# **Chapter 12: Power Electronics – Instructor Notes**

Chapter 12 introduces the subject of power electronics. The importance of power electronics cannot be overemphasized, considering the widespread industrial application of electric machines, and other high current loads in practical engineering applications. The chapter discusses the basic characteristics and limitations of power amplifiers, practical voltage regulators, inductive loads (such as electric motors), and SCRs. The aim is to give the student sufficient understanding of the device characteristics to be able to complete simple "order of magnitude" calculations to be able to size a device for a given application. **This chapter** is somewhat more practically oriented than some of the others in the text, and **may be used to accompany a course in electric power and machines based on Chapters 7, 18, 19 and 20.** 

After Sections 12.1 and 12.2 present a classification of power electronic devices (Figure 12.1, p. 622) and circuits (Table 12.1, p. 623), the discussion is divided into the topics of: Voltage Regulators (Section 12.3); Power Amplifiers and Transistor Switches (Section 12.4), which includes two new examples on push pull amplifiers and one on the efficiency of three different amplifier output stages; Rectifiers and Controlled Rectifiers (Section 12.5), and Electric Motor Drives (Section 12.6)

Homework problems are divided into four major sections. The first, on voltage regulators, includes problems two different voltage regulator circuits (12.2, 12.3). The second offers two new power amplifiere problems. The third, on rectifiers and controlled rectifiers, illustrates a battery charging circuit (12.9) and two simple motor speed control problems (12.11, 12.12). The last section, on electric motor drives, introduces choppers, and more advanced problems on DC motor supplies based on controlled rectifiers (12.28 to 12.31) and on switched power supplies (12.32 – 12.38). The 5th Edition of this book includes 14 new problems, increasing the end-of-chapter problem count from 24 to 38.

#### **Learning Objectives**

- 1. Learn the classification of power electronic devices and circuits. *Sections 12.1 and 12.2.*
- 2. Analyze the operation of practical voltage regulators. Section 12.3.
- 3. Understand the principal limitations of transistor power amplifiers. Section 12.4.
- 4. Analyze the operation of single- and three-phase controlled rectifier circuits. *Section* 12.5.
- 5. Understand the operation of power converters used in electric motor control, and perform simplified analysis on DC-DC converters. *Section 12.6*.

# Section 12.3: Voltage Regulators

# Problem 12.1

#### Solution:

#### Find:

Repeat Example 12.1 for a 7-V Zener diode.

# Analysis:

Calculating the collector and base currents according to:

$$I_E = \frac{7 - 1.3}{10} = 0.57 \,\text{A}$$

$$I_B = \frac{I_E}{11} = 51.8 \,\text{mA}$$

We find:

$$I_R = \frac{20-7}{47} = 0.277 \,\mathrm{A}$$

$$I_Z = I_R - I_B = 0.225 \,\mathrm{A}$$

$$V_{CE} = 20 - V_L = 20 - 5.7 = 14.3 \,\mathrm{V} > 0.6 \,\mathrm{V}$$

Thus, the transistor is in the active region. The Zener power is:  $I_Z \cdot V_Z = 1.576 \, \mathrm{W}$  .

# Problem 12.2

#### Solution:

#### **Known quantities:**

The current regulator circuit shown in Figure P12.2.

#### Find:

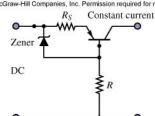
The expression for  $R_S$ .

#### Analysis:

Assuming that the Zener voltage is  $V_Z$ , that  $V_{BE} = V_{\gamma} = 0.6 \, \text{V}$ , and that the required current is I, we have:

$$R_S = \frac{V_Z + V_{BE}}{I} = \frac{V_Z + 0.6}{I}$$
.





#### Solution:

#### **Known quantities:**

The shunt-type voltage regulator shown in Figure P12.3.

#### Find:

The expression for the output voltage,  $V_{out}$ .

# Zener Vou

# Analysis:

If the Zener diode is to be in the regulator mode, the CB junction must be forward biased; in this case, both the CB and the BE junctions are forward biased, since a substantial base current will be generated through the Zener diode (depending on the value of the shunt resistor in the output circuit). Thus, the collector-emitter

voltage is equal to: 
$$V_{CEsat} \approx 0.2 \, \text{V}$$
, and the source current will be:  $I_S = \frac{V_S - V_{CEsat}}{R_S} = I_C + I_Z$ .

The voltage across the shunt resistor will therefore be  $V_{\gamma}$ , and the output voltage is:  $V_{out} = V_Z + V_{\gamma}$ .

# **Section 12.4: Power Amplifiers**

# Problem 12.4

# Solution:

# **Known quantities:**

The circuit shown in Figure P12.4.

#### Find:

Determine the power delivered to the 1.2 V rechargeable battery in the circuit.

#### **Analysis:**

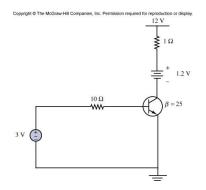
$$I_B = \frac{3 - 0.7}{10} = \frac{2.7}{10} = 0.27 \text{ A}$$

$$I_C = \beta I_B = 25(0.27) = 6.75 \text{ A}$$

$$V_{CE} = 12 - 1.2 - 6.75(1) = 4.05 \text{ V}$$

Power delivered to the 1.2 V rechargeable battery

$$P_{Batt} = 1.2 \times I_C = 1.2 \times 6.75 = 8.1 \text{ W}$$



# Problem 12.5

#### Solution:

#### **Known quantities:**

The circuit shown in Figure P12.5.

#### Find

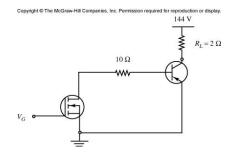
Determine the current through  $R_L$  and the voltage across it.

