EVERYTHING YOU'VE ALWAYS WANTED TO KNOW ABOUT TIME-HARMONIC FIELDS, BUT WERE AFRAID TO ASK

(A Solution Manual to TIME-HARMONIC ELECTROMAGNETIC FIELDS, by R. F. Harrington, McGraw-Hill Book Co.)

bу

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PREFACE

Most of the problems in "Time Harmonic Electromagnetic Fields" by R. F. Harrington are solved in this manual. The problems not done are 5-43, 5-44, 7-28, 8-11, 8-12, and 8-23. A few problems are referenced to other books or journal articles where the authors feel they are done more extensively than they would have been if presented here. During the course of this work various mistakes in the problems were found and corrections should appear in these solutions.

Finally the authors would like to express their grateful acknowledgement to Dr. Joseph Mautz for his invaluable help when consulted and to Dr. Roger Harrington for encouraging the project.

e- uty

$$\frac{1-2}{J} = \frac{cond}{cond}$$

$$\frac{1}{J} = \frac{cond}{cond}$$

$$\frac{1}{J} = \frac{cond}{cond}$$

$$\frac{1}{J} = \frac{1}{J} = \frac{1}{J}$$

$$\frac{1-5}{\sqrt{3}} = \frac{-3e}{\sqrt{3}} \quad \text{(continuity eqn.)}$$

$$\int 0.5 dV = \int 5 dA = I = dQ$$

$$C = Q/V \text{ at } Q = CV$$

$$\therefore i = \frac{d(CV)}{dt} = C \frac{dV}{dt}$$

$$\int \overline{E} \cdot dI = -\frac{3t}{\sqrt{3}} = -V \quad \text{(Fanaday)}$$

$$L = \frac{4}{\sqrt{3}} \quad \text{(Li)} = L \frac{di}{dt}$$

$$I - Q \quad \text{(I)} = 10 + j5 \quad \text{(place)}$$

$$i = \sqrt{2} \, R \quad \text{(Ieint)}$$

$$= \sqrt{2} \, R \quad \text{(Ieint)}$$

$$i = \sqrt{2} R (Ie^{i\omega t})$$

$$= \sqrt{2} R (Ie^{i\omega t})$$

$$= \sqrt{2} R [(10+i5)(\cos\omega t + i\sin\omega t)]$$

$$= \sqrt{2} [10\cos\omega t - 5\sin\omega t]$$

$$= 0$$

b.)
$$C = \sqrt{2} \operatorname{Re} \left(\operatorname{E} e^{i\omega t} \right)$$

$$= \sqrt{2} \operatorname{Re} \left[\overline{u}_{x} \left(s + i s \right) e^{i\omega t} + \overline{u}_{y} \left(z + i s \right) e^{i\omega t} \right]$$

$$= \sqrt{2} \left[\overline{u}_{x} \left(s \cos \omega t - 3 \sin \omega t \right) + \overline{u}_{y} \left(z \cos \omega t - 3 \sin \omega t \right) \right]$$

$$c.) H = \sqrt{2} \operatorname{Re} \left(\operatorname{He} i \omega t \right)$$

$$- \left[\overline{u}_{x} \left(s \cos \omega t - 3 \sin \omega t \right) \right]$$

$$\frac{1-7 \text{ (cont.)}}{\frac{2}{3\pi} \text{ Re}(A)} = \frac{2a}{3\pi} = \text{Re}\left(\frac{3A}{3\pi}\right)$$

$$\int \text{Re}(A) dx = \int a dx = \text{Re}\left[\int A dx\right]$$

$$1-8 = \frac{7}{3\pi} = \frac{7}{3\pi}$$

$$\frac{1-8}{H} = \overline{u_x} \sin y$$

$$\overline{H} = \sqrt{2} \, \overline{u_x} \sin y \cos \omega t$$

$$\overline{C} = \frac{1}{50} \int \mathbb{R} \times \overline{H} \, dt$$

$$= \int \frac{-\sqrt{2} \cos y \cos \omega t}{50 \, \xi_y} \cot \omega t$$

$$= \frac{\sqrt{2} \cos y \sin \omega t}{50 \, \xi_y \omega}$$

$$\overline{C} = \sqrt{2} \, R_z \left[\overline{E} e^{i \, \omega t} \right]$$

:. E = - j woy

D. ds = (02-5,). A where A is the area of each face Since the region is some free , D, = D2 and thus Asids = 0 = Qv.

E = 4 (1+j) Complex power supplied by Somer = E.I' = 10 (1+j) Time average power = 5 (1+j) Conduction curent = 6 = 5 × 16 - 4 4, Free space displacement jw Mo FF inge = 1 5 x10 4 = नंशाय Polaryation jw (û-40) H= jw (ê- E) E = (5×105+j 4.5×108) & 4x (2x107+j2,6x108) No Uy Displacement するなみ in SE Dissipatrie wm" H (6 +we") B Reading jw E'E iwm'H Induced -JWAH (6+jw 2)E 1-14 C= 300pt C= E'A . for air, C=A = 300 x10-12 Y = G+jwc = (500-j) x103

G= 1 = WE"A - 300 XID E"

$$\frac{1-14 \text{ (cond.)}}{\xi'' = \frac{1}{5 \times 10'' \times 300 \times 10^{-12}} = \frac{1}{150}$$

$$\xi' = \frac{1}{(250)(300)}$$

$$\hat{\gamma} = \omega \xi'' + \hat{\gamma} \omega \xi'$$

Power loss in wire ~ 6E.E#

Power loss in core ~ wp1"H.H*

H, for example, copper is used;

6 = 5,8 ×107

From Fig. 1.12, it is clear-that

6 >> wp1" for all frequencies

shown.

... Power loss in wire >>

Power loss in core.

1-16

Casume $\hat{\xi} = \xi' - j\xi''$ is an analytic function in w and i. ξ', ξ'' satisfy Canchy-Riemann equations.

Cauchy ditegral thin is: $\frac{\hat{\beta}}{z-j\omega} \frac{\hat{\xi}(\omega)}{z-j\omega} dz = 0 , z = x+jy$ $C = C_1 + C_2 + C_3$ $C = C_1 + C_2 + C_3$

1-16 (cont.) The contribution from pole at Z= iw = 217i [+ resider at that ; This is girst & (jw) To compute of let Z=RACIO $\frac{\int \hat{z}(\omega)}{z-i\omega} dz = \int \frac{\hat{z}(\omega)R_0e^{i\phi}}{R_0e^{i\phi}-i\omega} \frac{1}{i}R_0e^{i\phi}$ Roto 1 E(0) 5 do = iTE(0) Since In (E(w)) = 0 at an, €(0) = Re { €(0)} = E0 For the rest of the contour, Roto (14) dy + Seliy dy - Roto y-w $=\int \frac{2(i\eta)}{(i\eta)} d\eta = i\pi [2(i\eta) - 2(i\eta)]$ writing E(iw) and E(ix) in tens of real and imaginary parts, E'(w) = E0 + I / E"(y) dy $\xi''(\omega) = -\frac{1}{\pi} \int \frac{\xi'(y)}{y-\omega} dy$ E'= own function of w E"= odd function of w $E''(\omega) = -\frac{1}{\pi} \int \frac{E'(\gamma)}{\gamma - \omega} d\gamma - \frac{1}{\pi} \int \frac{E'(\gamma)}{\gamma - \omega} d\gamma$ J ε'(y) dy = J ε'(-y) (-dy) = - \(\frac{\xi'(4)}{4+\w} dy

$$\xi''(\omega) = -\frac{1}{\pi} \int_{0}^{\infty} \xi'(y) \left[\frac{1}{y-\omega} - \frac{1}{y+\omega} \right] dy$$

$$= -\frac{2\omega}{2\omega} \int_{0}^{\infty} \xi'(y) \left[\frac{1}{y-\omega} - \frac{1}{y+\omega} \right] dy$$

$$= -\frac{2\omega}{\pi} \int_{0}^{\infty} \frac{\epsilon'(4)}{y^2 - \omega^2} dy$$

which is the first egn. to be proven. E''(w) can be changed to the form given in the text by noting:

$$\varepsilon''(\omega) = -\frac{2}{\pi} \int_{0}^{\infty} \frac{\omega \left[\varepsilon'(y) - \varepsilon_{0} \right] dy}{y^{2} - \omega^{2}} dy$$

1-17 Chatric curents:

$$n \times (\overline{H_1} - \overline{H_2}) = \overline{J_3}$$

magnetic aments:

1-17 (cont.)

$$\oint \vec{E} \cdot \vec{a} = -\int \vec{M} \cdot \vec{A} \vec{A}$$

$$(\vec{E}_1 - \vec{E}_2) \cdot b \cdot \vec{A} = -\vec{M}_S \cdot b \cdot \vec{A}$$

$$\therefore \hat{\Lambda} \times (\vec{E}_2 - \vec{E}_1) = \vec{M}_S$$

$$\Delta \cdot (\vec{E}_1 - \vec{E}_2) \times \hat{\Lambda} = \vec{M}_S$$

$$\frac{2-1}{E_2 = E_0}e^{-ik\mathbf{R}} \quad \text{satisfies} \quad \nabla^2 \overline{E} + k^2 \overline{E} = 0$$
but does not satisfy $\nabla \cdot \overline{E} = 0$

$$\nabla^2 E_2 + k^2 E_2 = (-ik)^2 E_0 e^{-ik^2} + k^2 \overline{E}_0 e^{-ik^2}$$

It does not satisfy VXVXE -KZE =0

= P(P. E) -k2E - P2E #0 (Since P.E 40)

i. This is not a possible electromagnetic field.

$$\frac{2-3}{a_1} = \sqrt{-\hat{s}\hat{q}} = \sqrt{-iw\mu(6+i\omega\xi)}$$

$$k = \sqrt{w^2\mu_0 \& \xi_r} = k_0\sqrt{\xi_r}$$

inhomogeneous media.

$$\frac{2-3(\omega n t_{i})}{b_{i}} = \sqrt{\frac{3}{6}} = \sqrt{\frac{i\omega M_{o}}{i\omega \xi_{o}\xi_{f}}}$$

$$= \sqrt{\frac{M_{o}}{\xi_{o}}} \frac{1}{\sqrt{\xi_{f}}} = \frac{\gamma_{o}}{\sqrt{\xi_{f}}}$$

$$c_{i} = \frac{2\pi}{k} = \frac{2\pi}{k_{o}\sqrt{\xi_{f}}} = \frac{\lambda_{o}}{\sqrt{\xi_{f}}}$$

d.)
$$N_p = \frac{\omega}{R} = \frac{\omega_o}{R_o \sqrt{\xi_r}} = \frac{c_o}{\sqrt{\xi_r}}$$

$$P_{S} = P_{f} + P_{d} + \frac{\partial}{\partial t} (\omega_{e} + \omega_{m})$$

$$M - (\partial_{-} \vec{J}^{i} + \vec{H} \cdot \vec{m}^{i}) = 0.5 + 6e^{2}$$

$$+ \frac{\partial}{\partial t} (\omega_{e} + \omega_{m})$$

$$0.\overline{S} = \frac{4k}{\eta} E_0^2 \omega(\omega t - k_{\bar{s}}) \sin(\omega t - k_{\bar{s}})$$

$$\frac{4k}{\eta} = \frac{4\omega\sqrt{\mu} \sqrt{\xi}}{\sqrt{\mu}} = 4\omega\xi$$

Thus 1 is satisfied for Egn 2-18.

Egno 2-21:

+2wEEZcooZkzsin ut conut

1 becomes :

- kEo coo zkz sin zwt - w EE wowt sinut

+ w EE 2 Coo zkz (coowt sin wt)

+ w EE 2 coo wt sin wt

+ w EE 2 coo zkz sin wt coo wt

= - k Es cos ekz sin zwt + WEED COO ZKZ COO ZWT

=0 since k =wE

Egno. 2-27:

₽. \$ = 0

3 (we+wm) =0

in Agra(1) is Satisfied.

Equo. 2-29:

Q. 3 = 0

.. egn (1) is satisfied. of (wm+we)=0

2-5 relocity of energy propagation = ve

Ne = J We+Wm = Exx

Cx = JZ Eo sinkz wout

Ay = - 12 Eo us ky sinut

We = EE = EE Sin 2 kg cos 2 wit

Wm = MAT = EE 2 coo 2 kg sin 2 wt

d = - Eo sin zkz sin zwt

Ne= - sin zky sin zut /27

 $A = \left\{ \frac{1}{4} \left[1 - \omega_0 z k_{\delta} \right] \left[1 + \omega_0 z \tilde{\omega} t \right] \right\}$

+ 1 [1 + coo zlez][1 - coo zwt]

Ne= - sinzkz sinzwt TIME (1- 600 2 kg coo 2 wt)

sin zkz sinzwit

1- coo zkz coo zwt

= 1- 1-2000 (kg-wt) 1- coo (kz+wt)-coo (kz-wt)

= 1- 2

:. Ne & __

 $\frac{2-6}{E_X} = Ae^{-ikz} + Ce^{ikz}$

= A wokt - jA sinkt + Cwo kz +j Csinkz

=(A+C) cooke +j(C-A) sinke

= ((A+c) 2002 kz + (C-A) 2 sin 2 kz) 2 1 tam 6

= (A2+c2+2AC) con2k2 +(c2+A2-2AC) sin2k2

= (A2+c2+2Ac coo2kz) 1/2 1 tan-10

0 - (C-A) sinke $=-\frac{(A-c)}{(A+c)}\tan kz$ (A+c) cookz

$$\frac{2-6 (cont.)}{e^{i \tan \theta}} = e^{-i \tan^{-1} \frac{(A-C)}{(A+C)} \tan k \epsilon}$$

For phase velocity Np:

$$w - \frac{d}{dt} \left[\frac{(A-c)}{(A+c)} tankz \right] = 0$$

$$w - \frac{(A-c)k \sec^2 k + \nu \rho}{(A+c)} = 0$$

$$1 + \left(\frac{A-c}{A+c}\right)^2 \tan^2 k + \frac{(A-c)^2}{(A+c)^2} \cot^2 k + \frac{(A-c)^2}{(A+c$$

$$\frac{1}{\sqrt{A}} = \frac{1}{\sqrt{A}} \left[\frac{A+C}{A-C} \cos^2 kz + \frac{A-C}{A+C} \sin^2 kz \right]$$

$$\frac{2-7}{H} = (\overline{u_x} + j\overline{u_y}) E_0 \sin kz$$

$$H = (\overline{u_y} + j\overline{u_y}) E_0 \cos kz$$

$$z_{yx} = -\underline{E}_{y} = -\underline{j}E_{o}\sin kz = -\underline{j}\eta \tan kz$$

$$\frac{E_{o}\cos kz}{\eta} = -\underline{j}\eta \tan kz$$

Zxy = Zxx only in isotropic, homogenesus media.

The wave is wisularly polarized

In either case, A = Be 0 = 18

$$\frac{2-9}{\text{Let } A=B=|A|e^{i\pi/2}}$$
Let $A=B=|A|e^{i\pi/2}$

$$E=\left[\overline{u_x} A(i+i) + \overline{u_y} A(i+i)\right] e^{-ikz}$$

$$=\left[\overline{u_x} A + i\overline{u_y} A\right] e^{-ikz}$$
There represent two aircularly

polarized waves and thus E is a uniform plane wave,

$$\frac{2-10}{E} = (\overline{u}_{X}A + \overline{u}_{Y}B) e^{-jkz}$$

$$E = (\overline{u}_{X}A + \overline{u}_{Y}B) e^{-jkz}$$

$$1 \text{ at } A = a + jb$$

$$B = c + jd$$

$$E = \left[(a + jb)\overline{u}_{X} + (c + jd)\overline{u}_{Y} \right] e^{-jkz}$$

$$= \left[(a\overline{u}_{X} + c\overline{u}_{Y}) + j(b\overline{u}_{X} + d\overline{u}_{Y}) \right] e^{-jkz}$$

$$= \left(\overline{E}_{1} + j\overline{E}_{2} \right) e^{-jkz}$$

$$= (\overline{E}_{1} + j\overline{E}_{2}) e^{-jkz}$$
where $\overline{E}_{1} = Re(A)\overline{u}_{X} + Re(B)\overline{u}_{Y}$

$$\overline{E}_{2} = Im(A)\overline{u}_{X} + Im(B)\overline{u}_{Y}$$

$$\frac{2-11}{E} = Re(E) + j Im(E)$$

$$\frac{2}{\pi} = A \cos(k z) e^{j\omega t}$$

$$\frac{2}{\pi} = B \sin(kz) e^{j\omega t}$$

$$\frac{2}{\pi} = C_x C_x^2 = (a^2 + b^2) \cos^2(kz)$$

$$\frac{2}{\pi} = C_y C_y^2 = (c^2 + d^2) \sin^2(kz)$$

$$\frac{(a^2+b^2)}{(a^2+b^2)} + \frac{(c^2+d^2)}{(c^2+d^2)} = 1$$
which shows the elliptic preparty.

If $a = c$, $b = d$; the wave would be circularly planted.

$$\frac{2-12}{a.)} Pohystyren: \mathcal{E}' = \{0, \mathcal{E}'_{k}, \mathcal{E}''_{k}, \mathcal{E}$$

$$7 = \sqrt{\frac{M_0}{\varepsilon_0(\varepsilon' - j\varepsilon'')}} = \frac{1.31 \times 10^8}{\sqrt{x - j\gamma}}$$

	•	70	
1 XIDE	10	100	1000
٤4/٤.	.0003	.0005	,0448
٤'/٤	2.6	2.55	2.55
Jx-14	2.6 6003	2.5 1.0056	2.55 6.009
KXIET	16,2 -j.000\$	159-1.016	1591-1.25
7 × 108	.5 +j.000024	.514 +j5,x10 ⁻⁵	·514 +j 8.x/05

b.) Phrighas:

fx106	10	150	1000
٤٠/٤	2.7	2.67	2.65
٤"/٤,	. 627	.020	.015
- Jx-17	1.64 £.286	1.63 /215	1.627 1162
kx10-7	10.23-1.05	101.7-j.381	1015-12.87
7 x 108	·799+j.004	.804 +j.003	.865 +j.062

C.) Ferrance A:

2-12 (conti)

1			
£ ×106	10	100	1000
" / yo	25	7	1.5
פון "ון	2	10	2
Jx-14	25,08 <u>L2.28</u>	12.2 <u>-27.5</u>	2.5 /-26.6
kx1012	5.89-j.234	25,4-j13.24	52.5~j 26.3
7	2.33 -j,133	1,44-j.749	·297-j.149
	,		

~	0		
4×106	10	100	1000
kx104	4.78-14.78	15.1-j15.1	47.8-147.8
7×10-3	1.17+31.17	3, 28 + 3,28	11.7+11.7
	·		
	I		

$$\frac{2-13}{7} = \sqrt{\frac{3}{9}}$$

and
$$y = \frac{1}{7}$$

$$\eta = \frac{jk}{y} = \frac{jk}{j\omega \varepsilon}$$
 $\varepsilon = c$

$$= \frac{k}{w \varepsilon} = \frac{k' - j k''}{w \varepsilon}$$

$$k = \sqrt{-j w \mu (6 + j w \varepsilon)}$$

$$=\omega\sqrt{\mu\epsilon'}\left[1-3\left(\frac{e+n\epsilon''}{e+n\epsilon''}\right)\right]^{1/2}$$

=
$$\omega \sqrt{\mu \xi'} \left[1 - \frac{1}{2} \frac{1}{Q} + \frac{\frac{1}{2} \left(-\frac{1}{2} \right)}{2!} \left(-\frac{1}{Q^2} \right) \right]$$

$$-\frac{\frac{1}{2}\left(-\frac{1}{2}\right)\left(-\frac{3}{2}\right)}{3!}\frac{\cancel{j}}{\cancel{Q}^{3}}$$

$$\simeq \frac{\omega \, \epsilon''}{2} \sqrt{\frac{\epsilon'}{\epsilon'}} \left(1 - \frac{1}{8Q^2}\right)$$

$$7 = \sqrt{\frac{3}{9}} = \sqrt{\frac{j \omega \mu}{6 + j \omega (\epsilon' - j \epsilon'')}} = \sqrt{\frac{j \omega \mu}{6 + \omega \epsilon'' + j \omega \epsilon'}}$$

$$\frac{2-14 \ (cond.)}{7} = \sqrt{\frac{M}{E'}} \left[\frac{1}{1+\frac{1}{1Q}} \right]^{1/L} \qquad Q = \frac{WE'}{5+WE''}$$

$$= \sqrt{\frac{M}{E'}} \left[1+\frac{1}{1Q} \right]^{-1/2}$$

$$= \sqrt{\frac{M}{E'}} \left[1+\frac{1}{1Q} \right]^{-1/2}$$

$$+ \left(-\frac{1}{2} \right) \left(-\frac{1}{2} - 1 \right) \left(-\frac{1}{2} - 1 \right) \left(-\frac{1}{2} - 1 \right)$$

$$+ \left(-\frac{1}{2} \right) \left(-\frac{1}{2} - 1 \right) \left(-\frac{1}{2} - 1 \right) \left(-\frac{1}{2} - 1 \right)$$

$$+ \left(-\frac{1}{2} \right) \left(-\frac{1}{2} - 1 \right) \left(-\frac{1}{2} - 1 \right) \left(-\frac{1}{2} - 1 \right)$$

$$= \sqrt{\frac{M}{E'}} \left(1-\frac{3}{2Q} \right) \left(1-\frac{5}{2Q^2} \right)$$

$$= \sqrt{\frac{M}{E'}} \left(\frac{1}{2Q} \right) \left(1-\frac{5}{2Q^2} \right)$$

$$= \sqrt{\frac{M}{E'}} \left(\frac{1}{2Q} \right) \left(1-\frac{5}{2Q^2} \right)$$

$$\approx \frac{\xi''}{2\xi'} \sqrt{\frac{M}{\xi'}} \left(1 - \frac{5}{8Q^2} \right)$$

$$k = \sqrt{-j \omega \mu (\sigma + j \omega \varepsilon)}$$

$$= \sqrt{-j \omega \mu (\sigma + j \omega \varepsilon')}$$

$$= \sqrt{-j \omega \mu (\sigma + j \omega \varepsilon')}$$

$$k = k' - j k''$$

$$= \sqrt{-i} \omega_{MG} \left[1 + \frac{i}{2} + \frac{1}{2} \left(\frac{1}{2} - 1 \right) (iq)^{2} \right]$$

$$+\frac{1}{2}(\frac{1}{2}-1)(\frac{1}{2}-2)(10)^{3}$$

$$\frac{2-15 \left(\omega n t_{i} \right)}{k = \sqrt{\omega \mu \delta} \left(\left(\frac{1}{\sqrt{2}} + \frac{\alpha}{2\sqrt{2}} \right) + i \left(-\frac{1}{\sqrt{2}} + \frac{\alpha}{2\sqrt{2}} \right) \right]}$$

$$k' = \sqrt{\frac{\omega \mu \delta}{2}} \left(1 + \frac{\alpha}{2} \right)$$

$$k'' = \sqrt{\frac{\omega\mu\delta}{2}} \left(1 - \frac{\alpha}{2}\right)$$

$$\gamma = \sqrt{\frac{j\omega\mu}{6+j\omega\epsilon}} = \sqrt{\frac{\omega\mu}{6}} \left(\frac{1}{72} + \frac{1}{42}\right) \left[1 + \frac{1}{3}\alpha\right]^{-1/2}$$

$$R = \sqrt{\frac{\omega_{\mu}}{26}} \left(1 + \frac{Q}{2} \right)$$

$$\chi = \sqrt{\frac{\omega_H}{26}} \left(1 - \frac{Q}{2}\right)$$

$$\frac{2-16}{7} = \sqrt{\frac{j\omega\mu}{6+j\omega\xi}} \quad (\xi=0)$$

$$\delta = \frac{1}{\sqrt{\pi 4 \mu 6}} = \frac{\sqrt{2}}{\sqrt{\mu \mu 6}}$$

$$=\frac{1}{r}\left(i-j\right)$$

$$\frac{2-17}{8} = \frac{1}{65} = \sqrt{\frac{1}{12}} = \sqrt{\frac{1}{12}}$$

$$R = \frac{\sqrt{\omega \mu}}{\sqrt{25}} = 1.49 \times 0^{-2} \frac{\sqrt{4}}{\sqrt{5}}$$

Values of 6 used were:

Thus the formulas for the intrisse wars resistance easily obtained.

2-19

For a circularly polarized want in discipating media the amplitude decays exponentially with 2 and Mx lags Ex by TT/2-3.

From Egn 2-47;

From Egn 2-45:

$$\Gamma = \sqrt{\frac{M_0}{\xi_0 \xi_r}} - \sqrt{\frac{M_0}{\xi_0}} = \frac{1 - \sqrt{\xi_r}}{1 + \sqrt{\xi_r}}$$

$$\sqrt{\frac{M_0}{\xi_0 \xi_r}} + \sqrt{\frac{M_0}{\xi_0}} = \frac{1 + \sqrt{\xi_r}}{1 + \sqrt{\xi_r}}$$

Browster angle = Oi = tan / Ei

$$\xi_z = \xi_r \xi_r$$

$$\theta_{\xi} = \omega \sigma^{-1} \left[\frac{\xi_{i}}{\xi_{2}} \sqrt{\frac{\xi_{i}}{\xi_{i} + \xi_{2}}} \right]$$

$$\theta_c = \sin^{-1}\left[\frac{\xi_1 \mu_1}{\xi_2 \mu_2} - \sin^{-1}\left[\frac{\xi_2}{\xi_1} - \sin^{-1}$$

	WATER	GLASS	POLYSTYRENE
٤۴	8।	9	2.54
ð.	83.7°	71.60	58.00
bt	89,3*	83,9°	70.70
٥	4.38*	19.5°	₹ 8 .7°

$$\xi_2 = \xi_{\tau} \xi_{l}$$

angle greater than bi will be totally reflected. Values of & at critical angle are:

Water: ±jx=kcos6,380=.994k

Slass: ±jd= k 600 19,5°= .943 k

Polystyne: + j x = k coo 38,7° = .78k

e-dt => 1 is the value of t

which gives an attenuation of 1/e

water: $z = \frac{1.006}{k} = .16 \lambda_{\omega}$

Slass: $\frac{2}{k} = \frac{1.06}{k} = .17 \lambda q$

Pohystyrene: == 1.28 = , 2 hp

When the I's are the different wandlingths in the various materials.

 $\frac{2-24}{7} = \chi + j\beta = \sqrt{(R+jwL)(6+jwC)}$

= (jwL jwc) 1/2 [(1+ R)(1+ 6)]

= jwILC [1+ R + 6 Ziwc]

R= R + 6 JL/C

Ba WALC

2-25 = R+jwl

7 = 1(R+jwL) 66+jwc)

2-25 (cont.)

Similarly for a slave un

Similarly for a plane wave: $\gamma = \sqrt{2}$ $\gamma = \sqrt{2}$

7=97

6+iwc= \frac{gn}{z_0} = \frac{\gamma}{z_0} iw (\gamma' - i\gamma')

6 = W7E" C = W7E'

6 = we"c

It the frequency is high enough the curvature of the wire is unimportant so for a conductor of thickness of;

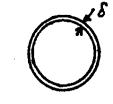
Rh.t. = Rs = 1 ohmo/meter

Roc. = 1 ohmo/miles

Rh.t. = Ro.c. (for wires of thickness = 8).

2-26

a.) $R \approx \frac{1}{\pi dGS}$



which for two

 $R \approx \frac{2}{\pi d 6 \delta}$ but $R = \frac{1}{6 \delta}$

So R = 2R

$$\frac{2-26 \left(\text{cont.}\right)}{\text{b.}}$$

$$R_1 = \frac{1}{2\pi a 66}$$

$$R_2 = \frac{1}{2\pi b 66}$$



$$R = R_1 + R_2 = \frac{\pi 6 \delta (a+6)}{2\pi^2 \delta^2 \sigma^2 a^6}$$

$$= \frac{R(a+6)}{2\pi a \delta}$$

c.)

