

# ECE 102 HW3

SANJIT SARDA

TOTAL POINTS

**92 / 100**

## QUESTION 1

### Linear Systems 15 pts

1.1 1 a) 5 / 5

- ✓ - 0 pts Correct
- 2 pts The system is linear

1.2 1 b) 5 / 5

- ✓ - 0 pts Correct
- 3 pts The system is linear

1.3 1 c) 5 / 5

- ✓ - 0 pts Correct
- 3 pts The system is not linear

## QUESTION 2

### LTI Systems 20 pts

2.1 2 a) 10 / 10

- ✓ - 0 pts Correct
- 4 pts Partial Correct
- 7 pts Missing sketch
- 10 pts Missing/ Incorrect

2.2 2 b) 10 / 10

- ✓ - 0 pts Correct
- 4 pts Correct  $h(t)$  but incorrect/ unsimplified  $y_2(t)$
- 6 pts Final ans for  $y_2(t)$  missing
- 10 pts Missing/ Incorrect

## QUESTION 3

### Convolution 36 pts

3.1 3 a) 9 / 12

- 0 pts Correct
- ✓ - 2 pts 2 partially correct

✓ - 1 pts 1 partially wrong

- 6 pts 2 mostly wrong
- 4 pts 2 partially correct (2 cases missing/wrong)
- 10 pts did not attempt/complete part 2
- 2 pts did not attempt/complete part 1

3.2 3 b) 12 / 12

- ✓ - 0 pts Correct
- 4 pts partially correct
- 12 pts did not attempt

3.3 3 c) 12 / 12

- ✓ - 0 pts Correct
- 2 pts part 2 partially correct
- 4 pts part 2 mostly wrong
- 10 pts part 2 missing
- 1 pts 1 partially correct
- 12 pts did not attempt

## QUESTION 4

### LTI Systems and Impulse response 12 pts

4.1 4 a) 4 / 4

- ✓ - 0 pts Correct
- 2 pts Partial Correct
- 4 pts Incorrect/ Missing

4.2 4 b) 2 / 2

- ✓ - 0 pts Correct
- 1 pts Adder block missing/ incomplete block diagram from  $x(t)$  to  $w(t)$
- 2 pts Incorrect

4.3 4 c) 2 / 2

- ✓ - 0 pts Correct
- 1 pts Partial Correct

- **2 pts** Incomplete

4.4 4 d) **4 / 4**

✓ - **0 pts** Correct

- **2 pts** Partial Correct

- **4 pts** Incorrect/ Missing

#### QUESTION 5

### Python Tasks 17 pts

5.1 5 a) **5 / 10**

- **0 pts** Correct

✓ - **5 pts** Incorrect convolution (improper inputs to  
nconv)

- **2 pts** Please show the y axis tick marks

- **10 pts** Missing

5.2 5 b) **7 / 7**

✓ - **0 pts** Correct

- **4 pts** Incorrect convolution (incorrect inputs to  
nconv)

- **2 pts** Should be  $\text{rect}(t) * \text{rect}(t)$  (rectangles both  
centered at the origin)

- **7 pts** Missing

## 1) Linear Systems

$$a) y(t) = x(t)e^{-j\omega t} \approx -x(t)$$

To show linearity:  $S[ax_1(t) + bx_2(t)] =$   
 $a S[x_1(t)] + b S[x_2(t)]$

$$\begin{aligned} \therefore S[ax_1(t) + bx_2(t)] &= (ax_1(t) + bx_2(t)) \cdot e^{-j\omega t} \\ &= (ax_1(t) + bx_2(t)) \cdot e^{-j\omega t} = ax_1(t)e^{-j\omega t} + bx_2(t)e^{-j\omega t} \\ &= a S[x_1(t)] + b S[x_2(t)] \\ &\quad \therefore \text{It is linear} \end{aligned}$$

$$b) y(t) = \int_{-\infty}^{\infty} [x(t)]^2 + x(t) dt$$

$$\therefore S[ax_1(t) + bx_2(t)] \approx a S[x_1(t)] + b S[x_2(t)]$$

$$S[ax_1(t) + bx_2(t)] = \int_{-\infty}^{\infty} [ax_1(t) + bx_2(t)]^2 + ax_1(t) + bx_2(t) dt$$

$$\int_{-\infty}^{\infty} [ax_1(t) + bx_2(t)]^2 + ax_1(t) + bx_2(t) dt = \int_{-\infty}^{\infty} a^2 x_1(t)^2 + 3ax_1(t) + bx_2(t) + b^2 x_2(t)^2 dt$$

$\hookrightarrow \neq \checkmark$

$$a S[x_1(t)] + b S[x_2(t)] = a \int_{-\infty}^{\infty} x_1(t)^2 + x_1(t) dt + b \int_{-\infty}^{\infty} x_2(t)^2 + x_2(t) dt$$

$\therefore$  It is not linear.

$$c) y(t) = e^{x(t)}$$

$$S[ax_1(t) + bx_2(t)] = e^{ax_1(t) + bx_2(t)}$$

$\hookrightarrow \neq \checkmark$

$$a S[x_1(t)] + b S[x_2(t)] = ae^{x_1(t)} + be^{x_2(t)}$$

$\therefore$  It is not linear.