#### **Analysis:**

$$I_D = K(V_{GS} - V_T)^2 = 0.01(8 - 4)^2 = 0.16 \text{ A}$$
  
 $I_B = I_D$   
 $I_C = \beta I_B = 200(0.16) = 32 \text{ A}$ 

The current through 
$$R_L$$
 is just  $I_C = 32 \text{ A}$ 

and the voltage across it is  $V_R = R \times I_C = 2 \times 32 = 64 \text{ V}$ 



# Section 12.5: Rectifiers and Controlled Rectifiers (AC-DC Converters)

#### Problem 12.6

#### Solution:

#### **Known quantities:**

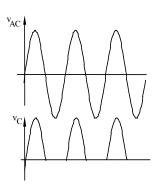
The circuit shown in Figure 12.19.

#### Find:

If the LR load is replaced by a capacitor, draw the output waveform and label the values.

# **Analysis:**

When the sinusoidal source voltage is in the positive half cycle, the series diode conducts, and the shunt diode is an open circuit; thus, the positive half cycle appears directly across the capacitor (assuming ideal diodes). During the negative half cycle, the series diode is open, and therefore the voltage across the capacitor remains zero, shown in the sketches on the right.



#### Problem 12.7

#### Solution:

#### **Known quantities:**

The circuit shown in Figure 12.19.

#### Find:

If the diode forward resistance is 50  $\Omega$ , the forward bias voltage is 0.7 V, and the load consists of a resistor  $R = 10\Omega$  and an inductor L = 2 H, draw  $v_L(t)$  and label the values for the given circuit.

#### Analysis:

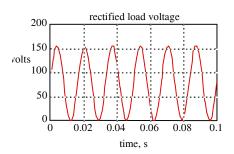
To obtain exact numerical values, we assume a 111  $V_{rms}$  source,  $R = 10\Omega$ , and L = 2H, then:

 $v_{AC}(t) = A\sin(\omega t) = 155.6 \cdot \sin(377t)$ , and from Equation 12.6,

the average load current is:  $I_L = \frac{155.6}{\pi R} = 4.95\,\mathrm{A}$  . Using the

approximation:  $v_L(t) \approx \frac{A}{2} + \frac{A}{2} \sin(\omega t)$  we have:

 $v_L(t) \approx \frac{A}{2} + \frac{A}{2}\sin(\omega t) = 77.8 + 77.8\sin(377t)$ . The waveform is shown on the right:



#### Solution:

# **Known quantities:**

For the circuit shown in Figure P12.8,  $v_{AC}$  is a sinusoid with 11 V peak amplitude,  $R = 2 \text{ k}\Omega$  and the forward-conducting voltage of D is 0.7 V.

#### Find:

- a. Sketch the waveform of  $v_L(t)$ .
- b. Find the average value of  $v_L(t)$ .

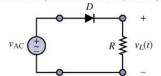
# Analysis:

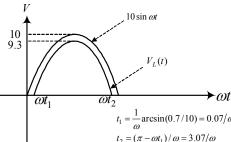
a. Assume  $v_{AC}(t) = 10\sin(\omega t)V$ . The output voltage is:  $v_L(t) = \begin{cases} 10\sin(\omega t) - 0.7 & \text{if } 10\sin(\omega t) - 0.7 \ge 0\\ 0 & \text{Otherwise} \end{cases}$ .

The waveform is shown on the right:

b. 
$$\langle v_L \rangle = \frac{\omega}{2\pi} \int_{t_1}^{t_2} (10\sin(\omega t) - 0.7) dt = \frac{1}{2\pi} (-\cos\omega t - 0.7\omega t)|_{t_1}^{t_2} = 2.841 \text{ V}.$$

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# Problem 12.9

#### Solution:

#### **Known quantities:**

The vehicle battery charge circuit shown in Figure P12.9.

#### Find:

Describe the circuit, and draw the output waveform ( $L_1$  and  $L_2$  represent the inductances of the windings of the alternator).

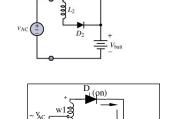
# Analysis:

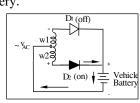
The positive half cycle from w1 is conducted by diode  $D_1$ . Diode  $D_2$  does not conduct due to negative bias at w2. The first half cycle is passed through to the battery.

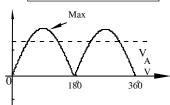
The second half cycle finds w2 positive and diode D<sub>2</sub> conducts current to the battery while diode D<sub>1</sub> is negatively biased and is off.

The full-wave rectified output waveform is shown below.

The average DC value  $\boldsymbol{V}_{AV}$  is 63% of the peak value.







# Solution:

#### Find:

Repeat Example 12.4 for  $\alpha = \pi/3$  and  $\pi/6$ .

# **Analysis:**

a) For  $\alpha = \pi/3$  we have:  $v_L \left(\frac{\pi}{3}\right) = \frac{120\sqrt{2}}{2} \sqrt{1 - \frac{1}{3} + \sin\frac{2\pi}{3}} = 105 \,\text{V}$ .

The power is:  $P = \frac{v_L^2}{R} = 45.94 \,\text{W}$ .

b) For  $\alpha = \pi/6$  we have:  $v_L \left(\frac{\pi}{6}\right) = \frac{120\sqrt{2}}{2} \sqrt{1 - \frac{1}{6} + \sin\frac{\pi}{3}} = 110.6 \text{ V}$ .

The power is:  $P = \frac{v_L^2}{R} = 50.97 \,\text{W}$ .

# **Problem 12.11**

#### Solution:

# **Known quantities:**

For the circuit shown in Figure P12.11, assume the thyristors are fired at  $\alpha = 60^{\circ}$  and that the motor current is 20 A and is ripple free. The supply is 111  $V_{AC}$  (rms).



- a) Sketch the output voltage waveform,  $v_0$ .
- b) Compute the power absorbed by the motor.
- c) Determine the volt-amperes generated by the supply.

