

#### EEN452 - Control and Operation of Electric Power Systems

Part 6: Power system stability fundamentals

https://sps.cut.ac.cy/courses/een452/

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#### Today's learning objectives



After this part of the lecture and additional reading, you should be able to ...

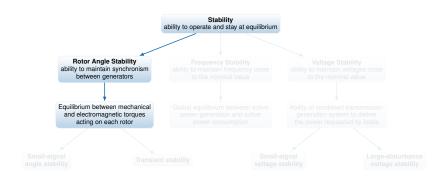
- 1 ... understand the basic classifications of power system stability;
- 2 ... be able to identify and perform stability analysis problems; and,
- ...propose methods for stabilizing power systems.

#### 1 Outline



- 1 Introduction
- 2 Voltage Stability
- **3 Rotor Angle Stability**
- 4 References

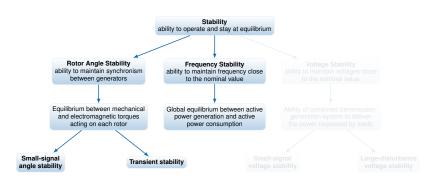




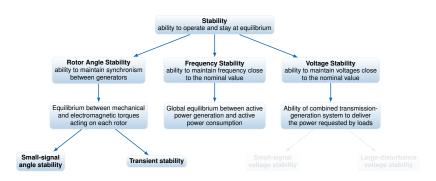




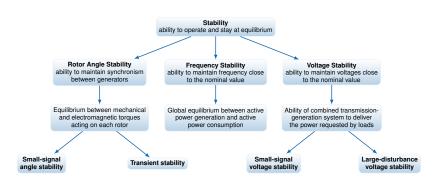








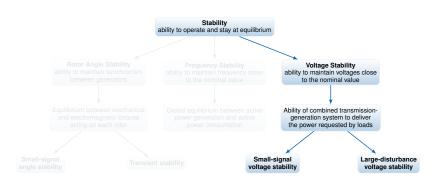










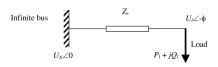


#### 2 Outline



- 1 Introduction
- 2 Voltage Stability
- 3 Rotor Angle Stability
- 4 References





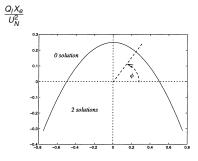
$$P_{l} = \frac{U_{l}U_{N}}{X_{e}}\sin\phi$$
  $Q_{l} = \frac{U_{l}U_{N}\cos\phi - U_{l}^{2}}{X_{e}}$ 

$$P_{I}^{2} + \left(Q_{I} + \frac{U_{I}^{2}}{X_{e}}\right)^{2} = \left(\frac{U_{I}U_{N}}{X_{e}}\right)^{2} \Rightarrow \left(U_{I}^{2}\right)^{2} + \left(2Q_{I}X_{e} - U_{N}^{2}\right)U_{I}^{2} + X_{e}^{2}(P_{I}^{2} + Q_{I}^{2}) = 0$$
(2.1)



To have (at least) one solution:

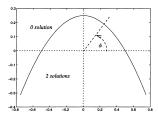
$$\left(2Q_{l}X_{e}-U_{N}^{2}\right)^{2}-4X_{e}^{2}\left(P_{l}^{2}+Q_{l}^{2}\right)\geq0\Rightarrow-\left(\frac{P_{l}X_{e}}{U_{N}}\right)^{2}-\frac{Q_{l}X_{e}}{U_{N}^{2}}+0.25\geq0$$



$$\frac{P_I X_e}{U_N^2}$$







 $\frac{P_I X_e}{U_N^2}$ 

- any P<sub>i</sub> can be reached provided Q<sub>i</sub> is adjusted (but U<sub>i</sub> may be unacceptable)
- dissymmetry between P<sub>I</sub> and Q<sub>I</sub> due to reactive transmission impedance
- locus symmetric w.r.t. Q<sub>i</sub> axis; this does no longer hold when transmission resistance is included



Under a constant load power factor  $\cos \phi$  (i.e.,  $Q_l = P_l \tan \phi$ ), we get from Eq. (2.1):

$$P_I^2+rac{U_N^2}{X_e} an\phi P_I-rac{U_N^4}{4X_e^2}=0$$

Then, we get the maximum power:

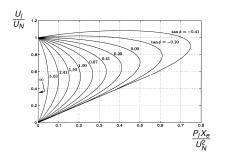
$$P_{l}^{max} = \frac{\cos\phi}{1+\sin\phi} \frac{U_{N}^{2}}{2X_{e}} \qquad Q_{l}^{max} = \frac{\sin\phi}{1+\sin\phi} \frac{U_{N}^{2}}{2X_{e}} \qquad U_{l}^{max} = \frac{U_{N}}{\sqrt{2}\sqrt{1+\sin\phi}}$$

Or, for the extreme cases:

$$\cos \phi = 1: \qquad \qquad P_l^{max} = \frac{U_N^2}{2X_e} \qquad Q_l^{max} = 0 \qquad U_l^{max} = \frac{U_N}{\sqrt{2}}$$

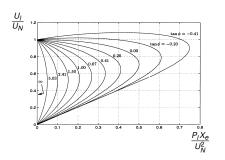
$$\cos \phi = 0: \qquad \qquad P_l^{max} = 0 \qquad Q_l^{max} = \frac{U_N^2}{4X_e} \qquad U_l^{max} = \frac{U_N}{2}$$





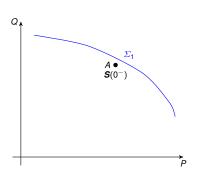
- for given power (P<sub>I</sub>)
  - 1 solution with "high" voltage and "low" current (normal operating point)
  - 1 solution with "low" voltage and "high" current
- compensating the load increases the maximum power but the "critical" voltage approaches normal values
- curves that provide similar information:
  - Q V or S V under constant  $\tan \phi$ , Q V under constant P, etc.





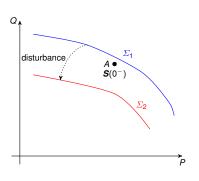
- in real systems, much more complicated
  - no infinite bus, voltage control by generators (AVR)
  - multiple loads and generators
  - complex, meshed transmission system with resistive components as well
  - voltage sensitive loads and restorative behavior
  - etc.





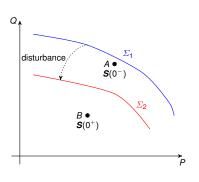
- Pre-fault loadability curve of the system  $(\Sigma_1)$
- Fault occurs in the system
  - 1 Loadability curve is shrunk to  $\Sigma_2$
  - ② Post-fault consumption is decreased due to depressed voltages (voltage sensitive loads). If point B is outside  $\Sigma_2$  then there is no solution and we have short-term voltage instability
- Loads try to restore consumption to pre-fault point A now outside the loadability curve
- Long-term instability leading to voltage collapse





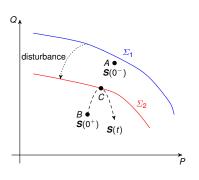
- Pre-fault loadability curve of the system  $(\Sigma_1)$
- Fault occurs in the system
  - f 1 Loadability curve is shrunk to  $m \Sigma_2$
  - 2 Post-fault consumption is decreased due to depressed voltages (voltage sensitive loads). If point B is outside Σ<sub>2</sub> then there is no solution and we have short-term voltage instability
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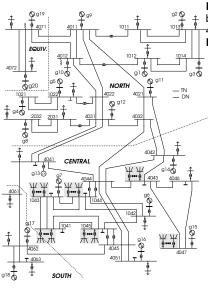




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# 2 Voltage instability: example (RAMSES with Nordic system)

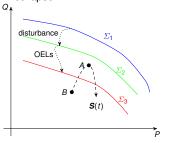




**Disturbance:** 5-cycle short-circuit near bus 4032, cleared by opening line 4032-4042

#### Result:

- the loadability curve is shrunk but the pre disturbance point A is still within the feasible region
- a series of generator
   OverExcitation Limiter (OEL)
   actions further shrink the feasible
   region leading to a system
   collapse



#### 2 Voltage instability countermeasures



- Series compensation: very effective but expensive
- Shunt compensation: cheapest mechanism
- SVC and STATCOM devices
- Adjustment of generator active power productions
- Adjustment of generator voltages
- Block load restoration (e.g., through load tap changers): effective but sometimes too slow
- Undervoltage load shedding : effective but expensive, last resort





#### 3 Outline



- 1 Introduction
- 2 Voltage Stability
- 3 Rotor Angle Stability
  - Transient stability
  - Small-disturbance angle stability
  - Summary
- 4 References

