

# CitizenGW-Compute: Turning Computational Operations into a Gravitational Wave Sensor (P-II)

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## Abstract

We propose a citizen science framework to explore gravitational wave (GW) properties by exploiting modern computing systems as auxiliary detectors. Instead of measuring fractional deviations in raw clock frequency, which are minuscule (order  $10^{-12}$  Hz for GHz clocks), we leverage the computational rate (operations per second) as an amplification mechanism. This approach scales deviations by factors of  $10^3$  to  $10^6$ , potentially rendering otherwise undetectable effects into measurable statistical signals. The framework includes dummy artificial neural network (ANN) workloads, distributed logging across commodity PCs, and extensions to high-performance computing (HPC) clusters such as EAGLE, with real-time correlation against LIGO/Virgo alerts.

## 1 Introduction

Gravitational waves (GWs) induce minute spacetime perturbations. Current detectors (LIGO, Virgo) are highly specialized laser interferometers optimized for  $10\text{--}10^3$  Hz bands. Here, we explore whether distributed computing systems—with stable clocks, GPUs, and CPUs—could serve as complementary sensors by monitoring deviations in computational throughput.

## 2 Theoretical Framework

### 2.1 Weak-field approximation

In linearized general relativity, the spacetime metric is approximated as:

$$g_{\mu\nu} = \eta_{\mu\nu} + h_{\mu\nu}, \quad |h_{\mu\nu}| \ll 1, \quad (1)$$

where  $h_{\mu\nu}$  is the GW perturbation. For a monochromatic GW of strain amplitude  $h \sim 10^{-21}$ , the fractional frequency shift on a stable clock is

$$\frac{\Delta f}{f} \sim h(t). \quad (2)$$

For a GHz-scale clock ( $f \sim 10^9$  Hz), this yields

$$\Delta f \sim 10^{-12} \text{ Hz}, \quad (3)$$

far below detection thresholds.

## 2.2 Principle of Amplification

We propose shifting from frequency  $f$  to computational throughput  $R$  (operations per second). If each core performs  $A$  operations per cycle, then

$$R = Af. \quad (4)$$

The fractional effect is

$$\frac{\Delta R}{R} \sim h(t), \quad \Delta R \sim Rh(t). \quad (5)$$

Thus,

$$\Delta R \sim Afh(t). \quad (6)$$

## 3 Numerical Worked Examples

### 3.1 Consumer-grade example

Consider a single core with  $f = 10^9$  Hz and per-cycle arithmetic capacity  $A = 10^4$ . Then

$$R = Af = 10^{13} \text{ ops/s}, \quad (7)$$

$$\Delta R = Rh = 10^{13} \times 10^{-21} = 10^{-8} \text{ ops/s}. \quad (8)$$

If  $A = 10^6$  (GPU-like logical throughput) the absolute change becomes  $10^{-6}$  ops/s though the fractional remains  $10^{-8}$ .

### 3.2 HPC-scale example (EAGLE/Azure)

Assume aggregate throughput  $R_{\text{tot}} = 10^{17}$  FLOP/s and  $h = 10^{-21}$ . Then

$$\Delta R_{\text{tot}} = R_{\text{tot}}h = 10^{-4} \text{ FLOP/s}. \quad (9)$$

Per-node expectation with  $N_{\text{CPU}} = 1.5 \times 10^5$  is

$$\Delta R_{\text{per}} = \frac{\Delta R_{\text{tot}}}{N_{\text{CPU}}} \approx 6.7 \times 10^{-10} \text{ FLOP/s}. \quad (10)$$

Coherent stacking across nodes improves SNR by  $\sqrt{N}$ , but absolute rates remain extremely small compared to telemetry quantization and environmental noise.

## 4 Computational Cost: Big- $\mathcal{O}$

For a cascaded ANN with  $L$  dense layers and layer width  $n$ , and  $I$  inference iterations, the flop count is approximately

$$\text{FLOPs} = I \cdot L \cdot 2n^2 = \mathcal{O}(ILn^2). \quad (11)$$

If using batched matrix–matrix kernels cost can approach  $\mathcal{O}(ILn^3)$  depending on blocking and algorithmic choices. Given per-core throughput  $R$ , the run time is

$$T_{\text{run}} \approx \text{FLOPs}/R. \quad (12)$$

Longer runs reduce fractional variance via averaging but reduce temporal sampling resolution.

## 5 Resolving Power and Bandwidth Dependencies

### 5.1 Definitions

Define:

- $f$ : per-core clock frequency (Hz).
- $A$ : ops-per-cycle amplification factor (dimensionless).
- $R = Af$ : per-core operations/sec.
- $y(t) = (R_{\text{meas}}(t) - R)/R$ : fractional throughput time series.
- $S_y(f)$ : one-sided PSD of  $y(t)$ .
- $T$ : coherent integration time.
- $B$ : effective analysis bandwidth (Hz).
- $N$ : number of independent nodes.