1.11 a) 5 / 5

✓ - 0 pts Correct

- 2 pts The system is linear

## 1) Linear Systems

$$a) y(t) = x(t)e^{-j\omega t} \approx -x(t)$$

To show linearity:  $S[ax_1(t) + bx_2(t)] =$   
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$$\begin{aligned} \therefore S[ax_1(t) + bx_2(t)] &= (ax_1(t) + bx_2(t)) \cdot e^{-j\omega t} \\ &= (ax_1(t) + bx_2(t)) \cdot e^{-j\omega t} = ax_1(t)e^{-j\omega t} + bx_2(t)e^{-j\omega t} \\ &= a S[x_1(t)] + b S[x_2(t)] \\ &\quad \therefore \text{It is linear} \end{aligned}$$

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$$\therefore S[ax_1(t) + bx_2(t)] \approx a S[x_1(t)] + b S[x_2(t)]$$

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$$\int_{-\infty}^{\infty} [ax_1(t) + bx_2(t)]^2 + ax_1(t) + bx_2(t) dt = \int_{-\infty}^{\infty} a^2 x_1(t)^2 + 2ax_1(t)bx_2(t) + b^2 x_2(t)^2 + ax_1(t) + bx_2(t) dt$$

$$a S[x_1(t)] + b S[x_2(t)] = a \int_{-\infty}^{\infty} x_1(t)^2 + x_1(t) dt + b \int_{-\infty}^{\infty} x_2(t)^2 + x_2(t) dt$$

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$$c) y(t) = e^{x(t)}$$

$$S[ax_1(t) + bx_2(t)] = e^{ax_1(t) + bx_2(t)}$$

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$\therefore$  It is not linear.

1.2 1 b) 5 / 5

✓ - 0 pts Correct

- 3 pts The system is linear

## 1) Linear Systems

$$a) y(t) = x(t)e^{-j\omega t} \approx -x(t)$$

To show linearity:  $S[ax_1(t) + bx_2(t)] =$   
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$$\int_{-\infty}^{\infty} [ax_1(t) + bx_2(t)]^2 + ax_1(t) + bx_2(t) dt = \int_{-\infty}^{\infty} a^2 x_1(t)^2 + 3ax_1(t) + bx_2(t) + b^2 x_2(t)^2 dt$$

$\hookrightarrow \neq \checkmark$

$$a S[x_1(t)] + b S[x_2(t)] = a \int_{-\infty}^{\infty} x_1(t)^2 + x_1(t) dt + b \int_{-\infty}^{\infty} x_2(t)^2 + x_2(t) dt$$

$\therefore$  It is not linear.

$$c) y(t) = e^{x(t)}$$

$$S[ax_1(t) + bx_2(t)] = e^{ax_1(t) + bx_2(t)}$$

$\hookrightarrow \neq \checkmark$

$$a S[x_1(t)] + b S[x_2(t)] = ae^{x_1(t)} + be^{x_2(t)}$$

$\therefore$  It is not linear.

1.3 1 c) 5 / 5

✓ - 0 pts Correct

- 3 pts The system is not linear



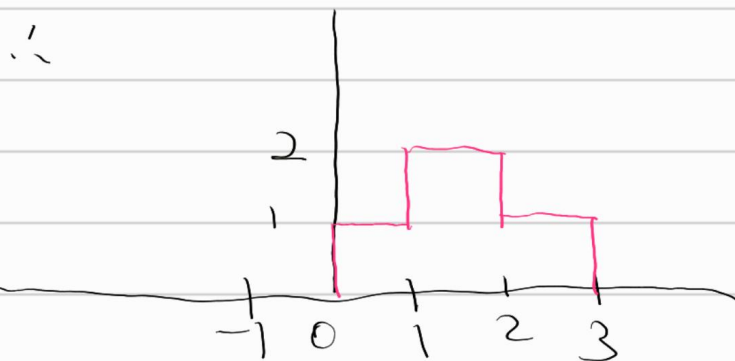
$$2) a) y(t) = S[x_1(t)]$$

$$\therefore \text{rect}\left(\frac{1}{2}(t-2)\right) = S[x_1(t)]$$

$$\text{by Observation, } x_2(t) = x_1(t) + x_1(t+1)$$

$$\therefore y_2(t) = S[x_2(t)] = S[x_1(t) + x_1(t+1)] \text{ since LTI?}$$

$$\therefore y_2(t) = \text{rect}\left(\frac{1}{2}(t-2)\right) + \text{rect}\left(\frac{1}{2}(t-1)\right)$$



$$b) y(t) = S[x(t)]$$

$$a^t \cos(t) = S[u(t)]$$

$$\therefore S[0.5 \delta(t+1) + \delta(t-1)]$$

$$= S\left[0.5 \frac{d}{dt} (u(t+1) + u(t-1))\right]$$

$$= 0.5 \frac{d}{dt} [S[u(t+1) + u(t-1)]]$$

$$= 0.5 \frac{d}{dt} (a^{t+1} \cos(t+1) + a^{t-1} \cos(t-1)) =$$

$$0.5 [a^{t+1} \ln(a) \cdot \cos(t+1) - a^{t+1} \sin(t) + a^{t-1} \ln(a) \cos(t-1) - a^{t-1} \sin(t-1)]$$

Non causal:

The output of the system uses future values!  $\therefore$  it is not causal.