Resistance of one plate is R, = 1 WSS

= K

For both plates: R= 2R

$$\frac{\partial}{\partial x} = \frac{\partial 3'}{\partial x'}, \frac{\partial}{\partial 3'} = n_{x'}, \frac{\partial}{\partial 3'}, \text{ ale.}$$

$$\bar{\nabla} = \bar{n} \frac{\partial}{\partial 3'}, \frac{\partial}{\partial 3'} = n_{x'}, \frac{\partial}{\partial 3'}, \text{ ale.}$$

$$\bar{n} \times \frac{\partial}{\partial 3'} = -j\omega_{x}H$$

$$\bar{n} \times \frac{\partial}{\partial 3'} = (6+j\omega E) = 0$$

assuming the g' variation is e^{ik_g} $k = \sqrt{-j \omega_{\mu} (6 + j \omega_{E})}$

$$H = -\frac{ik}{i} \, \bar{n} \, \times \bar{E} = \frac{1}{7} \, \bar{n} \, \times \bar{E}$$

$$\frac{\partial^2 E}{\partial x^2} + \frac{\partial^2 E}{\partial y^2} + \frac{\partial^2 E}{\partial z^2} + \frac{\partial^2 E}{\partial z^2} + \frac{\partial^2 E}{\partial z^2} = 0$$

assuming variation of Euroliz

Filel is also uniform in x-directors.

on
$$\frac{\partial^2 E}{\partial \eta^2} + k_{\eta}^2 E = 0$$
 $k_{\eta}^2 = k_{\eta}^2 + k_{\eta}^2 E = 0$
Since $E_2 = 0$,

En = 0 aty = 0 and y = 6

$$\frac{2-28(cont.)}{ky} = \frac{n\pi}{b}, B=0.$$

$$E_{\chi} = A \sin \frac{n\pi\eta}{b} e^{-\frac{\pi}{2}} \text{ for the } TE \text{ mostles.}$$

$$For the TM case, H_{2}=0 \text{ and } the equation satisfied by H_{χ} is:
$$\frac{\partial^{2}H_{\chi}}{\partial \eta^{2}} + (k^{2}+3^{2})H_{\chi} = 0$$

$$k_{1}^{2} = k^{2}+3^{2}$$

$$H_{\chi} = [A \sin k_{1}y + 8 \cos k_{2}y]e^{-\frac{\pi}{2}}$$

$$\frac{\partial H_{\chi}}{\partial \eta} = i\pi E_{\chi}$$

$$E_{\chi} = 0 \text{ at } \eta = 0, \eta = 6$$

$$\therefore A = 0 \text{ and } k_{\chi} = n\pi/6$$

$$H_{\chi} = B \cos \frac{n\pi}{b} e^{-\frac{\pi}{2}}$$

$$\operatorname{Cut}_{\chi} = \lim_{\chi \to \infty} \lim_{\chi \to \infty} \lim_{\chi \to \infty} |\chi = k^{2} - k^{2} - k^{2}$$

$$\ker_{\chi} = (\frac{n\pi}{b})^{2} = k^{2} - k^{2} - k^{2}$$

$$\ker_{\chi} = (\frac{n\pi}{b})^{2} = k^{2} - k^{2} - k^{2}$$

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$$\ker_{\chi} = (\frac{n\pi}{b})^{2} = k^{2} - k^{2} - k^{2}$$$$

$$f_{c} = \frac{N}{2b\sqrt{\mu}E}$$

$$\frac{2-29}{50\pi} \quad For \quad TE_{h} \quad modes,$$

$$E_{x} = E_{0} \sin\left(k_{c}y\right)e^{-\gamma t} \quad \gamma=j\beta$$

$$H_{y} = \frac{\gamma}{jw\mu} \quad E_{0} \sin\left(k_{c}y\right)e^{-\gamma t}$$

$$P = \int_{0}^{b} E_{x}H_{y}^{*} d\eta = \int_{-iw\mu}^{b} E_{0} \sin^{2}k_{c}y d\eta$$

$$= \frac{\gamma}{iw\mu} \frac{E_{0}}{2} \quad \gamma^{2} = k_{c}^{2} - k^{2}$$

$$= \frac{\gamma}{iw\mu} \frac{E_{0}}{2} \quad \gamma^{2} = -k^{2} + k_{c}^{2}$$

$$\frac{2-2?(\omega + 1)}{\beta^{2}z}\left(\frac{2\pi}{c_{0}}\right)^{2}\left(f^{2}-f_{c}^{2}\right)$$

$$\beta = \omega_{0} \delta_{0} \left[1-\frac{f_{c}}{f}\right]^{2}\left[1-\frac{f_{c}}{f}\right]^{2}$$

$$P = \frac{E_{0}}{2}\frac{\beta}{b} = \frac{E_{0}}{2\pi}\left[1-\frac{f_{c}}{f}\right]^{2}$$

$$Fon The works,$$

$$H_{x} = H_{0} \cos k_{e} y e^{-\frac{3}{2}z}$$

$$E_{y} = \frac{1}{1}\frac{\partial H_{x}}{\partial z} = -\frac{\gamma}{1}\frac{H_{0}}{\omega} H_{0} \cos k_{e} y e^{\frac{3}{2}z}$$

$$P = -\int_{0}^{b} E_{y} H_{x}^{*} dy = \frac{H_{0}}{2}\frac{b}{\omega} S$$

$$= \frac{H_{0}}{2}\frac{2\pi b}{1}\left[1-\frac{f_{c}}{f}\right]^{2}$$

$$= \frac{H_{0}}{2}\frac{2\pi b}{1}\left[1-\frac{f_{c}}{f}\right]^{2}$$

2-30 For TM modio: (Hx, F7, E++0) Pd = J'(Hx/2 R Hydz (over unit are) x2 = 21HolZR Kc = P.1 = 2R by J1- (Pc/4)2 For TE modes: (Ex, Hy, H, 70) Pd = 2/Hz/2 R = 2/kc/EDR = (tc) | k | 2R x = \frac{Pd}{2P4} = \left(\frac{f_c}{f}\right)^2 \right| \frac{k}{\white \frac{1}{27}\left(1 - \left(\frac{f_c}{f}\right)^2\right)^{1/2}} $= \left(\frac{f_c}{f}\right)^2 \frac{2R}{67\sqrt{1-\left(\frac{f_c}{f}\right)^2}}$

For Two mode:

$$|A_{\chi}| = |A_{\chi}| =$$

: we have a TEM work and to=0

$$\therefore \ \alpha_{c} = \frac{Pd}{2Pf} = \frac{R |H_{0}|^{2}}{\left(\frac{B}{WE}\right) |H_{0}|^{2}b} = \frac{R}{b\eta}$$

From Egn. 2-24:

$$K = \frac{R}{2\sqrt{4/c}} = \frac{RWE}{2B} = \frac{2RWE}{2BW}$$
(using Egm 2-26) = $\frac{R}{\gamma W}$

2-32 For TEO, mod:

It there is no dissipation in the

2-32 (conti) quide wallo we can choose a Surface 5 at some pt, in the quists when E=0, A 5. 25 =0

and Wm = We. This is most easily seen for a pure standing wave.

5= EXHX = NER [Es sin Tige ilut- per

= 2 Eo Sin 2 My cos (wt- Bt) A wy $\overline{S}_2 = \frac{1}{2} \int \int S_2 dy dx$

$$N_e = \frac{E_0^2 ba\beta^2}{2w_{jk} E_0 |^2 ab} = \frac{\beta}{w_{jk} E}$$

$$=\frac{(\gamma)^{-1}}{\xi}\left[1-\left(\frac{f_c}{f}\right)^2\right]^{1/2}$$

$$\frac{2-34 \text{ (cond.)}}{|I|} = \int_{0}^{1} \frac{E_{0}}{2\pi} \sin \frac{\pi y}{b} e^{-\frac{y^{2}}{2}} dy$$

$$= -\frac{b}{\pi} \frac{E_{0}}{2a} \cos \frac{\pi y}{b} e^{-\frac{y^{2}}{2}} \int_{0}^{1} e^{-\frac{y^{2}}{2}} dy$$

$$= \frac{2bE_{0}}{\pi^{2}} e^{-\frac{y^{2}}{2}} \int_{0}^{1} e^{-\frac{y^{2}}{2}} dy$$

$$= \frac{2bE_{0}}{\pi^{2}} e^{-\frac{y^{2}}{2}} \int_{0}^{1} e^{-\frac{y^{2}}{2}} dy$$

P = VI* because of the arbitraries in the definitions of U and I.

$$E_{X}^{1} = E_{0} \left(e^{-ik_{1}z} + \Gamma e^{ik_{1}z} \right) \sin \frac{\pi y}{6}$$

$$E_{x}^{2} = E_{0} T e^{-jk_{z}t} \sin \pi y$$

Now apply continuity of wave

impedance at interface:
$$\frac{2}{202} \left| \frac{E_{X}}{2E_{0}} \right| = \frac{E_{X}}{H_{Y}^{2}} \left| \frac{1+\Gamma}{2E_{0}} \right| = \frac{2}{1-\Gamma}$$

$$\frac{1+\Gamma}{2E_{0}} \left| \frac{1+\Gamma}{2E_{0}} \right| = \frac{2}{202} \left$$

ware when 202 = 201.

$$\frac{M_1}{\xi_1} \left(1 - \frac{M_1 \xi_1}{M_2 \xi_2} \left(\frac{f_{el}}{f} \right)^2 \right) = \frac{M_2}{\xi_2} \left[1 - \left(\frac{f_{el}}{f} \right)^2 \right]$$

$$\frac{M_1}{\xi_1} - \frac{M_2}{\xi_2} = \left(\frac{\xi_{c_1}}{\xi}\right)^2 \left(\frac{M_1^2}{M_2 \xi_2} - \frac{M_2}{\xi_2}\right)$$

$$\frac{\mu_1 \, \xi_2 - \mu_2 \, \xi_1}{\xi_1 \, \xi_2} = \left(\frac{4c_1}{4}\right)^2 \left[\frac{\mu_1^2 - \mu_2^2}{\mu_2 \, \xi_2}\right].$$

$$\frac{2-37}{201} \quad \text{For TM mode:}$$

$$\frac{2-37}{201} = \frac{\beta_1}{WE} = \frac{W\sqrt{ME}}{E} \left[1 - \left(\frac{f_{CI}}{4}\right)^2\right]^{1/2}$$

$$= 7, \left[1 - \left(\frac{f_{CI}}{4}\right)^2\right]^{1/2}$$

$$= \frac{7}{2} = \frac{2}{2} \left[1 - \left(\frac{f_{CI}}{4}\right)^2\right]^{1/2}$$

$$\frac{M_1}{\xi_1} \left[1 - \left(\frac{f_{c1}}{f} \right)^2 \right] = \frac{M_2}{\xi_2} \left[1 - \left(\frac{f_{c1}}{f} \right)^2 \frac{M_1 \xi_1}{M_2 \xi_2} \right]$$

$$\frac{M_1}{\xi_1} - \frac{M_2}{\xi_2} = \left(\frac{f_{cl}}{f}\right)^2 \left[\frac{M_1}{\xi_1} - \frac{M_1 \xi_1}{\xi_2^2}\right]$$

$$\frac{\mu_1 \, \xi_2 - \mu_2 \, \xi_1}{\xi_1 \, \xi_2} = \left(\frac{f_{cl}}{f}\right)^2 \left(\frac{\xi_2^2 - \xi_1^2}{\xi_1 \, \xi_2^2}\right) \mu_1$$

$$\frac{1}{f_{c_1}} = \sqrt{\frac{\mu_1(\xi_1^2 - \xi_2^2)}{\xi_2(\mu_1 \xi_2 - \mu_2 \xi_1)}}$$

Assume walls are made of copper, $R = 2.61 \times 10^{-7} \text{ Jf} = 8.25 \times 10^{-3}$

$$b = \frac{3 \times 10^8}{(10^9) \sqrt{2}} = .212 \text{ motor}$$

 $(10^9) \sqrt{2}$ so $q = .106 \text{ motor}$

$$Q_c = \frac{(1.11)(377)}{(8.25 \times 10^{-3})(2)} = 2.54 \times 10^{4}$$

$$b = \frac{3 \times 10^8}{(10^9) \sqrt{2(2.52)}} = .133 \text{ m}$$

$$Q = \frac{(1.11)(377)}{(8.25 \times 10^{-3})(2)\sqrt{2.56}} = 1.58 \times 10^{4}$$

Formulas for 6 and R are obtained directly.

$$\frac{2-40}{A_0} = A_2 \cos \theta$$

$$A_0 = -A_2 \sin \theta$$

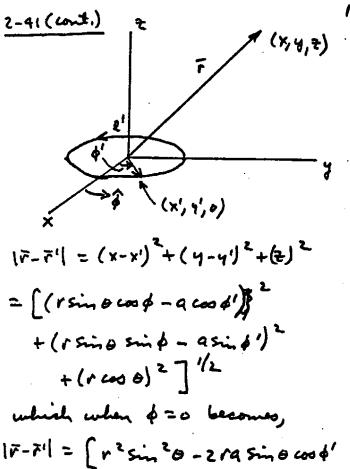
$$H = \frac{1}{\sqrt{2}\sin\theta} \begin{vmatrix} \overline{u}_{x} & r\overline{u}_{\theta} & r\sin\theta \, \overline{u}_{\phi} \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \theta} & \frac{\partial}{\partial \phi} \\ A_{2}\cos\theta - rA_{2}\sin\theta & 0 \end{vmatrix}$$

=
$$\frac{1}{r} \left[-A_2 \sin \theta - r \frac{\partial A_2}{\partial r} \sin \theta + A_2 \sin \theta \right]$$

$$H\phi = -\frac{\partial A_2}{\partial r} \sin \theta = -\sin \theta \frac{\partial e^{-ikr}}{\partial r}$$

$$= - C \sin \theta \left[-\frac{i k e^{-i k r}}{r^2} + e^{-i k r} \right]$$

$$\overline{A} = A\phi = \frac{1}{4\pi} \oint \frac{J(\phi')}{|\overline{r} - \overline{r}'|} e^{-jk|\overline{r} - \overline{r}'|} d\ell'$$



$$J(\phi') = I , dl' = a \cos \phi' d\phi'$$

$$A\phi = Iq \int_{0}^{2\pi} e^{-ik \int r^{2} + a^{2} - 2ra \sin \phi \cos \phi'} d\phi'$$

$$o \int_{0}^{2\pi} + a^{2} - ra \cdot 2 \sin \phi \cos \phi'}$$

$$f(a) = \frac{e^{-ikx}}{\alpha}, \alpha = \sqrt{r^2 + a^2 - 2ra \sin \theta \cos \theta}$$

$$af'(a)\Big|_{a=0} = \frac{\partial f}{\partial x} \frac{\partial x}{\partial a}$$

= $af'(a)\Big|_{a=0} = \frac{\partial f}{\partial x} \frac{\partial x}{\partial a}$

$$= \alpha \left\{ \frac{-ik\alpha e^{-ikx} - ikx}{\alpha^2 \cdot 2\alpha} \right\}$$

x(0) = r

Computing second term of maclaurin series me get:

$$= ae^{-ikr}(ikr+1)(2rsinecoop')$$

$$= ae^{-ikr}(\frac{ik}{r}+\frac{1}{r^2})sinecoop'$$

$$= ae^{-ikr}(\frac{ik}{r}+\frac{1}{r^2})sinecoop'$$

$$A_{\phi} = \frac{\prod_{i=1}^{2} \int_{0}^{2\pi} e^{-ikr} \left(\frac{ik}{r} + \frac{1}{r^{2}}\right) \sin \theta \cos^{2}\theta}{\left(-d\phi'\right)}$$

$$= \frac{I\pi\alpha^2}{4\pi} e^{-ikr} \left(\frac{ik}{r} + \frac{1}{r^2} \right) \sin\theta$$

$$\frac{2-42}{4\pi} A_4 = \frac{IS}{4\pi} e^{-ikr} \left(\frac{ik}{r} + \frac{1}{r^2} \right) \sin \theta$$

$$= \frac{\text{IS}}{4\pi} e^{-jkr} \left(\frac{jk}{r^2} + \frac{1}{r^3} \right) \cos \theta$$

$$+\frac{IS}{4\pi}e^{-jkr}\left(\frac{jk}{r^2}+\frac{1}{r^3}\right)\omega\theta$$

$$= \frac{IS}{2\pi} e^{-ikr} \left(\frac{jk}{r^2} + \frac{1}{r^3} \right) \omega \omega \omega$$

Ho =
$$-\frac{1}{r}\frac{\partial}{\partial r}(rA_{\phi}) = -\frac{A\phi}{r} - \frac{\partial}{\partial r}$$

$$= \frac{IS}{4\pi} e^{-ikr} \sin \theta \left[-\frac{ik}{r^2} - \frac{1}{r^3} + \frac{ik}{r^2} + \frac{3}{r^3} + \frac{ik}{r^2} + \frac{3}{r^3} + \frac{ik}{r^3} + \frac{3}{r^3} + \frac{3}{r^$$

$$R_{r} = E_{4} \times H_{0}^{*} / I^{2}$$

$$= \eta \left(\frac{H_{0}}{I} \right)^{2}$$

$$= \int_{0}^{2\pi} \int_{0}^{\pi} \frac{\eta S^{2}}{(4\pi)^{2}} \left(\frac{k^{2}}{r} \right)^{2} \sin^{2}\theta r^{2} \sin^{2}\theta d\theta d\theta$$

$$= \eta \frac{2\pi}{3} \left(\frac{kS}{\lambda} \right)^{2} \int_{0}^{\pi} \sin^{3}\theta = \frac{4}{3}$$

$$H\phi = \frac{Il}{4\pi} e^{-ikr} \left(\frac{ik}{r} + \frac{1}{r^2} \right) \sin \theta$$

$$E_{\phi} = \frac{kIS}{4\pi} e^{-ikr} \left(\frac{k\eta}{r} - \frac{i\eta}{r^2} \right) \sin\theta$$

$$H_{\theta} = \frac{kIS}{4\pi} e^{-j\frac{kr}{r}} \left(-\frac{k}{r} + \frac{j}{r^2} + \frac{l}{kr^3} \right) \sin \theta$$

The frist two terms of

Ev, Ex and Ho, H& indicate that they are cricularly polarized.

$$\frac{2-44}{\pi R_r} = \frac{1}{2} \int_0^{\pi} \frac{\left[\cos\left(\frac{kC}{L}\cos\theta\right) - \cos\frac{kC}{L}\right]_{d\theta}^2}{\sin\theta}$$

- 1 S cooke coox + sink Lsinx dex

= } [1- coor dor (1+ cook L) - Jelsin v sinklder + 1 [Sinkel] Zhl sin v do - cookel] 1-600 v do Tike = 1 (c+ logkl - cikl +sim hb(= 5i kk - 5i kl) + { cook ((+ for k) + 4 261 - 261 k) I(2) = In sin k(2+4/2) Az = 1 Typink(2+4/2) = 1kR E= jwysin & At for larger · J In sink (2'+ 2/2) eikz' coso dz' at 7 = 5 sin k(2'+4/2) eik2'000 $=\int_{-R/2}^{R/2} \frac{e^{ik(z'+l/z)}-ik(z'+l/z)}{e^{ikz'+l/z}} e^{ikz'+l/z}$ = (seik 2'(1+100) ille - e-ik 2' (1-400) ikl/2 } d 2'

$$\frac{1}{2-45(cont.)}$$

$$\frac{1}{3-\frac{1}{2}} \frac{e^{ikl/2}}{ik(1+cont)} e^{ik2'(1+cont)}$$

$$+ \frac{e^{-ik2'(1-cont)}}{ik(1-cont)} e^{ik/2}$$

$$+ \frac{e^{-ik2'(1-cont)}}{ik(1-cont)} e^{ik/2}$$

$$- \frac{e^{ikl/2}}{i+cont} e^{-ikl/2} (1+cont)$$

$$- \frac{e^{ikl/2}}{i+cont} e^{-ikl/2}$$

$$- \frac{e^{-ikl/2}}{i+cont} e^{-ikl/2}$$

$$- \frac{e^{-ikl/2}}$$

$$\frac{2-45}{1000} \left(\frac{kl}{2} \cos \theta \right) \left\{ \frac{2}{2} \sin^2 \theta - \frac{2}{2} \frac{kl}{2} \sin^2 \theta + \frac{2}{2} \frac{kl}{2} \cos \theta \right\} \left\{ \frac{2}{2} \sin \frac{kl}{2} \cos \theta \right\} \left\{ \frac{2}{2} \cos \theta \right\} \left\{ \frac{$$

$$=\int_{0}^{T} \frac{nI_{m}^{2}}{2\pi} \frac{(60)^{2} \left(\frac{n\pi}{2} (60)^{2}\right) d\theta}{\sin \theta} d\theta, \quad n \text{ even}$$

$$= \int_{0}^{\pi} \gamma I_{m}^{2} \frac{\sin^{2}(\frac{n\pi}{2}\cos\theta)}{\sin\theta} d\theta, \text{ n odd}$$

$$=\frac{1}{2}\int \frac{(1+\cos n\pi n)}{1-u^2} du$$

$$R_r = \frac{\rho}{T_n^2} = \frac{27}{4\pi} \left[log 7 + log 2 n \pi \right]$$

$$\frac{2-46 \left(cont_{1}\right)}{R_{1}} = \frac{R_{1}}{sin^{2}k\left(2+L/2\right)}$$

$$= \frac{R_{1}}{sin^{2}k\left(4+L/2\right)}$$

$$= \frac{R_{2}}{sin^{2}k^{2}} \left(4k + \frac{n\lambda}{4}\right)$$

$$= 4n\lambda$$

$$= 4n\lambda$$

Let $E_{x} = \begin{cases} -\frac{\eta J_{0}}{2} e^{jk^{2}}, & 2 > 0 \\ -\frac{\eta J_{0}}{2} e^{jk^{2}}, & 2 < 0 \end{cases}$

to show that there exists a current sheet $\overline{\mathcal{T}} = \overline{\mathcal{U}}_{\times} \overline{\mathcal{J}}_{o}$.

 $H_{y} = -\overline{E}_{x} \times \overline{U}_{z}$ $= -\frac{J_{0}}{2} e^{-jkz} , 270 \text{ (1)}$

 $=+\frac{J_0}{2}e^{jkz}, z<0$

 $\overline{J}\Big|_{z=o} = \overline{u}_z \times \left[\overline{H}^{0} - \overline{H}^{0}\right]\Big|_{z=o}$ $= \overline{u}_x \left[\frac{J_o}{z} + \frac{J_o}{z}\right] = J_o \overline{u}_x$

 $\frac{3-2}{2} \text{ Let } E_{x} = -\frac{m_{0} \sin \pi y}{2} e^{-i\beta t}, 270$ $= \frac{m_{0} \sin \pi y}{2} e^{i\beta t}, 270$

 $\overline{m} = \left[\overline{E}^{0} - \overline{E}^{0}\right] \times \overline{n}$ $= \left[\overline{E}^{0} - \overline{E}^{0}\right] \overline{u}_{x} \times \overline{u}_{z}$

my = mo sin 114 + mo sin 114
= mo sin 114

= Mo sin Tig

3-3 Since Ms = Uy A sin ITy from 3-2;

 $E_{\chi} = -\frac{A}{2} \sin \frac{\pi y}{b} e^{-i\beta \frac{2}{b}}, \pm 20$ = $\frac{A}{2} \sin \frac{\pi y}{b} e^{i\beta \frac{2}{b}}, \pm 20$

and the electric field due to current sheet $\overline{J} = \overline{u_y} \stackrel{A}{+} \sin \overline{u_y}$

3-3 (conti) can be obtained from Egn. 3-3:

 $E_{x} = -\frac{A}{2} \sin \frac{\pi y}{6} e^{-i\beta^{2}}, 200$

= -A sin 114 etipt /2<0

adding the two fields:

Ex = - A sin Try e-1/82 , 270

= 0 ,2<0

 $\frac{3-4}{E_{x}^{0}} = -\frac{J_{0}z_{0}}{2} \sin \frac{\pi y}{b} e^{-j\beta z}$

= - João sin Try e1 62, 240

 $E_{x}^{@} = \frac{J_{0}^{2}}{2} \sin \frac{\pi y}{6} e^{-i\beta(2+2d)}$

Ex = Ex + Ex (no discontinuity at 2=0)

 $E_X = 0$ at e = -d

Ex = -j Jo Zo sin Try sin p (zed) = ipd

Ex = - 30 20 sin my e-ip?

+ Jo 20 sing = j f(2+2d) = 1 270

= $\frac{J_0 \stackrel{?}{=}_0}{2} sin \frac{\pi y}{6} \left(e^{-j \frac{2\beta d}{-1}} \right) e^{-j \frac{\beta^2}{2\beta 0}}$

3-5 For the configuration of problem 2-28, the fields are written as :

(TEn) Ex = Essin ntry e , n= 12,...

(TM) Hx = Ho coo nTTy e 72, n=0,1,...

The dual fields for TEn configuration are

Hx = Ho Sin MTTY e M=1,2,... This would be a field for the

TMm case when the field strength

is maximum at ±6/2 or · conductors are placed at y = + 6/2

Similarly the dual fields for the

TM, configuration are:

Ex = - Fo wontry e 1 = 0,1,...

which would be the field for a

TEn case only when the field

is Ex=0 achieved by placing

conductors at y = ±6/2

The dual fields from a magnetic current logs come from an electric wire loop (Fig. 2-24). Hence the fields for the magnetic current loop having a 2- directed moment K5 are;

Er= - KS = ikr (ik + 1) coo

 $E_0 = -\frac{KS}{4\pi r} e^{-jRr} \left(-\frac{k^2}{r} + \frac{jk}{r^2} + \frac{1}{r^3} \right) \sin \theta$

3-6 (cont.) $H_4 = \frac{KS}{4\pi\eta} e^{-jkr} \left(\frac{k^2}{r} - \frac{jk}{r^2} \right) \sin\theta$

The fields due to a current dipole of moment Il are quen by Egn. 2-113;

H4 = Il eik (ik +12) sind

The two Hy's can only be equal if K5 (-jk) = IP 4TT

on Il = -jwEKS

3-7 7 Trs = impedance of twoin stat him Zep = impedance of collinear plate

For twin slot line :

I = 2/Hrsdl, V= SErsdl

275 = \(\in \in \text{E}_{75} \) \(\overline{U} \) 2 SHTS at

For collinear plate:

I = 2 \int \overline{H}_{cp} \overline{Al} , V = -\int \overline{E}_{cp} \overline{Al}

tep= - JEcp. II 2 \(\overline{H}_{C,p} \). \(\overline{A} \)

as r-> 0, ETS = MHTS and Ecp = -7Hcp

$$\frac{2\pi s^{2}}{4\pi s^{2}} = \int_{C_{L}} \frac{E_{L}s^{2}}{4\pi s^{2}} \frac{dl}{dl} \int_{C_{L}} \frac{E_{L}s^{2}}{4\pi s^{2}} \frac{dl}{dl}$$

$$= \frac{\eta^{2}}{4\pi s^{2}}$$

$$= \frac{\eta^{2}}{4\pi s^{2}}$$

For both the slotted line and collinear plate problems we need to solve the field equations:

on the collinear plate, (2) is solved subject to:

 $H_Y = H_Z = 0$ at x = 0 from symmetry. $H_X = 0$ at x = 0 because $\hat{H} : H = 0$ at perfect conductor.

For twin slotted line D is solved subject to:

Ey=Ez=0 at x=0 because

Etan=0 at a perfect conductes and

Ex=0 at x=0 from symmetry.

$$\frac{3-8}{H_y} = -\frac{J_0 z_0}{2} \sin \frac{\pi y}{b} e^{i\beta t}, \ \frac{2}{2} = 0$$

$$\frac{H_y^{\dagger}}{J_s} = -\frac{J_0 z_0}{2} \sin \frac{\pi y}{b} e^{i\beta t}, \ \frac{2}{2} = 0$$

$$\frac{J_s}{J_s} = -\frac{J_0 z_0}{2} \sin \frac{\pi y}{b} e^{i\beta t}, \ \frac{2}{2} = 0$$

These are the incoming waves instead of the antigoing waves developed in Section 3-2. As in Section 2-9 it was reasoned that waves must travelantword from the source, not inward. The above solutions do not vanish as 2-9 or and-therefore cannot be solutions in loss-free media. Any H field of the form:

Hy = $-\frac{J_0 z_0}{2} \sin \frac{\pi y}{b} f(\beta z)$, z>0

Hy = To20 sin Try + (-82), 200

is also a solution which gives the required current distributions

but are not physically realizable unless li flas - a: (1)

3-9 Start with fields given by Egn 2-113 and find the current sheets.

$$E_r = \frac{IR}{2\pi} e^{-jkr} \left(\frac{\eta}{r^2} + \frac{1}{j\omega E r^3} \right) \cos \theta$$

Eo = Il eier (iwh + m + 1 / ivers) sino

Hy= IP = jkr (ik + 1) sin 0

Js = Rx (H, -Hz) = RxH4 = -UBH4

$$\frac{3-9(\text{cont.})}{H_2 = 0 \text{ inoids current sphere}}$$

$$\overline{T} = -U_0 \underline{I} \underbrace{I}_{4T} e^{-ika} \underbrace{\left(\frac{ik}{a} + \underline{I}_{a^2}\right)}_{\text{Sino}}$$

$$\overline{M} = (E_1 - E_2) \times \widehat{M} \quad (E_2 = 0)$$

$$\overline{M} = U_0 \times E_r = -U_0 E_0$$
Then (1)

$$\bar{m} = -\bar{u}_{4} = -ika \left(\frac{iw_{1}}{a} + \frac{\gamma_{1}}{a^{2}} + \frac{1}{iw_{2}a^{2}} \right)$$

$$\frac{1}{2} \left(\frac{2iw_{1}}{a} + \frac{\gamma_{1}}{a^{2}} + \frac{1}{iw_{2}a^{2}} \right)$$

 $\frac{3-10}{4}$ Given a surface 5 enclosing a source \overline{J} :

Fields E_1 , H_1 are uniquely specified since $\hat{H} \times E$ is known one S from problem statement and \overline{J} is given. Also since \overline{H}_2 $\overline{H}_1 = -\nabla \times \overline{E}_1$,

- DXH,

Thus (E, H,) is supported by F of me now have:

3-11 Ex = Eo sin Try sinh 82 atwall surface, z=c, ñ=-uz Egno 1-86 are: $\left[E^{1}-E^{2}\right]\times\hat{n}=\overline{u}_{s} \frac{E_{2}}{\hat{n}} \frac{1}{E_{1}}$ $E_{s}^{2}=0$ E'=ExUx ms = (Exy)x(-uz) | == c = - Ty Ex | == c = - Ty Eo sin Try such TC Ms uniquely specifies the field Ex as guen. For a good distante (low loss): k = k - jk" = WJUE' - 1WE" J4 from Egnz-39.