### 5.2 Per-core resolving power

We define per-core resolving power at frequency  $f_{\text{gw}}$  as

$$\eta_{\text{core}}(T; f_{\text{gw}}) = \frac{1}{\sigma_y(T; f_{\text{gw}})}, \quad (13)$$

where, for white fractional noise,  $\sigma_y(T; f_{\text{gw}}) \simeq \sqrt{S_y(f_{\text{gw}})/(2T)}$ .

### 5.3 Network resolving power

Assuming independent noise realizations,

$$\eta_{\text{net}}(T; f_{\text{gw}}) = \sqrt{N} \eta_{\text{core}}(T; f_{\text{gw}}). \quad (14)$$

### 5.4 Sensitivity vs. bandwidth

For a band-limited estimator of ENBW  $B$ , variance scales as  $\sigma_y^2 \propto S_y B/T$ . Narrowing  $B$  improves sensitivity as  $\sigma_y \propto \sqrt{B/T}$  at the cost of responsiveness.

## 6 Narrowband Tuning: Digital Antenna / Lock-in Amplifier Analogy

To focus sensitivity at a target narrowband  $f_0$ , implement a deterministic gating waveform  $g(t)$  (sinusoid, multi-tone, PRBS) that modulates the computational load. The measurement chain is:

1. Acquire  $y(t)$  while applying  $g(t)$ .
2. Multiply  $y(t)$  by a coherent reference  $\cos(2\pi f_0 t)$  (mixing/heterodyne).
3. Low-pass filter the product with cutoff  $B$  and integrate for  $T$ .

The resulting demodulated output behaves like a lock-in amplifier: the effective SNR scales as

$$\text{SNR} \propto \frac{h_0 \sqrt{T}}{\sqrt{S_y(f_0) B}}. \quad (15)$$

Design choices for  $g(t)$  determine the effective “Q” (selectivity) of the digital antenna and its trade-off between gain and cyclostationary noise.

## 7 HPC / Cloud Extension: Practical Considerations

For a production-level experiment on EAGLE or Azure:

- Reserve nodes and ensure job start/stop synchronization (scheduler barriers, job arrays).
- Pin CPU frequency states (disable DVFS), fix SMT/Hyper-Threading, and isolate cores where possible.
- Instrument each node with high-resolution timers, temperature sensors, and interrupt counters.
- Run deterministic kernels with prebuilt weights and fixed data to avoid variability.
- Collect fractional throughput  $y_i(t)$  at a chosen sampling cadence and forward to an aggregation node for whitening and coherent stacking.

A practical coherent estimator is the inverse-PSD weighted average in Fourier space:

$$\hat{h}(f) = \frac{\sum_{i=1}^N \tilde{y}_i(f)/S_{y,i}(f)}{\sum_{i=1}^N 1/S_{y,i}(f)}. \quad (16)$$

Assuming  $S_{y,i} \approx S_y$  this reduces to the arithmetic mean in the frequency band of interest.

## 8 Scaling Laws with $A$ and $N$

Recall  $R = Af$  and  $\Delta R = Rh$ . While increasing  $A$  increases absolute  $\Delta R$ , fractional detectability depends on  $S_y$ . For design decisions the useful scaling is:

$$h_{\min} \sim \frac{\kappa}{\sqrt{N}} \sqrt{\frac{S_y(f_{\text{gw}})B}{T}}, \quad (17)$$

where  $\kappa$  accounts for windowing and mismatch factors. Increasing  $N$  or  $T$ , or reducing  $S_y$  and  $B$ , lowers detectable  $h_{\min}$ .

## 9 Hypothetical Ultra-High-Frequency Timebase

If a timebase at  $f_\gamma = 10^{18}\text{--}10^{20}$  Hz were available, then

$$\Delta f_\gamma = f_\gamma h. \quad (18)$$

For  $f_\gamma = 10^{19}$  Hz and  $h = 10^{-21}$ ,  $\Delta f_\gamma = 10^{-2}$  Hz. If such devices achieved a fractional PSD  $S_{y,\gamma}$  comparable to the best optical clocks, then detection becomes feasible at much smaller  $N$  and shorter  $T$ . This remains hypothetical but indicates frequency scaling benefits.

## 10 Statistical Correlation, Background Estimation and False-Alarm Control

The analysis pipeline should include:

1. Per-node noise characterization (estimate  $S_{y,i}(f)$  using long off-source runs).
2. Whitening by dividing by  $\sqrt{S_{y,i}(f)}$  in Fourier space.
3. Coherent stacking across nodes with inverse-PSD weights.
4. Background estimation by time-shifted coincidence trials (circular shifting node streams by offsets larger than causal windows) to compute false-alarm rates.
5. Multiply-trial correction using pre-defined band/session selections or FDR procedures.

## 11 Conclusion

The computational-amplification idea is internally consistent and offers a way to amplify absolute changes induced by GWs by using large aggregate computational throughput. Practical detection of astrophysical strains requires dramatic reductions in fractional noise or very large, well-synchronized ensembles in controlled environments (HPC/cloud). Nonetheless, this framework is scientifically useful for upper limits, noise studies, and citizen-science engagement. HPC-triggered narrowband experiments aligned with LVK alerts are the most promising near-term experimental path.

## References

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