



第三讲

短时Fourier变换

Short Time Fourier Transform (STFT)

彭志科

Email: z.peng@sjtu.edu.cn

上海交通大学

机械系统与振动国家重点实验室



● 发展简史



Jean Baptiste Joseph Fourier (1768 - 1830)

➤ 1807 - "All periodic functions can be expressed as a weighted sum of trigonometric function", Denied publication by Lagrange, Legendre and Laplace

➤ 1822 - The Analytic Theory of Heat published, J.C Maxwell - Fourier Theory: a great mathematical poem

. . .

· · · · Peter Gustav Lejeune Dirichlet, Bernhard Riemann

• • •

➤ 1965, Cooley & Tukey: Fast Fourier Transform - King of transforms





● 简介

Fourier级数

$$f(t) = \sum_{n = -\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$f(t) = \sum_{n = -\infty}^{\infty} c_n e^{jn\omega_0 t} \qquad c_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jn\omega_0 t} dt \qquad \omega_0 = 2\pi/T$$

Fourier变换

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) e^{j\omega t} d\omega \qquad F(j\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

$$F(j\omega) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t}dt$$

本质思想:线性空间的正交基分解和重构





● 主要性质

$$a_1 f_1(t) + a_2 f_2(t) \stackrel{\mathcal{F}}{\longleftrightarrow} a_1 F_1(j\omega) + a_2 F_2(j\omega)$$

时移性质

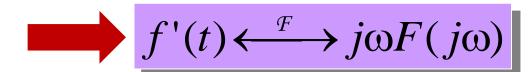
$$f(t-t_0) \longleftrightarrow F(j\omega)e^{-j\omega t_0}$$

频移性质

$$f(t)e^{j\omega_0} \stackrel{\mathcal{F}}{\longleftrightarrow} F[j(\omega-\omega_0)]$$

微分性质

$$f(t) \stackrel{\mathcal{F}}{\longleftrightarrow} F(j\omega)$$
 and $\lim_{t \to \pm \infty} f(t) = 0$







● 主要性质

微分性质

$$f(t) \stackrel{\mathcal{F}}{\longleftrightarrow} F(j\omega)$$
 and $\lim_{t \to \pm \infty} f(t) = 0$

$$f^{(n)}(t) \stackrel{\mathcal{F}}{\longleftrightarrow} (j\omega)^n F(j\omega)$$

积分性质

$$f(t) \stackrel{\mathcal{F}}{\longleftrightarrow} F(j\omega)$$
 and $\int_{-\infty}^{\infty} f(t)dt = F(0) = 0$

$$\mathcal{F}\left[\int_{-\infty}^{t} f(x)dx\right] = \frac{1}{j\omega} F(j\omega)$$

利用Fourier变换,可将微积分方程求解问题变成简单代数运算问题





● 示例

$$m\ddot{x}(t) + kx(t) = g(t)$$

$$m\left(\sum_{n=-\infty}^{\infty}\left(-n^2\omega_0^2c_ne^{jn\omega_0t}\right)\right)$$

$$+k\sum_{n=-\infty}^{\infty}c_{n}e^{jn\omega_{0}t}=\sum_{n=-\infty}^{\infty}d_{n}e^{jn\omega_{0}t}$$

$$\left\langle m \left(\sum_{n=-\infty}^{\infty} \left(-n^2 \omega_0^2 c_n e^{jn\omega_0 t} \right) + k \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t} \right) \right\rangle$$

$$\langle e^{jr\omega_0 t} \rangle = \left\langle \sum_{n=-\infty}^{\infty} d_n e^{jn\omega_0 t}, e^{jr\omega_0 t} \right\rangle$$

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$\ddot{x}(t) = \sum_{n=-\infty}^{\infty} \left(-n^2 \omega_0^2 c_n e^{jn\omega_0 t} \right)$$

$$g(t) = \sum_{n=-\infty}^{\infty} d_n e^{jn\omega_0 t}$$

$$c_r = d_r / (k - r^2 \omega_0^2 m)$$



$$-r^2\omega_0^2c_r m + kc_r = d_r$$







● 主要性质

卷积性质

$$f_1(t) * f_2(t) \stackrel{\mathcal{F}}{\longleftrightarrow} F_1(j\omega) F_2(j\omega)$$

乘积性质

$$f_1(t)f_2(t) \stackrel{\mathcal{F}}{\longleftrightarrow} \frac{1}{2\pi} F_1(j\omega) * F_2(j\omega)$$

卷积定义

$$f(t) = \int_{-\infty}^{\infty} f_1(\tau) f_2(t - \tau) d\tau$$

$$= f_1(t) * f_2(t)$$





Matlab 实现

fft(X) is the discrete Fourier transform (DFT) of vector X.

For matrices, the FFT operation is applied to each column.