 $= \omega \sqrt{\frac{M}{\epsilon'}} \left[\xi' - i \frac{\omega \xi''}{2} \right] = \xi'' \langle \langle \xi' \rangle$

3-11 (cont.) So
$$k^2 = \omega^2 \mu E'$$
 $7 = \sqrt{\frac{\pi}{6}}^2 - k^2 = \sqrt{\frac{\pi}{6}}^2 - \omega^2 \mu E'$

at resonance,
$$\omega_{\mu} = \frac{2\pi i}{2bc} \sqrt{\frac{b^2 + c^2}{\mu E'}}$$
 $8 = \sqrt{\frac{\pi}{6}}^2 - \sqrt{\frac{\pi}{6}}^2 (6^2 + c^2)$

sinh $8c = \sinh \sqrt{\frac{c\pi}{6}}^2 - \sqrt{\frac{c\pi}{6}}^2 - \pi^2$
 $= \sinh i\pi = 0$

So as E'' becomes much smaller than E' , $M_S \to 0$.

 $\frac{3-(2)}{(a)} = \frac{2}{d} = \frac{2}{d}$

 $A_{z} \approx \frac{e^{-ikr}}{4\pi r}$ It $e^{ikd\sin\theta\sin\phi}$

Es = jw/ sin & Az in the far-filed adding the contribution from the mage we obtain:

Es = jwh Ile ikr [ikd sin & soon for sind sind sind sind sind sind]

= jy Ile ikr (kd sind sind) sind

Eo = - 7 Il e 1 sin (kd sino sin b) sino Rr = Rr ob single about plus R, & its mage. Fe of single current about is from Egn 2-116: $\widetilde{P}_{4} = \frac{\gamma 2\pi}{3} \left| \frac{IP}{\lambda} \right|^{2}, R_{r} = \frac{\widetilde{P}_{4}}{|I|^{2}}$ Rr = 2 7716 Now from reciprocity, Ff of mage = Re { JEq | Jo dr} $J_b = IR \delta(y-d)$ $P = \int E_a \cdot J_b dr$ from Egn. 2-113 at 0=11/2, r=2d. Note that ky = wy, wE = k P= -I's (cookd-jsinzkd) iky +(cos eled-ismeled) 7 (2d) 2 +(w= 2kd-jsin2kd) 7
jk(2d)3 $P_f = Re(P) = \frac{\Gamma^2 k^2}{4\pi} \left[-\sin 2k d \left(\frac{k \gamma}{2d} \right) \right]$ $-\cos 2kd \frac{\gamma}{(2d)^2}$ + Sin 2 kd m

$$\hat{P}_{f} = \frac{\eta \pi I^{2} l^{2} \left[-\frac{\sin 2kd}{2kd} - \frac{\cos 2kd}{(2kd)^{2}} \right]$$

and adding this P, to that of a single element and dividing by III :

$$R_{r} = \frac{\eta \pi l^{2} \left\{ \frac{2}{3} - \frac{\sin 2kd}{2kd} - \frac{\cos 2kd}{(2kd)^{2}} + \frac{\sin 2kd}{(2kd)^{3}} \right\}$$

Let
$$A = -\frac{\sin z k d}{z k d} - \frac{\cos z k d}{(z k d)^2} + \frac{\sin z k d}{(z k d)^3}$$

$$A = \left[-1 + \frac{(2 \text{ kol})^2}{3!} \pm \dots - \frac{1}{(2 \text{ kol})^2} + \frac{1}{2!}\right]$$

$$-\frac{(2kd)^{2}}{4!}+\dots+\frac{1}{(2kd)^{2}}-\frac{1}{3!}+\frac{(2kd)^{2}}{5!}-\dots$$

$$= -\frac{2}{3} + \frac{8}{15} (kd)^2$$
 as $kd \to 0$

$$R_r = \frac{7\pi L^2}{\lambda^2} \left[\frac{2}{3} - \frac{2}{3} + \frac{8}{15} \frac{(2\pi d)^2}{\lambda^2} \right]$$

$$= \frac{732\pi^3l^2d^2}{15\lambda^4}$$
 as kd > 0

For small d,

$$= \frac{4\pi r^2 \eta \left| \frac{I}{\lambda r} \right|^2 \sin^2 kd}{2\pi \left| \frac{I}{\lambda} \right|^2 \left[\frac{2}{3} + A \right]}$$

$$g \approx \frac{4(kd)^2}{\frac{8}{15}(kd)^2} = 7.5$$

In d= 1/4, kd= TT/2

$$q = \frac{4(1)}{\left[\frac{2}{3} - 0 + \frac{1}{4} + 0\right]} = 5.21$$

For a large,

$$g = \frac{4 \sin^{2} kd}{\left[\frac{2}{3} - \frac{\sin^{2} kd}{2kd} - \frac{\cos^{2} 2kd}{(2kd)^{2}} + \frac{\sin^{2} kd}{(2kd)^{3}}\right]}$$

$$= \frac{4}{2/2} = 6$$

3-13 The dual of a small loop of electric current is a magnetic current of moment Kl such that Kl= jwy IS.

 $E_{\phi} = \frac{Kl}{\lambda r} e^{-jkr} \sin(kd \cos \theta) \sin \theta$

The results of the electric and magnetic currents are related by Egn, 3-19 (its dual).

as led so,

Friding time average power Pf as:

$$\frac{3-13(cont.)}{+(2kd)^{2}} + \frac{1}{(2kd)^{2}} +$$

 $F_{z} = \frac{Kl}{4\pi} \frac{e^{-ikr}}{r} \left[e^{ikd sin \phi sin \theta} + e^{-ikd sin \phi sin \theta} \right]$ $= \frac{Kl}{2\pi r} \frac{e^{-ikr}}{r} \cos \left(kd sin \phi sin \theta \right)$ Kl = i wh TS from alual of Egn 3-19. $E_{\phi} = -iw E \gamma \sin \theta F_{z} from dual$

E4 = -jω εη sub = from anal & Egn. 2-123. -jkr

Eq = (-iwez) i wh I se sing

= WZMENISE-jkr Sin O coo (kd sin \$ sin 6)

Rr = Rr B mage plus Rr B single current loops with dual electric current despo with dual magnetic current elements. Rr B single current loop is from problem 2-42: Rr = 72TT (k5)2

as in prob. 3-12,

Ha from dual of Egn 2-113:

$$H_0 = -\frac{Kl}{4\pi} e^{-ikr} \left(\frac{j\omega \varepsilon}{r} + \frac{1}{7r^2} + \frac{1}{j\omega \mu r^3} \right)$$

at 0=11/2 , 1=2d

$$=\frac{\omega^2\mu^2(IS)^2-ik^2d\left(\frac{jk}{7r}+\frac{1}{7r^2}\right)}{4\pi}$$

$$\omega^2 \mu^2 = \frac{k^2 \mu}{\xi} = (k \gamma)^2 \qquad \frac{1}{j k \gamma r^3}$$

$$P = \frac{k^2(kIS)^2\eta^2}{\gamma_{ATT}} \left(\cos 2kd\right)$$

adding this to Tf for single plument and dividing by | I 2 | me get:

the applied voltage for the dipole (monopole) to turie that of the monopole alone. Fields of monopole are same as those of dipole in region () conly,

$$R_b = \frac{\widetilde{P}_{fb}}{|I_{in}|^2} = \frac{2|V_{in}|}{|I_{in}|}$$

D - denotes dipolo m - denotes monopolo

$$R_{m} = \frac{\tilde{p}_{q_{m}}}{|I_{in}|^{2}} = \left| \frac{V_{ii}}{I_{in}} \right|$$

$$\frac{90}{9m} = \frac{P_{tm}}{P_{to}}$$
 Since fields are some for both in region O .

This is a loop of magnetic amount

which, if b << \lambda, acts as an electric dipole. The current can be represented as a continuous distribution of magnetic current filaments of strength dK = Mydp. The total moment of the source is then:

 $KS = \int_{a}^{b} \pi e^{2} dK = \pi V (b^{2} - a^{2})$ $= \frac{1}{2} \ln b / a$

The dual equivalent electric current element must satisfy the equality: Il = -jwEKS

Ho = iII e-ikr sin 0

 $= \frac{WETTV(b^2-a^2)e^{-ikr}}{4\lambda r \ln b/a}$

= 1 & Egn, 3-20.

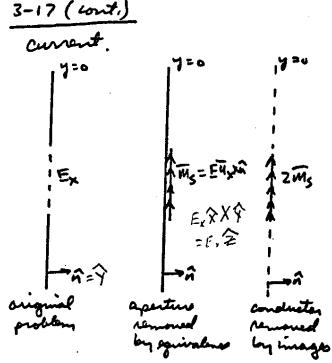
Radiated power = P = Jdf /7 | Hol Sin do

$$P = \frac{\left| \omega \epsilon \pi V \left(b^2 - a^2 \right) \right|^2}{4 \lambda \ln b / a} \left| \frac{2 \pi \gamma \left(\frac{4}{3} \right)}{3} \right|$$

= 1 Pf (& Egn 3-21)

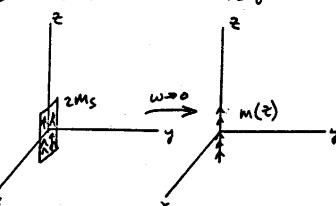
Gr = Pf = 1 & Egm 3 - 23.

3-17 loing the equivalence principle we can replace the aperture by a patch of magnetic



 $M_S = \frac{2V_m}{w} \sin\left[k\left(\frac{L}{2} - |2|\right)\right]$

as w- o the equivalent magnetic current behaves as a line filament:



M(2) replaces I(2)

Using the dual of Egn. 2-125 and making the following substitutions;

Im -> 2Vm, Eo -> Ho 4>0

7 -> 1/7 Eo-> -Ho y co

$$F_{2} = \frac{V_{M}}{\pi} \frac{e^{-ikr}}{r} \frac{\sin(\frac{k\omega\cos\phi\sin\phi}{2}\cos\phi)}{(\frac{k\omega\sin\phi\cos\phi}{2}\sin\phi\cos\phi)}$$

Ho=jwEsinoFz

THY (KW (00 psino)

$$f(\theta, \phi) = \frac{\sin(\frac{k\omega}{2}\cos\phi\sin\phi)(\omega(\frac{k\omega}{2}\cos\phi)-\omega\frac{k\omega}{2})}{(\omega(\frac{k\omega}{2}\cos\phi)-\omega\frac{k\omega}{2})}$$

$$= \frac{e^{-ikr} \sin \left[\frac{kq}{2} \cos q \sin \theta\right]}{(x)}$$

This expressions analystist from

$$F_2 = \frac{e^{-ikr} \sin(\frac{ka}{2} \cos \phi \sin \phi)}{2\pi r}$$

$$= \frac{e^{-jkr} Sin(\frac{ka}{2}\omega \phi Sin \theta)}{\frac{k}{2}\omega \phi Sin \theta} \frac{b\omega o(\frac{kb}{2}\omega o \theta)}{(\pi^2 - k^2b^2\omega o^2\theta)}$$

$$= \frac{2j}{7} \frac{e^{-jkr} b sin(\frac{kq}{2} cos \phi sin \phi) cos(\frac{kb}{2} cos \phi)}{cos \phi}$$

$$\frac{3-20}{E_{z}} = E_{0}e^{ik(x\cos\phi_{0} + y\sin\phi_{0})}$$

$$d\overline{F} = -\frac{1}{4\pi r} \overline{m} e^{-ik|r-r'|}$$

$$dE_{z}^{(s)} = -\nabla_{(r)} \times dF_{y}$$

$$= -\frac{i}{k} e^{-ikr} = e^{iky'(\sin \phi + \sin \phi_{0})}$$

$$= -\frac{i}{2\pi r} E_{0} e^{iky'(\sin \phi + \sin \phi_{0})}$$

$$= -\frac{i}{2\pi r} (\pi T_{0} + \phi) dS$$

$$E_{2}^{S} = -ikE_{0} e^{-ikr}$$

$$= -ikE_{0} e^{-ikr}$$

$$= -ik E_{0} be^{-ikr}$$

$$= -ik E_{0} be^{-ikr}$$

$$= -ik E_{0} be^{-ikr}$$

$$= -ik E_{0} be^{-ikr}$$

$$= -ik (\sin \phi + \sin \phi)$$

$$= -ika (\sin \phi + \sin \phi)$$

$$= -ika (\sin \phi + \sin \phi)$$

$$= -ika (\sin \phi + \sin \phi)$$

=
$$\frac{kE_0 abe^{-ikr} \omega \phi}{i 2\pi r} \frac{\left[\frac{ka}{2}(\sin\phi + \sin\phi)\right]}{\frac{ka}{2}(\sin\phi + \sin\phi)}$$

jk(sin 4+sin do)

at d=do, maximum backscatter

$$\frac{3}{3} = \frac{1}{3} \left| \frac{kE_0 ab sin \left(ka sin \phi_0\right) \cos \phi_0}{2\pi r} \right|$$

$$\overline{M}_S = 2 \hat{R} \times \overline{E}^{\hat{A}}$$

$$\overline{E}^{\hat{A}} = \gamma H_{\hat{A}}^{\hat{A}}$$

$$\frac{3-21(\text{cont})}{\text{E}_{2}^{\lambda}} = \gamma H_{0} e^{ik(x\cos\phi_{0} + y\sin\phi_{0})}$$

$$E_{2}^{\lambda} = \gamma H_{0} e^{ik(y\sin\phi_{0})}$$

$$M_{2} = 2\gamma H_{0} e^{ik(y\sin\phi_{0})}$$

$$F_{3} = \int_{-b/2}^{42} \frac{e^{-ikr}}{4\pi r} \frac{e^{ik(y\sin\phi_{0})}}{2\gamma H_{0}} e^{ik(y\sin\phi_{0})}$$

$$e^{ik(y\sin\phi_{0})}$$

$$e^{ik(y\sin\phi_{0})}$$

$$\cos\phi_{0} dzdy$$

$$E_{z}^{S} = \eta H_{z}^{S} = -\nabla x F_{y}$$

$$At \Theta = \frac{\pi}{2} \text{ in far field, } E_{\Theta} \Rightarrow -E_{z}$$

$$-E_{z} = j k F_{\varphi} = j k F_{y}$$

$$k_{\Theta} = \frac{\pi}{2} = \frac{\pi}{2} = \frac{\pi}{2}$$

from Egn 3-97. :. after earrying out integral for Fy we have:

at \$ = \$0 for max. backscatter:

$$\vec{S}^{i} = \eta \left| \frac{k H_0 a b}{2 \pi r} \frac{\sin[k a \sin \phi_0] \cos \phi_0}{k a \sin \phi_0} \right|$$

$$A_{c} = \lim_{n \to \infty} \left(\frac{4\pi r^{2} \tilde{S}^{5}}{\tilde{S}^{i}} \right)$$

which is the same as problem 3-20.

SSE. Jzdr = SSEz. J. dr

(1) E10 = \(\int_{22} \cdot \I_{2} dz' \) \(\text{since source } Z \) is a delta function)

For field of a Hertzian dipolo is:

Ezz = nj Ileie ikz woo

Substituting into (1), $E_0 = \frac{e^{-ikr}}{r} \int_{-L/2}^{L/2} \frac{\eta_j(T)e^{ikz'(\omega)6}}{2\lambda r}$

· In Sin [le(= -124)] Singly case II E10 = 17 Ine 1 (600 (16 400) - 600 (16)

which is Egn. 2-125.

3-23 From Egn 3-39, (a,b) = VbIa (b is a wettage

I is the current at b when some a is applied. When Egn 3-38 is satisfied, (a,6) = (6,0)

3-23 (cont.) on ub I'a = va I'b When port 1 (Fig 3-18) is excited and the short circuit current at port z is observed, 19421 = I'V Similarly, Vy = Ib

: . Yz = Y , z

Case I

いべ

が、

Lu moet,

Vi= V at () when In is applied at (2) Uzi = V at(2) when In is applied at(1) ViIn = Vi In from reciprocity

for case I, In! = VzIn

In(vi+vi) = (vi+vi) In

UZIL= USI4 V2 = V15

$$z = -\langle a, a \rangle$$

$$E_{x}^{a} = -\frac{J_{0} z_{0}}{2} \sin \frac{\pi y}{b} \quad \text{at } z = 0$$

$$\langle a, a \rangle = \iint_{\mathbb{R}^{2}} \mathbb{E}_{x}^{a} \cdot J_{x}^{a} ds$$

$$= \iint_{\mathbb{R}^{2}} -J_{b}^{2} \mathcal{E}_{0} \sin^{2} \pi y dx dy$$

$$=-\frac{J_0^2ab}{4}$$

$$T^2 = \frac{4(abJ_a)^2}{\pi^2}$$

$$\frac{\frac{1}{4}}{4(abJ_0)^2} = \frac{2000^2}{16ab}$$

$$\langle a,a\rangle = \iint E_x^q \cdot J_x^q dS$$

$$= \iint_{a} -\frac{J_{\delta}^{2}}{2} \sin^{2} \frac{\pi y}{b} \left(1 - e^{-\frac{1}{2}Ad}\right) dxdy$$

$$= -\frac{J_b^2 ab }{4} \left(1 - e^{-i 2\beta d} \right)$$

$$\frac{2}{1600} = \frac{20\pi^2}{1600} \left(1 - e^{-i2\beta d}\right)^2$$

Use identity:

II (E^qxH^b-E^bxH^q) dS = III (E^b·DXOXE^q-E^q·DXOXE^b) d7 which from results of problem 3-27 reduces to Eqn. 3-25.

Green's second identity, SS (Ā×♥×Ē-Ē×♥×Ā)dS = SSS(B.OXDXA-A.DXDXB)dY Let A = E , B = G S(ExOXG)-(GXOXE)dS = SSS(G, OXOXE-E. DXDXG,) dr From Egn 3-47, 3-48, $G_1 = \bar{C} \phi$ $\phi = \frac{e^{-ik|r-r|}}{e^{-ik|r-r|}}$ Volumo integral becomes; JJG1.(O(O.E)+k2E)2T - III E. (0(0,6,)+k26, dr In the region enclosed by the volume D.E=0. aso 6: (k2E) = E. (k261)

If E. (P(P.G.) + k²G. dt In the region enclosed by the volume P. E = 0. Also G: (k²E) = E. (k²Gi) So the volume integral further reduces to: - ISE. (P(P.G.)) dt = - SSP. (EP.G.) dt - SSE.C. dt = - SSP. (EP.G.) dt - SSE.C. dt by diseigned then by approx. as |r-r'| -> 0

Remembering this equals original surface integral about we drain

$$\frac{3-29 \text{ (cont.)}}{\text{a final result as:}}$$

$$-4\pi \overline{\text{c}} \cdot \text{E} =$$

$$\iint (\text{E} \times \nabla \times \text{G}_1 - \text{G}_1 \times \nabla \times \text{E}_1 + \text{EP.G}_1) dS$$

$$\overline{C} = \overline{U}_{2}$$

$$(x_{i} - y_{i}' e^{i})$$

rinding the vector potential A,

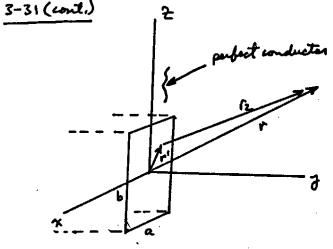
$$A_{z} = \frac{1}{4\pi} \iiint \frac{\delta(x', y', z')}{r_{i}} e^{-\frac{1}{2}k|r_{i}-r'|} - \frac{\delta(x', y', z')}{r_{i}} e^{-\frac{1}{2}k|r_{i}-r'|} e^{-\frac{1}{2}k|r_{i}-r'|} dr'$$

$$= \frac{e^{-ikr_i}}{4\pi r_i} - \frac{e^{-ikr_i}}{4\pi r_i^{2k}}$$

H = 0xA = 64

$$\overline{G_4} = 0 \times \overline{u_2} \left(\frac{e^{-ikr_1}}{r_1} - \frac{e^{-ikr_2}}{r_2} \right)$$

$$\overline{G}_{4} = \nabla \times \overline{u}_{2} \left(\frac{-e^{-ik\Gamma_{2}}}{r_{1}} \right)$$



Egn 3-75: $4\pi \overline{c} \cdot \nabla' \times \overline{E} = \iint (\overline{G}_{4} \times \overline{\nabla} \times \overline{E}) \cdot dS$ $\overline{c} = \overline{U}_{2}$ $\nabla \times \overline{E} = \frac{1}{7} \sin \frac{\pi y}{b} \overline{u}_{y}$ $\iint (\overline{G}_{4} \times \overline{\nabla} \times \overline{E}) \cdot dS$ $= \iint \nabla \times \overline{u}_{2} \left(\frac{-e^{-ikr_{2}}}{r_{2}} \right) \times \frac{1}{7} \sin \frac{\pi y}{b} \overline{u}_{y} \cdot dS$ $= \iint \nabla \times \overline{u}_{x} \left(\frac{e^{-ikr_{2}}}{r_{2}} \sin \frac{\pi y}{b} \right) \cdot dS$ $= \nabla \times \iint \frac{e}{r_{2}} e^{-ikr_{2}} \sin \frac{\pi y}{b} \cdot dS$ $= \nabla \times \iint \frac{e}{r_{2}} e^{-ikr_{2}} \sin \frac{\pi y}{b} \cdot dS$ $= \nabla \times \iint \frac{e}{r_{2}} e^{-ikr_{2}} \sin \frac{\pi y}{b} \cdot dS$ $= \nabla \times \iint \frac{e}{r_{2}} e^{-ikr_{2}} \sin \frac{\pi y}{b} \cdot dS$

which is the same integral as for ATTF in problem 3-19 and if the far field expression for our is used (jw E sin 0) we get the same answer as in problem 3-19.

$$\frac{3-32}{F = \frac{KRe^{-ik|r-r|}}{4\pi|r-r|}}$$

$$\psi = \frac{e^{-ik|r-r|}}{4\pi|r-r|}$$

By duality:

jup -> jwE

jue -> jun

E-> H

4-7 F

Egno. 3-65 become:

13 = 1 224 x 723 x 73

This can be very easily seen by taking the duals of Egus. 2-111 which are:

E = - 0xF

and finding H for FR. Ux, Uy, and

H = [1] Ke where [17]

is given above.

of Il is Ux directed,

$$A_{x} = \frac{Ie^{-ik|r-r'|}}{4\pi |r-r'|} \quad \psi = \frac{e^{-ik|r-r'|}}{4\pi |r-r'|}$$

 $\overline{H} = O \times \overline{A} = \frac{\partial A}{\partial z} \overline{u}_y - \frac{\partial A}{\partial y} \overline{u}_z$

For Il Ty directed,

$$H_X = -\frac{\partial Y}{\partial z}$$
 Ily

Hz= JY Ily

For Il is directed,

: we can write :

3-34 E= = E0 eik (x coods + y sin 40)

テニ Z於× Fi

Jy = - Fo eiky'sin do sin (7/2+ 40)

= - Eo eiky'sin \$ cos \$ 6

Ay = -eikr Eowo of eiky'sinho eik

E = 5×4 =

The scho area is found at \$=\$0=0 and the same procedure followed as in problem 3-20 which gives the same result.

 $J = \hat{n} \times \hat{H}^{i} = H_{0}e^{iky'sin}\phi_{0} = -J_{y}$ $A_{y} = \frac{-e^{-ikr}}{4\pi r} H_{0} \int \int e^{iky'sin}\phi_{0}$

$$H_{2}^{5} = -ik Aycoop$$

$$= ik H_{0}abe^{-ikr} sin\left[\frac{ka}{2}(sin\phi + sind_{0})\right]_{coop}$$

$$= \frac{ik H_{0}abe^{-ikr}}{2\pi r} \frac{ka}{2}(sin\phi + sind_{0})$$

The echo area is found in the same way as problem 3-21 at $\phi = \phi_0 = 0$ after which the same Ae is obtained.

This field is Tim to ±. It is a uniform plans wome travelling in the + if direction and is linearly polarized. It is also TEM to its direction of propagation.

 $\frac{3-37}{\gamma = e^{-ikx}} \qquad F = U_2 \gamma$ $E_x = 0 \qquad H_x = 0$ $E_y = -ike^{-ikx} \qquad H_y = 0$ $E_z = 0 \qquad H_z = -\hat{\gamma} e^{-ikx}$ $F_z = 0 \qquad H_z = -\hat{\gamma} e^{-ikx}$ $F_z = 0 \qquad H_z = -\hat{\gamma} e^{-ikx}$

This field is TEM to the Xdirection. It is a uniform plane wave travelling in the + X direction and linearly polarized.

z= ux , x = e-ikz ++= je-ikt Dx (24t) = ke-iktuy DXDX(TY4) = k2e-ik2 Ux E = -ke-iktuy - 3e-iktux $\nabla x (\xi \gamma^{a}) = -i k e^{-ik^{2}} \overline{u}_{y}$ DXDX(Ext) = jk2e-jkt ux H = -jkejktuy - gjejktux Ex = -3e-ikt Hx = -9je-ikt Ey = -ke-ikt Hy = -ike-ikt $\frac{E_{x}}{H_{y}} = \frac{-\hat{s}}{-ik} = \sqrt{\hat{s}'} = \gamma$

 $\frac{E\gamma}{H_{x}} = \frac{-k}{-9j} = \sqrt{-\frac{2}{3}\frac{2}{9}} = -7$

Thus we have a uniform

plane wave TEM to Z

travelling in the + & direction which is circularly polarized.

3-39

In the radiation zone, Egns 3-4 become:

$$\bar{H} = Q \times \bar{A} + \frac{1}{3} (Q \times Q \times \bar{F})$$

$$\exists x = -\overline{u_0} + \left[\frac{\partial}{\partial r}(rF_0)\right] + \overline{u_0} + \left[\frac{\partial}{\partial r}(rF_0)\right]$$

So,

$$E = -\overline{u}_{\theta}(ik)F_{\theta} + \overline{u}_{\phi}(ik)F_{\theta}$$

$$+ \overline{u}_{\theta} \frac{k^{2}}{\widehat{g}}A_{\theta} + \overline{u}_{\phi} \frac{k^{2}}{\widehat{g}}A_{\phi}$$

3-39 (cont.)

Separating components me get: Eo = -iwx Ao -ik Fo Eo = -iwx Ao + ikFo

$$\frac{4-1}{4} = \int \int f(k_x k_y) h(k_x x) h(k_y y) h(k_z z) dk_x dk_y$$

$$k_x k_y$$

$$\frac{\int^2 \gamma}{\partial x^2} = k_x \iint f(k_x, k_y) h''(k_x x) h(k_y y) h(k_z x) dk_x dk_y$$

$$k_x k_y$$

$$\frac{\partial^2 \psi}{\partial x^2} + k_x^2 \psi =$$

Now h"(kxx) +kx h(kxx) =0

from separation of variables.

