8^{14} \\ = 1 + 0.707 + j0.707 + j \\ = \underline{\underline{1.707 + j1.707}}$$

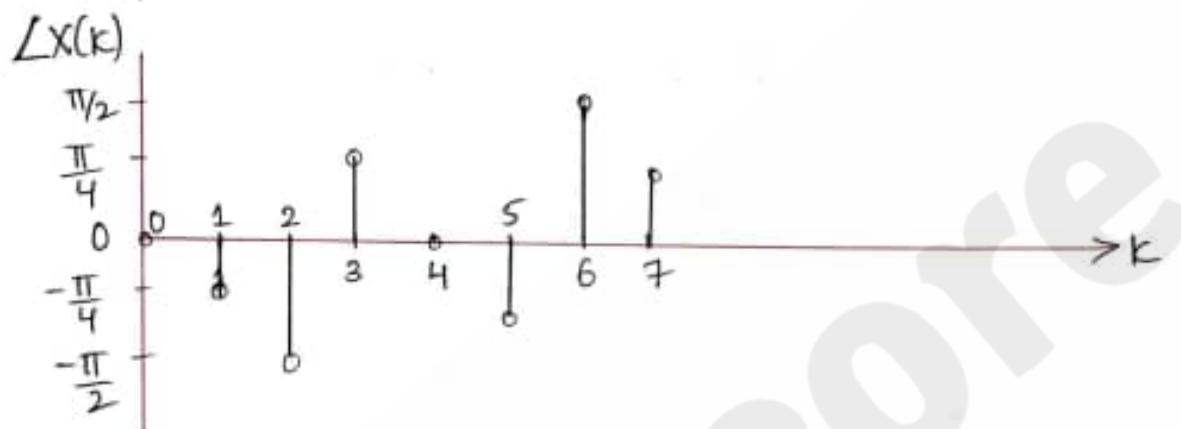
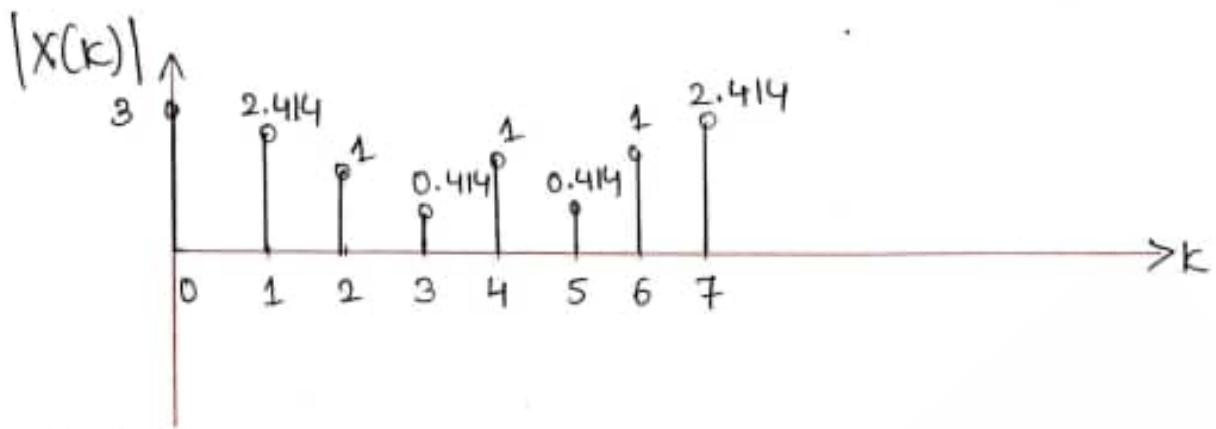
$$\therefore X(k) = \left\{ 3, (1.707 - j1.707), -j, (0.293 + j0.293), 1, (0.293 - j0.293), j, (1.707 + j1.707) \right\}$$

For magnitude Plot,

$$|X(k)| = \{ 3, 2.414, 1, 0.414, 1, 0.414, 1, 2.414 \}$$

Phase Plot,

$$\angle X(k) = \left\{ 0, -\frac{\pi}{4}, -\frac{\pi}{2}, \frac{\pi}{4}, 0, \frac{\pi}{4}, \frac{\pi}{2}, \frac{\pi}{4} \right\}$$



PROPERTIES OF DFT

1) LINEARITY :-

STATEMENT:

$$\text{If } x_1(n) \xrightarrow{\text{DFT}} X_1(k)$$

$$x_2(n) \xrightarrow{\text{DFT}} X_2(k)$$

$$\text{then } a x_1(n) + b x_2(n) \xrightarrow{\text{DFT}} a X_1(k) + b X_2(k)$$

Proof: W.K.T., $X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$

Let $x(n) = x_1(n)$,

$$\text{then } X_1(k) = \sum_{n=0}^{N-1} x_1(n) W_N^{kn} \quad \text{--- ①}$$

Let $x(n) = x_2(n)$,

$$\text{then } X_2(k) = \sum_{n=0}^{N-1} x_2(n) W_N^{kn} \quad \text{--- ②}$$

Let $x(n) = a x_1(n) + b x_2(n)$, then

$$\begin{aligned} \text{DFT } \left\{ a x_1(n) + b x_2(n) \right\} &= \sum_{n=0}^{N-1} [a x_1(n) + b x_2(n)] W_N^{kn} \\ &= a \sum_{n=0}^{N-1} x_1(n) W_N^{kn} + b \sum_{n=0}^{N-1} x_2(n) W_N^{kn} \\ &= a X_1(k) + b X_2(k) \end{aligned}$$

Hence the proof.

2) PERIODICITY

STATEMENT: If $x(n) \xleftrightarrow{\text{DFT}} X(k)$

$$\text{then } x(n+N) = x(n)$$

$$X(k+N) = X(k)$$

$$\text{or } x(n+N) \xleftrightarrow{\text{DFT}} X(k)$$

Proof: a) TPT $x(n+N) = x(n)$

$$\text{W.K.T } x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}$$

replace n by $n+N$,

$$x(n+N) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-k(n+N)}$$

$$= \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} \cdot W_N^{-kN}$$

$$\text{W.K.T } W_N^{-kN} = 1$$

$$\therefore x(n+N) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}$$

$x(n+N) = x(n)$

Hence proved.

b) T.P.T $X(k+N) = X(k)$

WKT $X(k) = \sum_{m=0}^{N-1} x(m) W_N^{km}$

replace k by $k+N$,

$$\begin{aligned} X(k+N) &= \sum_{m=0}^{N-1} x(m) W_N^{(k+N)m} \\ &= \sum_{m=0}^{N-1} x(m) W_N^{km} \cdot W_N^{Nm} \end{aligned}$$

$$X(k+N) = \sum_{m=0}^{N-1} x(m) W_N^{km} \quad \because W_N^{Nm} = 1$$

$$\therefore \boxed{X(k+N) = X(k)}$$

Hence proved.

c) T.P.T $x(n+N) \xleftrightarrow{\text{DFT}} X(k)$

WKT, DFT $\{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$

Let $x(n) = x(n+N)$, then

$$\text{DFT } \{x(n+N)\} = \sum_{n=0}^{N-1} x(n+N) W_N^{kn}$$

Put $\ell = n+N \Rightarrow n = \ell - N$

limits : $n=0, \ell=N$

$$n=N-1, \ell=N-1+N = 2N-1$$

$$= \sum_{l=N}^{2N-1} x(l) w_N^{k(l-N)}$$

$$= \sum_{l=N}^{2N-1} x(l) w_N^{kl} \cdot w_N^{-KN}$$

since $w_N^{-KN} = 1$ & limits N to $2N-1$ can be replaced by 0 to $N-1$ using periodicity.

$$\text{DFT } \{x(n+N)\} = \sum_{l=0}^{N-1} x(l) w_N^{kl}$$

$$\therefore \boxed{\text{DFT } \{x(n+N)\} = X(k)}$$

Hence proved.

③ CIRCULAR TIME-SHIFT

Circular time shift operation on an N-point sequence $x(n)$ is given by $x((n-m))_{\text{mod } N}$, $x((n-m))_N$ or $x(n-m, \text{mod } N)$.

STATEMENT: If $x(n) \xleftrightarrow{\text{DFT}} X(k)$

$$\text{then } x((n-m))_N \xleftrightarrow{\text{DFT}} X(k) e^{-j \frac{2\pi}{N} km}$$

$$\text{or } x((n-m))_N \xleftrightarrow{\text{DFT}} X(k) W_N^{km}$$

$$\text{||| by } x((n+m))_N \xleftrightarrow{\text{DFT}} X(k) e^{j \frac{2\pi}{N} km}$$

$$\text{or } x((n+m))_N \xleftrightarrow{\text{DFT}} X(k) W_N^{-km}$$

Proof: W.K.T $X(k) = \sum_{m=0}^{N-1} x(m) W_N^{km}$

$$\text{Let } x(m) = x((n-m))_N$$

$$\text{then DFT } \{x((n-m))_N\} = \sum_{m=0}^{N-1} x((n-m))_N \cdot W_N^{km}$$

$$\text{Let } n-m = l \Rightarrow m = l+n, \text{ limits:}$$

$$\text{at } m=0, l=-n$$

$$m=N-1, l=N-1-n$$

$$\text{DFT } \{x((n-m))_N\} = \sum_{l=-n}^{N-1-n} x(l) W_N^{k(l+n)}$$

$$= \sum_{l=0}^{N-1} x(l) W_N^{kl} \cdot W_N^{kn} \quad \text{using periodicity}$$

$$\therefore \boxed{\text{DFT } \{x((n-m))_N\} = X(k) W_N^{km}}$$

Example 15 If $x(n) = \{1, 2, 3, 4\}$ find $y(n)$, given $y(n) = x((n-3))_4$.

