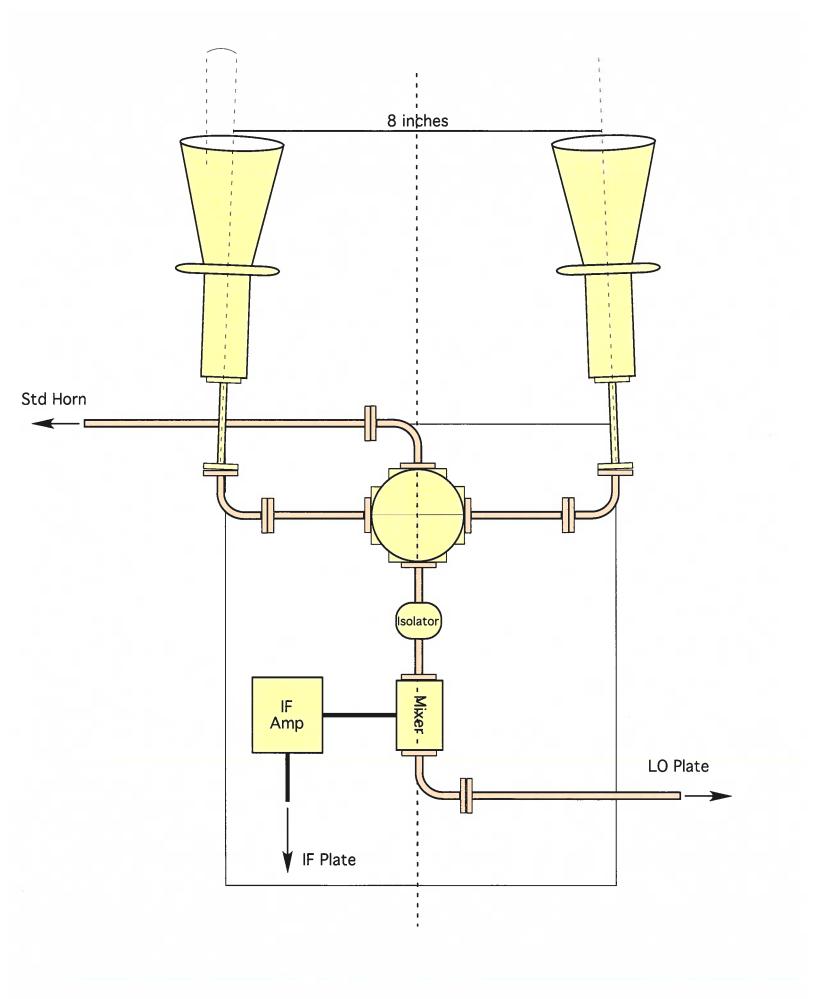
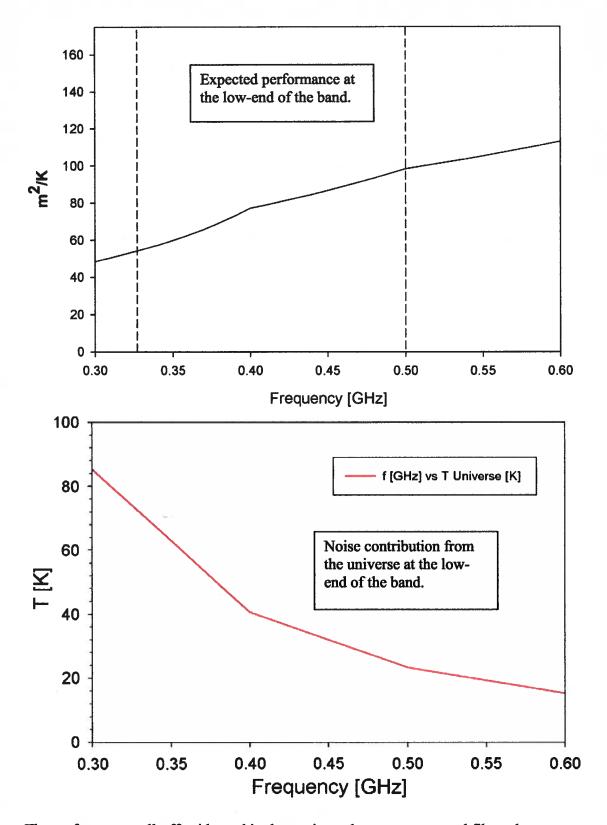
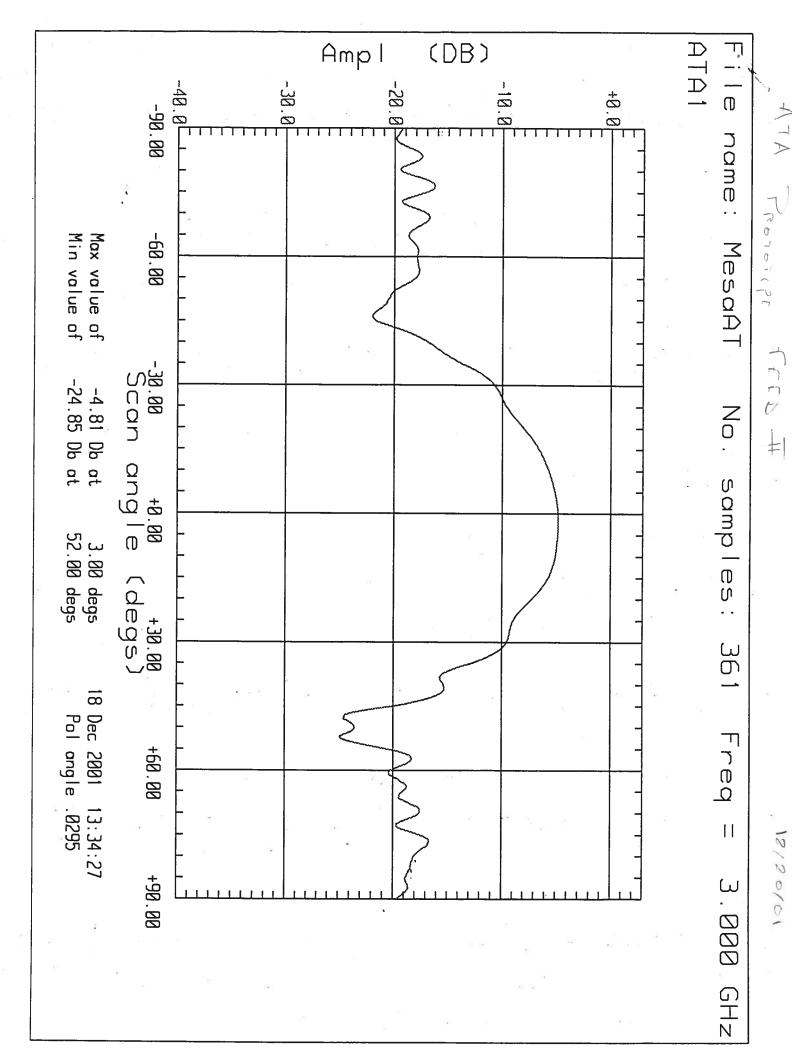
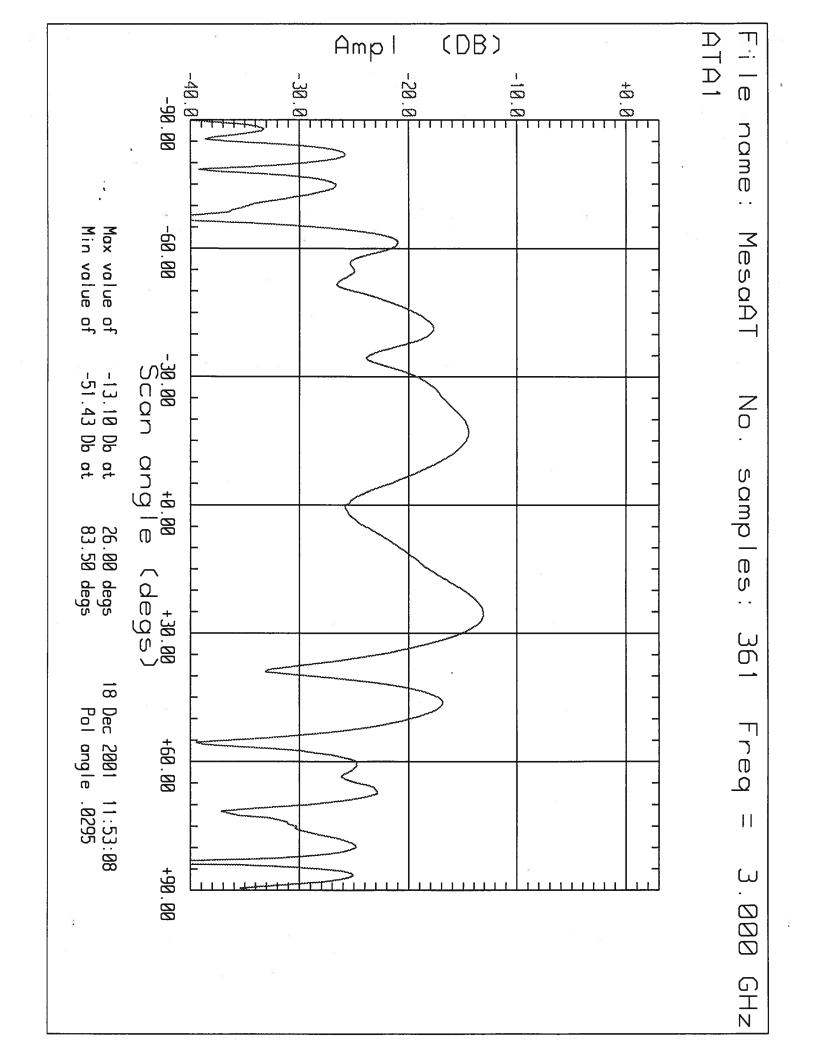
But "start button the wait 2) press "use preset button" 3) dick on "all modes" (4) press "trace" bother on the face emables trace choice 5 to vernoue purple trace; select trace? 6 select 'view blank" Deick on "blank", no traces left on seveen (8 select trace 1. Then select the trace you want. 9) select "amplitude" to vent cetive trace of (18) adjust level with knob (1) hit trace again (12) "trace average" gives an average (3) to print this: Save if en Moscreen: view again (14) to save: poess "save" away the buttons. (15) Save as Tack 2013 and the screen listing in light that; type something in (18) to shufdown

