AN EXPONENTIALLY CONVERGENT NONPOLYNOMIAL FINITE ELEMENT METHOD FOR TIME-HARMONIC SCATTERING FROM POLYGONS

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Abstract. In recent years nonpolynomial finite element methods have received increasing attention for the efficient solution of wave problems. As with their close cousin the method of particular solutions, high efficiency comes from using solutions to the Helmholtz equation as basis functions. We present and analyze such a method for the scattering of two-dimensional scalar waves from a polygonal domain that achieves exponential convergence purely by increasing the number of basis functions in each element. Key ingredients are the use of basis functions that capture the singularities at corners, and representing the scattered field towards infinity by a combination of fundamental solutions. The solution is obtained by minimizing a least-squares functional, which we discretize in such a way that a matrix least-squares problem is obtained. We give computable exponential bounds on the rate of convergence of the least-squares functional that are in very good agreement with the observed numerical convergence. Challenging numerical examples, including a nonconvex polygon with several corner singularities, and a cavity domain, are solved to around 10 digits of accuracy with a few seconds of CPU time. The examples are implemented concisely with MPSpack, a MATLAB toolbox for wave computations with nonpolynomial basis functions, developed by the authors. A code example is included.

Key words. scattering, polygons, Helmholtz equation, fundamental solutions

AMS subject classifications. 65N30, 35J05

1. Introduction. Nonpolynomial finite element methods for wave problems such as the PUFEM [3], the Least-Squares Finite Element Method [24] or the Ultra-Weak Variational Formulation [8] have recently received much attention in the mathematics and engineering community. The idea of these methods is to use Trefftz-type basis functions, i.e. exact solutions to the partial differential equation inside each element, then to solve for the coefficients which best satisfy the inter-element matching and/or boundary conditions. This better captures solution behavior than standard polynomial bases, and is closely related to the method of particular solutions or collocation methods [23]. Typical choices of basis functions are plane waves [26], Fourier-Bessel expansions [14, 27], and fundamental solutions of the Helmholtz equation [6, 16, 4].

The main advantage of nonpolynomial finite element methods over standard Galerkin methods is that they allow one to choose element sizes which are many wavelengths in diameter, while standard Galerkin methods with linear elements typically need of order 10 elements per wavelength to resolve a wave to any reasonable accuracy. Thus at high frequency the total number of degrees of freedom is much reduced, and due to their high-order (often exponential) convergence, this advantage is even more pronounced if high accuracy is required.

Great progress has also been made in recent years in the development of wavenumber independent boundary integral equation methods for the solution of scattering problems (see for example [7, 18, 10, 13]). An excellent overview of these new methods is given in [9].

In this paper we analyze a nonpolynomial finite element method for time-harmonic scattering from a sound-soft polygonally-shaped bounded obstacle $\Omega \subset \mathbb{R}^2$. That is, we are looking for the solution u of the problem

$$\Delta u + k^2 u = 0 \quad \text{in } \mathbb{R}^2 \backslash \overline{\Omega} \tag{1.1}$$

$$u = 0 \quad \text{on } \partial\Omega$$
 (1.2)

$$u = 0 \quad \text{on } \partial\Omega$$

$$\frac{\partial u_s}{\partial r} - iku_s = o(r^{-1/2}),$$
(1.3)

where $u = u_{inc} + u_s$ is the total field, u_{inc} is the incident wave and u_s the scattered field, and r is the radial coordinate. The wavenumber is $k=2\pi/\lambda$, where λ is the wavelength. The Sommerfeld radiation condition (1.3) is to be understood to hold uniformly in all directions.

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The main idea of the presented method is to use fractional order Fourier-Bessel functions at the corners of the polygon to match the asymptotic behavior of u there, and to represent the scattered field u_s towards infinity by fundamental solutions (i.e. effective sources), which automatically satisfy (1.3). The matching conditions between different elements are implemented using a least-squares finite element formulation. A related approach was proposed without analysis by Stojek in [27]. The main differences in our formulation are the choice of basis functions towards infinity and the setup of the discretized problem.

Least-squares finite element methods were also analyzed by Monk and Wang in [24]. Their analysis focused on Helmholtz problems on smooth domains with local absorbing boundary conditions. For such problems they also proved interior error estimates based on the least squares functional.

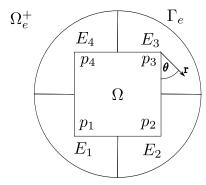
In this paper we choose each element as large as possible in such a way that each of the finite subdomains contains one corner of the polygon. The accuracy is then improved by increasing the number of basis functions in each subdomain. Hence, h-refinement is not necessary. This is similar to the domain decomposition method of Descloux and Tolley for the Laplace eigenvalue problem on polygonal domains [12].

We give exponential convergence bounds with a rate that can be numerically computed using techniques from conformal maps (see also [5]). The asymptotic convergence rate is wavenumber-independent. However, we observe that the minimum number of basis functions N required to enter into the regime of exponential convergence depends linearly on k, leading to a complexity of $O(k^3)$ for the fully discrete scheme due to the need for solving dense linear systems. But due to the fast exponential convergence, and the relatively small preprocessing effort, the overall computation time for this method is very attractive, and is competitive for low to medium frequencies with the wavenumber independent approach recently introduced by Chandler-Wilde and Langdon [10]. Furthermore, the method presented here works on convex and nonconvex polygons while the approach in [10] is restricted to convex polygons. We will demonstrate that high accuracy (up to 10 digits) is easily achievable.

Another advantage of the proposed approach compared to boundary element (or boundary integral) methods is that no quadrature of singular integrals close to the corners is necessary. Furthermore, the proposed method can be easily adapted to the sound-hard scattering case by a small change in the basis functions close to the corners. In contrast, combined integral formulations for sound-hard scattering involve the evaluation of hypersingular integral operators.

The paper is organised as follows. In Section 2 we present the formulation of the method. Section 3 describes how we discretize this into a linear system. The approximation of the scattered field u_s towards infinity is done using fundamental solutions approximations, which are presented in more detail in Section 4. In particular, a combined monopole-plus-dipole formulation is given that avoids problems with interior resonances. The convergence analysis in Section 5 depends on approximation theoretic results for Bessel functions and an analysis of the convergence of the fundamental solution approximations. A detailed analysis of the exponential convergence of fractional order Fourier-Bessel functions in cornered domains based on techniques from complex analysis was given by Betcke in [5]. Here, we only state the main convergence result of that paper adapted to our situation. Numerical examples are given in Section 6. All computations are implemented quite simply in MPSpack, a MATLAB toolbox for wave computations developed by the authors; a short example code to compute scattering from the square and plot Fig. 6.3 is given in Appendix A. Finally, since maintaining reasonably small coefficients is important for high accuracy, the numerical stability of the method is discussed in Section 7. We conclude in Section 8.

2. Formulation of the Method. Let Ω be a polygonal domain with boundary Γ having q corners at the points $p_1,\ldots,p_q\in\mathbb{R}^2$ and associated exterior angles $0<\pi/\alpha_j<2\pi,\ j=1,\ldots,q$. We define the artificial exterior boundary $\Gamma_e:=\partial\Omega_e$ to be an analytic Jordan curve whose simply connected interior Ω_e completely contains $\overline{\Omega}$, i.e. $\mathrm{dist}(\Gamma,\Gamma_e)>0$. We will also frequently need the exterior domain $\Omega_e^+=\mathbb{R}^2\backslash\overline{\Omega_e}$. Also, denote by $E:=\Omega_e\backslash\overline{\Omega}$ the part of Ω_e exterior to Ω . The geometry and notation are shown in Figure 2.1.



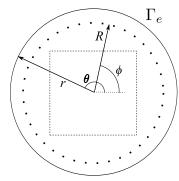


FIG. 2.1. Geometry of subdomains, illustrated for the square scatterer. For the exterior angles $\pi/\alpha_1, \ldots, \pi/\alpha_4$ at the corners p_1, \ldots, p_4 we have in this example that $\alpha_1 = \cdots = \alpha_4 = 2/3$. Left plot: The local coordinate systems of the Fourier-Bessel basis functions are oriented along the boundary lines adjacent to the corners of the polygon. Right plot: Fundamental solutions basis coordinate system, showing the charge points y_i lying on a circle.

We subdivide E into q simply connected subdomains E_i , i = 1, ..., q with boundaries $\Gamma_i = \partial E_i$ in such a way that the following conditions are satisfied:

- $E_i \cap E_j = \emptyset$ for all $i \neq j$,
- $\bigcup_i \overline{E_i} = \overline{E}$,
- $\Gamma_i \cap \partial \Omega$ consists of two straight lines which meet at the corner p_i and whose continuation towards infinity does not intersect E_i .¹

Remark 2.1. The requirement that every subdomain E_i contains a corner of Ω is not strictly necessary and is only used to simplify the description. An extension to the case of "free" elements that have no intersection with Ω is straightforward. This is used in the cavity example in Section 6.3.

In each subdomain E_i we approximate the total field u by a linear combination of Fourier-Bessel functions of the form

$$u(r,\theta) \approx \sum_{j=1}^{N_i} c_j^{(i)} J_{j\alpha_i}(kr) \sin j\alpha_i \theta.$$
 (2.1)

The local polar coordinate systems of the Fourier-Bessel functions are shown in Figure 2.1. This guarantees that the basis functions automatically satisfy the zero boundary conditions on Γ . We denote by V_i the finite dimensional linear space spanned by all Fourier-Bessel functions of the form (2.1) in the subdomain E_i .

The approximation by Fourier-Bessel functions is motivated by the corner behavior of Helmholtz solutons. One can show that in a wedge with interior angle π/α any Helmholtz solution u that satisfies zero Dirichlet boundary conditions on the sides of the wedge has close to the corner the absolutely convergent expansion

$$u(r,\theta) = \sum_{j=1}^{\infty} c_j J_{\alpha j}(kr) \sin \alpha j\theta$$

for certain coefficients $c_j \in \mathbb{C}$ [12]. If $\alpha \notin \mathbb{N}$ then any nontrivial u has a singularity at the corner of the wedge. Otherwise, if $\alpha \in \mathbb{N}$ then u can be analytically continued across the corner of the wedge [23].