### **Analysis:**

a) The output voltage waveform is on the right:

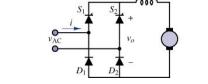
$$\alpha = 60 \times \frac{\pi}{180} = \frac{\pi}{3} \,, \ R_a = 0.2\Omega \,. \label{eq:alpha}$$

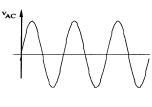
b) 
$$V_{O_{rms}} = \frac{120\sqrt{2}}{2} \left[ 1 - \frac{1}{3} + \sin 120^{\circ} \right]^{\frac{1}{2}} = 105 \text{ V} \implies$$

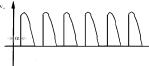
$$P_m = I_o V_{o_{rms}} = (20 \,\mathrm{A})(105 \,\mathrm{V}) = 2.1 \,\mathrm{kW}$$
.

c) 
$$P_R = I_o^2 R_a = (20)^2 (0.2) = 80 \,\text{W}$$
;  $P_S = P_m + P_R = 2180 \,\text{W}$ .









#### Solution:

### **Known quantities:**

The circuit of Figure 12.2, replacing the resistive load with a DC motor. The motor operates at 110 V and absorbs 4 kW of power. The AC supply is 80 V, 60 Hz. Assume that the motor inductance is very large (i.e., the motor current is ripple free), and that the motor constant is 0.055 V/rev/min.

#### Find:

If the motor runs at 1,000 rev/min at rated current:

- a) Determine the firing angle of the converter.
- b) Determine the rms value of the supply current.

#### **Analysis:**

Back emf =  $k_T \times N = 0.055 \times 1000 = 55 \text{ V}$ 

$$I_{DC} = \frac{P}{V} = \frac{4000}{110} = 36.4 \,\mathrm{A}$$

$$V = 110 \,\text{V}$$
. Assume  $R = 1\Omega$ . 
$$I_{DC} = \frac{1}{\pi R} \left[ \sqrt{2} V(\cos \alpha) - V_B(\pi - \alpha) \right]$$

a) 
$$36.4 = \frac{1}{\pi} \left[ \sqrt{2} (110) \cos \alpha - 55(\pi - \alpha) \right]$$
. Solving yelds:  $\alpha \approx \frac{7}{12} \pi$  rad.

b) With zero ripple,  $I_{rms} = I_{DC} = 36.4 \,\mathrm{A}$ .

#### Solution:

#### Find:

**NOTE:** Typo in problem statement, referring to the wrong example problem. For the light dimmer circuit of Example 12.6, determine the load power at firing angles  $\alpha = 0^{0}, 30^{0}, 60^{0}, 90^{0}, 120^{0}, 150^{0}, 180^{0}$ , and plot the load power as a function of  $\alpha$ .

# **Analysis:**

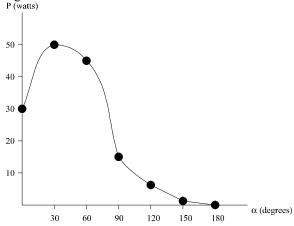
and

 $V_{L_{rms}} = \frac{120}{\sqrt{2}} \sqrt{1 - \frac{\alpha}{\pi} + \sin 2\alpha}$ , for  $\alpha \le 90^{\circ}$ , Note that

 $V_{L_{rms}} = \frac{120}{\sqrt{2}} \sqrt{1 - \frac{\alpha}{\pi} + \frac{\sin 2\alpha}{2\pi}}$ , for  $a > 90^0$ 

α	$V_{L_{rms}}$	$P = \frac{V_{L_{rms}}^2}{R_{BULB}}$
0°	84.85	30
30°	111.60	50.98
60°	115.00	45.98
90°	60.00	15.00
120°	37.53	5.87
150°	14.41	0.87
180°	0	0

A sketch of power vs. firing angle is shown below:

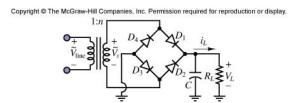


#### Solution:

# **Known quantities:**

For the circuit shown in Figure P12.14:

$$V_L = 10 \text{ V}$$
  $V_r = 10\% = 1 \text{ V}$ 
 $I_L = 650 \text{ mA}$   $v_{line} = 170 \cos \omega t \text{ V}$ 
 $\omega = 2513 \frac{\text{rad}}{\text{s}}$ 

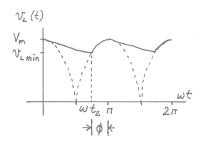


#### Find:

Determine the conduction angle of the diodes, if the diodes are fabricated form silicon.

# **Analysis:**

$$V_m = V_L + \frac{1}{2}V_r = 10 \text{ V} + \frac{1}{2}[1 \text{ V}] = 10.5 \text{ V}$$
  
 $v_{L\text{-min}} = V_L - \frac{1}{2}V_r = 10 \text{ V} - \frac{1}{2}[1 \text{ V}] = 9.5 \text{ V}$ 

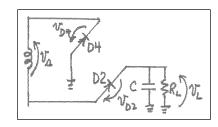


D2 and D4 conduct during  $\omega t_2 < \omega t < \pi$ .

