## 21-10-2017 BILKENT UNIVERSITY

## Department of Electrical and Electronics Engineering EEE313 Electronic Circuit Design Midterm Exam I

SOLUTION

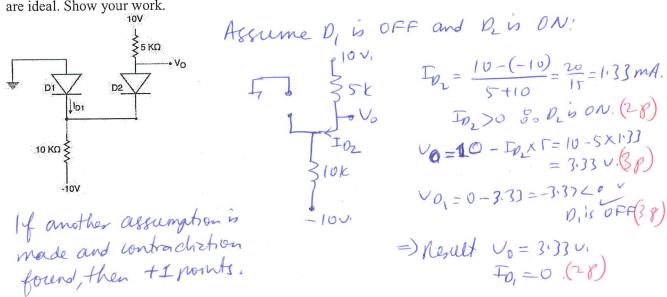
Surname:	,
Name:	
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Duration is 110 minutes. Solve all 6 questions. Show all your work. No books or notes.

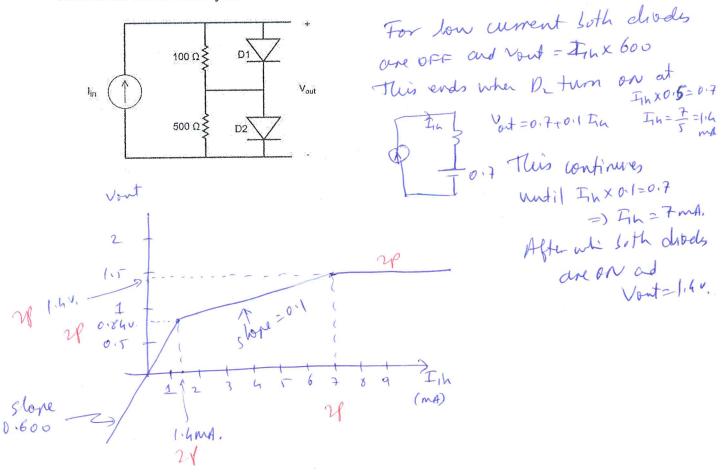
Q1 (20 points)	
Q2 (20 points)	
Q3 (10 points)	
Q4 (20 points)	
Q5 (20 points)	
Q6 (10 points)	
Total (100 points)	

Q1.

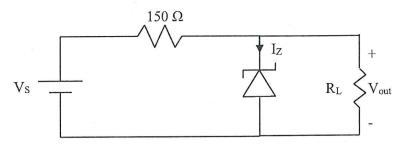
(a) Find  $V_0$  and  $I_{D1}$  in the circuit below. Justify the states of the diodes. Diodes in this circuit are ideal. Show your work.



(b) Find and plot  $V_{out}$  versus  $I_{in}$  transfer characteristics for the circuit below (Horizontal axis is  $I_{in}$  and the vertical axis is  $V_{out}$ ). The cutin voltage for the diodes is  $V_{\gamma} = 0.7$  V. Justify the states of the diodes. Show your work.



O2. For the circuit below,  $V_S = 60 \text{ V}$ , the Zener diode has a Zener voltage of 15 V and an incremental resistance of 0 Ω. The Zener current Iz needs to be more than 15 mA and the power rating of the Zener is 4 W. Show your work.



- a) Determine the range of Iz for proper operation of the Zener.
- b) Find the range of the DC voltage source Vs such that the current limitations of the Zener diode are not violated, for  $R_L = 150 \Omega$ .
- Determine the range of the load resistor, R<sub>L</sub>, for when V<sub>S</sub> = 60 V, such that the current limitations of the Zener diode are not violated.

b) 
$$R_{L} = 170 - \Omega = 0.17 \text{ K.R.}$$

$$T_{Z} = \frac{V_{S} - 15}{0.15} - \frac{15}{0.15} = \frac{V_{S} - 30}{0.15} \text{ mA.}$$