This is also true for the other

Two components thus the Halmboltz

equation is satisfied.

sinkx = sin (B-jk) X

= Simpx coo(j x) x

- coopx Sin (jk) x

= sin prosh ax - j cos px sinh ax

sin(ix) = j sinhx

cos(jx) = cosh x

Similarly

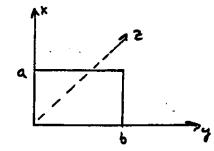
 $\omega_0 kx = \omega_0 (\beta x - j kx)$

= cospx coohax + jsin px sinhax

= -0 x 42 (e ikx - iky y - ikz)

=
$$(k_x^2 + k_y^2 +) \overline{u}_2 - k_y k_2 + \overline{u}_y$$

- $k_x k_2 + \overline{u}_x$



For TMmn mode:

$$\frac{4^{4} \left(\operatorname{cont}_{i} \right)}{|a^{2}|} = \omega^{2} M \, E - \beta^{2} = \omega^{2} M \, E$$

$$\left(\frac{m\pi}{a} \right)^{2} + \left(\frac{m\pi}{b} \right)^{2} + \beta^{2} = \omega^{2} M \, E$$

$$\kappa = \frac{Power}{2} hoso on wallo$$

$$\frac{a}{2} \cdot Pawer flow$$

$$eta = \int_{0}^{a} \frac{b}{E \times H^{4}} dx dy$$

$$= \int_{0}^{a} \left(E_{X} H_{y}^{4} - E_{y} H_{x}^{4} \right) dx dy$$

$$E_{X} H_{y}^{4} = \frac{C^{2} \beta \omega E}{h^{2} \cdot h^{2}} \left(\frac{m\pi}{a} \right)^{2} \cos^{2} \frac{m\pi\pi x}{a} \sin^{2} \frac{n\pi\pi y}{b}$$

$$E_{Y} H_{x}^{4} = \frac{C^{2} \beta \omega E}{h^{2} \cdot h^{2}} \left(\frac{m\pi}{b} \right)^{2} \sin^{2} \frac{m\pi\pi x}{a} \cos^{2} \frac{m\pi y}{b}$$

$$P_{\xi} = \frac{C^{2} \beta \omega E}{h^{4}} \frac{a}{2} \cdot \frac{b}{2}$$

$$P_{\xi} = R \int_{0}^{a} \left(\frac{(H_{x})^{2} + (H_{y})^{2}}{h^{4}} \right) dx$$

$$+ \int_{0}^{a} \left(\frac{\omega E C m\pi}{h^{2}} \right)^{2} \sin^{2} \frac{m\pi\pi x}{a} dx$$

$$+ \int_{0}^{a} \left(\frac{\omega E C m\pi}{h^{2}} \right)^{2} \sin^{2} \frac{m\pi\pi x}{a} dx$$

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$$+ \int_{0}^{a} \left(\frac{\omega E C m\pi}{h^{2}} \right)^{2} dx$$

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$$+ \int_{0}^{a} \left(\frac{\omega E C m\pi}{h^{2}} \right)^{2} dx$$

$$+ \int_{$$

$$\frac{4-4 \left(\operatorname{cont}_{1} \right)}{\left| \mathbf{h}^{2} \right|^{2} + \left(\operatorname{mtt}_{1} \right)^{2} + \left(\operatorname{mtt}_{2} \right)^{2} + \left(\operatorname{mt}_{2} \right)^{2} + \left(\operatorname{mt}$$

$$\frac{4-4 (cont.)}{\beta = k}$$

$$\frac{1-(\frac{k}{e})^2}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1-(\frac{k}{e})^2}{ab} \frac{1-(\frac{k}{e})^2}{m^2b^2+n^2a^2}$$

$$\frac{2R}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{2}{ab} \frac{m^2b^3+n^2a^3}{m^2b^2+n^2a^2}$$

$$\frac{2R}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1}{ab} \frac{1-(\frac{k}{e})^2}{m^2b^2+n^2a^2}$$

$$\frac{1}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab}$$

$$\frac{1}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab}$$

$$\frac{1}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab}$$

$$\frac{1}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab} \frac{1}{ab}$$

$$\frac{1}{\gamma \sqrt{1-(\frac{k}{e})^2}} \frac{1}{ab} \frac$$

$$\frac{4-4(cond.)}{+(\frac{\beta C}{h^{2}} \frac{n\pi}{b})^{2}} + (\frac{\beta C}{h^{2}} \frac{n\pi}{b})^{2} \sin^{2} \frac{n\pi}{b} \int_{a}^{b} d\eta$$

$$= \frac{2RC^{2}}{2} \left[a + \frac{\beta^{2}}{h^{4}} \frac{m\pi}{a} \right]^{2} a$$

$$+ b + \frac{\beta^{2}}{h^{4}} \frac{(n\pi)^{2}}{b} \int_{a}^{b}$$

$$= RC^{2} \left[(a+b) + \frac{\beta^{2}\pi^{2}}{h^{4}} \frac{m^{2}b + n^{2}a}{ab} \right]$$

$$= \frac{2R(a+b)h^{2}}{w_{\mu}\beta ab} + \frac{RC^{2}\beta^{2}\pi^{2} \cdot 2(n^{2}b + n^{2}a)}{h^{4}} \frac{ab \cdot ab}{ab \cdot ab}$$

$$= \frac{m^{2}}{w_{\mu}\beta} = \frac{w^{2}\mu^{2} - \beta^{2}}{w_{\mu}\beta} - \frac{w}{\beta} - \frac{\beta}{w_{\mu}\beta}$$

$$= \frac{m^{2}}{\sqrt{1-(4c/4)^{2}}} - \frac{\sqrt{1-(4c/4)^{2}}}{\sqrt{1-(4c/4)^{2}}}$$

$$= \frac{(4c/4)^{2}}{ab \gamma \sqrt{1-(4c/4)^{2}}}$$

$$+ \frac{2R(a+b)(4c/4)^{2}}{ab \gamma \sqrt{1-(4c/4)^{2}}}$$

$$+ \frac{m^{2}b + n^{2}a}{n^{2}b^{2}} \cdot 2R\beta \frac{\omega}{\omega}$$

$$\alpha_{TE} = \frac{2R}{2R} \frac{a+b}{ab} \frac{(f_{1}/e)^{2}}{\sqrt{1-f_{1}/e}}$$

$$+ \frac{2h}{m^{2}b^{2}+n^{2}a^{2}} \sqrt{1-f_{1}/e}$$

$$+ \frac{2h}{m^{2}b^{2}+n^{2}a^{2}} \sqrt{1-f_{1}/e}$$

$$+ \frac{n}{m^{2}b^{2}+n^{2}a^{2}} \sqrt{1-f_{1}/e}}$$

$$+ \frac{n}{m^{2}b^{2}+n^{2}a^{2}} \sqrt{$$

For single mode operation over a 2:1 frequency range, b/a=2 $t=10^{10}$ Hz => $\lambda=3$ Cm.

to b = zam, hc = 4 am for TEOI make

$$K_{c} = \frac{2.61 \times 10^{-7} \sqrt{10^{10}}}{a(377)\sqrt{1-(\frac{.75}{1})^{2}}} \left[1 + \frac{29}{6}(.75)^{2}\right]$$

= 1.64×10-4 nepero

then $k_c = 6$ cm on $f_c = 5 \times 10^9$ Hz which is way above cutoff so we no longer have single mode propagation since TE_{10} mode starts at 10^{10} Hz.

Acourse no variation in X-direction, i.e., $\frac{2}{2} = 0$ The equation to $\frac{2}{2}$ by satisfied is:

 $\frac{\partial^2 t}{\partial y^2} + \frac{\partial^2 t}{\partial z^2} + k^2 t = 0$

Let 3244 + ky 450 and 324 + k2 4 =0

Py = A'wokyy + B'sin ky y

4-6 (cont.) $\psi_z = C' e^{-ikz^2} + b' e^{ikz^2}$ Cosume the wave travels only in the +2 direction and form the product solution $\psi = \psi_y \psi_z$: $\psi = (A cooky y + B sin ky y) e^{-ikz^2}$ The ψ for τ and τ coses

One obtained by matching the fields

The of for TM and TE coses are obtained by matching the fields at the boundaries. For the TEM mode the Halmholtz equation Separates into the following form assuming a variation of e^{-ik_2t} in the 2-direction.

 $\frac{\partial^2 f}{\partial y^2} y = 0 \qquad \frac{\partial^2 f}{\partial z^2} + k_z^2 f_z = 0$

Total solution is:

Y = Aye-jkz=

4-7

From Egn 4-33 and 4-34:

1/mm = coo milk sin nitry e-iket

ATEX = sin mix wo my e-iket

Since there is no K- variation for the parallel plate quiet these reduce to the equation given in the problem statement.

4-8 X

Emm = jyk xux since F= 4ux

Emm = j +k xuz since F = +uz

$$\frac{4-8(2011)}{E_{mn}} = \frac{1}{3}(-k_{e}k + u_{e}k^{2})\psi \quad \text{sines } \overline{A} = \psi \overline{u}_{e}$$

$$= \frac{1}{3}(-k_{e}k + u_{e}k^{2})\psi \quad \text{sines } \overline{A} = \psi \overline{u}_{e}$$

$$= \frac{1}{3}(-k_{e}k + u_{e}k^{2})\psi \quad + D_{3}\psi \overline{u}_{x} \times \overline{k}$$

$$= \frac{1}{3}(-k_{e}k + u_{e}k^{2})\psi \quad + D_{3}\psi \overline{u}_{x} \times \overline{k}$$

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$$= \frac{1}{3}(-k_{e}k + u_{e}k^{2})\psi \quad + D_{3}\psi \overline{u}_{x} \times \overline{k}$$

$$M O = -\frac{B}{3} k_2 k_x + A k_y j$$

$$jkz = -Ak_xj - \frac{B}{9}k_2k_y$$

$$-jky = (k_x^2 + k_y^2)\frac{B}{9} \Rightarrow B = -\frac{9}{9}jk_y$$

$$k_x^2 + k_y^2$$

$$\therefore A = \frac{-k_2 k_x}{k_x^2 + k_y^2}$$

Restating Emy in problem statement

form:
$$E_{m_n}^{TE_y} = A' \left(E_{m_n}^{TE} + B' E_{m_n}^{TM} \right)$$

$$A' = \frac{-k_2 k_x}{k_x^2 + k_y^2}, B' = \frac{j \hat{\gamma} k_y}{k_2 k_x}$$

$$\frac{4-8 \text{ (cont.)}}{j + \overline{u}_{x} \times \overline{k}} = \frac{c}{s} \left(-k_{z}\overline{k} + \overline{u}_{z} k^{2}\right) + b + b + \overline{u}_{x} \times \overline{k}$$

$$+ b + \overline{u}_{x} \times \overline{k}$$

$$j + (k_{y}\overline{u}_{z} - k_{z}\overline{u}_{y}) = \frac{c}{s} \left(-k_{z}k_{x}\overline{u}_{x} - k_{z}k_{y}\overline{u}_{y}\right)$$

$$+ k_{x}\overline{u}_{z} + k_{y}\overline{u}_{z}\right) + b + j b + (k_{x}\overline{u}_{y} - k_{y}\overline{u}_{x})$$

$$D = -\frac{Ck_{2}k_{x}}{3} - \frac{Ck_{2}k_{y}}{3} - \frac{10}{3}k_{y}$$

$$-\frac{1}{3}k_{z} = -\frac{Ck_{2}k_{y}}{3} + \frac{1}{3}0k_{x}$$

$$-\frac{1}{3}k_{y} = \frac{C}{3}(k_{x}^{2} + k_{y}^{2}) \Rightarrow C = -\frac{1}{3}\frac{3}{3}k_{y}$$

$$k_{x}^{2} + k_{y}^{2}$$

$$k_{x}^{2} + k_{y}^{2}$$

$$\therefore D = \frac{-k_x k_z}{k_x^2 + k_y^2}$$

Restating Hum in problem statement form : Home = C' (Home + DHome)

$$C' = -\frac{i \hat{\delta} k_{y}}{k_{x}^{2} + k_{y}^{2}}, \quad D' = \frac{k_{x} k_{z}}{i \hat{\delta} k_{y}}$$

of The = Sin mil Sin nily too prite for a rectangular cavity. For this particular resonator there are no zeros in the 2 direction so 1/mm = sin milk sin mily

The resonant frequencies are fr = 2 / (mt) 2 + (nt) 2

which are the cutoff frequencies

4-10

Only the general map cases will be computed.

 $\frac{4-10(cont.)}{+(n\pi)^{2}ab} + (\frac{m\pi}{a})^{2}ab \over 4}$ $P_{d}^{550} = \frac{R\pi^{2}}{2a^{2}b^{2}} \left[n^{2}a(b+c) + m^{2}b(c+a) \right]$

$$= M \left[\left(\frac{\mu \pi}{b} \right)^2 \frac{abc}{8} + \left(\frac{m\pi}{a} \right)^2 \frac{abc}{8} \right]$$

Hy = 1 2 m = -1 mt pir comitix sin ntily coopire

=
$$\frac{1}{5} \left(k^2 - \frac{p^2 \pi^2}{c^2} \right) \cos \frac{m \pi x}{a} \cos \frac{n \pi y}{b} \sin \frac{p \pi x}{c}$$

$$|P_{cl}|_{x=0} = \frac{2R}{8^{2}} \left[\frac{|h\pi|^{2}}{b} \frac{2p\pi}{c} \frac{2bc}{4} + \left(k^{2} - \frac{p^{2}\pi^{2}}{c^{2}} \right)^{2} \frac{bc}{4} \right]$$

$$|P_{d}|_{y=0} = \frac{2R}{3^{2}} \left(\frac{P\Pi}{C} \right)^{2} \left(\frac{m\Pi}{A} \right)^{2} \frac{aC}{4} +$$

$$\frac{4-10(cond.)}{+(k^{2}-\rho^{2}\pi^{2})^{2}} + \frac{ac}{4}$$

$$\frac{1}{2} + \frac{1}{2} + \frac$$

expression,

TITM = cookix x e ikz = 0 = x = d 12Tm = wokzx(a-x) e-ikz= dsxca k1x+ k2 = 6,2 = 62/1, 8, kzx + k2 = k2 = w > 12 {2 $E_{21} = \frac{1}{9} \frac{\partial^2 \psi}{\partial x \partial x_1} = \frac{1}{9} \frac{k_2 k_1 x}{2} \sin k_1 x e^{-\frac{1}{9} k_2 x}$

50
$$\frac{4-12(cont.)}{H_{y_1}} = \frac{34}{32} = -\frac{1}{9}k_2 \cos k_{1x} e^{-\frac{1}{9}k_2t}$$
 $H_{y_2} = -\frac{1}{9}k_2 \cos k_{1x} (a-x) e^{-\frac{1}{9}k_2t}$
 $H_{y_2} = -\frac{1}{9}k_2 \cos k_{1x} (a-x) e^{-\frac{1}{9}k_2t}$
 $E_{z_1} \Big|_{x=d} = E_{z_2} \Big|_{x=d}$
 $E_{z_2} \Big|_{x=d} = \frac{1}{2} \exp \left(\frac{a-d}{2}\right)$
 $\lim_{x \to d} = \frac{1}{2} \exp \left(\frac{a-d}{2}\right)$
 $\lim_{x \to d} \frac{1}{2} \exp \left($

Sin
$$k_1 pl = \sin k_{2x}(a-d)$$

Har $k_1 pl = \sin k_{2x}(a-d)$

Har $k_2 l = \lim_{k \to \infty} \lim_$

$$C = \varepsilon_2 d + \varepsilon_1(a-d)$$

For dominant mode, kx, = kx2=0

: kz = w Luz which is the

transmission line mode.

$$\frac{4-14}{\varepsilon_{i}} \approx -\frac{k_{\chi_{1}}^{2}(a-d)}{\varepsilon_{i}}$$

$$f(k_{z},d) = \frac{k_{x_1}d}{\epsilon_1} + \frac{k_{x_2}^2(a-d)}{\epsilon_2} = 0$$

$$k_{\chi_1} = k_1^2 - \left(\frac{\pi}{6}\right)^2 - k_2^2$$

$$k_{xz} = k_z^2 - \frac{\pi}{4})^2 - k_z^2$$

$$+\frac{(a-d)}{\xi_2}\left[\beta_2^2-k_2^2\right]$$

$$f_2 = \frac{\partial f}{\partial k_2} = \frac{-2k_2d}{\xi_1} = \frac{2k_2(a-d)}{\xi_2}$$

Heglecting second order and

$$f(k_2,d) \approx f(\beta_0,0) + f_d(\beta_0,0)d + f_2(\beta_0,0)(k_2-\beta_0)$$

$$\frac{4-14 \left(\text{conth}_{1} \right)}{f \left(k_{2}, d \right) = 0 + \left(k_{1}^{2} - k_{2}^{2} \right) \frac{d}{\xi_{1}}} + \frac{1}{\xi_{2}} \left(-2\beta_{0} \right) \left(a \right) \left(k_{2} - \beta_{0} \right)$$

$$\frac{\left(k_1^2-k_2^2\right)d}{\xi_1}=\frac{2\beta_0}{\xi_2}\left(k_2-\beta_0\right)a$$

$$\therefore k_2 = \beta_0 + \frac{\varepsilon_2(k_1^2 - k_2^2)d}{\varepsilon_1}$$

$$\frac{4-15}{k_{x_1}^2 = k_1^2 - k_2^2}$$

$$k_{x_2}^2 = k_2^2 - k_2^2 = \left(\frac{\pi}{a}\right)^2 + \beta_0^2 - k_2^2$$

$$\beta_0^2 = k_2^2 - \left(\frac{\pi}{a}\right)^2$$

Resignocal & Egn. 4-47;

$$\mu_i\left(d+\frac{d^3k_{y_i}^2}{3}\right) =$$

$$\mu_1 k_{xz} \left[d + \frac{d^3}{3} \left(k_{xz} + k_1^2 - k_2^2 \right) \right] =$$

$$-\mu_2 \left[k_{xz} (a - d) - \overline{\tau}_1 + \left(\frac{k_{xz} (a - d) - \overline{\tau}}{3} \right) \right]$$

Let $k_{\times 2}\alpha - \pi = C_0 + C_1d + C_2d^2 + C_3d^2$ Hegleting fourth ander and higher terms and solve for coefficients by setting terms of

like poners equal.

24-15 (cont) $C_2 = \frac{C_1^2}{\pi} = \frac{(\mu_2 - \mu_0)^2 \pi}{\mu_2} \frac{2}{a^2}$ C = 12-1/1 TT a $C_{3} = -\left(\frac{C_{1} - \frac{\pi}{4}}{3}\right)^{3} + \frac{C_{2}\left(\mu_{2} - \mu_{3}\right)}{a} - \frac{\mu_{1}\pi}{39\mu_{2}}\left(\frac{\pi}{4}\right)^{2} + k_{1}^{2} - k_{2}^{2}\right]$ $C_{3} = \left(\frac{\pi}{a}\right)^{3} \frac{\mu_{1}^{3}}{3\mu_{2}^{3}} + \left(\frac{\mu_{2} - \mu_{1}}{\mu_{2}^{3}}\right)^{\frac{\pi}{a}} - \frac{\mu_{1}\pi}{39\mu_{2}} \left(\frac{\pi}{a}\right)^{2} + k_{1}^{2} - k_{2}^{2}\right)$ $k_{2} = \frac{\pi + \zeta_{1}d + \zeta_{2}d^{2} + \zeta_{3}d^{3}}{a} \text{ and } k_{2} = \sqrt{k_{1}^{2} - k_{1}^{2}}$ $k_{2} = \left[k_{1}^{2} - \left(\frac{\pi}{a}\right)^{2} - \frac{2\zeta_{1}\pi d + \left(\zeta_{1}^{2} + 2\zeta_{2}\pi\right)d^{2} + \left(2\zeta_{3}\pi + 2\zeta_{1}\zeta_{2}\right)d^{3}\right]$ $= \beta_{6} \left[1 - \frac{2c_{1}\pi d + (c_{1}^{2} + 2c_{2}\pi)d^{2} + (2c_{3}\pi + 2c_{1}c_{2})d^{3}}{\beta_{2}^{2}a^{2}} \right]^{1/2}$ $C_1^2 + 2C_2\pi = 3G^2$ $C_1^2 + 2C_1\pi = 3G^2$ Simplifying and using Burismal expansion,

 $k_{2} = \beta_{0} + \frac{(\mu_{1} - \mu_{2})}{\mu_{2} \beta_{0}} \left(\frac{\pi}{a}\right)^{2} \frac{d}{a} - \frac{(\mu_{1} - \mu_{2})^{2}}{\mu_{2}^{2}} \left(\frac{3\beta_{0}^{2} a^{2} + \pi^{2}}{2\beta_{0}^{3} a^{2}}\right) \frac{\pi^{2}}{a^{2}} \left(\frac{d^{2}}{a^{2}}\right)$

which reduces to the result in the book if (M, -M2) =0 in coefficients of higher powers of d. This problem was also done by straight differentiation thus obtaining derivatives of order 3. The results obtained were exactly the same,

4,TM = C, cookx,x sin north sin port 1/2 TM = (260) [kx2(a-x]) Sin 1977 Sin 1978 Ey = -1 (, kx, MTT sin kx, x coonTy sin pro Eyz = 1 (2 kx2 hit sin [kx2 (a-x)] · Coo MTY Sin PTT 2 Ez = = C kx, pm sin kx, x sin mm 4 coopme $E_{22} = \frac{1}{jw} C_2 k_{x_1} P^{\pi} sin(k_{x_2}(e-x))$ · Sin MITY wo PITE From continuity of field quantities: $\frac{1}{\xi_{1}}C_{1}K_{x_{1}}Sink_{x_{1}}d = -\frac{1}{\xi_{2}}C_{2}k_{x_{2}}Sin[k_{x_{1}}(q-d)]$ Hy = PTTC1 cookx, x sin ntry coopTTE Hy = $\frac{\rho\pi}{c}$ coo [kx2(a-x)] sin $\frac{n\pi y}{b}$ coo $\frac{\rho\pi}{c}$ $\frac{1}{2}$ $\frac{1}{2}$ Hz = NTTC1 cookx1x coontry sin prize Hez = MTCz coo[kxz(a-x)] co mty sispte again, from continuity, C, cookxid = Cz coo[kxz(a-d)] Egn. 4-45 is now easily obtained. Similarly for the TEcase, variation in the

gining Egn. 4-47.

2- direction cancels out

From pro6, 4-14; $k_2 \approx \beta_0 + \frac{\epsilon_2}{\epsilon_1} \left(\frac{k_1^2 - k_2^2}{2\beta_0} \right) \frac{d}{a}$ k2 ≈ k2 - (T) 2 + d ε, (k,2-k2) $k_2 = \left(\frac{\pi}{c}\right)^2$ k= k2 = w2 (11, E, -12, E2) (1) + (1) = w/2 Ez + d \(\xi_1 \omega^2 \left(\mu_1 \xi_1 - \mu_2 \xi_2 \right) W2[M2 E2Q E, + d E2 M, E, - d E2 M2 E2] = $a = \left(\left(\frac{\pi}{6} \right)^2 + \left(\frac{\pi}{6} \right)^2 \right)$ $\omega^{2}\mu_{2}\xi_{2} = \frac{q \xi_{1}}{\alpha \xi_{1} + d \left[\frac{\mu_{1}\xi_{1} - \mu_{2}\xi_{2}}{\mu_{2}}\right]} \left[\left(\frac{\pi}{b}\right)^{2} + \left(\frac{\pi}{c}\right)^{2}\right]$ Using binomial theorem : $\int_{\mathcal{W}_{2}^{2}/2} \left[1 - \frac{1}{2} \left(\frac{\mu_{1}}{\mu_{2}} - \frac{\varepsilon_{2}}{\varepsilon_{1}} \right) \frac{d}{d} \right] \left(\frac{\pi}{a} \right)^{2} + \left(\frac{\pi}{c} \right)^{2}$ $\omega_r = \omega_0 \left[1 - \frac{1}{2} \left(\frac{M_1}{M_2} - \frac{\varepsilon_2}{\varepsilon_1} \right) \frac{d}{q} \right]$ where $w_{0} = \frac{1}{\sqrt{M_{2} \xi_{2}}} \left(\frac{\pi}{b} \right)^{2} + \left(\frac{\pi}{c} \right)^{2} \right]^{1/2}$

- TIZM, CZ 3M2 (a24C2) [1-M2](d)

```
Ez= -MITA sink x sin MITY e-1kz for

by= -kxo A nTT cookxoX sin nTT e 1kz 2 0 10
 Ez = - not Book x /x- 2/sin hory = 2/22 )
 Hy = - ky, B NTT sinky (x-2) siny = ike2 (2)
     For continuity at x= a-d , x= a+d ;
         -\sin k_{xo}\left(\frac{a-d}{2}\right) = -\cos k_{xi}\frac{d}{2}
    -\frac{k_{xo}}{M_0} \cos k_{xo} \left(\frac{a-d}{2}\right) = -\frac{k_{xi}}{M_0} \sin \frac{k_{xi}d}{2}
      Dividue egns. gives :
               kxo cot (kxo (a-d)) = kxi tan (kxid)
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                           -N2+k2 = k2 = w/10 80
 E_{z} = \frac{u^{2}}{jw\xi_{x}} A \omega u \times e^{-jk_{z}z}
H_{y} = Au \sin u \times e^{-jk_{z}z}
|X| < \frac{a}{2}
  Hy = NB e NIXI e-iket
E_{z} = \frac{-\sqrt{2}B}{j\omega \xi_{\alpha}} e^{-\nu |x|} e^{-jk_{z}z} \begin{cases} |x|>q \\ 0 \end{cases}
Continuity & Ez & Hy at x=± ques:
                  Au2 com = - + 28 = va/2
                 Ausinua - NBE
           == -1-2
```

मुक्रम्य = ध्रुष्

TM, TEO, TE, . Refer to Figure 4-11.

$$\frac{u_1 a}{2} = \frac{371}{8} \quad \frac{u_2 a}{2} = .7277$$

$$k_2 = \sqrt{k_d^2 - u^2}$$

4-22 a Taylow series expansion To the left side of Equ. 4-56 about a =0 and the right side about v=0 gives:

$$\frac{u^2a^2}{4} = \frac{\xi_d}{\xi_d} \frac{va}{2}$$
on $v = \frac{\xi_0}{\xi_d} \frac{a}{2} u^2$

From Egn 4-55, u2= kg- k2

But
$$k_2 = k_0 + \frac{N^2}{2k_0}$$

So
$$N = \frac{\xi_0}{\xi_d} \frac{\alpha}{2} (k_d^2 - k_0^2)$$

Similarly for TE case Egn: 4-59

$$\frac{u^2a^2}{4} = \frac{\mu d}{\mu_0} \frac{va}{2}$$

$$\Delta v = \frac{\mu_0}{2} \frac{a}{u^2}$$

No 22 27/kg
$$\left(1-\frac{1}{8}\right)$$
 (0.044) (using formula for t small).

N 2 1.158 = attenuation constant

in the x direction. Ot . 86

wavelengths the field decays to

36.870 Bits walno at the

surface. It would be reasonable

to say a tightly bound surface

wous is possible.