In Ω_e^+ we approximate the scattered field u_s by a linear combination of fundamental solutions whose origins lie in Ω_e , as follows. Let $\Gamma_F \subset E$ be a closed analytic Jordan curve, and choose

¹The continuation condition is a technical condition needed for the convergence estimates on the finite subdomains (see [5, Figure 6] for an example violating this condition).

points $\mathbf{y}_j \in \Gamma_F$. Then our approximation, which satisfies the Helmholtz equation in Ω_e^+ , is

$$u_s(\mathbf{x}) \approx \sum_{j=1}^{N_e} c_j^{(e)} \left(\frac{i}{4} \frac{\partial}{\partial \nu(\mathbf{y}_j)} H_0^{(1)}(k|\mathbf{x} - \mathbf{y}_j|) - \frac{\eta}{4} H_0^{(1)}(k|\mathbf{x} - \mathbf{y}_j|) \right), \quad \mathbf{x} \in \Omega_e^+, \tag{2.2}$$

where $\nu(\mathbf{y}_j)$ is the outward-facing unit normal direction of the curve Γ_F at the point \mathbf{y}_j , $H_0^{(1)}$ the outgoing Hankel function of zeroth order, and $\eta \neq 0$ a real parameter. Typically, by a scaling argument we choose $\eta = k$. The formulation (2.2) can be interpreted as a discrete version of an integral equation of the first kind for approximating u (see Section 4). The finite dimensional space of all linear combinations of the fundamental solutions in Ω_e^+ of the form (2.2) is denoted by V_e .

Remark 2.2. One could alternatively approximate the scattered field u_s by a multipole expansion of the form

$$u_s(r,\theta) \approx \sum_{m=-N_c}^{N_c} \alpha_m H_m^{(1)}(kr) e^{im\theta}.$$

This was proposed by Stojek in [27]. The disadvantage of this expansion is that when Γ_e is anything other than a circle about the origin, severe numerical stability problems arise at large wavenumber due to the huge dynamic range of Hankel functions at large m. Fundamental solutions do not suffer from this problem, hence allow more flexibility (e.g. see Section 6.2).

Combining the above basis sets, the trial space V of the finite element method is the space of functions v such that $v_i := v|_{E_i} \in V_i$, $\forall i$, and $v_e := v|_{\Omega_e^+} \in V_e$. It is useful to express the number of basis functions in each subdomain as a multiplier of a common factor N. Let $N_i = n_i N$, $i = 1, \ldots, q$ and $N_e = n_e N$. The total dimension of the trial space is then

$$N_V = \left[n_e + \sum_{i=1}^q n_i \right] N =: n_V N.$$
 (2.3)

We now explain the least-squares finite element formulation as proposed by Stojek in [27] and Monk and Wang in [24]. Let Γ_{ij} be the internal boundary between two elements E_i and E_j . Let $\nu(\mathbf{x})$ be a unit normal to the curve Γ_{ij} , i < j, defined for almost every $\mathbf{x} \in \Gamma_{ij}$, pointing away from domain E_i . The jump of a function u defined on E_i and E_j is

$$[u](\mathbf{x}) := \lim_{\epsilon \to 0} u(\mathbf{x} + \epsilon \nu(\mathbf{x})) - u(\mathbf{x} - \epsilon \nu(\mathbf{x})), \quad \mathbf{x} \in \Gamma_{ij}$$

Denoting by $\partial_{\nu} := \nabla \cdot \nu$ the normal derivative with the sense explained above, and by $[\partial_{\nu} u]$ its jump, define the inter-element matching error functional

$$J(v) := \sum_{i < j} \int_{\Gamma_{ij}} |[\partial_{\nu} v]|^2 ds + k^2 |[v]|^2 ds + \sum_{i=1}^q \int_{\Gamma_i \cap \Gamma_e} |[\partial_{\nu} (\hat{u}_{inc} + v)]|^2 + k^2 |[\hat{u}_{inc} + v]|^2 ds, \quad (2.4)$$

where

$$\hat{u}_{inc}(\mathbf{x}) := \begin{cases} u_{inc}(\mathbf{x}), & \mathbf{x} \in \Omega_e^+ \\ 0, & \mathbf{x} \in \Omega_e \end{cases}$$
 (2.5)

Here we need the restriction \hat{u}_{inc} of the incoming wave u_{inc} to Ω_e^+ since inside Ω_e we directly approximate the full field u, while in Ω_e^+ we approximate the scattered field u_s .

The least-squares finite element approximation v_{LS} is now defined as

$$v_{LS} = \arg\min_{v \in V} J(v). \tag{2.6}$$

3. Implementation. In this section we give details of the numerical implementation of the least-squares formulation (2.6). Denote by $c = \left[c^{(1)}, \ldots, c^{(q)}, c^{(e)}\right]^T \in \mathbb{C}^{n_V N}$ the coefficient vector associated with all basis functions in the elements E_i , $i = 1, \ldots, q$ and Ω_e^+ . Since the basis functions in each element are linearly independent every $v \in V$ has a unique representation in terms of a vector $c \in \mathbb{C}^{n_V N}$. We can therefore reformulate (2.6) equivalently as

$$c_{LS} = \arg\min_{c \in \mathbb{C}^{N_V}} J(c).$$

For notational convenience we use the same name for the least-squares functional J(v) defined in (2.4) and its parameterized version in terms of a coefficient vector c.

We now derive a matrix representation of J(c). Consider first two elements E_i and E_j , within which basis functions are $g_1^{(i)}, \ldots, g_{N_i}^{(i)}$ and $g_1^{(j)}, \ldots, g_{N_j}^{(j)}$ respectively. We assume these functions and their derivatives are also defined on Γ_{ij} . Denote by ξ_{τ} , $\tau = 1, \ldots, m_{ij}$, quadrature points on Γ_{ij} with corresponding weights $\omega_{\tau} > 0$ appropriate for integration with respect to arc length (e.g. in Section 6 we use the Clenshaw-Curtis rule.) Then from Γ_{ij} the contribution to J(c) is

$$\begin{split} & \int_{\Gamma_{ij}} k^2 \left| \sum_{p=1}^{N_i} c_p^{(i)} g_p^{(i)}(s) - \sum_{p=1}^{N_j} c_p^{(j)} g_p^{(j)}(s) \right|^2 + \left| \sum_{p=1}^{N_i} c_p^{(i)} \partial_{\nu} g_p^{(i)}(s) - \sum_{p=1}^{N_j} c_p^{(j)} \partial_{\nu} g_p^{(j)}(s) \right|^2 ds \\ & \approx \left\| \begin{bmatrix} W_{ij} & \\ & W_{ij} \end{bmatrix} \begin{bmatrix} kA_i & -kA_j \\ \partial_{\nu} A_i & -\partial_{\nu} A_j \end{bmatrix} \begin{bmatrix} c^{(i)} \\ c^{(j)} \end{bmatrix} \right\|_2^2, \end{split}$$

where for each $i=1,\ldots,q$, A_i is the matrix with elements $(A_i)_{\tau p}=g_p^{(i)}(\xi_{\tau})$, and $(\partial_{\nu}A_i)_{\tau p}=\partial_{\nu}g_p^{(i)}(\xi_{\tau})$ is the matrix of normal derivatives. Quadrature weights now reside in the diagonal matrix W_{ij} with elements $(W_{ij})_{\tau\tau}=\omega_{\tau}^{1/2}$, $\tau=1,\ldots,m_{ij}$. On each outer boundary the contribution to J(c) is

$$\int_{\Gamma_{i}\cap\Gamma_{e}} k^{2} \left| \sum_{p=1}^{N_{i}} c_{p}^{(i)} g_{p}^{(i)}(s) - \sum_{p=1}^{N_{e}} c_{p}^{(e)} g_{p}(s) - u_{inc}(s) \right|$$

$$+ \left| \sum_{p=1}^{N_{i}} c_{p}^{(i)} \partial_{\nu} g_{p}^{(i)}(s) - \sum_{p=1}^{N_{e}} c_{p}^{(e)} \partial_{\nu} g_{p}(s) - \partial_{\nu} u_{inc}(s) \right|^{2} ds$$

$$\approx \left\| \begin{bmatrix} W_{ie} \\ W_{ie} \end{bmatrix} \left(\begin{bmatrix} kA_{i} & -kA_{e} \\ \partial_{\nu}A_{i} & -\partial_{\nu}A_{e} \end{bmatrix} \begin{bmatrix} c^{(i)} \\ c^{(j)} \end{bmatrix} - \begin{bmatrix} b_{i} \\ \tilde{b}_{i} \end{bmatrix} \right) \right\|_{2}^{2},$$

where g_{ℓ} is the ℓ th fundamental solutions basis function with source at \mathbf{y}_{ℓ} , and, analogous to the above, $(A_e)_{\tau p} = g_p(\xi_{\tau})$, $(\partial_{\nu} A_e)_{\tau p} = \partial_{\nu} g_p(\xi_{\tau})$, and quadrature points now live on $\Gamma_i \cap \Gamma_e$. The inhomogeneous vector is filled with the known values $(b_i)_{\tau} = u_{inc}(\xi_{\tau})$ and $(\tilde{b}_i)_{\tau} = \partial_{\nu} u_{inc}(\xi_{\tau})$ for $i = 1, \ldots, m_{ie}$.

Note that, by construction, all of the above quadratures involve smooth (in fact analytic) functions on the respective segments; no quadrature near the corner singularities is needed. However, the integrand is oscillatory on the wavelength scale, and this will inform the choice of the number of quadrature nodes.

Stacking up all contributions from the different interfaces we obtain

$$J(c) \approx ||W(Ac - b)||_2^2$$
.

The matrix A is built up from the subblocks A_i and $\partial_{\nu}A_i$ as described above and the right-handside vector b is zero apart from the contributions of the incident wave at the quadrature points on Γ_e . In [27, 24, 17] the authors discretize the variational derivative of J(v) to obtain a linear system of equations that they solve. This approach is equivalent to solving the normal equation

$$A^H W^2 A c = A^H W^2 b. (3.1)$$

However, as long as the problem size is not too large we can, and do, equivalently solve directly the least squares problem

$$\min_{c \in \mathbb{C}^{N_V}} \|W\left(Ac - b\right)\|_2 \tag{3.2}$$

using a dense least-squares solver, such as the backslash (mldivide) in MATLAB. This is numerically more stable than solving (3.1) if WA is ill-conditioned.

