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Signature: \_\_\_\_\_

EEE 313 Spring 2015

Bilkent University  
Department of Electrical and Electronics Engineering  
EEE 313 Electronic Circuit Design

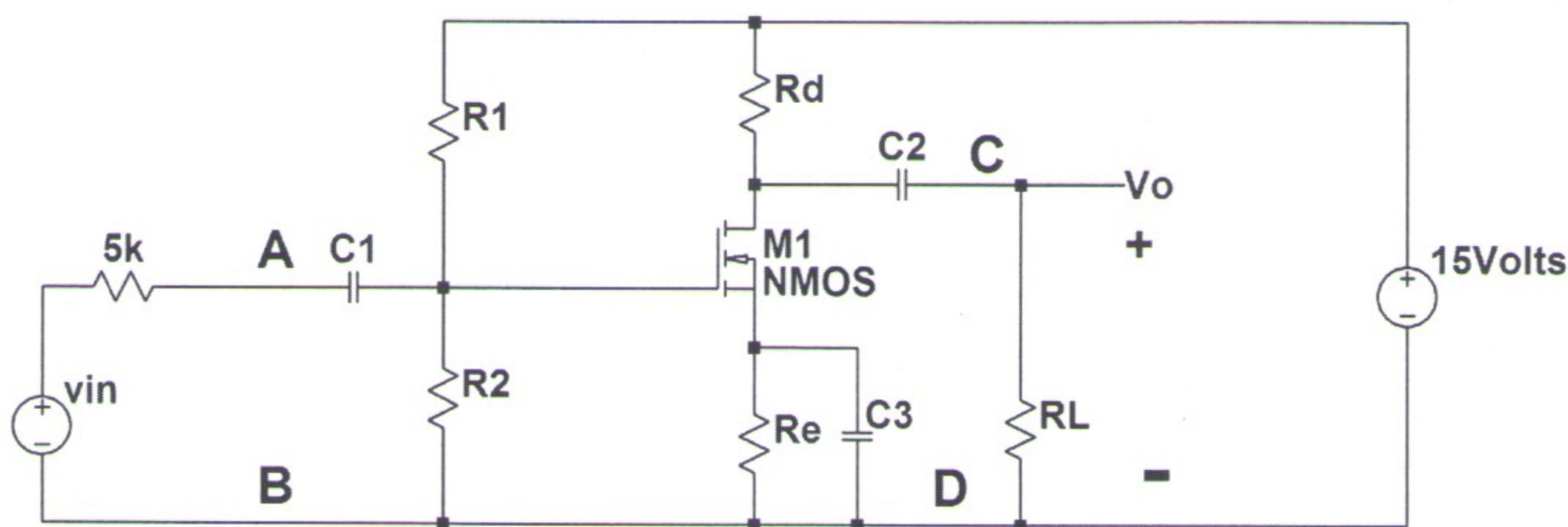
**Midterm 2**  
20 April 2015, 17:50  
(4 questions, 120 minutes)

- This is a **closed book**, closed notes exam. No cheat sheet allowed.
- All cell-phones should be completely **turned off**.
- Use a calculator for numerical computations. Carry at least **4 significant digits** during calculations. Your final answer should be at least **3 significant digits**.
- Be sure to write the **units** of all numerical results.
- **Show** all work clearly.
- Please put your **final answer** for each part inside a box for easy identification. Do not give multiple answers, they will not be graded.
- Do not remove the **staple** from the exam sheets or separate pages of the exam. All extra pages must be stamped to your exam.
- You may leave the exam room when you are done. However, please do not leave during the **last five minutes** of the exam.
- At the end of the exam, please stay seated until **all** exam papers are collected.

Please do not write below this line

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1. 25 pts.	
2. 25 pts.	
3. 20 pts.	
4. 30 pts.	
Total 100 pts.	



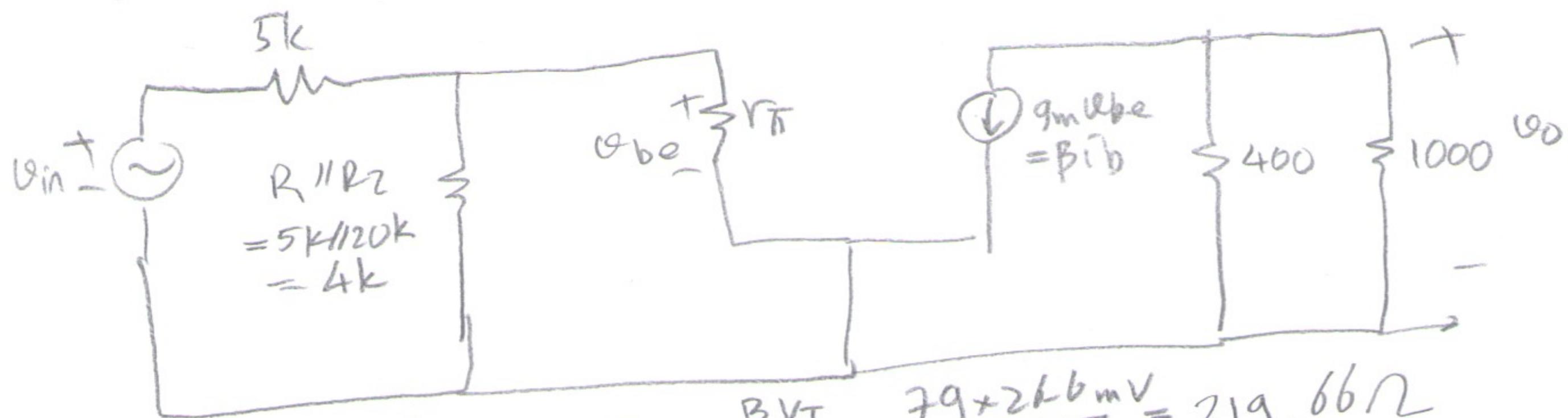
- 1) (25 points) At the circuit given above,  $K_n = 10 \times 10^{-3} \text{ A/V}^2$ ,  $V_{TN} = 2 \text{ Volts}$ . Assume that the DC solution was calculated before and the drain current,  $I_D$  was found to be equal to  $10 \text{ mA}$  and the state of the transistor was found to be saturation. The amplifier is going to be used as microphone amplifier with a bandwidth of  $20 \text{ Hz} - 20 \text{ kHz}$ . The values of  $R_1$  and  $R_2$  are set to  $12 \text{ k}\Omega$ ,  $R_E = 300 \Omega$ ,  $R_d = 500 \Omega$  and  $R_L = 1 \text{ k}\Omega$ .
- (7 points) Assuming that the rest of the capacitors are very large, find the value of  $C_1$  in order to have  $3 \text{ dB}$  lower cut-off frequency.
  - (8 points) Assuming that the rest of the capacitors are very large, find the value of  $C_2$  in order to have  $3 \text{ dB}$  lower cut-off frequency.
  - (10 points) Assuming that the rest of the capacitors are very large, find the value of  $C_3$  in order to have  $3 \text{ dB}$  lower cut-off frequency.