First, determine the amplitude of the source voltage. [The secondary of the transformer acts as a source.] Then use the same KVL to determine the angle at which the diodes start conducting.

$$KVL: +_{VD4} +_{Vs} +_{VD2} +_{VL} = 0$$

At 
$$\omega t = \pi$$
:  $v_{D2} = v_{D4} = v_{D-on} = 0.7 \text{ V [Si]}$   
 $v_s = -V_{so}$   $v_L = V_m$   
 $V_{so} = v_{D-on} + v_{D-on} + V_m = 0.7 \text{ V} + 0.7 \text{ V} + 10.5 \text{ V} = 11.9 \text{ V}$ 



At 
$$\omega t = \omega t_2$$
:  $v_{D4} = v_{D2} = v_{D-on} = 0.7 \text{ V}$   $v_s = V_{so} \cos \omega t_2$   $v_L = v_{L-min}$ 

$$\cos \omega t_2 = -\frac{v_{D-on} + v_{D-on} + v_{L-min}}{V_{so}} = -\frac{0.7 \text{ V} + 0.7 \text{ V} + 9.5 \text{ V}}{11.9 \text{ V}} = -0.91597$$

$$\omega t_2 = 156.3^{\circ} \qquad \therefore \phi = 180^{\circ} - 156.3^{\circ} = 23.66^{\circ}$$

#### Solution:

#### **Known quantities:**

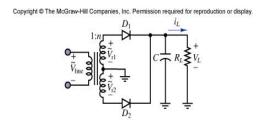
For the circuit shown in Figure P12.15, assume that the conduction angle of the diodes shown (which are Silicon) is:

$$\Phi = 23^{o}$$

$$v_{S} I[t] = v_{S} 2[t] = 8V \cos wt$$

$$w = 377 \frac{1}{s} \quad R_{L} = 20 \text{ K}\Omega$$

$$C = 0.5 \mu f$$



#### Find:

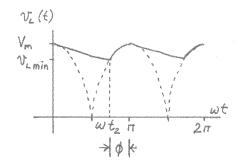
The ripple voltage.

$$KVL: -v_{s1}(t) + v_{D1} + v_{L}(t) = 0$$

$$At t = 0:$$

$$v_{L}(0) = V_{m} = V_{so} - v_{D-on} = 8 \text{ V} - 0.7 \text{ V} = 7.3 \text{ V}$$

$$\omega t_{2} = \pi - \Phi \implies t_{2} = \frac{\pi - 23^{\circ} \frac{\pi}{180^{\circ}}}{377 \frac{\text{rad}}{\text{s}}} = 7.268 \text{ ms}$$



#### **Analysis:**

$$v_L(t) = v_L(\infty) + (v_L(0) - v_L(\infty))e^{-\frac{t}{T_C}} = 0 + [V_m - 0]e^{-\frac{t}{R_LC}}$$

$$v_L(t_2) = v_{L-\min} = 7.3 \cdot e^{-\frac{7.286 \cdot 10^{-3}}{[20 \cdot 10^3][0.5 \cdot 10^{-6}]}} = 3.529 \text{ V}$$

$$V_r = V_m - v_{L-\min} = 7.3 - 3.529 = 3.771 \text{ V}$$

Note the ripple is quite large. This is primarily due to the very small (for this type circuit) value of the capacitance. Also the conduction angle assumed above is not correct for this circuit.

#### Solution:

# **Known quantities:**

The diodes in the full-wave DC power supply shown are Silicon. If:

$$I_L = 85 \text{ ma}$$
  $V_L = 5.3 \text{ V}$   $V_r = 0.6 \text{ V}$   $\omega = 377 \frac{\text{rad}}{\text{s}}$   $v_{line} = 156 \cos \omega t \text{ V}$   $C = 1023 \,\mu\text{F}$   $\phi = Conduction \, Angle = 23.90^{\circ}$ 

#### Find:

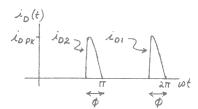
The value of the average and peak current through each diode.

# **Analysis:**

Diodes D1 and D3 will conduct half of the load current and Diodes D2 and D4 will conduct the other half. Therefore:

$$i_{D-ave} = \frac{1}{2}I_L = \frac{1}{2}[85 \text{ mA}] = 42.5 \text{ mA}$$

The waveforms of the diode currents are complex but can be roughly approximated as triangular [recall area of triangle = bh/2]:



$$I_{L} = \left[ i_{D1,3} + i_{D2,4} \right]_{ave} = \frac{1}{2\pi} \int_{0}^{2\pi} \left[ i_{D1,3}(\omega t) + i_{D2,4}(\omega t) \right] d[\omega t] = \frac{1}{2\pi} \left[ \frac{\phi i_{D-pk}}{2} + \frac{\phi i_{D-pk}}{2} \right]$$

$$i_{D-pk} = \frac{2\pi I_{L}}{\frac{1}{2} \Phi + \frac{1}{2} \Phi} = \frac{2\pi I_{L}}{\phi} = \frac{\left[ 2\pi \operatorname{rad} \right] \left[ 85 \operatorname{ma} \right]}{\left[ 23.90^{\circ} \right] \left[ \frac{\pi \operatorname{rad}}{180^{\circ}} \right]} = 1.280 \,\mathrm{A} \,.$$

#### Solution:

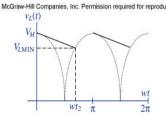
# **Known quantities:**

The diodes in the full-wave DC power supply shown in Figure P12.15 are Silicon. If:

$$I_L = 600 \,\text{mA}$$
  $V_L = 50 \,\text{V}$   $V_r = 8\% = 4 \,\text{V}$   $V_{line} = 170 \cos \omega t \,\text{V}$   $\omega = 377 \,\frac{\text{rad}}{\text{s}}$ 

#### Find:

The value of the conduction angle for the diodes and the average and peak current through the diodes. The load voltage waveform is shown in Figure P12.17.