#### 3 (Rotor) angle stability



- most of the electrical energy today is generated by synchronous machines
- in normal system operation:
  - all synchronous machines rotate at the same electrical speed  $\omega_0 = 2\pi f_n$
  - the mechanical and electromagnetic torques acting on the rotating masses of each generator balance each other

$$\dot{\omega}_i = \frac{\omega_0}{2H_i}(T_{mi} - T_{ei})$$

- the phase angle differences between the internal e.m.f.'s of the various machines are constant (synchronism)
- following a disturbance, there is an imbalance between the two torques and the rotor speed varies
- rotor angle stability deals with the ability to keep/regain synchronism after being subject to a disturbance

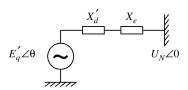
#### 3.1 Transient (angle) stability



- Transient (angle) stability deals with the ability of the system to keep synchronism after being subject to a large disturbance
- typical "large" disturbances:
  - short-circuit cleared by opening of circuit breakers
  - more complex sequences: backup protections, line autoreclosing, etc.
- the nonlinear behavior of the generator and its controllers must be taken into account
  - numerical integration of the differential-algebraic equations is used
- unacceptable consequences of transient instability:
  - generators tripped due to loss of synchronism (to avoid equipment damages)
  - long-lasting voltage dips created by large angle swings (disturb customers)

#### 3.1 Transient (angle) stability

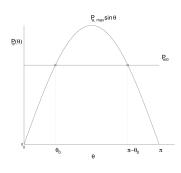




$$\frac{2H}{\omega_0} \frac{d^2 \theta}{dt^2} = P_{m0} - P_{\theta, max} \sin \theta \Rightarrow$$

$$M\ddot{\theta} = P_{m0} - P(\theta)$$
 (3.1)

where in steady-state  $P_{m0}=P_{e,max}\sin\theta_0$ 



## 3.1 Transient (angle) stability: equal area criterion



Multiplying both sides of Eq. (3.1) with  $\dot{\theta}$  and integrating:

$$\frac{1}{2}M\dot{\theta}^2 - \int_0^t (P_{m0} - P(\theta))\dot{\theta}dt = C$$

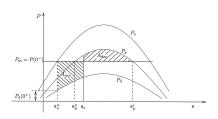
Changing the integration variable  $(x = \theta(t))$ 

$$\frac{1}{2}M\dot{\theta}^2 + \int\limits_{\theta_0}^{\theta} \left(P(x) - P_{m0}\right)dx = C$$

"kinetic" energy + "potential" energy = Constant

## 3.1 Transient (angle) stability: equal area criterion





# The system is stable if there exists an angle $\theta_p^i$ such that the areas are equal $(A_{acc} = A_{dec})$

Or, for a given  $\theta_e$ ,  $A_{acc} - A_{dec} < 0$ 

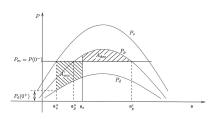
Туре	symbol	time
pre-fault	и	<i>t</i> < 0
during	d	$0 \le t < t_e$
post-fault	р	$t \geq t_e$

$$A_{acc} = \int_{\theta_u^0}^{\theta_e} (P_d(x) - P_{m0}) dx$$

$$A_{dec} = \int\limits_{ heta_e}^{ heta_p'} \left(P_{
ho}(x) - P_{m0}
ight) dx$$

### 3.1 Transient (angle) stability: equal area criterion





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#### **Curves:**

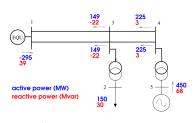
- during fault: capability of evacuating power on the network decreased due to low voltages
- post-fault: system weaker owing to the fault clearing actions (e.g., line tripping)

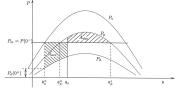
#### Critical clearing time ( $t_e = t_c$ ):

- Maximum fault duration so that the system returns to equilibrium
- When the system is at the stability limit :  $A_{acc} A_{dec} = 0$  and  $\theta_{\theta} = \theta_{c} = \theta(t_{c})$

# 3.1 Transient (angle) stability: example (RAMSES with 5-bus system)







#### Disturbances:

- 6-cycle (120ms) short-circuit without impedance on line "1-3", next to bus 3, cleared by opening that line, when the generator produces 450 MW
- the same fault cleared without line opening, when the generator produces 450 MW
- the same sequence as above, but with the generator producing 400 MW

#### 3.1 Transient (angle) instability countermeasures



- Modifying the pre-disturbance operating point:
  - reducing the active power generation
  - operating with higher excitation
- Automatic emergency controls:
  - actions on network: line auto-reclosing, fast series capacitor reinsertion, fast fault clearing - single pole breaker operation
  - actions in generators: (turbine) fast valving, generation shedding
  - action on "load": dynamic braking
- Other means:
  - equip generators with fast excitation system
  - control voltage at intermediate points in a long corridor: through synchronous condensers or static var compensators.

