

# EE-229 (Signal Processing-I)

## Homework-4

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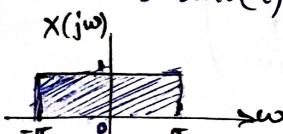
Sol 1] T.P.T  $\int_{-\infty}^{\infty} \text{sinc}(t) dt = 1$  and  $\int_{-\infty}^{\infty} \text{sinc}^2(t) dt = 1$  where  $\text{sinc}(t) = \frac{\sin(\pi t)}{\pi t}$ .

Proof: I use the following properties:

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega$$

$$\frac{\sin(\pi t)}{\pi t} \xrightarrow{\text{FT}} X(j\omega) = \begin{cases} 1, & |\omega| < \pi \\ 0, & |\omega| > \pi \end{cases}$$

(<sup>o</sup>Parseval's theorem)

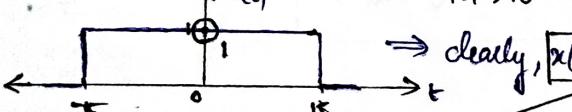
$$\because \text{sinc}(t)$$
 is a real-valued fn,  $\text{sinc}^2(t) = |\text{sinc}(t)|^2$ .  $\therefore \int_{-\infty}^{\infty} |\text{sinc}(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} 1 d\omega$ 


$$\Rightarrow \int_{-\infty}^{\infty} \text{sinc}^2(t) dt = \frac{1}{2\pi} (1 \cdot \pi - 1 \cdot (-\pi))$$

$$= 2\pi/2\pi = 1$$

∴ Proved

Now, consider  $x(t) = \begin{cases} 1, & |t| \leq \pi \\ 0, & |t| > \pi \end{cases}$  Now,  $X(\omega) = \frac{2\sin(\omega \cdot \pi)}{\omega} \times \frac{\pi}{\pi} = 2\pi \text{sinc}(\omega)$ .



$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} 2\pi \text{sinc}(\omega) e^{j\omega t} d\omega$$

$$\Rightarrow x(t) = \int_{-\infty}^{\infty} \text{sinc}(\omega) e^{j\omega t} d\omega$$

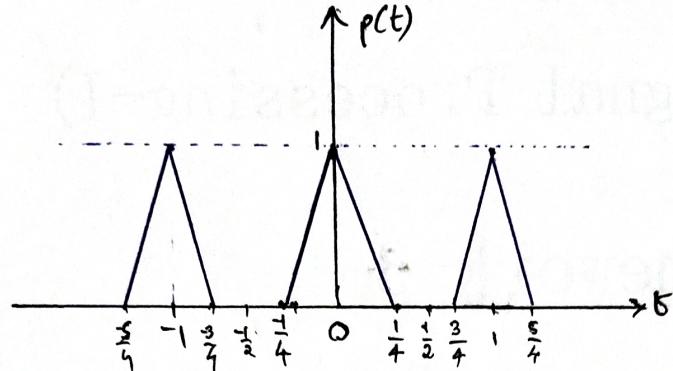
$$\Rightarrow x(0) = \int_{-\infty}^{\infty} \text{sinc}(\omega) e^0 d\omega$$

$$= 1 = \int_{-\infty}^{\infty} \text{sinc}(t) dt$$

∴ proved

Put t = 0 !!

Sol. 2] We know that FT of a triangular pulse is sinc<sup>2</sup> fn, let's anyways derive its



$\rightarrow$

$$p(t) = \begin{cases} 4t+1 & -\frac{1}{4} \leq t \leq 0 \\ -4t+1 & 0 < t \leq 1 \\ 0 & \text{otherwise} \end{cases}$$

(considering one period)