4-24 Fild attemates as e-VX

For 36.80% attenuation, NX=1

From Egn 4-70,

value of & for this attenuation.

$$\frac{4-25}{\text{Fer a distance field}}$$
, corrugated slot worse guide,
$$E_{X} = \frac{k_{z}}{WE_{z}}H_{y}$$

$$E_{z} = -\frac{B}{WE_{z}}V^{z}e^{-VX}e^{-jk_{z}z}$$

$$\frac{2}{(-x)} = \frac{E_z}{H_y} = \frac{jv}{w E_d} = j \gamma_d \tan k_d d$$

4-26 Taking a superposition of the modes given in problem 4-7:

$$\frac{4-26 \text{ (cond.)}}{E_{n} = \text{ Neumann number}}$$

$$= 1, N = 0$$

$$= 2, N > 1$$

$$\therefore A_{n} = \underbrace{E_{n}}_{b x_{n}} \int_{b}^{b} \underbrace{E_{n}}_{b} \underbrace{coon\pi u}_{b} dy$$

$$H_{x} = \underbrace{\frac{1}{3} \left(k^{2} - \frac{n^{2}\pi^{2}}{b^{2}}\right)}_{h_{x}} A_{x} \underbrace{coon\pi u}_{b} e^{-\delta_{x} \frac{1}{2}}$$

$$\frac{1}{2} = -\underbrace{E_{y}}_{h_{x}} = \underbrace{\frac{7}{n}}_{n} \underbrace{\left(k^{2} - \frac{n^{2}\pi^{2}}{b^{2}}\right)}_{h_{x}} A_{x} \underbrace{coon\pi u}_{b} e^{-\delta_{x} \frac{1}{2}}$$

Using Equo. 4-32, $E_{ij} = -\frac{\partial Y}{\partial x} = 7_{ij} \sum_{n=0}^{\infty} A_n \overline{v}_n \cos n \overline{v}_i + e^{-7n^2}$

Power per unit width,

From Egn 4-73, & contry dy

For was A. To = CEDEN

From Egn. 4-74, P = (2) (Y0) 0 + 2 = (x) 1 (sin nIIc) }

Vettage accross center of aperture is:

$$P = \frac{c^{2}(Y_{0})_{0}}{b} \left\{ 1 + 2 \frac{g^{2}}{n} \frac{j 2b}{\lambda_{0} \sqrt{n^{2} - (2b/\lambda_{0})^{2}}} \right\}$$

$$\left\{ \frac{\sin n \pi c}{\sqrt{n \pi c/b}} \right\}^{2}$$

This susceptance is twice that given by Egn. 4-18.

$$7_{10}A_{10} = \frac{2}{2} c \frac{a}{2} = \frac{c}{L}$$

$$P = -\int_{0}^{\infty} \frac{\partial}{\partial x} \frac{\partial}{\partial x} A_{n} \cos \frac{n\pi \eta}{b} \frac{1}{b} k^{2} \frac{\partial}{\partial x} \frac{\partial}{\partial$$

which one nonzero only for even values of u.

$$B_{\alpha} = \frac{4\alpha}{2 \eta \lambda_{0}} \sum_{n=1}^{\infty} \frac{1}{\left[(2n)^{2} - \frac{(2b)^{2}}{\lambda_{0}}\right]^{1/2}} \left(\frac{\sin \frac{n\pi c}{b}}{\left(\frac{n\pi c}{b}\right)^{2}}\right)^{1/2} \left(\frac{\sin \frac{n\pi c}{b}}{b}\right)^{2}$$

$$\left(\sin \frac{n\pi c}{b}\right)^{2}$$

$$\frac{4-29 (cont.)}{2_{mn} A_{mn}^{ab}} = \int_{-\frac{\pi}{2}}^{4/2} \sin \frac{\pi \pi}{2} \sin \frac{m \pi x}{a} dx \int_{0}^{2\pi} \cos \frac{n \pi y}{6} dy \frac{4-29 (cont.)}{1 \cdot B_{a}} = \frac{2\pi}{m} - \frac{8c^{2}\lambda}{\pi^{2} b a^{2} \gamma} \sqrt{\frac{m}{2} - \frac{2m}{\lambda}^{2}}$$

$$\frac{3_{mn} A_{mn}}{2 \cdot \epsilon_{m}} = \frac{2\pi}{2} \frac{-8c^{2}\lambda}{\pi^{2} b a^{2} \gamma} \sqrt{\frac{m}{2} - \frac{2m}{\lambda}^{2}}$$

$$= \frac{b}{2\pi} \left[\frac{\sin\left(\frac{m\pi c}{2a} - \frac{\pi}{2}\right)}{\frac{m}{a} - \frac{1}{c}} - \frac{\sin\left(\frac{m\pi c}{2a} + \frac{\pi}{2}\right)}{\frac{m}{a} + \frac{1}{c}} \right]$$

 $= \frac{b}{2} \int_{-\infty}^{\infty} \left[\cos \left(\frac{m\pi x}{a} - \frac{\pi x}{c} \right) - \cos \left(\frac{m\pi x}{a} + \frac{\pi x}{c} \right) \right] dx$

$$-\frac{\sin\left(-\frac{m\pi c}{2a} + \frac{\pi}{2}\right)}{\frac{m}{a} - \frac{1}{c}} + \frac{\sin\left(-\frac{m\pi c}{2a} - \frac{\pi}{2}\right)}{\frac{m}{a} + \frac{1}{c}}$$

$$= \frac{b}{2\pi} \left[-\frac{2\cos\left(\frac{m\pi c}{2a}\right)}{\frac{m}{a} - \frac{1}{c}} + \frac{2\cos\left(\frac{m\pi c}{2a}\right)}{\frac{m}{a} + \frac{1}{c}} \right]$$

$$= \frac{b}{\pi} \cos\left(\frac{\pi\pi c}{2a}\right) \left[\frac{2/c}{\frac{1}{a^2}}\right]$$

$$= \sum_{m=1}^{\infty} (y_0^*)_{mo} \left(\frac{\cos \frac{m\pi c}{2a}}{1 - (\frac{mc}{a})^2} \right)^2 \frac{16c^2ab}{\pi^2a^22}$$

$$y_a = \frac{P}{|v|^2} = \frac{P}{6^2} = 6a + j Ba$$

$$\frac{4-24 (cont.)}{1-\frac{4}{2}}$$

$$\frac{1}{1-\frac{4}{2}} = \frac{\frac{8c^{2}\lambda}{\pi^{2}ba^{2}\eta} \sqrt{\frac{m}{2}}^{2} - \frac{4}{\lambda}^{2}}{\frac{2q}{1-\frac{4mc}{2}}^{2}}$$

$$\frac{4-30}{-678_{A}} = \frac{2}{\pi^{2}} \left(\frac{c}{a}\right)^{2} \frac{c}{2} \frac{\sin \frac{m\pi c}{a}}{1-\frac{mc}{a}^{2}}$$

$$\sqrt{\frac{m}{2}^{2}-\frac{c}{A}^{2}}$$

$$f = -\frac{67Ba}{\lambda} \quad \frac{mC}{a} \sim x$$

$$\frac{c}{a} \sim dx$$

$$dx = x$$

Let
$$u = \sin^2 \pi x$$
 dut = $\frac{x dx}{(1-x^2)^2}$

$$\int \frac{\left(\sin \tau \tau x\right)^{2}}{1-x^{2}} dx$$

$$=\frac{2\sin^2\pi x}{(1-x^2)}\left|-\frac{x}{2}\int_{0}^{\infty}\frac{\sin 2\pi x}{(1-x^2)}dx\right|$$

0 4-30 (und)

 $\frac{1}{\lambda} - \frac{676a}{\lambda} \rightarrow \frac{77}{277} Si(277) = \frac{Si(277)}{277}$

4-31 Chaose mode functions as :

4+ = = 3+ contry = 7/2 , 2 >0

7-= = B B CONTTY = 3n2 , 200

Continuity of Ex and Ey at ==0

=> Bn = Bn = Bn.

Ju = [Hx - Hx] | 2=0

Hx = 34 => Jy = \ Z 2 on By coontry D

Let An = Vn Bn

hen & An coo MITTY e -7/12/ = Hx , 2>0

=-H*, 2 <0

and $A_n = \frac{\epsilon_n}{2b} \int_{\mathbb{R}} J_y \cos \frac{n\pi y}{b} dy$ from 0.

4-32

4+= = A+ sin nmg e - 2,2 >0

N= 2 An Sin ninge on?

 $H_{\mathcal{G}} = \frac{\partial}{\partial x} A_{x}$

Jx = [HJ -Hb] = = = 27, An SinnTy

=x = jwy. Ax

An = The Jb sin ntry e - on tel

on if B = jum An then

4-32 (cont.)

Bu = jw/ Job Sin MITH dy

and Ex = \frac{2}{2} B_n \sin \frac{n\pi}{b} e^{-\frac{2}{3}(1+1)}

4-33 # 220

4+= 2 2 8 8mm coominx sin nity e - 7mm2

For ZCO,

7 = 2 5 Bmm women singly e mm2

at 2=0, B+ = Bmn = Bmn

Jx = [Hy - Hy] 2=0

Ty = [++ + +=] y =0

Hy = 24 = - 70, 80, sin my

Hz = - dx = T Box 00 Ty

Jx = 2 80, 80, sin Ty

J, =0

 $J_x = \omega \kappa \times \delta(y-c)$

Jax J sin Ty Jx dy = To, Bo, ba

To, Bo1 = 1 Scookx dx Simmy S(y-c) dy

= ab [sinked sin TC]

 $P = -\iint E \cdot J_s^* ds$ $= -\iint dx \int dy J_k^* E_x |_{z=0}$

$$\frac{4-33 \ (cont.)}{E_{X}} = \frac{1}{9} \frac{k^{2} + 4}{k^{2}}$$

$$= \frac{1}{9} \frac{1}{9} \frac{k^{2} + 4}{k^{2}}$$

$$= \frac{1}{9} \frac$$

$$\frac{1}{R_{yh}} = \frac{\rho^{2}}{I_{in}} = \cos ky \Big|_{y=0} = 1$$

$$\frac{1}{R_{x}} = \frac{2\sigma}{ab} \Big[\frac{\sin \pi \zeta}{k} \Big[\frac{\sin k(d+c) - \sin kc}{k} \Big] \Big]$$

$$\frac{4 - 35}{\sin^{2}x} \approx x^{3} - \frac{x^{3}}{6} + \frac{x^{5}}{120} - \frac{x^{7}}{5040}$$

$$\frac{1}{\sin^{2}x} \approx x^{2} - \frac{x^{4}}{3} + \frac{2x^{6}}{45} - \frac{x^{6}}{315}$$

$$\frac{\sin^{2}x}{x^{2}} \approx 1 - \frac{x^{2}}{3} + \frac{2x^{4}}{45} - \frac{x^{6}}{315}$$

$$\frac{1}{x^{2}} \approx 1 - \frac{x^{2}}{3} + \frac{2x^{4}}{45} - \frac{x^{6}}{315} dx$$

$$\frac{1}{x^{2} - x^{2}} \approx 1 - \frac{x^{2}}{3} + \frac{2x^{4}}{45} - \frac{x^{6}}{315} dx$$

$$\frac{1}{x^{2} - x^{2}} \approx 1 - \frac{x^{2}}{3} + \frac{2x^{4}}{45} - \frac{x^{6}}{315} dx$$

$$\frac{1}{x^{2} - x^{2}} = \frac{\pi}{2} \left[\frac{ka}{2} \right]^{2} - \frac{1}{2} \left[\frac{ka}{2} \right]^{2} - \frac{1}{2} \left[\frac{ka}{2} \right]^{4} - \frac{1}{2} \left[\frac{ka}$$

$$A \eta B_{\alpha} = I = \int_{\omega}^{\omega} \frac{d\omega}{\sqrt{\omega^{2} + 6^{2}}} - Re \int_{\omega}^{\omega} \frac{e^{2j\omega}}{\sqrt{\omega^{2} - 6^{2}}} d\omega$$
where $b = \frac{k\alpha}{2}$
there $b = \frac{k\alpha}{2}$
there $b = \frac{k\alpha}{2}$
there $b = \frac{k\alpha}{2}$
there $b = \frac{1}{2} - Re \int_{\omega}^{\omega} \frac{e^{j\omega}}{\sqrt{\omega^{2} - 6^{2}}} d\omega$

$$C_{1} = \frac{1}{6^{2}} - Re \int_{\omega}^{\omega} \frac{e^{j\omega}}{\sqrt{\omega^{2} - 6^{2}}} d\omega$$
The probability of $\frac{1}{2} - Re \int_{\omega}^{\omega} \frac{e^{j\omega}}{\sqrt{\omega^{2} - 6^{2}}} d\omega$

$$C_{1} = \frac{1}{6^{2}} - Re \int_{\omega}^{\omega} \frac{e^{j\omega}}{\sqrt{\omega^{2} - 6^{2}}} d\omega$$
Where $\omega = 6 + j\alpha$
Naw as $b = \frac{1}{6^{2}} - Re \int_{\omega}^{\omega} \frac{e^{-2\alpha}}{\sqrt{j}} dx$

$$I = \frac{1}{6^{2}} - Re \int_{\omega}^{\omega} \frac{e^{-2\alpha}}{\sqrt{j}} dx$$
Using Duright 860.05,
$$I = \frac{1}{6^{2}} - \frac{\sqrt{\pi}}{2 \sqrt{572}} \left(\cos 2b - \sin 2b \right)$$

$$I = \frac{1}{6^{2}} - \frac{\sqrt{\pi}}{2 \sqrt{572}} \cos \left(2b + \pi/4 \right)$$

$$I = \frac{1}{6^{2}} - \frac{\sqrt{\pi}}{2 \sqrt{572}} \cos \left(2b + \pi/4 \right)$$

$$I = \frac{1}{2} - \frac{\sqrt{\pi}}{2 \sqrt{572}} \cos \left(2b + \pi/4 \right)$$

$$I = \frac{1}{2} - \frac{\sqrt{\pi}}{2 \sqrt{572}} \cos \left(2b + \pi/4 \right)$$

$$I = \frac{1}{2} - \frac{\sqrt{\pi}}{2 \sqrt{572}} \cos \left(2b + \pi/4 \right)$$

$$I = \frac{1}{2} - \frac{1}{2}$$

Expanding
$$[(\pi)^2 - \omega^2]^2$$
 about $\omega = 0$, we get:

 $f = a^4 \left(1 + 2a^2 \omega^2 + 3a^4 \omega^4 + 4a^4 \omega^6 + 5a^8 \omega^8 + ... \right)$
 $600^2 \omega \approx 1 - \omega^2 + \frac{\omega^4}{3} - \frac{2\omega^6}{45} + \frac{\omega^8}{315}$
 $f \cos^2 \omega \approx a^4 \left(1 + \omega^2 \left[2a^2 - 1 \right] \right)$
 $+ \omega^4 \left[\frac{1}{3} - 2a^2 + 3a^4 \right]$
 $+ \omega^6 \left[\frac{1}{3} - 2a^2 + 3a^4 \right]$
 $+ \omega^6 \left[\frac{1}{3} - \frac{4a^2}{45} + a^4 - 4a^6 + 5a^8 \right]$
 $= a^4 \left\{ 1 + \omega^2 \left(-.1894305 \right) + \omega^6 \left(-.000740432 \right) + \omega^6 \left(-.000740432 \right) + \omega^6 \left(-.000740432 \right) \right\}$
 $f t d \omega = \frac{b^2}{2} \left(\frac{\pi}{2} \right) = \left(\frac{a}{3} \right)^2 \frac{\pi^3}{4}$

$$\int_0^b t d \omega = \frac{b^4}{3} \left(\frac{\pi}{2} \right) = \left(\frac{a}{3} \right)^4 \frac{\pi^5}{16}$$

$$\int_0^b t \omega^2 d \omega = \frac{b^4}{3} \left(\frac{\pi}{2} \right) = \left(\frac{a}{3} \right)^6 \frac{\pi^7}{16}$$

$$\int_0^b t \omega^4 d \omega = \frac{b^4}{3} \left(\frac{\pi}{2} \right) = \left(\frac{a}{3} \right)^6 \frac{\pi^7}{32}$$

$$\int_0^b t \omega^4 d \omega = \frac{b^4}{3} \left(\frac{\pi}{2} \right) = \left(\frac{a}{3} \right)^6 \frac{\pi^7}{32}$$

$$\int_0^b t \omega^4 d \omega = \frac{5b^8}{16} \left(\frac{\pi}{2} \right)$$

To find confined the above identities are used to integral

62-w2 f 6002w dw

Factor out $\frac{2}{\pi}$ and compute first four coefficients, $b_1 = \frac{1}{2} \left(\frac{2}{\pi}\right)^3 \frac{\pi^3}{4} = 1.0$ $b_2 = \frac{1}{2} \left(\frac{2}{\pi}\right)^3 \frac{\pi^5}{16} \left(-.1894305\right)$

$$b_3 = \frac{1}{2} \left(\frac{2}{\pi} \right)^3 \frac{\pi}{32} (.01553101)$$

$$b_4 = \frac{1}{2} \left(\frac{2}{\pi} \right)^3 \frac{5\pi^9}{(128)2} \left(-.000740432 \right)$$

· · etc.

Let u = 100 2 W dv = wdw

du = - zcoswsinwdw [(E) -w2]

Use identity:

$$\int_{0}^{\infty} \frac{\sin 2x}{(\pi)^{2}-x^{2}} dx = \frac{2}{\pi} \operatorname{Si}(\pi)$$

$$\frac{4-38\cdot(cont.)}{-\lim_{\omega\to0}\left[\frac{\cos^2\omega}{2([\frac{\omega}{2})^2-\omega^2]}\right]=-\frac{2}{\pi^2}$$

$$\frac{\eta_{G_q}}{\lambda} = \frac{1}{2} \int_{0}^{\frac{h_q}{2}} \sqrt{\frac{(\frac{n_q}{2})^2 - \omega^2}{(\frac{n_q}{2})^2 - \omega^2}} \cos^2 \omega d\omega$$

large even when w gets close to kea because the denormiates is increasing much faster than the numerator in the integrand.

$$\frac{1}{\lambda} \frac{\gamma_{6a}}{\frac{1}{2} + 2} = \frac{ka}{8} \operatorname{Re} \int \frac{(1 + e^{2\delta \omega})}{(\sqrt{(\frac{\pi}{2})^2 - \omega^2)^2}} d\omega$$

$$\frac{1+e^{2j\omega}}{\left(\left(\frac{\pi}{2}+\omega\right)\left(\frac{\pi}{2}-\omega\right)\right)^2}$$
 has a pole of
$$\left(\left(\frac{\pi}{2}+\omega\right)\left(\frac{\pi}{2}-\omega\right)\right)^2$$
 and a 1 at $\omega=\frac{\pi}{2}$

$$\int_{C_0}^{\infty} = -\pi j \left[\text{Residue} \right]$$

$$=-\pi i\left(\frac{-2i}{\pi i}\right)=-\frac{2}{\pi}$$

Over Cz, w= 0+jx

$$Re \int = -\frac{ka}{8} Re \int \frac{(1+e^{2i(jx)})}{[[Ty_2]^2-(jx)^2]} dx$$

Integrand is a real no.

Co+C1+C2+C2 the integrand is analytic inside the contour

and
$$\frac{\eta Ga}{\lambda} = \frac{ka}{8} \left(\frac{2}{\pi t}\right) = \frac{ka}{4\pi t}$$

at y = 0 assume Ex=1, ocxca.

$$\overline{E}_{x} = \int_{0}^{\infty} E_{x}(x,o) e^{-jk_{x}x} dx$$

(bar above quantity indicates transform) Ex= J-e-8 kxx + Jae-3 kxx $= \frac{2}{ib} \int (-\sin k_x a) = \frac{4}{ib} \sin^2 \frac{k_x a}{2}$

$$\frac{4-40 \text{ (cond.)}}{P=-\int \left[\bar{E}_{x} \bar{H}_{e}^{+}\right]_{y=0}} \frac{dk_{x}}{2\pi i}$$

$$= -\frac{16}{100} \int_{0}^{\infty} \sin^{4} \frac{k_{x}q}{2} dk_{x}$$

$$= -\frac{16}{\lambda \eta} \int_{-\sigma}^{\sigma} \frac{\sin^4 \frac{k_x q}{2}}{k_y^2 k_x^2} dk_x$$

Vottage of one him is
$$V = \int_{0}^{a} E_{x} dx = a$$

$$y_{a} = \frac{p^{*}}{|V|^{2}} = \frac{-16}{\lambda \pi a^{2}} \int_{-\infty}^{\infty} \frac{\sin^{4}k_{x}a}{k_{y}k_{x}^{2}} dk_{x}$$

$$G_a = \frac{-16}{\lambda \pi a^2} \int \frac{\sin^4 k_x a}{2} dk_x$$
 $-k - (\sqrt{k^2 - k_x^2}) k_x^2$

$$Ga = \frac{32}{\lambda \gamma} \int_{\omega}^{ka} \frac{\sin^4 \omega}{2} d\omega$$

$$Ba = \frac{-16}{\lambda \eta a^2} \left(\int_{-\infty}^{k} k \int_{k}^{\infty} \frac{\sin^4 \frac{k_x a}{2}}{\sqrt{k_x^2 - k^2 k_x^2}} dk_x \right)$$

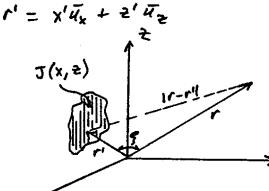
$$B_{a} = \frac{32}{\lambda^{2}} \int_{ka}^{a} \frac{\sin^{4} \frac{\omega}{2}}{\omega^{2} \sqrt{\omega^{2} - (ka)^{2}}} d\omega$$

4-41 It is not clear to the authors how breen's folentity can be used to show this equivalence to the two 4's. The extension to the far field as stated in the problem, however, is trivial as follows.

of by potential integral method

$$\psi = \int \int \frac{J(x,z)e^{-jk|r-r'|}}{4\pi |r-r'|} dx'dz'$$

$$r = x \overline{u}_x + y \overline{u}_y + z \overline{u}_z$$



Specializing to far field,

≈ r + x'sin 0 coo \$ + 2' coo 0

$$\psi = \frac{e^{-jkr}}{4\pi r} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} J(x,z) e^{-jk(x\sin\theta\cos\phi + 2\cos\phi)} dxdz$$

$$\overline{J}_{z} = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} J(x,z) e^{-jk_{x}x} e^{-jk_{z}z} dxdz$$

$$= \overline{J}_{z}(k_{x},k_{z})$$

kx = k sin o coop

kz = k 000 0

$$\therefore \psi = \frac{e^{-j kr}}{4\pi r} \overline{J}_{2}(-k\omega\phi \sin\phi - k\omega\phi)$$

$$\frac{4-42}{2} = \frac{p}{|\mathbf{II}|^2}$$

J_k= 1次1 くま

False, ribb, = $\frac{\eta^2}{2}$ Yapant

4-42 (cont.)

Yapent is given by Egn. 4-114 because the Efield (in Fig. 4-23, in the aperture is in the same direction as the current in the ribbors. (dual to aperture).

Hence
$$\frac{1}{p} \frac{\partial}{\partial p} \left(P \frac{\partial \psi}{\partial p} \right) = \sum_{n} \int_{n}^{\infty} q_{n}(k_{p}) B_{n}(k_{p}) h(n\phi) h(k_{3}3) dk_{p}$$

$$+ \sum_{n} \int_{n}^{\infty} q_{n}(k_{p}) B_{n}(k_{p}) h(n\phi) h(k_{3}3) dk_{p}$$

$$+ \sum_{n} \int_{n}^{\infty} q_{n}(k_{p}) B_{n}(k_{p}) h(n\phi) h(k_{3}3) dk_{p},$$

$$\frac{1}{p} \frac{\partial^{2}\psi}{\partial \phi^{2}} = \sum_{n} \frac{n^{2}}{p^{2}} \int_{n}^{\infty} q_{n}(k_{p}) B_{n}(k_{p}p) h''(n\phi) h(k_{3}3) dk_{p}, \quad \text{and}$$

$$\frac{\partial^{2}\psi}{\partial z^{2}} = \sum_{n} \int_{n}^{\infty} k_{p} g_{n}(k_{p}) g_{n}(k_{p}p) h(n\phi) h''(k_{3}3) . \quad \text{So},$$

$$\frac{1}{p} \frac{\partial}{\partial p} \left(P \frac{\partial \psi}{\partial p} \right) + \frac{1}{p} \frac{\partial^{2}\psi}{\partial \phi^{2}} + \frac{\partial^{2}\psi}{\partial z^{2}} + k^{2} \psi$$

$$= \sum_{n} \int_{n}^{\infty} q_{n}(k_{p}) \left[\frac{k_{p}}{p} \frac{B_{n}(k_{p}p)}{B_{n}(k_{p}p)} + \frac{k_{p}^{2}}{p} \frac{B_{n}'(k_{p}p)}{h(n\phi)} + \frac{k_{p}^{2}}{p^{2}} \frac{h''(k_{p}3)}{h(n\phi)} + k_{p}^{2} \frac{h''(k_{p}3)}{h(k_{p}3)} + k^{2} \right]$$

$$\times g_{n}(k_{p}p) h(n\phi) h(k_{3}3) dk_{p}$$

$$= \sum_{n} \int_{n}^{\infty} q_{n}(k_{p}p) \left[\frac{k_{p}}{p} \frac{B_{n}'(k_{p}p)}{B_{n}(k_{p}p)} + \frac{k_{p}^{2}}{p} \frac{h''(k_{p}3)}{h(k_{p}3)} + k^{2} \frac{h''(k_{p}3)}{h(k_{p}3)} + k^{2} \right]$$

$$\times g_{n}(k_{p}p) h(n\phi) h(k_{3}3) dk_{p}$$

$$= \sum_{n} \int_{n}^{\infty} q_{n}(k_{p}p) \left[\frac{k_{p}}{p} \frac{B_{n}'(k_{p}p)}{B_{n}(k_{p}p)} + \frac{k_{p}^{2}}{p} \frac{h''(k_{p}3)}{h(k_{p}3)} + k^{2} \frac{h''(k_{p}3)}{$$

The above equation satisfies Helmholtz equation because the component equations satisfy equation (5-7).

5.2 Since
$$\psi = (\log \rho) e^{-jk^2}$$
 hence $\frac{\partial \psi}{\partial \rho} = \frac{1}{\rho} e^{-jk^2}$; $\frac{\partial \psi}{\partial s} = -jke (\log \rho)$ and so $\frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho \frac{\partial \psi}{\partial \rho}) = 0$; $\frac{\partial \psi}{\partial s^2} + k^2 \psi = 0$.

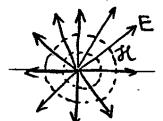
Hence Y satisfies the scalar Helmholtz equation.

Here The case

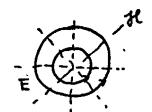
$$\begin{aligned}
&\text{Hom The case} \\
&\text{Hom The case} \\
&\text{Hom The case} \\
&\text{Hom The case} \\
&\text{Ep = $\frac{1}{9}$} \left(\frac{-jk}{p} \right) e^{-jk2} \\
&\text{Ep = 0} \\
&\text{Hom The case} \\
&\text{Ep = 0} \\
&\text{Hom The case} \\
&\text{Ep = 0} \\
&\text{Ep = 0} \\
&\text{Hom Ep = 0} \\
&\text{Hom Ep = 0} \\
&\text{Hom Ep = 0}
\end{aligned}$$

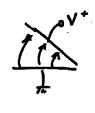
$$\begin{aligned}
&\text{Hom The case} \\
&\text{Ep = $\frac{1}{2}$} \left(-\frac{jk}{p} \right) e^{-jk2} \\
&\text{Ep = $\frac{1}{p}$} e^{-jk2} \\
&\text{Hom Ep = 0} \\
&\text{Hom Ep = 0}
\end{aligned}$$

$$\begin{aligned}
&\text{Hom The case} \\
&\text{Ep = 0} &\text{Hom Ep = 0} \\
&\text{Hom Ep = 0}
\end{aligned}$$



Sketch of Ed Il lines





The coaxial cable supports the above fields This system would support such fields

$$\overline{A} = \overline{u}_1 P Y_1$$
 and $\overline{f} = \overline{u}_1 P Y_2$ and $\frac{\partial Y_1}{\partial z} = \frac{\partial Y_2}{\partial \overline{z}} = 0$

$$\overline{E} = \overline{u^3} \frac{3h^2}{3h^2} + \frac{1}{3} \left[-\overline{u^4} \frac{1}{1} \frac{1}{3h^1} + \overline{u^6} \frac{3h^2}{3h^1} \right] - \cdots 0$$

$$\vec{H} = -\vec{u}_3 \frac{\partial \phi}{\partial \phi} + \frac{3}{3} \left[-\vec{u}_0 + \frac{1}{2} \frac{\partial \phi_2}{\partial \phi_2} + \vec{u}_0 + \frac{\partial \phi_3}{\partial \phi_2} \right] \cdots (3)$$

Since 4, and 42 must be solutions to

$$\nabla^2 \left(\frac{Ap}{p} \right) + k^2 \left(\frac{Ap}{p} \right) = 0$$
 and

Townify gauge conditions, take $\hat{\rho}$ and $\hat{\phi}$ component

of equation
$$(3-78)$$
.

of equation (3-18).

Hor A,
$$-\frac{1}{p}\frac{5\psi_{1}}{7p^{2}} - \psi^{2}\psi_{1} = -\frac{9}{3}\frac{3\phi^{2}}{7p^{2}}$$
...(3)

and
$$\frac{\partial \varphi_0}{\partial \varphi_1} = -\frac{1}{2}\frac{\partial}{\partial \varphi} \varphi^{\alpha}$$
 or $\frac{\partial \varphi_1}{\partial \varphi} = -\frac{1}{2}\varphi^{\alpha} + c(\varphi)$

[after integrating once with respect to p]

Substitution of (4) into (3),

Integrating with respect to p

48 [5-3] continued
$$\rho\left(\frac{\partial t_1}{\partial \rho}\right) = -\frac{1}{2}\rho^2 + C_1(\rho)$$
; $C_1(\rho) = C(\rho)$

Hence $C_1(\rho) = C(\rho) = 0$; and $\rho^2 = -\frac{1}{2}\rho\frac{\partial t_1}{\partial \rho} = -\frac{1}{2}\rho\frac{\partial}{\partial \rho}\left(\frac{A\rho}{\rho}\right)$

We must show that $\frac{\partial t}{\partial \rho}$ is an arbitrary solution to the wave equation.