Soln:- Given $y(n) = x((n-3))_4$

$$\begin{aligned} \text{at } n=0, \quad y(0) &= x((-3))_4 = x(4-3) \\ &= x(1) = \underline{\underline{2}} \end{aligned}$$

$$\begin{aligned} \text{at } n=1, \quad y(1) &= x((1-3))_4 = x((-2))_4 \\ &= x(4-2) = x(2) = \underline{\underline{3}} \end{aligned}$$

$$\begin{aligned} \text{at } n=2, \quad y(2) &= x((2-3))_4 = x((-1))_4 \\ &= x(4-1) = x(3) = \underline{\underline{4}} \end{aligned}$$

$$\begin{aligned} \text{at } n=3, \quad y(3) &= x((3-3))_4 = x(0) \\ &= \underline{\underline{1}} \end{aligned}$$

$$\therefore y(n) = \boxed{\{2, 3, 4, 1\}}$$

16 If $x(n) = \{1, 2, 2, 1\}$ find the DFT of $y(n) = x((n-2))_4$.

Soln:- First find $X(k)$.

$$X_N = x_N W_N$$

$$X(k) = \begin{bmatrix} 1 & 2 & 2 & 1 \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$\therefore X(k) = [6 \quad -1-j \quad 0 \quad -1+j]$$

Given $y(n) = x((n-2))_4$

Taking DFT and applying circular time-shift property,

$$Y(k) = X(k) W_4^{2k}, \quad k=0, 1, 2, 3$$

$$\text{at } k=0, \quad Y(0) = X(0) W_4^0 = (6)(1) = 6$$

$$\text{at } k=1, \quad Y(1) = X(1) W_4^2 = (-1-j)(-1) = 1+j$$

$$\text{at } k=2, \quad Y(2) = X(2) W_4^4 = 0$$

$$\text{at } k=3, \quad Y(3) = X(3) W_4^6 = (-1+j)(-1) = 1-j$$

$$\therefore \boxed{Y(k) = \{6, 1+j, 0, 1-j\}}$$

4) TIME REVERSAL

STATEMENT: If $x(n) \xrightarrow[N]{\text{DFT}} X(k)$

$$\text{then. } x((-n))_N = x(N-n) \xrightarrow[N]{\text{DFT}} X(-k)$$

$$X((-k))_N = X(N-k)$$

Proof: - From def'n of DFT, we have

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$$\text{Hence, DFT } \{x(N-n)\} = \sum_{m=0}^{N-1} x(N-n) W_N^{km}$$

$$\text{Put } m = N-n \Rightarrow n = N-m$$

limits : at $m=0, m=N$

$$m=N-1, m=N-(N-1) = 1$$

$$\therefore \text{DFT } \{x(N-m)\} = \sum_{m=N}^1 x(m) W_N^{k(N-m)}$$

$$= \sum_{m=0}^{N-1} x(m) W_N^{KN} \cdot W_N^{-km}$$

using
periodicity
property.

$$\text{Also, } W_N^{KN} = 1$$

$$\text{DFT } \{x(N-m)\} = \sum_{m=0}^{N-1} x(m) W_N^{(-k)m}$$

$$\boxed{\text{or DFT } \{x(N-m)\} = X((-k))_N}$$

(17) Example : $x(n) = \{5, 2, -3, 7\}$. Find
 $y(n) = x((-n))_4$.

Sol'n: Consider $y(n) = x((-n))_4$

$$\text{or } y(n) = x(4-n)$$

$$\text{at } n=0, \quad y(0) = x(4) = x(0) = 5$$

$$n=1, \quad y(1) = x(4-1) = x(3) = 7$$

$$n=2, \quad y(2) = x(4-2) = x(2) = -3$$

$$n=3, \quad y(3) = x(4-3) = x(1) = 2$$

$$\therefore y(n) = \boxed{\{5, 7, -3, 2\}}$$

(18) Given $x(n) = \{1, 2, 3, 4\}$ with 4-point DFT $X(k) = \{10, -2+2j, -2, -2-2j\}$. Find the 4-point DFT of the sequence $y(n) = \{1, 4, 3, 2\}$.

Sol'n:- Observing $x(n) & y(n)$ we find that

$$y(n) = x((-n))_4$$

Taking DFT on both sides and applying time reversal property

$$Y(k) = X((-k))_4 = X(4-k)$$

$$\text{at } k=0, \quad Y(0) = X(0) = 10$$

$$k=1, \quad Y(1) = X(4-1) = X(3) = -2-2j$$

$$k=2, \quad Y(2) = X(4-2) = X(2) = -2$$

$$k=3, \quad Y(3) = X(4-3) = X(1) = -2+2j$$

$$\therefore Y(k) = \{ 10, -2-2j, -2, -2+2j \}$$

5) CIRCULAR FREQUENCY SHIFT

STATEMENT: $\sum x(n) \xrightarrow[N]{\text{DFT}} X(k)$

then $x(n) e^{\frac{j2\pi}{N} mn} \xrightarrow[N]{\text{DFT}} X((k-m))_N$

or $x(n) W_N^{-mn} \xrightarrow{\text{DFT}} X((k-m))_N$

Similarly $x(n) e^{-\frac{j2\pi}{N} mn} \xleftrightarrow{\text{DFT}} X((k+m))_N$

or $x(n) W_N^{mn} \xleftrightarrow{\text{DFT}} X((k+m))_N$

Proof:- $X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$

$$\begin{aligned} \text{DFT} \left\{ x(n) e^{\frac{j2\pi}{N} mn} \right\} &= \sum_{n=0}^{N-1} x(n) e^{\frac{j2\pi}{N} mn} \cdot e^{-\frac{j2\pi}{N} kn} \\ &= \sum_{n=0}^{N-1} x(n) e^{-\frac{j2\pi}{N} (k-m)n} \end{aligned}$$

$$= X((k-m))_N$$

Hence the proof.

(19) Find N-point DFT of $x_1(n) = \cos\left(\frac{2\pi}{N}k_0n\right)x(n)$.

Soln:-

$$\text{Given } x_1(n) = \cos\left(\frac{2\pi}{N}k_0n\right)x(n)$$

$$= \left[\frac{e^{j\frac{2\pi}{N}k_0n} + e^{-j\frac{2\pi}{N}k_0n}}{2} \right] x(n)$$

If $x(n) \xleftrightarrow{\text{DFT}} X(k)$

using circular frequency shift property,

$$e^{j\frac{2\pi}{N}k_0n} x(n) \xleftrightarrow{\text{DFT}} X((k - k_0))_N$$

$$e^{-j\frac{2\pi}{N}k_0n} x(n) \xleftrightarrow{\text{DFT}} X((k + k_0))_N$$

Hence DFT $\{x_1(n)\} = X_1(k)$

and
$$X_1(k) = \frac{1}{2} \left[X((k - k_0))_N + X((k + k_0))_N \right]$$

(20) Find N-point DFT of $x_2(n) = \sin\left(\frac{2\pi}{N}k_0n\right)x(n)$

ANS :
$$X_2(k) = \frac{1}{2j} \left[X((k - k_0))_N - X((k + k_0))_N \right]$$

Also, $x(n) \cos\left(\frac{4\pi}{N}n\right) \xleftrightarrow{\text{DFT}_N} \frac{1}{2} \left[X((k - 2))_N + X((k + 2))_N \right]$

$$x(n) \sin\left(\frac{8\pi}{N}n\right) \xleftrightarrow{\text{DFT}_N} \frac{1}{2j} \left[X((k - 4))_N - X((k + 4))_N \right]$$

6) CIRCULAR CONVOLUTION / TIME-DOMAIN CONVOLUTION OR FREQUENCY DOMAIN MULTIPLICATION

STATEMENT:

$$\text{If } x(n) \xleftrightarrow{\text{DFT}} X(k)$$

$$\text{& } h(n) \xleftrightarrow{\text{DFT}} H(k)$$

$$\text{then } y(n) = x(n) \circledast h(n) \xleftrightarrow{\text{DFT}} X(k) H(k) = Y(k)$$

Proof:

Consider two sequences $x(n)$ & $h(n)$ of length N , then circular convolution is defined as,

$$y(n) = x(n) \circledast h(n) = \sum_{m=0}^{N-1} x(m) h((n-m))_N$$

$$\text{DFT } \{y(n)\} = Y(k) = \sum_{n=0}^{N-1} y(n) W_N^{kn}$$

$$= \sum_{n=0}^{N-1} \sum_{m=0}^{N-1} x(m) h((n-m))_N \cdot W_N^{kn}$$

Interchanging order of summations,

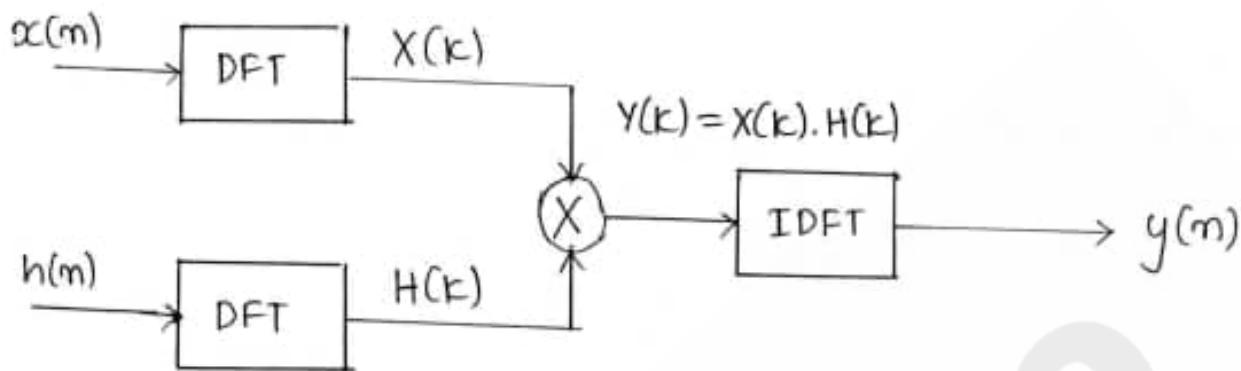
$$= \sum_{m=0}^{N-1} x(m) \sum_{n=0}^{N-1} h((n-m))_N W_N^{kn}$$

$$= \sum_{m=0}^{N-1} x(m) W_N^{km} H(k) \quad \left| \begin{array}{l} \text{Using Time-shift} \\ \text{property,} \\ h((n-m))_N \xleftrightarrow{\text{DFT}} \\ W_N^{km} H(k) \end{array} \right.$$

$$= X(k) H(k)$$

Hence the proof.

II METHOD : Circular convolution using DFT & IDFT
 (stockham's method)



The method involves taking the N point DFTs of $x(n)$ & $h(n)$ both of length N points.

The respective DFTs are multiplied elementwise.

Then taking IDFT of the sequence $y(k)$ to obtain $y(n)$.

- ② Compute the circular convolution of sequences
 $x(n) = \{1, 2, 3, 4\}$ and $h(n) = \{1, 2, 2\}$.

Soln:- Given $x(n) = \{1, 2, 3, 4\}$

& $h(n) = \{1, 2, 2\}$

Length of $x(n)$, $N_1 = 4$. Length of $h(n)$, $N_2 = 3$.

The convolution length $N = \max(N_1, N_2)$
 $= \max(4, 3)$

$N = 4$

Since $h(n)$ is of length 3, pad one zero.

$$\therefore h(n) = \{1, 2, 2, 0\}$$

I-METHOD: Time-Domain Approach or Concentric circle Method

From the def'n of circular convolution,

$$y(n) = x(n) \circledast h(n) = \sum_{m=0}^{N-1} x(m) h((n-m))_N$$

where $n = 0$ to $N-1$.

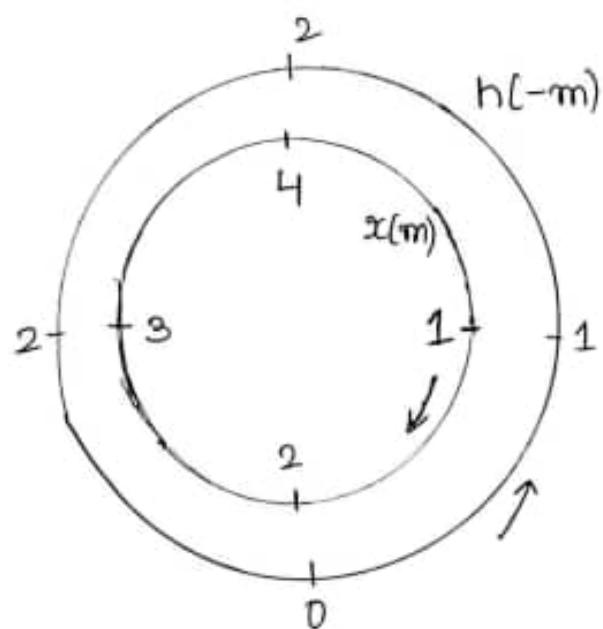
Given $N=4$,

$$\begin{aligned}\therefore y(n) &= \sum_{m=0}^3 x(m) h((n-m))_4, \quad m=0, 1, 2, 3 \\ &= x(0) h((n))_4 + x(1) h((n-1))_4 \\ &\quad + x(2) h((n-2))_4 + x(3) h((n-3))_4\end{aligned}$$

$$\text{at } n=0, \quad y(0) = x(0) h(0) + x(1) h(-1)_4 + x(2) h(-2)_4 \\ + x(3) h(-3)_4$$

$$\begin{aligned}y(0) &= x(0) h(0) + x(1) h(3) \\ &\quad + x(2) h(2) + x(3) h(1) \\ &= (1)(1) + (2)(0) + (3)(2) \\ &\quad + (4)(2) \\ &= 1 + 0 + 6 + 8\end{aligned}$$

$$y(0) = \underline{\underline{15}}$$

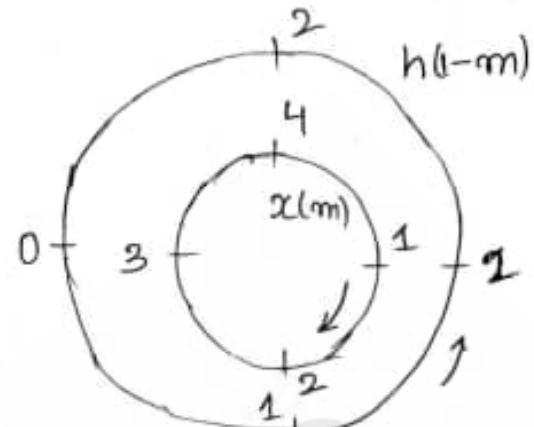


$$\text{at } m=1, \quad y(1) = x(0) h(1)_4 + x(1) h(0) + x(2) h(3)$$

$$+ x(3) h(2)$$

$$y(1) = 2 + 2 + 0 + 8$$

$$y(1) = \underline{\underline{12}}$$

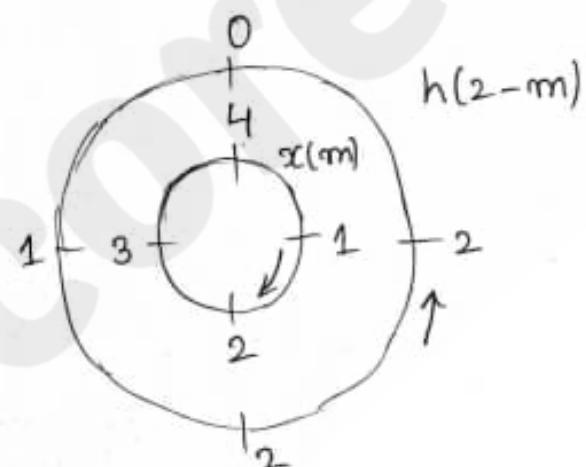


$$\text{at } m=2, \quad y(2) = x(0) h(2) + x(1) h(1) + x(2) h(0)$$

$$+ x(3) h(3)$$

$$y(2) = 2 + 4 + 3 + 0$$

$$y(2) = \underline{\underline{9}}$$

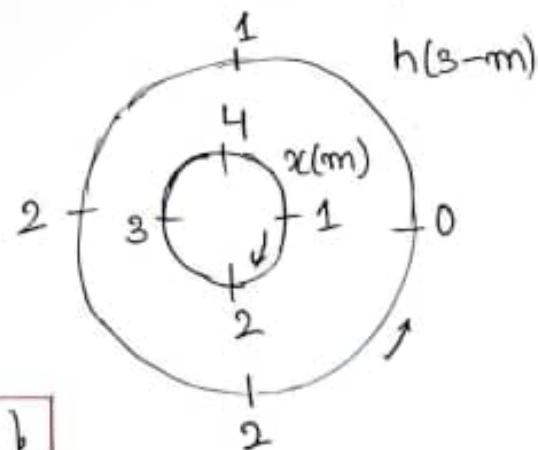


$$\text{at } m=3, \quad y(3) = x(0) h(3) + x(1) h(2) + x(2) h(1)$$

$$+ x(3) h(0)$$

$$y(3) = 0 + 4 + 6 + 4$$

$$y(3) = \underline{\underline{14}}$$



$$\therefore y(m) = \{ 15, 12, 9, 14 \}$$

Verification:

$$\begin{bmatrix} 1 & 4 & 3 & 2 \\ 2 & 1 & 4 & 3 \\ 3 & 2 & 1 & 4 \\ 4 & 3 & 2 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 2 \\ 0 \end{bmatrix} = \begin{bmatrix} 15 \\ 12 \\ 9 \\ 14 \end{bmatrix}$$

II - METHOD : Using DFT & IDFT equations or
Transform Domain Approach or Stockham's method

$$y(n) = x(n) \circledast h(n)$$

Applying DFT,

$$Y(k) = X(k) \cdot H(k)$$

(i) To find $X(k)$. $X_N = x_N W_N$

$$X(k) = [1 \ 2 \ 3 \ 4] \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$X(k) = \{10, -2+2j, -2, -2-2j\}$$

(ii) To find $H(k)$. $H_N = h_N W_N$

$$H(k) = [1 \ 2 \ 2 \ 0] \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$H(k) = \{5, -1-2j, 1, -1+2j\}$$

(iii) Find the product of the sequence $x(k)$ & $h(k)$.