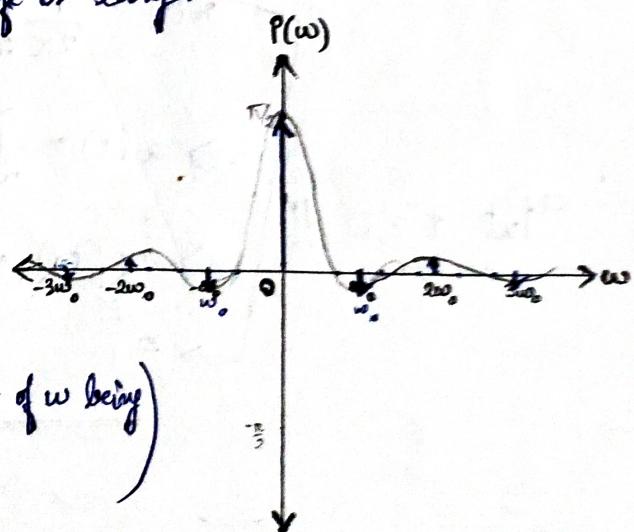
$$\begin{aligned} a) \quad a_k &= \frac{1}{T} \int_{-T/2}^{T/2} p(t) e^{-j\omega_0 k t} dt = \frac{1}{(1)} \left( \int_{-\frac{1}{4}}^0 (4t+1) dt + \int_0^{\frac{1}{4}} (-4t+1) dt \right) \quad \left( \because \int_{-\frac{1}{2}}^{\frac{1}{2}} 0 dt = 0 \right. \\ &\quad \left. \text{and } \int_{\frac{1}{4}}^{\frac{1}{4}} 0 dt = 0 \right) \\ &= \left[ \frac{(4t+1)e^{-j\omega_0 t}}{(-j\omega_0)} - \int 4 \cdot \frac{e^{-j\omega_0 t}}{(-j\omega_0)} dt \right]_{-1/4}^0 + \left[ \frac{(-4t+1)e^{-j\omega_0 t}}{(-j\omega_0)} - \int (-4) \cdot \frac{e^{-j\omega_0 t}}{(-j\omega_0)} dt \right]_{0}^{1/4} \\ &= \left( \frac{-1+4t}{j\omega_0} e^{-j\omega_0 t} + \frac{4e^{-j\omega_0 t}}{\omega_0^2} \right)_{-1/4}^0 + \left( \frac{(4t-1)e^{-j\omega_0 t}}{j\omega_0} - \frac{4e^{-j\omega_0 t}}{\omega_0^2} \right)_{0}^{1/4} \\ &= \left( \frac{-1}{j\omega_0} + \frac{4}{\omega_0^2} - \left( 0 + \frac{4e^{+j\omega_0/4}}{\omega_0^2} \right) \right) + \left( 0 - \frac{4e^{-j\omega_0/4}}{\omega_0^2} - \left( \frac{4e^{-j\omega_0(0)}}{\omega_0^2} - \frac{e^0}{j\omega_0} \right) \right) \\ &= \left( \cancel{\frac{-1}{j\omega_0}} + \frac{4}{\omega_0^2} - 4 \frac{e^{j\omega_0/4}}{\omega_0^2} \right) + \left( \cancel{\frac{-4e^{-j\omega_0/4}}{\omega_0^2}} + \frac{4}{\omega_0^2} + \frac{1}{j\omega_0} \right) \\ &= \frac{8}{\omega_0^2} - \frac{8}{\omega_0^2} \left( \frac{e^{j\omega_0/4} + e^{-j\omega_0/4}}{2} \right)^2 = \frac{8}{\omega_0^2} \left( 1 - \cos \frac{\omega_0}{4} \right) \Big|_{\omega_0 = \omega_0} \\ \Rightarrow a_k &= \frac{8}{\omega_0^2} \cdot 2 \sin^2 \frac{\omega_0}{8} = \frac{1}{4} \left( \frac{\sin(\omega_0/8)}{(\omega_0/8)} \right)^2 = \boxed{\frac{1}{4} \operatorname{sinc}^2 \left( \frac{k\omega_0}{8\pi} \right)} \quad \text{where } \omega_0 = \frac{2\pi}{T} \end{aligned}$$

b) Now that we've got Fourier coefficients, life is easy:

$$X(j\omega) = \sum_{k=-\infty}^{+\infty} 2\pi a_k \delta(\omega - k\omega_0)$$

$$\therefore X(j\omega) = \frac{\pi}{2} \sum_{k=-\infty}^{\infty} \operatorname{sinc}^2 \left( \frac{k\omega_0}{8\pi} \right) \delta(\omega - k\omega_0)$$

$\left( \{a_k\} \text{ scaled by } 2\pi \text{ and sampled at values of } \omega \text{ being integral multiples of } \omega_0 (= 2\pi/T) \right)$

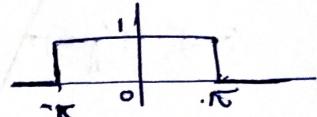


2.c) Let  $X(j\omega)$  be the Fourier Transform of the aperiodic signal  $x(t)$ .