Solutions

a-) Drawing the equivalent circuit of the input circuit;

$\frac{v_{gs}}{v_{in}} = \frac{6000}{6000 + 500 + \frac{1}{j\omega C_1}} \Rightarrow \omega_o = 2\pi \times 20 = \frac{1}{11000 C_1} \Rightarrow \omega_o = 125.66$ 
 $C_1 = \frac{1}{125.66 \times 11000} = 7.23 \times 10^{-7} \text{ F} = 0.723 \mu\text{F}$ 
 $R_1 \parallel R_2 = 6 \text{ k}$

b-) Drawing the AC equivalent circuit:



$$g_m = \frac{I_c}{V_T} = \frac{10\text{mA}}{26\text{mV}} = 0.3768$$

$$r_{\pi} = \frac{B V_T}{I_c} = \frac{79 \times 26\text{mV}}{10\text{mA}} = 219.66\Omega$$

$$\alpha_o = \left( \frac{r_{\pi} // 4k}{(r_{\pi} // 4k) + 5k} \right) g_m \cdot \frac{r_{\pi} // 4k}{(r_{\pi} // 4k) + 5k} \cdot \frac{1000}{1000 + 219.66} = \frac{219.66 \times 1000}{4219.66} = 208.2\Omega$$

$$\frac{V_o}{V_{in}} = -285.7\Omega \times 0.3768 \times \frac{208.2}{5208.2} = -4.295$$

c-)  $r_{\pi} // R_1 // R_2 = R_{in} = 208.2\Omega$

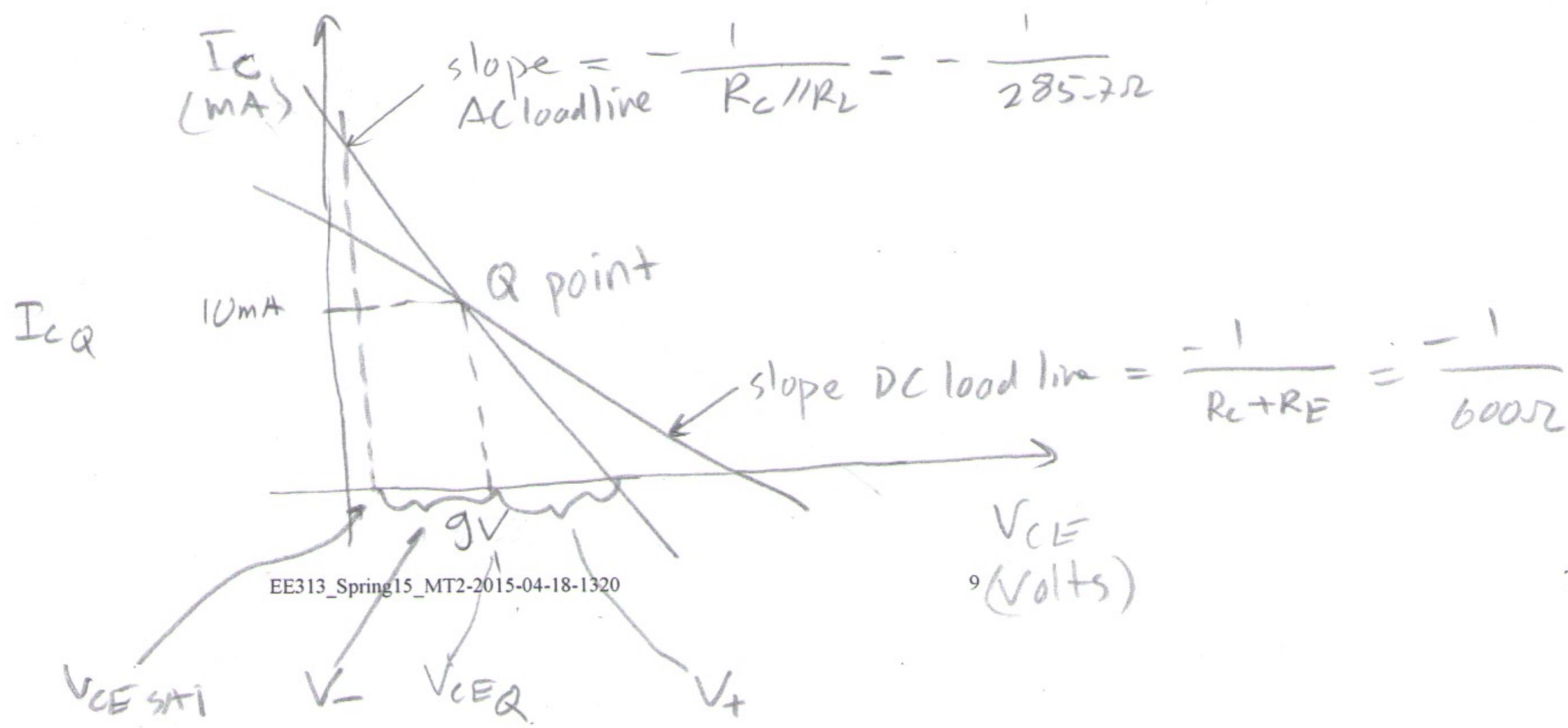
d-)  $R_D = R_C = 400\Omega$

e-)  $V_+ = I_{CQ} \times R_C // R_L = 10\text{mA} \times 285.7\Omega = 2.857\text{V}$

$$V_- = V_{CEQ} - V_{CESAT} = 9\text{V} - 0.4\text{V} = 8.6\text{V}_{\text{diff}}$$

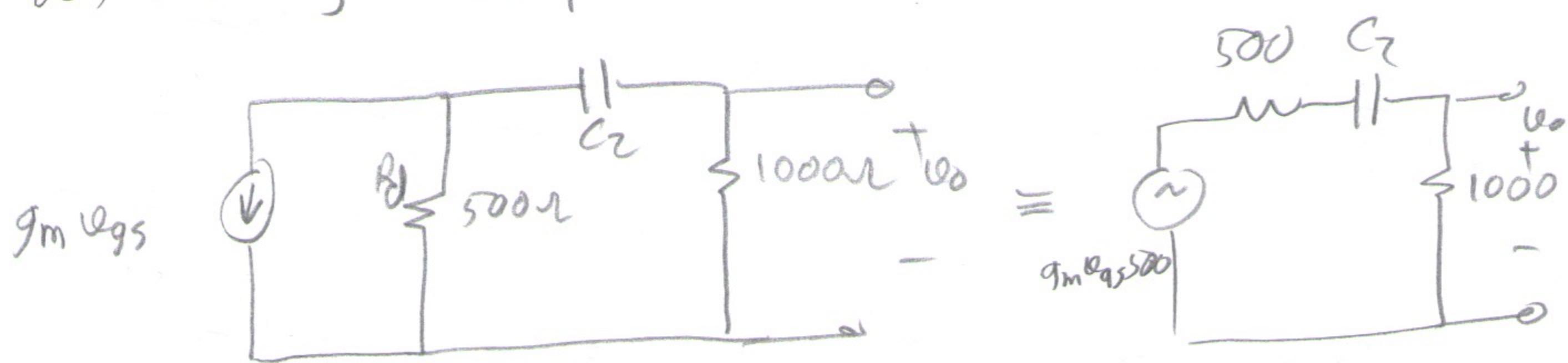
where  $V_{CEQ} = 15 - 10\text{mA} \times 600\Omega = 9\text{V}$

$$V_{P-P \text{ undistorted}} = 2 \times \min \{ V_+, V_- \} = 2 \times 2.857 = 5.714\text{V}$$