$$y(k) = x(k) \cdot h(k)$$

$$y(k) = \{ 50, 6+2j, -2, 6-2j \}$$

(iv) To find $y(n)$

$$y(n) = IDFT \{ y(k) \} = \frac{1}{N} Y_N W_N^*$$

$$y(n) = \frac{1}{4} [50 \quad 6+2j \quad -2 \quad 6-2j] \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix}$$

$$y(n) = \{ 15, 12, 9, 14 \}$$

7) MULTIPLICATION OF TWO SEQUENCES

OR MODULATION PROPERTY OR CONVOLUTION
IN FREQUENCY DOMAIN

STATEMENT: If $x_1(n) \xleftrightarrow{\text{DFT}} X_1(k)$

& $x_2(n) \xleftrightarrow{\text{DFT}} X_2(k)$

then $x_1(n)x_2(n) \xleftrightarrow{\text{DFT}} \frac{1}{N} [X_1(k) \odot X_2(k)]$

Proof:— Consider two sequences $x_1(n)$ & $x_2(n)$ of length N , with DFT $X_1(k)$ & $X_2(k)$.

$$\text{IDFT } \{X_1(k)\} = x_1(n) = \frac{1}{N} \sum_{l=0}^{N-1} X_1(l) W_N^{-ln}$$

$$\text{IDFT } \{X_2(k)\} = x_2(n) = \frac{1}{N} \sum_{m=0}^{N-1} X_2(m) W_N^{-mn}$$

Let $y(n) = x_1(n) \cdot x_2(n)$

then,

$$\begin{aligned} \text{DFT } \{y(n)\} &= Y(k) = \sum_{n=0}^{N-1} y(n) W_N^{kn} \\ &= \sum_{n=0}^{N-1} x_1(n) x_2(n) W_N^{kn} \end{aligned}$$

$$= \sum_{n=0}^{N-1} \left\{ \frac{1}{N} \sum_{l=0}^{N-1} X_1(l) W_N^{-ln} \right\} \left\{ \frac{1}{N} \sum_{m=0}^{N-1} X_2(m) W_N^{-mn} \right\}$$

$\times W_N^{km}$

$$= \frac{1}{N^2} \sum_{l=0}^{N-1} X_1(l) \sum_{m=0}^{N-1} X_2(m) \sum_{n=0}^{N-1} W_N^{(k-l-m)n}$$

$$\text{If, } k-l-m = PN$$

$$\text{or } m = k-l-PN = ((k-l))_N$$

$$\text{or } l = k-m-PN = ((k-m))_N$$

then,

$$\sum_{n=0}^{N-1} W_N^{(k-l-m)n} = \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N} PN n}$$

$$= \sum_{n=0}^{N-1} 1 = N$$

$$\therefore Y(k) = \frac{1}{N^2} \sum_{l=0}^{N-1} X_1(l) X_2((k-l))_N \cdot N$$

$$Y(k) = \frac{1}{N} [X_1(k) \textcircled{\texttimes} X_2(k)]$$

8) SYMMETRY PROPERTY OF A COMPLEX VALUED SEQUENCE

STATEMENT: If $x(n) \xrightarrow{\text{DFT}} X(k)$

$$\text{then } x^*(n) \xrightarrow{\text{DFT}} X^*(N-k) = X^*((-k))_N$$

$$\text{or } x^*(-n)_N = x^*(N-n) \xrightarrow{\text{DFT}} X^*(k)$$

Proof: From def'm of DFT, we know that

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) w_N^{kn}$$

Taking complex conjugate on both sides,

$$X^*(k) = \sum_{n=0}^{N-1} x^*(n) w_N^{-kn}$$

replace k by N-k,

$$\begin{aligned} X^*(N-k) &= \sum_{n=0}^{N-1} x^*(n) w_N^{-(N-k)n} \\ &= \sum_{n=0}^{N-1} x^*(n) w_N^{-Nm} \cdot w_N^{km} \\ &= \sum_{n=0}^{N-1} x^*(n) w_N^{km} \quad \because w_N^{-Nm} = 1 \end{aligned}$$

$$X^*(N-k) = \text{DFT } \{x^*(n)\}$$

$$\text{or } \boxed{x^*(n) \xrightarrow[N]{\text{DFT}} X^*(N-k)}$$

9) SYMMETRY PROPERTY OF REAL-VALUED SEQUENCE

STATEMENT: If $x(n) \xleftrightarrow{\text{DFT}} X(k)$

and $x(n)$ is real, then

$$X(k) = X^*(N-k)$$

Proof:- W.K.T, $X(k) = \sum_{n=0}^{N-1} x(n) w_N^{kn}$

Taking complex conjugate on both sides,

$$X^*(k) = \sum_{n=0}^{N-1} x^*(n) w_N^{-kn}$$

Since $x(n)$ is real, $x^*(n) = x(n)$

$$\begin{aligned} X^*(k) &= \sum_{n=0}^{N-1} x(n) w_N^{-kn} \\ &= \sum_{n=0}^{N-1} x(n) w_N^{-kn} \cdot w_N^{Nm} \quad \because w_N^{Nm} = 1 \\ &= \sum_{n=0}^{N-1} x(n) w_N^{(N-k)m} \end{aligned}$$

$$X^*(k) = X(N-k)$$

or
$$\boxed{X(k) = X^*(N-k)}$$

(22) The first five points of the 8-point DFT of a real-valued sequence are

$$\{ 0.25, (0.125 - j0.3018), 0, (0.125 - j0.0518), 0 \}$$

Determining the remaining 3 points.

Soln: Given,

$$x(0) = 0.25$$

$$x(3) = 0.125 - j0.518$$

$$x(1) = 0.125 - j0.3018$$

$$x(4) = 0$$

$$x(2) = 0$$

since $x(n)$ is real valued

$$x^*(k) = x(N-k)$$

$$\text{or } x(k) = x^*(N-k)$$

Here, $N=8$,

$$\text{Then, for } k=5, x(5) = x^*(8-5) = x^*(3) \\ = (0.125 - j0.0518)^*$$

$$x(5) = \underline{\underline{0.125 + j0.0518}}$$

$$\text{for } k=6, x(6) = x^*(8-6) = x^*(2) \\ = \underline{\underline{0}}$$

$$\text{for } k=7, x(7) = x^*(8-7) = x^*(1) \\ = (0.125 - j0.3018)^* \\ = \underline{\underline{0.125 + j0.3018}}$$

10) CIRCULAR CORRELATION

For complex valued sequences $x(n)$ and $y(n)$

$$\text{If } x(n) \xleftrightarrow{\text{DFT}} X(k)$$

$$y(n) \xleftrightarrow{\text{DFT}} Y(k)$$

$$\text{then } \tilde{r_{xy}}(l) \xleftrightarrow{\text{DFT}} \tilde{R_{xy}}(l) = X(k) \cdot Y^*(k)$$

where $\tilde{r_{xy}}(l)$ is the circular cross-correlation sequence defined by

$$\tilde{r_{xy}}(l) = \sum_{n=0}^{N-1} x(n) y^*((n-l))_N$$

Proof: We can express $\tilde{r_{xy}}(l)$ as the circular convolution of $x(n)$ with $y^*(-n)$.

$$\text{i.e., } \tilde{r_{xy}}(l) = x(l) \circledast y^*(-l)$$

taking DFT on both sides,

$$\text{DFT} \left\{ \tilde{r_{xy}}(l) \right\} = \text{DFT} \left\{ x(l) \circledast y^*(-l) \right\}$$

$$\boxed{\tilde{R_{xy}}(k) = X(k) \cdot Y^*(k)}$$

using circular convolution & complex conjugate property

If $x(n) = y(n)$,

$$\tilde{r_{xx}}(l) \xleftrightarrow{\text{DFT}} \tilde{R_{xx}}(k) = |X(k)|^2$$

11) PARSEVAL'S THEOREM

For complex valued sequence $x(n)$ & $y(n)$

$$\text{If } x(n) \xleftrightarrow{\text{DFT}} X(k)$$

$$y(n) \xleftrightarrow{\text{DFT}} Y(k)$$

then $\sum_{n=0}^{N-1} x(n) \cdot y^*(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) \cdot Y^*(k)$

Proof:- Circular cross correlation of two sequences

is defined as

$$\tilde{r}_{xy}(l) = \sum_{m=0}^{N-1} x(m) y^*((m-l))_N$$

at $l=0$, $\tilde{r}_{xy}(0) = \sum_{m=0}^{N-1} x(m) y^*(m)$ —①

W.K.T, IDFT $\{ \tilde{R}_{xy}(k) \} = \tilde{r}_{xy}(l)$

or $\tilde{r}_{xy}(l) = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{R}_{xy}(k) e^{\frac{j2\pi}{N} kl}$ from def'n of IDFT

$$= \frac{1}{N} \sum_{k=0}^{N-1} X(k) Y^*(k) e^{\frac{j2\pi}{N} kl}$$

from circular correlation property

at $l=0$,

$$\tilde{r}_{xy}(0) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) Y^*(k)$$
 —②

Equating RHS of eqn's @ & ⑥,

$$\sum_{m=0}^{N-1} x(m) y^*(m) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) Y^*(k)$$

if $x(m) = y(m)$, then

$$\sum_{m=0}^{N-1} |x(m)|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X(k)|^2$$

which is the energy in the finite duration sequence $x(m)$ in terms of the frequency components $X(k)$.