#### 3.2 Small-disturbance angle stability



- Small-signal (or small-disturbance) angle stability deals with the ability of the system to keep synchronism after being subject to a "small disturbance"
- "small disturbances" are those for which the system equations can be linearized around an equilibrium point
  - tools from linear system theory can be used (in particular eigenvalue and eigenvector analysis)
- following a small disturbance, the variation in electromagnetic torque T<sub>e</sub>
  can be decomposed into:

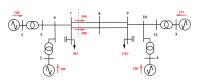
$$\triangle T_e = K_s \triangle \delta + K_d \triangle \omega$$

 $K_s \triangle \delta$ : synchronizing torque  $K_d \triangle \omega$ : damping torque

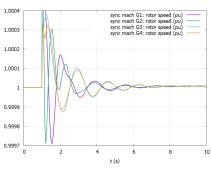
- a decrease in synchronizing torque will eventually lead to aperiodic instability (machine "going out of step")
- a decrease in damping torque will eventually lead to oscillatory instability (growing oscillations)

#### 3.2 Electromechanical oscillations





- A small "nudge" to the system (1 ms fault at bus 7) to excite the interarea modes.
- Oscillation of machines 1 and 2 against machines 3 and 4
- Period ~ 2 s





#### Local modes (involve a small part of the system)

- rotor angle oscillations of a single generator or a single plant against the rest of the system: local plant mode
  - can be studied using a one-machine infinite-bus system
- oscillations between rotors of a few generators close to each other: inter-machine or inter-plant mode oscillations
- typical range of frequencies of local plant and inter-plant modes: 0.7 to 2
   Hz
- may also be associated with inappropriate tuning of a control equipment (excitation system, HVDC converter, SVC, etc.): control mode

#### 3.2 Electromechanical oscillations



#### Global modes (involve large areas of the system, widespread effects)

- oscillations of a large group of generators in one area swinging against a group of generators in another area: interarea mode
- usually, the larger the group of generators, the slower the oscillations
- typical range of frequencies of interarea modes: 0.1 to 0.7 Hz
- more complex to analyze and to damp

# 3.2 Small-signal Stability



 Let's consider an autonomous system described by the differential equations:

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x})$$

and  $\mathbf{x}^*$  is an equilibrium point:  $\mathbf{f}(\mathbf{x}^*) = \mathbf{0}$ 

 If we linearize the system around the operating point and ignore higher order terms:

$$\dot{\triangle x} = \dot{x} = \left. \frac{\partial f}{\partial x} \right|_{x = x^*} x = Ax$$

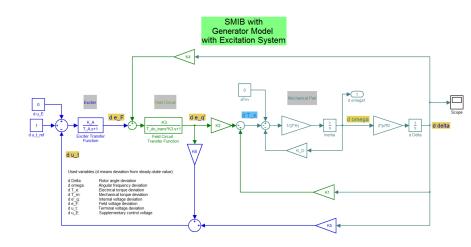
where  $\frac{\partial \mathbf{f}}{\partial \mathbf{x}}$  is the Jacobian of  $\mathbf{f}$  with respect to  $\mathbf{x}$ , and  $\mathbf{A} = \left. \frac{\partial \mathbf{f}}{\partial \mathbf{x}} \right|_{\mathbf{x} = \mathbf{x}^*}$  is the state matrix of the linearized system.