$$15 \angle T_{Z} \angle 267$$

$$15 \angle \frac{V_{S} - 30}{0.11} \angle 267$$

$$2.27 \angle V_{S} - 30 \angle 40$$

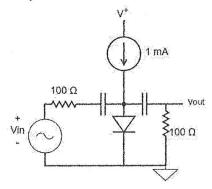
$$\frac{2.2r < V_{s} - 30 < 40}{4P}$$

$$\frac{4P}{32.2r} < V_{s} < 70 v,$$

$$e) V_{s} = 60$$

c) 
$$V_{S}=60$$
 $T_{Z}=\frac{60-15}{0.15}-\frac{15}{R}=\frac{45}{0.15}-\frac{15}{R}=300-\frac{15}{R}$ 
 $15 \leq T_{Z} \leq 267$ 
 $15 \leq 300-\frac{15}{R} \leq 2267$ 
 $-285 \leq -\frac{15}{R} \leq 33.33$ 
 $285 \geq \frac{15}{R} \geq 33.33$ 
 $285 \leq \frac{15}{R} \leq 32.33$ 
 $285 \leq \frac{15}{R} \leq 32.33$ 

- Q3. In the circuit below,  $v_{in} = 10$ sinwt mV. Capacitors are very large.
- a) Derive the formula for the incremental (small signal) resistance of the diode  $r_d = V_T/I_{DO}$ .
- b) Find the small signal component of the output voltage at  $300^{\circ}$ K ( $V_T = 26 \text{ mV}$ ). Show your work.



Solution

a) 
$$i_0 = I_s(e^{\frac{\sqrt{p}}{\sqrt{1}}}) \approx I_s e^{\frac{\sqrt{p}}{\sqrt{1}}} = I_s e^{\frac$$

b)

Note

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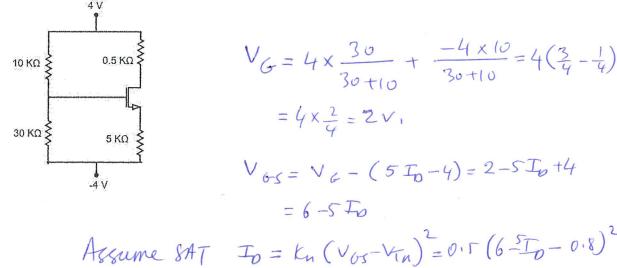
The root out 
$$d = \frac{0.026}{1mA} = 262$$
 (4p)

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The root out  $d = \frac{26.2}{1mA} = 26.2$  (4p)

Fout  $d = \frac{20.6}{20.6 + 100} \times 105 \text{ mwt (mv)}$ 
 $d = \frac{20.6}{20.6 + 100} \times 105 \text{ mwt (mv)}$ 
 $d = 1.71 \text{ shwt (mv)}$  (4p)

Q4. The nMOS transistor in the circuit below has the parameters  $V_{TN} = 0.8V$  and  $K_n = 0.5mA/V^2$ . Determine I<sub>D</sub>, V<sub>GS</sub>, and V<sub>DS</sub>. Justify your results. Show your work.



Assume SAT 
$$F_0 = K_n (V_{OS} - V_{Tn})^2 = 0.5 (6 - T_0 - 0.8)^2$$
  
= 0.5 (5.2 - 5  $T_0$ )<sup>2</sup>  
 $2T_0 = 5.2^2 - 52 T_0 + 25 T_0^2$ 

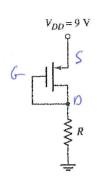
$$25 I_0^2 - 54 I_0 + 27.04 = 0$$

$$I_0 = \begin{cases} 1.37 \text{ mA} \Rightarrow V_{65} = 6 - 5 \times 1.37 = -0.85 \times 0.8 \\ > 0.79 \text{ mA} \Rightarrow V_{65} = 6 - 5 \times 0.79 = 2.056 > 0.8 \end{cases}$$

VDS = 4 - 0.79 ×0.5 - 5×0.79 +4=3.66 v. 3.66≥7.056-0.8 V 3. SAT.

Q5.

(a) The pMOS transistor in the figure below has  $V_{TP}$  = -0.7V and  $\lambda$  = 0. Find the values of R and the  $K_p$  of the transistor so that  $I_D$  = 0.1 mA and  $V_{SD}$  = 2.5 V. Justify the state of the transistor. Show your work.

