

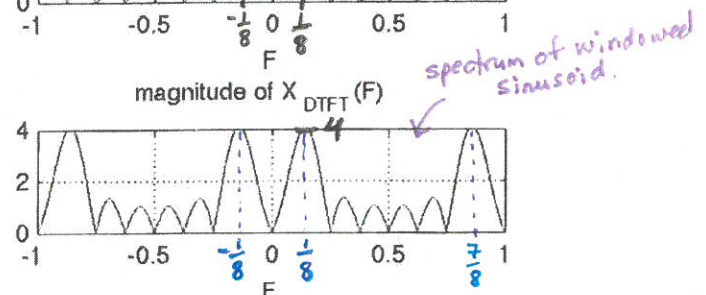
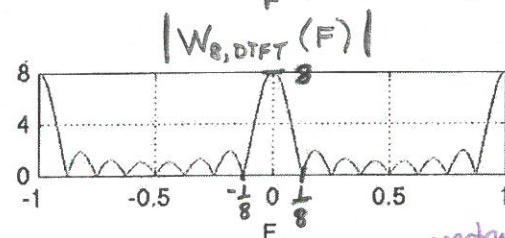
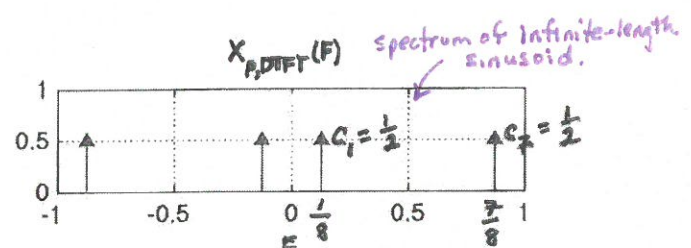
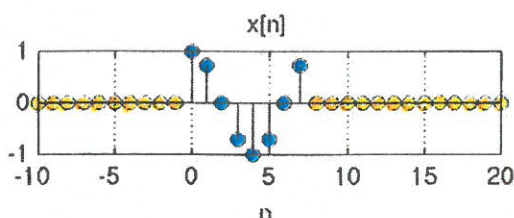
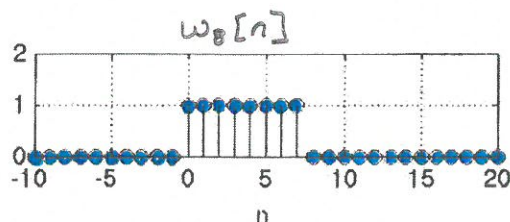
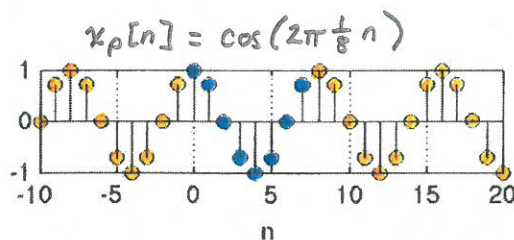
It is particularly instructive to compare the DTFT of an infinitely long sinusoid to the DTFT of a windowed version of the sinusoid.

Example

$$\text{Let } x_p[n] = \cos(2\pi \tfrac{1}{8} n) \longleftrightarrow X_{p, \text{DTFT}}(F) = \left(\tfrac{1}{2} \delta(F + \tfrac{1}{8}) + \tfrac{1}{2} \delta(F - \tfrac{1}{8}) \right) * \text{comb}(F)$$

$$\text{let } x[n] = x_p[n] w_8[n] \quad \text{where } w_8[n] = \begin{cases} 1, & 0 \leq n \leq 7 \\ 0, & \text{otherwise} \end{cases}$$

$$\begin{aligned} \text{Then } X_{\text{DTFT}}(F) &= [X_{p, \text{DTFT}}(F) \text{rect}(F)] * W_{8, \text{DTFT}}(F) \\ &= \left(\tfrac{1}{2} \delta(F + \tfrac{1}{8}) + \tfrac{1}{2} \delta(F - \tfrac{1}{8}) \right) * W_{8, \text{DTFT}}(F) \\ &= \tfrac{1}{2} W_{8, \text{DTFT}}(F + \tfrac{1}{8}) + \tfrac{1}{2} W_{8, \text{DTFT}}(F - \tfrac{1}{8}) \end{aligned}$$