2.1 2 a) 10 / 10

✓ - 0 pts Correct

- 4 pts Partial Correct

- 7 pts Missing sketch

- 10 pts Missing/ Incorrect

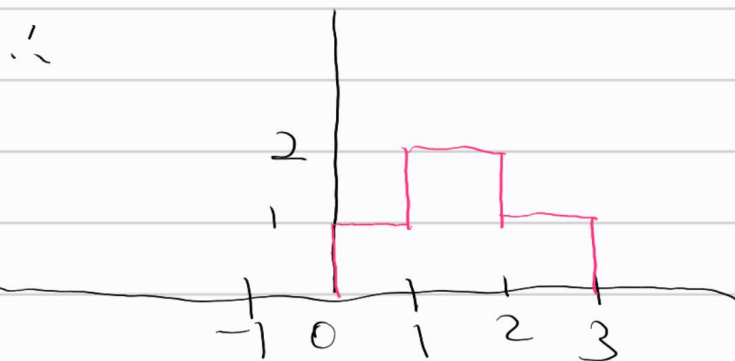
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$$\therefore y_2(t) = \text{rect}\left(\frac{1}{2}(t-2)\right) + \text{rect}\left(\frac{1}{2}(t-1)\right)$$



$$b) y(t) = S[x(t)]$$

$$a^t \cos(t) = S[u(t)]$$

$$\therefore S[0.5 \delta(t+1) + \delta(t-1)]$$

$$= S\left[0.5 \frac{d}{dt} (u(t+1) + u(t-1))\right]$$

$$= 0.5 \frac{d}{dt} [S[u(t+1) + u(t-1)]] =$$

$$= 0.5 \frac{d}{dt} (a^{t+1} \cos(t+1) + a^{t-1} \cos(t-1)) =$$

$$0.5 [a^{t+1} \ln(a) \cdot \cos(t+1) - a^{t+1} \sin(t) + a^{t-1} \ln(a) \cos(t-1) - a^{t-1} \sin(t-1)]$$

Non causal:

The output of the system uses future values!  $\therefore$  it is not causal.

2.2 2 b) 10 / 10

✓ - 0 pts Correct

- 4 pts Correct  $h(t)$  but incorrect/ unsimplified  $y_2(t)$

- 6 pts Final ans for  $y_2(t)$  missing

- 10 pts Missing/ Incorrect

$$3a) f(t) = \delta(t+1) + 5\delta(t-2), g(t) = e^{-t} u(t)$$

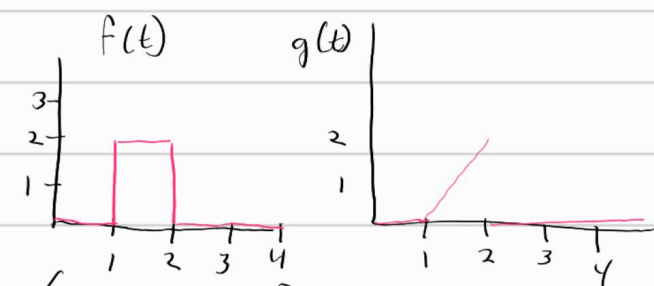
$$f(t) * g(t) = \delta(t+1) * g(t) + 5(\delta(t-2) * g(t)) = g(t+1) + 5g(t-2)$$

↑  
Distributive property

↑  
Delay via convolution

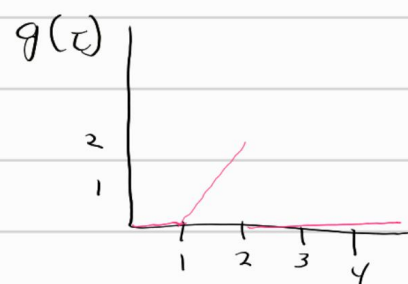
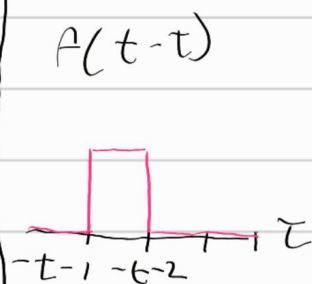
$$f(t) * g(t) = e^{-t-1} u(t+1) + e^{2-t} u(t-2)$$

ii)  $f(t) = 2 \text{rect}(t-3/2)$   $g(t) = 2r(t-1) \text{rect}(t-3/2)$   
 Flip n Drag:

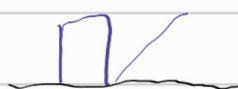


$$(f * g)(t) =$$

$$\int_{-\infty}^{\infty} g(\tau) f(t-\tau) d\tau$$



Case 1:  $t < 2$



$$\int_{-\infty}^{\infty} g(\tau) f(t-\tau) d\tau = 0 \quad \therefore \text{No overlap}$$

