

Data sheet acquired from Harris Semiconductor SCHS043B – Revised July 2003

# CMOS Micropower Phase-Locked Loop

Phase-Locked Loop (PLL) consists of a low-power, linear voltage-controlled oscillator (VCO) and two different phase comparators having a common signal-input amplifier and a common comparator input. A 5.2-V zener diode is provided for supply regulation if necessary.

The CD4046B types are supplied in 16-lead hermetic dual-in-line ceramic packages (F3A suffix), 16-lead dual-in-line plastic packages (E suffix), 16-lead small-outline packages (NSR suffix), and 16-lead thin shrink small-outline packages (PW and PWR suffixes).

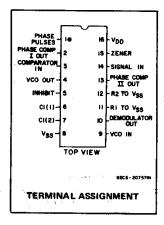
#### VCO Section

The VCO requires one external capacitor C1 and one or two external resistors (R1 or R1 and R2). Resistor R1 and capacitor C1 determine the frequency range of the VCO and resistor R2 enables the VCO to have a frequency offset if required. The high input impedance (1012 $\Omega$ ) of the VCO simplifies the design of low-pass filters by permitting the designer a wide choice of resistor-tocapacitor ratios. In order not to load the low-pass filter, a source-follower output of the VCO input voltage is provided at terminal 10 (DEMODULATED OUTPUT). If this terminal is used, a load resistor (RS) of 10  $k\boldsymbol{\Omega}$  or more should be connected from this terminal to VSS. If unused this terminal should be left open. The VCO can be connected either directly or through frequency dividers to the comparator input of the phase comparators. A full C'MOS logic swing is available at the output of the VCO and allows direct coupling to frequency dividers such as the RCA-CD4024, CD4018, CD4020, CD4022, CD4029, and CD4059 One or more CD4018 (Presettable Divide-by-N Counter) or CD4029 (Presettable Up/Down Counter), or CD4059A (Programmable Divide-by-"N" Counter), together with the CD4046B (Phase-Locked Loop) can be used to build a micropower low-frequency synthesizer. A logic 0 on the INHIBIT input "enables" the VCO and the source follower, while a logic 1 "turns off" both to minimize stand-by power consump-

## CD4046B Types

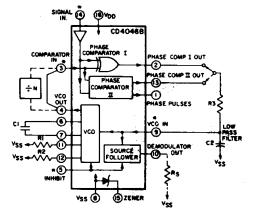
#### Features:

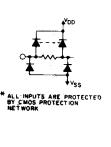
- Very low power consumption:
   70 μW (typ.) at VCO f<sub>o</sub> = 10 kHz, V<sub>DD</sub> = 5 V
- Operating frequency range up to 1.4 MHz (typ.) at  $V_{DD} = 10 \text{ V}$ , RI = 5 k $\Omega$
- Low frequency drift: 0.04%/°C (typ.) at VDD = 10 V
- Choice of two phase comparators:
   Exclusive-OR network (I)
   Edge-controlled memory network with phase-pulse output for lock indication (II)
- High VCO linearity: <1% (typ.) at VDD = 10 V</p>
- VCO inhibit control for ON-OFF keying and ultra-low standby power consumption
- Source-follower output of VCO control input (Demod. output)
- Zener diode to assist supply regulation
- Standardized, symmetrical output characteristics
- 100% tested for quiescent current at 20 V
- 5-V, 10-V, and 15-V parametric ratings
- Meets all requirements of JEDEC Tentative Standard No. 13B, "Standard Specifications for Description of 'B' Series CMOS Devices"



#### Applications:

- FM demodulator and modulator
- Frequency synthesis and multiplication
- Frequency discriminator
- Data synchronization
- Voltage-to-frequency conversion
- Tone decoding
- FSK Modems
- **■** Signal conditioning
- (See ICAN-6101) "RCA COS/MOS Phase-Locked Loop — A Versatile Building Block for Micropower Digital and Analog Applications"





9203 - 29172

Fig.1 - CMOS phase-locked loop block diagram. ....