## 3.2 Stability of the equilibrium $x^*$



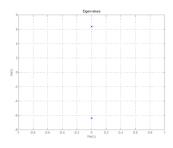
- Let  $\lambda$  be a real eigenvalue of matrix  $\boldsymbol{A}$ :
  - $\lambda$  < 0: The corresponding mode is stable (decaying exponential).
  - $\lambda > 0$ : The corresponding mode is unstable (growing exponential).
  - ullet  $\lambda=0$ : The corresponding mode has integrating characteristics.
- Let  $\lambda_{1,2} = \sigma \pm j\omega$  be a complex conjugate pair of eigenvalues of **A**:
  - $\Re(\lambda_{1,2}) < 0$ : The corresponding mode is stable (decaying oscillation).
  - $\Re(\lambda_{1,2}) > 0$ : The corresponding mode is unstable (growing oscillation).
  - \$\R(\lambda\_{1,2}) = 0\$: The corresponding mode is critically stable (undamped oscillation).
  - The following dynamic properties can be established:
    - Oscillation frequency:  $f = \frac{\omega}{2\pi}$
    - $\qquad \text{Damping ratio: } \zeta = \frac{-\sigma}{\sqrt{\sigma^2 + \omega^2}}$







# SMIB with classical generator model (mechanical damping torque $K_D=0$ )



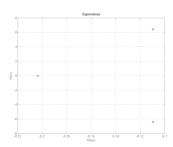
# Eigenvalues, synchronizing and damping torque coefficients

	$\sigma \pm j\omega$	ζ	f [Hz]	K <sub>sync</sub>	K <sub>damp</sub>
$\lambda_{1,2}$	0 ± <i>j</i> 6.39	-	1.02	0.76	0

Eigenvalues on imaginary axis → system is critically stable



#### SMIB including field circuit dynamics



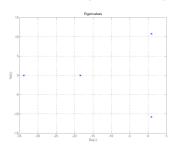
# Eigenvalues, synchronizing and damping torque coefficients

		$\sigma\pm j\omega$	ζ	f [Hz]	K <sub>sync</sub>	K <sub>damp</sub>
	$\lambda_{1,2}$	$-0.11 \pm j6.41$	0.02	1.02	-0.0008	1.53
ĺ	$\lambda_3$	$-0.20 \pm j0$	1.0	-	-0.77	0

# Eigenvalues moved to the left because field circuit adds damping torque



#### SMIB including excitation system



# Eigenvalues, synchronizing and damping torque coefficients

		$\sigma\pm j\omega$	ζ	f [Hz]	K <sub>sync</sub>	K <sub>damp</sub>
L	$\lambda_{1,2}$	0.88 ± <i>j</i> 10.79	-0.08	1.72	0.27	-10.60
	$\lambda_3$	$-33.83 \pm j0$	1.0	-	-19.81	0
ſ	$\lambda_4$	−18.46 ± <i>j</i> 0	1.0	-	-7.01	0

Eigenvalues moved to the right by the excitation system → System is unstable!

# 3.2 Left and right eigenvectors



Let's assume again the linearized system with the state matrix  $\mathbf{A}$  ( $n \times n$ ) and  $\lambda_i$  is one of its non-zero eigenvalues. Then:

•  $\mathbf{v}_i$  is the right eigenvector of  $\lambda_i$ :

$$\mathbf{A}\mathbf{v}_i = \lambda_i \mathbf{v}_i$$

•  $\mathbf{w}_i$  is the left eigenvector of  $\lambda_i$ :

$$\mathbf{A}^T \mathbf{w}_i = \lambda_i \mathbf{w}_i$$

In matrix form:

$$V = [v_1 \dots v_n]$$
  $W = \begin{bmatrix} w_1^T \\ \vdots \\ w_n^T \end{bmatrix}$ 

• It can be shown that:  $\mathbf{W} = \mathbf{V}^{-1}$  and  $\mathbf{WAV} = diag(\lambda_i) = \mathbf{\Lambda}$ 

# 3.2 Controllability and observability



Now, consider a system with state vector  $\mathbf{x}$ , input vector  $\mathbf{u}$  and a scalar output  $\mathbf{z}$ :

$$\dot{x} = Ax + Bu$$
 $z = Cx + Du$ 

• We change the variables (y = Wx)

$$\dot{y} = WAVy + WBu = \Lambda y + WBu$$
  
 $z = CVy + Du$ 

- For the *i*-th "mode"  $\lambda_i$  of the state matrix **A**:
  - the larger  $(\mathbf{W}\mathbf{B})_i = \mathbf{w}_i^T \mathbf{B}$ , the more the mode can be controlled by  $\mathbf{u}$
  - the larger  $(CV)_i = Cv_i$ , the more the mode can be observed in z