Case 3:  $3 \leq t < 4$

$$\int_{-\infty}^{\infty} g(\tau) f(t-\tau) d\tau:$$



$$= \int_2^{t-2} 2(2\tau-2) d\tau$$

$$= 4 \int_2^{t-2} \tau-1 d\tau$$

$$= 4 \left( \frac{\tau^2}{2} - \tau \right) \Big|_2^{t-2} = 2(t-4)(t-2)$$

Case 4:  $4 < t$

$$\int_{-\infty}^{\infty} g(\tau) f(t-\tau) d\tau = 0$$



Case 2:  $2 \leq t < 3$



$$\int_{-\infty}^{\infty} g(\tau) f(t-\tau) d\tau$$

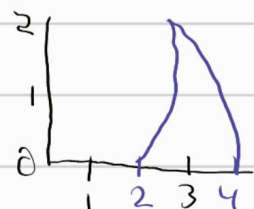
$$= \int_1^{t-1} 2 \cdot (\tau-1) \cdot 2 d\tau$$

$$= 4 \cdot \int_1^{t-1} \tau-1 d\tau = 4 \left( \frac{\tau^2}{2} - \tau \right) \Big|_1^{t-1}$$

$$= 2(t-1)^2 - 4(t-1) + 2$$

Putting it together:

$$f * g(t) = \begin{cases} 2(t-1)^2 - 4(t-1) + 2, & 2 \leq t < 3 \\ 2(t-4)(t-2), & 3 \leq t < 4 \\ 0, & \text{else} \end{cases}$$



3.13 a) 9 / 12

- 0 pts Correct
- ✓ - 2 pts 2 partially correct
- ✓ - 1 pts 1 partially wrong
- 6 pts 2 mostly wrong
- 4 pts 2 partially correct (2 cases missing/wrong)
- 10 pts did not attempt/complete part 2
- 2 pts did not attempt/complete part 1

$$b) y(t) = \int_{t-T}^t (t-\tau)^2 x(\tau) d\tau$$

$$\therefore \text{Impulse Response} = \int_{t-T}^t (t-\tau)^2 \delta(\tau) d\tau$$

$$\therefore x(\tau) \delta(\tau) = x(0) \delta(\tau) = t^2 \delta(\tau)$$

$$\therefore y(t) = \int_{t-T}^t t^2 \delta(\tau) d\tau = \begin{cases} t^2, & 0 < t < T \\ 0, & \text{else} \end{cases}$$

$$\therefore = t^2 \begin{cases} 1, & 0 < t < T \\ 0, & \text{else} \end{cases} = \boxed{t^2 \cdot \text{rect}\left(\frac{x}{T} - \frac{1}{2}\right)} \quad h(t)$$

$$c) i) e^t * \sum_{k=0}^{\infty} \delta(t-k)$$

$$= \sum_{k=0}^{\infty} e^t * \delta(t-k) = \sum_{k=0}^{\infty} e^{t-k} = e^t \sum_{k=0}^{\infty} \left(\frac{1}{e}\right)^k = \frac{e^t}{1 - \frac{1}{e}}$$

$$= e^t \cdot \frac{1}{(e-1)e^{-1}} = \frac{e^t \cdot e}{e-1} = \frac{e^{t+1}}{e-1}$$

$$ii) \frac{d}{dt} [(v(t) - v(t-1)) * v(t-2)]$$

$$\begin{aligned} &\rightarrow (v(t) - v(t-1)) * v(t-2) \\ &= v(t) * v(t-2) - v(t-1) * v(t-2) \\ &= r(t-2) - r(t-1-2) \\ &= r(t-2) - r(t-3) \end{aligned}$$

$$v(t) * v(t) = \int_{-\infty}^{\infty} v(\tau) v(t-\tau) d\tau$$

$$\begin{aligned} &\frac{d}{dx} (r(t-2) - r(t-3)) \\ &= v(t-2) - v(t-3) \end{aligned}$$

Case 1:

$x < 0, (v * v)(t) = 0$  : No overlap

$$x \geq 0, (v * v)(t) = \int_0^t 1 \cdot 1 d\tau = t$$

$$\therefore = \begin{cases} t, & x \geq 0 \\ 0, & \text{else} \end{cases} = \text{rel}v(t)$$

3.2 3 b) 12 / 12

✓ - 0 pts Correct

- 4 pts partially correct

- 12 pts did not attempt



$$b) y(t) = \int_{t-T}^t (t-\tau)^2 x(\tau) d\tau$$

$$\therefore \text{Impulse Response} = \int_{t-T}^t (t-\tau)^2 \delta(\tau) d\tau$$

$$\therefore x(\tau) \delta(\tau) = x(0) \delta(\tau) = t^2 \delta(\tau)$$

$$\therefore y(t) = \int_{t-T}^t t^2 \delta(\tau) d\tau = \begin{cases} t^2, & 0 < t < T \\ 0, & \text{else} \end{cases}$$

$$\therefore = t^2 \begin{cases} 1, & 0 < t < T \\ 0, & \text{else} \end{cases} = \boxed{t^2 \cdot \text{rect}\left(\frac{x}{T} - \frac{1}{2}\right)} \quad h(t)$$

$$c) i) e^t * \sum_{k=0}^{\infty} \delta(t-k)$$

$$= \sum_{k=0}^{\infty} e^t * \delta(t-k) = \sum_{k=0}^{\infty} e^{t-k} = e^t \sum_{k=0}^{\infty} \left(\frac{1}{e}\right)^k = \frac{e^t}{1 - \frac{1}{e}}$$

$$= e^t \cdot \frac{1}{(e-1)e^{-1}} = \frac{e^t \cdot e}{e-1} = \frac{e^{t+1}}{e-1}$$

$$ii) \frac{d}{dt} [(v(t) - v(t-1)) * v(t-2)]$$

$$\begin{aligned} & \rightarrow (v(t) - v(t-1)) * v(t-2) \\ & = v(t) * v(t-2) - v(t-1) * v(t-2) \\ & = r(t-2) - r(t-1-2) \\ & = r(t-2) - r(t-3) \end{aligned}$$

$$v(t) * v(t) = \int_{-\infty}^{\infty} v(\tau) v(t-\tau) d\tau$$

$$\begin{aligned} & \frac{d}{dx} (r(t-2) - r(t-3)) \\ & = v(t-2) - v(t-3) \end{aligned}$$