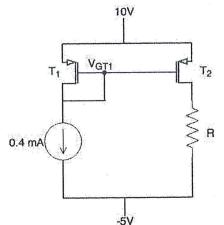


$$V_{SG} = V_{SO} \implies V_{SG} + V_{TP} \implies Tr is SHT or OFF$$

Since  $I_{0} = 0.1 \text{ mat } > 0$ 
 $T_{0} = K_{p} (V_{SG} - 0.7)^{2}$ 
 $0.1 = K_{p} (2.\Gamma - 0.7)^{2} = K_{p} \times 1.8^{2}$ 
 $K_{p} = \frac{0.1}{(1.8)^{2}} = 0.0309 \text{ mA}/U^{2}$ 
 $Q - V_{SD} - I_{D} \times R = 0 \implies Q - 2.\Gamma = 0.1 R$ 
 $R = \frac{6.\Gamma}{0.1} = 6.\Gamma K$ 

(b) In the circuit below  $T_1$  and  $T_2$  are identical (matched pair) with  $V_{TP} = -1$  V and  $K_p = 0.1$  mA/V<sup>2</sup>.

Find V<sub>GT1</sub>. Find the range of R for the current mirror to operate properly.



T1 in SAT:  $0.4 = 0.1 (V_5 G_1^{-1})^2$   $4 = (V_5 G_1^{-1})^2 \Rightarrow \pm 2 = V_5 G_1^{-1}$   $V_5 G_1 = 1 \pm 2$   $0.7 \times G_1 = 3 \times 1$   $0.7 \times G_2 = 3 \times 1$  $0.7 \times G_1 = 3 \times 1$ 

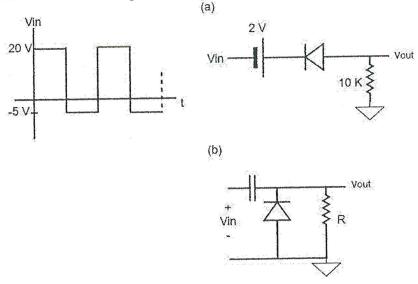
Assume Tz is SAT: => IDz = 0.4 mA. => VSOZ = 10 - (0.4 R-5) = 15-0.4 R

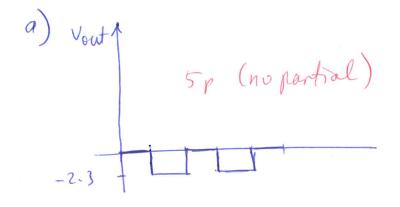
We need  $VSD_2 > VSG_2 + VTP$ => 15-0.4 R > 3-1

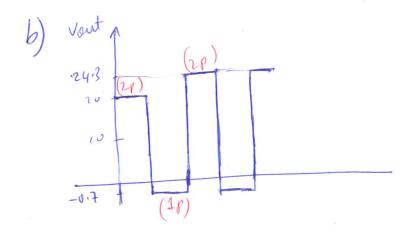
13 > 0.4 R

R <  $\frac{13}{0.4} = 32.5$  K.

Q6. For the input signal  $V_{in}$  as drawn below, find and draw the output signals,  $V_{out}$ , for the circuits in (a) and (b). The cutin voltage for the diodes is  $V_{\gamma} = 0.7 \text{ V}$ . Show your work. Initial value of the capacitor voltage is zero. R in (b) is very large.







### Table 3.1 Summary of the MOSFET current-voltage relationships

# NMOS Nonsaturation region $(v_{DS} < v_{DS}(sat))$

 $i_D = K_n[2(v_{GS} - V_{TN})v_{DS} - v_{DS}^2]$ Saturation region  $(v_{DS} > v_{DS}(\text{sat}))$ 

 $i_D = K_n (v_{GS} - V_{TN})^2$ 

Transition point

 $v_{DS}(\text{sat}) = v_{GS} - V_{TN}$ 

Enhancement mode

 $V_{TN} > 0$ 

Depletion mode

 $V_{TN} < 0$ 

#### **PMOS**

Nonsaturation region ( $v_{SD} < v_{SD}(\text{sat})$ )

 $i_D = K_P[2(v_{SG} + V_{TP})v_{SD} - v_{SD}^2]$ 

Saturation region  $(v_{SD} > v_{SD}(\text{sat}))$ 

 $i_D = K_P (v_{SG} + V_{TP})^2$ 

Transition point

 $v_{SD}(\text{sat}) = v_{SC} + V_{TP}$ 

Enhancement mode

 $V_{TP} < 0$ 

Depletion mode

 $V_{TP} > 0$ 

### Equation for diode current is:

$$I_D = I_S \left[ e^{\left(\frac{V_D}{V_T}\right)} - 1 \right]$$