4. The Method of Fundamental Solutions for scattering problems. In this section we give a brief introduction to the Method of Fundamental Solutions (MFS) for Helmholtz problems in exterior domains and motivate the combined MFS formulation (2.2). Overviews of the MFS, its theory and applications are given in [6, 15]. The results of this section form the basis for the convergence analysis in Section 5.1.

Let D be a simply connected domain with analytic boundary ∂D . Let u_s be the unique solution [11] to the Helmholtz equation (1.1) in $\mathbb{R}^2 \setminus \overline{D}$ and the Sommerfeld condition (1.3) with boundary value data $u_s = f$ on ∂D . The general idea of the MFS is to choose charge points in D, with strengths (coefficients), whose combined field can well approximate u_s in $\mathbb{R}^2 \setminus \overline{D}$ as the number of charges grows. We focus on the special case where these charges lie on a closed curve $\Gamma_F \subset D$ some distance inside D. Then the MFS may be viewed as a quadrature approximation to a single-layer potential of the form

$$u_s(\mathbf{x}) \approx \int_{\Gamma_F} \frac{i}{4} H_0^{(1)}(k|\mathbf{x} - \mathbf{y}|) g(\mathbf{y}) ds_{\mathbf{y}}.$$
 (4.1)

However, the MFS is more general than merely a discretization of a standard boundary integral representation: depending on the smoothness of u_s (or singularities of its continuation into D), there might not exist any layer density g that can represent u_s exactly. Nevertheless there still may exist a sequence of densities in suitable spaces whose field converges to u_s . The discrete version of this statement is that, with the MFS, there exists a sequence of charge points and strengths whose field converges to u_s . If Γ_F is suitably chosen one can show that there are such sequences that give rapid exponential convergence in certain norms (see [4], and Section 5.1).

Unlike boundary integral methods which enable a second kind formulation, (4.1) is necessarily first kind. However, it has the advantages of easily provable convergence rate, of simplicity of evaluation arbitrarily close to ∂D , and that no singular quadrature schemes are required. Disadvantages of the MFS are that the right choice of Γ_F and the quadrature points \mathbf{y}_j on Γ_F can be difficult for complicated domains, and that the MFS typically leads to ill-conditioned linear systems. In our case the exterior boundary Γ_e is typically a very simple shape (circles or simple analytic curves), making it easy to automate the choice of Γ_F , and the effects of the ill-conditioning can be controlled in such a way that computing u to many digits of accuracy is still possible (see Section 7).

We now focus on the case that D is a disk with radius r and Γ_F is a circle with radius R < r (see Figure 2.1 which shows this geometry for the case $D = \Omega_e$ that we will use in Section 5.1). We identify the plane \mathbb{R}^2 with \mathbb{C} . Define the single-layer operator $S: L^2[0, 2\pi] \to L^2[0, 2\pi]$ by

$$(Sg)(\theta) = \int_0^{2\pi} \frac{i}{4} H_0^{(1)}(k|re^{i\theta} - Re^{i\phi}|)g(\phi)d\phi.$$

The task is to find a density g such that the boundary data is matched, that is, $Sg \approx f$. For the interior Helmholtz problem on the disk this was analyzed in [4]. Here, we focus on the eigenvalue expansion of S in order to reveal some stability properties for exterior problems and to lay the groundwork for Section 5.1.

Using Graf's addition formula [1, Eq. 9.1.79] we obtain

$$\frac{i}{4}H_0^{(1)}(k|re^{i\theta} - Re^{i\phi}|) = \sum_{m \in \mathbb{Z}} \frac{i}{4}H_m^{(1)}(kr)e^{-im\phi}J_m(kR)e^{im\theta}.$$

Defining the Fourier coefficient $\hat{g}(m)$ of a function $g \in L^2[0, 2\pi]$ by $\hat{g}(m) = \frac{1}{2\pi} \int_0^{2\pi} g(\phi) e^{-im\phi} d\phi$ it follows that

$$(Sg)(\theta) = \sum_{m \in \mathbb{Z}} \hat{s}(m)\hat{g}(m)e^{im\theta},$$

where the eigenfunctions of S are the complex exponentials, and the eigenvalues of S are

$$\hat{s}(m) = \frac{i\pi}{2} H_m^{(1)}(kr) J_m(kR).$$

If kR hits a zero of the Bessel function of order m then $\hat{s}(m) = 0$. This occurs if and only if k^2 is a Dirichlet eigenvalue of the disk of radius R. But the exterior Helmholtz problem has a unique solution for all k > 0, therefore we need a formulation that is not affected by interior resonances. We achieve this by a combined formulation analogous to those for boundary integral equations [11].

We define the modfied operator $S_{\eta}: L^{2}[0,2\pi] \to L^{2}[0,2\pi]$ by

$$(S_{\eta}g)(\theta) = \frac{i}{4} \int_0^{2\pi} \left(\frac{\partial}{\partial R} H_0^{(1)}(k|re^{i\theta} - Re^{i\phi}|) + i\eta H_0^{(1)}(k|re^{i\theta} - Re^{i\phi}|) \right) g(\phi)d\phi. \tag{4.2}$$

This combines single and double layer densities. Its eigenvalues $\hat{s}_{\eta}(m)$ are easily computed as

$$\hat{s}_{\eta}(m) = \frac{\pi}{2} H_m^{(1)}(kr) \left[i \frac{\partial}{\partial R} J_m(kR) - \eta J_m(kR) \right] = \frac{\pi}{2} H_m^{(1)}(kr) \left[\frac{ik}{2} \left(J_{m-1}(kR) - J_{m+1}(kR) \right) - \eta J_m(kR) \right]$$
(4.3)

LEMMA 4.1. Let $\eta \in \mathbb{R} \setminus \{0\}$. Then $\hat{s}_{\eta}(m) \neq 0$ for all $m \in \mathbb{Z}$.

Proof. For Hankel functions of the first kind it holds that $|H_m^{(1)}(z)| \neq 0$ for all $z \in \mathbb{C}$ [25, Chapter 7]. Then from the first expression in (4.3) we have $\hat{s}_{\eta}(m) = 0$ if and only if $\frac{\partial}{\partial R} J_m(kR) = 0$ and $J_m(kR) = 0$. But this is not possible since otherwise by Green's representation theorem the function $v(r,\theta) := J_m(kr)e^{im\theta}$ would be identically zero in the disk with radius R. \square

We now claim that the eigenvalues $\hat{s}_{\eta}(m)$ and their derivatives with respect to the outer disc radius r decay exponentially with a rate that depends on the radius ratio of R to r.

LEMMA 4.2. For $\eta \neq 0$ and $\epsilon > 0$ arbitrarily small there exist constants $c_s > 0$ and $C_s > 0$ such that for $m \in \mathbb{Z}$ both the following hold,

$$c_s \left(\frac{r}{R}\right)^{-|m|} \le |\hat{s}_{\eta}(m)| \le C_s \left(\frac{r}{R}\right)^{-|m|}, \tag{4.4}$$

$$\left| \frac{\partial}{\partial r} \hat{s}_{\eta}(m) \right| \le C_s \left[\left(\frac{r}{R} \right) - \epsilon \right]^{-|m|}, \tag{4.5}$$

where c_s and C_s depend on k, R and r, and C_s additionally depends on ϵ .

Proof. Consider the three terms in (4.3). Large-order asymptotics for Bessel functions [1, 9.3.1] yield

$$J_m(kR) \sim \frac{1}{\sqrt{2\pi m}} \left(\frac{ekR}{2m}\right)^m, \quad Y_m(kr) \sim -\sqrt{\frac{2}{\pi m}} \left(\frac{ekr}{2m}\right)^{-m}$$

for fixed z and $m \to \infty$. Since $H_m(z) = J_m(z) + iY_m(z)$ it follows that $H_m(z) \sim iY_m(z)$ and therefore

$$H_m^{(1)}(kr)J_m(kR) \sim -\frac{i}{\pi m} \left(\frac{R}{r}\right)^m,$$

$$H_m^{(1)}(kr)J_{m-1}(kR) \sim -\frac{2i}{ekR\pi} \left(\frac{R}{r}\right)^m,$$

$$H_m^{(1)}(kr)J_{m+1}(kR) \sim -\frac{ekRi}{2\pi} \frac{1}{m^2} \left(\frac{R}{r}\right)^m.$$

Inserting these into (4.3) we get

$$\hat{s}_{\eta}(m) \sim \frac{1}{2eR} \left(\frac{R}{r}\right)^m - \frac{Rek^2}{8m^2} \left(\frac{R}{r}\right)^m + \frac{i\eta}{2m} \left(\frac{R}{r}\right)^m \sim \frac{1}{2eR} \left(\frac{R}{r}\right)^m.$$

Together with the reflection laws $J_{-m}(z) = (-1)^m J_m(z)$ and $H_{-m}^{(1)}(z) = (-1)^m H_m^{(1)}(z)$ the upper bound in (4.4) follows. Furthermore, by Lemma 4.1 there exists $c_s > 0$ such that also the lower bound in (4.4) is valid. We apply similar estimates to the expression

$$\frac{\partial}{\partial r}\hat{s}_{\eta}(m) = \frac{k\pi}{4} \left[H_{m-1}^{(1)}(kr) - H_{m+1}^{(1)}(kr) \right] \left[\frac{ik}{2} \left(J_{m-1}(kR) - J_{m+1}(kR) \right) - \eta J_m(kR) \right]. \tag{4.6}$$

to obtain

$$\frac{\partial}{\partial r}\hat{s}_{\eta}(m) \sim Cm \left(\frac{r}{R}\right)^{-|m|},$$

for some constant $C \in \mathbb{C}$. The dominant contribution is $H_{m+1}^{(1)}(kr)J_{m-1}(kR)$. Absorbing the algebraic factor m into the exponential bound and choosing C_s sufficiently large yields (4.5). \square