 $\nabla^2 \left[\frac{\partial t}{\partial \rho}\right] + k^2 \left[\frac{\partial t}{\partial \rho}\right] = 0$ for either t_1 , or t_2 .

or $\frac{\partial}{\partial \rho} \left[\nabla^2 t + k^2 t^2\right] = 0$; hence $\nabla^2 t + k^2 t = f(\rho)$

or $\frac{1}{2}\rho^2 \left(\frac{\partial t}{\partial \rho}\right) + \frac{1}{2}\frac{\partial^2 t}{\partial \rho^2} + k^2 t = f(\rho)$

The above equation has a solution consisting of two parts of the solution of the solution of the solution is equal.

(1) the homogeneous equation (2) particular integral : Y = R(P) + Yhombgeneous

5-4 fc = 9000 MHz. Hence λc = 10/3 cm = 2πα/1.841 $a = \frac{18.41}{(6\pi)} = 0.977$

The cut off frequencies for the next ten lowest order modes are 14.3, 19.1, 20.6, 23.8, 26.2, 28.3, 31.4, 32.3, 32.7 and 36.4 GHZ. For En= 9 aff the fe's are divided by 2

5-5 lynen
$$Y = B_n(k_p p) \ell(n\phi) e^{\pm jk_3 3}$$

The TM modes,

 $E_p = \frac{1}{99} (\pm jk_3) B_n(k_p p) e^{\pm jk_3 3} n \ell'(n\phi) \quad E_\phi = 1$
 $= 0 |_{p=a}$. Hence $B_n(k_p a) = 0$.

Now since the fields do not go to zero inside the waveguide Bn (kpp) = A Jn (kpp) + 8 Yn (kpp) sence Bn (kpp)=[Yn (kpa) Jn (kpp) -Jn(kpa) Yn(kpl)]

times a constant. The & variation could be simp or comp or could be a combination of bothe depending on the boundary conditions

 $E_{\phi} = B_{n}'(k_{\rho}P) k_{\rho} h(n\phi) e^{\pm jk_{\beta}\delta}$ = 0|p=a Hence Bn (kpa) = 0. Following orguments of the TM case Bn (&pp) = [Yn' (kpa) Jn (&pp) - J'(k,a) Y(k, ?)

For TEmodes

tures a constant. The proviotion could be sin mp or comp or a combination of both, depending on the · han dage conditions.

5-5 condd.

Since also B_n(kpb) = 0 or Ep=0

Hence kp is a roof of the equation

Y_n(kpb) J_n(kpa) - J_n(kpb) Y_n(kpa)=0.

Similarly as $E_p=0|_{b=0}$ Hence k_p is a roof of the equation $J'_n(k_pb)Y'_n(k_pa)-Y'_n(k_pb)J'_n(k_pa)=$

[5-6] For the coaxial wavefuide with a baffle $h(n\phi)$ is assumed to be $h(n\phi) = A \cos n\phi + B \sin n\phi$

Hence A cosn ϕ + B sin ϕ = 0 at ϕ = 0. Thence A cosn ϕ + B sin ϕ = 0 at ϕ = 0. A = 0. and B sin 2π n = 0, so $\sin 2n\pi = \sin \frac{\pi}{2}$. $\therefore n = \frac{\pi}{2}$ for $\frac{\pi}{2}$, $\frac{\pi}{2}$, $\frac{\pi}{2}$... For TE modes, $E_f = 0 \text{ ad } \phi = 0 \text{ d } \phi = 2\pi$ So $k'(m\phi) = 0$.

or $n[A\sin n\phi + 6\cos n\phi] = 0$ $\therefore B = 0$ and $\sin n2\pi = 0 = \sin \frac{k\pi}{2\pi}$ $n = \frac{k}{2} \text{ for } k = 1,2,3,\cdots$ hence $h(n\phi) = \cos n\phi$

[5-7] For the wedge waveguide the boundary conditions are for TH modes i) $E_p = 0$ at p = 0 for TE modes

 $f_{N} = 0$ at f = 0Thence $f_{N} = 0$ at f = 0Thence $f_{N} = 0$

hence $h(n\phi) = \sin n\phi$

ii) $E_p = 0$ at $\phi = 0$; Acom $\phi + B$ sui $\phi = 0$: A = 0 at $\phi = \phi$. : $A = \frac{k\pi}{\phi_0}$

Hence $\psi^{TM} = J_n(k_p f) sin p e^{\pm jk_g g}$ when $n = \frac{k\pi}{60}$ for $k = 1, 2, 3, \cdots$

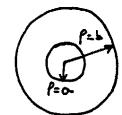
i) $\xi p = 0$ at p = 0 hence $J_n'(kpa) = 0$ and $\frac{d}{dp} [Accomp + Beinp] = 0$ at $\frac{dp}{dp}$ Hence B = 0 and $\sin np = \sin k\pi$ so $\psi^{TE} = J_n(kpp) \cos np e^{J_1k_3}$ for n = 0, $\frac{\pi}{p_0}$, $\frac{2\pi}{p_0}$, ...

Since the dominate mode is the TE mode $\psi^{TE} = J_n(k_p p) e^{-jk_3 s} \left[A \sin \phi + B \sin n \phi \right]$ $E_p = -\frac{1}{p} J_n(k_p p) e^{-jk_3 s} \left[nA \cos n \phi - Bn \sin n \phi \right] = 0 \text{ at } \left\{ \phi = 2\pi \right\}$ Hence $n = \frac{k}{2}$; A = 0. Again $E_{\phi} = 0$ at p = 0 and f = a $\vdots J_n'(k_p a) = 0 = J'(k_p a)$ $i'(a) \left[\pi - i' \right] = 0$

 $\frac{1}{10}(2) = \sqrt{\frac{17}{22}} J_{n+\frac{1}{2}}(2) = 0$ for 2 = 1.16; $\lambda_c = \frac{2\pi a}{1.16}$

$$\frac{1}{5-9} \frac{1}{4\pi} \frac{TM \text{ modes}}{TM} = J_n(k_p f) \cos np e^{-jk_3} \\
E_p = \frac{1}{9} J_n'(k_p f) k_p (-jk_3) \cos np e^{-jk_3} \\
E_p = -\frac{n}{9} \sin np J_n(k_p f) (-jk_3) e^{-jk_3} \\
E_p = -\frac{n}{9} \sin np J_n(k_p f) (-jk_3) e^{-jk_3} \\
H_p = -\frac{n}{p} J_n(k_p f) \sin np e^{-jk_3} \\
H_p = -\frac{n}{p} J_n(k_p f) k_p \cos np e^{-jk_3} \\
H_p = -J_n'(k_p f) k_p \cos np e^{-jk_3} \\
H_p = -J_n'(k_p f) k_p \cos np e^{-jk_3} \\
E_p = \frac{n}{9} \int_{0}^{2\pi} dp \left\{ (H_p)^2 + (H_p)^2 \right\}_{p=a;3=0}^{2} \\
= k_p^2 a R \pi J_n'^2 (k_p a) \\
P_p = \int_{0}^{2\pi} f dp \int_{0}^{2\pi} dp \left[E_p H_p^* - E_p H_p^* \right] \\
= k_3 \pi k_p \int_{0}^{2\pi} \{ k_p f \right) J_n'^2 (k_p f) + \frac{n^2}{k_p f} J_n'(k_p f) \int_{0}^{2\pi} e^{-jk_3} \\
= \frac{\pi k_3}{9} \frac{k_p^2 a^2}{2} J_n'^2 (k_p a) \\
\therefore \alpha = \frac{P_d}{2P_f} = \frac{R^2}{4k_3} = \frac{R}{7a \sqrt{1 - (\frac{f_c}{f_c})^2}}$$

5-10



4 = [A J, (k1) + BY, (k1)] e-)

For TM to 3

: A
$$J_n(ka) + B Y_n(ka) = 0$$

A $J_n(kb) + B Y_n(kb) = 0$

Hence
$$-\frac{B}{A} = \frac{J_n(ka)}{Y_n(ka)} = \frac{J_n(kb)}{Y_n(kb)}$$

 $\frac{\text{Ter TEmodes}}{\Psi^{TE}} = J_n(k_p p) con p e^{-jk_3 3}$ $E_p = \frac{n}{p} J_n(k_p p) suin p e^{-jk_3 3}$ $E_{\varphi} = J_n'(k_p p) k_p con \varphi e^{-jk_3}$ $Hp = \frac{1}{3} k_p J_n'(k_p p)(-jk_3) cosn p e^{-jk_3}$ $H_{\varphi} = -\frac{\eta}{3} \sum_{k} sump J_{n}(k_{p}p)(-jk_{3})e^{-jk_{3}}$ Hz = 1 Jn (kpf) & comp e - 1k3 } P2 = Q { [p d p ([Hp]2+[H3]2)] 3=0, p=0 = \frac{\pi Ra [J_n^2(kpa) kp4 - \frac{n^2 k_3^2 J_n^2(ka)}{a^2}] Pf = \ \ ap | \ Ep Hp - Ep Hp + | = $\frac{\pi k_3}{3} \left[\int_{0}^{\infty} \left[\frac{n^2}{p} J_n^{-1}(k_p a) + \rho J_n^{-1}(k_p a) k_p^{-1} \right] \right]$ = $\frac{\pi k_1}{3} \frac{k_1 a^2}{2} \left(1 - \frac{n^2}{k_1 a^2}\right) J_n^2(k_1 a)$ $d = \frac{Pd}{zPf} = \frac{R}{\eta \alpha \sqrt{1 - (\frac{f_c}{f})^2}} \left[\frac{kp^2}{k^2} + \frac{n^2}{kp^2a^2 - n^2} \right]$ since R= bp+ + & 2 & & & pa = x'np 1 x = R na \(\frac{fc}{f} \) = \[\left(\frac{fc}{f}\right)^2 + \left(\frac{n^2}{x'_{np}}\right)^2 - n^2 \] \(\frac{fc}{f} \)

For TE to 3 $E p = \frac{\partial \Psi}{\partial p} = 0 \text{ at } p = a \& P = b.$

:. A
$$J_{n}'(ka) + B Y_{n}'(ka) = 0$$
A $J_{n}'(kb) + B Y_{n}'(kb) = 0$
... - B $J_{n}'(ka) = J_{n}'(ka)$

Hence
$$-\frac{B}{A} = \frac{J_n'(ka)}{Y_n'(ka)} = \frac{J_n'(kb)}{Y_n'(kb)}$$

5-11 By the terms of the problem
$$Y_{mn} = cos(\frac{m\pi z}{a})cosnp H_n^{(1)}(k_p p)$$

$$:E_{z} = \frac{\left[k^{2} - \frac{m^{2}\pi^{2}}{a^{2}}\right]}{9} cos(\frac{m\pi^{2}}{a})cosnpH_{n}^{(1)}(kf)$$

$$H_{\phi} = -\cos\left(\frac{m\pi^2}{\alpha}\right)\cos \varphi H_n^{(1)}(k_{\rho}\rho)$$

From (5.36)

$$\beta_{p} \text{ of } E_{z} = \frac{\partial}{\partial p} \left[\frac{1}{\text{ten}} \frac{Y_{n}(k_{p}p)}{J_{n}(k_{p}p)} \right]$$

$$= \frac{Y_{n}'J_{n} - J_{n}'Y_{n}}{J^{2} \left[1 + \frac{Y_{n}^{2}}{J^{2}} \right]} \cdot k_{p}$$

Using the Wronskiaus

$$\beta_p \circ f \in_{\Sigma} = \frac{2}{\pi p} \frac{1}{J_n^2(k_p f) + Y_n^2(k_p f)}$$

$$\beta_{f}$$
 of $H_{p} = \frac{Y_{n}'' J_{n}' - J_{n}'' Y_{n}'}{[J_{n}'^{2} + Y_{n}'^{2}]} \cdot k_{f}$

5-13 epien ν(f)=- α Ez; I(1)=2π f H p

:
$$j\omega\mu H_{\phi} = \frac{\partial E_{z}}{\partial \rho} = -\frac{1}{\alpha} \frac{\partial V}{\partial \rho}$$

$$\frac{j\omega\mu}{2\pi\rho} I(\rho) = -\frac{1}{a} \frac{\partial V}{\partial \rho} \frac{\partial V}{\partial \rho} = -\frac{j\omega\mu\alpha}{2\pi\rho} I(\rho)$$

$$\stackrel{\triangle}{=} -j\omega L I(\rho)$$

 $j\omega \in E_3 = \frac{1}{p} \frac{\partial}{\partial p} \left[f H_{\phi} \right] = \frac{1}{2\pi p} \frac{\partial I}{\partial p}$

$$-j\omega \in \frac{V}{\alpha} = \frac{1}{2\pi \rho} \frac{\partial \Gamma}{\partial \rho}$$

$$\therefore L = \frac{\mu a}{2\pi r} \quad 2 \quad C = \frac{2\pi r \epsilon}{a}$$

$$\frac{5-12}{2} \text{ We know,}$$

$$\frac{H_n^{(1)}(x)}{H_n^{(1)}(x)} = -\frac{1}{3}$$

$$\frac{2+}{x\to\infty} \frac{H_n^{(2)}(x)}{H_n^{'(2)}(x)} = \frac{1}{4}$$

$$Z + \frac{1}{k_p p} = -\frac{E_z}{H_p} = \frac{k_p}{j\omega \epsilon} \frac{H_n^{(s)}(k_p p)}{H_n^{(s)}(k_p p)}$$

Similarly
$$Z_p^{TM} = \gamma$$

$$Z_{+p}^{TM} = \frac{\eta}{j} \frac{H_0^{(2)}(k_p p)}{H_0^{(2)}(k_p p)} fr n = 0$$

$$H_0^{(2)}(k_p p) = 1 - \frac{2j}{\pi} \log \frac{7k_p p}{2}$$

$$\therefore \stackrel{\text{TH}}{=} \frac{1}{i} \frac{\pi \, k_{p} p}{(-2j)} \left[1 + \frac{j^{2}}{\pi} \log \frac{2}{\tau \, k_{p} f} \right]$$

$$= \frac{\eta k \rho P \left[\pi + 2j \log \frac{2}{\tau k \rho P} \right]}{2}$$

[note a factor of 2 is missing]

$$H_{y}^{(2)}(s) = \frac{1}{y!} \left(\frac{x}{2} \right)^{3} + j \frac{(y-1)!}{\pi} \left(\frac{x}{2} \right)^{3}$$

$$H_{\nu}^{1(2)}(x) = \frac{1}{\nu!} \frac{\nu x^{\nu-1}}{2^{\nu}} + j \frac{(\nu-1)!}{\pi} \frac{2^{\nu}(-\nu)}{x^{\nu+1}}$$

$$\therefore \frac{H_{n}^{(2)}(x)}{H_{n}^{(2)}(x)} = \frac{x}{n} \frac{\left(\frac{x}{2}\right)^{2n} \frac{\pi}{(n!)^{2}} + \frac{j}{n}}{\left(\frac{x}{n!}\right)^{2n} \frac{\pi}{(n!)^{2}} - \frac{j}{n}}$$

$$\therefore \mathcal{Z}_{+\rho}^{TM} = \frac{\eta_{k\rho} P}{j n} \left[\frac{\left(\frac{k_{\rho} P}{2}\right)^{2n} \frac{1}{(n!)^{2}} + \frac{j}{n!}}{\left(\frac{k_{\rho} P}{2}\right)^{2n} \frac{1}{(n!)^{2}} - \frac{j}{n}} \right]$$

$$\frac{f_{\pi}}{f} = \frac{1}{2\pi\alpha\sqrt{\mu\epsilon}} \times np$$

$$\frac{f_{\pi}}{f} = \frac{1}{2\pi\alpha\sqrt{\mu\epsilon}}$$

$$\frac{5-17}{(f_r)^{TM_{010}}} = \frac{1}{2\pi a \sqrt{\xi \mu}} \sqrt{\chi_{np}^{2} + 2^{2}\pi^{2}}$$

$$= \frac{10^{10}}{2\pi} \sqrt{(2.405)^{2}} = 3.94 \text{ GHz}$$

The other 9 resonant frequencies are obtained by multiplying for of the dominant mode by 1.5, 1.59, 1.63, 1.80, 2.05, 2.05, 2.13, 2.29 and 2.73.

$$Q_{010}^{TM} = \frac{\eta x_{01}}{4Q} = 1.37 \times 10^4$$

5-19 equien ε,=4ε, and μ,=μ, α=λ.

The cut off frequencies are given by (4.63)

$$f_{c} = \frac{\pi}{2\lambda_{0}\sqrt{\epsilon_{d}}\mu_{d}-\epsilon_{0}M_{0}} = \frac{\pi c}{2\lambda_{0}\sqrt{3}}$$

$$= \frac{\pi f_{0}}{2\sqrt{3}} \text{ where } f_{0} = \frac{\zeta}{\lambda_{0}} = \frac{\zeta}{a}$$

for m=0,1,2, ...

Hence ter TEO and TMO modes propagate unattenuated, no matter how tein/thick the slab is.

For hig. 5-9c , the characteristic

equation 4 (from 4.64)

$$f_{c} = \frac{\pi}{4 + \sqrt{\xi_{1} \mu_{d} - \xi_{0} \mu_{0}}} = \frac{\pi f_{0}}{2\sqrt{3}}$$

where for = 4a.

for TM modes m = 0, 2,4, ...

TE modes n = 1, 3, 5, .

The dominant mode is then the TMo mode which propagates unattemated at all frequencies.

So B = WILC

5-18 we know kp + k2 = ω μ,ε = k,2 kp+ k2 = ωμ.ε. = k2 and $\frac{k_3}{E_1}$ for $k_3 d = -\frac{k_{32}}{E_2}$ for $\left[k_3 (a-d)\right]$ $f(k_3,d) = \frac{k_{31}}{E_1} + ank_3d + \frac{k_{32}}{E_2} + an[k_3(a-d)]$ $\approx \frac{k_3 d}{\epsilon} + \frac{k_3 (a-d)}{\epsilon}$ 123 = w 11, E, - &p2 ε; = ω, ε, - kp $f(\beta,0) = \frac{k_{3_1}a}{\epsilon_1} = \frac{(\omega \mu_2 \epsilon_2 - \beta^2)a}{\epsilon_2}$ $f_{d}'(\beta,0) = \frac{k_{3}}{\xi_{1}} - \frac{k_{3}}{\xi_{2}}$ also f(kz,d) = f(p,0) + dfd(p,0) $O = \left(\frac{k_2^2 - \beta^2}{\varepsilon_1}\right) a + d \left(\frac{k_1^2 - \beta^2}{\varepsilon_1} - \frac{k_2^2 - \beta^2}{\varepsilon_2}\right)$ $\frac{\left(\beta^{2}-\frac{1}{2}\right)\alpha}{\overline{\epsilon_{2}}} = \frac{\lambda}{\alpha\overline{\epsilon_{1}}}\left(\omega^{2}\mu_{1}\overline{\epsilon_{2}} - \beta^{2}\frac{\overline{\epsilon_{1}}}{\overline{\epsilon_{1}}} - \omega^{2}\mu_{2}\overline{\epsilon_{1}} + \beta^{2}\right)$ $\therefore \hat{\beta} = k_2^2 \left[1 + \left(\frac{\mu_1}{\mu_2} - 1 \right) \frac{1}{\alpha} \right]$ [1+(=-1)4] $\beta \approx k_2 \sqrt{\frac{1+(\mu_1/\mu_2-1)d/a}{1+(\epsilon_1/\epsilon_1-1)d/a}}$ $LC = \frac{(\mu_1 - \mu_2)d + \mu_2 \alpha}{\xi_1 \alpha + d(\xi_1 - \xi_1)}$ 2 h2 = w Ju282

$$+ \sum_{k=1}^{2} ke_{2} F_{3} \begin{vmatrix} 0 & \mu_{2} ke_{1} F_{2} & \mu_{1} ke_{2} F_{4} \\ ke_{1} F_{1}' & \frac{k_{2} F_{2}}{\omega \mu_{1} a} & \frac{k_{2} F_{4}}{\omega \mu_{2} a} \end{vmatrix} = 0$$

$$\frac{ke_{1} F_{1}}{\omega \epsilon_{1} a} ke_{1} F_{2}' ke_{2} F_{4}'$$

$$(F_1 = F_2)$$
 $k_1^2 = \omega^2 \mu_1 \xi = k_2^2 + k_{e_1}^2$
 $k_2^2 = \omega^2 \mu_2 \xi_2 = k_2^2 + k_{e_2}^2$

$$F_1 = F_2 = J_1(k_0 a)$$
 $F_3 = J_1(k_0 a)N_1(k_0 b) - N_1(k_0 a)J_1(k_0 b)$
 $F_3' = J_1'(k_0 a)N_1(k_0 b) - N_1'(k_0 a)J_1(k_0 b)$
 $F_4 = J_1(k_0 a)N_1'(k_0 b) - N_1(k_0 a)J_1'(k_0 b)$
 $F_4' = J_1'(k_0 a)N_1'(k_0 b) - N_1'(k_0 a)J_1'(k_0 b)$

Chanael. Egm. becomes:

Now charact. Egn. becomes:

{ HI (kezb) [Eikez Ji'(keia) Ji(keza) - Ezkej Ji(keia) Ji'(keza)] + J, (kezb) [Ezke, J, (ke, a) N, (keza) - E, kez J, (ke, a) N, (keza)] } x

Charack Egn. is:

$$\left[AN_{1}(ke_{2}b) + BJ_{1}(ke_{2}b)\right]\left[CN_{1}'(ke_{2}b) + DJ_{1}'(ke_{2}b)\right] \\
-k_{2}^{2}(k_{2}^{2}-k_{2}^{2})^{2}J_{1}^{2}(ke_{1}c)\left[J_{1}(ke_{2}a)N_{1}(ke_{2}b) - N_{1}(ke_{2}c)J_{1}(ke_{2}b)\right] \\
-k_{1}^{2}k_{2}^{2}\omega^{2}a^{2} \\
-k_{1}^{2}k_{2}^{2}\omega^{2}a^{2}$$

$$\left[J_{1}(ke_{2}a)N_{1}(ke_{2}b) - N_{1}(ke_{2}a)J_{1}(ke_{2}b)\right] = 0$$

[5-22] using small argument approximation for Ko(2) = ln 2/72

From [5.21] - (problem nos)

 $u^2a^2 = \frac{\epsilon_1 + \epsilon_2}{\epsilon_2 k_0(va)} = \frac{\epsilon_1 + \epsilon_2}{\epsilon_2 \ln \frac{2}{\gamma va}}$

Ter E1 = 9E2 , a = 0.12,

 $\frac{K_1(VP)}{K_1(Va)} \approx 10$ when VP = 0.9

 $\frac{15-24}{5-24} \text{ for the dominant mode}$ $V_{opq}^{TM} = AJ_{opq}(k_{1}, p) \cos(k_{2}, 2) \text{ in reduin(1)}$

40 pg = 85. (k2p) cos [k2 (d-2)] in madin 2

Hence Construity of Ex at 2=6 implies

- ke, Ak, J. (8,7) sin (83,6)

= kez B & 70/(8,7) sin[83(d-b)]-(1)

Similarly continuity of Ho at 3= b

implies

Ak, J. (8, p) co(k3, b) = Bk2 J. (62) co [k3 (d-b)] - (1) New (k-k0) a J. (ka) = J. (ka)

Hence dividing 1 by 1

- = ten [k3 b] = k32 ten [k3 (d-b)]

tor small d (See problem 5-18)

it is very similar to that

5.23 The fields are completely specified by the magnetic vector potentials

 $(Y_{010}^{TH})^{air} = \left[J_{o}(k_{0}t) - \frac{J_{o}(k_{0}a) Y_{o}(k_{0}t)}{Y_{o}(k_{0}a)}\right]$ $(65\frac{\pi^{2}}{d})$ $(65\frac{\pi^{2}}{d})$

 $\left(Y_{010}^{TM}\right)^{\text{dislectric}} = J_0\left(kp\right) \cos\frac{\pi^2}{4}$

Hence from the continuity of Ez,

 $\frac{k_{p_1}^2}{3} \psi^{air} = \frac{k_{p_2}^2}{3} \psi^{d}$

and from the constinuity of Hps k_1, ψ' air = k_1, ψ' d

Hence $\frac{\psi'd}{\psi^d} \frac{\dot{y}_2}{k_p} = \frac{\dot{y}_1}{k_p} \frac{\psi'air}{\psi air}$

To(kc) = 1 Jo(kc) / (ka)-Jo(ka) / (kc)

Jo(kc) - 7. (ka) / (ka) - Jo(ka) / (kc)

Jo(kc) / Jo(ka) - Jo(ka) / (ka)

J.'(kc)[J.(kc)4.(ka) - J.(ka)4.(kc)

V μοξο Jo(k, c) [Jo(kc) Yo (ka) -Jo (ka)

where & = DOV MOE,

R = WHOER

Ro = WOV HOEO

and using the small argument, approximations for the Bestel functions one obtains

Yo(k, a) π c²(k²-k²) = (k-ko) a Jo(ka

Livere all see functions have been expanded which has a C in the argument]

: 4-40 = # 1/(4a) c2 x01(Ex-1)

5.25 For the circular country with a conducting wedge the general solution is of the form

$$J_n(k_p p) \begin{cases} \sin n \beta \\ \cos n \beta \end{cases} e^{\pm jk_3 3}$$

for
$$E_p = 0$$
,

A simp + 8 comp = 0 for $\begin{cases} \phi = 0 \\ \phi = 2\pi - \phi \end{cases}$

$$2n\pi - n\phi = k\pi$$

$$\therefore n = \frac{k\pi}{2\pi - \phi}$$

and for TM modes $J_n(k_p a) = 0$, where $m = \frac{\pi}{2\pi - \alpha}$ for the dominant k=1 mode

$$F_{3} = F_{3}'(x, y - \frac{5}{2}) - F_{3}'(x, y + \frac{5}{2})$$

$$= -\frac{5}{2}\frac{F_{3}'}{2y} = -\frac{K5}{4j}\frac{3}{2y}H_{0}^{(2)}(kp)$$

$$= \frac{KKS}{4j}H_{1}^{(2)}(kp)\frac{3p}{2} = \frac{KKS}{4j}H_{1}^{(2)}(kp)\frac{5inp}{2}$$
and $\vec{E} = -\vec{\nabla} \times \vec{3}\vec{F}_{3}$

5.26 Let
$$A_x = CH_0^{(2)}(kp)$$
where C is a constand.

and by the terms of the problem H_3 . $2\Delta z = I$ and

$$H_3 = -\frac{\partial A_x}{\partial y} = -\frac{\partial Q}{\partial y} \left[1 - \frac{2j}{\pi} \log \frac{fx}{2}\right]$$

$$= \frac{2jC}{x\pi}$$

$$\therefore C = \frac{T}{4j} = \frac{J\ell}{4j}$$

$$F_{2} = F_{3}'(x, y + \frac{5}{2})$$

$$- F_{3}'(x, y - \frac{5}{2}) \xrightarrow{-1} K$$

They are identical when

$$I = \frac{k^2}{j\omega\mu} kS = -j\omega E KS$$

$$A_{3}^{2} = S_{2} \frac{\partial A_{3}^{1}}{\partial x}$$

$$= -\frac{k S_{1} S_{2} I \left[k H_{1}^{(2)}(k f) \omega_{1}^{2} \beta + H_{1}^{(2)}(k f) \frac{y^{2}}{x^{2} + y^{2}}\right]}{\beta}$$

$$\frac{2}{kdp} \left[H_{1}^{(2)}(kp) = H_{0}^{(2)}(kp) - H_{2}^{(2)}(kp) \right]$$

$$k H_{1}^{(2)}(kp) = \frac{k}{2} \left[H_{0}^{(2)}(kp) - H_{2}^{(2)}(kp) \right] \text{ and }$$

$$H_{1}^{(2)}(kp) = \frac{kp}{2} \left[H_{0}^{(2)}(kp) + H_{2}^{(2)}(kp) \right]$$

$$\therefore A_{3}^{2} = \underbrace{k S_{1}S_{2}I}_{G_{1}^{2}} \left[-H_{0}^{(2)}(kp) + H_{2}^{(2)}(kp) + H_{2}^{(2)}(kp) \right]$$

current filaments are represented as

$$J_{3} = \frac{I}{2\pi\alpha} \sum_{p=1}^{2m} (-1)^{p+1} \delta \left[\phi - \frac{\pi}{2n} - (p-1) \frac{\pi}{n} \right]$$

$$= \frac{I}{2\pi\alpha} \left[\delta \left(\frac{\pi}{2n} \right) - \delta \left(\frac{3\pi}{2n} \right) + \delta \left(\frac{5\pi}{2n} \right) + \cdots \right]$$

Because of symmetry, the 4 potentials can be represented as $\psi^- = c_1 J_n(\ell_p) sin p$ (Inside region)

4+ = C2 Hn (kp) sin np (external to the current elements)

· From the continuity of Ez (from 5.18) at p=a

C, Jn (ka) = C2 Hn (ka), and from the discontinuity of

 $H_{\rho}^{+} - H_{\rho} = J_{\chi}$

-C, k Jn' (ka) sin p + C2 k Hn (ka) sui np = Jz

:.
$$J_3 = k.C_2. sim \neq \left[H_n^{(2)}(ka) - \frac{H_n^{(2)}(ka)}{J_n(ka)} \right]$$

 $J_3 = -\frac{2jk}{\pi ka} \sin n\phi \frac{C_2}{J_n(ka)}$

Multiplying each side by sin no and integrating from

 $2\pi \sum_{n=1}^{2\pi} (-1)^{n+1} \delta \left[\phi - \frac{\pi}{2n} - (p-1)\frac{\pi}{n} \right] \sin n \phi d \phi$ $2\pi \sum_{n=1}^{2\pi} (-1)^{n+1} \delta \left[\phi - \frac{\pi}{2n} - (p-1)\frac{\pi}{n} \right] \sin n \phi d \phi$

= $-\frac{2j}{\pi a} \cdot \frac{c_2}{J_n(ka)}$. π . Expanding $J_n(ka)$ for small

 $C_{2} = -\frac{nI}{2j\pi} \frac{1}{n!} \left(\frac{ka}{2}\right)^{n} \left[as \ a \rightarrow 0\right]$ argument approximations.