## MAXIMUM RATINGS, Absolute-Maximum Values: DC SUPPLY-VOLTAGE RANGE, (VDD)

Voltages referenced to VSS Terminal)	0.5V to +20V
INPUT VOLTAGE RANGE, ALL INPUTS	0.5V to V <sub>DD</sub> +0.5V
DC INPUT CURRENT, ANY ONE INPUT	±10mA
POWER DISSIPATION PER PACKAGE (PD):	
For T <sub>A</sub> = -55°C to +100°C	500mW
For T <sub>A</sub> = +100°C to +125°C	
DEVICE DISSIPATION PER OUTPUT TRANSISTOR	
FOR TA = FULL PACKAGE-TEMPERATURE RANGE (All Package	Types) 100mW
OPERATING-TEMPERATURE RANGE (TA)	55°C to +125°C
STORAGE TEMPERATURE RANGE (Tstg)	65°C to +150°C
LEAD TEMPERATURE (DURING SOLDERING):	
At distance 1/16 $\pm$ 1/32 inch (1.59 $\pm$ 0.79mm) from case for 10s m	ax+265 <sup>o</sup> C

### Phase Comparators

The phase-comparator signal input (terminal 14) can be direct-coupled provided the signal swing is within CMOS logic levels [logic "0" ≤30% (VDD-VSS), logic "1" ≥ 70% (VDD-VSS)]. For smaller swings the signal must be capacitively coupled to the self-biasing amplifier at the signal input.

Phase comparator I is an exclusive-OR network; it operates analagously to an overdriven balanced mixer. To maximize the lock range, the signal- and comparator-input frequencies must have a 50% duty cycle. With no signal or noise on the signal input, this phase comparator has an average output voltage equal to VDD/2. The low-pass filter connected to the output of phase comparator

RECOMMENDED OPERATING CONDITIONS at  $T_A$  = Full Package-Temperature Range For maximum reliability, nominal operating conditions should be selected so that operation is always within the following ranges:

CHARACTERISTIC	LIN	UNITS	
	Min.	Max.	]
Supply-Voltage Range VCO Section:			
As Fixed Oscillator	3	18	1
Phased-Lock-Loop Operation	5	18	l v
Supply-Voltage Range Phase Comparator Section:			1
Comparators	3	18	1
VCO Operation	5	. 18	1 - 1

#### **DESIGN INFORMATION**

This information is a guide for approximating the values of external components for the CD4046B in a Phase-Locked-Loop system.

The selected external components must be within the following ranges:

 $5~\text{k}\Omega \leqslant \text{R1, R2, R}_{\text{S}} \leqslant 1~\text{M}\Omega$  C1  $\geqslant$  100 pF at  $\text{V}_{\text{DD}} \geqslant 5~\text{V};$  C1  $\geqslant$  50 pF at  $\text{V}_{\text{DD}} \geqslant$  10 V

Characteristics	Phase Comparator Used	Design Inf	ormation						
VCO Frequency	1	VCO WITHOUT OFFSET  R2 = ∞   MAX  To  To  To  To  To  To  To  VCO INPUT VOLTAGE	VCO WITH OFFSET						
For No Signal Input	1 2	Same as for No.1  VCO will adjust to center frequency, fo  VCO will adjust to lowest operating frequency, f							
Frequency Lock Range, 2 f	1	2 fL = full VCO frequency r. 2 fL = fmax-fmin							
Frequency Capture Range, 2 f <sub>C</sub>	2	Same as for No.1	(1), (2) $2 f_{\mathbb{C}} \approx \frac{1}{\pi} \sqrt{\frac{2\pi f_{\mathbb{L}}}{\sigma 1}}$						
Loop Filter Component Selection	1	IN N3 OUT	For 2 f <sub>C</sub> , see Ref. (2)						
	2	$f_C = f_L$							
Phase Angle Between Signal and Comparator	1	90° at center frequency (f <sub>0</sub> ) approximating 0° and 180° at ends of lock range (2 f <sub>L</sub> )							
	2	Always 00 in lock							
Locks On Harmonic of Center Frequency	. 1	Yes .							
Signal Input	1	No High							
Noise Rejection	- 2	Lov	<del></del>						

For further information, see

- (1) F. Gardner, "Phase-Lock Techniques" John Wiley and Sons, New York, 1966
- (2) G. S. Moschytz, "Miniaturized RC Filters Using Phase-Locked Loop", BSTJ, May, 1965.