典型用法:
$$Y = fft(X)$$
, % $size(Y) = size(X)$

$$Y = fft(X,N)$$
 % the N-point FFT

X - 1×xLen; %信号的长度为xLen

Fs - Sampling Frequency

Y-1×xLen;%频谱

频率分辨率

$$\Delta F = \frac{Fs}{xLen} = \frac{Fs}{s\Delta t} = \frac{1}{s}$$
 s- 采样时间长度 / sec

如果某分量 的频率为f

$$I = \frac{f}{\Delta F} = fs$$

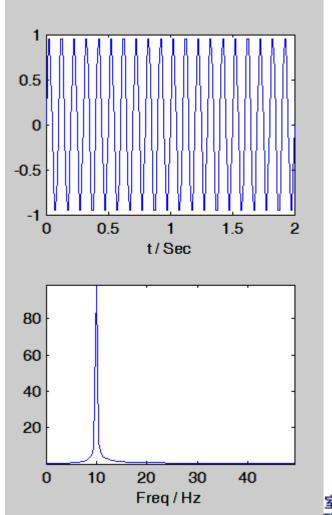
如果I非整数,则出 现能量泄漏



Matlab - 示例

```
function Test_FFT(),
SampFreq = 100;
td = 2;
t = 0.1/SampFreq:td;
%t = t(1:end-1);
Sig = sin(2*pi*10*t);
SigLen = length(Sig);
Spec = fft(Sig, SigLen);
% the lenght of sig is td*SampFreq + 1;
SpecLen = length(Spec);
Freq = 0:SampFreq/SpecLen:SampFreq;
Freq = Freq(1:SpecLen);
subplot(211), plot(t, Sig); xlabel('t / Sec')
subplot(212), plot(Freq(1:end/2), abs(Spec(1:end/2)));
xlabel('Freq / Hz')
axis('tight')
```

I非整数

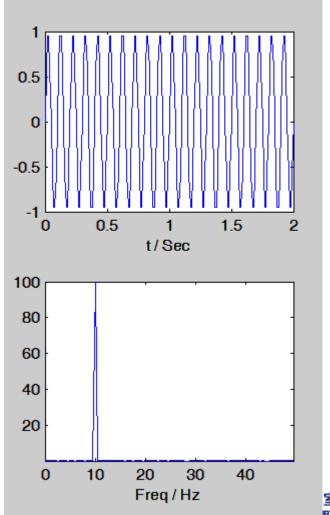




Matlab - 示例

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I整数

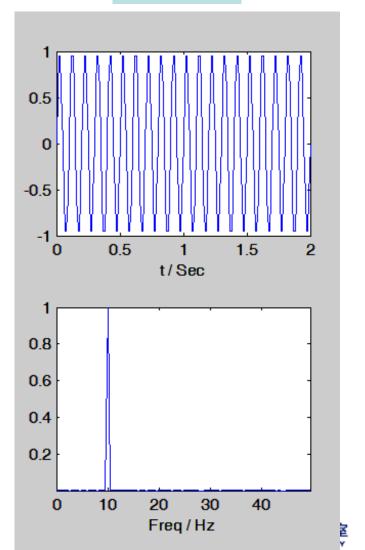




● Matlab - 示例

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function Test_FFT(),
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Sig = sin(2*pi*10*t);
SigLen = length(Sig);
Spec = fft(Sig, SigLen-1);
% the lenght of sig is td*SampFreq + 1;
SpecLen = length(Spec);
Spec = 2*Spec/ SpecLen;
Freq = 0:SampFreq/SpecLen:SampFreq;
Freq = Freq(1:SpecLen);
subplot(211), plot(t, Sig); xlabel('t / Sec')
subplot(212), plot(Freq(1:end/2), abs(Spec(1:end/2)));
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```