Tank Reyhen

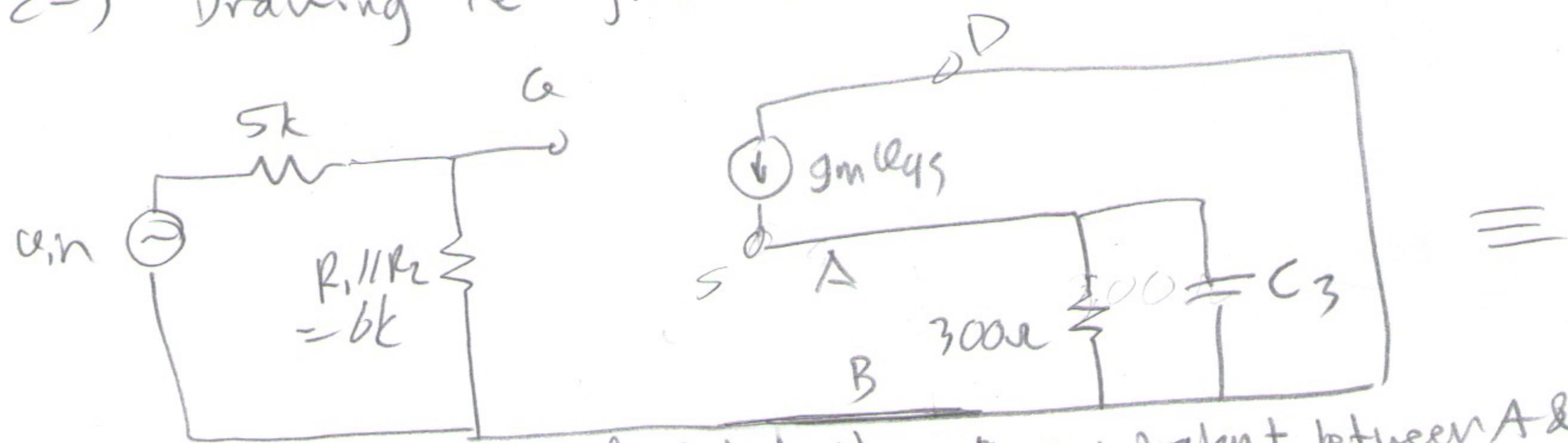
b) Drawing the output circuit



$$\frac{v_o}{g_m V_{GS} 500} = \frac{1000}{1000 + 500 + \frac{1}{j\omega C_2}} \Rightarrow \omega_0 = 126.66 \text{ rad/sec} = \frac{1}{1500 C_2}$$

$$\Rightarrow C_2 = \frac{1}{126.66 \times 1500} = 5.305 \times 10^{-6} \text{ F} \approx 5.3 \mu\text{F}$$

c) Drawing the gate-to-source circuit

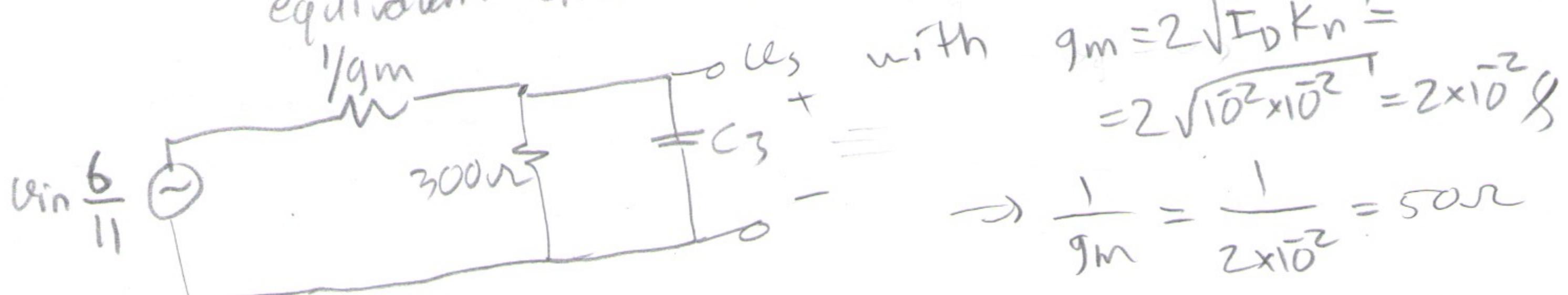


in order to find the Thévenin's equivalent between A & B  
find the open circuit voltage between A & B with \$300\Omega\$ and \$C\_3\$ removed.  $\Rightarrow$

$$V_{open} = u_{in} \times \frac{6k}{5k+6k} \quad \text{as source voltage will be equal to the gate voltage if no current passes through $g_m V_{GS}$}$$

i.e.  $g_m V_{GS} = 0 \Rightarrow V_{GS} = 0$

$R_o$  between A & B is equal to  $1/gm$   $\Rightarrow$  the equivalent circuit becomes



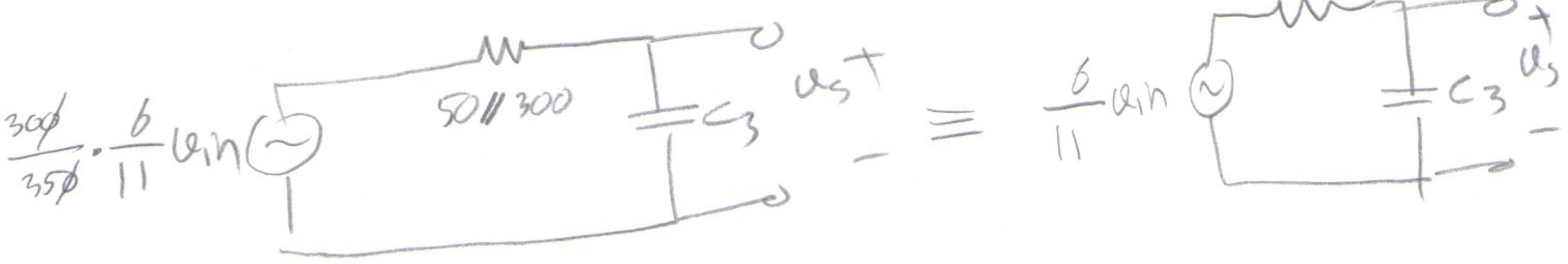
$$gm = 2\sqrt{I_D K_n} = 2\sqrt{10^2 \times 10^2} = 2 \times 10^2 \text{ S}$$

$$\Rightarrow \frac{1}{gm} = \frac{1}{2 \times 10^2} = 50\Omega$$

continued

(page 3-b)