Then, for  $y(t) = x(t) \cdot p(t)$ ,  $\mathcal{Y}(j\omega) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\theta) P(j(\omega-\theta)) d\theta$

$$\therefore \mathcal{Y}(j\omega) = \frac{1}{2\pi} \cdot \frac{1}{2} \int_{-\infty}^{+\infty} X(j\theta) \cdot \sum_{k=-\infty}^{+\infty} \operatorname{sinc}^2\left(\frac{k\omega_0}{8\pi}\right) \delta(\omega-\theta-k\omega_0) d\theta$$

2.d) For  $x(t) = \operatorname{sinc}(t)$ ,  $X(j\omega) = \begin{cases} 1, & |\omega| \leq \pi \\ 0, & |\omega| > \pi \end{cases}$



$\because P(\omega)$  is a train of impulses ( $\delta(\omega) + \delta(\omega - k\omega_0), k \in \mathbb{Z}$ )

$\therefore$  when it is convolved with  $X(j\omega)$ ,

it reproduces itself!  $\therefore$  only change is the scaling by  $\frac{1}{2\pi}$ .

$$\therefore \mathcal{Y}(j\omega) = \frac{1}{4} \sum_{k=-\infty}^{+\infty} \operatorname{sinc}^2\left(\frac{k\omega_0}{8\pi}\right) \delta(\omega - k\omega_0)$$



Sol. 4]

LTI system:  $x(t) * h(t) = y(t)$  and  $X(j\omega) \cdot H(j\omega) = Y(j\omega)$ .

Also,  $x(t) = e^{-at} u(t) \xrightarrow{\text{FT}} X(j\omega) = \frac{1}{a+j\omega} \quad (\operatorname{Re}(a) > 0)$ ,

$$\begin{aligned} \text{a)} \quad H(j\omega) &= \frac{Y(j\omega)}{X(j\omega)} \quad \text{Now, } X(j\omega) = \frac{1}{1+j\omega} + \dots = \frac{1}{3+j\omega} \\ &\text{and } Y(j\omega) = \frac{2}{1+j\omega} + \frac{(-2)}{4+j\omega} \\ \Rightarrow H(j\omega) &= \frac{3(3+j\omega)}{(2+j\omega)(4+j\omega)} \end{aligned}$$

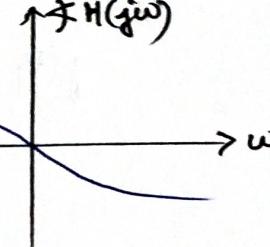
$$\begin{aligned} \text{b)} \quad &\text{Decomposing } H(j\omega) \text{ into partial fractions, } \frac{3}{(2+j\omega)(4+j\omega)} \times \frac{1}{2} \\ &= \frac{3}{2} \left( \frac{1}{4+j\omega} + \frac{1}{2+j\omega} \right) \end{aligned}$$

$$\therefore h(t) = \frac{3}{2} (e^{-4t} + e^{-2t}) u(t)$$

$$\begin{aligned} \text{c)} \quad &\because \frac{Y(j\omega)}{X(j\omega)} = \frac{(9+3j\omega)}{8+6j\omega-w^2} \Rightarrow 8Y + 6j\omega \cdot Y - w^2 \cdot Y = 9 \cdot X + 3j\omega \cdot X \\ &\Rightarrow 8y(t) + 6 \frac{dy}{dt} y(t) + \frac{d^2}{dt^2} y(t) = 9x(t) + 3 \frac{dx}{dt} x(t) \end{aligned}$$

$$[Sol. 5] \text{ a) } |H(j\omega)| = \frac{|a-j\omega|}{|a+j\omega|} = \frac{\sqrt{a^2+\omega^2}}{\sqrt{a^2+\omega^2}} = 1 \quad \therefore \quad \frac{|H(\omega)|}{1} \xrightarrow{\omega}$$

Now,  $\angle H(j\omega) = \angle \left( \frac{a-j\omega}{a+j\omega} \right) = \tan^{-1} \left( \frac{-\omega}{a} \right) - \tan^{-1} \left( \frac{\omega}{a} \right)$   
 $= -2\tan^{-1} (\omega/a)$ .



b) for  $a=1$ ,  $|H(j\omega)| = 1$ ,  $\angle H(j\omega) = -2\tan^{-1}(\omega) \Rightarrow H(j\omega) = \frac{1-j\omega}{1+j\omega}$ .

