

EE302 Tutorial Sheet 2, Section S2, Tutorial 1st Feb 2024.

Q-1: Plot in the same figure the step responses of following transfer fns-

$$\frac{s-1}{2s+3}, \frac{1}{s+2}, \frac{1}{s+3}, \frac{1}{1+0.3s}, \frac{1}{1+0.2s}, \frac{s+2}{4s+5}$$

Q-2: Suppose $G(s)$ is proper and 1st order.

then check if $G(0) = \lim_{t \rightarrow \infty} \text{step response}$: under what assumptions?

Insightful!

Q-3(a) Suppose $G(s)$ is proper and 1st order and stable (i.e. poles in OLHP) \equiv OLHP

If $y(t)$ is the step response, then $y(0^+) \neq 0 \Leftrightarrow G(s)$ has a zero.
(b) Show that $y(0^+) \cdot y(\infty) < 0 \Leftrightarrow G(s)$ has a non-minimum phase zero.

Q-4: Use initial value theorem for finding

initial value (i.e. $t \rightarrow 0^+$ for finding not at $t=0$) of impulse response and step response of $\frac{10}{3s+61}, \frac{10s+9}{3s+61}, \frac{10}{s-3}$

Q-5: For proper $G(s)$, relative degree of $G(s) := \deg d - \deg n$

Consider step response of $G(s)$.

$$G(s) = \frac{n(s)}{d(s)} \quad (d \neq 0). \quad n, d \equiv \text{polynomials.}$$

Suppose relative degree of $G(s) = 7$.

then what can be told about $y(0), \dot{y}(0^+), \ddot{y}(0^+), \dots$

$$\left. \frac{d^6}{dt^6} y \right|_{0^+}, \left. \frac{d^7}{dt^7} y \right|_{0^+}, \left. \frac{d^8}{dt^8} y \right|_{0^+}$$

Q-6: Suppose $G(s)$ is biproper and

has pole, zero in OLHP, G is 1st order.

Suppose $G(s) > 0$. (DC gain $\equiv G(0)$)
for stable $G(s)$.

depending on step response $y(t)$'s initial value $y(0^+)$ & $\dot{y}(0^+)$ (+ve or -ve), what can be told about whether

pole or zero is closer to $j\mathbb{R}$?

Q-7:

Suppose impulse response of $G(s)$ (proper, stable) is $a\delta + b e^{-t/T}$ (for $t \geq 0$). $T > 0$.

Find relative condition between $a, b, T \in \mathbb{R}$ for

- zero to be closer than pole (and both in OLHP) from $j\mathbb{R}$.
- pole to be closer than zero from $j\mathbb{R}$ (& both in OLHP)
- zero at origin $s=0$.

Q 8 onwards: see following pages.

Q8) Find steady state value for the following

(a) $\frac{100}{(s+6)^2 + 64}$ for unit step input

(b) $\frac{1}{s^2 + 3s + 2}$ for unit impulse input

Q9) Find the step response for $\frac{-1}{s+10}$. Also find the step response when a zero at $s=10$ is added.

[i.e.] step response of the non-minimum phase system $\frac{-(s-10)}{s+10}$

Q10) For unit step input find initial rise rate for the following $\frac{5}{s+5}$ & $\frac{20}{s+20}$. Also find rise time & settling time.

Q. :- For the following transfer functions, write the general form of step response

a) $G(s) = \frac{400}{s^2 + 12s + 400}$

c) $G(s) = \frac{225}{s^2 + 30s + 225}$

b) $G(s) = \frac{900}{s^2 + 90s + 900}$

d) $G(s) = \frac{625}{s^2 + 625}$

Q. :- For the given transfer functions, find ξ & ω_n

a) $G(s) = \frac{36}{s^2 + 4.2s + 36}$

b) $\frac{20}{s^2 + 6s + 44}$

c) $G(s) = \frac{s+2}{s^2 + 9}$

d) $\frac{5}{(s+3)(s+6)}$

Also state the nature of each response (overdamped, underdamped, & so on).

Q. :- For each of the second order systems given below find ξ , ω_n , T_d , T_p , T_s & % overshoot.

a) $T(s) = \frac{16}{s^2 + 3s + 16}$

b) $\frac{0.04}{s^2 + 0.02s + 0.04}$

c) $T(s) = \frac{1.05 \times 10^7}{s^2 + 1.6 \times 10^3 s + 1.05 \times 10^7}$

Q. :- Find the transfer function of a second-order system that yields a 12.3% overshoot and a settling time of 1 second.

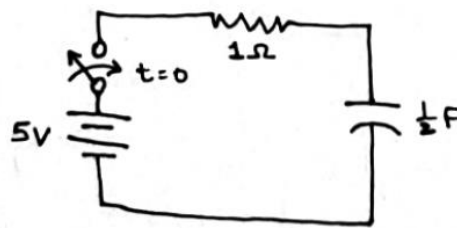
Q Consider a first order system of the form

$$\tau \frac{dx}{dt} = -x + u, \quad y = x$$

We say that the parameter τ is the time constant for the system, since when the input is zero, the system approaches the origin as $e^{-t/\tau} x(0)$. For this model, show the following:-

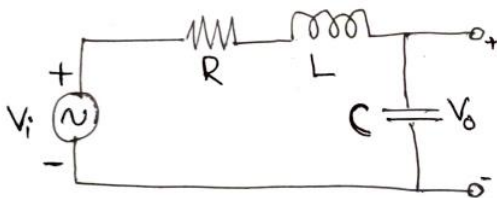
- i) Show that the rise time for a step response of the system is approximately 2τ .
- ii) Show that the 1%, 2% and 5% settling times are approximately equal to 4.6τ , 4τ and 3τ respectively.

Q Find the capacitor voltage in the network shown below, if the switch closes at $t=0$. Assume zero initial conditions. Also find the time constant, rise time, and settling time for the capacitor voltage.



Q. Consider a series RLC circuit:

For $R = 1\Omega$, $L = 1H$ and $C = 1F$, the time constant of circuit (in sec) is



Simple example to notice how feedback is more robust than open loop controllers.

This problem is for concluding the robustness of feedback controller w.r.t. changes in system pole and/or initial condition $x(0)$.

Consider $\dot{x} = 3x + u$ & $x(0) = 4$.
model

— Check input $u(t)$ that gives $x(t) = x(0)e^{-2t}$
(after control action)
($u = -5x$ would have sufficed; but obtain $u(t)$ explicitly.)

— Now apply $u(t)$ from above explicitly to
 $\dot{x} = 3x + u$ & $\mathcal{L}(\quad) = \mathcal{L}(\quad)$ to get
 $sX(s) - 3X(s) - x(0) = U(s)$.

thus $X(s) = \frac{1}{(s-3)} [\quad]$ & check that the $u(t)$ indeed gives $x(0)e^{-2t}$.

Now use same $u(t)$ for actual system $\dot{x} = 3.1x + u$
and/or actual initial condition $x(0) = 3.9$

check if $x(t)$ is still $x(0)e^{-2t}$ or how different $x(t)$ could be due to the mismatch in system equation/initial condition.