Case 1:

$x < 0, (v * v)(t) = 0$  : No overlap

$$x \geq 0, (v * v)(t) = \int_0^t 1 \cdot 1 d\tau = t$$

$$\therefore = \begin{cases} t, & x \geq 0 \\ 0, & \text{else} \end{cases} = \text{rel}v(t)$$

### 3.3 3 c) 12 / 12

✓ - **0 pts** Correct

- **2 pts** part 2 partially correct
- **4 pts** part 2 mostly wrong
- **10 pts** part 2 missing
- **1 pts** 1 partially correct
- **12 pts** did not attempt

$$w(t) = H(\delta(t))$$

4)

- $S_1: y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau;$
- $S_2: y(t) = \int_{-\infty}^{t-2} x(\tau) d\tau$
- $S_3$  is characterized by its impulse response:  $h_3(t) = \delta(t-3).$

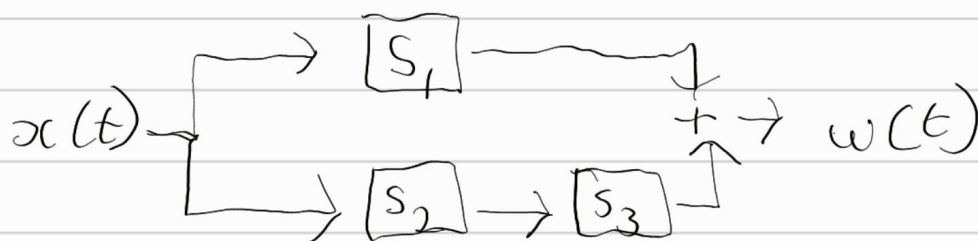
(a) (4 points) Compute the impulse response  $h_1(t)$  of  $S_1$ .

$$a) \quad h_1(t) = S_1[\delta(t)]$$

$$h_1(t) = \int_{-\infty}^t e^{-3(t-\tau)} \delta(\tau) d\tau = e^{-3t} \int_{-\infty}^t \delta(\tau) d\tau = e^{-3t} u(t)$$

b) (2 points) Define  $w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$ . Represent this relationship using a block diagram where  $x(t)$  is the input and  $w(t)$  is the output.

$$w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$$



$$c) \quad S_{eq} = S_1[\text{signal}] - S_3[S_2[\text{signal}]]$$

$$h_1(t) = e^{-3t} u(t), \quad h_2(t) = \int_{-\infty}^{t-2} \delta(\tau) d\tau = u(t-2)$$

$$\therefore h_{eq} = e^{-3t} u(t) - S_3[S_2[\delta(t)]] = e^{-3t} u(t) - S_3[u(t-2)]$$

$\swarrow$  simpl.

$$\therefore h_{eq} = e^{-3t} u(t) - \delta(t-3) * u(t-2) = e^{-3t} u(t) - u(t-3-2)$$

$$\therefore h_{eq}(t) = e^{-3t} u(t) - u(t-5)$$

$$d) \quad h_y(t) = S_{eq}[\delta(t) + 2\delta(t-3)]$$

$$\therefore h_y(t) = \delta(t) * h_{eq}(t) + 2\delta(t-3) * h_{eq}(t)$$

$$= e^{-3t} u(t) - u(t-5) + 2(e^{-3(t-3)} u(t-3) - u(t-8))$$

4.14 a) 4 / 4

✓ - 0 pts Correct

- 2 pts Partial Correct

- 4 pts Incorrect/ Missing

$$w(t) = H(\delta(t))$$

4)

- $S_1: y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau;$
- $S_2: y(t) = \int_{-\infty}^{t-2} x(\tau) d\tau$
- $S_3$  is characterized by its impulse response:  $h_3(t) = \delta(t-3).$

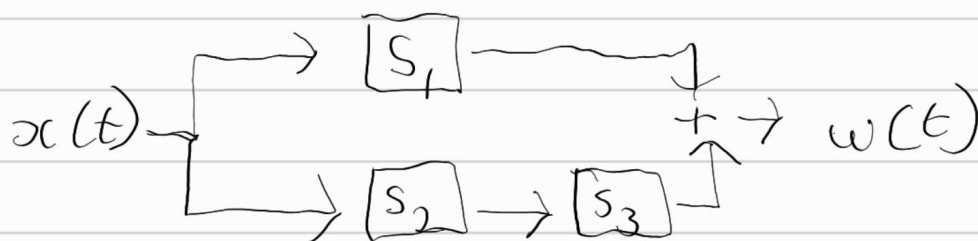
(a) (4 points) Compute the impulse response  $h_1(t)$  of  $S_1$ .

$$a) \quad h_1(t) = S_1[\delta(t)]$$

$$h_1(t) = \int_{-\infty}^t e^{-3(t-\tau)} \delta(\tau) d\tau = e^{-3t} \int_{-\infty}^t \delta(\tau) d\tau = e^{-3t} u(t)$$

b) (b) (2 points) Define  $w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$ . Represent this relationship using a block diagram where  $x(t)$  is the input and  $w(t)$  is the output.