(23) Determine N-point circular correlation of

$$x(m) = \cos \frac{2\pi}{N} m \quad \& \quad y(m) = \sin \frac{2\pi}{N} m.$$

Soln:- Given $x(m) = \cos \frac{2\pi m}{N}$

Taking N-point DFT,

$$X(k) = \frac{N}{2} [\delta(k-1) + \delta(k+1)]$$

$$\text{Given } y(m) = \sin \frac{2\pi m}{N}$$

$$\text{Its N-point DFT, } Y(k) = \frac{N}{2j} [\delta(k-1) - \delta(k+1)]$$

From circular correlation property,

$$\tilde{R_{xy}}(k) = X(k) \cdot Y^*(k)$$

$$\begin{aligned} \tilde{R_{xy}}(k) &= \frac{N}{2} [s(k-1) + s(k+1)] \cdot \left\{ \frac{N}{2j} [s(k-1) - s(k+1)] \right\}^* \\ &= \frac{N}{2} [s(k-1) + s(k+1)] \left\{ -\frac{N}{2j} [s(k-1) - s(k+1)] \right\} \\ &= -\frac{N^2}{4j} \left\{ [s(k-1)]^2 + s(k+1)s(k-1) - s(k-1)s(k+1) \right. \\ &\quad \left. - [s(k+1)]^2 \right\} \end{aligned}$$

W.E.T., $s(k+1)s(k-1) = 0$, $[s(k-1)]^2 = s(k-1)$
and $[s(k+1)]^2 = s(k+1)$.

$$= -\frac{N^2}{4j} [s(k-1) - s(k+1)]$$

$$\tilde{R_{xy}}(k) = -\frac{N}{2} \left\{ \frac{N}{2j} [s(k-1) - s(k+1)] \right\}$$

taking IDFT,

$$Y_{xy}(l) = -\frac{N}{2} \sin \frac{2\pi n}{N}$$

(24) Find the circular autocorrelation of $x(n) = \cos \frac{2\pi}{N} n$

Soln:- Given $x(n) = \cos \left(\frac{2\pi}{N} n \right)$

Taking DFT,

$$X(k) = \frac{N}{2} [\delta(k-1) + \delta(k+1)]$$

WKT $\tilde{R_{xy}}(k) = X(k) \cdot Y^*(k)$

since $x(n) = y(n)$,

$$\tilde{R_{xx}}(k) = X(k) \cdot X^*(k)$$

$$\tilde{R_{xx}}(k) = \left\{ \frac{N}{2} [\delta(k-1) + \delta(k+1)] \right\} \left\{ \frac{N}{2} [\delta(k-1) + \delta(k+1)] \right\}^*$$

$$= \frac{N}{2} \cdot \frac{N}{2} [\delta(k-1) + \delta(k+1)]$$

taking IDFT,

$$\tilde{R_{xx}}(n) = \frac{N}{2} \cos \left(\frac{2\pi}{N} n \right)$$

(25) Find circular autocorrelation of the sequence
 $x(n) = \{1, 2, 3, 4\}.$

Soln:- Given $x(n) = \{1, 2, 3, 4\}$, $N=4$

Find the 4-point DFT,

$$\text{we get } X(k) = \{10, -2+2j, -2, -2-2j\}$$

Autocorrelation of the sequence $x(n)$ is

$$\begin{aligned} \tilde{r}_{xx}(l) &\xleftarrow{\text{DFT}} \tilde{R}_{xx}(k) = X(k) \cdot X^*(k) \\ &= |X(k)|^2 \end{aligned}$$

$$\tilde{R}_{xx}(0) = |X(0)|^2 = 10^2 = 100$$

$$\begin{aligned} \tilde{R}_{xx}(1) &= |X(1)|^2 = |(-2+2j)|^2 \\ &= |\sqrt{4+4}|^2 = 8 \end{aligned}$$

$$\tilde{R}_{xx}(2) = |X(2)|^2 = |-2|^2 = 4$$

$$\tilde{R}_{xx}(3) = |X(3)|^2 = |(-2-2j)|^2 = 8$$

$$\therefore \boxed{\tilde{R}_{xx}(k) = \{100, 8, 4, 8\}}$$

To find $\tilde{r}_{xx}(l)$:

$$\tilde{r}_{xx}(l) = \text{IDFT} \left\{ \tilde{R}_{xx}(k) \right\}$$

$$\tilde{r}_{xx}(l) = \frac{1}{4} \begin{bmatrix} 100 & 8 & 4 & 8 \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix}$$

$$= \frac{1}{4} \begin{bmatrix} 120 & 96 & 88 & 96 \end{bmatrix}$$

$$\boxed{\tilde{r}_{xx}(l) = \{ 30, 24, 22, 24 \}}$$

(26) Find the circular correlation given

$$x(n) = \{ 1, 2, 3, 4 \} \text{ and } y(n) = \{ 1, 2, 2, 0 \}$$

Soln:- Find $X(k)$

$$X(k) = \{ 10, -2+2j, -2, -2-2j \}$$

Find $Y(k)$

$$Y(k) = \{ 5, -1-2j, 1, -1+2j \}$$

$$\Rightarrow Y^*(k) = \{ 5, -1+2j, 1, -1-2j \}$$

circular correlation property,

$$R_{xy}(\kappa) = x(\kappa) \cdot y^*(\kappa)$$

$$= \{ 50, -2-6j, -2, -2+6j \}$$

Taking IDFT,

$$r_{xy}(\ell) = \{ 11, 16, 13, 10 \}$$

- Q27 Given $x(n) = \{ 1, 2, 3, 4 \}$ find the energy
and hence verify Parseval's theorem.

Soln:- Given $x(n) = \{ 1, 2, 3, 4 \}$

Energy of the signal $x(n)$ is given by

$$E = \sum_{n=-\infty}^{+\infty} |x(n)|^2$$

$$= \sum_{n=0}^3 |x(n)|^2 = |x(0)|^2 + |x(1)|^2 + |x(2)|^2 + |x(3)|^2$$

$$= 1^2 + 2^2 + 3^2 + 4^2$$

$$E = 30$$

Find DFT of $x(n)$,

$$X(k) = \{ 10, -2+2j, -2, -2-2j \}$$

From Parseval's Theorem,

$$\sum_{n=0}^{N-1} |x(n)|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X(k)|^2$$

$$E = \frac{1}{4} \sum_{k=0}^3 |X(k)|^2$$

$$= \frac{1}{4} \left[|X(0)|^2 + |X(1)|^2 + |X(2)|^2 + |X(3)|^2 \right]$$

$$= \frac{1}{4} [100 + 8 + 4 + 8]$$

$$E = \frac{120}{4} = \underline{\underline{30}}$$

Hence Parseval's Theorem is verified.

Definition of Even and Odd Symmetry and time reversal.

An N -point sequence is said to be circularly even if it is symmetric about the point zero on the circle.

$$\text{ie, } x(N-m) = x(m), \quad 1 \leq m \leq N-1$$

An N -point sequence is said to be circularly odd if it is antisymmetric about the point zero on the circle.

$$\text{ie, } x(N-m) = -x(m), \quad 1 \leq m \leq N-1$$

The time reversal of an N -point sequence is attained by reversing its samples about the point zero on the circle. Thus the sequence $x((-n))_N$ is given by

$$x((-n))_N = x(N-n), \quad 0 \leq n \leq N-1$$

For periodic sequence $x_p(n)$, def'n of even and odd sequences :

$$\text{even: } x_p(n) = x_p(-n) = x_p(N-n)$$

$$\text{odd: } x_p(n) = -x_p(-n) = -x_p(N-n)$$

If $x_p(n)$ is complex valued, then

$$\text{conjugate even: } x_p(n) = x_p^*(N-n)$$

$$\text{conjugate odd: } x_p(n) = -x_p^*(N-n)$$

Sequence $x_p(n)$ can be decomposed as

$$x_p(n) = x_{pe}(n) + x_{po}(n)$$

where

$$\boxed{x_{pe}(n) = \frac{1}{2} [x_p(n) + x_p^*(N-n)]}$$

$$\boxed{x_{po}(n) = \frac{1}{2} [x_p(n) - x_p^*(N-n)]}$$

Symmetry properties of the DFT

Let us assume that $x(n)$ and its DFT $X(k)$ are both complex valued.