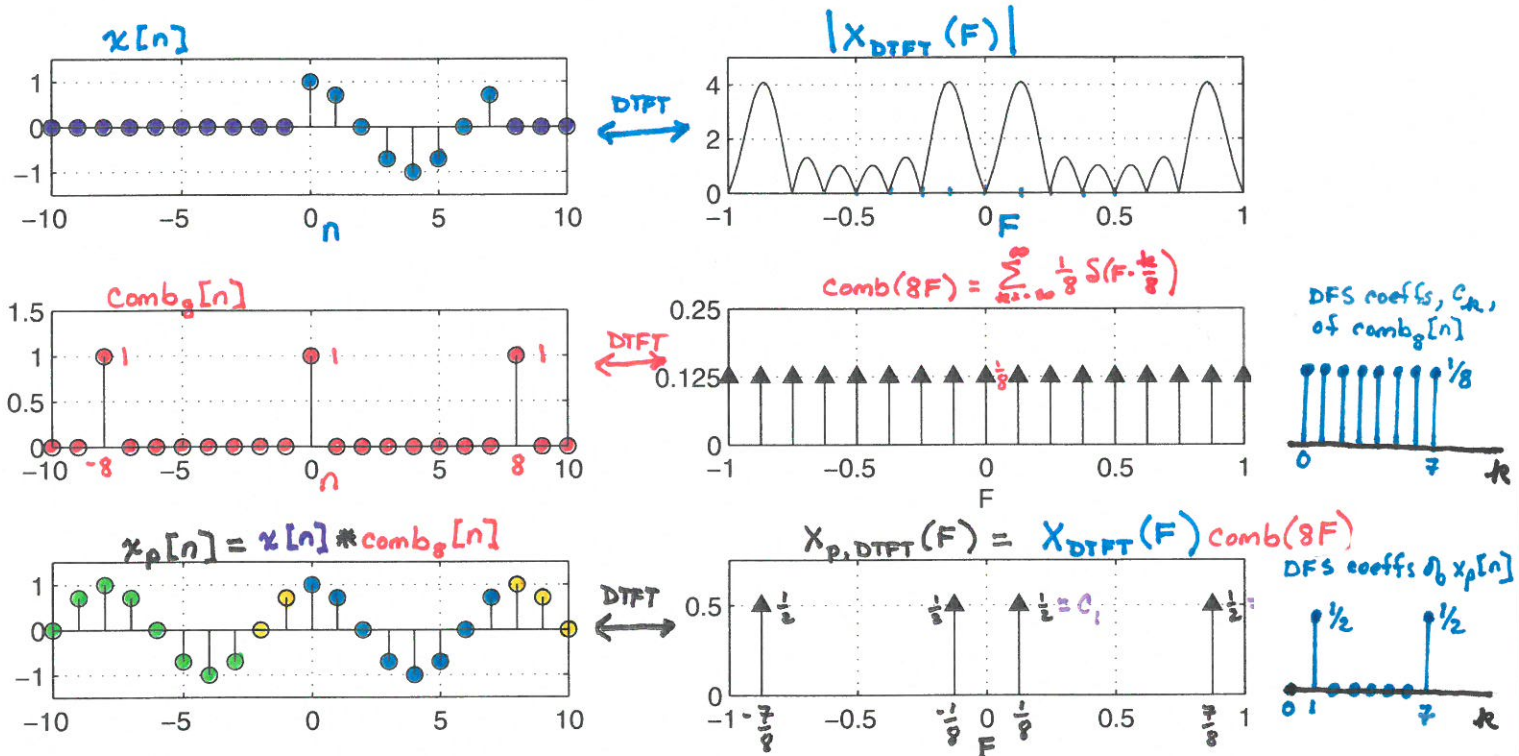


Note the spectral smearing that took place as a result of truncating/windowing the time-domain signal. The result is a loss of frequency resolution. The longer the time window, the better will be the frequency resolution.

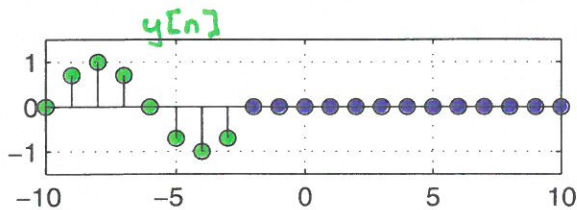
Since $x_p[n] = x[n] * \text{comb}_8[n]$ ← see illustrations in left-side column below,
← see illustrations in right-side column below,

We should find: $X_{p, \text{DTFT}}(F) = X_{\text{DTFT}}(F) \text{comb}(8F) = X_{\text{DTFT}}(F) \sum_{k=-\infty}^{\infty} \frac{1}{8} \delta(F - \frac{k}{8})$

$$\Rightarrow X_{p, \text{DTFT}}(F) = \sum_{k=-\infty}^{\infty} \frac{1}{8} X_{\text{DTFT}}(\frac{k}{8}) \delta(F - \frac{k}{8})$$



Note that $y[n]$, shown below, has the same 8-periodic extension as $x[n]$.



Does $Y_{\text{DTFT}}(F) = X_{\text{DTFT}}(F)$?