: $Y_2 = -\frac{I}{2j\pi(n-1)!} \left(\frac{ka}{2}\right)^n H_n^{(2)}(kr) \sin n\rho$

$$Y = \begin{cases} \frac{\pi a}{2j} \sum_{n=-\infty}^{\infty} A_n J_n(ka) H_n(kp) e^{jn\phi}, & p > a \\ \frac{\pi a}{2j} \sum_{n=-\infty}^{\infty} A_n H_n(ka) J_n(kp) e^{jn\phi}, & p < a \end{cases}$$

Since $H_{\phi}^{+} - H_{\phi}^{-} = J$ and $H_{\phi} = -\frac{\partial A_{3}}{\partial \rho}$

$$H_{\beta}^{+} = -\frac{\pi a}{2j} \sum_{n=-\infty}^{\infty} A_{n} J_{n}(ka) k H_{n}^{(2)}(kp) e^{jn\beta}$$

$$T_{2} = H_{p}^{+} \Big|_{f=a} - H_{p}^{-} \Big|_{f=a}$$

$$= \frac{\pi a k}{2 j} \sum_{n=-\infty}^{\infty} A_{n} e^{jnp} \Big[J_{n}^{\prime}(ka) H_{n}^{(2)}(ka) - J_{n}(ka) H_{n}^{\prime(2)}(ka) \Big]$$

$$J_{n}(ka) H_{n}^{\prime(2)}(ka)$$

Using the Wronskians

$$J_{2} = \frac{\pi a k}{2j} \cdot \frac{2j}{\pi k a} \sum_{n=-\infty}^{\infty} A_{n} e^{jn\phi} = \sum_{n=-\infty}^{\infty} A_{n} e^{jn\phi}$$

$$A_n = \frac{1}{2\pi} \int_{0}^{2\pi} J_3 e^{-jn\phi} d\phi$$

5-33 Since e-19000 = = 5 1 Jn (7)e 176

$$cos(\varphi cos \phi) - j sui (\varphi cos \phi)$$

= $5.(\varphi) - J_3(\varphi) \left[e^{j2\phi} - j2\phi \right]$ $i\phi - i\phi$

Let 0 - 1/2 - p and equating real and imaginary parts

5.31 The vector potential due to a ribbon of magnetical current is quien by (5-98)

$$A_3 = \frac{e}{\sqrt{8j\pi kR}} \int_{-\frac{\alpha}{2}}^{\sqrt{2}} \int_{-\frac{\alpha}{2}}^{\sqrt{2}} e^{jk x' \cos \theta} dx'$$

$$= \frac{e^{-jk\pi}}{\sqrt{8j\pi k\pi}} \int_{Z} a \frac{\sin\left(\frac{ka}{2}\cos\beta\right)}{\left(\frac{ka}{2}\cos\beta\right)}$$

Since Ez = - jum Az

$$\frac{1}{100} = \frac{-j \mu_{n} e^{-j k \pi}}{\sqrt{8\pi j k \pi}} \frac{\sin(\frac{k\alpha}{2} \cos \phi)}{(\frac{k\alpha}{2} \cos \phi)}$$

and the = - Ez in ten for zone.

5.32 Assuming that Langential E in the shot is $\hat{\mathbf{U}}_{\mathbf{x}}$ Eo, a constant,

 $\overline{M}_{S} = \overline{E}^{i} \times \overline{n} = (\widehat{x} E_{o}) \times \widehat{y} = \widehat{g} E_{o}$ The field in produced by $2\overline{M}_{S}$ and the electric vector potential is given by (5-98)

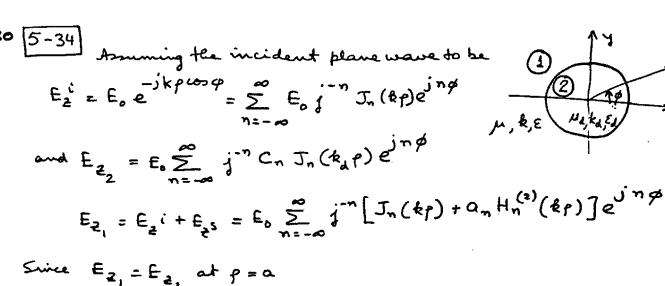
$$F_{2} = \frac{e^{-jk\rho}}{\sqrt{8j\pi k\rho}} = \int_{-9/2}^{-9/2} jkx \cos \rho \, dx'$$

$$= \frac{e^{-jkp}}{\sqrt{8j\pi kp}} 2E_0 a \frac{\sin(\frac{ka}{2}\cos\phi)}{(\frac{ka}{2}\cos\phi)}$$

Since $H_3 = -j\omega \epsilon F_3$

$$H_{3} = -jωε α e^{-jκρ} = ω (kα ωρ)$$

$$\sqrt{8jπkρ} = (kα ωρ)$$



Since
$$E_{Z_1} = E_{Z_2}$$
 at $p = a$

$$\therefore C_n = \frac{J_n(ka) + a_n H_n^{(2)}(ka)}{J_n(k_d a)}$$

Also
$$H_{\phi_{1}} = E \cdot \sum_{n=-\infty}^{\infty} j^{-n} \frac{p}{j\omega\mu} k [J_{n}'(kp) + a_{n} H_{n}'^{(2)}(kp)] e^{jn\phi}$$

and
$$H_{\phi_2} = E_0 \sum_{n=-\infty}^{\infty} j^{-n} c_n \frac{p}{jw\mu_1} k_1 J_n'(k_1 p) e^{jnp}$$

Since
$$H_{\phi_{i}} = H_{\phi_{2}} \text{ at } \rho = \alpha$$

$$\frac{k}{\mu} \left[J_{n}'(ka) + a_{n} H_{n}'^{(2)}(ka) \right] = \frac{J_{n}(ka) + a_{n} H_{n}'(ka)}{J_{n}(ka)} \frac{k_{d}}{\mu_{d}} J_{n}'(ka)$$

$$\frac{L}{\mu} \left[J_{n}'(ka) + a_{n} H_{n}'^{(2)}(ka) \right] = \frac{J_{n}(ka) + a_{n} H_{n}'(ka)}{J_{n}(ka)} \frac{k_{d}}{\mu_{d}} J_{n}'(ka)$$

$$\frac{J_{n}(ka)}{H_{n}^{(2)}(ka)} = \frac{J_{n}'(ka)}{\frac{J_{n}'(ka)}{J_{n}(ka)}} - \frac{1}{kl} \frac{J_{n}'(ka)}{J_{n}(ka)}$$

$$\frac{M}{kM_{2}} \frac{J_{n}'(ka)}{J_{n}(ka)} - \frac{1}{kl} \frac{H_{n}^{(2)}(ka)}{H_{n}^{(2)}(ka)}$$

$$a_{n} = -\frac{J_{n}(ka)}{H_{n}^{(2)}(ka)} - \frac{\frac{EL}{Ek_{d}a} \frac{J_{n}'(k_{d}a)}{J_{n}(k_{d}a)} - \frac{1}{ka} \frac{J_{n}(ka)}{J_{n}(ka)}}{\frac{EL}{Ek_{d}a} \frac{J_{n}'(k_{d}a)}{J_{n}(k_{d}a)} - \frac{1}{ka} \frac{H_{n}'^{(2)}(ka)}{H_{n}^{(2)}(ka)}}$$

The fields internal to the cylinder is grienby E_{Z_2} . As $E_1 \rightarrow 0$, $E_{Z_2} \rightarrow 0$ and $\Omega_n \rightarrow -\frac{J_n(k\alpha)}{H_n^{(2)}/(k\alpha)}$. Hence

this reduces to the solution of a conducting cylinder.

5-35 Since the problem is duel to the problem of 5-34 the answer is signifian

$$a_0 = -\frac{1}{H_0^{(2)}(k\alpha)} \left[\frac{-\frac{\epsilon_k k_k \alpha}{2\epsilon_k k_k \alpha} + \frac{k\alpha}{2k\alpha}}{-\frac{\epsilon_k k_k \alpha}{2\epsilon_k} + (\frac{2}{k\alpha})^2} \right]$$

$$= \frac{\pi_j^2}{2} \left[1 - \frac{\xi A}{\xi} \right] (ka)^2$$

$$E_{2}^{S} = -\frac{i\pi}{4} E_{0} (\xi_{n} - i) H_{0}^{(2)} (49) (8a)^{2}$$

$$5-38$$
 H_2 as quien by (5.134)

$$= H_0 \sum_{n=0}^{\infty} \sum_{n} j^{n} J_{n}(kp) \cos \frac{n\pi}{2} \cos \frac{n\pi}{2}$$

= 2 Ho \sum_{n=0}^{\infty} \(\x_1 \) \(\x_2 \) \(\x_1 \) \(\x_2 \) \(\x_1 \) \(\x_2 \)

In this case
$$jkp\cos(\phi-\phi')$$

$$E_{z}^{i} = E_{0}e$$

$$= 2E_{0}\sum_{n=1}^{\infty} j^{n/2} J_{n/2}(k_{1}) \sin \frac{n\phi}{2} \sin \frac{n\phi}{2}$$

$$= 2E_{0}\sum_{n=1}^{\infty} j^{n/2} J_{n/2}(k_{1}) \sin \frac{n\phi}{2} \sin \frac{n\phi}{2}$$

$$from (5.124)$$

Hoom p. 233

$$H\phi = \frac{1}{\sqrt{\omega_i}} \frac{\partial E_3}{\partial \rho} \quad \text{and} \quad H\rho = \frac{1}{\sqrt{\omega_i}} \frac{\partial E_3}{\partial \rho}$$

$$H_{p} = \frac{2E_{o}}{-j\omega\mu\rho} \sum_{n=1}^{\infty} i^{n/2} J_{\gamma_{2}}(R\rho) n \sin \frac{n\rho}{2}$$

$$E_2^{i} = 2E_0 \sqrt{\frac{2kpi}{\pi}} \sin \frac{\phi}{2} \sin \frac{\phi}{2}$$

[This is norgers for
$$m=1$$
 when $kp \rightarrow 0$]

32
$$5-39$$
 As explained in $5-40$ $F^i = dual of Ψ_i and ho $F^i = F^i = F^i = \frac{Kl e^{-jk\pi}}{4\pi\pi} \frac{jk3^i \cos\theta}{2\pi} \sum_{n=0}^{\infty} \epsilon_n(j)^n \left[J_n(k_f^i \sin\theta) + b_n H_n^{(2)}(k_f^i \sin\theta) \right] \cos n$
By the terms of the problem $E_p = 0 |_{P=a} = \frac{3F_3}{3P} |_{F=a} = 0$$

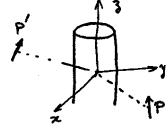
Hence
$$b_n = -\frac{J_n'(kasino)}{H_n'^{(2)}(kasino)}$$
. For far field,

$$\left[J_{n}(kasin0) - \frac{J_{n}'(kasin0) H_{n}^{(2)}(kasin0)}{H_{n}'^{(2)}(kasin0)}\right]$$

$$E_{\phi} = -\frac{jkR}{4\pi R} \sin \theta \sum_{n=0}^{\infty} E_{n} j^{n} \cos n\phi \frac{-2j}{\pi ka \sin \theta} \cdot \frac{1}{H_{n}^{1(2)}(ka \sin \theta)}$$

$$= -\frac{\text{Kle}}{2\pi^{2}\alpha} \sum_{n=0}^{\infty} \operatorname{En}(j)^{n} \frac{\cos^{n}\beta}{H_{n}^{(u)}(\text{Rasino})}$$

5-40 The radiation pattern of a dipole at P near a conducting cylinder is obtained as a receiving antenna instead of as a transmitting antenna. The wave received from a distance source will be essentially a plane wave that



is easily expanded into a sum of standing cylindrical waves. Equating the sum of the incident and the scattered tangential electric fields to zero at the cylinder surface gives the magnitude of the secondary or re-radiated warrs.

So, the incident field in its jkg' cosp sino
$$y' = \frac{Tleikn}{4\pi n}e$$

=
$$\frac{\text{Il } e^{-jkn}}{4\pi n}$$
 $e^{jk3^2\cos\theta}$ $\sum_{n=0}^{\infty} \varepsilon_n(j)^n J_n(kj \sin\theta) \cos n\phi$

5-40 contd.

$$\psi s = \frac{T\ell}{4\pi\pi} e^{-jk\pi} e^{jkg'\cos\theta} \sum_{n=0}^{\infty} b_n \epsilon_n (j)^n H_n^{(2)}(k\rho' \sin\theta) \cos n\phi'$$

Since $E_3 = 0$ at $f = a$, $b_n = \frac{J_n(Ra \sin\theta)}{H_n^{(2)}(Ra \sin\theta)}$
 $\vdots E_0 = \left[E_p \cos\theta' - E_g \sin\theta'\right] p' = b, \theta' = 90', 3' = 0$
 $= -E_2 | p' = b = -j\omega \mu \sin\theta A_3 | p' = b \quad (in the far fields)$
 $\vdots E_0 = j\omega \mu \frac{J_1 e^{-jk\pi}}{4\pi\pi} \sin\theta \sum_{n=0}^{\infty} \left[J_n(kb \sin\theta) - \frac{J_n(Ra \sin\theta)}{H_n^{(2)}(Ra \sin\theta)} H_n^{(2)}(Rb \sin\theta)\right]$
 $= \frac{1}{3}(\pi) \sin\theta \sum_{n=0}^{\infty} j^n e^{jn\phi} \left[J_n(kb \sin\theta) - \frac{J_n(Ra \sin\theta)}{H_n^{(2)}(Ra \sin\theta)} H_n^{(2)}(Rb \sin\theta)\right]$
 $= \frac{1}{3}(\pi) \sin\theta \sum_{n=0}^{\infty} j^n e^{jn\phi} \left[\frac{J_n(Ra \sin\theta) N_n(Rb \sin\theta) - J_n(Rb \sin\theta)}{H_n^{(2)}(Ra \sin\theta)} H_n^{(2)}(Ra \sin\theta)\right]$

Since $J_{-n}(3) = (-1)^n J_n(3)$; $N_{-n}(3) = (-1)^n N_n(3)$ and $H_{-n}^{(3)}(3) = e^{-in\pi} (3)$; hence

 $E_0 = A \cdot f(x) \sin\theta \sum_{n=0}^{\infty} j^n e^{jn\phi} \left[\frac{J_n(x) N_n(\beta) - J_n(\beta) N_n(\alpha)}{H_n^{(2)}(\alpha)} \right]$
 $= \frac{J_n(3) \sin\theta}{J_n(3)} = e^{-jn\phi} \left[\frac{J_n(x) N_n(\beta) - J_n(\beta) N_n(\alpha)}{H_n^{(2)}(\alpha)} \right]$
 $= \frac{J_n(3) \sin\theta}{J_n(3)} = \frac{J_n(3) \sin\theta}{J_n(3)} = \frac{J_n(3) N_n(\alpha)}{J_n(3)} = \frac{J_n(3) \sin\theta}{J_n(3)} = \frac{J_n(3) N_n(\alpha)}{J_n(3)} = \frac{J_n(3) \sin\theta}{J_n(3)} = \frac{J_n(3) N_n(\alpha)}{J_n(3)} = \frac{J_n(3) N_n(\alpha)$

For this configuration.

$$\begin{aligned}
Y_{x}' &= \frac{I \cdot e^{-jkx}}{4\pi x} e^{jkp'} \cos \varphi \sin \theta \\
&= \frac{I \cdot e^{-jkx}}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \int_{n} (kp' \sin \theta) \cos n\varphi \\
&= \frac{I \cdot e^{-jkx}}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{I \cdot e^{-jkx}}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{I \cdot e^{-jkx}}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{I \cdot e^{-jkx}}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{4\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{2\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp' \sin \theta) \right] \\
&= \frac{1}{2\pi x} \sum_{n=0}^{\infty} \epsilon_{n}(j)^{n} \left[\int_{n} (kp' \sin \theta) + b_{n} H_{n}^{(2)} (kp$$

$$\frac{5-41}{4\pi \pi} contd. \quad \text{In the case, } \theta' = 90^{\circ}, \quad \theta = 90^{\circ}$$

$$\psi_{X} = \frac{\text{Il } e^{-jKx}}{4\pi \pi} \sum_{n=0}^{\infty} E_{n}(j)^{n} \left[J_{n}(kf') - \frac{J_{n}'(ka)}{H_{n}^{(2)}(ka)} + H_{n}^{(2)}(kf') \right] cos \pi \phi$$

$$\vdots E_{\phi} = E_{y} cos \phi' - E_{x} cos \phi' \int_{f'=0}^{f} = E_{y} = \frac{1}{9} \frac{d^{4}k}{f \cos \theta' d \phi}$$

$$= -\frac{Il e^{-jKx}}{4\pi \pi} \sum_{n=0}^{\infty} E_{n}(j)^{n} \frac{m \sin n\phi}{H_{n}^{(2)}(ka)} \cdot \frac{2}{\pi ka}$$

$$= f(f) \sum_{n=1}^{\infty} \frac{n(j)^{n} \sin n\phi}{H_{n}^{(2)}(ka)}$$

The field from a magnetic dipole at angle 0 Here is $f(x) = \frac{K e^{-jkx}}{4\pi x^2} e^{-jkx} e^{-jkx}$ F = Kle-jkr ejk8'cos0 = Enj^1/2 Jn/2 (kpsino) cos m# Ex = -jk F = ino (from 3.97). Ep = - ikkle sino = En j 1/2 Jn/2 (kasino) cos nos 2

$$\begin{array}{ll}
\cdot \overline{\nabla} \cdot \overline{A} = \frac{1}{\gamma^{2}} \frac{\partial}{\partial n} (n^{2} A_{\Lambda}) + \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (A_{\theta} \sin \theta) + \frac{1}{n \sin \theta} \frac{\partial A_{\theta}}{\partial \theta} \\
&= \frac{\cos \theta}{n^{2}} \frac{\partial}{\partial n} (n^{2} \psi) - \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin \theta) \\
\overline{\nabla} (\overline{\nabla} \cdot \overline{A}) = \overline{u}_{V} \frac{\partial}{\partial n} (\overline{\nabla} \cdot \overline{A}) + \overline{u}_{\theta} \frac{1}{n \frac{\partial}{\partial \theta}} (\overline{\nabla} \cdot \overline{A}) + \overline{u}_{\theta} \frac{1}{n \frac{\partial}{\partial \theta}} (\overline{\nabla} \cdot \overline{A}) + \overline{u}_{\theta} \frac{\partial}{\partial \theta} (\overline{\nabla} \cdot \overline{A}) \\
&= \overline{u}_{V} \frac{\partial}{\partial n} \left[\frac{\cos \theta}{n^{2}} \frac{\partial}{\partial n} (n^{2} \psi) - \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin \theta) \right] \\
&+ \overline{u}_{\theta} \frac{1}{n \frac{\partial}{\partial \theta}} \left[\frac{\cos \theta}{n^{2}} \frac{\partial}{\partial n} (n^{2} \psi) - \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin \theta) \right] + \overline{u}_{\theta} \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin \theta) \\
&= \frac{\cos \theta}{n^{2}} \frac{\partial}{\partial n} (n^{2} \psi) - \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin \theta) \\
&= \frac{\cos \theta}{n^{2}} \frac{\partial}{\partial n} (n^{2} \psi) - \frac{1}{n \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin \theta) \right]$$

$$E_0 = j\omega_{\mu} + \sin \theta + \frac{1}{j\omega_{ER}} \frac{3}{3\theta} \left[\frac{\cos \theta}{R^2} \frac{3}{3R} (R^2 \psi) - \frac{1}{n \sin \theta} \frac{3}{3\theta} (\Psi \sin \theta) \right]$$

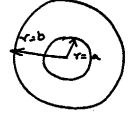
$$f_{\phi} = \frac{1}{\text{jwersing}} \frac{\partial}{\partial p} \left[\frac{\cos \theta}{\eta^2} \frac{\partial}{\partial x} (x^{\nu} \psi) - \frac{1}{x \sin \theta} \frac{\partial}{\partial \theta} (\psi \sin^2 \theta) \right]$$

$$\frac{1}{2} \left[\frac{\partial^{2} A_{1}}{\partial x^{2}} - A_{1} \right] + \frac{1}{2} \left[\frac{\partial^{2} A_{1}}{\partial x^{2}} + \frac{\partial^{2} A_{1}}{\partial x^{2}} + \frac{\partial^{2} A_{1}}{\partial x^{2}} + \frac{\partial^{2} A_{1}}{\partial x^{2}} + \frac{\partial^{2} A_{1}}{\partial x^{2}} \right] = 0.$$

$$\frac{1}{2} \left[\frac{3}{2} \left[\frac{3}{2} \frac{A}{A} - A_{A} \right] + \frac{1}{2} \frac{3}{2} \left(\frac{3}{2} \frac{A}{A} \right) + \frac{1}{2} \frac{3}{2} \frac{3}{2} \frac{A}{A} + \frac{3}{2} \frac{A}{A} = 0$$

36 [6-3] r= 5 cm. The dominant mode is the TMOII mode. The frist ten resonant frequencies are $(f_n)_{mnp}^{TM} = \frac{u'np}{2\pi a \sqrt{\epsilon_{\mu}}} = \frac{3 \times 10^9}{77} u'np$ in GHz, 2.62; 3.7; 4.3; 4.75; 5.5; 5.8 TM_{m,1,1} TM_{m,2,1} TE_{m,1,1} TM_{m,3,1} TE_{m,2,1} TM_{m,4,1} 5.85; 6.67; 6.83; 7.1 TMm,1,2 TEm,3,1 TMm,5,1 TMm,4,2 $Q = \frac{1.01}{R} = \frac{1.01 \times 120 \, \pi}{2.61 \times 10^{-7} \times \sqrt{2.62 \times 10^{-7}}} = 2.85 \times 10^{5}$ 6-4 for TM to se modes $F_R = \hat{J}_n(kx) P_n^m(\cos\theta) \cos m\phi$ $E_R = 0$ and $\hat{J}_n(ka) = 0$ $A_{R} = \frac{\hat{J}_{R}(k_{R}) P_{R}^{m}(\omega = 0)}{\omega + \omega}$ $E_0 = -\frac{1}{9\pi} k \int_{\pi}^{\pi} (4\pi) \sin \theta_n^{m'}(\cos \theta) \cos m\phi$ Et = m In(kx) Pm (coro) simp $E_{\phi} = -\frac{mk}{9\pi\sin\theta} \int_{n}^{\infty} (k\pi) P_{n}^{m}(\cos\theta) \sin m\phi$ $E_{\phi} = -\frac{1}{\pi} \hat{J}_{n}(kx) P_{n}^{m}(\omega \theta) \sin \theta \cos m \theta$ Ho = $-\frac{m}{\pi \sin \theta} \hat{J}_n(kz) P_n^m(\cos \theta) \sin m\phi$ $H_{\theta} = -\frac{k}{2\pi} \hat{J}_{n}'(kx) p_{n}^{m}'(con\theta) \sin \theta \cos m p$ $H_{\beta} = \frac{1}{\pi} \hat{J}_{n}(kx) P_{n}^{m'}(\cos 0) simp sin \theta$ $H\phi = -\frac{km}{2\pi \sin \theta} \hat{J}_{n}'(kr) P_{n}^{m}(\cos \theta) \sin m \theta$ 4n = 0 and $\frac{\hat{J}_n}{J_n}(ka) = 0$. $W = 2We = E \int d\varphi \int d\theta \int d\theta \int d\theta \int d\theta \left[E_{\theta}^{2} + E_{\varphi}^{2} \right]$ W = 2 Wm = m [dq [d0] 2 sin d dr [Ho + Ho] and $\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} (kx) dx = \frac{\pi}{2k} (kx)^2 \left\{ J_n^2(kx) - J_{n-1}(kx) J_{n+1}(kx) \right\}$ $W = \frac{E\pi}{k} \frac{2n(n+1)}{2n+1} \frac{(n+m)!}{(n-m)!} \frac{ka}{2} \left\{ I_n(ka) \right\}$ $.. W = \frac{\mu \pi}{R} \left[\frac{2\eta(n+1)}{2n+1} \frac{(n+m)!}{(n-m)!} \right] \left[\frac{k\alpha}{2} \right] \left[1 - \frac{\eta(n+1)}{(k\alpha)^2} \right]$ P = RT Jn/2 (kg) 2n(n+1) (n+m)! Pa = R dp [do a2 sino [Ho + Ho] n=a = $R \tilde{J}_{n}^{2}(4a) \frac{2n(n+1)}{2n+1} \frac{(n+m)!}{(n-m)!}$ Q = we ka = 1 unp $Q = \frac{\omega W}{P_{\perp}} = \frac{\omega \mu}{2 k R} \left(= \frac{\eta}{2 R} \right) \left[\frac{u'_{np}}{u'_{np}} - \frac{(n+1)^n}{u'_{np}} \right]$

where Unp = Ra



Since Eo = 0 of r=a & r=b, Hence

$$B = -\frac{\hat{J}_n(ka)}{\hat{N}_n(ka)} = -\frac{\hat{J}_n(kb)}{\hat{N}_n(kb)}$$
 & so the characteristic eqⁿ is

$$\frac{\hat{J}_n(ka)}{\hat{J}_n(kb)} = \frac{\hat{N}_n(ka)}{\hat{N}_n(kb)}$$

For TM modes

$$E_{\theta} = -\frac{k \sin \theta A}{j \omega \epsilon r} \left[\hat{J}_{n}'(kx) + B \hat{N}_{n}'(kx) \right] P_{n}^{m}'(\cos \theta) \cos m \phi$$

Again since Eo = 0 at r = a & r = b, hence

$$B = -\frac{\hat{J}_n'(ka)}{\hat{N}_n'(ka)} = -\frac{\hat{J}_n'(kb)}{\hat{N}_n'(kb)} & \text{ so the characteristic ag}^n is$$

$$\frac{\hat{J}_{n}'(ka)}{\hat{J}_{n}'(kb)} = \frac{\hat{N}_{n}'(ka)}{\hat{N}_{n}'(kb)}$$

$$\hat{\sigma}_{1}(x) \approx \frac{x^{3}}{3}$$

$$\frac{2}{3}$$

$$(k-\ell_0)bJ_1''(\ell_0b)\frac{1}{\ell_0^2a^2}=\frac{2}{3}k_0aN_1''(\ell_0b)$$

$$\frac{\omega - \omega_0}{\omega_0} = \frac{2}{3} (k.b)^2 \frac{\hat{N}'(k.b)}{\hat{J}''(k.b)} (\frac{a}{b})^3$$
 where $k.b = 2.744$