$\Rightarrow$  The equivalent circuit then becomes



The output current is  $g_m v_{egs} \Rightarrow i_o = g_m (v_g - v_s) \Rightarrow$

$$\hat{i}_o = g_m \left[ \frac{6}{11} v_{in} - \frac{6}{11} v_{in} \left( \frac{\frac{1}{j\omega C_3}}{42.9 + \frac{1}{j\omega C_3}} \right) \cdot \frac{6}{350} \right]$$

$$\Rightarrow \hat{i}_o = \frac{6 g_m v_{in}}{11} \left[ 1 - \frac{6}{7} \frac{1}{1 + j 42.9 \omega C_3} \right] \Rightarrow 1, 42.9 \omega C_3$$

$$\frac{\hat{i}_o}{v_{in}} = \frac{6 g_m}{11} \left[ \frac{1}{7} + \frac{6}{7} \left[ 1 - \frac{1}{1 + j 42.9 \omega C_3} \right] \right]$$

$$\frac{\hat{i}_o}{v_{in}} = \frac{6 g_m}{11} \left[ \frac{1}{7} + \frac{6}{7} \left[ \frac{j + j 42.9 \omega C_3}{1 + j 42.9 \omega C_3} \right] \right] \Rightarrow$$

$$\frac{\hat{i}_o}{v_{in}} = \frac{6 \times 0.02}{11 \times 7} \left[ 1 + 6 \frac{j \omega 42.9 C_3}{1 + j \omega 42.9 C_3} \right] \Rightarrow \text{The } 3\text{dB}$$

point is defined by

$$\omega_0 = 2\pi \times 20 = 126.66 = \frac{1}{42.9 C_3} \Rightarrow$$

$$C_3 = \frac{1}{42.9 \times 126.66} = 1.855 \times 10^{-4} F = 185.5 \mu F$$

or one can simply say that  $\omega_0$  is defined by

$$\omega_0 = \frac{1}{\left(\frac{1}{g_m} / R_E\right) C_3} \quad \text{as mentioned in the classroom}$$

(page 3-c)

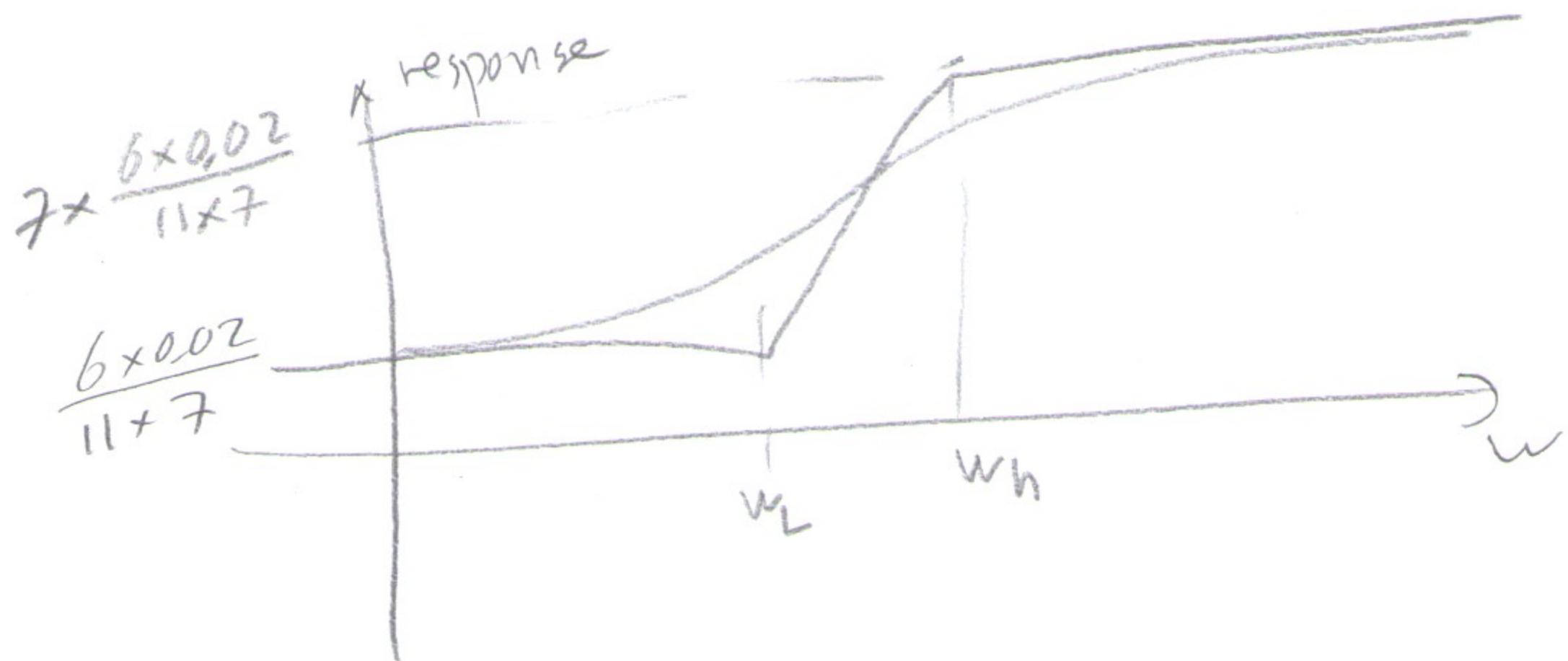
To be more precise:

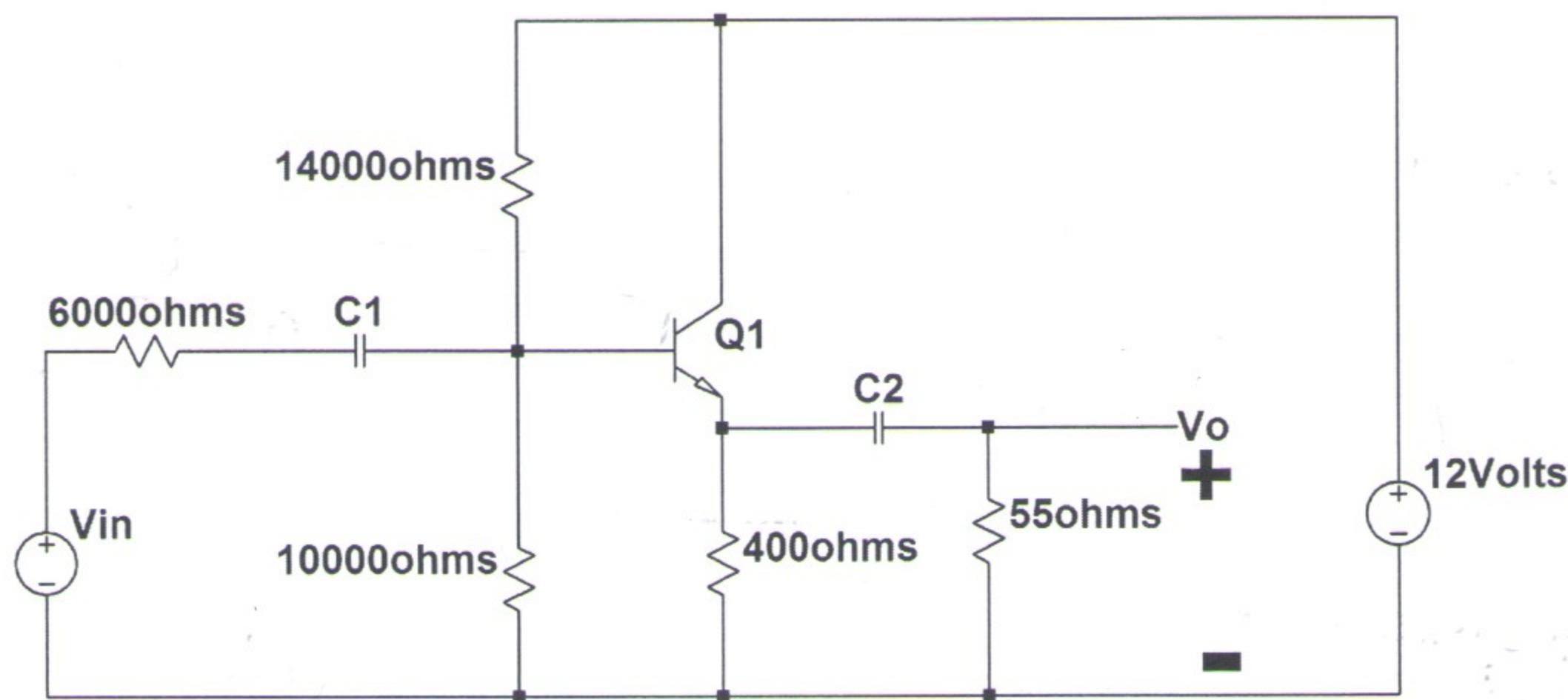
$$\frac{i_o}{v_{in}} = \frac{6 \times 0.02}{11 \times 7} \left[ 1 + b \frac{jw 42.9 C_3}{1 + jw 42.9 C_3} \right]$$