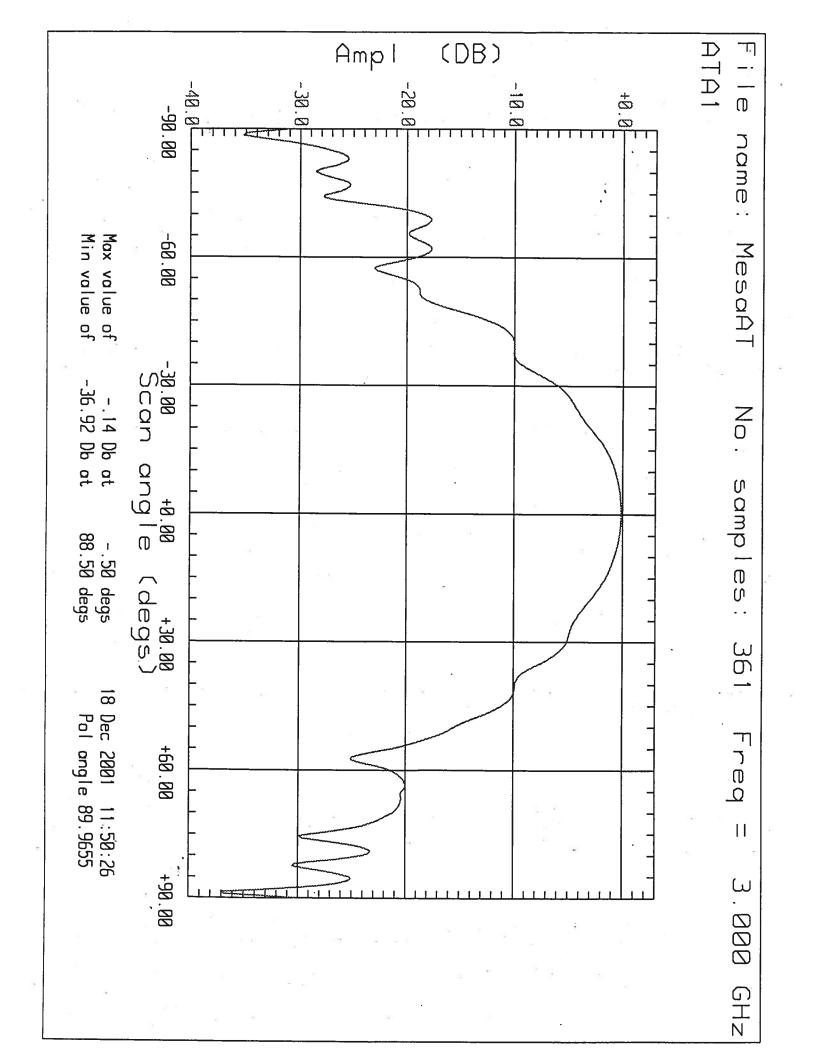


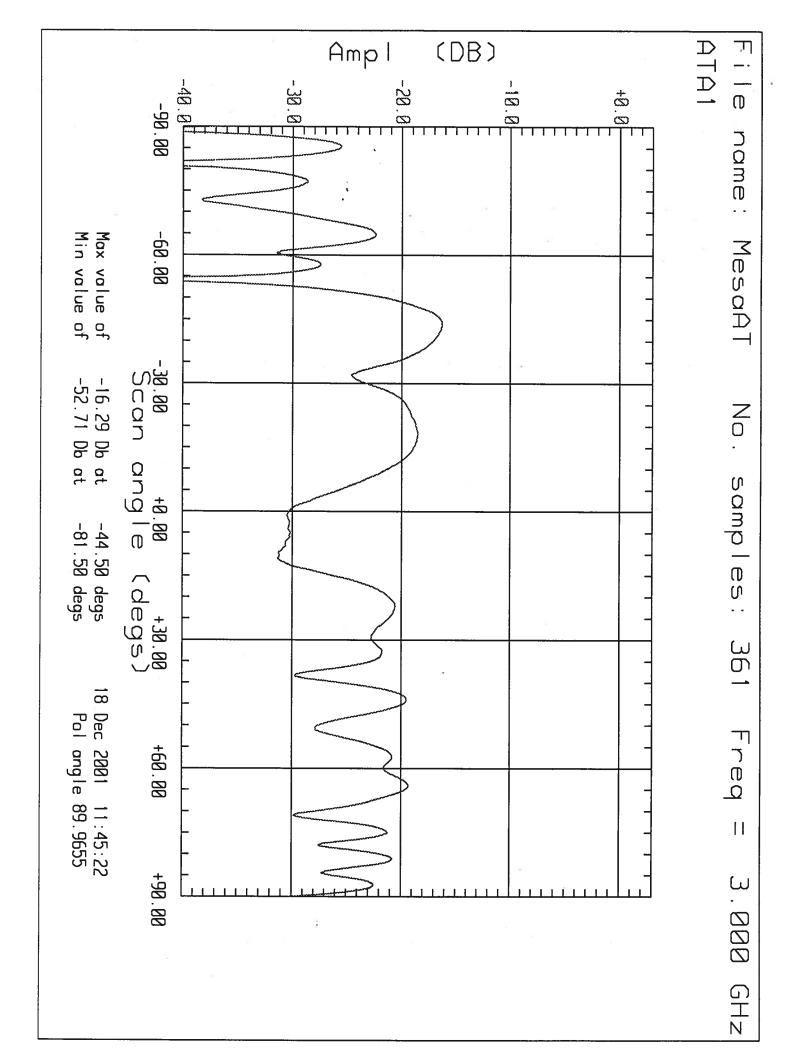


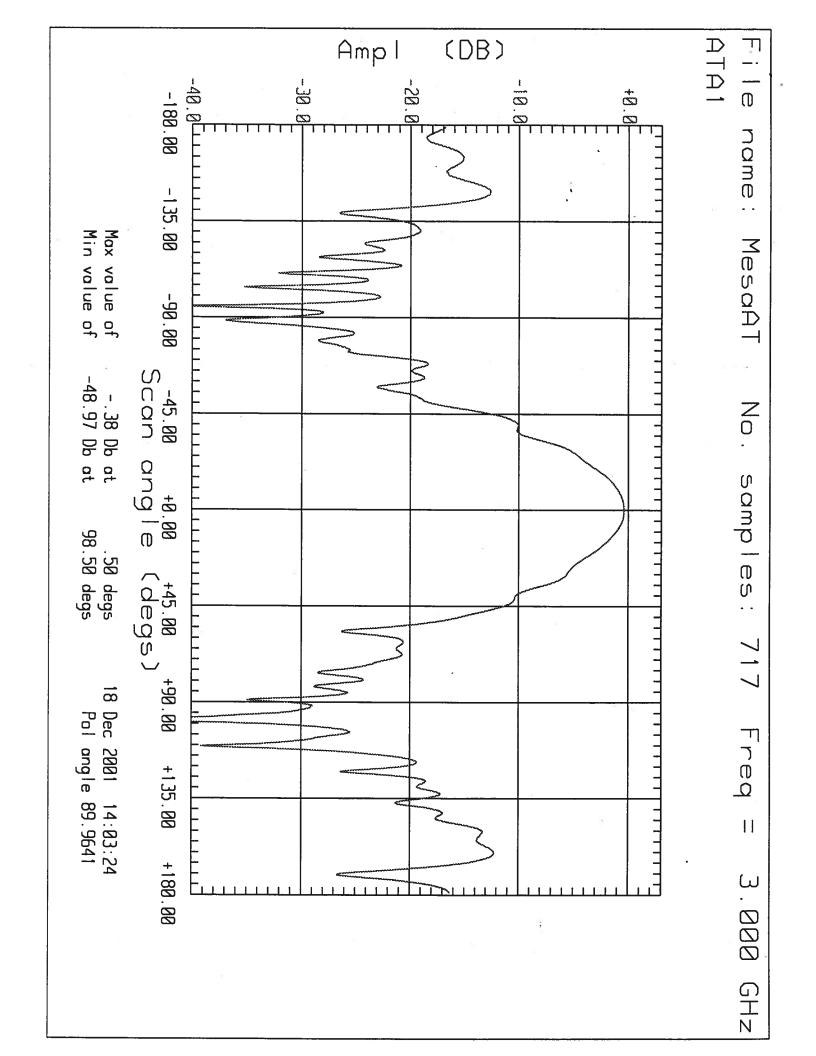
The performance roll-off evidenced in the top is partly components and filters, but predominantly the increased noise from the universe at lower frequencies (bottom figure).