$$w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$$



$$c) \quad S_{eq} = S_1[\text{signal}] - S_3[S_2[\text{signal}]]$$

$$h_1(t) = e^{-3t} u(t), \quad h_2(t) = \int_{-\infty}^{t-2} \delta(\tau) d\tau = u(t-2)$$

$$\therefore h_{eq} = e^{-3t} u(t) - S_3[S_2[\delta(t)]] = e^{-3t} u(t) - S_3[u(t-2)]$$

$\swarrow$   $S_1 \rightarrow h_1$

$$\therefore h_{eq} = e^{-3t} u(t) - \delta(t-3) * u(t-2) = e^{-3t} u(t) - u(t-3-2)$$

$$\therefore h_{eq}(t) = e^{-3t} u(t) - u(t-5)$$

$$d) \quad h_y(t) = S_{eq}[\delta(t) + 2\delta(t-3)]$$

$$\therefore h_y(t) = \delta(t) * h_{eq}(t) + 2\delta(t-3) * h_{eq}(t)$$

$$= e^{-3t} u(t) - u(t-5) + 2(e^{-3(t-3)} u(t-3) - u(t-8))$$

4.2 4 b) 2 / 2

✓ - 0 pts Correct

- 1 pts Adder block missing/ incomplete block diagram from  $x(t)$  to  $w(t)$

- 2 pts Incorrect

$$w(t) = H(\delta(t))$$

4)

- $S_1: y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau;$
- $S_2: y(t) = \int_{-\infty}^{t-2} x(\tau) d\tau$
- $S_3$  is characterized by its impulse response:  $h_3(t) = \delta(t-3).$

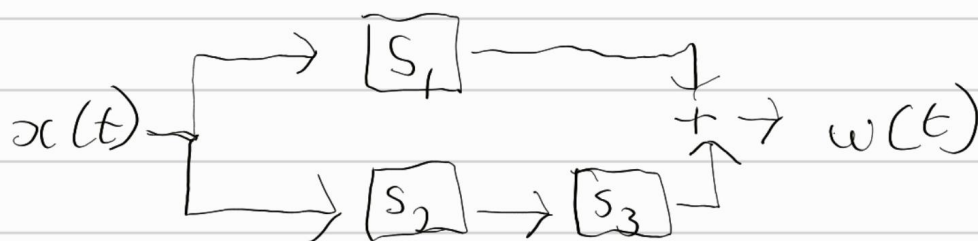
(a) (4 points) Compute the impulse response  $h_1(t)$  of  $S_1$ .

$$a) \quad h_1(t) = S_1[\delta(t)]$$

$$h_1(t) = \int_{-\infty}^t e^{-3(t-\tau)} \delta(\tau) d\tau = e^{-3t} \int_{-\infty}^t \delta(\tau) d\tau = e^{-3t} u(t)$$

b) (b) (2 points) Define  $w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$ . Represent this relationship using a block diagram where  $x(t)$  is the input and  $w(t)$  is the output.

$$w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$$



$$c) \quad S_{eq} = S_1[\text{signal}] - S_3[S_2[\text{signal}]]$$

$$h_1(t) = e^{-3t} u(t), \quad h_2(t) = \int_{-\infty}^{t-2} \delta(\tau) d\tau = u(t-2)$$

$$\therefore h_{eq} = e^{-3t} u(t) - S_3[S_2[\delta(t)]] = e^{-3t} u(t) - S_3[u(t-2)]$$

$\swarrow$   $S_1 \rightarrow h_1$

$$\therefore h_{eq} = e^{-3t} u(t) - \delta(t-3) * u(t-2) = e^{-3t} u(t) - u(t-3-2)$$

$$\therefore h_{eq}(t) = e^{-3t} u(t) - u(t-5)$$

$$d) \quad h_y(t) = S_{eq}[\delta(t) + 2\delta(t-3)]$$

$$\therefore h_y(t) = \delta(t) * h_{eq}(t) + 2\delta(t-3) * h_{eq}(t)$$

$$= e^{-3t} u(t) - u(t-5) + 2(e^{-3(t-3)} u(t-3) - u(t-8))$$

4.3 4 c) 2 / 2

✓ - **0 pts** Correct

- **1 pts** Partial Correct

- **2 pts** Incomplete



$$w(t) = H(\delta(t))$$

4)

- $S_1: y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau;$
- $S_2: y(t) = \int_{-\infty}^{t-2} x(\tau) d\tau$
- $S_3$  is characterized by its impulse response:  $h_3(t) = \delta(t-3).$

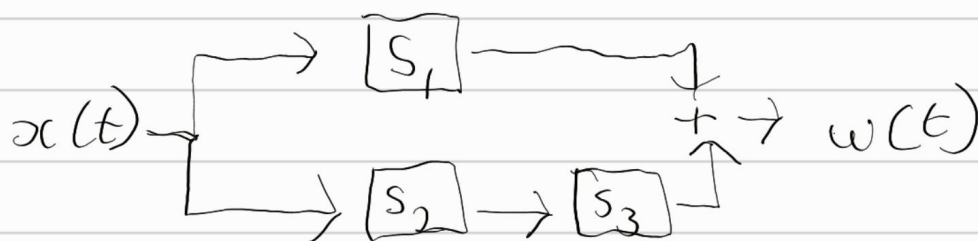
(a) (4 points) Compute the impulse response  $h_1(t)$  of  $S_1$ .