$$\frac{i_o}{v_{in}} = \frac{6 \times 0.02}{11 \times 7} \left[ \frac{1 + jw 42.9 C_3 + b jw 42.9 C_3}{1 + jw 42.9 C_3} \right]$$

$$= \frac{6 \times 0.02}{11 \times 7} \cdot \frac{1 + 7 \times 42.9 j w C_3}{1 + 42.9 j w C_3}$$

← higher cut-off  $w_L$   
← lower cut-off  $w_H$





2. (25 points) The transistor in the emitter follower circuit shown below has the following parameters:

$$\beta = 99, V_{BEON} = 0.7 \text{ Volts}, V_{CESAT} = 0.4 \text{ Volts}$$

The amplifier is used as a preamplifier in a music set. Therefore the frequency range is equal to 20Hz-20kHz. Answer the following:

- a) (7 points) Find the state and the bias of the transistor and verify it.

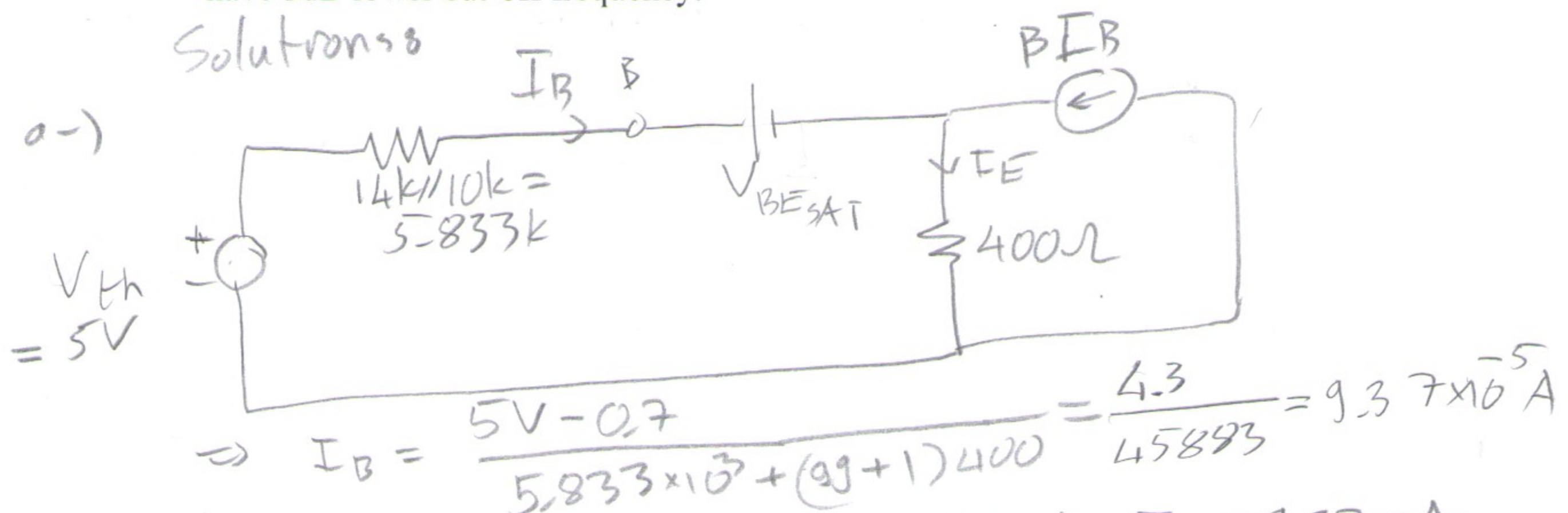
For the items of question (2) below, please take  $I_E$  as 10mA.

- b) (8 points) Assuming that  $C_2$  is very large, find the value of  $C_1$  in order to have 3dB lower cut-off frequency.

- c) (10 points) Assuming that  $C_1$  is very large, find the value of  $C_2$  in order to have 3dB lower cut-off frequency.

Solutions

a-)



$$I_B = \frac{5V - 0.7}{5.833 \times 10^3 + (99+1)400} = \frac{4.3}{45833} = 9.37 \times 10^{-5} \text{ A}$$

$$\Rightarrow I_B = 93.7 \mu\text{A} \Rightarrow I_C = 9.28 \text{ mA}, I_E = 9.37 \text{ mA}$$

The transistor is at FA because of the circuit configuration

$$V_{CE} = V_{CC} - I_E R_E = 12 - 9.37 \times 400 \Omega = 12 - 3.748 \text{ V} = 8.252 \text{ V}$$

b-) Drawing the AC equivalent circuit:

$$r_{\text{tube}} = \frac{\beta}{g_m} = \frac{99 \times V_T}{I_c} = \frac{99 \times 26.6 \text{mV}}{10 \text{mA}} = 263.34 \Omega$$