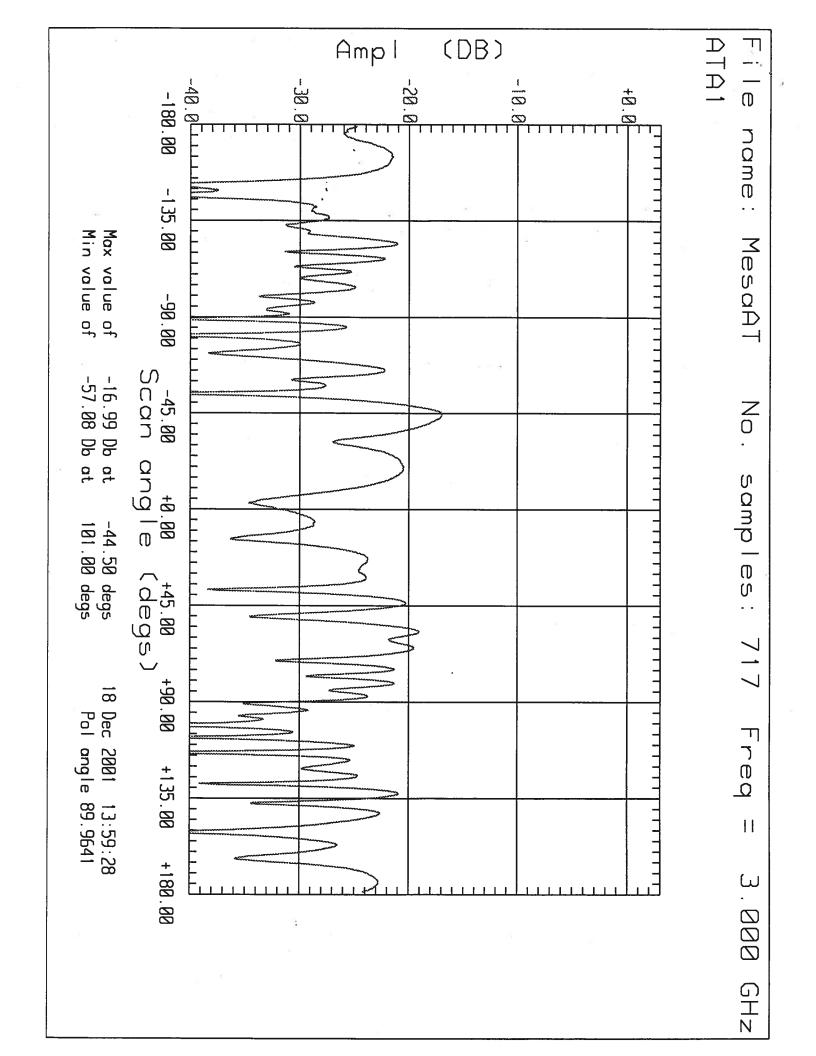












JL

intransa.mcd

jbl

28jun02

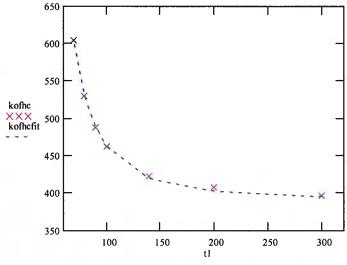
data from CIC database

t1 :=
$$\begin{bmatrix} 300 \\ 200 \\ 140 \\ 100 \\ 90 \\ 80 \\ 70 \end{bmatrix}$$
 ·K kofhcfit :=
$$\begin{bmatrix} \frac{520}{t1} - 22 \\ \frac{520}{t1 \cdot K} - 22 \\ \frac{520}{t1 \cdot K} - 22 \end{bmatrix} \cdot \frac{W}{m \cdot K}$$
 kofhcf(t) :=
$$\begin{bmatrix} \frac{520}{t1} - \frac{520}{t1} \\ \frac{520}{t1} - \frac{520}{t1} \\ \frac{520}{t1} - \frac{520}{t1} \end{bmatrix} \cdot \frac{W}{t1}$$

kofhcf(t) :=
$$\left[\frac{520}{\frac{t}{1 \cdot K} - 22} \right]^{2.25} + 391 \right] \cdot \frac{W}{m \cdot K}$$
 t := 200·K kofhcf(t) = 402.157 • $\frac{W}{m \cdot K}$

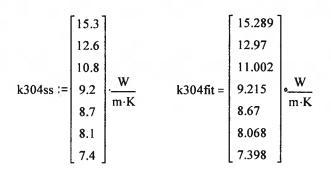
kofhc :=
$$\begin{bmatrix} 396 \\ 407 \\ 422 \\ 462 \\ 487 \\ 529 \\ 604 \end{bmatrix} \cdot \frac{W}{m \cdot K} \qquad \text{kofhcfit} = \begin{bmatrix} 395.092 \\ 402.157 \\ 419.137 \\ 462.416 \\ 488.244 \\ 530.09 \\ 603.919 \end{bmatrix} \cdot \frac{W}{m \cdot K}$$

kofhcfit is pretty good from 70 to 300 K

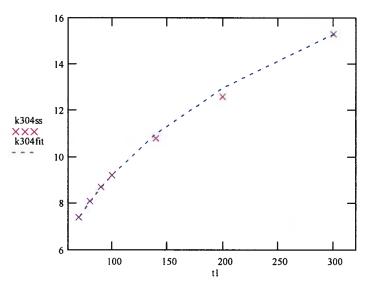


k304fit :=
$$\left[18. \cdot \left(\frac{t1}{1 \cdot K}\right)^{.15} - 0.0018 \cdot \frac{t1}{1 \cdot K} - 26.52\right] \frac{W}{m \cdot K}$$

$$k304f(t) := \left[18. \cdot \left(\frac{t}{1 \cdot K}\right)^{.15} - 0.0018 \cdot \frac{t}{1 \cdot K} - 26.52\right] \cdot \frac{W}{m \cdot K} \quad t := 100 \cdot K \quad k304f(t) = 9.215 \cdot \frac{W}{m \cdot K}$$



k304fit is pretty good from 70 to 300 K



odss := 0.014 · in

 $idss := .010 \cdot in$

Ass := $\frac{\pi}{4}$ (odss·odss – idss·idss)