The MFS is now obtained by applying to the above layer potential formulation the discrete sources ansatz $g(\phi) = \sum_{j=1}^{N} c_j \delta(\phi - \phi_j)$, where $\phi_j = \frac{2\pi j}{N}$, j = 1, ..., N. It follows that

$$(S_{\eta}g)(\theta) = \sum_{j=1}^{N} c_{j} \left[\frac{i}{4} \frac{\partial}{\partial R} H_{0}^{(1)}(k|re^{i\theta} - Re^{i\phi_{j}}|) - \frac{\eta}{4} H_{0}^{(1)}(k|re^{i\theta} - Re^{i\phi_{j}}|) \right]. \tag{4.7}$$

Note $g \notin L^2[0, 2\pi]$, but S_η has a smooth bounded kernel so is well defined with g merely integrable. We will also need the Fourier series for this choice of g, which we note converges only distributionally. Denote by \hat{c} the discrete Fourier transform of the coefficient vector $c = [c_1, \ldots, c_N]$ defined by

$$\hat{c}_s = \frac{1}{N} \sum_{i=1}^{N} c_j e^{-is\phi_j}, \quad -\frac{N}{2} < s \le \frac{N}{2}.$$

Then applying our ansatz to the definition of the Fourier coefficients,

$$\hat{g}(m) = \frac{1}{2\pi} \sum_{j=1}^{N} c_j e^{im\phi_j} = \frac{N}{2\pi} \hat{c}_{(m \mod N)}, \tag{4.8}$$

where $m \mod N$ denotes the unique integer lying in the range $-N/2+1,\ldots,N/2$ which differs from m by an integer multiple of N. This 'wrapping' of the vector \hat{c} is due to the delta-comb in g.

The above (4.7) is a special case of the general monopole-plus-dipole formulation (2.2), with Γ_F circular and the evaluation restricted to the concentric circle ∂D . The basis functions in (2.2), viewed as functions on $\mathbb{R}^2 \backslash D$, automatically satisfy the Sommerfeld radiation conditions. Thus the only remaining task to solve the boundary value problem laid out at the start of this section would be to choose the coefficients c_j to match the boundary data f, e.g. by collocation or a least-squares method on ∂D . We remind the reader that in this work we instead couple the MFS (with the choices $D = \Omega_e$ and $N = N_e$) to other basis representations in E.

5. Convergence analysis. Let u be the exact solution of (1.1)-(1.3) with given u_{inc} . Since u and its derivative are continuous across element boundaries, we may study the convergence of approximations to u independently in each subdomain. In particular, we will make use of the rate of approximation by Fourier-Bessel functions in each subdomain E_i , and the rate of approximation on the boundary Γ_e by fundamental solutions. Convergence to u then automatically leads to a

minimization of the jump on the interfaces, as we now show. Using the triangle inequality and estimating the boundary L^2 norm by the L^{∞} norm over the whole domain, we estimate (2.4) by

$$J(v) = \sum_{i < j} \int_{\Gamma_{ij}} |[\partial_{\nu} v - \partial_{\nu} u]|^{2} ds + k^{2} |[v - u]|^{2} ds$$

$$+ \sum_{i=1}^{q} \int_{\Gamma_{i} \cap \Gamma_{e}} |[\partial_{\nu} (\hat{u}_{inc} + v - u)]|^{2} + k^{2} |[\hat{u}_{inc} + v - u]|^{2} ds$$

$$\leq C_{1} \sum_{i} \left\{ k^{2} ||u_{s} - v_{e}||_{L^{2}(\Gamma_{i} \cap \Gamma_{e})}^{2} + ||\partial_{\nu} u_{s} - \partial_{\nu} v_{e}||_{L^{2}(\Gamma_{i} \cap \Gamma_{e})}^{2} \right\}$$

$$+ C_{2} \left\{ k^{2} \sum_{i} ||u - v_{i}||_{L^{\infty}(E_{i})}^{2} + \sum_{i < j} ||\partial_{\nu} u - \partial_{\nu} v_{i}||_{L^{\infty}(\Gamma_{ij})}^{2} \right\},$$

$$(5.1)$$

where $v \in V$ (recalling the definitions of v_i , v_e from Section 2), and C_1, C_2 are mesh-dependent constants.

The analysis of the rate of convergence of approximating u by $v_i \in V_i$ in each subdomain E_i is based on transforming the approximation problem involving Bessel functions into a polynomial approximation problem in the complex plane. For this it is useful to work in the L^{∞} norm. On Γ_e the rate of convergence of the fundamental solutions will be analyzed using Fourier series estimates, for which the natural space is L^2 .

The convergence analysis for approximating solutions of $\Delta u + \lambda u = 0$ with fractional Bessel functions on domains with one corner singularity and zero Dirichlet boundary conditions at the wedges forming the singular corner was given by Betcke in [5]. The only notable new feature here is that now complex rather than real functions are approximated. This can easily be achieved by splitting up the approximation problem on the subdomains E_i into separate problems for the approximation of the real and imaginary parts of u.

The following theorem summarizes the results of [5, Theorem 5.2 and Lemma 5.5] adapted to our notation in this paper. We refer to [5, Section 5] for the proof.

THEOREM 5.1. For every element E_i , $i=1,\ldots,q$ the following statement holds. There exists $\rho_i > 1$ such that for every $1 < \tau < \rho_i$ and each interface Γ_{ij} between E_i and a neighboring subdomain E_j it holds that

$$\min_{v \in V_i} \left(\|\nabla u - \nabla v\|_{L^{\infty}(\Gamma_{ij})}^2 + k^2 \|u - v\|_{L^{\infty}(E_i)}^2 \right) = O(\tau^{-2N_i})$$

as $N_i \to \infty$.

The rates ρ_i may be computed; they are the conformal distance of the nearest singularity in u to (a conformal map of) the domain E_i [5].

To estimate the convergence on Γ_e of the fundamental solutions approximation to the scattered field u_s we consider (as in Section 4) only the case of concentric circles.² The source points for the fundamental solutions are given by $\mathbf{y}_j = Re^{i\phi_j}$, $j=1,\ldots,N_e$ with $\phi_j = \frac{2\pi j}{N_e}$, and the exterior circle Γ_e has radius r. We impose $\max_i |p_i| < R < r$. The proof of the following theorem is given below in Section 5.1.

Theorem 5.2. Let $\epsilon > 0$ and $\tilde{\epsilon} > 0$ be arbitrarily small. Define $\rho := \min_i \frac{r}{|p_i|}$ and $\rho_{\epsilon} := \rho - \epsilon$. Let ν be the outward normal direction at the disk with radius r and assume that N_e is even. Then

$$\min_{v \in V_e} \left(\|\partial_{\nu} u_s - \partial_{\nu} v\|_{L^2(\Gamma_e)}^2 + k^2 \|u_s - v\|_{L^2(\Gamma_e)}^2 \right) = \begin{cases} O\left(\left(\frac{r}{R} - \tilde{\epsilon}\right)^{-2N_e}\right), & \frac{r}{R} < \rho_{\tilde{\epsilon}}^{\frac{1}{2}} \\ O\left(\rho_{\epsilon}^{-N_e}\right), & \frac{r}{R} > \rho_{\tilde{\epsilon}}^{\frac{1}{2}} \end{cases} \tag{5.2}$$

²The convergence analysis of the MFS for integral equations with periodic kernels on more general domains has recently been considered by Kangro [22]. For the understanding of the method presented in this paper the case of the MFS in the exterior of a disk is sufficient for many applications and can be treated completely using arguments from Fourier analysis.

as $N_e \to \infty$.

Remark 5.3. If we only consider the rate of convergence for approximating the boundary value of u_s by $v \in V_e$ and not the normal derivatives then $\tilde{\epsilon} = 0$ is possible.

Combining (5.1), Theorem 5.1 and Theorem 5.2 we obtain immediately the following.

THEOREM 5.4. Let ρ_1, \ldots, ρ_q be the measures of the exponential convergence rates given in Theorem 5.1, $\rho_e := \min\left\{\left(\frac{r}{R}\right)^2, \min_i \frac{r}{|p_i|}\right\}$ and let n_1, \ldots, n_q and n_e be the multipliers for the number of basis functions as defined in (2.3). Assume that n_e is even and let $\omega > 1$ be defined by

$$\omega^2 := \min\{\rho_1^{2n_1}, \dots, \rho_q^{2n_q}, \rho_e^{n_e}\}.$$

Then

$$\min_{v \in V} J(v) = O\left((\omega - \epsilon)^{-2N}\right)$$

for any $\epsilon > 0$ arbitrarily small.

Remark 5.5. An ideal choice of the multipliers n_1, \ldots, n_q and n_e is such that

$$\rho_1^{2n_1} \approx \cdots \approx \rho_q^{2n_q} \approx \rho_e^{n_e}.$$

This choice keeps the number of basis functions required for a certain accuracy to a minimum. Remark 5.6. It is an open question whether there exists C > 0 such that

$$||u - v||_{L^2(E)}^2 \le CJ(v). \tag{5.3}$$

In [24, Theorem 3.1] such an estimate was proved for the case that Ω is a sufficiently smooth domain and Γ_e has the impedance condition $\partial_{\nu}u - iku = 0$. The crucial difference here is that we use fundamental solutions approximations instead of local absorbing boundary conditions. Similar estimates have also been investigated by Hiptmair, Moiola and Perugia in the context of Plane Wave Discontinuous Galerkin Methods [21]. In Section 6 we present numerical examples, which support the conjecture (5.3).

5.1. Convergence of the MFS approximations. In this section we give a proof of Theorem 5.2. The proof proceeds along similar lines as the proof of [4, Theorem 3]. We proceed by finding a specific function $v \in V_e$, for which we show that (5.2) holds. In the following C > 0 will always denote an unspecified constant that may change in the course of the derivation and that depends on k, R, r and ϵ but not on N_e .