Then, $x(n) = x_R(n) + j x_I(n)$, $n=0, 1, \dots, N-1$ ①

$X(k) = X_R(k) + j X_I(k)$, $k=0, 1, \dots, N-1$ ②

From the def'n of DFT,

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad \text{--- ③}$$

Substituting eqn ① in eqn ③,

$$X(k) = \sum_{n=0}^{N-1} [x_R(n) + j x_I(n)] e^{-j \frac{2\pi}{N} kn}$$

and since, $e^{j\theta} = \cos\theta + j \sin\theta$

$$\begin{aligned}
 X(k) &= \sum_{n=0}^{N-1} \left[x_R(n) + j x_I(n) \right] \left[\cos \frac{2\pi}{N} kn - j \sin \frac{2\pi}{N} kn \right] \\
 &= \sum_{n=0}^{N-1} \left[x_R(n) \cos \left(\frac{2\pi}{N} kn \right) - j x_R(n) \sin \left(\frac{2\pi}{N} kn \right) \right. \\
 &\quad \left. + j x_I(n) \cos \left(\frac{2\pi}{N} kn \right) + x_I(n) \sin \left(\frac{2\pi}{N} kn \right) \right]
 \end{aligned}$$

L(4)

Comparing eqn's (2) & (4),

$$X_R(k) = \sum_{m=0}^{N-1} \left[x_R(m) \cos \left(\frac{2\pi}{N} km \right) + x_I(m) \sin \left(\frac{2\pi}{N} km \right) \right]$$

L(5)

$$X_I(k) = \sum_{m=0}^{N-1} \left[-x_R(m) \sin \left(\frac{2\pi}{N} km \right) + x_I(m) \cos \left(\frac{2\pi}{N} km \right) \right]$$

L(6)

IIIrd, WKT $x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) w_N^{-kn}$

Substituting eqn (2) in above equation,

$$\begin{aligned}
 x(n) &= \frac{1}{N} \sum_{k=0}^{N-1} \left[X_R(k) + j X_I(k) \right] e^{j \frac{2\pi}{N} kn} \\
 &= \frac{1}{N} \sum_{k=0}^{N-1} \left[X_R(k) + j X_I(k) \right] \left[\cos \left(\frac{2\pi}{N} kn \right) + j \sin \left(\frac{2\pi}{N} kn \right) \right]
 \end{aligned}$$

$$= \frac{1}{N} \sum_{k=0}^{N-1} \left[X_R(k) \cos\left(\frac{2\pi}{N} km\right) + j X_I(k) \sin\left(\frac{2\pi}{N} km\right) \right. \\ \left. + j X_I(k) \cos\left(\frac{2\pi}{N} km\right) - X_I(k) \sin\left(\frac{2\pi}{N} km\right) \right]$$

L(7)

Comparing eqns (1) & (7),

$$x_R(n) = \frac{1}{N} \sum_{k=0}^{N-1} \left[X_R(k) \cos\left(\frac{2\pi}{N} km\right) - X_I(k) \sin\left(\frac{2\pi}{N} km\right) \right]$$

L(8)

$$x_I(n) = \frac{1}{N} \sum_{k=0}^{N-1} \left[X_R(k) \sin\left(\frac{2\pi}{N} km\right) + X_I(k) \cos\left(\frac{2\pi}{N} km\right) \right]$$

L(9)

case 1) Real Valued sequences

If $x(n)$ is real, then

$$X(n-k) = X^*(k) = X(-k)$$

And, $|X(n-k)| = |X(k)|$ also $\angle X(n-k) = -\angle X(k)$

For real $x(n)$, $X_I(n) = 0$, hence $x(n)$ can be determined from eqn (8).

Case 2) Real and even sequence

If $x(n)$ is real and even, that is

$$x(n) = x(N-n), \quad 0 \leq n \leq N-1$$

then eqn.⑥ yields $X_I(k) = 0$. Hence the DFT reduces to

$$X(k) = \sum_{m=0}^{N-1} x(m) \cos\left(\frac{2\pi km}{N}\right), \quad 0 \leq k \leq N-1$$

where $X(k)$ is real valued and even.

since $X_I(k) = 0$, the IDFT reduces to

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) \cos\left(\frac{2\pi kn}{N}\right), \quad 0 \leq n \leq N-1$$

Case 3) Real and odd sequence

If $x(n)$ is real and odd, that is

$$x(n) = -x(N-n), \quad 0 \leq n \leq N-1$$

then eqn.⑤ yields $X_R(k) = 0$. Hence

$$X(k) = -j \sum_{m=0}^{N-1} x(m) \sin\left(\frac{2\pi km}{N}\right), \quad 0 \leq k \leq N-1$$

which is purely imaginary and odd.

since $X_R(k) = 0$, the IDFT reduces to

$$x(n) = j \frac{1}{N} \sum_{k=0}^{N-1} X(k) \sin\left(\frac{2\pi kn}{N}\right), \quad 0 \leq n \leq N-1$$

Case 4) Purely Imaginary sequence

In this case $x(n) = j x_I(n)$.

Eqs (5) & (6) reduces to

$$X_R(k) = \sum_{n=0}^{N-1} x_I(n) \sin \frac{2\pi kn}{N}$$

$$X_I(k) = \sum_{n=0}^{N-1} x_I(n) \cos \frac{2\pi kn}{N}$$

We observe that $X_R(k)$ is odd and $X_I(k)$ is even.

If $x_I(n)$ is odd, then $X_I(k) = 0$, hence $X(k)$ is purely real.

If $x_I(n)$ is even, then $X_R(k) = 0$, hence $X(k)$ is purely imaginary.

The symmetry properties above may be summarised as follows :

$$\begin{aligned} x(n) &= x_R^e(n) + x_R^o(n) + j x_I^e(n) + j x_I^o(n) \\ X(k) &= X_R(k) + X_R^o(k) + j X_I^e(k) + j X_I^o(k) \end{aligned}$$

PROPERTIES OF THE DFT

PROPERTY	TIME DOMAIN	FREQUENCY DOMAIN
Periodicity	$x(n) = x(n+N)$	$X(k) = X(k+N)$
Linearity	$a_1 x_1(n) + a_2 x_2(n)$	$a_1 X_1(k) + a_2 X_2(k)$
Time Reversal	$x(N-n)$	$X(N-k)$
Circular Time Shift	$x((n-l))_N$	$X(k) \cdot e^{-j\frac{2\pi}{N}kl}$
Circular Frequency shift	$x(n) e^{j\frac{2\pi}{N}lm}$	$X((k-l))_N$
Complex Conjugate	$x^*(n)$	$X^*(N-k)$
Circular Convolution	$x_1(n) \textcircled{N} x_2(n)$	$X_1(k) X_2(k)$
Circular Correlation	$x(n) \textcircled{N} y^*(-n)$	$X(k) Y^*(k)$
Multiplication of two sequences	$x_1(n) x_2(n)$	$\frac{1}{N} [X_1(k) \textcircled{N} X_2(k)]$
Parseval's Theorem	$\sum_{n=0}^{N-1} x(n) y^*(n)$	$\frac{1}{N} \sum_{k=0}^{N-1} X(k) Y^*(k)$

- (28) A 498 point DFT of a real valued sequence $x(n)$ has the following DFT samples

$$X(0) = 2$$

$$X(249) = 2.9$$

$$X(11) = 7 + j3.1$$

$$X(309) = -4.7 - j1.9$$

$$X(k_1) = -2.2 - j1.5$$

$$X(k_3) = 3 - j0.7$$

$$X(112) = 3 + j0.7$$

$$X(412) = -2.2 + j1.5$$

$$X(k_2) = -4.7 + j1.9$$

$$X(k_4) = 7 - j3.1$$

The other samples have a value zero. Find the value of k_1, k_2, k_3 & k_4 .

Sol'n:- For real valued sequence,

$$\text{WKT, } X(k) = X^*(N-k)$$

Given $N = 498$.

a) For $k = 112$, $X(412) = X^*(498 - 412)$

$$= X^*(86)$$

$$\text{or } X(86) = X^*(412)$$

$$= (-2.2 + j1.5)^*$$

$$X(86) = -2.2 - j1.5 = X(k_1)$$

$$\therefore \boxed{k_1 = 86}$$

b) for $k = 309$, $X(309) = X^*(498 - 309)$
 $= X^*(189)$

or $X(189) = X^*(309)$
 $= (-4.7 - j1.9)^*$

$$X(189) = -4.7 + j1.9 = X(k_2)$$

$\therefore \boxed{k_2 = 189}$

c) for $k = 112$, $X(112) = X^*(498 - 112)$
 $= X^*(386)$

or $X(386) = X^*(112)$
 $= (3 + j0.7)^*$

$$X(386) = 3 - j0.7 = X(k_3)$$

$\therefore \boxed{k_3 = 386}$

d) for $k = 11$, $X(11) = X^*(498 - 11)$
 $= X^*(487)$

or $X(487) = X^*(11)$
 $= (7 + j3.1)^*$

$$X(487) = 7 - j3.1 = X(k_4)$$

$\therefore \boxed{k_4 = 487}$

Hence, $k_1 = 86$, $k_2 = 386$, $k_3 = 189$ & $k_4 = 487$

- (29) Find the 4-point DFTs of the two sequences $x(n)$ & $y(n)$ using a single 4-point DFT.
 $x(n) = \{1, 2, 0, 1\}$ & $y(n) = \{2, 2, 1, 1\}$.