Ass = $4.864 \cdot 10^{-8}$ ·m²

tCu := 0.000150·in

 $ACu := \pi \cdot (odss + tCu) \cdot tCu$

 $ACu = 4.302 \cdot 10^{-9} \cdot m^2$

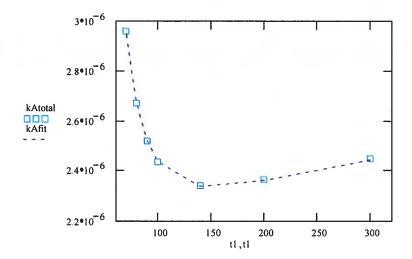
 $kAtotal := kofhc \cdot ACu + k304ss \cdot Ass$

 $kAfit := kofhcfit \cdot ACu + k304fit \cdot Ass$

 $kAf(t) := kofhcf(t) \cdot ACu + k304f(t) \cdot Ass$

$$kAtotal = \begin{bmatrix} 2.448 \cdot 10^{-6} \\ 2.364 \cdot 10^{-6} \\ 2.341 \cdot 10^{-6} \\ 2.435 \cdot 10^{-6} \\ 2.518 \cdot 10^{-6} \\ 2.67 \cdot 10^{-6} \\ 2.958 \cdot 10^{-6} \end{bmatrix} \circ W \cdot \frac{m}{K} \quad kAfit = \begin{bmatrix} 2.443 \cdot 10^{-6} \\ 2.361 \cdot 10^{-6} \\ 2.338 \cdot 10^{-6} \\ 2.438 \cdot 10^{-6} \\ 2.522 \cdot 10^{-6} \\ 2.673 \cdot 10^{-6} \\ 2.958 \cdot 10^{-6} \end{bmatrix} \circ W \cdot \frac{m}{K}$$

.00015in = ,00038cm = 38/0m



Thot
$$:= 310 \cdot K$$

Tmid $:= 80 \cdot K$

Tcold := $77 \cdot K$

lhot $= 2.6 \cdot in$

lhot = 6.604 °cm

 $lcold := 4.5 \cdot in$

lcold = 11.43 °cm

 $\frac{kAf(t)}{lhot}dt$?

 $Qdot = 8.34 \cdot 10^{-3} \quad \text{°W}$

Now estimate the required copper cross section on the balun ignoring substrate conductivity.

$$I := \int_{-T_{cold}}^{T_{mid}} kofhcf(t) dt$$

 $I = 1.616 \cdot 10^3 \cdot \frac{W}{m}$

a little rough (no)

$$Acold := \frac{Qdot \cdot lcold}{I}$$

Acold = $9.143 \cdot 10^{-4}$ •in²

$$dcold := \sqrt{\frac{Acold \cdot 4}{\pi}}$$

temp :=
$$\begin{bmatrix} 60 \\ 70 \\ 80 \\ 90 \end{bmatrix}$$
 ·K kalumina := $\begin{bmatrix} 155 \\ 162 \\ 155 \\ 135 \end{bmatrix}$ · $\frac{W}{m \cdot K}$

US Propheros.

$$kqtzpar := \begin{bmatrix} 78 \\ 62 \\ 53 \\ 46 \\ 42 \end{bmatrix} \frac{W}{m \cdot K} \qquad kqtzperp := \begin{bmatrix} 40 \\ 32 \\ 27 \\ 25 \\ 24 \end{bmatrix} \frac{W}{m \cdot K} \qquad kqtzfused := \begin{bmatrix} 0.46 \\ 0.52 \\ 0.57 \\ 0.65 \\ 0.71 \end{bmatrix} \frac{W}{m \cdot K} \qquad 90 \quad K$$

for a trapezoidal substrate shape:

$$wcold := 0.800 \cdot in$$

$$area(z) := thk \cdot \left[wcold - (wcold - wmid) \cdot \frac{z}{lcold} \right]$$

areaint :=
$$\int_{0}^{\bullet} \operatorname{lcold} \frac{1}{\operatorname{area}(z)} dz$$

areaint =
$$3.101 \cdot 10^4 \cdot m^{-1}$$

for alumina take k = 150 w/m/K for whole temperature range

ksub :=
$$150 \cdot \frac{W}{m \cdot K}$$

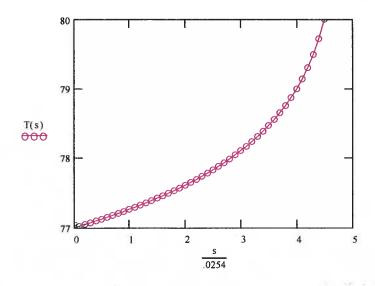
$$Qdotsub := \frac{\frac{m \cdot K}{ksub \cdot (Tmid - Tcold)}}{areaint} \quad ?$$

$$Qdotsub = 0.015 \circ W$$

calculate the temperature profile

$$s := 0 \cdot in_{1} \cdot in_{2} \cdot in_{3} \cdot in_{4}$$

$$T(s) := \begin{bmatrix} \int_{0}^{s} & \frac{Qdotsub}{ksub \cdot area(z)} dz \\ \end{bmatrix} + Tcold$$



Now look at electrical properties.