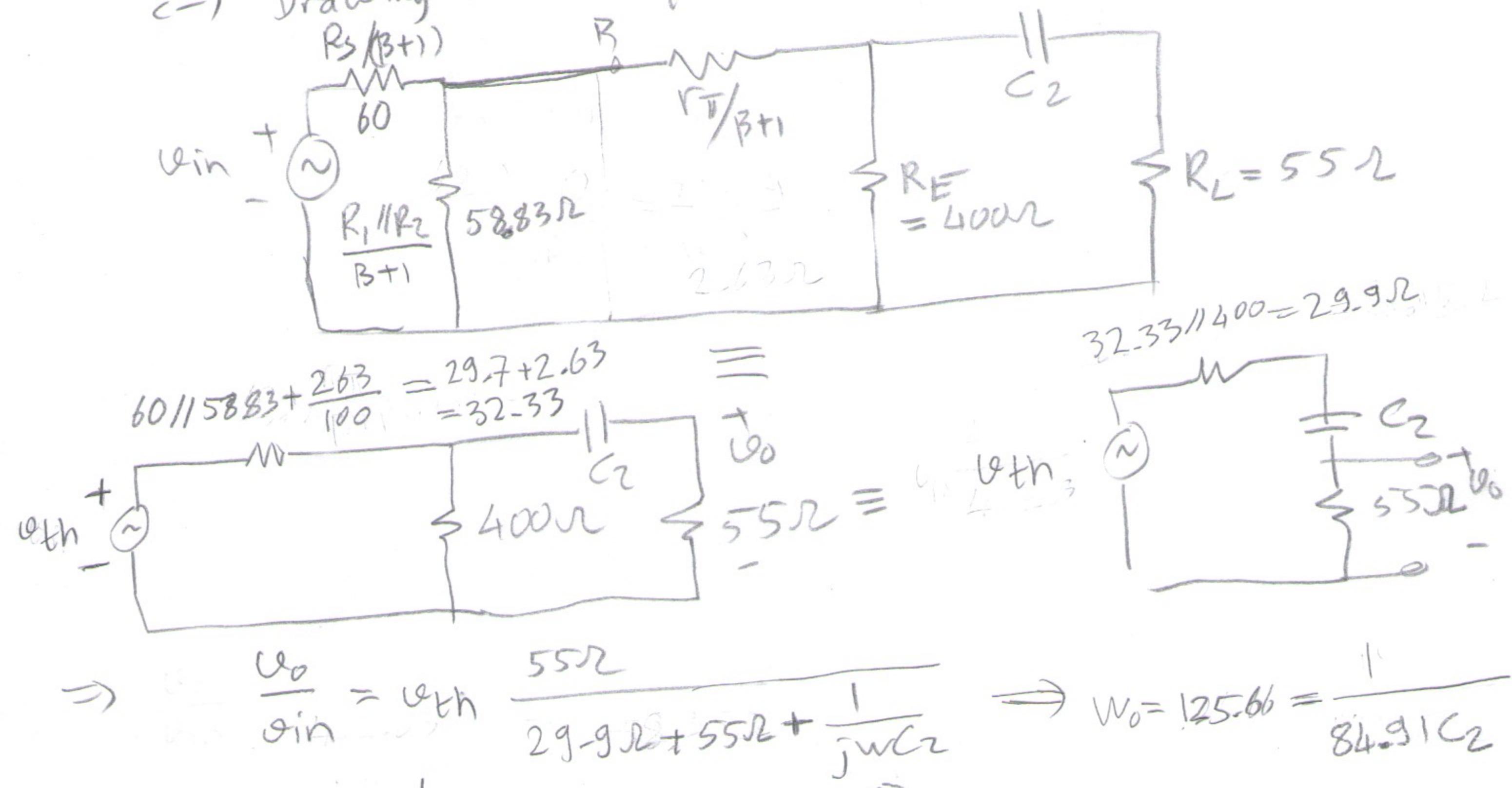
$$\frac{v_{be}}{v_{in}} = \frac{5883/(263 + 40000/15500)}{6000 + (5883/(263 + 40000/15500)) + \frac{1}{j\omega C_1}} = \frac{(5883/(263 + 4835))}{6000 + (5883/(263 + 4835)) + \frac{1}{j\omega C_1}}$$

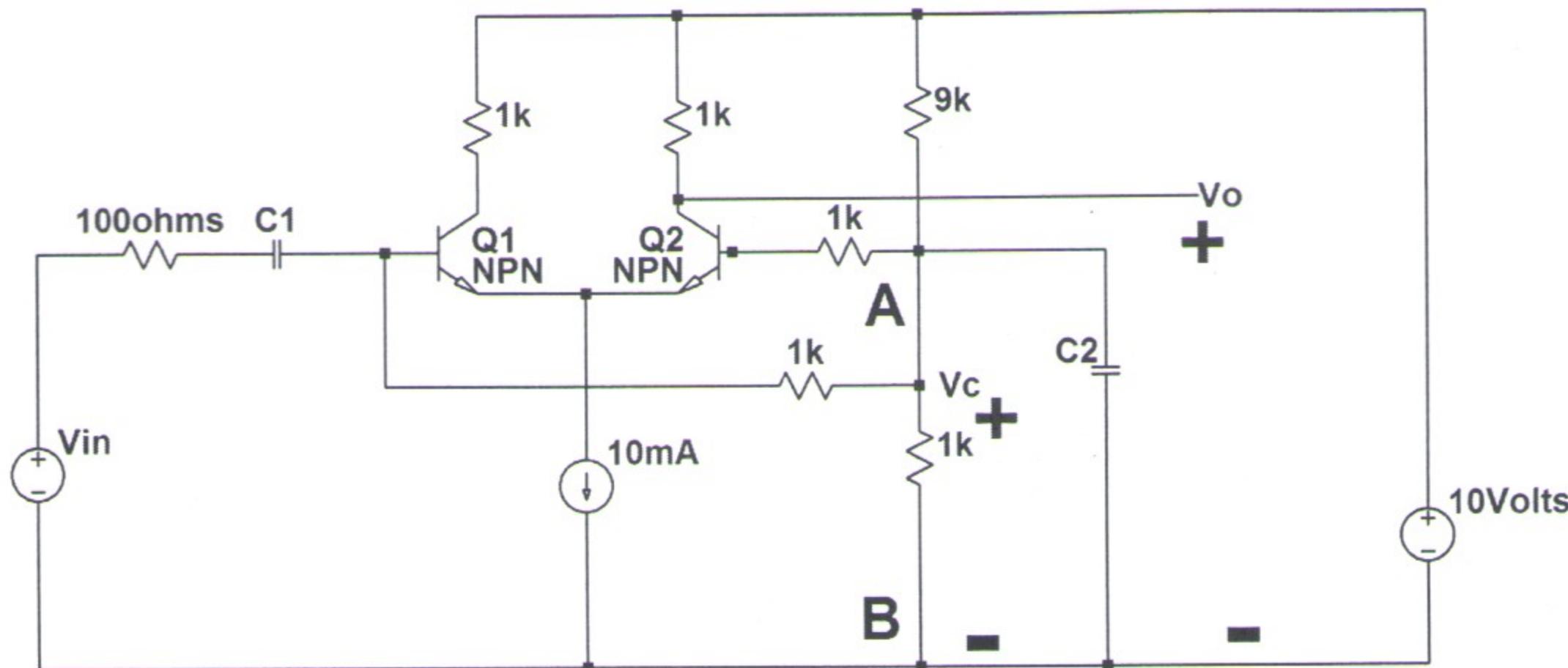
$$\frac{v_{be}}{v_{in}} = \frac{5883/(5098)}{6000 + (5883/(5098)) + \frac{1}{j\omega C_1}} = \frac{2731.2}{6000 + 2731.2 + \frac{1}{j\omega C_1}}$$

$$w_0 = 2\pi \times 20 = 125.66 = \frac{1}{8231.2 C_1} \Rightarrow$$

$$C_1 = \frac{1}{125.66 \times 8231.2} = 9.668 \times 10^{-7} \text{F} = 0.9668 \mu\text{F}$$