Soln:-

From the given, combining the two sequences $x(n)$ and $y(n)$ to create a complex sequence $h(n)$.

$$\text{With, } h(n) = x(n) + jy(n), 0 \leq n \leq 3$$

$$\therefore h(n) = \{(1+2j), (2+2j), j, (1+j)\}$$

$$\text{Taking DFT, } H_N = h_N W_N$$

$$\therefore H(k) = \begin{bmatrix} 1+2j & 2+2j & j & 1+j \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$H(k) = [4+6j \quad 2 \quad -2 \quad 2j]$$

$$\Rightarrow H^*(k) = [4-6j \quad 2 \quad -2 \quad -2j]$$

$$\text{Using the relation, } X(k) = \frac{H(k) + H^*(-k)}{2}$$

$$\text{for } k=0, \quad X(0) = \frac{H(0) + H^*(0)}{2} = \frac{(4+6j) + (4-6j)}{2} \\ = \frac{8}{2} = \underline{\underline{4}}$$

$$\text{for } k=1, \quad X(1) = \frac{H(1) + H^*(-1)}{2} = \frac{H(1) + H^*(3)}{2} \\ = \frac{2-2j}{2} = \underline{\underline{1-j}}$$

$$\text{for } k=2, \quad X(2) = \frac{H(2) + H^*(-2))_4}{2} = \frac{H(2) + H^*(2)}{2}$$

$$= \frac{-2 - 2}{2} = \underline{\underline{-2}}$$

$$\text{for } k=3, \quad X(3) = \frac{H(3) + H^*(-3))_4}{2} = \frac{H(3) + H^*(1)}{2}$$

$$= \frac{2j+2}{2} = \underline{\underline{1+j}}$$

$$\therefore X(k) = \left\{ 4, (1-j), -2, (1+j) \right\}$$

$$\text{Also, } Y(k) = \frac{H(k) - H^*(-k))_N}{2j}$$

$$\text{for } k=0, \quad Y(0) = \frac{H(0) - H^*(0)}{2j} = \frac{4+6j - 4+6j}{2j} = \underline{\underline{6}}$$

$$\text{for } k=1, \quad Y(1) = \frac{H(1) - H^*(-1))_4}{2j} = \frac{H(1) - H^*(3)}{2j}$$

$$= \frac{2+2j}{2j} = \underline{\underline{1-j}}$$

$$\text{for } k=2, \quad Y(2) = \frac{H(2) - H^*(-2))_4}{2j} = \frac{H(2) - H^*(2)}{2j}$$

$$= \frac{-2 + 2}{2j} = \underline{\underline{0}}$$

$$\text{for } k=3, \quad Y(3) = \frac{H(3) - H^*(-3))_4}{2j} = \frac{H(3) - H^*(1)}{2j}$$

$$= \frac{2j - 2}{2j} = \underline{\underline{1+j}}$$

$$\therefore Y(k) = \left\{ 6, (1-j), 0, (1+j) \right\}$$

- (30) Let $x_p(n)$ be a periodic sequence with fundamental period N . If the N -point DFT $\{x_p(n)\} = X_1(k)$ and $3N$ -point DFT $\{x_p(n)\} = X_3(k)$.
- Find the relationship between $X_1(k)$ & $X_3(k)$.
 - Verify the above result for $\{2, 1\} \leftarrow \{2, 1, 2, 1, 2, 1\}$.

Soln:- (i) WKT N -point DFT $\{x(n)\} = X(k)$

$$= \sum_{m=0}^{N-1} x(m) W_N^{km}$$

$$N\text{-point DFT } \{x_p(n)\} = X_1(k) = \sum_{m=0}^{N-1} x_p(m) W_N^{km}$$

$$3N\text{-point DFT } \{x_p(n)\} = X_3(k) = \sum_{m=0}^{3N-1} x_p(m) W_{3N}^{km}$$

$$X_3(k) = \sum_{m=0}^{N-1} x_p(m) W_{3N}^{km} + \sum_{m=0}^{N-1} x_p(m+N) W_{3N}^{k(m+N)}$$

$$+ \sum_{m=0}^{N-1} x_p(m+2N) W_{3N}^{k(m+2N)}$$

$$= \sum_{m=0}^{N-1} x_p(m) W_{3N}^{km} + \sum_{m=0}^{N-1} x_p(m) W_{3N}^{km} \cdot W_{3N}^{KN}$$

$$+ \sum_{m=0}^{N-1} x_p(m) W_{3N}^{km} \cdot W_{3N}^{2KN} \quad ; \text{ From periodicity of input.}$$

$$x_p(m+N) = x_p(m)$$

$$\therefore x_p(m+2N) = x_p(m)$$

$$\text{Also, } W_{3N}^{NK} = e^{-j\frac{2\pi}{3N}NK} = e^{-j\frac{2\pi}{3}K} = W_3^K$$

$$\text{Hence } W_{3N}^{2NK} = W_3^{2K} \quad \text{and} \quad W_{3N}^{NK} = W_N^{\frac{NK}{3}}.$$

$$\therefore X_3(K) = \sum_{n=0}^{N-1} x_p(n) [1 + W_3^K + W_3^{2K}] W_N^{\frac{nK}{3}}$$

$$= [1 + W_3^K + W_3^{2K}] \sum_{n=0}^{N-1} x_p(n) W_N^{\frac{nK}{3}}$$

$$X_3(K) = [1 + W_3^K + W_3^{2K}] X_1\left(\frac{K}{3}\right)$$

i) Given $x_1(n) = \{2, 1\}$

Its DFT, $X_1(K) = 2 + W_2^K$

$$x_3(n) = \{2, 1, 2, 1, 2, 1\} \quad \text{with DFT}$$

$$X_3(K) = 2 + W_6^K + 2W_6^{2K} + W_6^{3K} + 2W_6^{4K} + W_6^{5K}$$

$$W_6^K = W_{2 \times 3}^K = W_2^{K/3}$$

$$\text{Then, } X_3(K) = \left[2 + W_2^{K/3}\right] + W_6^{2K} \left[2 + W_2^{K/3}\right]$$

$$+ W_6^{4K} \left[2 + W_2^{K/3}\right]$$

$$= \left[2 + W_2^{K/3}\right] \left[1 + W_6^{2K} + W_6^{4K}\right]$$

$$X_3(K) = X_1\left(\frac{K}{3}\right) [1 + W_3^K + W_3^{2K}]$$

Hence verified.

(31) Suppose that you are given a program to find the DFT of a complex-valued sequence $x(n)$. How can we use this program to find the inverse DFT of $X(k)$.

Soln:- WKT,

$$\text{DFT } \{x(n)\} = X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad \text{--- (1)}$$

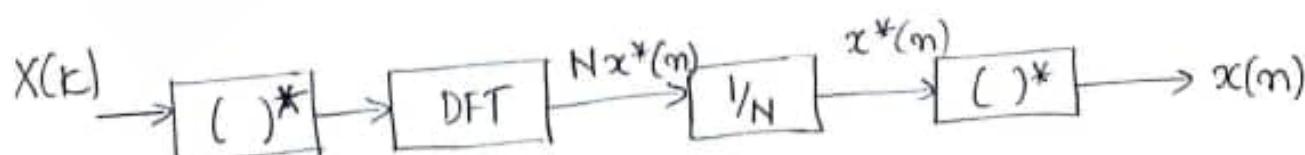
$$\text{IDFT } \{X(k)\} = x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} \quad \text{--- (2)}$$

Taking complex conjugate of eqn (2),

$$x^*(n) = \frac{1}{N} \sum_{k=0}^{N-1} X^*(k) W_N^{kn}$$

$$\text{or } N x^*(n) = \sum_{k=0}^{N-1} X^*(k) W_N^{kn}$$

From observing the above eqn, we can use DFT program to compute IDFT by calculating $X^*(k)$ and finding its DFT. The output is $N x^*(n)$. scale it by $\frac{1}{N}$ and take its conjugate.



- (32) Consider the sequence $x(n) = 4s(n) + 3s(n-1) + 2s(n-2) + s(n-3)$.

Let $X(k)$ be the 6-point DFT of $x(n)$. Find the finite length sequence $y(n)$ that has a 6-point DFT $Y(k) = W_6^{4k} X(k)$.

Soln:- Given, $x(n) = 4s(n) + 3s(n-1) + 2s(n-2) + s(n-3)$

Its DFT,

$$X(k) = 4 + 3W_6^k + 2W_6^{2k} + W_6^{3k}$$

Also, given $Y(k) = W_6^{4k} X(k)$

$$Y(k) = W_6^{4k} \left[4 + 3W_6^k + 2W_6^{2k} + W_6^{3k} \right]$$

$$= 4W_6^{4k} + 3W_6^{5k} + 2W_6^{6k} + W_6^{7k}$$

WKT, $W_6^{6k} = W_6^0 = 1$, $W_6^{7k} = W_6^k$

$$\therefore Y(k) = 4W_6^{4k} + 3W_6^{5k} + 2 + W_6^k$$

$$Y(k) = 2 + W_6^k + 4W_6^{4k} + 3W_6^{5k}$$

Taking IDFT,

$$y(n) = 2s(n) + s(n-1) + 4s(n-4) + 3s(n-5)$$

$$\therefore \boxed{y(n) = \{2, 1, 0, 0, 4, 3\}}$$

(33) If $x(n) = \{1, 2, 0, 3, -2, 4, 7, 5\}$,

Evaluate the following (i) $x(0)$ (ii) $x(4)$ (iii) $\sum_{k=0}^7 x(k)$

$$(iv) \sum_{k=0}^7 |x(k)|^2$$

Soln:- (i) From the def'n of DFT we have

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

$N=8$,

$$\text{at } k=0, \quad X(0) = \sum_{n=0}^7 x(n) W_8^n$$

$$= \sum_{n=0}^7 x(n) = 1 + 2 + 0 + 3 \\ -2 + 4 + 7 + 5$$

$$\boxed{X(0) = 20}$$

$$(ii) \quad X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

for $N=8$, Let $k=4$

$$X(4) = \sum_{n=0}^7 x(n) W_8^{4n}$$

$$\therefore W_8^{4n} = e^{\frac{-j2\pi}{8} \times 4n}$$

$$= \sum_{n=0}^7 x(n) (-1)^n$$

$$= (e^{-j\pi})^n = (-1)^n$$

$$X(4) = x(0) - x(1) + x(2) - x(3) + x(4) - x(5) \\ + x(6) - x(7)$$

$$\boxed{X(4) = -8}$$

iii) To find $\sum_{k=0}^7 x(k)$

Consider the IDFT equation,

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn}$$

for $N=8$, put $n=0$,

$$x(0) = \frac{1}{8} \sum_{k=0}^7 X(k)$$

$$\therefore \sum_{k=0}^7 X(k) = 8 x(0) = 8$$

iv) To find $\sum_{k=0}^7 |X(k)|^2$

From Parseval's Theorem,

$$\sum_{n=0}^{N-1} |x(n)|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X(k)|^2$$

or $\sum_{k=0}^{N-1} |X(k)|^2 = N \sum_{n=0}^{N-1} |x(n)|^2$

for $N=8$, $\sum_{k=0}^7 |X(k)|^2 = 8 \left[\sum_{n=0}^7 |x(n)|^2 \right]$

$$= 8 \left[|1|^2 + |2|^2 + |0|^2 + |3|^2 + |-2|^2 + |4|^2 + |7|^2 + |5|^2 \right]$$

$$= 8 [1 + 4 + 9 + 4 + 16 + 49 + 25]$$

$$\sum_{k=0}^7 |X(k)|^2 = \underline{\underline{864}}$$

(34) Let $x(n)$ be a finite length sequence with
 $x(k) = \{0, (1+j), 1, (1-j)\}$ using the properties
of DFT, find DFTs of the following sequences.