归一化

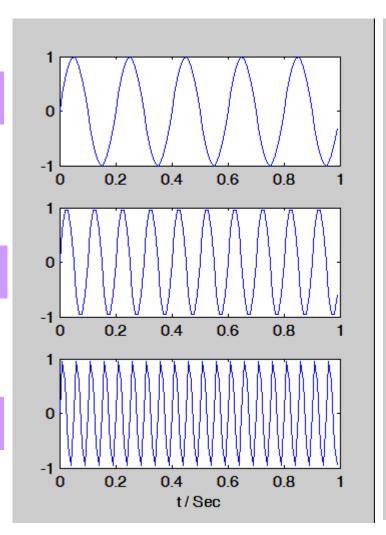


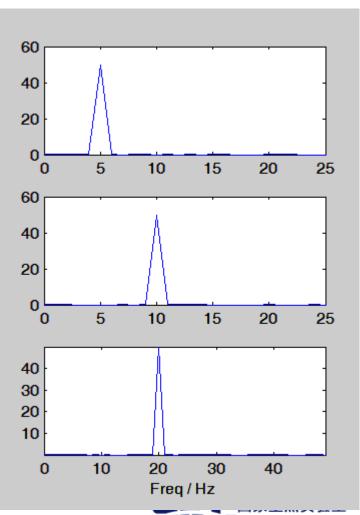


$$s_1 = \sin(10\pi t)$$

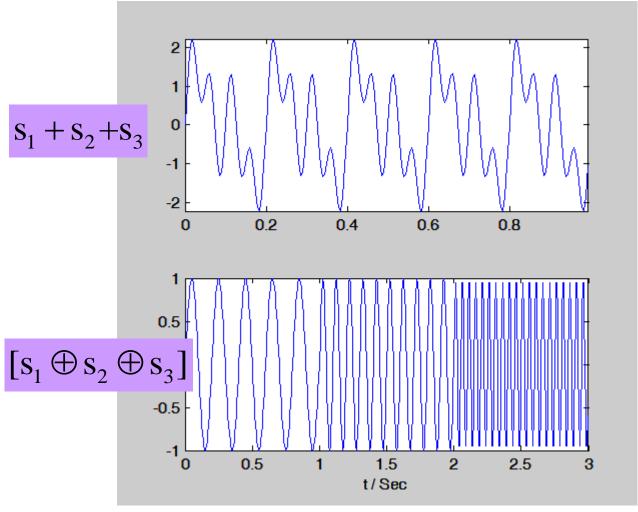
$$s_2 = \sin(20\pi t)$$

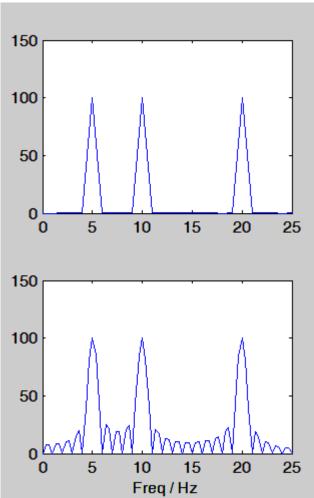
$$s_3 = \sin(40\pi t)$$







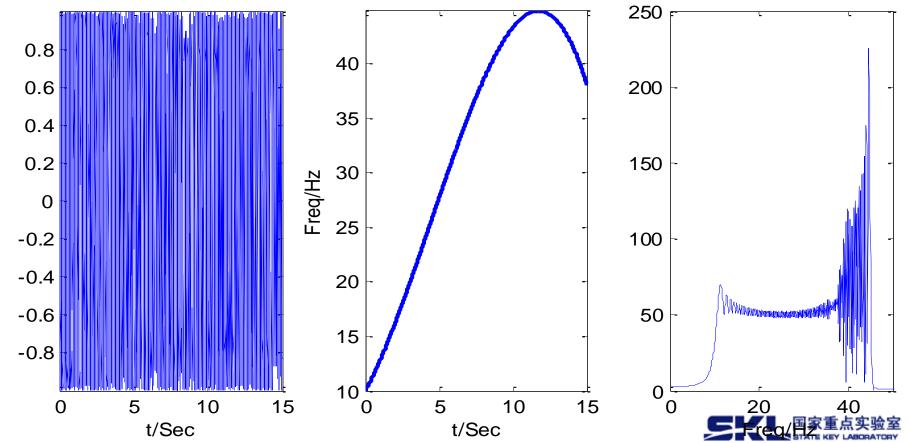




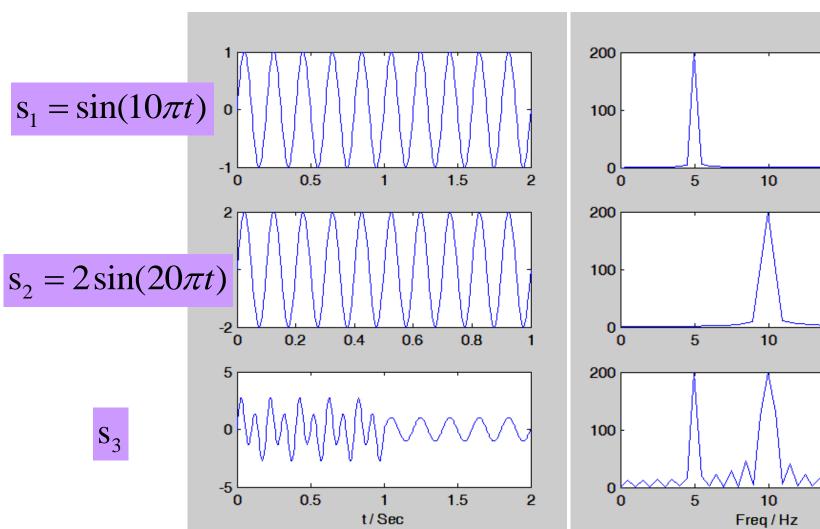




$$s(t) = \sin\left(2\pi\left(10t + \frac{5}{4}t^2 + \frac{1}{9}t^3 - \frac{1}{120}t^4\right)\right) \frac{IF(t) = 10 + 2.5t + t^2/3 - t^3/30 \text{ (Hz)}$$









- 1) 准确反映信号所含频率分量,不能反映频率分量存在的时间段
- 2) 能反映调频信号的频率范围,不能反映频率 随时间的变化规律
- 3) 能准确反映信号中某频率分量的总能量,却不能显示频率分量强度随时间的变化规律