#### 3.2 Transfer function and residues



We can now build the transfer function of the system:

$$H(s) = \frac{Z(s)}{U(s)}$$

$$= CV(sI - \Lambda)^{-1}WB + D$$

$$= [Cv_1 \dots Cv_n] diag(\frac{1}{s - \lambda_i}) \begin{bmatrix} w_1^T B \\ \vdots \\ w_n^T B \end{bmatrix} + D$$

$$= \sum_{i=1}^n \frac{Cv_i w_i^T B}{s - \lambda_i} + D = \sum_{i=1}^n \frac{R_i}{s - \lambda_i} + D$$

The residue  $R_i$  relative to the *i*-th mode  $\lambda_i$ :

- depends on both the observability and the controllability of the mode
- would be zero in case of exact zero-pole cancellation

# 3.2 Synthesis of a stabilizing feedback using residues





Consider a compensator using z as input and acting on u:

• Which condition should be satisfied by the transfer function of the PSS in order to stabilize the critical mode  $\lambda_c$  of the uncompensated system?

The closed-loop transfer function is:

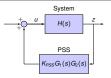
$$\frac{H(s)}{1 - K_{PSS}H(s)G_1(s)G_2(s)}$$

If  $\tilde{s}$  is one of the closed-loop poles:

$$1 - \textit{K}_{\textit{PSS}}\textit{H}(\tilde{s})\textit{G}_{1}(\tilde{s})\textit{G}_{2}(\tilde{s}) = 0 \Leftrightarrow 1 - \textit{K}_{\textit{PSS}}\left[\sum_{i=1}^{n}\frac{\textit{R}_{i}}{\tilde{s} - \lambda_{i}} + \textbf{\textit{D}}\right]\textit{G}_{1}(\tilde{s})\textit{G}_{2}(\tilde{s}) = 0$$

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The closed-loop transfer function is:

$$\frac{H(s)}{1 - K_{PSS}H(s)G_1(s)G_2(s)}$$

If  $\tilde{s}$  is one of the closed-loop poles:

$$1 - K_{PSS}H(\tilde{s})G_1(\tilde{s})G_2(\tilde{s}) = 0 \Leftrightarrow 1 - K_{PSS}\left[\sum_{i=1}^n \frac{R_i}{\tilde{s} - \lambda_i} + \mathbf{D}\right]G_1(\tilde{s})G_2(\tilde{s}) = 0$$

# 3.2 Synthesis of a stabilizing feedback using residues





Let's consider a closed-loop pole  $\tilde{s}$  lying on the branch of the root locus which starts from the open-loop pole  $\lambda_c$ . When the compensator gain  $K_{PSS}$  tends to zero,  $\tilde{s}$  tends to  $\lambda_c$ .

• Keeping only the dominant terms:

$$1 - R_c G_1(\lambda_c) G_2(\lambda_c) \lim_{K_{PSS} \to 0} \frac{K_{PSS}}{\tilde{s} - \lambda_c} = 0 \Leftrightarrow R_c G_1(\lambda_c) G_2(\lambda_c) = \lim_{K_{PSS} \to 0} \frac{\tilde{s} - \lambda_c}{K_{PSS}}$$

- In the complex plane  $\lim_{K_{PSS} \to 0} \frac{\tilde{s} \lambda_c}{K_{PSS}}$  is a vector tangent to the branch of the root locus starting from  $\lambda_c$ .
- In order to shift the eigenvalue to the left:
  - the branch of the root locus should leave  $\lambda_c$  at an angle of 180 degrees. Thus,  $G_1(\lambda_c)G_2(\lambda_c)$  must be such that  $\angle G_1(j\omega_c)G_2(j\omega_c) = \pm 180 - \angle R_c$
  - $R_cG_1(\lambda_c)G_2(\lambda_c)$  should be a real negative number



#### • Purpose:

Provide additional *damping torque* component in order to prevent the system from becoming unstable.

#### Approach:

Insert a feedback between *angular frequency* and *voltage setpoint* to "stabilize" a critical mode  $\lambda_c$ .