First check consistency between thermal and electrical conductivity.

$$\rho \text{ fit } := \boxed{7.2 \cdot 10^{-9} \cdot \boxed{\frac{(13 - 25)}{132}}^{1.24}}$$

$$\rho \text{ fit } 1 := \boxed{8.4 \cdot 10^{-9} \cdot \boxed{\frac{(13 - 40)}{132}}^{1.1}}$$

$$\rho \text{ fit } 1 := \boxed{8.4 \cdot 10^{-9} \cdot \boxed{\frac{(13 - 40)}{132}}^{1.1}}$$

$$\rho \text{ fit } 1 := \boxed{8.4 \cdot 10^{-9} \cdot \boxed{\frac{(13 - 40)}{132}}^{1.1}} \cdot \text{ ohm \cdot m}$$

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$$\rho \text{ fit } 1 := \boxed{8.4 \cdot 10^{-9} \cdot \boxed{\frac{(13 - 40)}{132}}^{1.1}} \cdot \text{ ohm \cdot m}$$

$$\rho \text{ fit } 1 := \boxed{\frac{(1.771 \cdot 10^{-8})}{3.529 \cdot 10^{-9}}} \cdot \text{ ohm \cdot m}$$

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$$\rho \text{ fit }$$

now check using Weideman-Franz law:

50

100

Kittel Solid State Physics gives L = 2.33e-8 at 373K, 2.23e-8 at 273K, and ~0.24e-8 at 15K.

200

250

300

150

t3

Lorentz :=
$$\frac{1}{\text{kofhcfit} \cdot \rho \text{fit1a}}$$

$$\text{Lmeas} := \begin{bmatrix} 2.33 \cdot 10^{-8} \\ 2.23 \cdot 10^{-8} \\ 0.24 \cdot 10^{-8} \end{bmatrix} \quad \text{tL} := \begin{bmatrix} 373 \\ 273 \\ 15 \end{bmatrix}$$

$$\text{Lorentz} = \begin{bmatrix} 2.332 \cdot 10^{-8} \\ 2.087 \cdot 10^{-8} \\ 1.853 \cdot 10^{-8} \\ 1.632 \cdot 10^{-8} \\ 1.566 \cdot 10^{-8} \\ 1.497 \cdot 10^{-8} \\ 1.42 \cdot 10^{-8} \end{bmatrix}$$

$$\text{Lorentz} = \begin{bmatrix} 2.33 \cdot 10^{-8} \\ 2.23 \cdot 10^{-8} \\ 0.24 \cdot 10^{-8} \end{bmatrix} \quad \text{tL} := \begin{bmatrix} 373 \\ 273 \\ 15 \end{bmatrix}$$

Now calculate electrical loss in the copper plated stainless steel section at 10 GHz.

$$Z0 := 204$$

Now recall from page 3 that the thermal conductivity of this section of transmission line is nearly constant from 310 K down to 90K. Therefore assume a linear temperature profile. Use pfit1.

$$T(z) := \frac{\text{Thot}}{1 \cdot K} - \left(\frac{\text{Thot} - \text{Tmid}}{1 \cdot K} \right) \cdot \frac{z}{\text{lhot}}$$

TB :=
$$\frac{2 \cdot \text{sod}}{\pi \cdot \text{d} \cdot Z0 \cdot \sqrt{\text{sod}^2 - 1}} \cdot \sqrt{\pi \cdot 4 \cdot \pi \cdot 10^{-7} \cdot \text{f} \cdot 8.4 \cdot 10^{-9} \cdot \left[\frac{(T(z) - 40)}{132} \right]^{1.1}} \cdot T(z) \cdot \text{Kdz}$$

$$TB = 2.49 \cdot K$$

nov 27, 01 jbl

```
input transmission line calculations
```

summary

0.010 dia kovar pins with 0.000150 in thick Cu plate, 5 cm long

```
ave thermal cond. for 347 ss for 100-320K = 125 mW/cm/K ave thermal cond. for constantan 100-320K = 230 mW/cm/K ave thermal cond. for monel for 100-320K = 200 mW/cm/K (all from my cryo notes; monel and constantan magnetic constantan constant
```

composi	ition:	Fe	Ni	Cu	Co	Cr	be sw	K@20C (W/m.K)	rho (ohm.cm)
	monel	.01	.655	.30					
	inconel 718	.185	.525			.19		11.4	
	constantan								
	incoloy909	.42	.38	0	.13	0		14.8	
	Kovar	.53	.29		.17			17.3	4.9e-5
	(Carpenter Technologies)								
	NILO alloy K	.53	.29		.17			16.7	4.3e-5
	(Inco Alloys International)								
	301 ss annea	led .75	.07			.17		~15.3	
	302 ss annea	led .72	.09			.18		~15.3	
	304 ss annea	led .71	.092			.19		~15.3	
	347 ss annea	led .70	.11	•		.18		~15.3	

Cu plated Kovar for 5 cm:

take pin .010 inches OD, gives $A = 5.07e-4 \text{ cm}^2$

take 1 = 5 cm

take deltaT = 320-100 = 220 K

take Kave = 150 mW/cm.K (167 - 10%)

Qdot = 3.35 mW

for copper plating thickness of 0.00015 inches, get A = 3.04e-5 cm²; the plating thickness is almost 2 skin depths at 1 GHz and 295 K, increasing to 5 skin depths at 1 GHz and 77 K. see skindepth.xls

take 1 = 5 cm

take deltaT = 320-100 = 220 K

take Kave = 4400 mW/cm.K

Qdot = 5.89 mW

for 4 lines, get Qdot =37.0 mW

solid Cu for 23 cm:

take OD = 0.020, $A = 2.03e-3 \text{ cm}^2$

1 = 23 cm

Kave (90 K) = 5000 mW/cm.K

Qdot = 9.25 mW per wire

get deltaT = 21 K

input transmission line resonant frequency

from Mark's handbook: for simply supported ends f = (pi/2)*x/1/1 where x = sqrt(E*I*g/A/rho), l = length, $g/rho = in^3/slug$ for l = 2 in, get f = 215 Hz