$$A(\omega) = |H(\omega)| \quad \therefore H(\omega) = |H(\omega)| e^{j\phi(\omega)} \\ = e^{j\phi(\omega)} = e^{j(-2\tan^{-1}(\omega))}.$$

Now, for  $x(t) = \cos(t/\sqrt{3}) + \cos(b) + \cos(\sqrt{3}t)$

$$= \frac{e^{jt/\sqrt{3}} + e^{-jt/\sqrt{3}}}{2} + \frac{e^{jb} + e^{-jb}}{2} + \frac{e^{j\sqrt{3}t} + e^{-j\sqrt{3}t}}{2}$$

$$\Rightarrow X(j\omega) = \frac{1}{2}\pi \left[ \underbrace{\delta(\omega - 1/\sqrt{3}) + \delta(\omega + 1/\sqrt{3})}_{\rightarrow} + \underbrace{\delta(\omega - b) + \delta(\omega + b)}_{\rightarrow} + \underbrace{\delta(\omega - \sqrt{3}) + \delta(\omega + \sqrt{3})}_{\rightarrow} \right]$$

$$\therefore Y(j\omega) = X(j\omega) \cdot H(j\omega) = \pi e^{-2j\tan^{-1}\omega} \left( \delta(\omega - 1/\sqrt{3}) + \delta(\omega + 1/\sqrt{3}) + \delta(\omega - b) + \delta(\omega + b) + \delta(\omega - \sqrt{3}) + \delta(\omega + \sqrt{3}) \right)$$

$$\therefore y(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} Y(j\omega) e^{j\omega t} d\omega. \quad \text{But } \because \int_{-\infty}^{+\infty} f(x) \delta(x-a) dx = f(a).$$

$$\therefore y(t) = \frac{1}{2} \left( e^{j(t/\sqrt{3} - 2\tan^{-1}(1/\sqrt{3}))} + e^{j(-t/\sqrt{3} - 2\tan^{-1}(-1/\sqrt{3}))} + e^{j(t/b - 2\tan^{-1}(b))} + e^{j(t/\sqrt{3} - 2\tan^{-1}(\sqrt{3}))} + e^{j(-t/\sqrt{3} - 2\tan^{-1}(-\sqrt{3}))} \right) \\ = \frac{1}{2} \left( 2\cos \left( \frac{t}{\sqrt{3}} - 2\pi/6 \right) + 2\cos \left( t/b - \pi/4 \times 2 \right) + 2\cos \left( \sqrt{3}t - 2\sqrt{3} \right) \right)$$

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$$y(t) = \cos \left( \frac{t}{\sqrt{3}} - \frac{\pi}{3} \right) + \cos \left( t - \frac{\pi}{2} \right) + \cos \left( \sqrt{3}t - \frac{2\pi}{3} \right).$$

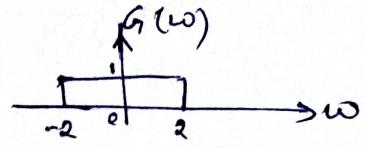

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Sol 6] Given,  $g(t) = x(t) \cos(t)$  and  $\text{FT}(g(t)) : G(\omega) = \begin{cases} 1 & |\omega| \leq 2 \\ 0 & |\omega| > 2 \end{cases}$

a)

$$g(t) = F^{-1}(G(\omega))$$

$$= \frac{\sin \frac{\omega^2 t}{2}}{\pi t} = \frac{2 \sin t \cos t}{\pi t}$$



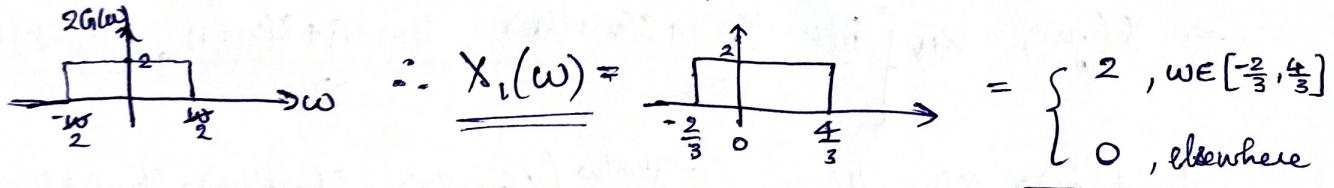
$$\therefore x(t) = \frac{g(t)}{\cos(t)} = \boxed{\frac{2 \sin t}{\pi t}}$$

b) Modulation property of CTFT:

If  $x(t) \xrightarrow{\text{FT}} X(\omega)$  then  $x(t) \cos(\omega_0 t) \xrightarrow{\text{FT}} \frac{1}{2} [X(\omega - \omega_0) + X(\omega + \omega_0)]$ .