Let $t[v] := ||u_s - v||_{L^2(\Gamma_e)}$ for $v \in V_e$. The multipole (Laurent-type) representation of u_s is

$$u_s(r,\theta) = \sum_{m \in \mathbb{Z}} \alpha_m H_m^{(1)}(kr) e^{im\theta}$$

for $r > \max_i |p_i|$ and $\theta \in [0, 2\pi]$. It follows immediately that the Fourier coefficients $\hat{u}_s(m)$ of the angular function $u_s(r, \cdot)$ are

$$\hat{u}_s(m) = \alpha_m H_m^{(1)}(kr).$$

The following lemma gives a bound on the decay of $\hat{u}_s(m)$ as $m \to \infty$.

LEMMA 5.7. Let $\rho := \min_{i \mid p_i \mid} \text{ and } \rho_{\epsilon} := \rho - \epsilon$. Then

$$|\hat{u}_s(m)| \le C\rho_{\epsilon}^{-|m|} \tag{5.4}$$

for any arbitrarily small $\epsilon > 0$, where C depends on k, R, r and ϵ .

Proof. Let $f(t) := u_s(r, 2\pi t)$ for $t \in [0, 1]$. We need to show that f can be analytically continued as a holomorphic function to the strip $t + i\tau$ for $|\tau| < \frac{1}{2\pi} \log \rho$. Denote by $\tilde{p} := \arg \max_i |p_i|$ the corner of $\partial \Omega$ with largest absolute value. Reparameterize the scattered field $u_s(x,y)$ in the cartesian (x,y) coordinates as $U_s(z,\overline{z}) := u_s(x,y)$, where z = x + iy. Let $\Omega_{\tilde{p}} := u_s(x,y)$

 $\{z: |\tilde{p}| < |z| < \frac{r^2}{|\tilde{p}|}, \text{ Im}\{z\} > 0\}$ be a half annulus and define $\Omega_{\tilde{p}}^* := \{z: \overline{z} \in \Omega_{\tilde{p}}\}$. By a result from Vekua [19, 28] U_s can be analytically continued as a holomorphic function of two independent complex variables z and z^* into the whole of $\Omega_{\tilde{p}} \times \Omega_{\tilde{p}}^*$, such that $U_s(z,z^*) = u_s(x,y)$ if and only if $z^* = \overline{z}$. From the definition of f it follows that $f(t) = U_s(re^{i2\pi t}, re^{-i2\pi t})$. Hence, by the analyticity of U_s we have that $f(t+i\tau) = U_s(re^{-2\pi\tau+i2\pi t}, re^{2\pi\tau-i2\pi t})$ is analytic for 0 < t < 1/2 and $-\frac{1}{2\pi}\log\rho < \tau < \frac{1}{2\pi}\log\rho$. By rotating the half annulus and repeating the argument the analyticity in this strip can be extended to the whole interval $t \in [0,1]$. The estimate (5.4) now follows from the decay of Fourier coefficients of analytic functions (see for example [20, §13.2]).

The Fourier series representation of $v \in V_e$, associated with density g, restricted to the circle of radius r is given by (see Section 4)

$$\hat{v}(m) = \widehat{S_{\eta}g}(m) = \hat{s}_{\eta}(m)\hat{g}(m),$$

where $g(\theta) = \sum_{j=1}^{N_e} c_j \delta(\phi - \phi_j)$ and $\hat{g}(m) = \frac{N_e}{2\pi} \hat{c}_{(m \mod N_e)}$. By reparamaterizing the L^2 integral over Γ_e and using Parseval's identity it follows that

$$t[v]^{2} = 2\pi r \sum_{m \in \mathbb{Z}} |\hat{u}_{s}(m) - \hat{v}(m)|^{2}.$$
(5.5)

We now take a specific $v \in V_e$ by choosing the MFS coefficients of v such that for all $-N_e/2$ $m \leq N_e/2$ we have,

$$\hat{g}(m) = \frac{\hat{u}_s(m)}{\hat{s}_{\eta}(m)} = \frac{\alpha_m H_m^{(1)}(kr)}{\frac{\pi}{2} H_m^{(1)}(kr) \left[\frac{ik}{2} \left(J_{m-1}(kR) - J_{m+1}(kR)\right) - \eta J_m(kR)\right]}$$
(5.6)

Then the terms involving $-N_e/2 < m \le N_e/2$ cancel out in (5.5) and we can estimate

$$t[v]^2 \le C \left(\sum_{m \not\in [-\frac{N_e}{2}+1,\dots,\frac{N_e}{2}]} |\hat{u}_s(m)|^2 + \sum_{m \not\in [-\frac{N_e}{2}+1,\dots,\frac{N_e}{2}]} |\hat{s}_{\eta}(m)\hat{g}(m)|^2 \right) =: C(E_u^2 + E_s^2)$$

Using (5.4) we can now estimate E_u^2 as

$$E_u^2 \le C \sum_{m \notin [-\frac{N_e}{2} + 1, \dots, \frac{N_e}{2}]} \rho_{\epsilon}^{-2|m|} \le C \rho_{\epsilon}^{-N_e}.$$

$$(5.7)$$

In order to estimate E_s^2 we rewrite it as

$$E_s^2 = \sum_{n = -\frac{N_e}{2} + 1}^{\frac{N_e}{2}} \hat{g}(n)^2 \sum_{b \in \mathbb{Z} \setminus \{0\}} \hat{s}_{\eta}(bN_e + n)^2.$$

Bounding the inner sum and substituting $|\hat{g}(n)| = \left|\frac{\hat{u}_s(n)}{\hat{s}_{\eta}(n)}\right| \leq C \left(\frac{r}{R\rho_{\epsilon}}\right)^{|n|}$ gives

$$E_s^2 \le C \sum_{n = -\frac{N_e}{2} + 1}^{\frac{N_e}{2}} |\hat{g}(n)|^2 \left(\frac{r}{R}\right)^{-2N_e + 2|n|} \le C \left(\frac{r}{R}\right)^{-2N_e} \sum_{n = -\frac{N_e}{2} + 1}^{\frac{N_e}{2}} \left(\frac{r^2}{R^2 \rho_{\epsilon}}\right)^{2|n|}. \tag{5.8}$$

If $\rho_{\epsilon} > \frac{r^2}{R^2}$ it follows from (5.8) that $E_s^2 \leq C \left(\frac{r}{R}\right)^{-2N_e}$. This decays slower than the estimate in (5.7). Hence, we have $t[v]^2 \leq C \left(\frac{r}{R}\right)^{-2N_e}$. If $\rho_{\epsilon} < \frac{r^2}{R^2}$ it follows from (5.8) that

$$E_s^2 \le C \left(\frac{r}{R}\right)^{-2N_e} \left(\frac{r^2}{R^2 \rho_\epsilon}\right)^{N_e} = C \rho_\epsilon^{-N_e},$$

which has the same exponential rate of decay as (5.7) and therefore $t[v]^2 \leq C\rho_{\epsilon}^{-N_e}$ in this case.

We now show that up to an arbitrarily small correction in rate $\tilde{\epsilon} > 0$ the error $t'[v] := \|\partial_{\nu}u_s - \partial_{\nu}v\|_{L^2(\Gamma_e)}$ has the same exponential bounds as t[v]. Similar to (5.5) we have

$$t'[v] = 2\pi r \sum_{m \in \mathbb{Z}} |\widehat{\partial_{\nu} u_s}(m) - \widehat{\partial_{\nu} v}(m)|^2, \tag{5.9}$$

where

$$\widehat{\partial_{\nu} u_s}(m) = \frac{\partial}{\partial r} \hat{u}_s(m)$$

and

$$\widehat{\partial_{\nu}v}(m) = \frac{\partial}{\partial r}\widehat{v}(m) = \left(\frac{\partial}{\partial r}\widehat{s}_{\mu}(m)\right)\widehat{g}(m).$$

The second equality follows from the fact that $\hat{g}(m)$ is independent of r (see (5.6)). Furthermore, from (5.6) it follows that

$$\frac{\frac{\partial}{\partial r}\hat{u}_s(m)}{\frac{\partial}{\partial r}\hat{s}_{\mu}(m)} = \frac{\hat{u}_s(m)}{\hat{s}_{\eta}(m)} = \hat{g}(m)$$

for $-N/2 < m \le N/2$. Hence, the Fourier coefficients involving $-N/2 < m \le N/2$ also cancel out for t'[v]. By Lemma 4.2, $\frac{\partial}{\partial r} \hat{s}_{\eta}(m)$ has up to an arbitrarily small rate correction $\tilde{\epsilon}$ the same exponential upper bound as $\hat{s}_{\eta}(m)$. Also, since $\frac{\partial}{\partial r} u_s$ has its only possible singularities at the corners p_i its Fourier coefficients $\frac{\partial}{\partial r} \hat{u}_s(m)$ have the same exponential bound (5.4) as those of u_s . We can therefore apply the same arguments to estimate t'[v] as we did for t[v] and arrive (up to $\tilde{\epsilon}$ coming from the bound on $\frac{\partial}{\partial r} \hat{s}_{\eta}(m)$) at the same exponential bounds as for t[v], which concludes the proof of Theorem 5.2.

- **6. Numerical examples.** In this section we demonstrate the numerical properties of the method and present computational results for scattering on a square, a snowflake domain and a cavity. All computations were performed under Matlab 2009a on a dual processor workstation with two 2.8 GHz Intel Xeon quad-core processors and 64GB RAM. However, MATLAB's multithreading rarely utilized more than one CPU core. Hence, the timings will be comparable on a standard desktop machine.
- **6.1. Scattering from a square.** Let Ω be the unit square given by $\Omega = (-0.5, 0.5)^2$. We use a domain decomposition as shown in Figure 2.1. Γ_e is a circle with radius r and the source points for the fundamental solutions are given by $\mathbf{y}_j = Re^{i\phi_j}$, $\phi_j = \frac{2\pi j}{N_e}$, $j = 1, \ldots, N_e$. We found the fundamental solutions basis size $N_e := 2N$ was sufficient, where N is the number of Fourier-Bessel sine functions in each element E_i . For the fundamental solutions we use the combined formulation (2.2) with $\eta = k$. For quadrature we used the Clenshaw-Curtis rule with a number of nodes chosen high enough that further increases had negligible effect; typically this number is a small multiple of N per interface Γ_{ij} . (Code for this setup is given in the Appendix).

