

- For AWGN channel, $\underline{h} = \delta[n] = [1]$, $\underline{Q} = \underline{I}_{64 \times 64}$ (assume 64 subcarriers)

$$\underline{y} = \underline{Q} \cdot \underline{x} + \underline{w} \quad , \quad \underline{x}: \text{An OFDM symbol}, \underline{y}: \text{received signal}, \underline{w} \sim \mathcal{CN}(0, 2\sigma^2 \underline{I})$$

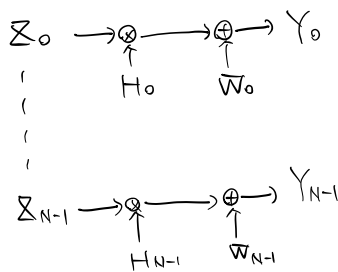
After FFT: ($\underline{\tilde{F}}$: Unitary FFT matrix)

$$\underline{Y} = \underline{\tilde{F}} \underline{y} = \underline{\tilde{F}} (\underline{Q} \underline{x} + \underline{w}) = \underline{\tilde{F}} (\underline{\tilde{F}}^H \underline{H} \underline{\tilde{F}} \underline{x} + \underline{w}) = \underline{H} \underline{x} + \underline{\bar{w}}$$

where \underline{H} is a diagonal matrix, where $\underline{H} = \text{diag}(\text{DFT}\{\underline{h}\})$

$$\left(\begin{array}{l} \text{Find } \underline{\bar{w}} \text{'s mean and covariance matrix:} \\ \underline{w} \sim \mathcal{CN}(0, 2\sigma^2 \underline{I}), \underline{\bar{w}} = \underline{\tilde{F}} \underline{w}, \text{ then mean: } E[\underline{\bar{w}}] = E[\underline{\tilde{F}} \underline{w}] = \underline{0}, \\ \text{Covariance matrix: } K_{\underline{\bar{w}}} = E[\underline{\bar{w}} \underline{\bar{w}}^H] = E[\underline{\tilde{F}} \underline{w} \underline{w}^H \underline{\tilde{F}}^H] = \underline{\tilde{F}} 2\sigma^2 \underline{I} \underline{\tilde{F}}^H = \underline{\tilde{F}} \underline{\tilde{F}}^H 2\sigma^2 \underline{I} = 2\sigma^2 \underline{I} \end{array} \right)$$

Since \underline{H} is always diagonal and $\underline{\bar{w}}$ is independent between each channel, every channel can detect independent:



For BPSK, $x_i = \pm 1$, by ML:

$$\begin{array}{l} \frac{Y_i}{H_i} > 0 \rightarrow +1 \\ \frac{Y_i}{H_i} < 0 \rightarrow -1 \end{array}$$

↑
One-tap equalizer

Since we only care about real part, define $\text{SNR} \equiv \frac{E[|x_i|^2]}{E[|\text{Real}\{\bar{w}_i\}|^2]} = \frac{1}{\sigma^2}$

In AWGN, $H_i = 1$, every subcarrier has same BER: $\frac{Y_i}{H_i} = x_i + \bar{w}_i$

$$P[\mathcal{E}_i] = \frac{1}{2} \int_{-\infty}^0 \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{(p-1)^2}{2\sigma^2}} dp + \frac{1}{2} \int_0^{\infty} \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{(p+1)^2}{2\sigma^2}} dp$$

$$= \int_{-\infty}^0 \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{(p-1)^2}{2\sigma^2}} dp = \int \frac{1}{\frac{1}{\sigma}} \frac{1}{\sqrt{2\pi}} e^{-\frac{\alpha^2}{2}} d\alpha = Q\left(\frac{1}{\sigma}\right)$$

$$= Q(\sqrt{\text{SNR}}) = \text{Average SNR} : \text{Same as BPSK}$$

- In Multipath channel, H_i is different in different subcarrier :

$$\frac{Y_i}{H_i} = \sum \bar{x}_i + \frac{\bar{W}_i}{H_i}, \text{ result in different SNR and BER}$$

$$\text{If define } \text{SNR}_i = \frac{E[|x_i|^2]}{E[|\text{Real}\{\frac{\bar{W}_i}{H_i}\}|^2]} = \frac{1}{\frac{1}{|H_i|^2} E[|\text{Real}\{\bar{W}_i\}|^2]} = |H_i|^2 \frac{1}{\sigma^2}$$

$$\text{then BER of } i\text{th channel} = Q(\sqrt{\text{SNR}_i}) = Q\left(\frac{|H_i|}{\sigma}\right)$$

$$\text{Average BER} = \frac{1}{N} \sum_{i=0}^{N-1} Q\left(\frac{|H_i|}{\sigma}\right)$$

$$\left(\begin{array}{l} \cdot \text{ If we let } \bar{W} = A + Bj, \text{ where } A = B \sim N(0, \sigma^2), \text{ and let } H = a + bj \\ \frac{\bar{W}}{H} = \frac{(A+Bj)(a-bj)}{(a+bj)(a-bj)} = \frac{Aa+Bb}{a^2+b^2} + \frac{-Ab+Ba}{a^2+b^2} j, \text{ Real}\left\{\frac{\bar{W}}{H}\right\} = \frac{Aa+Bb}{a^2+b^2} \\ \text{Var}\left(\frac{Aa+Bb}{a^2+b^2}\right) = \frac{a^2}{(a^2+b^2)^2} \text{Var}(A) + \frac{b^2}{(a^2+b^2)^2} \text{Var}(B) = \frac{a^2+b^2}{(a^2+b^2)^2} \text{Var}(A) \\ \quad \quad \quad \uparrow \\ \quad \quad \quad \text{if } A \text{ and } B \text{ are i.i.d.} \\ = \frac{1}{a^2+b^2} \text{Var}(A) = \frac{1}{|H|^2} \text{Var}(A) \end{array} \right)$$

$$\begin{bmatrix} 1 \\ 1.5 \\ 2 \\ 1 \end{bmatrix}$$

$$[1 \ 2 \ 3 \ 4 \ 5 \ 6 \ 7 \ 8 \ 9 \ 10 \ \dots]$$

$$\begin{bmatrix} 1x \ 1234 \dots 10 \\ 15x \ 1234 \dots 10 \end{bmatrix}$$

64x25