I supplies the averaged voltage to the VCO input, and causes the VCO to oscillate at the center frequency ( $f_0$ ).

The frequency range of input signals on which the PLL will lock if it was initially out of lock is defined as the frequency capture range (2f<sub>C</sub>).

The frequency range of input signals on which the loop will stay locked if it was initially in lock is defined as the frequency lock range ( $2f_L$ ). The capture range is  $\leq$  the lock range.

With phase comparator I the range of frequencies over which the PLL can acquire lock (capture range) is dependent on the low-pass-filter characteristics, and can be made as large as the lock range. Phase-comparator I enables a PLL system to remain in lock in spite of high amounts of noise in the input signal.

One characteristic of this type of phase comparator is that it may lock onto input frequencies that are close to harmonics of the VCO center-frequency. A second characteristic is that the phase angle between the signal and the comparator input varies between  $0^{\rm O}$  and  $180^{\rm O}$ , and is  $90^{\rm O}$  at the center frequency. Fig. 2 shows the typical, triangular, phase-to-output response characteristic of phase-comparator I. Typical waveforms for a CMOS phase-locked-loop employing phase comparator I in locked condition of  $f_{\rm O}$  is shown in Fig. 3.

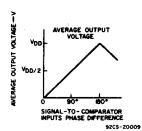


Fig.2 - Phase-comparator I characteristics at low-pass filter output.

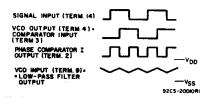


Fig. 3—Typical waveforms for CMOS phase-locked loop employing phase comparator in locked condition of f<sub>0</sub>.

Phase-comparator II is an edge-controlled digital memory network. It consists of four flip-flop stages, control gating, and a three-state output circuit comprising p- and n-type drivers having a common output node. When the p-MOS or n-MOS drivers are ON they pull the output up to VDD or down to VSS, respectively. This type of phase comparator acts only on the positive edges of the signal and comparator inputs. The duty cycles of the signal and comparator inputs are not important since positive transitions