# **Analysis:**

$$v_{L-\min} = V_L - \frac{1}{2}V_r = 50 - \frac{1}{2}4 = 48 \text{ V}$$

$$V_m = V_L + \frac{1}{2}V_r = 50 + \frac{1}{2}4 = 52V$$

$$KVL: -v_{s1}(t)+v_{D1}+v_{L}(t)=0$$

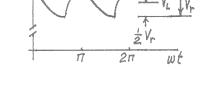
At 
$$t = 0$$
:  $-V_{so}\cos(0) + v_{D-on} + V_m = 0 \implies V_{so} = 0.7 + 52 = 52.7 \text{ V}$ 

$$KVL: +v_{s2}(t)+v_{D2}+v_{L}(t)=0$$

At 
$$t = t_2$$
:  $+V_{so}\cos(\omega t_2) + v_{D-on} + v_{L-min} = 0$ 

$$\omega t_2 = \cos^2 \left[ -\frac{v_{D-on} + V_{L-min}}{V_{so}} \right] = \cos^2 \left[ -\frac{0.7V + 48V}{52.7 V} \right] = 2.749 \text{ rad}$$

$$\Phi = \pi - \omega t_2 = \pi - 2.749 \text{ rad} = 392.1 \text{ mrad} \frac{180^{\circ}}{\pi \text{ rad}} = 22.47^{\circ}$$



The waveforms of the diode currents are complex but can be roughly approximated as triangular [recall the area of triangle = bh/2].

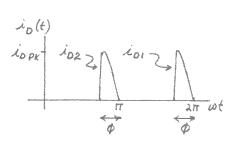
$$i_{D1-ave} = i_{D2-ave} = \frac{1}{2}I_L = \frac{1}{2}600 \text{ mA} = 300 \text{ mA}$$

$$[i_{D1} + i_{D2}]_{ave} = I_L$$

$$\frac{1}{\omega T} \int_0^{2\pi} [i_{D1}(\omega t) + i_{D2}(\omega t)] d[\omega t] = I_L$$

$$\frac{1}{\omega T} \left[ \frac{1}{2} \Phi_{iD-pk} + \frac{1}{2} \Phi_{iD-pk} \right] = I_L$$

$$i_{D-pk} = \frac{\omega T I_L}{\Phi} = \frac{[2\pi \text{ rad}][600 \text{ mA}]}{392.1 \text{ mrad}} = 9.615 \text{ A}$$



# Solution:

# **Known quantities:**

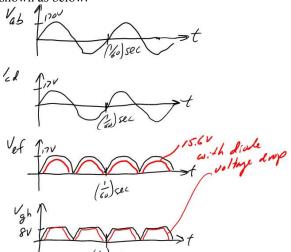
The circuit shown in figure P12.18.

#### Find:

Sketch the signals  $V_{ab}$ ,  $V_{,cd}$ ,  $V_{,ef}$  and  $V_{gh}$ .

# Analysis:

The voltage time curves are shown as below.



120 Vrms

20:1

# **Problem 12.19**

#### Solution:

# **Known quantities:**

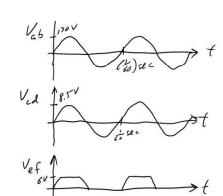
The circuit shown in Figure P12.19.

# Find:

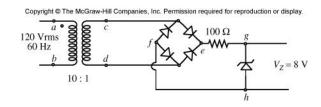
Sketch the signals  $V_{ab}$ ,  $V_{,cd}$ ,  $V_{,ef}$  and  $I_Z$ .

# **Analysis:**

The voltage time curves and are shown in the figure



No load



# Solution:

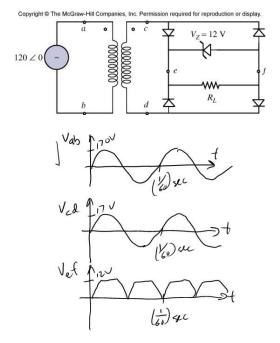
# **Known quanlities:**

The circuit shown in Figure P12.20.

#### Find:

Sketch the signals.

# Analysis:



# **Problem 12.21**

# Solution:

# Known quantities:

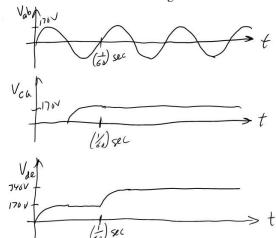
The circuit shown in Figure 12.21.

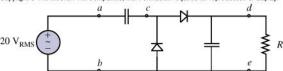
#### Find:

Sketch the signals  $V_{ab}, V_{,ca}, V_{,de}$  assuming that R is so large as to make any ripple not noticeable.

#### Analysis:

The voltage time curves and are shown as below. It is a voltage doubler.





#### Solution:

# **Known quantities:**

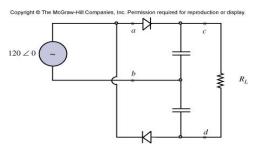
The circuit shown in Figure P12.22.

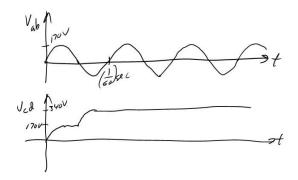
#### Find:

Sketch the signals  $V_{ab}$ ,  $V_{,cd}$  assuming that the capacitors are large enough that the ripple is not significant in the output voltage.

# **Analysis:**

The voltage time curves and are shown as below.





# Section 12.6: Electric Motor Drives

### **Problem 12.23**

#### Solution:

#### **Known quantities:**

For the chopper of Figure 12.37 in the text, the supply voltage is 120 V, and the armature resistance of the motor is 0.15  $\Omega$ . The motor back emf constant is 0.05 V/rev/min and the chopper frequency is 250 Hz. Assume that the motor current is free of ripple and equal to 125 A at 120 rev/min.

#### Find:

- a) The duty cycle of the chopper,  $\delta$ , and the chopper-on time,  $t_I$ .
- b) The power absorbed by the motor.
- c) The power generated by the supply.