$$a) \quad h_1(t) = S_1[\delta(t)]$$

$$h_1(t) = \int_{-\infty}^t e^{-3(t-\tau)} \delta(\tau) d\tau = e^{-3t} \int_{-\infty}^t \delta(\tau) d\tau = e^{-3t} u(t)$$

b) (2 points) Define  $w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$ . Represent this relationship using a block diagram where  $x(t)$  is the input and  $w(t)$  is the output.

$$w(t) = S_1[x(t)] - S_3\{S_2[x(t)]\}$$



$$c) \quad S_{eq} = S_1[\text{signal}] - S_3[S_2[\text{signal}]]$$

$$h_1(t) = e^{-3t} u(t), \quad h_2(t) = \int_{-\infty}^{t-2} \delta(\tau) d\tau = u(t-2)$$

$$\therefore h_{eq} = e^{-3t} u(t) - S_3[S_2[\delta(t)]] = e^{-3t} u(t) - S_3[u(t-2)]$$

$\swarrow$   $S_1 \rightarrow h_1$

$$\therefore h_{eq} = e^{-3t} u(t) - \delta(t-3) * u(t-2) = e^{-3t} u(t) - u(t-3-2)$$

$$\therefore h_{eq}(t) = e^{-3t} u(t) - u(t-5)$$

$$d) \quad h_y(t) = S_{eq}[\delta(t) + 2\delta(t-3)]$$

$$\therefore h_y(t) = \delta(t) * h_{eq}(t) + 2\delta(t-3) * h_{eq}(t)$$

$$= e^{-3t} u(t) - u(t-5) + 2(e^{-3(t-3)} u(t-3) - u(t-8))$$

4.4 4 d) 4 / 4

✓ - 0 pts Correct

- 2 pts Partial Correct

- 4 pts Incorrect/ Missing

## ▼ 5.

We start with nconv where

x : input signal vector

tx: times over which x is defined

h : impulse response vector

th: times over which h is defined

and the outputs are:

y : output signal vector

ty: times over which y is defined.

```
# imports
import matplotlib.pyplot as plt
import numpy as np
from scipy.integrate import odeint
import seaborn
import math

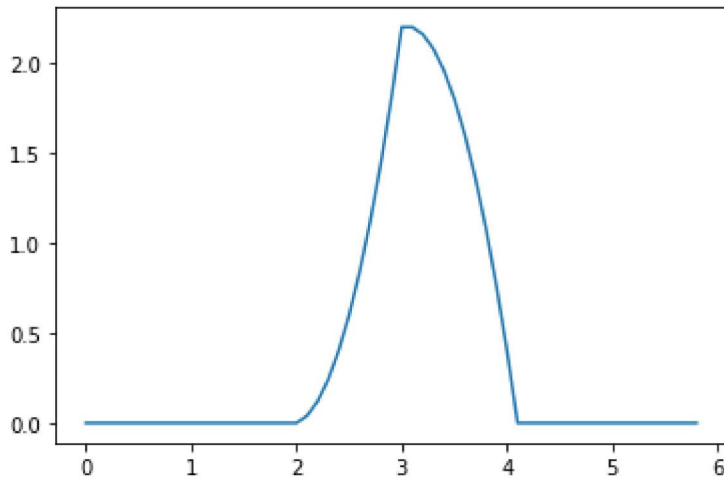
import numpy as np
def nconv(x, tx, h, th):
    y = np.convolve(x, h) * (th[1] - th[0])
    ty = np.linspace(tx[0] + th[0], tx[-1] + th[-1], len(y))
    return y, ty
```

```
# Define Rect and Relu
def rect(t):
    return 1 if abs(t)<= 1/2 else 0
def relu(t):
    return t if t > 0 else 0
```

a) Use nconv() to check your result for problem 3(a)(ii) and plot the output. Use the same step size for tx and th and label the plots.

```
def f(t):
    return 2*rect(t - 3/2)
def g(t):
    return 2*relu(t-1)*rect(t-3/2)
```

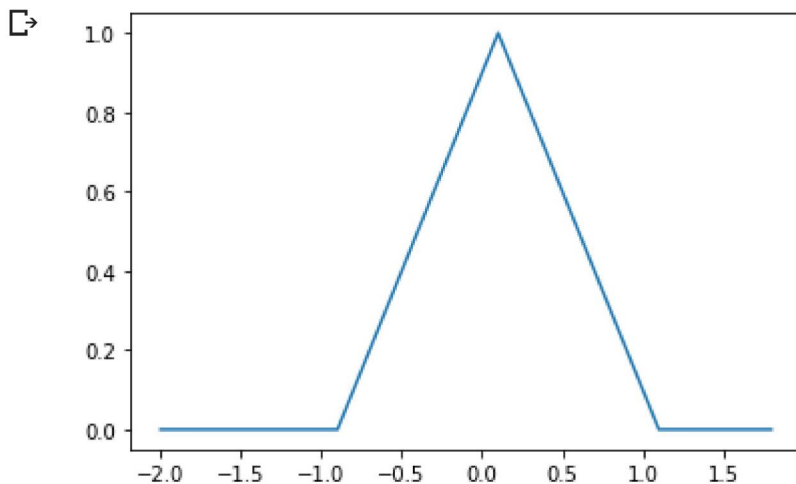
```
t = np.arange(0, 3, 0.1)
ty, y = nconv([f(te) for te in t], t, [g(te) for te in t], t)
plt.plot(y, ty)
plt.show()
```