CHARAC- TERISTIC	CONI	OITIO	NS	LIMITS AT INDICATED TEMPERATURES (°C)							
	ν <sub>ο</sub> (γ)	V <sub>IN</sub>	V <sub>DD</sub>	-55	-40	+85	+125	Min.	+25 Tue	Mex.	S
VCO Section	1141	(4)	107			700	7120	MAIN.	Тур.	MEX.	<u> </u>
	0.4	_ ^ E	5	0.64	0.01	0.40	0.00	A 54	· •		γ-
Output Low (Sink) Current	0.5	0,5 0,10	10	0.64	0.61 1.5	0.42	0.36	0.51	1		ł
IOL Min.	1.5	0.15	15	1.6 4.2	1.5	1.1 2.8	0.9 2.4	1.3 3.4	2.6 6.8	=	ł
	4.6	0,5	5	-0.64	-0.61	-0.42	-0.36	-0.51	-1	_	ļ,,
Output High (Source)	2.5	0,5	5	-2	-1.8	-1.3	-1.15	-1.6	-3.2	┢▔	<b>l</b> ‴
Current,	9.5	0.10	10	-1.6	-1.5	-1.1	-0.9	-1.3	-2.6	<u> </u>	ł
I <sub>OH</sub> Min.	13.5	0.15	15	-4.2	-4	-2.8	-2.4	-3.4	-6.8	┢ <u>╼</u>	ł
Output Voltage:	Term. 4	0.5	5		0	.05		_	0	0.05	t
Low-Level,	driving	0,10	10			.05			0	0.05	1
VOL Max.	смоѕ	0.15	15	<u> </u>		.05	-		0	0.05	ľ
Output	1	0,5	5		4	.95		4.95	5		ł۲
Voltage:	e.g.	0,10	10			95		9.95	10	_	
High-Level, V <sub>OH</sub> Min.	Term.3	0,15	15			.95		14.95	15	-	
Input Current	-	0,18	18	±0.1	±0.1	±1	±1	_	±10 <sup>—5</sup>	±0.,1	μ
Phase Comparator S	ection				:			-	·		
Total Device	I –	0,5	5	r		0.2		-	0.1	0.2	Г
Current, IDD Max.	_	0,10	10			1		-	0.5	1	1,,
Term. 14 open,	_	0,15	15	1.5			_	0.75	1.5	ľ	
Term. 5 = V <sub>DD</sub>	_	0,20	20	4			_	2	4	1	
	_	0,5	5			20		-	10	20	Γ
Term. 14 = V <sub>SS</sub>	_	0,10	10	<u> </u>		40		_	20	40	1,,
or V <sub>DD</sub> , Term. 5	_ :	0,15	15			80		_	40	80	ľ
= V <sub>DD</sub>	_	0,20	20			160		_	80	160	1
0.4.4	0.4	0,5	5	0.64	0.61	0.42	0.36	0.51	1	_	Γ
Output Low (Sink) Current	0.5	0,10	10	1.6	1.5	1.1	0.9	1.3	2.6		1
IOL Min.	1.5	0,15	15	4.2	4	2.8	2.4	3.4	6.8	_	1
Output High	4.6	0,5	5	-0.64	-0.61	-0.42	-0.36	-0.51	-1	_	l
(Source)	2.5	0,5	5	-2	-1.8	-1.3	-1.15	-1.6	-3.2	_	1
Current	9.5	0,10	10	-1.6	-1.5	-1.1	-0.9	-1.3	-2.6	-	1
OH Min.	13.5	0,15	15	-4.2	-4	-2.8	-2.4	-3.4	-6.8	_	1
DC-Coupled Signal Input and Comparator Input			5			1.5			_	1.5	
Voltage Sensitivity	0.5,4.5	<del>-</del>	10		<del></del>	3	<del> </del>	=	<del></del> -	3	1
Low Level VIL Max.	1,9 1.5,13.5	_	15			4		<u> </u>		<del></del>	v
High Level	0.5,4.5	_	5			3.5		3.5	I -	_	1
V <sub>IH</sub> Min.	1,9	-	10			7		7	<u> </u>	-	1
***	1.5,13.5		15		t	11	T	11	_		1

control the PLL system utilizing this type of comparator. If the signal-input frequency is higher than the comparator-input frequency, the p-type output griver is maintained ON most of the time, and both the n and p drivers OFF (3 state) the remainder

of the time. If the signal-input frequency is lower than the comparator-input frequency, the n-type output driver is maintained ON most of the time, and both the n and p drivers OFF (3 state) the remainder of the time. If the signal- and comparator-

input frequencies are the same, but the signal input lags the comparator input in phase, the n-type output driver is maintained ON for a time corresponding to the phase difference. If the signal- and comparator-input frequencies are the same, but

#### STATIC ELECTRICAL CHARACTERISTICS

CHARAC- TERISTIC	CO	NDITIO	NS	LIMITS AT INDICATED TEMPERATURES (°C)							
	V <sub>O</sub>	VIN	VDD	<b>–55</b>	<b>–40</b>	+85			s		
	(v)	(V)	(V)				+125	Min.	Тур.	Max.	]
Phase Comparator	Section	(cont'd	)								-
Input Current IJN Max. (except Term.14)	ı	0,18	18	±0.1	±0,1	±1	±1		±10 <sup>-5</sup>	±0.1	μΑ
3-State Leakage Current, IOUT Max.	0,18	0,18	18	±0.1	±0.1	±0.2	±0.2		±10 <sup>-5</sup>	±0.1	μА

<sup>\*</sup>Limit determined by minimum feasible leakage current measurement for automatic testing.