$\therefore$  if  $g(t) = x_1(t) \cos(\frac{2t}{3})$  then  $G(\omega) = \frac{1}{2} [X_1(\omega - \frac{2}{3}) + X_1(\omega + \frac{2}{3})]$

$$\Rightarrow 2G(\omega) = X_1(\omega - \frac{2}{3}) + X_1(\omega + \frac{2}{3})$$



Sol. 7] Considering Fourier transform of the entire eqn:

$$\Rightarrow j\omega Y(j\omega) + 10Y(j\omega) = X(j\omega)Z(j\omega) - X(j\omega) \quad (\because x * y \leftrightarrow X(1 \cdot Y))$$

$\Rightarrow$  a)  $H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \frac{Z(j\omega) - 1}{(10 + j\omega)}$ , Now,  $Z(j\omega) = \frac{1}{1+j\omega} + 3$

$$\therefore H(j\omega) = \frac{3 + 2j\omega}{(1+j\omega)(10+j\omega)} = \frac{1}{9} \cdot \frac{1}{1+j\omega} + \frac{17}{9} \cdot \frac{1}{10+j\omega}$$

b)  $h(t) = F^{-1}(H(j\omega))$

$$= \frac{1}{9} e^{-t} u(t) + \frac{17}{9} e^{-10t} u(t)$$

Sol. 8] a)

$$h(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(\omega) e^{j\omega t} d\omega$$

$$= \frac{1}{2\pi} \left( \int_{-\pi}^{\pi} e^{j\omega t} d\omega + \int_{\pi}^{2\pi} e^{j\omega t} d\omega \right)$$

$$= \frac{1}{2\pi} \left( \int_{2\pi}^{\pi} e^{-j\tilde{\omega}t} (-d\tilde{\omega}) + \int_{\pi}^{2\pi} e^{j\omega t} d\omega \right) = \frac{1}{2\pi} \left( \int_{\pi}^{2\pi} e^{-j\tilde{\omega}t} d\tilde{\omega} + \int_{\pi}^{2\pi} e^{j\omega t} d\omega \right)$$

$$= \frac{1}{2\pi} \left( \int_{\pi}^{2\pi} 2\cos(\omega t) d\omega \right) = \frac{1}{2\pi} \cdot \frac{2\sin \omega t}{t} \Big|_{\pi}^{2\pi}$$

$$= \frac{1}{\pi t} (\sin(2\pi t) - \sin(\pi t)) = \underline{\underline{2 \operatorname{sinc}(2t) - \operatorname{sinc}(t)}} .$$

b) No!

$$\text{BIBO stability} \equiv \int_{-\infty}^{\infty} |h(t)| dt < \infty$$

But  $\int_{-\infty}^{\infty} |2\operatorname{sinc}(2t) - \operatorname{sinc}(t)| dt = 2 \int_0^{\infty} |2\operatorname{sinc}(2t) - \operatorname{sinc}(t)| dt$