● 傅立叶变换缺点 全局变换,丢失局部信息

$$f(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$f(t) = \sum_{n = -\infty}^{\infty} c_n e^{jn\omega_0 t} \qquad c_n = \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jn\omega_0 t} dt \qquad \omega_0 = 2\pi/T$$

$$c_{n} = \frac{1}{T} \int_{-T/2}^{T/2} f(t)e^{-jn\omega_{0}t} dt = \frac{1}{T} \sum_{k=0}^{n-1} \int_{t_{k}}^{t_{k+1}} f(t)e^{-jn\omega_{0}t} dt;$$

$$t_0 = -T / 2; t_n = T / 2$$

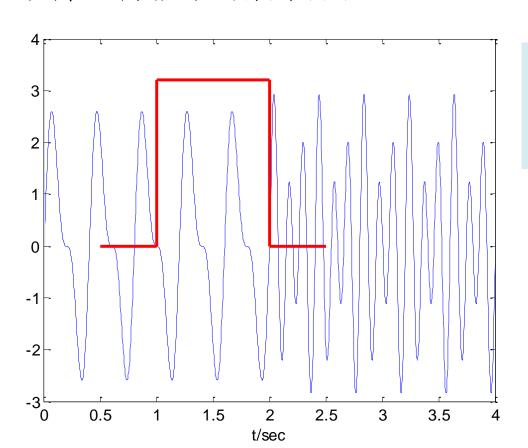
可看出,只要信号f(t)任一段 $[t_k,t_{k+1}]$ 内包含有频 率为 $n\omega_0$ 的分量,则 c_n 非零。





● 基本思想:加窗Fourier变换

通过加窗处理,取出信号片段进行Fourier变换,以观察该片段内信号的频率分量



矩形窗

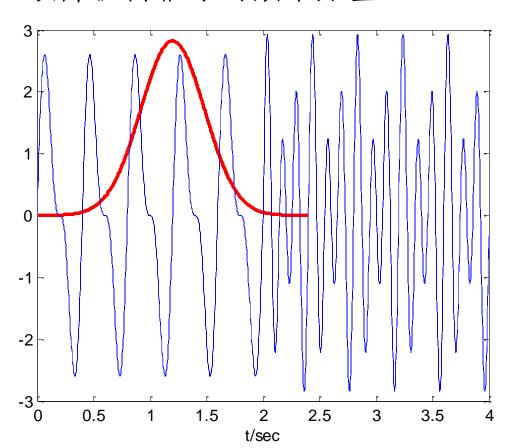
窗口中的每一点在计算中的贡献是等同的





● 基本思想:加窗Fourier变换

通过加窗处理,取出信号片段进行Fourier变换,以观察该片段内信号的频率分量



钟形窗

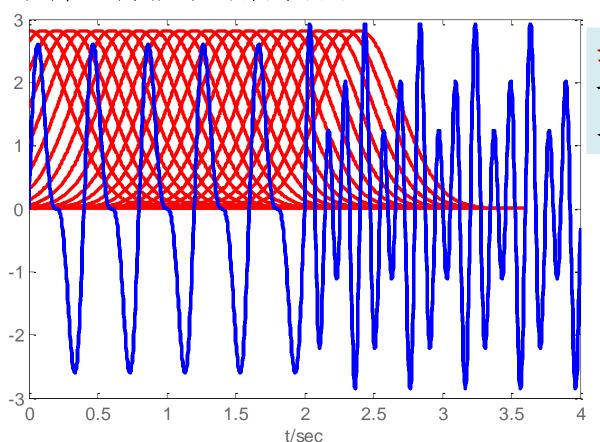
突出窗口中的中间点在计算中的贡献





● 基本思想:加窗Fourier变换

通过加窗处理,取出信号片段进行Fourier变换,以观察该片段内信号的频率分量

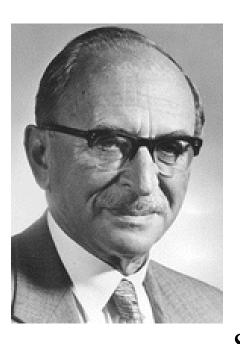


逐点移动钟形窗 计算各个点附近片 段的Fourier变换

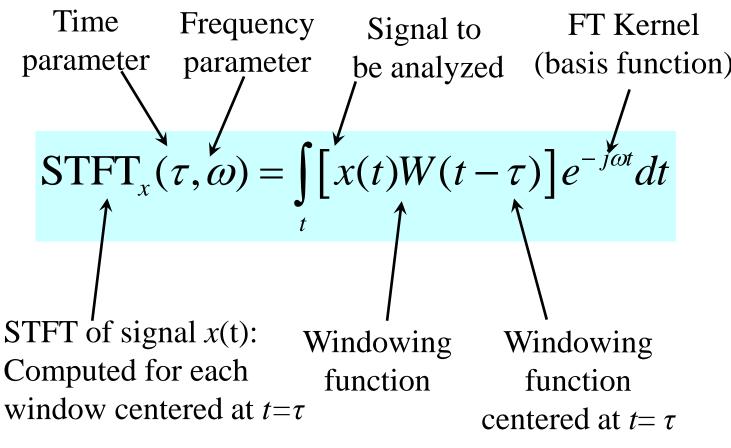




● STFT定义



Gabor (1946) 全息照相技 术1971获诺 贝尔奖







STFT性质

线性可加
$$a_1 f_1(t) + a_2 f_2(t) \leftarrow STFT \rightarrow a_1 STFT_{x1} + a_2 STFT_{x2}$$