Does $Y_{\text{DTFT}}(\frac{k}{8}) = X_{\text{DTFT}}(\frac{k}{8})$?

Since: $y[n] * \text{comb}_8[n] = x[n] * \text{comb}_8[n]$

we know that: $Y_{\text{DTFT}}(F) \text{comb}(8F) = X_{\text{DTFT}}(F) \text{comb}(8F)$

$$\Rightarrow \sum_{k=-\infty}^{\infty} \underbrace{\frac{1}{8} Y_{\text{DTFT}}(\frac{k}{8})}_{C_k^{(y)}} \delta(F - \frac{k}{8}) = \sum_{k=-\infty}^{\infty} \underbrace{\frac{1}{8} X_{\text{DTFT}}(\frac{k}{8})}_{C_k^{(x)}} \delta(F - \frac{k}{8})$$

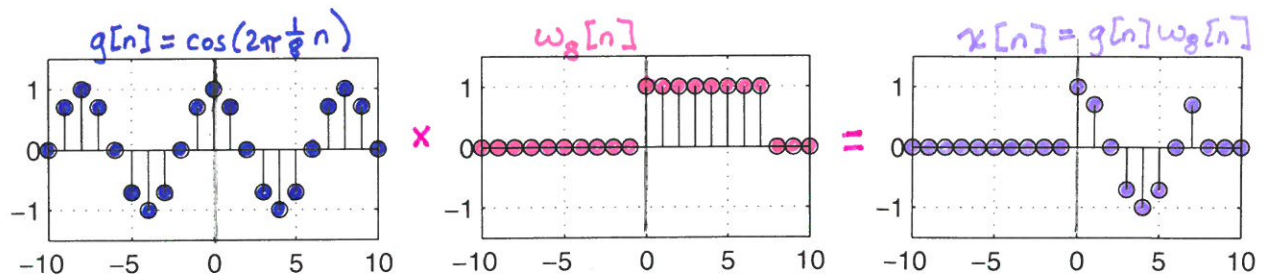
$$\Rightarrow Y_{\text{DTFT}}(\frac{k}{8}) = X_{\text{DTFT}}(\frac{k}{8})$$



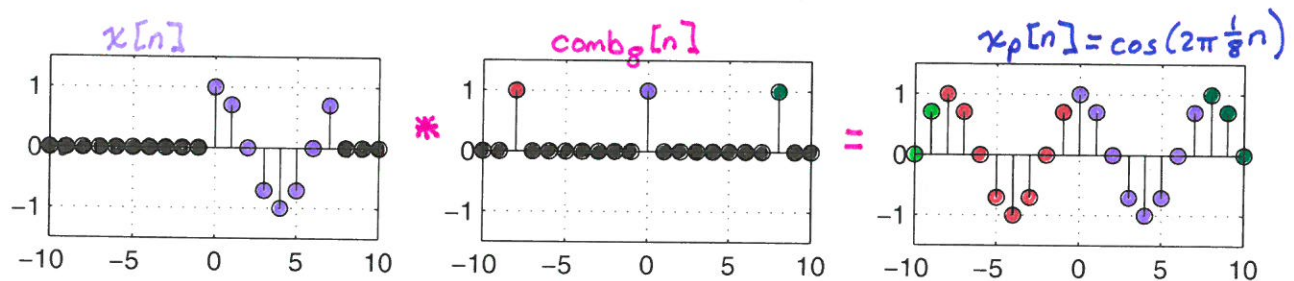
Important finding

If two signals, $x[n]$ and $y[n]$, have the same N -periodic extension, their DTFTs will agree in value at $F = k/N$, $k = 0, \pm 1, \pm 2, \dots$

Let $x[n]$ denote an energy signal whose nonzero values are contained within the interval $0 \leq n \leq N-1$. For example consider $x[n] = g[n]w_g[n]$ as shown below.



Let $x_p[n]$ denote the N -periodic extension of $x[n]$: $x_p[n] = x[n] * \text{comb}_N[n]$



Then the DTFT of $x[n]$ is:

$$X_{\text{DTFT}}(F) = \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi F n} = \sum_{n=0}^{N-1} x[n] e^{-j2\pi F n}$$

since $x[n] = 0$ for $n < 0$ and for $n \geq N$

And the DFS coefficients of $x_p[n]$ are given by:

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x_p[n] e^{-j2\pi \frac{k}{N} n} = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{k}{N} n} = \frac{1}{N} X_{\text{DTFT}}\left(\frac{k}{N}\right)$$

since $x_p[n] = x[n]$ for $0 \leq n \leq N-1$

The N -point DFT of a signal $x[n]$ is defined as:

$$X_{\text{DFT},N}[k] = \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{k}{N} n} = N c_k, \quad k=0, \dots, N-1$$