Rate of convergence. In Figure 6.1 we show the rate of convergence of $t[v_{LS}] := J(v_{LS})^{1/2}$, which measures the norm of the jumps in function values and normal derivatives on the interfaces. We fix R = 0.8 and the wavenumber k = 1. The left plot shows the case r = 0.9 and the right plot r = 1.3. The measured convergence (solid-dotted line) for the latter is much faster than for the former. In both cases we have also plotted the asymptotic bound ω^{-N} from Theorem 5.4. The aymptotic estimates for the convergence on the finite elements E_1, \ldots, E_4 can be computed using numerical conformal mapping techniques (see [5] for more details). The convergence of the MFS approximations is directly given by Theorem 5.2.

What is the optimal value of r? If r is small then the singular corners of the square have a large relative distance to the neighboring elements, suggesting a fast convergence on these elements. However, the rate of convergence for the MFS approximations will be slow since the radius r is

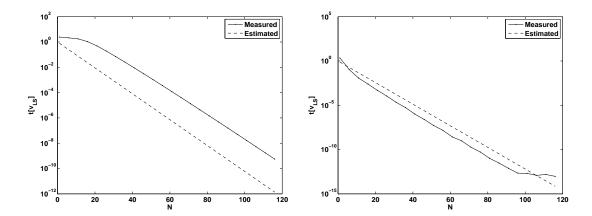


Fig. 6.1. Left plot: Convergence of $t[v_{LS}]$ (solid-dotted line) for growing N for the case r=0.9 compared to the estimated rate of convergence from Theorem 5.4 (dashed line). Right plot: Same as left plot but now r=1.3. The convergence is much faster for this choice of r.

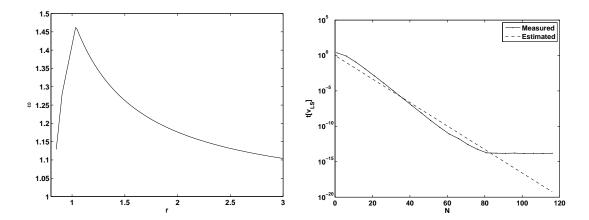


Fig. 6.2. Left plot: Asymptotic convergence factor ω in dependence of r. Large values of ω indicate faster convergence of $t[v_{LS}]$. Right plot: Measured (solid-dotted line) and estimated convergence (dashed line) for the optimal radius r=1.036.

close to the radius of the corners of the polygon. If r is large then the MFS approximations converge fast but the relative distance of a singular corner to a neighboring element E_i is very small leading to slow convergence on the finite elements. This suggests that there is an optimal radius $r = r_{opt}$, which is achieved when the asymptotic rate of convergence of the MFS is identical to the asymptotic rate of convergence on the finite elements.

In the left plot of Figure 6.2 we compute (using conformal mapping) the asymptotic exponential convergence factor ω from Theorem 5.4 as a function of the radius r of the outer circle Γ_e . The optimum asymptotic rate is achieved for $r \approx 1.036$. In that case the rate of convergence on the finite elements E_j and on Ω_e^+ is almost identical, with a rate of approximately 1.465^{-N} . The right plot of Figure 6.2 shows the convergence of $t[v_{LS}]$ in this case. For N=80 the value of $t[v_{LS}]$ is close to machine precision.

Sound-hard boundary conditions. In this paper we have analyzed the case of sound-soft boundary conditions for u, that is u=0 on $\partial\Omega$. However, the scheme presented in this paper can be trivially extended to sound-hard boundary conditions, that is $\partial_{\nu}u=0$, where ∂_{ν} denotes the outward pointing normal derivative on $\partial\Omega$. For this we switch from Fourier-Bessel sine function

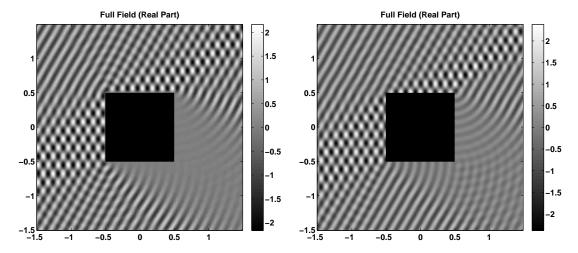


Fig. 6.3. Real part of the full field for sound-soft scattering on the square (left plot) and sound-hard scattering on the square (right plot). In both cases k = 50 and the incident wave is a plane wave with angle $-\frac{\pi}{6}$.

approximations to Fourier-Bessel cosine function approximations in each E_i , replacing (2.1) by

$$u(r,\theta) \approx \sum_{j=0}^{N_i} c_j^{(i)} J_{j\alpha_i}(kr) \cos j\alpha_i \theta.$$

It is simple to check that these functions satisfy the homogeneous boundary conditions for the normal derivative. The fundamental solutions basis functions in Ω_e^+ do not change. This is a great advantage compared to combined integral equation formulations, which usually involve the approximation of a hypersingular integral operator for the sound-hard scattering case.

Timing results. In Figure 6.3 we show the real part of the full field for the sound-soft scattering case (left plot) and the sound-hard scattering case (right plot) for k = 50, i.e. about 8 wavelengths on a side. The incident field is a plane wave with direction $-\frac{\pi}{6}$.

The solution time was approximately 11 seconds in both cases. This includes the setup of the problem and the time for solving the least squares problem $\min_c \|W(Ac-b)\|_2$, where A is of dimension 4816×600 (sound-soft case), or 4816×604 (sound-hard case), corresponding to N=100 and $m_{ij}=200$ Clenshaw-Curtis quadrature points on each interface Γ_{ij} . In the sound-soft case we have $t[v_{LS}] \approx 1.3 \times 10^{-10}$ and in the sound-hard case $t[v_{LS}] \approx 9.2 \times 10^{-11}$. Creating the solution plots takes longer, since the solution needs to be evaluated on a large number of grid points.

6.2. Scattering from a snowflake domain. We now show results for a snowflake domain, which is nonconvex but star-shaped. The domain and the corresponding decomposition are shown in the top two plots of Figure 6.4. Note that the external boundary Γ_E is not a circle but an equipotential line of the exterior (Laplace) Green's function of the domain. Although we have given the analysis only for Γ_E being a circle, we demonstrate in this example that domain-adapted analytic curves also work well.

In the bottom two plots of Figure 6.4 we show the scattered field (left) and the full field (right) for sound-soft scattering by an incident plane wave with angle $-\frac{\pi}{3}$ and wavenumber k=100. Setting up and solving the dense least-squares problem takes 15 seconds. The error $t[v_{LS}]$ is approximately $4 \cdot 10^{-9}$.

In Table 6.1 we show results for different wavenumbers. N was chosen such that the error $t[v_{LS}]$ is roughly of magnitude 10^{-7} to 10^{-9} . The number of fundamental solutions used was $N_e = 3.3N$. m and N_V denote the dimension of the matrix $A \in \mathbb{C}^{m \times N_V}$ in the least-squares problem $\min_c ||W(Ac - b)||_2$. The times are the total time for setting up the matrix and solving the least-squares problem. Up to around k = 200 the solution time is excellent. For higher k the $O(k^3)$ dependence of the solution time becomes a problem. Also at this high wavenumber, we have

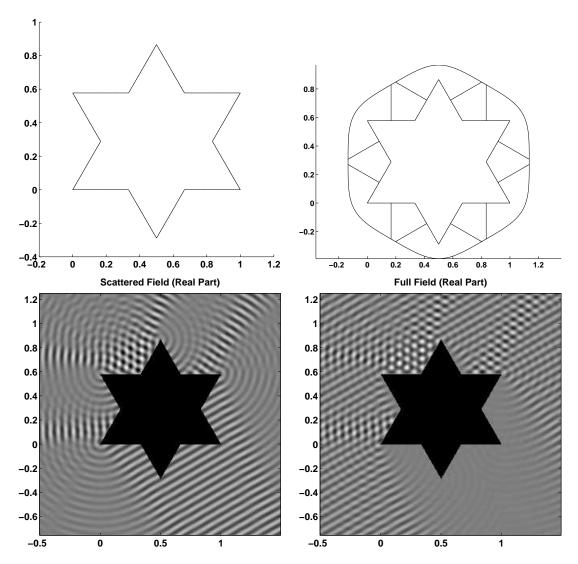


Fig. 6.4. Top plots: A snowflake shaped domain and the corresponding domain decomposition. Bottom plots: The scattered (left) and full field (right) for sound-soft scattering on the snowflake domain with k=100

k	N	m	N_V	$t[v_{LS}]$	time
50	70	1980	1071	$3 \cdot 10^{-8}$	8 s
100	90	2460	1377	$4 \cdot 10^{-9}$	$15 \mathrm{\ s}$
200	130	3660	1989	$5 \cdot 10^{-9}$	$44 \mathrm{\ s}$
500	260	8700	3978	$2 \cdot 10^{-7}$	$7 \mathrm{m}$
Table 6.1					

Results for different wavenumbers for sound-soft scattering on the snowflake domain.

noticed that the best achievable $t[v_{LS}]$ seems to be around 10^{-7} . By contrast, Figure 6.5 shows that, at lower k, accuracies approaching machine precision are achievable. This figure also shows that a small increase in the number of basis functions leads to a large decrease of $t[v_{LS}]$, once the number of basis functions is high enough to get into the regime of exponential convergence. From our experiments the number of basis functions required to enter this regime grows approximately linearly with k, as supported by Figure 6.5. Once this regime is entered, the convergence rate appears to be independent of k, as suggested by the analysis in Section 5.

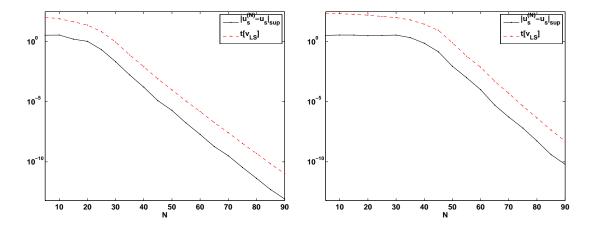


Fig. 6.5. Convergence of $t[v_{LS}]$ (dashed line) for k=50 (left plot) and k=100 (right plot). The plots also show an approximation to the convergence versus N of the L^{∞} norm of the approximate scattered field $u_s^{(N)}$ to the exact scattered field u_s (solid line).