$$(i) x_1(n) = e^{j\frac{\pi}{2}n} x(n)$$

$$(ii) x_2(n) = \cos \frac{\pi}{2} n x(n)$$

$$(iii) x_3(n) = x((n-1))_4$$

Sol'n:- (i) Given $x_1(n) = e^{j\frac{\pi}{2}n} x(n)$

$$\text{or } x_1(n) = e^{j\frac{2\pi}{4}n} x(n)$$

Applying DFT and using circular frequency
shift property

$$X_1(k) = X((k-1))_4$$

$$\therefore X_1(k) = \{ (1-j), 0, (1+j), 1 \}$$

(ii) Given $x_2(n) = \cos \frac{\pi}{2} n x(n)$

$$\text{or } x_2(n) = \frac{1}{2} \left[e^{j\frac{\pi}{2}n} + e^{-j\frac{\pi}{2}n} \right] x(n)$$

$$= \frac{1}{2} \left[e^{j\frac{2\pi}{4}n} + e^{-j\frac{2\pi}{4}n} \right] x(n)$$

Applying DFT on both sides and using circular
frequency shift property.

$$X_2(k) = \frac{1}{2} [X((k-1))_4 + X((k+1))_4]$$

$$\text{Given } X(k) = \{ 0, (1+j), 1, (1-j) \}$$

$$X((k-1))_4 = \{ (1-j), 0, (1+j), 1 \}$$

$$X((k+1))_4 = \{ (1+j), 1, (1-j), 0 \}$$

$$\therefore X_2(k) = \frac{1}{2} [2 \ 1 \ 2 \ 1]$$

$$X_2(k) = \{ \underline{\underline{1, 0.5}}, \underline{\underline{1, 0.5}} \}$$

$$\text{iii) Given } x_3(n) = x((n-1))_4$$

taking DFT on both sides and using circular time-shift property.

$$X_3(k) = W_4^k X(k)$$

$$\text{at } k=0, X_3(0) = W_4^0 X(0) = 0$$

$$\text{at } k=1, X_3(1) = W_4^1 X(1) = (-j)(1+j) = (1-j)$$

$$\text{at } k=2, X_3(2) = W_4^2 X(2) = (-1)(1) = -1$$

$$\text{at } k=3, X_3(3) = W_4^3 X(3) = j(1-j) = 1+j$$

$$\therefore X_3(k) = \{ \underline{\underline{0, 1-j}}, \underline{\underline{-1, 1+j}} \}$$

(35) Let $x(n) = \{1, 2, 3, 4\}$ with $X(k) = \{10, -2+2j, -2, -2-2j\}$. Find the DFT of $x_1(n) = \{1, 0, 2, 0, 3, 0, 4, 0\}$ without actually calculating the DFT.

$$\text{Soln:-- DFT } \{x_1(n)\} = X_1(k) = \sum_{n=0}^{N-1} x_1(n) W_N^{kn}$$

Dividing above summation into even and odd parts.

$$X_1(k) = \sum_{m=0}^3 x_1(2m) W_N^{2mk} + \sum_{m=0}^3 x_1(2m+1) W_N^{(2m+1)k}$$

$N=8,$

$$X_1(k) = \sum_{m=0}^3 x_1(2m) W_8^{2mk} + 0 \quad \because \text{odd numbered samples are zero.}$$

$$X_1(k) = \sum_{m=0}^3 x_1(m) W_{8/2}^{mk}$$

$$X_1(k) = \sum_{m=0}^3 x_1(m) W_4^{mk} \quad \text{or} \quad X_1(k) = \begin{cases} X(k), & m=0, 1, 2, 3 \\ X(k), & m=4, 5, 6, 7 \end{cases}$$

$$X_1(k) = X(k)$$

$$\text{i.e., } X_1(k) = \{10, -2+2j, -2, -2-2j, 10, -2+2j, -2, -2-2j\}$$

Using periodicity property,

$$X_1(4) = X(4) = X(0) \dots$$

- (36) Let $X(k)$ be a 14-point DFT of a length-14 real sequence $x(n)$. The first 8 samples of $X(k)$ are given by

$$\begin{aligned} X(0) &= 12, & X(1) &= -1 + j3, & X(2) &= 3 + j4 \\ X(3) &= 1 - j5, & X(4) &= -2 + j2, & X(5) &= 6 + j3 \\ X(6) &= -2 - j3, & X(7) &= 10 \end{aligned}$$

Determine the remaining samples of $X(k)$.

Evaluate the following functions of $x(n)$ without computing the IDFT of $X(k)$.

$$\begin{array}{lll} \text{(i)} \quad x(0) & \text{(ii)} \quad x(7) & \text{(iii)} \quad \sum_{n=0}^{13} x(n) \\ \text{(iv)} \quad \sum_{n=0}^{13} e^{\frac{j4\pi n}{14}} x(n) & & \text{(v)} \quad \sum_{n=0}^{13} |x(n)|^2 \end{array}$$

Soln:- For a real valued sequence $x(n)$, symmetry property of real valued sequence.

$$X(k) = X^*(N-k) = X^*((-k))_N$$

$$X(8) = X^*(14-8) = X^*(6) = -2 + j3$$

$$X(9) = X^*(14-9) = X^*(5) = 6 - j3$$

$$X(10) = X^*(14-10) = X^*(4) = -2 - j2$$

$$X(11) = X^*(14-11) = X^*(3) = 1 + j5$$

$$X(12) = X^*(14-12) = X^*(2) = 3 - j4$$

$$X(13) = X^*(14-13) = X^*(1) = -1 - j3$$

$$(i) \text{ We know } x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{\frac{j2\pi}{N} kn} ; 0 \leq n \leq N-1$$

$$N=14, \quad x(n) = \frac{1}{14} \sum_{k=0}^{13} X(k) e^{\frac{j\pi}{7} kn}$$

$$\text{Put } n=0, \quad x(0) = \frac{1}{14} \sum_{k=0}^{13} X(k)$$

$$= \frac{1}{14} [x(0) + x(1) + x(2) + \dots + x(13)]$$

$$x(0) = \frac{32}{14} = \underline{\underline{2.2857}}$$

$$(ii) \text{ We know, } x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{\frac{j2\pi}{N} kn}$$

$$N=14, \quad x(n) = \frac{1}{14} \sum_{k=0}^{13} X(k) e^{\frac{j2\pi}{14} kn}$$

$$\text{Put } n=7, \quad x(7) = \frac{1}{14} \sum_{k=0}^{13} (-1)^k X(k)$$

$$= -\frac{12}{14} = \underline{\underline{-0.8571}}$$

$$(iii) \quad X(k) = \sum_{n=0}^{13} x(n) e^{-\frac{j2\pi}{14} kn}$$

$$\text{Put } k=0, \quad X(0) = \sum_{n=0}^{13} x(n)$$

$$\therefore \sum_{n=0}^{13} x(n) = X(0) = \underline{\underline{12}}$$

iv) The DFT of $e^{j\frac{4\pi}{7}m} x(m)$ or $e^{j\frac{2\pi}{14}4m} x(m)$ is

$$X((k-4))_{14}$$

We know, $X(k) = \sum_{n=0}^{N-1} x(n) e^{-j\frac{2\pi}{N} km} ; 0 \leq k \leq N-1$

$$\therefore X((k-4))_{14} = \sum_{n=0}^{13} \left\{ e^{j\frac{4\pi}{7}m} x(m) \right\} e^{-j\frac{2\pi}{14}km}$$

Put $k=0$,

$$X((-4))_{14} = \sum_{n=0}^{13} e^{j\frac{4\pi}{7}n} x(n)$$

$$\therefore \sum_{n=0}^{13} e^{j\frac{4\pi}{7}n} x(n) = X(10) \\ = \underline{-2 - j2}$$

v) From Parseval's Theorem,

$$\sum_{n=0}^{N-1} |x(n)|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X(k)|^2$$

$$N=14, \quad \sum_{n=0}^{13} |x(n)|^2 = \frac{1}{14} \sum_{k=0}^{13} |X(k)|^2$$

$$= \frac{1}{14} \left[|X(0)|^2 + |X(1)|^2 + \dots + |X(13)|^2 \right]$$

$$= \frac{498}{14} = \underline{\underline{35.5714}}$$