时移性质
$$f(t-t_0) \longleftrightarrow STFT_x(\tau-t_0,\omega)$$

频移性质
$$f(t)e^{j\omega_0} \leftarrow STFT_x(\tau,\omega-\omega_0)$$

众多Fourier变换满足的性质





STFT性质

特别是,当W为高斯窗时 $W(t) = \frac{1}{2\sqrt{\pi a}} e^{-t^2/4a}$

$$W(t) = \frac{1}{2\sqrt{\pi a}} e^{-t^2/4a}$$

$$\int STFT_{x}(\tau,\omega)d\tau = \int \int_{t} [x(t)W(t-\tau)]e^{-j\omega t}dtd\tau$$
$$= \widehat{x}(\omega)$$

$$\int \int |STFT_{x}(\tau,\omega)|^{2} d\tau d\omega = \int |x(t)|^{2} dt$$

$$x(t) = \frac{1}{2\pi} \int \int STFT_{x}(\tau, \omega) W(t - \tau) e^{j\omega t} d\omega d\tau$$





● 离散STFT

$$T \rightarrow$$
 采样间隔 $f=1/s$ s-信号长度/sec

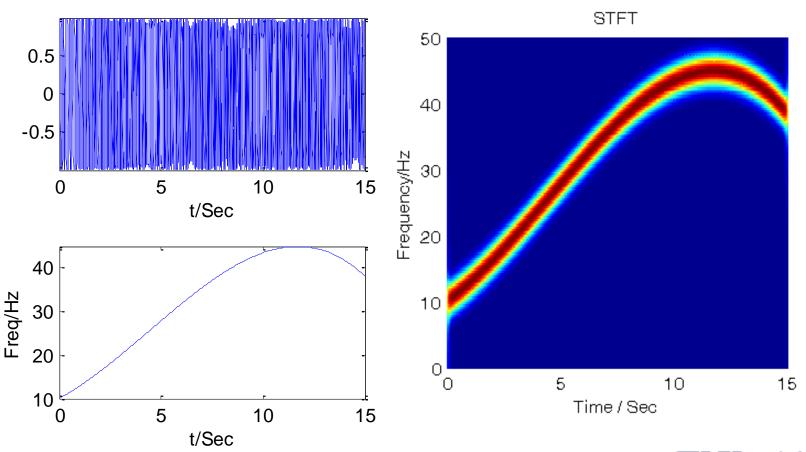
$$STFT_{x}(n,k) = \int x(t)W(t-nT)e^{-j2\pi kft}dt$$

$$x(t) = \sum_{n} \sum_{k} STFT_{x}(n,k)W(t-nT)e^{j2\pi kft}$$





$$s(t) = \sin\left(2\pi\left(10t + \frac{5}{4}t^2 + \frac{1}{9}t^3 - \frac{1}{120}t^4\right)\right)$$

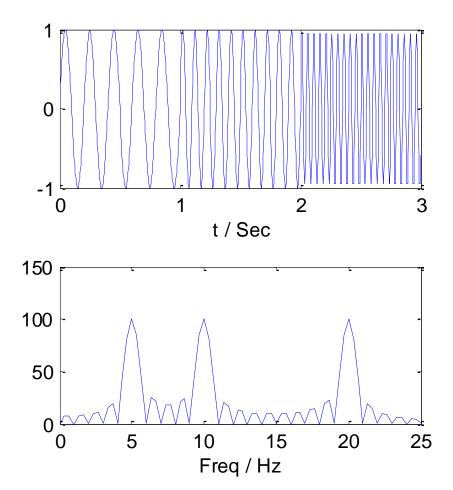


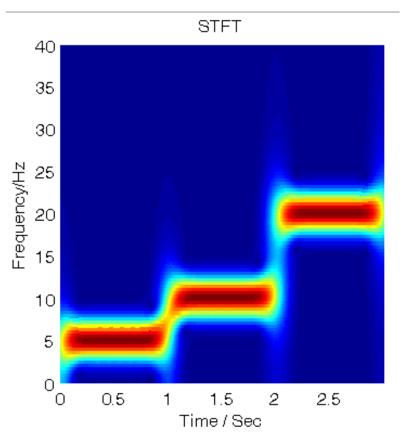




● 示例

 $[s_1 \oplus s_2 \oplus s_3]$



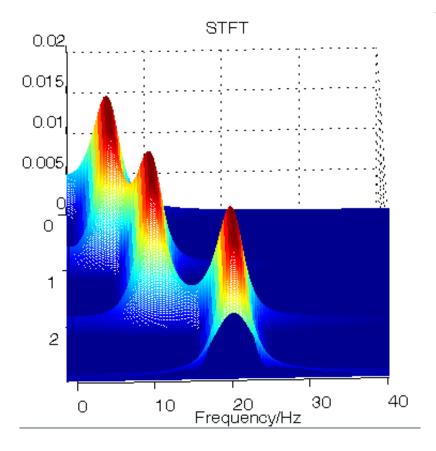


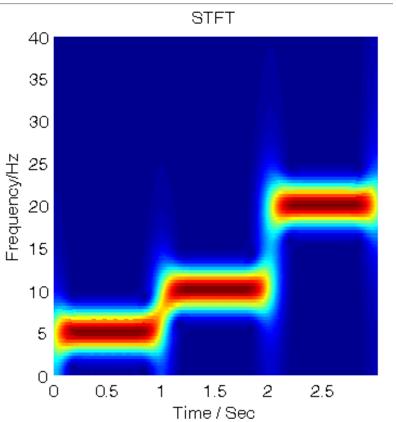




● 示例

$$[s_1 \oplus s_2 \oplus s_3]$$



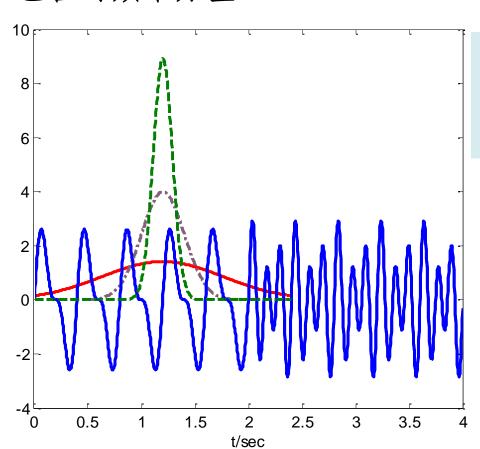






● 问题:窗口宽度的选择

是否可以将窗口取得非常小,以确定信号在每个瞬间所包含的频率分量?



极端情况

- 1. 窗宽为信号长度
- 2. 窗宽为单点($\Delta w = 1/Fs$)





● 问题:窗口宽度的选择

是否可以将窗口取得非常小,以确定信号在每个瞬间所 包含的频率分量?

$$STFT_{x}(\tau,\omega) = \int_{t} [x(t)W(t-\tau)]e^{-j\omega t}dt$$

$$f(t) = x(t)W(t-\tau)$$
 记窗口宽度 $\Delta \mathbf{w}$ $\omega_0 = 2\pi/\Delta \mathbf{w}$

$$\omega_0 = 2\pi/\Delta \mathbf{w}$$

$$STFT_{x}(\tau, n\omega_{0}) = \frac{1}{T} \int_{\tau - \Delta \mathbf{w}/2}^{\tau + \Delta \mathbf{w}/2} f(t) e^{-jn\omega_{0}t} dt$$

频域离散:
$$\{n\omega_0 << Fs/2; n=0,\pm 1,\pm 2,...\}$$

频域分辨率:
$$\Delta \omega = \omega_0 = 2\pi/\Delta \mathbf{w}$$





问题: 窗口宽度的选择

时域分辨率: $\Delta t = \Delta \mathbf{w}$

 $\Delta\omega = \omega_0 = 2\pi/\Delta \mathbf{w}$ 频域分辨率:

$$\Delta\omega\Delta t = 2\pi$$

- 提高时域分辨率,必然使频域分辨率下降
- > 提高频域分辨率,必然使时域分辨率下降

极端情况

1. 窗宽为信号长度

$$\Delta\omega = \frac{2\pi}{s}$$

 $\Delta \omega = \frac{2\pi}{}$ s- 信号长度 / sec

2. 窗宽为单点($\Delta w = 1/Fs$)

$$\Delta\omega = 2\pi Fs$$





● 问题: 窗口宽度的选择

极端情况

1. 窗宽为信号长度

$$\Delta\omega = \frac{2\pi}{s}$$

$$W(\tau) = 1 \rightarrow \text{STFT}_{x}(\tau, \omega) = \hat{x}(\omega)$$

2. 窗宽为单点($\Delta w = 1/Fs$) $\Delta \omega = 2\pi Fs$

$$W(\tau) = \delta(\tau) \rightarrow \text{STFT}_{x}(\tau, \omega) = x(\tau)e^{-j\omega\tau}$$



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Short Time Fourier Transform

Heisenberg测不准原理

如果 $W(t) \in L^2(R)$ and $tW(t) \in L^2(R)$

则W(t)为窗口函数,窗口中心点定义为 (t_0,ω_0)

$$\begin{cases} t_0 = \frac{1}{\|W(t)\|^2} \int_{-\infty}^{\infty} t |W(t)|^2 dt \\ \omega_0 = \frac{1}{\|\widehat{W}(\omega)\|^2} \int_{-\infty}^{\infty} \omega |\widehat{W}(\omega)|^2 d\omega \end{cases}$$

$$\widehat{W}(\omega) = F[W(t)]$$

窗口大小
$$\Delta W = \left(\frac{1}{\|W(t)\|^2} \int_{-\infty}^{\infty} (t - t_0)^2 |W(t)|^2 dt\right)^{1/2}$$

$$\Delta W \Delta \widehat{W} = \left(\frac{1}{\|\widehat{W}(\omega)\|^2} \int_{-\infty}^{\infty} (\omega - \omega_0)^2 |\widehat{W}(\omega)|^2 d\omega\right)^{1/2}$$





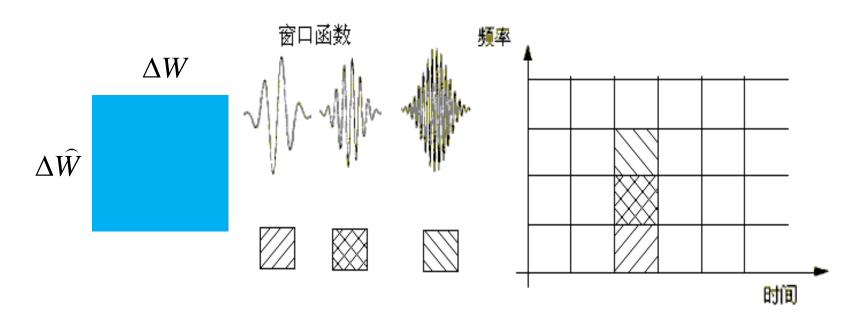


● Heisenberg测不准原理

> 窗口面积必须大于一个常数

$$\Delta W \Delta \widehat{W} \ge \frac{1}{2}$$

 \triangleright 窗口大小不受中心点位置影响 $(\Delta W, \Delta \hat{W})$ (t_0, ω_0)

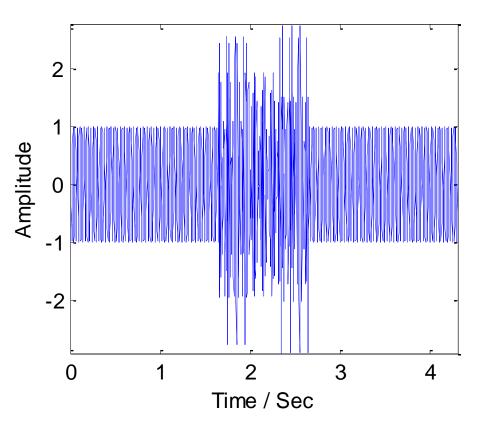




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Short Time Fourier Transform



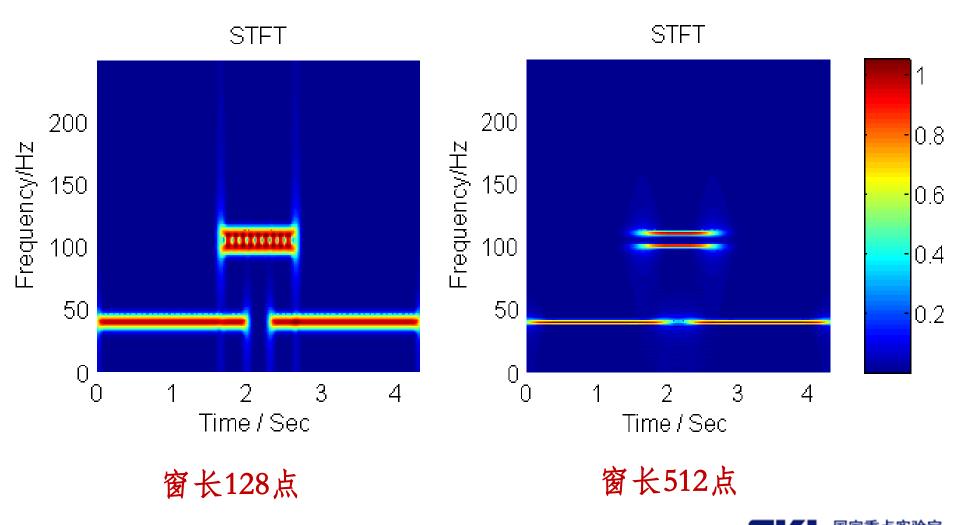


The three components are: 1) 40Hz over the whole time span apart from a narrow disconnection of 0.3 second in the middle, 2) 100Hz occupying the middle part with a 1 second duration, and 3) 110Hz occupying the middle part with a 1 second duration. The sampling frequency is 500Hz.

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Short Time Fourier Transform







STFT的滤波器观点解释

$$STFT_{x}(\tau,\omega) = \int_{t} [x(t)W(t-\tau)]e^{-j\omega t}dt$$

$$f(t,\tau) = x(t)W(t-\tau)$$



$$STFT_{x}(\tau,\omega) = \int_{t} f(t,\tau)e^{-j\omega t}dt$$



$$\mathbf{STFT}_{x}(\tau,\omega) = \widehat{x}(\omega) \otimes \left[\widehat{W}(\omega)e^{-j\omega\tau}\right]$$



$$STFT_{x}(\tau,\omega) = e^{-j\omega\tau} \int \widehat{x}(\bar{\omega}) \widehat{W}(\omega - \bar{\omega}) e^{j\bar{\omega}\tau} d\bar{\omega}$$





● STFT的滤波器观点解释

$$STFT_{x}(\tau,\omega) = e^{-j\omega\tau} \int \widehat{x}(\bar{\omega}) \widehat{W}(\omega - \bar{\omega}) e^{j\bar{\omega}\tau} d\bar{\omega}$$

$$x(t)$$
 $\xrightarrow{\mathcal{F}} \widehat{x}(\overline{\omega})$ 带通滤波 \widehat{W} $\widehat{$

Fourier 反变换

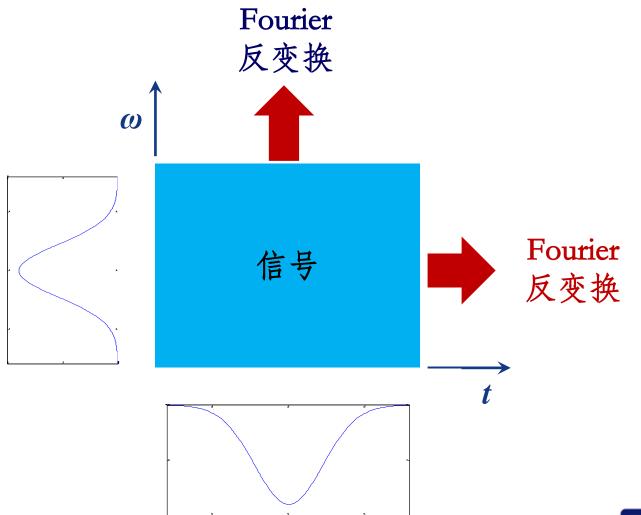
$$\mathbf{STFT}_{x}(\tau,\omega) \leftarrow \otimes \widehat{f}(\tau,\omega)$$

$$\uparrow e^{-j\omega\tau}$$





● STFT的滤波器观点解释

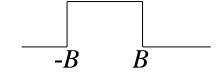




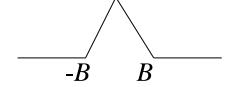


● 窗类型的影响

(1) Rectangle



(2) Triangle



(3) Hanning

$$W(t) = \begin{cases} 0.5 + 0.5\cos(\pi t / B) & \text{when } |t| \le B \\ 0 & \text{otherwise} \end{cases}$$

(4) Hamming

$$W(t) = \begin{cases} 0.54 + 0.46\cos(\pi t / B) & \text{when } |t| \le B \\ 0 & \text{otherwise} \end{cases}$$

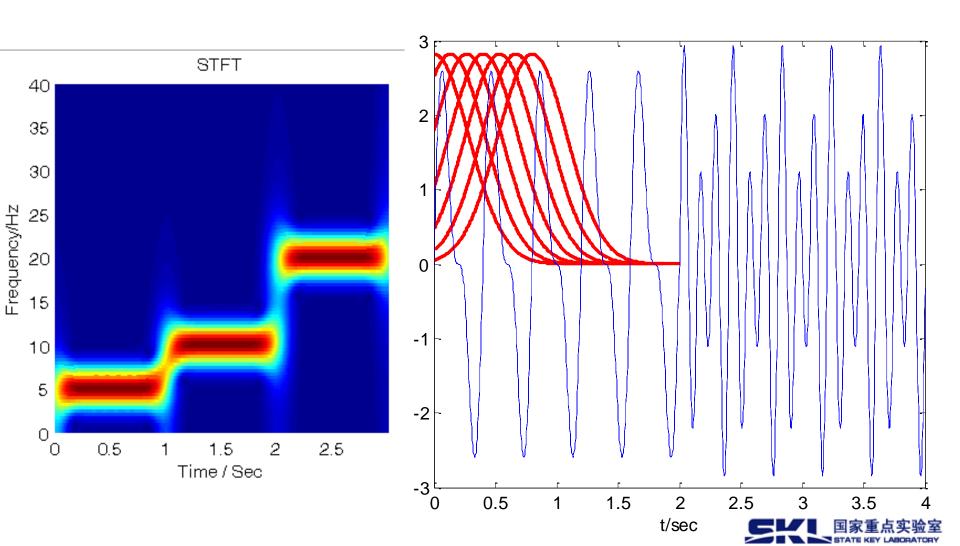
(5) Gaussian

$$W(t) = \frac{1}{2\sqrt{\pi a}} e^{-t^2/4a}$$



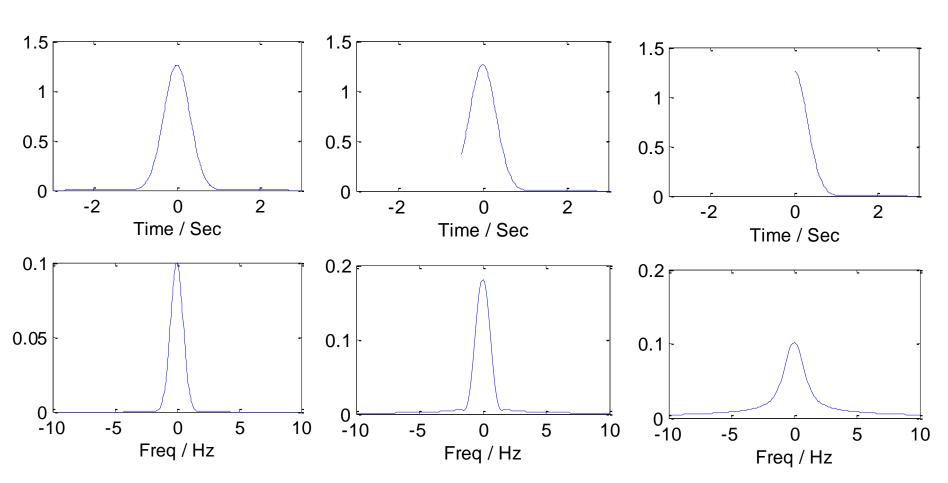


● STFT的边界扭曲





® STFT的边界扭曲



不完整的窗不再严格满足带通滤波器性质





● STFT的边界扭曲

改进方法

> 边界延拓

镜像拓展 补零 边界预测

• • •

有效?





● 讨论

- ➤ Fourier变换的适用性
- > STFT的适用性
- > 平稳信号
- > 非平稳信号





谢谢聆听欢迎交流