c-) Drawing the AC equivalent circuit of the output



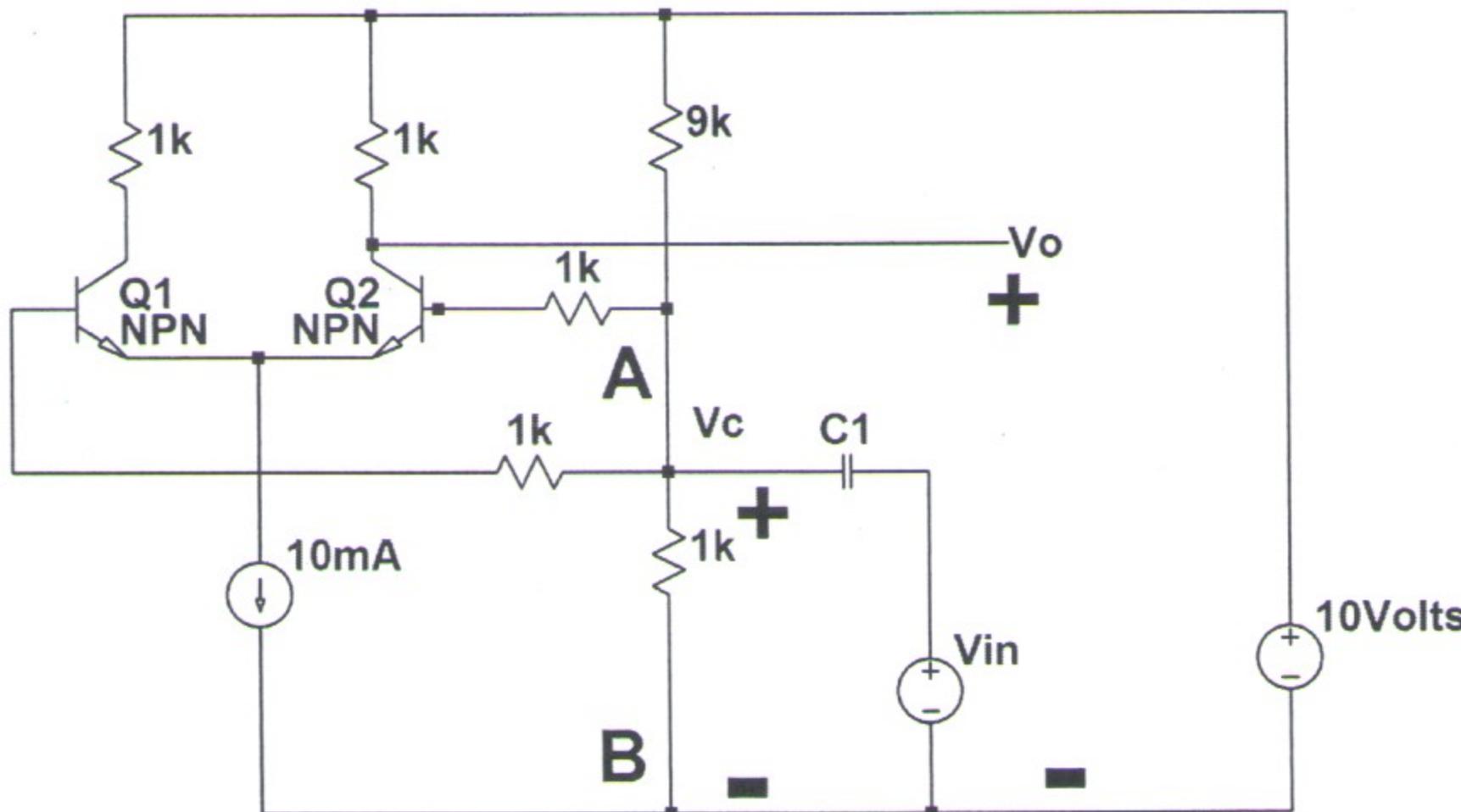


3. (20 points) At the circuit given above:

$\beta = 74$ ,  $V_{BEON} = 0.7$  Volts,  $V_{CESAT} = 0.3$  Volts,  $C_1$  and  $C_2$  are very large.

Please answer the following:

- (4 points) Find the DC bias of the transistors,
- (4 points) Draw the AC equivalent circuit,
- (4 points) Find the gain of the circuit defined as  $V_o/V_{in}$  assuming that the transistors are in forward active state.



- (4 points) The circuit is modified to measure the common-mode voltage. Find the common-mode gain defined as  $V_o/V_c$  and assuming that the transistors are in forward active state. Please note that  $C_1$  is very large again. The transistor parameters are still the same.
- (4 points) Write the definition of CMMR (Common-Mode Rejection Ratio).

### Solutions

a-) As the transistor biases are the same  $I_{E1} = I_{E2} = \frac{10mA}{2} = 5mA$

$$I_{B1} = I_{B2} = \frac{I_E}{(\beta+1)} = \frac{5mA}{75} = 0.066mA = 66.6\mu A$$

The base voltages are

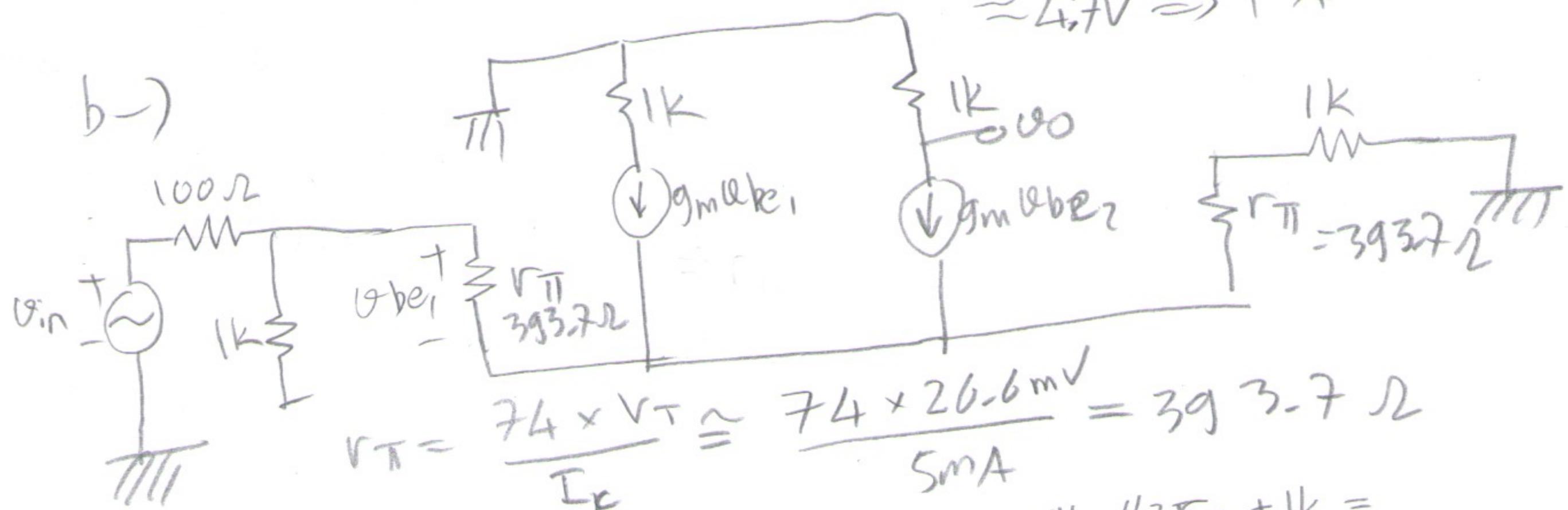
$$10V \times \frac{1k}{1k+9k} - I_B R_B = 10 \times \frac{1}{10} - 66.6 \times 10A \times 1000\Omega \\ = 1 - 0.66V = 0.933V$$