ELECTRICAL CHARACTERISTICS at TA = 25°C

CHARAG-					LIMITS	· :		
TERISTIC	TEST	CONDITIONS	V <sub>DD</sub>	-	LL TYP	E8	UNITS	
	L		(v)	Min.	Typ.	Max.		
VCO Section	<u> </u>	·						
Operating Power Dissipation, PD	f <sub>o</sub> = 10 kHz R <sub>2</sub> = ∞	$R_1 = 1 M\Omega$ $VCO_{IN} = \frac{V_{DD}}{2}$	5 10	_	70 800	140 1600	μW	
Dissipation, 1		VCO <sub>IN</sub> = 2	15	:	3000	6000		
Maximum	C <sub>1</sub> =50 pF		. 5	0.3	0.6	_		
Operating	R <sub>2</sub> = ∞	$R_1 = 10 k\Omega$	10	0.6	1.2	. –		
Frequency f <sub>max</sub>	ACO <sup>IM</sup> =A <sup>DD</sup>		15	8.0	1.6		MHz	
	C <sub>1</sub> = 50 pF		5	0.5	0.8	-		
	R <sub>2</sub> = ∞ VCO <sub>IN</sub> =V <sub>DD</sub>	$R_1 = 5 k\Omega$	10	1	1.4	· -·		
•	AGOIN-APP		15.	1.4	2.4		L	
Center Frequency (f <sub>O</sub> ) and Frequency Range (f <sub>max</sub> —f <sub>min</sub> )		Programmable with external components R1, R2, and C1 See Design Information						
	VCO <sub>IN</sub> = 2.5 V	$\pm$ 0.3V, R <sub>1</sub> =10 k $\Omega$	5		1.7			
	=5 V ±			_	0.5		-	
Linearity		$2.5 \text{ V}, =400 \text{ k}\Omega$	10		4		%	
		$\pm 1.5 \text{ V}, = 100 \text{ k}\Omega$			0.5	-		
-	= 7.5 V	$\pm 5 \text{ V}, = 1 \text{ M}\Omega$	15	_	7	· –		
Temperature - Frequency Stability: No Frequency			5 10 15	1 1 1	±0.12 ±0.04 ±0.015	<u>-</u> ,		
Offset f <sub>MIN</sub> = 0							%/°C	
Frequency Offset			5 10	1 -	±0.09 ±0.07	_	%/ C	
f <sub>MIN</sub> ≠ 0			15		±0.03			
Output Duty Cycle			5,10,15	_	- 50	_	%	
Output Transition Times,	effect of		5 10	-	100 50	200 100	ns	
tthL, ttlH			15	_	40	80		

the comparator input lags the signal in phase, the p-type output driver is maintained ON for a time corresponding to the phase difference. Subsequently, the capacitor voltage of the low-pass filter connected to this phase comparator is adjusted until the signal and comparator inputs are equal in both phase and frequency. At this stable point both pand n-type output drivers remain OFF and thus the phase comparator output becomes an open circuit and holds the voltage on the capacitor of the low-pass filter constant. Moreover the signal at the "phase pulses" output is a high level which can be used for indicating a locked condition. Thus, for phase comparator II, no phase difference exists between signal and comparator input over the full VCO frequency range. Moreover, the power dissipation due to the lowpass filter is reduced when this type of phase comparator is used because both the p- and n-type output drivers are OFF for most of the signal input cycle. It should be noted that the PLL lock range for this type of phase comparator is equal to the capture range. independent of the low-pass filter. With no signal present at the signal input, the VCO is adjusted to its lowest frequency for phase comparator II. Fig. 10 shows typical waveforms for a CMOS PLL employing phase comparator II in a locked condition.

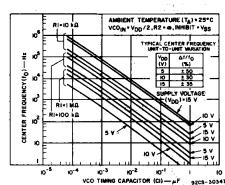


Fig. 4 - Typical center frequency as a function of C1 and R1 at V<sub>DD</sub> = 5 V, 10 V, and 15 V.

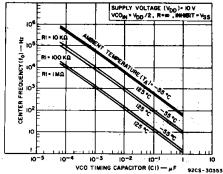


Fig. 5 — Center frequency as a function of C1 and R1 for ambient temperatures of -55°C to 125°C.