This looks like the plot of the answer for 3a thus verifying it.

b) Use `nconv()` to convolve two unit rectangles:  $\text{rect}(t) * \text{rect}(t)$ . Plot the result and label the axes.

```
t = np.arange(-1, 1, 0.1)
ty, y = nconv([rect(te) for te in t], t, [rect(te) for te in t], t)
plt.plot(y, ty)
plt.show()
```



5.1 5 a) 5 / 10

- 0 pts Correct

✓ - 5 pts Incorrect convolution (improper inputs to nconv)

- 2 pts Please show the y axis tick marks

- 10 pts Missing

## ▼ 5.

We start with nconv where

x : input signal vector

tx: times over which x is defined

h : impulse response vector

th: times over which h is defined

and the outputs are:

y : output signal vector

ty: times over which y is defined.

```
# imports
import matplotlib.pyplot as plt
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import seaborn
import math

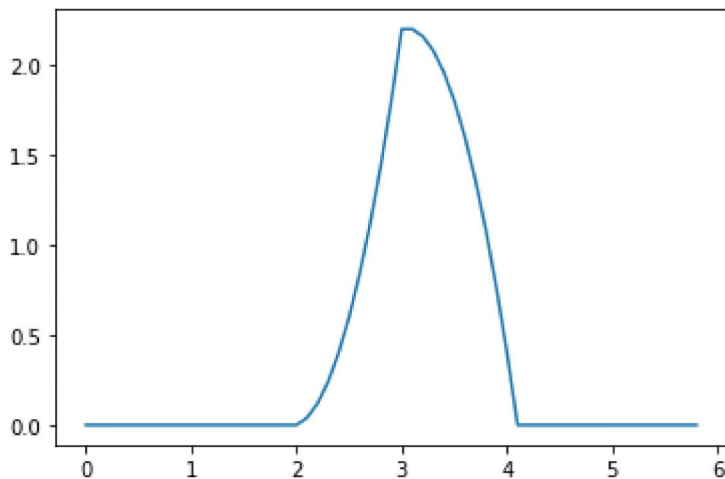
import numpy as np
def nconv(x, tx, h, th):
    y = np.convolve(x, h) * (th[1] - th[0])
    ty = np.linspace(tx[0] + th[0], tx[-1] + th[-1], len(y))
    return y, ty
```

```
# Define Rect and Relu
def rect(t):
    return 1 if abs(t) <= 1/2 else 0
def relu(t):
    return t if t > 0 else 0
```

a) Use nconv() to check your result for problem 3(a)(ii) and plot the output. Use the same step size for tx and th and label the plots.

```
def f(t):
    return 2*rect(t - 3/2)
def g(t):
    return 2*relu(t-1)*rect(t-3/2)
```

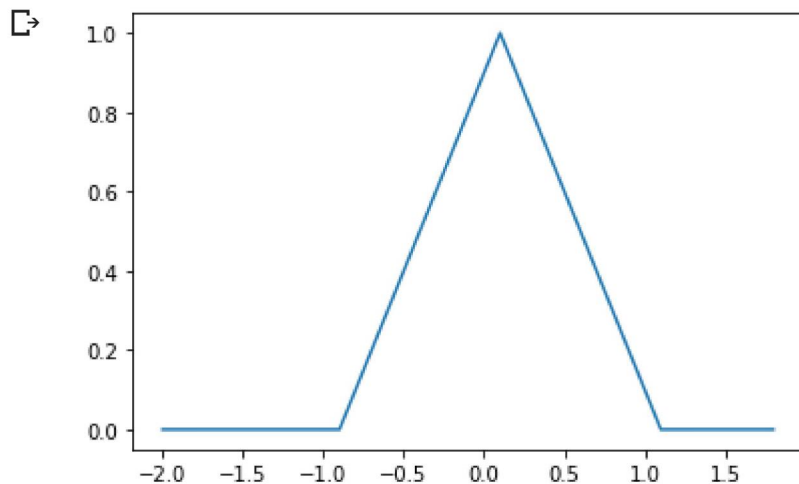
```
t = np.arange(0, 3, 0.1)
ty, y = nconv([f(te) for te in t], t, [g(te) for te in t], t)
plt.plot(y, ty)
plt.show()
```



This looks like the plot of the answer for 3a thus verifying it.

b) Use `nconv()` to convolve two unit rectangles:  $\text{rect}(t) * \text{rect}(t)$ . Plot the result and label the axes.

```
t = np.arange(-1, 1, 0.1)
ty, y = nconv([rect(te) for te in t], t, [rect(te) for te in t], t)
plt.plot(y, ty)
plt.show()
```



## 5.2 5 b) 7 / 7

✓ - 0 pts Correct

- 4 pts Incorrect convolution (incorrect inputs to nconv)

- 2 pts Should be  $\text{rect}(t) * \text{rect}(t)$  (rectangles both centered at the origin)

- 7 pts Missing



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✓ 0s completed at 10:11 PM