The collector voltages are

$$10V - 5mA \times 1k = 10V - 5 \times 10^3 \times 10^3 = 10 - 5 = 5V$$

$$\Rightarrow V_{CE} = 5V - (V_B - V_{BE(SAT)}) = 5 - (0.933) + 0.7 = \\ \approx 4.7V \rightarrow FA$$

b-)



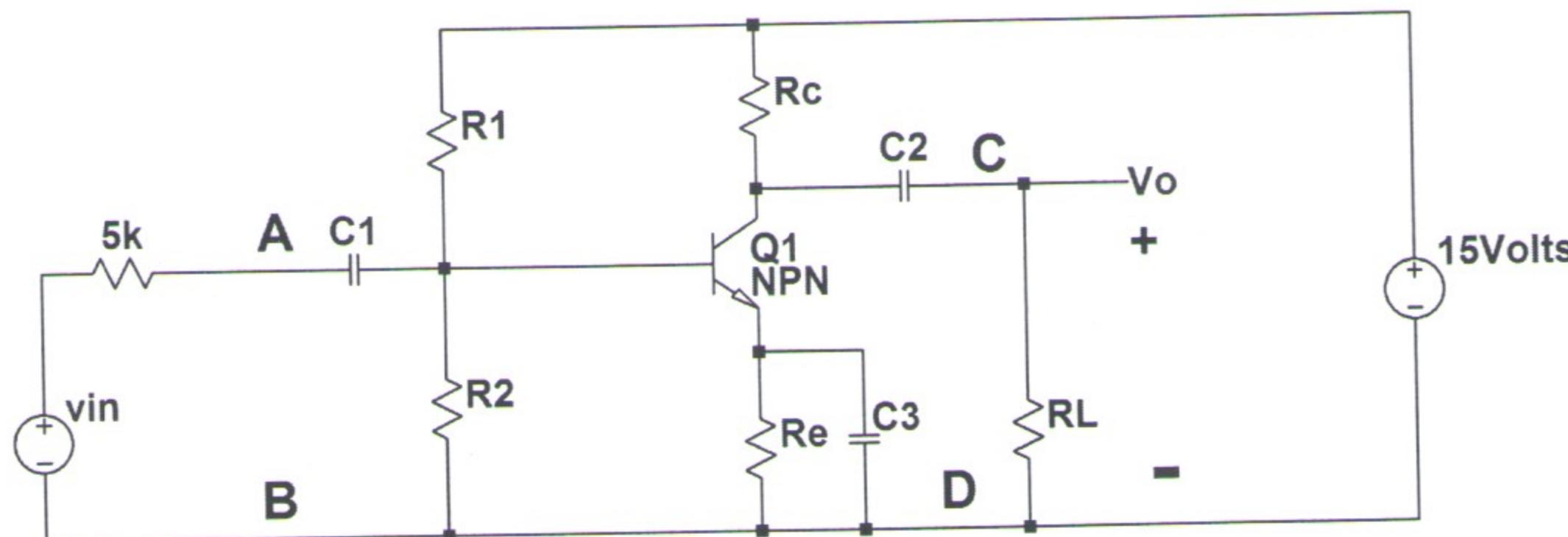
$$g_m = \frac{5mA}{26.6mV} = 0.1888$$

$$1k // 2r_{\pi} + 1k = \\ 1000 // (787 + 1000) = 641.2\Omega$$

$$c) \frac{V_o}{V_{in}} = \frac{641.2}{641.2 + 100} \cdot \frac{393}{1736} \cdot g_m \cdot R_C = 0.8651 \times 0.2264 \times 0.1888 \times 1000 \\ = 36.82$$

d-) As the common-mode voltage does not change the collector current because the emitters are biased through an ideal current source  $A_C \equiv \frac{V_o}{V_C} = 0$

$$e) CMRR = \frac{A_d}{A_c} = \frac{\text{differential mode gain}}{\text{common-mode gain}}$$



$V_{BE\text{SAT}}$

4. (30 points) At the BJT circuit given above:

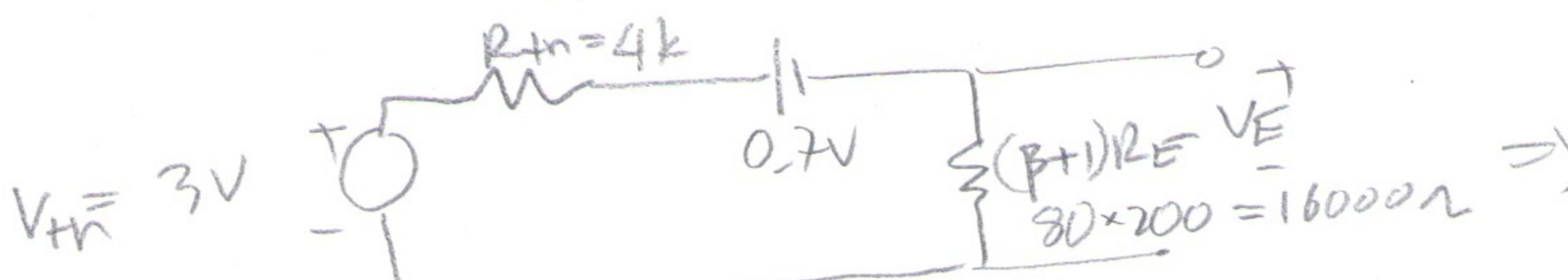
$\beta=79$ ,  $V_{BE\text{ON}} = 0.7$  Volts,  $V_{CE\text{SAT}} = 0.4$  Volts,  $C_1, C_2$  and  $C_3$  are very large,  $R_c=400\Omega$ ,  $R_L=1000\Omega$ ,  $R_1=20000\Omega$ ,  $R_2=5000\Omega$ ,  $R_e=200\Omega$ . Please answer the following:

- (6 points) Find the DC bias of the circuit,
- (6 points) Find the gain of the amplifier defined as  $V_o/V_{in}$  assuming that the transistor is at forward active and the collector current is equal to 10mA.
- (6 points) Find the input impedance seen between points A and B assuming that the transistor is at forward active and the collector current is equal to 10mA.
- (6 points) Find the output impedance seen between points C and D assuming that the transistor is at forward active and the collector current is equal to 10mA.
- (6 points) Find the maximum undistorted peak-to-peak output swing of the amplifier assuming that the transistor is at forward active and the collector current is equal to 10mA.

Solutions

$$o-1) 15V \times \frac{R_2}{R_1+R_2} = 15 \times \frac{5k}{5k+20k} = 15 \times \frac{5}{25} = 3V$$

$$R_{th} = \frac{5k \times 20k}{5k+20k} = 5k \times \frac{20}{25} = \frac{20}{5} = 4k \Rightarrow$$



$$I_B = \frac{3V - 0.7}{4000 + 16000} = \frac{2.3}{20000} = 1.15 \times 10^{-4} A = 115 \mu A$$

$$I_E = (\beta + 1)I_B = 80 \times 115 \mu A = 9.2 \text{ mA} \quad I_C = 79 \times 115 \mu A = 9.085 \text{ mA}$$

$$V_{CEQ} = 15V - 400 \times 0.009085A - 200 \times 9.2 \times 10^{-3} A =$$

$$V_{CEQ} = 15 - 3.634 - 1.84 = 9.526 V \Rightarrow FA$$