#### Analysis:

$$v_o = i_o R_a + E_a = (125)(0.15) + 6 = 24.75 V$$
A
$$) 24.75 = \delta(120) \Rightarrow \delta = \frac{24.75}{120} = 0.2063$$

$$\delta = \frac{t_1}{T} \Rightarrow t_1 = \delta T = (0.2063) \left(\frac{1}{250}\right) = 825 \mu s$$
b)
$$P_m = E_a i_o = (6)(125) = 750 W$$

$$P_R = R_a i_o^2 = (0.15)(125)^2 = 2.344 kW$$
c)
$$P_S = P_m + P_R = 3.094 kW \text{ or } P_S = \delta \cdot V_S \cdot i_o = (0.2063)(120)(125) = 3.094 kW .$$

#### **Problem 12.24**

#### Solution:

#### **Known quantities:**

For the circuit shown in Figure 12.41 in the text, the motor constant is 0.3 V/rev/min, the supply voltage is 600 V, and the armature resistance is  $R_a = 0.2\Omega$ .

#### Find:

If the motor speed is 800 rev/min and the motor current is 300 A, determine:

- a) The duty cycle of the chopper,  $\delta$ .
- b) The power fed back to the supply.

#### **Analysis:**

a) 
$$v_o = E_a + i_o R_a = 240 + (-300)(0.2) = 180 \text{ V} \implies \delta = \frac{180}{600} = 0.300 \text{ .}$$

b) 
$$P_{m} = E_{a}i_{o} = (240)(-300) = -72.0 \text{ kW}$$

$$P_{R} = R_{a}i_{o}^{2} = (0.2)(-300)^{2} = 18 \text{ kW}$$

$$P_{S} = P_{m} + P_{R} = -54.0 \text{ kW}$$
or 
$$P_{S} = \delta \cdot V_{S} \cdot i_{o} = (0.300)(600)(-300) = -54.0 \text{ kW}.$$

#### Solution:

#### **Known quantities:**

For the two quadrant chopper of Figure 12.42 in the text, assume the thyristors  $S_1$  and  $S_2$  are turned on for time  $t_1$  and off for time  $T - t_1$  (T is the chopping period).

#### Find:

An expression for the average output voltage in terms of the supply voltage,  $V_s$ , and the duty cycle,  $\delta$ .

# **Analysis:**

$$\left\langle v_o \right\rangle = \frac{t_1}{T} V_S = \frac{t_1}{t \cdot t_1} V_S = \frac{1}{t} V_S = \delta \cdot V_S \; .$$

# **Problem 12.26**

#### Solution:

#### **Known quantities:**

Supply voltage; chopper duty cycle.

#### Find:

Average and rms value of ideal switched supply voltage.

### **Analysis:**

The duty cycle is  $\delta = \frac{t_1}{T} = \frac{1}{2.5} = 0.4$  The average value is therefore  $\langle V_{\text{supply}} \rangle = \delta \times V_{\text{supply}} = 40 \text{ V}$ .

To compute the rms value we use the definition of eq. 4.24:

$$\widetilde{V}_{\text{supply}} = \sqrt{\int_{0}^{\delta} (V_{\text{supply}})^2 dt} = \sqrt{\delta (100^2)} = 63.25 \text{ V}$$

#### Solution:

#### **Known quantities:**

Load resistance and inductance; ideal supply voltage; duty cycle.

#### Find:

Average values of current and voltage; power drawn from battery supply.

#### **Analysis:**

From the data given,  $\delta = 0.333$ . Since the period of the waveform is 3 ms, we can calculate the switching frequency to be:

$$f = \frac{1}{T} = \frac{1}{3 \times 10^{-3}} = 333.33 \,\text{Hz}$$

 $\omega = 2,094.4 \text{ rad/s}$ 

The time constant of the load impedance is  $\tau = \frac{L}{R} = \frac{10^{-3} \text{ H}}{0.5 \,\Omega} = 2 \text{ ms}$ .

The average load voltage is:  $\langle V_L \rangle = \delta \times V_{supply} = 33.33 \text{ V}.$ 

The average load current is  $\langle I_L \rangle = \delta \frac{V_{\text{supply}}}{R} = 0.33 \frac{100 \text{ V}}{0.5 \Omega} = 66.67 \text{ A}$ .

To compute the power drawn from the battery supply (which is assumed equal to the load power if switching losses are held negligible), we really need to compute the rms load current, since  $P_L = (\widetilde{I}_L)^2 R$ ; however, this calculation cannot be completed without knowing exactly the shape of the load current. Without further analysis we can only state that the power drawn from the battery is greater than  $\langle P_L \rangle = \langle V_L \rangle \langle I_L \rangle = 2.22$  kW.

#### **Problem 12.28**

#### Solution:

#### **Known quantities:**

The converter of Problem 12.27 with a DC motor as load.

$$R_a = 0.2\Omega$$
,  $L_a = 1$  mH,  $E_a = 10$  V,  $T = 3$  ms,  $\delta = \frac{1}{3}$ 

#### Find:

The average load current and voltage.

#### Analysis:

The average load voltage is:  $\langle V_L \rangle = \delta \times V_{supply} = 33.33 \text{ V}.$ 

The average current is: 
$$\langle I_L \rangle = \frac{\langle V_L \rangle - E_a}{R_a} = \frac{33.33 - 10}{0.2} = 116.5 \text{ A}$$

# Solution:

#### **Known quantities:**

Load resistance and inductance; load current; motor armature constant; DC supply and desired rpm range.

#### Find:

Range of duty cycles required.

#### Analysis:

Assume steady-state operation, so that the effects of the load inductance may be neglected. When the rpm is zero, the back emf is also 0, so

$$V_L = I_a R_a = 25 \times 0.3 = 7.5 \text{ V}$$
.