In Remark 5.6 we stated the conjecture that $\|u-v\|_{L^2(E)}^2 \leq CJ(v)$ for some C>0. To support this conjecture we also plotted in Figure 6.5 the convergence of the scattered field u_s . This was done by first computing an approximate solution u_s for N=100 and then approximating the L^{∞} error $\|u_s^{(N)} - u_s\|_{L^{\infty}(\mathbb{R}^2 \setminus \Omega)}$, where $u_s^{(N)}$ is the approximate scattered field in step N, by evaluating it on the same grid that was used to plot the solution in Figure 6.4. Notice that the L^{∞} errors are always at least a factor of 50 smaller than $t[v_{LS}]$. Therefore this plot supports the stronger conjecture

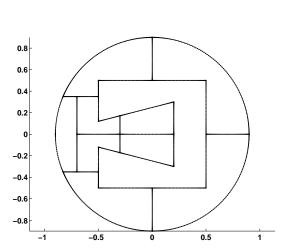
$$||u - \hat{u}_{inc} - v||_{L^{\infty}(\mathbb{R}^2 \setminus \Omega)}^2 \le CJ(v),$$

where \hat{u}_{inc} was defined in (2.5) and C > 0 is a constant, which is, however, expected to be large if exterior resonance effects occur.

6.3. A cavity domain. To conclude this section we demonstrate the example of scattering from a non-star-shaped cavity domain, which allows strong resonances to occur. The domain decomposition is shown in the left plot of Figure 6.6. Note that the left-most element does not share a corner with the cavity domain; in this element we use a regular Fourier-Bessel expansion, that is, (2.1) with $\alpha_i = 1$ and both sin and cos terms, whose origin is chosen inside the element. The reason is to reduce the size of the elements at the corners of the entrance to the cavity, which increases the relative distance to neighboring singularities and thereby increases their rate of exponential convergence. Even though an extra basis set is required, at high accuracy the total basis size is reduced, hence efficiency improved. The previous convergence analysis would easily extend to this case.

In the right plot of Figure 6.6 we show the solution of sound-soft scattering by an incoming plane wave with direction $-\pi/10$ for k=100, i.e. 16 wavelengths on the longest edge. The setup and solution of the associated least-squares problem takes 14 seconds (matrix size 1938×1521) for an error of $t[v_{LS}] \approx 3 \cdot 10^{-8}$. To evaluate and plot the solution on $6 \cdot 10^4$ points (a square grid of spacing 0.01) takes a further 46 seconds.

7. Numerical stability of the method. The method discussed in this paper requires the solution of least-squares problems of the form (3.2), where $A \in \mathbb{C}^{m \times N_V}$ is highly ill-conditioned and W is a diagonal matrix containing quadrature weights. Sources of the ill-conditioning in A are the large dynamic range of the Fourier-Bessel basis functions and the ill-conditioning of fundamental solution bases. The ill-conditioning of the Fourier-Bessel functions can be improved by rescaling these functions so that they have unit value at the subdomain's maximum radius from the corner, which is what we have done in all example calculations. By contrast, ill-conditioning



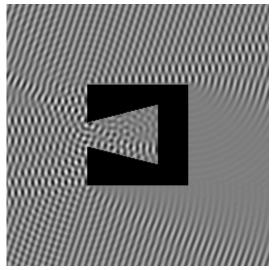


Fig. 6.6. Sound-soft scattering from a cavity domain.

of the fundamental solutions is inextricably linked to the fast exponential decay of the eigenvalues $\hat{s}_{\eta}(m)$ of the associated layer potential operator S_{η} , thus cannot be removed by scaling.

Denote by c_{LS} the solution of the least-squares problem (3.2). If we have sufficiently many quadrature points then $t[v_{LS}]$ is well approximated by the discrete error $t_d[v_{LS}] := ||W(Ac_{LS}-b)||_2$. Solving (3.2) using a backward stable least-squares solver—such as mldivide in MATLAB—returns a solution \tilde{c}_{LS} , which is the exact minimizer of

$$\min_{c \in \mathbb{C}^{N_V}} \| (WA + E)c - (Wb + f) \|_2$$

where E and f are small perturbations bounded by

$$||E||_2, ||f||_2 \le C\epsilon_{mach},\tag{7.1}$$

where the constant C>0 is small and the machine precision is ϵ_{mach} [2]. Because of ill-conditioning, \tilde{c}_{LS} may differ wildly from c_{LS} . However, the following shows that the corresponding discrete least-squares functional can only deviate from its true minimum under certain conditions.

LEMMA 7.1. Let \tilde{c}_{LS} be defined as above, and \tilde{v}_{LS} be the corresponding solution field. Then,

$$t_d[v_{LS}] \le t_d[\tilde{v}_{LS}] \le t_d[v_{LS}] + C(2 + ||c_{LS}||_2 + ||\tilde{c}_{LS}||_2)\epsilon_{mach}$$
 (7.2)

Proof. Using the property that c_{LS} minimizes the unperturbed least-squares problem, and the triangle inequality, results in

$$||WAc_{LS} - Wb||_{2} \le ||WA\tilde{c}_{LS} - Wb||_{2} \le ||(WA + E)\tilde{c}_{LS} - (Wb + f)||_{2} + C(1 + ||\tilde{c}_{LS}||_{2})\epsilon_{mach}.$$

Exchanging perturbed and unperturbed quantities in the above gives similarly

$$||(WA+E)\tilde{c}_{LS}-(Wb+f)||_2 \leq ||WAc_{LS}-Wb||_2 + C(1+||c_{LS}||_2)\epsilon_{mach}.$$

Combining the two estimates gives (7.2). \square

Thus, if the basis coefficient 2-norms at the approximate and exact minima are small, the numerical least-squares solution must converge at the same exponential rate (Theorem 5.4) as the exact least-squares solution. To rigorously apply this Lemma we need bounds on the norms of $||c_{LS}||_2$ and $||\tilde{c}_{LS}||_2$. A partial result towards deriving such bounds is given in the following

Theorem, which establishes under which conditions on Γ_F the coefficients of the fundamental solutions basis functions in Ω_e are bounded independently of N. A similar result for approximating interior Helmholtz problems with fundamental solutions was given in [4].

Theorem 7.2. Consider a sequence of fundamental solutions approximations, each of the form $v(\mathbf{x}) = \sum_{j=1}^{N_e} c_j^{(e)} \left(\frac{i}{4} \frac{\partial}{\partial \nu(\mathbf{y}_j)} H_0^{(1)}(k|\mathbf{x} - \mathbf{y}_j|) - \frac{\eta}{4} H_0^{(1)}(k|\mathbf{x} - \mathbf{y}_j|) \right)$, with growing numbers N_e (N_e even) of charge points, that attains the error bound from Theorem 5.2 as $N_e \to \infty$. Let each coefficient vector be written $c^{(e)} := \left[c_1^{(e)}, \dots, c_{N_e}^{(e)}\right]^T$. If $R > \max_i |p_i|$, then the sequence of norms $||c^{(e)}||_2$ is bounded independently of N_e .

Proof. In the following C > 0 denotes an unspecified constant that depends on k, R, r and ϵ

but not on N_e , and may change throughout the proof. By assumption $\max_i |p_i| < R < r$. Let $g(\theta) = \sum_{j=1}^{N_e} c_j^{(e)} \delta(\phi - \phi_j)$ and $\hat{g}(m) = \frac{N_e}{2\pi} \hat{c}_{(m \mod N_e)}^{(i)}$ the Fourier coefficients of g (see Section 5.1). The proof follows by showing that $\hat{g}(m)$ is boundeded for m in the interval $[-N_e/2 + N_e]$ $1, \ldots, N_e/2$ and applying the discrete version of Parseval's identity. For the specific choice of the coefficients $c^{(e)}$, such that $\hat{g}(m) = \frac{\hat{u}_s(m)}{\hat{s}_{\eta}(m)}$, which was used for the convergence analysis in Section 5.1, boundedness already follows for sufficiently small ϵ from

$$|\hat{g}(m)| \le C \left(\frac{r}{R\rho_{\epsilon}}\right)^{|m|},$$

and the assumption $R > \max_i |p_i|$.

To establish boundedness of $|\hat{q}(m)|$ for any sequence $c^{(e)}$ that attains the optimal rate of convergence from Theorem 5.2 we proceed as follows. From (5.2) and (5.5) we have that

$$k\sqrt{2\pi r}|\hat{u}_s(m) - \hat{s}_{\eta}(m)\hat{g}(m)| \le C\tau^{-N_e}, \quad m \in \mathbb{N}$$

where $\tau = \min\{\left(\frac{r}{R}\right), \rho_{\epsilon}^{1/2}\}$ (note that by Remark 5.3, $\tilde{\epsilon} = 0$ in Theorem 5.2 is possible since we only use the estimate for the function and not for the normal derivative). It follows that

$$|\hat{g}(m)| \le |\hat{s}_{\eta}(m)|^{-1} \left(C\tau^{-N_e} (2\pi r)^{-1/2} + |\hat{u}_s(m)| \right), \quad m \in \mathbb{N}$$