ELECTRICAL CHARACTERISTICS at TA = 25°C

CHARAC-			l .		LIMITS		ł
TERISTIC	TES	T CONDITIONS	VDD		LL TYP		UNITS
VCO Section (cont	<u> </u>		(V)	Min.	Тур.	Max.	
	<u>0)</u>		<del></del>				
Source-Follower Output (Demodulated Output): Offset Voltage  VCOIN-VDEM	RS	> 10 kΩ	5 10 15	-	1.8 1.8 1.8	2.5 2.5 2.5	· V
Linearity	R <sub>S</sub> =100 kΩ = 300 kΩ =500 kΩ	= 300 kΩ $= 5±2.5 V$			0.3 0.7 0.9	- -	%
Zener Diode Voltage (V <sub>z</sub> )	IZ	= 50 μΑ		4.45	5.5	6.15	v
Zener Dynamic Resistance, R <sub>z</sub>	l <sub>2</sub>	= 1 mA		_	40		Ω
Phase Comparator S	ection						
Term. 14 (SIGNAL IN) Input Resistance R <sub>14</sub>			5 10 15	1 0.2 0.1	2 0.4 0.2	- - -	МΩ
AC Coupled Signal Input Voltage Sensitivity* (peakto-peak)		= 100 kHz, wave	5 10 15	- - -	180 330 900	360 660 1800	mV
Propagation Delay Times, Terms. 14 to 1: High to Low Level, tpHL					225 100 65	450 200 130	ns
Low to High Level, tpLH			5 10 15	- - -	350 150 100	700 300 200	ns
3-State Propagation Delay Times, Terms. 3 to 13: High Level to High Impedance, tPHZ			5 10 15	-	225 100 95	450 200 190	ns
Terms. 14 to 13: Low Level to High Impedance, tpLZ			5 10 15	- - -	285 130 95	570 260 190	ns
Input Rise or Fall Times, t <sub>r</sub> , t <sub>f</sub> Comparator Input, Term. 3		See Fig. 5 for Phase Comp. II output loading			_ _ _	50 1 0.3	μs
Signal Input, Term. 14			5 10 15	_ _ _	- - -	500 20 2.5	μs
Output Transition Times, t <sub>THL</sub> , t <sub>TL</sub>	1		5 10 15	- - -	100 50 40	200 100 80	ns

<sup>\*</sup> For sine wave, the frequency must be greater than 10 kHz for Phase Comparator II.

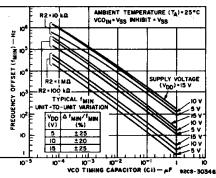


Fig. 6 — Typical frequency offset as a function of C1 and R2 for  $V_{DD}$  = 5 V, 10 V, and 15 V.

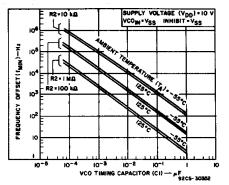


Fig. 7 — Frequency offset as a function of C1 and R2 for embient temperetures of -55°C to 125°C.

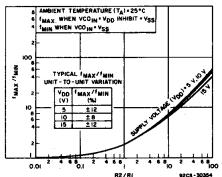


Fig. 8 — Typical f<sub>MAX</sub>/f<sub>MIN</sub> as a function of R2/R1.

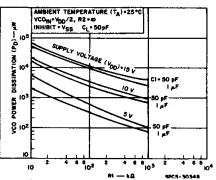


Fig. 9 — Typical VCO power dissipation at center frequency as a function of R1.

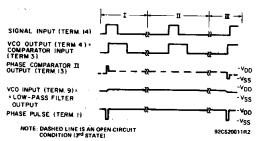


Fig. 10 - Typical waveforms for COS/MOS phase-locked loop employing phase comparator II in locked condition.

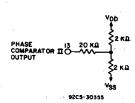


Fig. 11 — Phase comparator II output loading circuit.

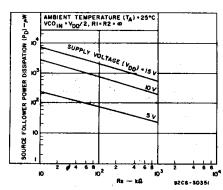


Fig. 13 – Typical source follower power dissipation as a function of Rs.

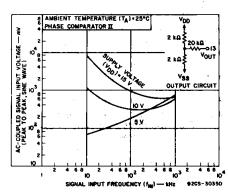
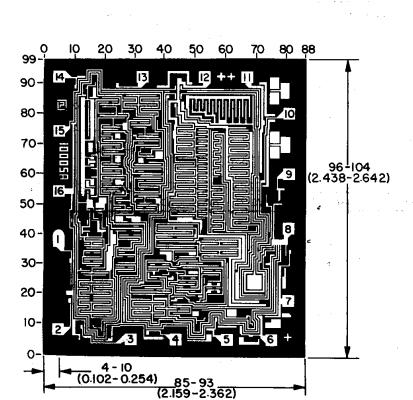


Fig. 14 — AC-coupled signal input voltage as a function of signal input frequency.



92CM-36467 Dimensions and pad layout for CD4046BH.

Dimensions in parentheses are in millimeters and are derived from the basic inch dimensions as indicated. Grid graduations are in mils ( $10^{-3}$  inch).

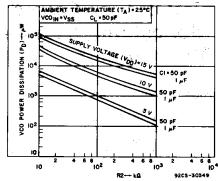


Fig. 12 – Typical VCO power dissipation at f<sub>MIN</sub> as a function of R2.

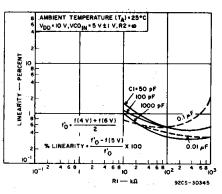


Fig. 15 — Typical VCO linearity as a function of R1 and C1 at V<sub>DD</sub> = 10 V.

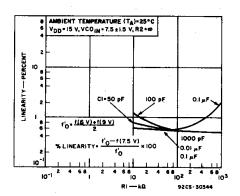


Fig. 16 – Typical VCO linearity as a function of R1 and C1 at  $V_{DD}$  = 15 V.





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#### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
5962-9466401MEA	ACTIVE	CDIP	J	16	1	Non-RoHS & Non-Green	SNPB	N / A for Pkg Type	-55 to 125	5962-9466401ME A CD4046BF3A	Samples
CD4046BE	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-55 to 125	CD4046BE	Samples
CD4046BEE4	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-55 to 125	CD4046BE	Samples
CD4046BF	ACTIVE	CDIP	J	16	1	Non-RoHS & Non-Green	SNPB	N / A for Pkg Type	-55 to 125	CD4046BF	Samples
CD4046BF3A	ACTIVE	CDIP	J	16	1	Non-RoHS & Non-Green	SNPB	N / A for Pkg Type	-55 to 125	5962-9466401ME A CD4046BF3A	Samples
CD4046BNSR	ACTIVE	SO	NS	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	CD4046B	Samples
CD4046BNSRE4	ACTIVE	so	NS	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	CD4046B	Samples
CD4046BPW	ACTIVE	TSSOP	PW	16	90	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	CM046B	Samples
CD4046BPWG4	ACTIVE	TSSOP	PW	16	90	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	CM046B	Samples
CD4046BPWR	ACTIVE	TSSOP	PW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	CM046B	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.





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- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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#### OTHER QUALIFIED VERSIONS OF CD4046B, CD4046B-MIL:

Catalog: CD4046B

Military: CD4046B-MIL

NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Military QML certified for Military and Defense Applications

## PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter	Reel Width	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CD4046BNSR	SO	NS	16	2000	(mm) 330.0	<b>W1 (mm)</b> 16.4	8.45	10.55	2.5	12.0	16.2	Q1
CD4046BPWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

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#### \*All dimensions are nominal

Device	Package Type	Package Type Package Drawing F		SPQ	Length (mm)	Width (mm)	Height (mm)
CD4046BNSR	SO	NS	16	2000	853.0	449.0	35.0
CD4046BPWR	TSSOP	PW	16	2000	853.0	449.0	35.0



SMALL OUTLINE PACKAGE



- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.



SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



## **MECHANICAL DATA**

## NS (R-PDSO-G\*\*)

## 14-PINS SHOWN

## PLASTIC SMALL-OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.



### 14 LEADS SHOWN



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. This package is hermetically sealed with a ceramic lid using glass frit.
- D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.
- E. Falls within MIL STD 1835 GDIP1-T14, GDIP1-T16, GDIP1-T18 and GDIP1-T20.

## N (R-PDIP-T\*\*)

## PLASTIC DUAL-IN-LINE PACKAGE

16 PINS SHOWN



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
- The 20 pin end lead shoulder width is a vendor option, either half or full width.



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