thermal expansion issues

deltaL/L = 2.9e-3 for 300K to 80K

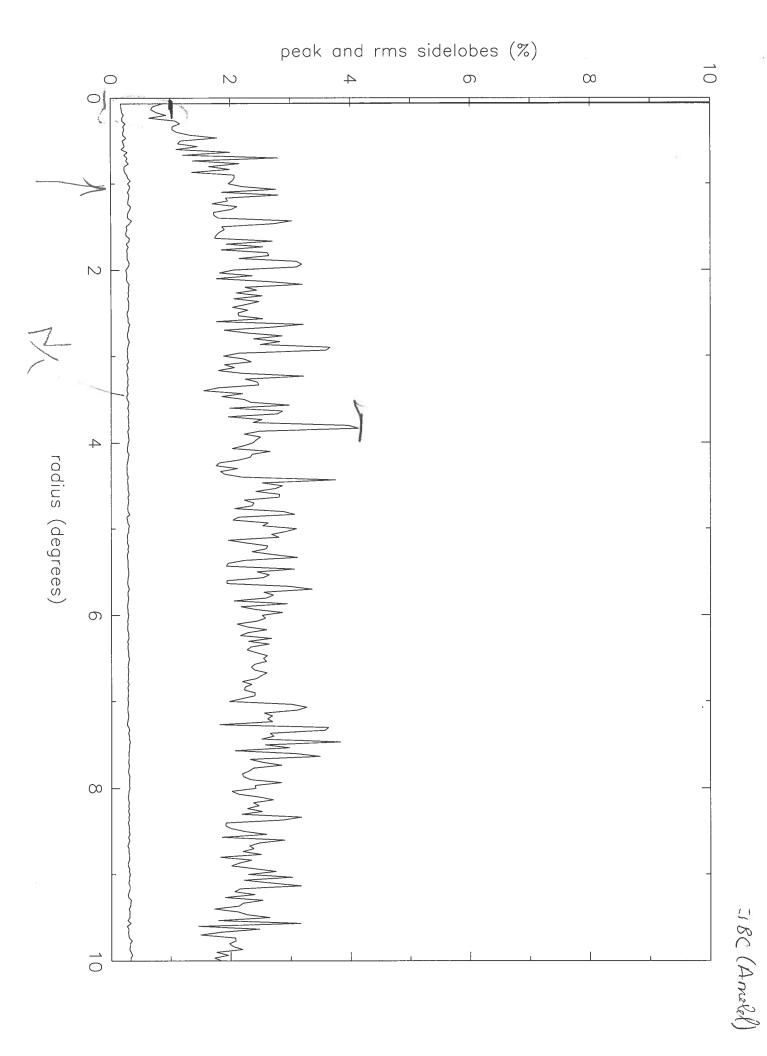
deltaL/L = 3.0e-3 for 300K to 60K for 304 SS

so, roughly, for a linear temp gradient from 300 to 60, and L=6 in expect deltaL = 1.5e-3*6=0.009 in. this could be accomodated in a diaphram dewar end as long as the diaphram is large enough.

heat sinking the Cu transmission lines to 70 K:

material	K@70K (mW/cm.K)	K@300K	epsilon (@ 1 MHz)	
<pre>fused quartz, SiO alumina, Coors AD-998</pre>	3.8	14. 300.	3.75 9.8	

for fused quartz standoffs,



ATA Front-End Critical Design Review Agenda

Wednesday, July 9, 2003

Campbell Hall 544

	9:30	Assemble/Coffee				
	10:00 Introductions		Tom Pierson/Leo Blitz			
	10:15	6 Overview	Dave DeBoer			
	10:45	PAX	Dave DeBoer			
	11:15	F/O Link/Cooler	Ed Ackerman			
1:00	12:15 Lunch		Room 501			
	1:00	Executive committee				
	1:30	Feed	Jack Welch			
	2:15	Coffee				
	2:30	LNA	Sandy Weinreb/Niklas Wadefalk			
	3:00	Dewar Integration and Cryocooler John Lugten				
	4:00	Feed Mechanicals	Matt Fleming			
	4:30	Lab Tour				
	5:00	Executive Session				
	7:00	Dinner for reviewers and presente	ers			

140k van gets 3000 WBA13 or 1400 WBAL2 Four wafers start out, expect 1/2 to work

com elevation effect? Evys cost estimate ~ 3K

5000 / 18 2.

ATA TAP Agenda Dec 17-18 2002

Space Sciences Laboratory UCB

Tuesday, December	17	SSL	Annex	Conference	Room
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- 8:30 Welcome: Leo Blitz and Tom Pierson
- 8:45 Overview of ATA Status (Dave Deboer)
- 10:15 Coffee
- 10:30 The ATA Antenna (Matt Fleming)
- 11:15 The Pulse Tube Cooler/Compressor (John Lugten)
- 11:45 Executive Session
- 12:00 Lunch (catered in)
- 1:00 First Operation and Tests (Dave Deboer)

 Pointing and tracking Tests (Geoff Bower)
- 2:00 Feed/Receiver Developments (Jack Welch)
- 2:30 ATA Control Software Archetecture (Gerry Harp/Rob Ackerman)
- 3:00 Coffee
- 3:15 Array Configuration (Douglas Bock)
- 3:45 Correlator Update (Leo Blitz)
- 4:00 Executive Session
- 7:00 Dinner at John Dreher's

Wednesday, December 18 Room 105 SSL (new building)

- 8:30 Hat Creek Permitting and site overview (Douglas Bock)
- 9:00 Report of the Antenna CDR (John Dreher)
- 9:30 ATA Schedule Recovery Plan (Mike Davis)
- 10:00 Coffee
- 10:15 Construction Plan and Budget (John Dreher)
- 11:30 Executive Committee
- 12:00 Lunch (catered in)
- 1:00 Executive Committee, Discussion and Report Drafting

Approach #3: Integrated Laser/EA Modulator

Measured Performance w/1 km SM fiber and Discovery Photodetector SFDR₃ \geq 95 dB·Hz^{2/3}; SFDR₂ \geq 79 dB·Hz^{1/2} detector bias = 6 V, fundamental frequencies \sim 5 GHz) (Modulator biased for minimum 2nd-order distortion, Measured Distortion Products Input Power at Fundamental (dBm) Fundamental frequencies $CDR \ge 137 dB \cdot Hz$ 2nd-order distortion $NF \le 42 dB$ 9 တ္ ဓ္က -100 -140 -120 Output Power (dBm) 12 (Modulator biased for minimum 2nd-order distortion) At bias for minimum 2nd-order distortion: Measured Link Frequency Response vs. 9 qВ Modulator Bias Voltage r Frequency (GHz) $\Delta G_{0.5-11 \text{ GHz}} < 4 \text{ dB}$ $BW_{3dB} \sim 9.5 GHz$ $G_{11~GHz}\sim -\,37~dB$ \rightarrow -58 -30 -32 -36 -34 4 42 38 Link Gain (dB)

Photonic Systems, Inc.