The motor emf constant is  $k_a \phi = 0.00167 \text{ V-s/rev}$ , or 60\*0.00167 = 0.1004 V-min/rev

At 2000 rpm, the back emf is:

$$E_a = k_a \phi n = 200.4 \text{ V}.$$

Thus, the total load voltage is

$$V_L = I_a R_a + E_a = 7.5 + 200.4 = 207.9 \text{ V}$$

From the range of voltages required by the motor for its proper operation, we conclude that the range of required duty cycles is:

$$\delta_{\min} = \frac{V_{\min}}{V_{\text{supply}}} = \frac{7.5}{220} = 0.0341$$

$$\delta_{\text{max}} = \frac{V_{\text{max}}}{V_{\text{supply}}} = \frac{207.9}{220} = 0.943$$

#### Solution:

#### **Known quantities:**

Motor ratings; motor armature resistance and armature constant; power supply ratings.

#### Find:

Motor speed, power factor and efficiency for  $\alpha = 0^{\circ}$  and  $\alpha = 20^{\circ}$ .

#### Analysis:

The nominal torque from the rated data is

$$T_m = \frac{P}{\omega_m} = \frac{10000}{\frac{2\pi 1000}{60}} = 95.49 \,\text{Nm}$$

It follows that the DC current is  $I_a = \frac{T_m}{K_a} = \frac{95.49}{2} = 47.75 \text{ A}$ 

The average load voltage for firing angle of  $0^{\circ}$  is

$$\langle V_{L,0^{\circ}} \rangle = \frac{2\sqrt{2}}{\pi} V_S = \frac{2\sqrt{2}}{\pi} 240 = 216 \text{ V}$$

The speed at zero degree of firing angle is given by

$$\langle V_{L.0^{\circ}} \rangle = E_{a.0^{\circ}} + R_a I_a \Rightarrow E_{a.0^{\circ}} = 216 - 0.42 \cdot 47.75 = 196 \text{ V}$$

$$\omega_{m, 0^{\circ}} = \frac{E_{a, 0^{\circ}}}{K_a} = \frac{196}{2} = 97.97 \text{ rad/s}$$

The efficiency is

$$\eta = \frac{E_{a,0^{\circ}}I_a}{E_{a,0^{\circ}}I_a + R_aI_a^2} = \frac{E_{a,0^{\circ}}}{E_{a,0^{\circ}} + R_aI_a} = \frac{196}{196 + 0.42 * 47.75} = 91\%$$

The rms voltage is

$$V_{rms,L} = V_S = 240$$

The power factor can be calculate as follows

$$pf(0^{\circ}) \cong \frac{E_{a,0^{\circ}}I_a + R_aI_a^2}{V_{rms, L}I_a} = \frac{E_{a,0^{\circ}} + R_aI_a}{V_{rms, L}} = \frac{196 + 0.42 * 47.75}{240} = 0.9$$

In the case of firing angle of 20°, we have

$$\langle V_{L,20^{\circ}} \rangle = \frac{2\sqrt{2}}{\pi} V_S \cos 20^{\circ} = 203 \text{ V}$$

$$E_{a, 20^{\circ}} = \langle V_{L, 20^{\circ}} \rangle - R_a I_a = 203 - 0.42 * 47.75 = 182.95 \text{ V}$$

$$\omega_{m, 20^{\circ}} = \frac{E_{a, 20^{\circ}}}{K_a} = \frac{182.95}{2} = 91.47 \,\text{rad/s}$$

$$\eta = \frac{E_{a, 20^{\circ}}}{E_{a, 20^{\circ}} + R_a I_a} = \frac{182.95}{182.95 + 0.42 \cdot 47.75} = 90\%$$

$$pf(20^{\circ}) \cong \frac{E_{a,20^{\circ}} + R_a I_a}{V_{rms,L}} = \frac{182.95 + 0.42 * 47.75}{240} = 0.846$$

#### Solution:

#### **Known quantities:**

Separately excited DC motor: P = 10 kW, V = 300 V,  $\omega = 1000 \text{ rev/min}$ ,  $R_a = 0.2 \Omega$ ,  $K_a = 1.38 \text{ V} \cdot \text{s/rad}$ 

Power supply:  $V_S = 220 \text{ (rms)V}$ , f = 60 Hz

Three phase controlled bridge rectifier.

#### Find:

Speed, power factor, and efficiency for a firing angle of 30 deg.

#### **Assumptions:**

Load torque is constant, and the DC motor deliver the power P at 0 deg of firing angle.

Additional inductance is present to ensure continuous conduction.

# Analysis:

The average voltage supplied by the rectifier at 0 deg of firing angle is

$$\langle V_{L,0^{\circ}} \rangle = \frac{3\sqrt{3}}{\pi} \sqrt{2} V_S = 514.6 \text{ V}$$

It follows

$$<\!V_{L,0^{\circ}}>=\!R_{a}I_{a}+E_{a,0^{\circ}}=R_{a}\frac{P}{E_{a,0^{\circ}}}+E_{a,0^{\circ}}\Rightarrow\!E_{a,0^{\circ}}^{2}-<\!V_{L,0^{\circ}}>\!E_{a,0^{\circ}}+R_{a}P=0\Rightarrow\!E_{a,0^{\circ}}^{2}-5146E_{a,0^{\circ}}+2000=0$$
 The

$$E_{a,0^{\circ}} = 51068 \,\mathrm{V} \Rightarrow I_a = I_{a,0^{\circ}} = I_{a,30^{\circ}} = \frac{P}{E_{a,0^{\circ}}} = \frac{10000}{51068} = 19.58 \,\mathrm{A}$$

average voltage supplied by the rectifier at 30° of firing angle is

$$\langle V_{L,30^{\circ}} \rangle = \langle V_{L,0^{\circ}} \rangle \cdot \cos 30^{\circ} = \frac{3\sqrt{3}}{\pi} \frac{\sqrt{3}}{2} \sqrt{2} V_{S} = 445.65 \text{V}$$

The emf of the DC motor in this condition is given by

$$E_{a,30^{\circ}} = \langle V_{L,30^{\circ}} \rangle - R_a I_a = 44565 - 0.2 \cdot 19.58 = 441.74 \text{ V}$$

Finally, the speed is given by

$$\omega_m = \frac{E_{a, 30^{\circ}}}{K_a} = 320.1 \frac{\text{rad}}{\text{s}}$$

If we assume that the inductance is big enough, then the current ripple is negligible.