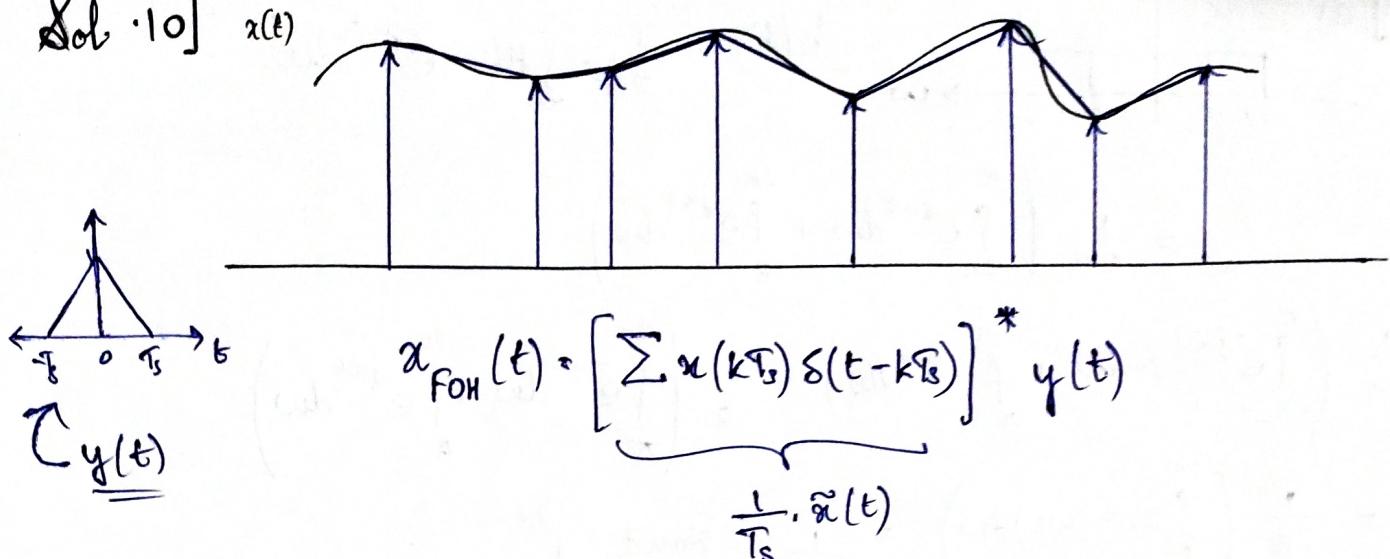
$\because |\operatorname{sinc}| \rightarrow \text{unstable} \therefore \text{it is an ideal filter,}$

$$\therefore 2 \int_0^{\infty} |2\operatorname{sinc}(2t) - \operatorname{sinc}(t)| dt > \infty$$

$\therefore \underline{\text{Unstable}} \quad \therefore \text{planned}$

Yes,  $\therefore$  every ideal low-pass filter is unstable.

Sol. 10]



$$\frac{1}{T_s} \tilde{x}(w) = \frac{1}{T_s} \sum X(w - kw_0)$$

$$\text{Now, } Y(w) = \int_{-\infty}^{\infty} y(t) e^{-jwt} dt = \int_{-T_s}^{0} \left( \frac{E}{T_s} + 1 \right) e^{-jwt} dt + \int_0^{T_s} \left( \frac{-E}{T_s} + 1 \right) e^{-jwt} dt$$

$$= \int_{-T_s}^{T_s} e^{-jwt} dt + \frac{1}{T_s} \left( \int_{-T_s}^0 te^{-jwt} dt + \int_0^{T_s} -te^{-jwt} dt \right)$$

$$= \frac{e^{-jwt}}{(-j\omega)} \Big|_{-T_s}^{T_s} + \frac{1}{T_s} \left( \int_0^{T_s} -te^{jwt} dt + \int_0^{T_s} -te^{-jwt} dt \right)$$

$$= \frac{-2j \sin(\omega T_s)}{-j\omega} + \frac{1}{T_s} \left( \int_0^{T_s} -t \times 2 \cos(\omega t) dt \right)$$

$$= \frac{2 \sin(\omega T_s)}{\omega} + \frac{1}{T_s} \left( -2 \left( \frac{t \sin(\omega t)}{\omega} + \frac{\cos(\omega t)}{\omega^2} \right) \right) \Big|_0^{T_s}$$

$$= \cancel{\frac{2 \sin(\omega T_s)}{\omega}} + \cancel{\frac{(-2)}{T_s} \left( \frac{T_s \sin(\omega T_s)}{\omega} + \frac{\cos(\omega T_s)}{\omega^2} - \frac{1}{\omega^2} \right)}$$

$$= \frac{+2}{\omega^2 T_s} \left( 1 - \cos(\omega T_s) \right) = \frac{4 \sin^2 \left( \frac{\omega T_s}{2} \right)}{\omega^2 T_s} = T_s \sin^2 \left( \frac{\omega T_s}{2\pi} \times \pi \right)$$

$$= T_s \sin^2 \left( \frac{\omega T_s}{2\pi} \right)$$

$$\therefore X_{FOH}(w) = \frac{1}{T_s} \sum X(w - kw_0) \times T_s \sin^2 \left( \frac{\omega T_s}{2\pi} \right)$$

$$= \sin^2 \left( \frac{\omega T_s}{2\pi} \right) \cdot \underbrace{\sum X(w - kw_0)}_{\text{Spectrum of } x(t)}$$