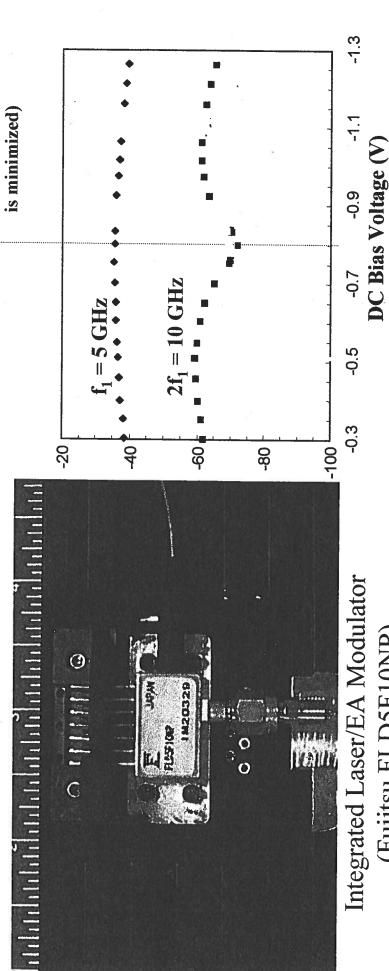
www.photonicsinc.com

Approach #3: Integrated Laser/EA Modulator

Measured Signal and Distortion Characteristics vs. DC Bias Voltage

Optimum bias voltage

(2nd-order distortion



(Fujitsu FLD5F10NP)

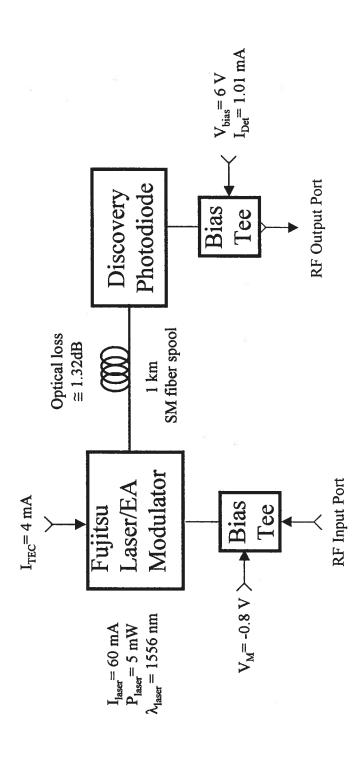
Photonic Systems, Inc.

www.photonicsinc.com

(Input signal: 0 dBm at $f_1 = 5$ GHz)

Approach #3: Integrated Laser/EA Modulator

Test Configuration (Room Temperature)



Single-mode fiber connection was achieved using FC/PC connectors.

Photonic Systems, Inc.

www.photonicsinc.com

PTA Single Dish Tests

0. Pedestal tests

Survival pull test 🗸

GPS XYZ coordinates of upper flange and comparison to survey stake on ground (nice but not essential)

1. General Mechanical Inspection

Wiring, Power, and limits

Verify code compliance of antennas

Brakes

Weather proofing and radome (may be kludge at first)

Grounding checks

antenna to ground

intra-antenna grounds

Motion (under local control?)

Basic motion, slew rates, both hot (day) & cool (night)

Antenna hard-stop motion tests

cable wrap range and operation

safety items: E-stop, stow, power failure, wind stow, limits

azimuth back-drive

Measure alignment of optics with photogrammetry

Voice communications from antenna to control room

2. Dynamical/Mechanical Tests

Tiltmeter checks for AZ rotational stability

Resonant frequency measurements

Acceleration measurements, slew rate check

Motor current, power consumption, power factor

Determine basic antenna mount coordinates (zero points and tilt)

3. Tracking Tests (under computer control)

Optical (with BIMA camera)

check and optimize servo loop for smooth tracking

optical pointing model (optional)

determine non-reproducible errors (if we can)

Radio (with RT front end) at several TBD frequencies

focus curves

measure beam pattern using geostationary

determine radio pointing model with strong sources

verify smooth tracking with edge of sun or other strong source

Aeff/Tsys measurements

Cal source calibration (if available)

Tsys measurements with ambient absorber and cold sky

4. Other Tests

Measure antenna RFI/EMI characteristics

Mesure sound levels

Try both large and small stepper motors

Oil leak test

Here are some are the PTA "single-dish" tasks from the PTA Test Plan with some notes on suitable RF systems.

focus curves

Maybe OK with Spectrum Analyzer (SA), but if SNR requires use of a satellite, using a second antenna with fixed focus as a reference might be much better. That would require the RPA backend. Also requires a focus drive (possibly manual)

measure beam pattern using geostationary

For near in beam pattern, we can use a spectrum analyzer. But it would be very nice to be able to measure the voltage pattern over the whole sky like you did for the RPA dishes. I'd like to do that before and after the shroud is installed. Since the voltage method provides the phase information, it should be possible to do holography as well, which would provide a useful check on the photogrammetry and also demonstrate that we are providing the correct illumination of the primary.

determine radio pointing model with strong sources

The System Equivalent Flux Density for a single PTA dish with the room-temperature front end will be about 14,000 Jy. Since the strongest sources are of order 100-1000 Jy at L-band, we'll want BW*tau >> 1E4, which we can achieve much faster with the 10 MHz BW of the RPA system than with the SA -- in fact I have doubts that the SA can do this at all.

verify smooth tracking with edge of sun or other strong source SA can do this

Aeff/Tsys measurements

The higher BW*tau of the RPA backend will be a big help with this.

Cal source calibration

Ditto

Tsys measurements with ambient absorber and cold sky SA can do this.

Measure antenna RFI/EMI characteristics

SA can do this over full range, but RPA can go deeper, faster over it's RF range.