Under this conditions

$$\eta = \frac{E_{a, 30^{\circ}} I_a}{E_{a, 30^{\circ}} I_a + R_a I_a^2} = \frac{8649.27}{8725.94} = 99.1\%$$

The rms voltage supplied for a firing angle of 30° is

$$V_{L,rms} = 453.03 \text{ V}$$

The power factor is

$$pf = \frac{P_{out}}{V_{L,rms}I_{L,rms}} \cong \frac{E_{a,30^{\circ}}I_a + R_aI_a^2}{V_{L,rms}I_a} = \frac{E_{a,30^{\circ}} + R_aI_a}{V_{L,rms}} = \frac{441.74 + 3.92}{453.03} = 0.984$$

#### Solution:

# **Known quantities:**

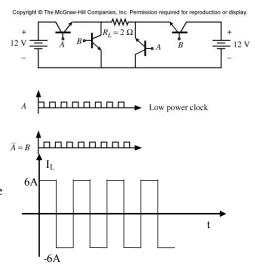
The circuit of Figure P12.32, the switch mode of A and B.

#### Find:

The load current waveform.

#### **Analysis:**

When A or B has a high level voltage, it works as a short-loop; when A or B has a low level voltage, it works as a open-loop. The current waveform of the load is as the below:



# Problem 12.33

Note: the battery symbol on the right hand side is upside down (the polarity is correct). And the right switch should be labeled  $\overline{A}$ .

# Solution:

# Known quantities:

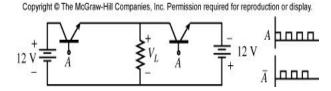
The circuit of Figure P12.33, the switch mode of A and B.

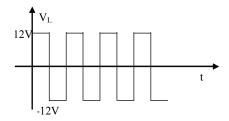
# Find:

The load voltage waveform.

#### Analysis:

The load voltage waveform is as shown in the figure:





Note: this problem has many solutions.

#### Solution:

#### **Known quantities:**

The circuit of Figure P12.34.

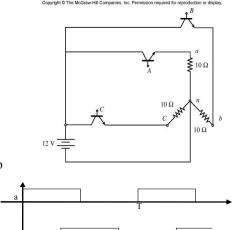
#### Find:

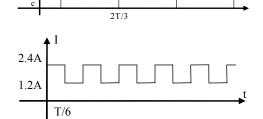
Sketch the timing diagrams for the three low power clock inputs to generate a balanced three phase source; sketch the current in the neutral return wire.

# **Analysis:**

When the output is a 3-phase alternate current, we know that each phase is T/3 delayed to the previous phase, T is the period of each phase. Let's assume A is fired at t=0, and followed by B and C, then the three low power clock inputs are as follows:

And the current in the neutral return wire is:



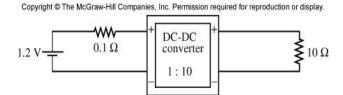


# Problem 12.35

#### Solution:

#### **Known quantities:**

A DC to DC converter as Figure P12.35. V=1.2 V,  $R=10\Omega$ 



#### Find:

Using the usual transformer "reflecting theorem", determine the power supply supplied by source and to load.

#### Analysis:

From "reflecting theorem", the reflected load circuit is:

So 
$$Z' = \frac{Z_L}{N^2} = \frac{10}{10^2} = 0.1 \ \Omega, \ P_S = \frac{V^2}{R} = \frac{(1.2)^2}{0.1 + 0.1} = 7.2 \ W$$

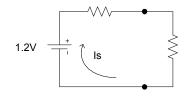
The power supplied by source is 7.2 W.

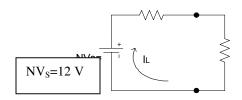
From "reflecting theorem", the reflected source circuit is:

So 
$$NV_s = 10 \cdot 1.2 = 12 \text{ V}$$
,  $Z'' = N^2 Z_s = 100 \cdot 0.1 = 10 \Omega$ 

$$P_L = \frac{V_L^2}{R_L} = \frac{(6)^2}{10} = 3.6 \text{ W}$$

The power supplied to the load is 3.6 W.





# Solution:

# **Known quantities:**

The circuit of P12.36

#### Find:

The voltage across a high resistance load R.

# **Assumptions:**

The frequency of clock is relatively high.

# Analysis:

When CLK is high, the lower capacitor sees 12 V. When  $\overline{CLK}$  is high, the upper capacitor sees 12 V. If the frequency is high, the load will see an almost constant voltage of 24 V.

# Problem 12.37

Note: there is a mistake in the figure, the right switch should be labeled  $\overline{A} \ \overline{B} \ \overline{C}$ 

# Solution:

# **Known quantities:**

The circuit of P12.37

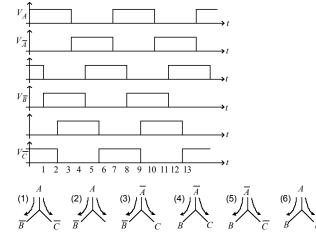
#### Find:

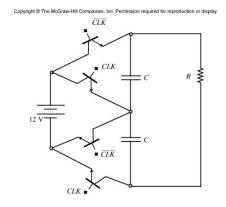
Sketch the low power periodic signals in time

#### **Analysis:**

A possible waveform of these low power periodic signals is as follows.

The desired current flow in the WYE load is from (1)-(6), and then repeats again and again.





# Solution:

# **Known quantities:**

The circuit of P12.38

#### Find:

Sketch the load voltage signal.

# **Analysis:**

According KVL, the following waveform of load voltage is easy to get:

