## Application of Z-transforms to Difference Equations

LCCDE: 
$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k]$$

Taking Z-transform of LCCDE:

$$\Rightarrow Y(z) \sum_{n=0}^{N} a_n z^{-n} = X(z) \sum_{n=0}^{M} b_n z^{-n}$$

$$\Rightarrow \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^{M} b_k z^{-k}}{\sum_{k=0}^{N} a_k z^{-k}} = H(z), \text{ the system's transfer function}$$

Note: The Z transform of the system's output is the product of the Z transform of the system's input and the system's transfer function.

$$Y(z) = H(z) X(z)$$
we also know:

The system's transfer function, H(Z), is the Z transform of the system's impulse response.

The impulse response of an LTI system characterized by an LCCDE has a rational Z-transform.

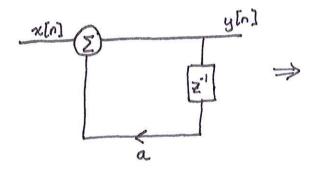
## Example

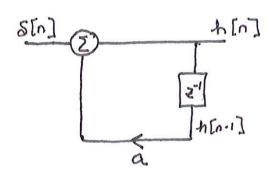
Let h[n] = a u[n].

Find a system which models hin]. That is, find a system which, when excited by a unit impulse, will generate the response hin].

Solution We Know that H(z) = 1-az 1 | 21 > |a|

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1-az^{-1}}$$





In general, the Z-transform of a causal signal (with rational Z-transform) can be viewed as a model for the signal; it is the transfer function of an LTI system whose impulse response is the signal.

A rational z-transform has the form of a ratio of polynomials in z or z!

$$H(z) = \frac{B(z)}{A(z)} = \frac{b_0 + b_1 z^{-1} + \dots + b_m z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}}$$

$$= \frac{b_0 z^{-M} \left(z^M + \frac{b_1}{b_0} z^{M-1} + \dots + \frac{b_m}{b_0}\right)}{a_0 z^{-N} \left(z^N + \frac{a_1}{a_0} z^{N-1} + \dots + \frac{a_N}{a_0}\right)}$$

$$= \frac{b_0}{a_0} z^{N-M} \frac{\left(z - z_1\right) \left(z - z_2\right) \dots \left(z - z_m\right)}{\left(z - \rho_1\right) \left(z - \rho_2\right) \dots \left(z - \rho_N\right)}$$

Assuming bo #0:

H(z) has M zeroes at  $z=z_1,...,z_M$ H(z) has N poles at  $z=\rho_1,...,\rho_N$ 

In addition to the poles and zeros mentionned above:

if N>M, X(2) has N-M zeroes at Z=0 if M>N, X(2) has M-N poles at Z=0

Let 
$$h[n] = a^n u[n]$$
.  
Sketch the pole-zero plot and ROC for  $H(z)$ .

Solution
$$H(z) = \frac{1}{1-az^{-1}} = \frac{z}{z-a}, |z| > |a|$$

$$Pac = \frac{1}{1-az^{-1}} = \frac{z}{z-a}, |z| > |a|$$

$$P_1 = a$$

For causal signals, the growing /decaying nature of the signal's envelope will depend on whether the poles are contained within the unit eircle, on the unit circle, or outside the unit circle.

## Referring to the example above

If 
$$|a|=1$$
, the pole is on the unit circle and the envelope of h[n] remains constant as  $n \to \infty$ 

Ex.  $h[n] = u[n] \Rightarrow h[n] = \{\frac{1}{n} \mid 1 \mid 1 \mid 1 \dots \}$ 

If 
$$|a| < 1$$
, the pole is inside the unit circle and  $n[n]$  decays to zero as  $n \to \infty$ 

$$Ex. h[n] = (\frac{1}{2})^n u[n] \Rightarrow h[n] = \{\frac{1}{4}, \frac{1}{8}, \dots\}$$

Recall also that a causal LTI system is stable iff all of its characteristic roots are inside the unit circle.

We now know that the poles of a system are the same as the system's characteristics roots, and thus a causal system is BIBO stable if and only if all of its poles are inside the unit circle.

The ROC for the Transfer Function of a stable causal system will always contain the unit circle.