The N -point IDFT is defined as:

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} \underbrace{X_{\text{DFT},N}[k]}_{N c_k} e^{j2\pi \frac{k}{N} n} = x_p[n], \quad 0 \leq n \leq N-1$$

$$\text{Recall: } x_p[n] = \sum_{k=0}^{N-1} c_k e^{j2\pi \frac{k}{N} n}$$

Assuming that $x[n]$ is equal to zero outside of the interval $0 \leq n \leq N-1$, we may claim that:

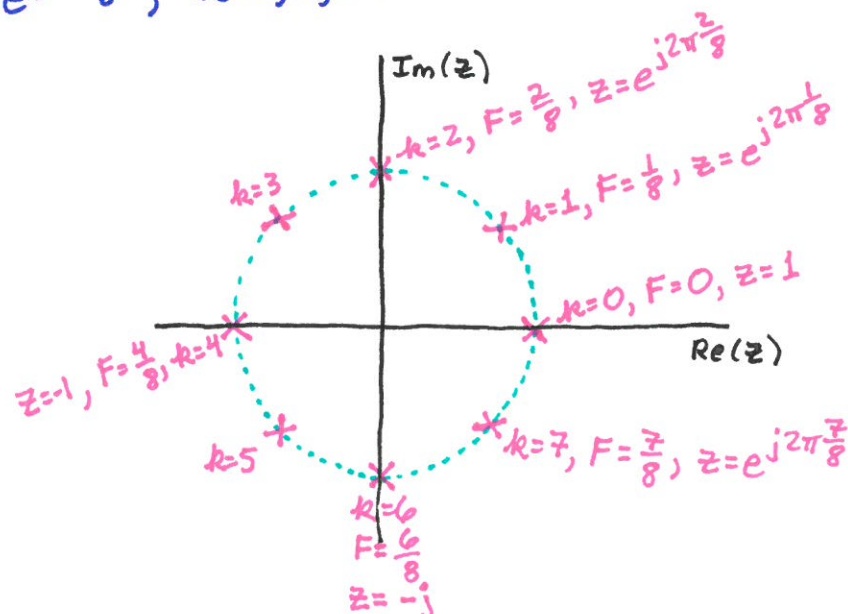
$$X_z(z) = \sum_{n=0}^{N-1} x[n] z^{-n}$$

$$X_{\text{DFT}}(F) = \sum_{n=0}^{N-1} x[n] e^{-j2\pi F n} = \sum_{n=0}^{N-1} x[n] (e^{j2\pi F})^{-n} = X_z(e^{j2\pi F})$$

$$X_{\text{DFT},N}[k] = \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{k}{N} n} = \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{k}{N} n} = X_{\text{DFT}}\left(\frac{k}{N}\right)$$

Thus:
$$X_{\text{DFT},N}[k] = X_{\text{DFT}}\left(\frac{k}{N}\right) = X_z\left(e^{j2\pi \frac{k}{N} n}\right)$$

Example: Assuming $x[n] = 0$ for $n < 0$ and $n \geq 8$, the 8-point DFT of $x[n]$ consists of a set of $N=8$ values (indexed by $k=0, 1, \dots, 7$) which may be viewed as samples of $X_{\text{DFT}}(F)$ at $F = \frac{k}{8}$, $k=0, \dots, 7$ or as samples of $X_z(z)$ at $z = e^{j2\pi \frac{k}{8}}$, $k=0, \dots, 7$.



Example : Find the 4-point DFT of the sequence

$$x[n] = \{ 0 \underset{\uparrow}{1} \ 1 \ 1 \ 0 \}$$

Solution

$$X_z(z) = 1 + z^{-1} + z^{-2}$$

$$X_{\text{DFT}}(F) = X_z(e^{j2\pi F}) = 1 + e^{-j2\pi F} + e^{-j2\pi 2F}$$

$$\begin{aligned} X_{\text{DFT},4}[k] &= X_{\text{DFT}}\left(\frac{k}{4}\right) = 1 + e^{-j2\pi \frac{k}{4}} + e^{-j2\pi 2\frac{k}{4}} \\ &= 1 + e^{-j\frac{\pi}{2}k} + e^{-j\pi k} \end{aligned}$$

k	1	$e^{-j\frac{\pi}{2}k}$	$e^{-j\pi k}$	$X_{\text{DFT},4}[k]$
0	1	1	1	3
1	1	-j	-1	-j
2	1	-1	1	1
3	1	j	-1	j

In matlab, the 4-point DFT of $x[n]$ may be computed as:

```
> x = [1 1 1];
```

```
> X = fft(x, 4)
```

```
X = [3 -j 1 +j]
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