#### Block diagram:

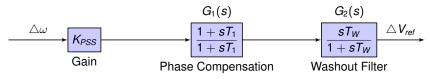


Figure: Block diagram of a simple PSS



#### Phase Compensation $G_1$ :

• shifts  $\lambda_c$  to the left in the complex plane by bringing a phase compensation according to the residue method :

$$\angle G_1(\lambda_c) \simeq \angle G_1(j\omega_c) = \pm 180 - \angle R_c$$

- $G_1(s)$  corresponds to one or several lead-lag filters
- the latter are "tuned" to provide their maximum phase shift  $\phi_m$  at the frequency  $\omega_c$

#### Washout Filter G<sub>2</sub>:

- in steady state and for slow variations, the PSS must not affect voltage regulation
- G<sub>2</sub>(s) is a high-pass filter
- $T_w$  is taken large enough to not modify the phase angle of  $G_1(s)$  for frequencies around  $\omega_c$ . For instance:

$$\frac{10}{T_w} \simeq \frac{\omega_c}{10}$$



#### Gain $K_{PSS}$ :

• adjusted until the corrected mode  $\tilde{\lambda}_c$  has a damping ratio :

$$\zeta = \frac{-\Re(\tilde{\lambda}_c)}{\sqrt{\Re(\tilde{\lambda}_c)^2 + \Im(\tilde{\lambda}_c)^2}} \ge 0.05 - 0.10$$

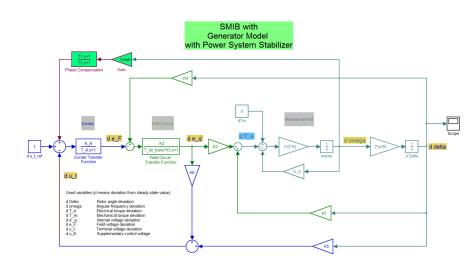
- while K<sub>PSS</sub> is increased, the other eigenvalues are monitored since they
  might move to the right (the residue method allows controlling a
  single mode)
- for excessive values of  $K_{PSS}$ , the branch of the root locus that starts from  $\lambda_c$  might "bend" to the right (the residue method focuses on a neighborhood of the mode to correct)



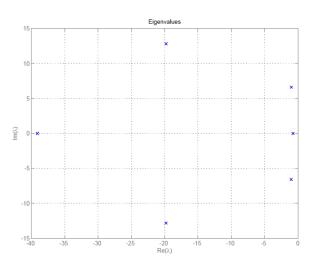
#### Low-pass Filter $G_3$ (optional):

- in a thermal power plant, the turbine stages, the generator and the exciter are mounted on a relatively long shaft. The latter has torsional oscillation frequencies in the range 10 – 15 Hz and higher
- the PSS must not excite those frequencies
- the risk is higher for a PSS using the rotor speed as input signal
- in this case,  $G_3$  is a low-pass filter so that the PSS contribution is negligible at the lowest torsional frequency and above.









## 3.3 (Rotor) angle stability: Remarks



#### Transient stability:

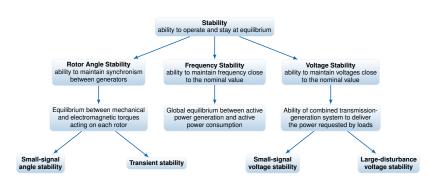
- depends on operating point and system parameters
- depends also on the disturbance
  - the system may be stable for disturbance 1 but not disturbance 2
  - if so, the system is insecure for 2, but as long as 2 does not happen, it can operate
  - usually, an N-1 security is required

#### Small-disturbance angle stability:

- depends on operating point and system parameters
- does not depend on the disturbance (assumed infinitesimal and arbitrary)
- is a necessary condition for operating a power system (small disturbances are always present)

## 3.3 Classification [1]





#### 4 Outline



- 1 Introduction
- 2 Voltage Stability
- **3 Rotor Angle Stability**
- 4 References

#### 4 Slides based on



- [1] P. Kundur, J. Paserba, V. Ajjarapu, G. Andersson, A. Bose, T. Van Cutsem, C. Canizares, N. Hatziargyriou, D. Hill, V. Vittal, A. Stankovic, and C. Taylor, "Definition and Classification of Power System Stability IEEE/CIGRE Joint Task Force on Stability Terms and Definitions," IEEE Trans. Power Syst., vol. 19, no. 3, pp. 1387–1401, Aug. 2004.
- [2] M. J. Gibbard, P. Pourbeik, and D. J. Vowles, "Small-signal stability, control and dynamic performance of power systems", University of Adelaide Press, Adelaide, 2015.