Now restrict m to the interval $[-N_e/2+1,\ldots,N_e/2]$. If $R>\sqrt{r\max_i|p_i|}$ then for sufficiently small ϵ it follows that $\tau = \frac{r}{R}$. Using (4.4) and (5.4) we have

$$|\hat{g}(m)| \le C \left(\frac{r}{R}\right)^{|m|} \left[\left(\frac{r}{R}\right)^{-N_e} + \rho_{\epsilon}^{-|m|} \right] \le C \left(\frac{r}{R}\right)^{|m|} \left[\left(\frac{r}{R}\right)^{-|m|} + \rho_{\epsilon}^{-|m|} \right]$$
(7.3)

since r > R and $|m| \le \frac{N_e}{2}$. From $\max_i |p_i| < R$ it follows for sufficiently small ϵ that $\rho_{\epsilon}^{-|m|} \le R$ $\left(\frac{r}{B}\right)^{-|m|}$. Inserting this into (7.3) gives

$$|\hat{g}(m)| \leq C$$

where all constants are absorbed into C. If $R < \sqrt{r \max_i |p_i|}$ then for sufficiently small ϵ it holds that $\tau = \rho_{\epsilon}$. We obtain, absorbing constants into C,

$$|\hat{g}(m)| \leq C \left(\frac{r}{R}\right)^{|m|} \left[\rho_{\epsilon}^{-N_e/2} + \rho_{\epsilon}^{-|m|}\right] \leq C \left(\frac{r}{R}\right)^{|m|} \left[\rho_{\epsilon}^{-|m|} + \rho_{\epsilon}^{-|m|}\right] \leq C \left(\frac{r}{R\rho_{\epsilon}}\right)^{|m|} \leq C$$

for ϵ sufficiently small since $\max_i |p_i| < R$ and therefore $\rho_{\epsilon} > \frac{r}{R}$. From (4.8) it follows that $\hat{g}(m) = \frac{N_e}{2\pi} \hat{c}^{(e)}(m)$ for $m \in [-N_e/2 + 1, \dots, N_e/2]$, where the vector $\hat{c}^{(e)} \in \mathbb{C}^{N_e}$ is the discrete Fourier transform of $c^{(e)}$. Hence, using the boundedness of $\hat{g}(m)$ for $m \in [-N_e/2+1,\ldots,N_e/2]$ we obtain together with the discrete version of Parseval's identity $\|c^{(e)}\|_2^2 = N_e \|\hat{c}^{(e)}\|_2^2$ that

$$\|c^{(e)}\|_2^2 = N_e \|\hat{c}^{(e)}\|_2^2 = \frac{(2\pi)^2}{N_e} \sum_{j=-\frac{N_e}{2}+1}^{\frac{N_e}{2}} |\hat{g}(m)|^2 \le (2\pi)^2 C^2,$$

which concludes the proof. \Box

Hence, the coefficient vector of the MFS approximations stays bounded in norm as long as $\max_i |p_i| < R$. In [4] this was studied numerically for interior Helmholtz problems, in the general case of analytic MFS and boundary curves. In that work we also showed that a large coefficient norm forces a corresponding loss of accuracy in solution evaluation due to round-off error.

Consider again the problem of sound-soft scattering on the square (Section 6.1). We have $|p_i| = \frac{\sqrt{2}}{2}$ for all corners of the square $\Omega = (-1/2, 1/2)^2$. If the radius of the source points \mathbf{y}_j is R = 0.4 then for k = 100 and N = 150 we find that the computed coefficient vector \tilde{c}_{LS} has $\|\tilde{c}_{LS}\| \approx 3 \cdot 10^9$. The measured boundary error $t_d[\tilde{v}_{LS}] \approx 3.6$ is terrible. On the other hand, if $R = 0.8 > \max_i |p_i|$ then $\|\tilde{c}_{LS}\| \approx 6.2$ and $t_d[\tilde{v}_{LS}] \approx 7 \cdot 10^{-10}$ is excellent. Note that in both cases the matrix WA is numerically singular with a condition number of around 10^{20} .

To summarize this section, the crucial factor limiting the accuracy of the method is not the condition number of WA, but the norm of the basis coefficients vector. This can be kept small by rescaling Fourier-Bessel functions and by an informed choice of the fundamental solutions curve.

8. Conclusions. In this paper we have demonstrated that efficient and highly accurate computation of scattering from a variety of convex and nonconvex polygons can be achieved by using a small number of elements, one for each corner, with the right kind of basis functions on each element. This allows one to tackle problems tens of wavelengths across quite rapidly, even with dense linear algebra. An advantage of the latter is that a dense factorization would allow multiple right-hand sides (e.g. incident wave directions) to be solved at minimal extra cost. Once the coefficients have been solved for, the user may choose where to evaluate the solution at a constant cost per point—this may be slow on large grids, but is trivially parallelizable.

For high-frequency sound-soft scattering of plane waves on convex polygons the method analyzed in this paper cannot compete with specialized high-frequency boundary element methods (e.g. [10]). However, the approach in this paper is much more general. It allows for the computation of convex and nonconvex polygons, sound-soft or sound-hard boundary conditions, and non-plane wave incident fields. Due to its fast exponential convergence, up to medium frequencies (< 50 wavelengths in size) it is very competitive.

Our implementation is also much simpler than boundary element or boundary integral methods. At high frequency these usually need a careful treatment of oscillatory integrals with complicated basis functions and weakly oscillatory kernels, or spectral quadrature schemes for (hyper)singular kernels. In addition they require careful mesh-refinement or quadrature schemes for corner singularities. None of this complexity is needed in our approach. Moreover, with MPSpack we have created a simple-to-use software tool that allows rapid and flexible implementation of our methods, and more (see Appendix A).

The approach in this paper can easily be generalized to multiple scattering problems. For example, Figure 8 shows the solution for sound-soft scattering with incident angle $-\frac{\pi}{3}$ from a disk near a snowflake domain, at wavenumber k=100. Setting up and solving the matrix problem takes around 7.5 minutes with an error $t[v_{LS}] \approx 4 \cdot 10^{-7}$. Here we used an additional basis of fundamental solutions with charge points lying on a circle inside the disk domain.

The exponential convergence analyzed in this paper relies on asymptotic series for Helmholtz solutions at corners, and their convergence properties, which are known analytically for sound-soft and sound-hard boundary conditions on straight boundaries. For impedance, transmission, and curved-wall corner problems, the solution is not separable close to a corner, and it is an open problem what kind of basis functions to choose here to obtain fast exponential convergence.

Another open problem is optimal mesh (subdomain) generation. The mesh should be chosen in such a way that fast exponential convergence on each element is guaranteed by keeping the relative distance of mesh elements to neighboring singularities large. But at the same time the mesh should not contain too many elements, so that the overall numbers of basis functions and matching points remains small.

Finally, the efficient solution of the least-squares linear algebra problem needs to be further investigated. At the moment we use a dense, backward stable solver, which works well for medium

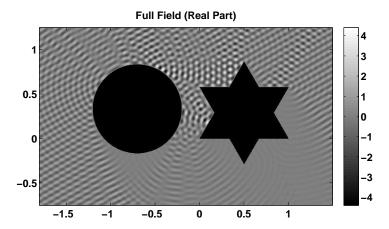


Fig. 8.1. Multiple scattering from a disk and a snowflake domain.

wavenumbers. However, it would be desirable to have iterative solvers for large, structured, ill-conditioned least-squares problems in order to remove the $O(k^3)$ bottleneck at high wavenumbers.

Acknowledgements. Timo Betcke wishes to thank Lehel Banjai (MPI Leipzig), Mahadevan Ganesh (Colorado School of Mines) and Ralf Hiptmair (ETH Zürich) for helpful discussions. Alex Barnett thanks Leslie Greengard (Courant) for helpful discussions. Both authors wish to thank the anonymous referees for helpful comments that improved the presentation of the paper.

Appendix A. MPSpack example: scattering from a square.

MPSpack is an object-oriented MATLAB toolbox developed by the authors to solve twodimensional Helmholtz/Laplace PDE problems with particular and fundamental solution methods. The code can be obtained from http://code.google.com/p/mpspack. The Download section of that webpage also includes an extensive tutorial and a technical manual.

The short human-readable code in Figure A.1 implements scattering from the square as in Section 6.1, giving least-squares error 8×10^{-11} in 6 seconds of CPU time, and a further 11 seconds to plot the solution. The code is explained in more detail in the tutorial. Here we will only mention that segments are first created as in Figure 2.1, then subdomains E_1 through E_4 and Ω_e^+ are created. Fractional-order Bessel function bases are added to subdomains with the command addcornerbases, fundamental solutions with addmfsbasis. Finally a scattering object is created which contains methods to construct the matrix WA from Section 3, solve the least-squares linear system, and plot the solution.

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```
% Exponentially accurate sound-soft time-harmonic scattering from the square
k = 50;
                       % Wavenumber
r = 1.0;
                       % Radius of outer artificial circle
M = 100;
                       % Number of quadrature points on each segment
N = 90;
                       % Number of basis funcs in each corner subdomain
a = 0.5;
                       % Half-size of the square
R = sqrt(0.5);
                      % Radius of the fundamental solutions curve
% Define segments...
s = segment.polyseglist(M, [1i*r 1i*a a+1i*a a r]);
                                                         % straight
s = [s(1:3) segment(3*M, [0 r 0 pi/2])];
                                                         % add arc
s = [s rotate(s, pi/2) rotate(s, pi) rotate(s, 3*pi/2)]; % add 3 copies
sext = s([4 8 12 16]);
                                    % segments forming outer circle
% Define domains...
for j=1:4, d(j) = domain(s(1+mod(4*(j-1)+[0 1 2 12 3],16)),[1 1 1 -1 1]); end
ext = domain([], [], sext(end:-1:1), -1);
sart.setmatch([k -k], [1 -1]);
                                   % matching conditions between elements
% Basis functions...
nuopts = struct('type','s', 'cornermultipliers',[0 0 1 0 0], 'rescale_rad',1);
                                                         % frac-order FB
for j=1:4, d(j).addcornerbases(N,nuopts); end
Z = Q(t) R*exp(2i*pi*t); Zp = Q(t) 2i*pi*R*exp(2i*pi*t);
                                                        % fund soln curve
ext.addmfsbasis({Z, Zp}, N, struct('eta',k, 'fast',2, 'nmultiplier',2.1));
p = scattering(ext, d);
                                    % now set up problem, solve, and plot...
p.setoverallwavenumber(k);
p.setincidentwave(-pi/6);
p.solvecoeffs;
fprintf('least-square err = %g, coeff norm = %g\n', p.bcresidualnorm, norm(p.co))
p.showfullfield(struct('bb',[-1.5 1.5 -1.5 1.5], 'dx',0.02));
```

Fig. A.1. MATLAB code for sound-soft scattering from a square using MPSpack toolbox.

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