Here:
$$E_{\theta_{1}} = -\frac{C_{1}k_{1} \sin \theta}{j \omega e_{1} R} \frac{1}{J_{n}'(k_{1}x)} P_{n}^{m}(\cos \theta) \cos m\phi$$

$$E_{\theta_{2}} = -\frac{C_{2}k_{2} \sin \theta}{j \omega e_{1} R} \left[\frac{1}{J_{n}'(k_{1}x)} - \frac{1}{N_{n}'(k_{2}b)} \frac{1}{N_{n}'(k_{1}b)} \frac{1}{N_{n}'($$

6-9 given
$$f(\theta, \phi) = \begin{cases} 1 & 0 < \theta < \pi/2 \\ 0 & \pi/2 < \theta < \pi \end{cases}$$

A fourier Legendre series for a function $f(\theta, \phi)$ on a spherical surface is guien a

surface is quien as
$$f(\theta, \phi) = \sum_{n=0}^{\infty} \sum_{m=0}^{\infty} (a_{mn} \cos m\phi + b_{mn} \sin m\phi) P_n^m (\cos \theta)$$

$$f(\theta,\phi) = \sum_{n=0}^{\infty} \sum_{m=0}^{\infty} \sum_{m=0}^{\infty} \sum_{m=0}^{\infty} \int_{0}^{\infty} d\theta f(\theta,\phi) P_{n}(\omega n\theta)$$
where $Q_{0n} = \frac{2n+1}{4\pi} \int_{0}^{\pi/2} P_{n}(\omega n\theta) d\theta$

$$= \frac{2n+1}{2} \int_{0}^{\pi/2} P_{n}(\omega n\theta) d\theta$$

$$= \frac{2\eta + 1}{2} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) d\theta}{d\theta} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{d\theta}{d\theta} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{d\theta}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{d\theta}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) P_{n}(\omega s \theta) Cos m \phi}{(n+m)!} \int_{0}^{2\pi} \frac{P_{n}(\omega s \theta) P_{n}(\omega s \theta)$$

$$= \frac{2\pi}{2\pi} \frac{(n+m)!}{(n+m)!} \int_{0}^{2\pi} u dd \int_{0}^{\pi/2} \rho_{n}^{m}(un\theta) \sin \theta d\theta = 0$$

$$= \frac{2n+1}{2\pi} \frac{(n-m)!}{(n+m)!} \int_{0}^{2\pi} con m \phi d\phi \int_{0}^{\pi/2} \rho_{n}^{m}(un\theta) \sin \theta d\theta = 0$$

$$= \frac{2\pi \tau}{2\pi} \frac{(n+m)!}{(n+m)!} \int_{0}^{2\pi} d\theta \int_{0}^{\pi} d\theta \int_{0}^{\pi} (0,p) P_{n}^{m} (\cos \theta) \sin m\phi \sin \theta$$

$$= \frac{2n+1}{2\pi} \frac{(n-m)!}{(n+m)!} \int_{0}^{2\pi} d\theta \int_{0}^{\pi} d\theta \int_{0}^{\pi} (0,p) P_{n}^{m} (\cos \theta) \sin \theta d\theta = 0.$$

$$= \frac{2\pi + (n+m)!}{2\pi} \int_{-\infty}^{\infty} \frac{(n+m)!}{(n+m)!} \int_{-\infty}^{\infty} \frac{\pi}{2\pi} \int_{-\infty}^{\infty} \frac{\pi}{2\pi} \int_{-\infty}^{\infty} \frac{\pi}{2\pi} \int_{-\infty}^{\infty} \frac{(n+m)!}{(n+m)!} \int_{-\infty}^{\infty} \frac{\pi}{2\pi} \int_$$

Hence
$$a_{00} = \frac{1}{2} \cdot \frac{\pi}{2}$$
 $a_{01} = \frac{3}{2} \cdot \frac{2}{8} \cdot \frac{8}{12} \cdot \frac{1}{8} \cdot \frac{1}{12} - \frac{1}{8} \cdot \frac{1}{12}$

6-10 Consider a family change at A

and field at B, here

$$\frac{1}{|\vec{x} - \vec{x}'|} = \frac{1}{\sqrt{A^2 + G^2 - 2AG\cos\theta}} \cdots 0$$

outside sphere of radius A.

$$\frac{1}{|\vec{n} \cdot \vec{n}'|} = \sum_{n=0}^{\infty} c_n \, n^n \, P_n \, (\omega_n \sigma) = \sum_{n=0}^{\infty} c_n \, B^{-(n+1)} P_n \, (\omega_n \sigma) \cdots \, O$$

Let 0 = 0 in 1 and @ expand 1 in Taylor series in B 1 i.e.

$$\boxed{\begin{array}{c}
\boxed{1} \frac{1}{|\vec{x} - \vec{x}'|} = \frac{1}{8-A} = \frac{1}{6} \left[1 + \frac{A}{6} + \frac{A^2}{8^2} + \cdots \right]}$$

$$\frac{1}{|\bar{x} - \bar{x}'|} = \frac{C_0}{B} + \frac{C_1}{B^2} + \frac{C_2}{B^3} + \cdots$$
 Equating (1) to (2)
 $C_0 = 1, C_1 = A, C_2 = A^2, \cdots$

$$\frac{1}{|\vec{x} - \vec{x}'|} = \frac{1}{|\vec{x} - \vec{x}'|}$$

76 6-11 $Z_{+n}^{TM} = j\eta \frac{\Lambda_{n}^{(2)}/(kn)}{\Lambda_{n}^{(2)}(kn)}$ = $i\eta \frac{\hat{J}_n'(kr) - j \hat{N}_n'(kr)}{\hat{N}_n'(kr)}$ $\hat{J}_n(kx) - j \hat{N}_n(kx)$ $Z_{-n}^{TM} = -j\eta \frac{\bigwedge_{n}^{(i)}(kn)}{\hat{H}_{n}^{(i)}(kn)}$ $=-j\eta \frac{\Im'_{n}(kx)+jN'_{n}(kx)}{\Im_{n}(kx)+jN'_{n}(kx)}$ I hence $Z_{+n}^{TM} = \left(Z_{-n}^{TM}\right)^{T}$ assuming of & to be real. Rt. $Z_{+n}^{TM} = j \eta \frac{dt}{k_n \rightarrow \infty} \frac{H_n^{(2)}(k_n)}{H_n^{(2)}(k_n)}$ $= j \eta \frac{\sqrt{2/(\pi z)} (-j) e^{-jz - \frac{\eta \pi}{2} - \frac{\pi}{4}}}{\sqrt{2/(\pi z)} e^{-jz - \eta \pi/2 - \frac{\pi}{4}}}$ Since, dt. $H_{y}^{(2)}(3) \Rightarrow \frac{1}{\pi} \Gamma(y) \left(\frac{3}{2}\right)^{-3}$ $\chi_{\lambda} = H_{\lambda}^{(1)}(3) \Rightarrow \frac{1}{\pi} \Gamma(\lambda)(-\lambda)(\frac{3}{2})$ $\therefore \mathcal{L}t \cdot \mathcal{Z}_{+n}^{TM} = j\eta \frac{(kn)^{-\gamma-1}}{(kn)^{-\gamma}}(-\gamma)$ $=-j \gamma \gamma /(kn)$ Also since $Z_{+n}^{TE} = -j \eta \frac{H_n^{(2)}(kn)}{H_n^{(2)}(kn)}$ 2 $Z_{+n}^{Tm} = j \gamma \frac{H_n^{(2)}(kn)}{H_n^{(2)}(kn)}$: Z+n Z+n = 72

guin by

(A.) mu = pm (coso) {cosmp} A, (k)

[from (6.61)]

For the dominant mode

m = 0 and for v = 1 the

spatial TM mode is the

field of an electric current

element as shown in (6.85)

on p.287.

Similarly for TE to re modes the potential is quien by

(Fr) = Pom (600) { (20) } Hi (Rn)

[from (6.63)]

For the dominant mode

m=0 and for V=1, the

spatial TE mode which

is the field of an magnetic

current element (This is

dual of (6.85) on p. 287.

[6-13] The field components are guin by

 $E_0^{+} = \frac{jk}{\omega \epsilon_{R} \sin \theta} e^{\pm jkR}$ Where are the upper signs refer to inward - travelling waves and the lower signs to outward - travelling waves.

 $P_{f} = \int_{0}^{2\pi} \pi \sin d\phi \int_{0}^{\pi} r d\phi \frac{jk}{\omega \epsilon \pi \sin \theta} e \frac{(\pm jk\pi)}{\pi \sin \theta} e$ $= \frac{2\pi k}{\omega \varepsilon} \int_{0}^{0} \frac{d\theta}{\sin \theta} = \frac{2\pi k}{\omega \varepsilon} \left(\log \tan \frac{\theta}{2} \right)_{0}^{0} = 2\pi \gamma \log \frac{\tan \frac{\theta_{2}}{2}}{\tan \frac{\theta_{1}}{2}}$ $P_{d} = \int_{0}^{2\pi} \sin \theta_{1} d\phi \frac{R}{\pi^{2} \sin \theta_{1}} + \int_{0}^{2\pi} \pi \sin \theta_{2} d\phi \frac{R}{\sin^{2}\theta_{2}} = \frac{2\pi R}{\pi} \left[\cos \theta_{1} + \cos \theta_{2} \right]$

 $\therefore \alpha = \frac{Pd}{2Pf} = \frac{R}{2\eta^{TL}} \frac{\left[cosec\theta_1 + cosec\theta_2\right]}{\log \frac{cot\theta_1/2}{cot\theta_2/2}}$

6-14 The dominant TE mode for spherical wedge wave quide (Fr) nw = Pm (woo) cos w & Hn (2) (Rr), where

and p=0 and n=1 20 the domainant mode. Hena,

 $(F_R)_{NW} = \cos \theta H_1^{(2)}(kr); := \xi = -\frac{\sin \theta}{R} H_1^{(2)}(kr)$

 $H_0 = -\frac{k}{\Lambda_-} \hat{H}_1^{(2)} / (k_A) sin .$

for the large argument approximation of $H_1^{(2)}(kz)$ and Ex & Sind e-jkr Hi (& u) one the obtains

HO or y sing e-jkr.

which are the anal fields of an electric dipole given by 2-114).

[6-15] The field components for the spherical bornquide many be written as $H_n = f(0) \left[\hat{H}_{y}^{(1)}(kn) + B \hat{H}_{y}^{(2)}(kn) \right] \cos y \beta$. If we think of the lowest order mode propagating radially inward it would be quite similar to the TE10 mode of a rectangular guide, although modified by the convergence of the sides. We would consequently expect a cut-off plenomenon at such a redino τ_c that the width $\tau_2 \phi$, becomes half a wavelength . .: $\rho_i \pi_c = \lambda/2$. Avery effective cut-off phenomenon at about the radius To the reactive energy for a given power transfer becomes very great for radii less than this.

 $Z_{r}(kr) = j\eta \frac{H_{r}^{(2)}(kr)}{r}$ Jee radial impedames

92 $Z_n^-(kn) = -j\eta \frac{H_0(kn)}{H_0(kn)}$ become predominantly reactive at a value kr 22 which is compatible with $\phi_{Tc} = 3/2$. Circumferential modes might exist for radial lines greater than a wowelength in maximum circumference 6-16 The first ten resonant frequencies are $(f_n)_{mnp} = c \frac{u_{np} \text{ or } u'_{np}}{2\pi a} = \frac{3 \times 10^{\circ}}{20\pi} u_{np} \text{ or } u'_{np}$ = 1.31 GHz; 1.85; 2.15; 2.38; 2.76; 2.9; 2.92 and 3.34. Q dominant mode = $\frac{0.573 \times 12077}{2.61 \times 10^7 \times \sqrt{1.31 \times 10^9}} = 2.29 \times 10^4$ $\omega_n = \frac{2\pi c}{2\alpha} = \frac{\pi c}{\alpha}$ $\therefore \lambda_c = 2\alpha$ · k = 21 = T/a and Q = 0.35 1 = 1.4 × 104 ·Zin = jZotanka = jZotan = 0 18 The following results can be obtained from (6-26) (6-30) and Robbin equations (6-33). [6-19] The result for this case is the dual of eq. (6-86) and the field is given by $\bar{E} = -\bar{\nabla} \times \bar{u}_2 F_2$

field is given by $\vec{E} = - \vec{\nabla} \times \vec{u_3} F_3$

Field is green - $\frac{1}{6-20}$ Since the potential from a phipping fource is $\left[\frac{k \pm l}{4\pi j} h_o^{(2)}(k r) = A_3\right]$ Hence A_3^l for the madrupole source = $S_1 S_2 \frac{\partial^2}{\partial 3^2} (A_3^l)$ $= \frac{k^3 S_1 S_2 I l}{4 \pi j} \left[l_0'''^{(2)}(kz) \cos \theta + \frac{l_0'(kz)}{k} (kz) \frac{\pi - \frac{3}{2} l_1}{k} \right]$ Since h'(2) = h(2) [-h(2)(kn)] = \frac{1}{3} [h(2)(kn) - 2h(2)(kn)] $\frac{k_{1}^{(2)}(kn)}{2n} = \frac{1}{3} \left[k_{1}^{(2)}(kn) + k_{2}^{(2)}(kn) \right]$

" $A_3' = -\frac{k^3 S_2 Il}{12\pi i} \left[R_0^{(2)} (kn) + R_2^{(2)} (kn) \left\{ \sin \theta - 2\cos^2 \theta \right\} \right]$

Also P2 (1000) = = [3 con20+1] = [1000 - Sin 0/2]

: $A_3' = \frac{k^3 s_1 s_2}{12 \pi j} \left[-l_0^{(2)}(kn) + 2l_2^2(\omega s_0) l_2^{(2)}(kn) \right]$

$$\frac{-21}{|\pi - \pi'|} \frac{1}{|\pi - \pi'|} = \frac{k^{(2)} (|\pi - \pi'|)}{|\pi - \pi'|} \frac{1}{|\pi - \pi'|} = \frac{k^{(2)} (|\pi - \pi'|)}{|\pi - \pi'|} \frac{1}{|\pi - \pi'|} = \frac{k^{(2)} (|\pi - \pi'|)}{|\pi - \pi'|} \frac{1}{|\pi - \pi'|} \frac{1}$$

The result is an expansion of a cylindrical

Bessel function in a series of

94 of aphenical Bassel function
$$J_{m}(kn) = \sum_{\ell=0}^{\infty} \frac{(-1)^{\ell}}{2^{2\ell+m-1}} \frac{(4\ell+2m-1)}{(2\ell+2m)} \frac{(2\ell)!}{\ell! (\ell+m-1)!} \frac{\rho^{m}}{2\ell+m} \frac{(600)}{2\ell+m} \frac{j}{2\ell+m}$$

[6-23] Since
$$\frac{\sum_{n=0}^{(2)}(2n+1)h_n(n)f_n(n$$

Since $\int_0^{\pi} P_n(\omega n \theta) P_2(\omega n \theta) \sin \theta d\theta = 0$ for $n \neq 2$ = $\frac{2}{2n+1}$ for n = 2

6-24 As seen from equation (6.109) on p-296 for the dominant n=1 mode and for $\theta=60^{\circ}$, $E_{g}^{S}=0$, and $E_{g}^{S}\neq0$ and hence the wave is <u>linearly polarized</u>. For a general situation, when n>1, the scattering of a plane-polarized wave by a small conducting sphere no longer produces a linearly polarized wave at $\theta=60^{\circ}$, since E_{g}^{S} is no longer zero.

6-25

"Scottering of Electromagnetic Waves from Iwo Concentric Spheres!" by A.L. Aden and M. Kerker, Journal of Applied Physics, Volume 22, Number 10, October 1951

[6-26] The incident wave is
$$H_{x}^{i} = H_{0} = \frac{-jkx'\cos\theta'}{3}$$

$$H_{n}^{i} = H_{0} \frac{\cos\rho'}{jkn} \frac{\partial}{\partial\theta'} \left(e^{-jkn\cos\theta'}\right)$$

where $H_{0} = -\frac{j\omega\epsilon}{4\pi\pi} \frac{kl}{e} = \frac{-jkx}{4\pi\pi} \left(\frac{\partial^{2}}{\partial\theta'} \left(e^{-jkn\cos\theta'}\right)\right)$

or $H_{0}^{i} = -\frac{jH_{0}\cos\rho'}{4\pi\pi} \frac{\partial\theta'}{\partial\theta'} = \frac{\partial^{2}}{\partial\theta'} \left(\frac{\partial\theta'}{\partial\theta'}\right)$

Hence the electric vector potential is $F_n' = \frac{\text{Ho}}{\omega e} \cos \phi \sum_{n=1}^{\infty} a_n J_n(kn') P_n^{1}(\cos \theta')$

and $F_n^S = \frac{H_0}{\omega \epsilon} \cos \phi \sum_{n=1}^{\infty} C_n H_n^{(2)}(R_{n'}) P_n^{(1)}(\cos \theta')$. Hence the total

Potential $F = \frac{H_0}{\omega \epsilon} \cos \phi' \sum_{n=1}^{\infty} \left[\alpha_n \frac{\hat{J}_n(kn') + c_n H_n^{(2)}(kn')}{\hat{J}_n(kn)} \right] P_n^{\frac{1}{2}}(\cos \theta')$ where $c_n = -\alpha_n \frac{\hat{J}_n(kn)}{\hat{H}_n^{(2)}(kn)}$ since $E\phi = E_0 = 0$ at x = a

 $: \overrightarrow{H} = \frac{1}{j\omega\mu} \left(\frac{\partial^2}{\partial x^{i2}} + k^2 \right) \overrightarrow{F} \qquad . \text{ Since, } \left(\frac{\partial^2}{\partial x^{i2}} + k^2 \right) \overrightarrow{B}_n(kx^i) = \frac{n(n+i)}{b^2} \overrightarrow{B}_n(kx^i)$

: H, = H, cop/ = m(n+1)[an Jn(&n/)+cn Hn(2)(&n/)] Pn (con 0)

(similar to 6-117)

where $A_{n'b} = \frac{1}{b^{2}b^{2}} \sum_{n=1}^{\infty} \frac{1}{n} (n+1) \left[a_{n} \int_{n}^{\infty} (kb) + C_{n} \int_{n}^{\infty} (kb) \right] (-1)^{n} P_{n}(cos\theta)$ $= \frac{1}{4\pi k \eta b^{2}} \sum_{n=1}^{\infty} \frac{1}{n} \sum_{n=1}^{\infty} \frac{1}{n} (n+1) \left[a_{n} \int_{n}^{\infty} (kb) + C_{n} \int_{n}^{\infty} (kb) \right] (-1)^{n} P_{n}(cos\theta)$

6-27) For the case of a dielectric sphere there are two regions, one region r < a characterized by E_{L} , μ_{L} and the region r > a by E_{0} , μ_{0} . From the continuity of the tangential electric and mapstic fields we find that the fields external to the dielectric sphere will be guinn by (6-117) by hym. The protentials internal and external to the spheres are given by (6-101) and (6-112). The coefficients then are given by (6-113). Then the radiation fields are quein by (6-117) and by (6-113).

Fr =
$$\sum_{i} b_{i} J_{n}(k_{i})$$
 for $n < a$.

I have $A_{n}^{-1} = \begin{bmatrix} A_{n}^{(2)}(k_{a}) & for n < a \\ \frac{1}{2} f_{n}(k_{a}) & \frac{1}{2} f_{n}^{(2)}(k_{a}) \end{bmatrix}$

$$= -\frac{1}{3} f_{n}(k_{a})$$
and the results of (6.29) applies with A_{n} replaced by $J_{n}(k_{a})$.

$$A_{n} = \frac{J_{n}(k_{a})}{J_{n}}$$

$$\therefore A_{n} = \frac{J_{n}(k_{a})}{J_{n}} \sum_{i} J_{n}^{-1}(k_{a}) P_{n}^{(1)}(a) P_{n}^{(1)}(a)$$

[6-30] For a magnetic current the potential is given by $A_n = \sum_{n} C_n P_n(\omega n \theta) \hat{H}_n^{(2)}(kn) \quad \text{for } n > a \quad (6.137)$

where u are ordined solutions to Pu (cor 0,)=0 (6.138)

I and $C_u = -\frac{K}{\eta M_u} \hat{J}_u(ka) \sin \theta_2 \frac{\partial}{\partial \theta_2} P_u(\omega_2) \cdots (6.145)$

and $M_u = \frac{u(u+i)}{2u+1} \left[\frac{\partial P_u}{\partial \theta} \frac{\partial P_u}{\partial u} \right]_{\theta=0}$, (6.142)

In the limit as a >0, Tu(ka) -> 1 only for the first u and for the rest u's Tu(ka) ->0 and hance

Eo = $\frac{1}{9\pi} \frac{3^2 Ar}{200} = \eta f(r) \sin \theta R'(\cos \theta)$ only for the first $u = \theta R(\cos \theta) = 0$

For TM modes the equivalent circuit is for the n=1 mode

EX

Ich = IL + IR and jwpx IL = MIR

Ich = IL + IR and jwpx IL = MIR

IL | IR = IL | IR = Jwpx IL

IR = Jwpx IL

since whe ky and WEY= R

Now $G_{d} = \eta I_{R} I_{R}^{*} = \frac{\eta |I_{c}|^{2}}{1 + \frac{1}{R^{2} \pi^{2}}}$

and $\overline{W}_e = \frac{1}{2} C V_e V_c^* = \frac{1}{2} E \pi \frac{|I_c|^2}{\omega^2 E^2 \pi^2} = \frac{|I_c|^2}{2 \omega^2 E \pi} = \frac{|I_c|^2 \gamma_c}{2 k \pi \omega}$

 $\therefore Q = \frac{2\omega \overline{U}_{e}}{\overline{Q}_{e}} \text{ since } \overline{U}_{e} > \overline{U}_{m}$ $= \frac{2\omega |I_{e}|^{2} Y}{2k\kappa \omega \eta |I_{e}|^{2}} \left[1 + \frac{1}{k^{2}\kappa^{2}}\right] = \frac{1}{k\kappa} + \frac{1}{k^{2}\kappa^{3}}\Big|_{\rho=0}$

Since
$$-\overline{\nabla} \times \overline{E}_0 = j\omega_0 \mu \overline{H}_0 - I$$

 $\overline{\nabla} \times \overline{H}_0 = (\sigma + j\omega_0 E) \overline{E}_0 - \underline{T}$

$$-\nabla \times \vec{E} = j\omega \mu \vec{H} - \vec{M}$$

$$\nabla \times \vec{H} = (\sigma + j\omega \epsilon) \vec{E} - \vec{N}$$

Multiplying (I) by Eo and conjugate of (I) by H we obtain

Similarly 7. (Ho XE) = E. VXHo - Ho. VXE = (G+ju.E) E. Eo +jum.H. Ho. .. (8)

Subtracting B from (1)

ecting (B) from (A)
$$\nabla \cdot (\vec{H} \times \vec{E} \cdot) = -j(\omega, -\omega) \cdot \vec{E} \cdot \vec{E}_{0} - j(\omega - \omega_{0}) \mu \cdot \vec{H} \cdot \vec{H}_{0}$$

Taking III on both Endes

Jaking III on both sides

III
$$\nabla \cdot (\overline{H}_0 \times \overline{E}) d\overline{\tau} = 0$$
, since $\pi \times \overline{E} = 0$ on S' (:: $\int_{S'} \overline{E} \times \overline{H}_0 dS = 0$)

III $\nabla \cdot (\overline{H}_0 \times \overline{E}) d\overline{\tau} = 0$, since $\pi \times \overline{E} = 0$ on S' (:: $\int_{S'} \overline{E} \times \overline{H}_0 dS = 0$)

$$\iint \nabla \cdot (\widetilde{H}_{x} \times \widetilde{E}) d\tau = 0, \quad \text{since } \pi \times \widetilde{E} = 0, \quad \text{since } \pi \times \widetilde{E}_{0} = 0, \quad \text{s$$

So
$$\iint \bar{H} \times \bar{E}_0 dS = \iiint_{\Sigma'} (\omega - \omega_0) \left[\mu \bar{H} \cdot \bar{H}_0 - \epsilon \bar{E} \cdot \bar{E}_0 \right] d\Sigma$$

$$\Delta S = \frac{\int \int \widetilde{H} \times \widetilde{E}_0 dS}{\int \int \widetilde{E} \cdot \widetilde{E}_0 - \mu \cdot \widetilde{H} \cdot \widetilde{H}_0 dS}$$

$$\omega - \omega_0 = \frac{\int \frac{1}{\Delta_0^2}}{\int \int \left[\epsilon \, \bar{\epsilon}_0 \, - \mu \, \bar{H}_0 \, \bar{H}_0 \right] d r}$$

[7-2] from parjectly conducting wall - having a wall impedance of Z hence $\vec{n} \times \vec{E} = \vec{x} \cdot \vec{H}_E$. From previous problem

Jaking the volume integral [][V.[-H. x E+ H x E.] dr = \$\frac{H}{H.} x \overline{E} ds Since \$\sin \tilde{E} = 0 and S. Since \$\sin \tilde{E} = \frac{7}{4} \tilde{H}_{\tilde{L}} and

$$(\omega-\omega_{\bullet}) = \frac{- \cancel{\sharp} \cancel{\xi} \overline{H}.\overline{H}_{\bullet}dS}{\iiint[\varepsilon \overline{E}.\overline{E}_{\bullet} - \mu \overline{H}.\overline{H}_{\bullet}]dS}$$

By Substituting £. 4 H. ; H& E in the above equation,

It is assumed Win = We If $\omega = \omega_R(1 + \frac{j}{2Q})$ and $\mathcal{Z} = \mathbb{R} + j\mathcal{X}$

By taking real and imaginary parts of equation (it is clear.

$$\omega_{\pi} - \omega_{o} = \frac{- \iint \chi |H_{o}|^{2} ds}{2 \iiint \mu |H_{o}|^{2} d\tau} \quad \text{and} \quad$$

replaced by Wo since the difference is mey a few pur cent]

7-4 The impedance for a metal wall is Z=R+jR

Hence
$$\frac{\omega_{n}-\omega_{0}}{\omega_{0}} = -\frac{\#R |H_{0}|^{2}dS}{2 \int \int \mu |H_{0}|^{2}dT}$$

Surce

and so ₽.(4×£°)=E°. Δ×# - H. 4× €° = [6+46+jw(E+AE)]E.E. + jw. M. H. H.

$$\nabla \cdot (\overline{H} \circ x \overline{E}) = \overline{E} \cdot \nabla x \overline{H} \circ - \overline{H} \circ \cdot \nabla x \overline{E}$$

$$= (\sigma + j \omega \cdot \varepsilon) \overline{E} \cdot \overline{E} \circ + j \omega (\mu + \Delta \mu) \overline{H} \cdot \overline{H}$$

$$= (\sigma + j \omega \cdot \varepsilon) \overline{E} \cdot \overline{E} \circ + j \omega (\mu + \Delta \mu) \overline{H} \cdot \overline{H}$$

Subtracting @ from ® ⊽. (ਜ × €°) - ੲ. (ਜ• × €) = j(w-w.) [E E. E. - M H. H.] - jwAM H. H. + jω (Δε - <u>jΔ6</u>) Ē.Ē.

Taking the volume integral of the above equation [[v.(H×€.)-v.(H.×Ē)] d~

Since nxE. = 0 = nxE on surfaces

$$\therefore \frac{\omega - \omega_0}{\omega} = \frac{\left[\left[\left(\Delta \epsilon - \frac{j \Delta \epsilon}{\omega} \right) \overline{\epsilon} \cdot \overline{\epsilon}_0 - \Delta \mu \overline{H} \cdot \overline{H}_0 \right] d\gamma}{\left[\left[\epsilon \overline{\epsilon} \cdot \overline{\epsilon}_0 - \mu \overline{H} \cdot \overline{H}_0 \right] d\gamma} \right]$$

7-6 Let $\tilde{E} \approx \tilde{E}_0 = |E_0|$ and $\tilde{H} \approx \tilde{H}_0 = j|H_0|$ and $\omega \approx \omega_x + j \frac{\omega_0}{2\alpha}$ and from 7-5 $\omega_x - \omega_0 + j \frac{\omega_0}{2\alpha} = \frac{\omega_0}{||[\Delta \epsilon |E_0|^2 + \Delta \mu |H_0|^2] d^2}$ $||[[\epsilon |E_0]^2 + \mu |H_0|^2] d^2$

+ SSS DE 12 d2 SSS [E1E0] + M | Ho 12] d2

Equating the Imaginary parts and assuming $\overline{W}_m = \overline{W}_e$

 $\frac{\omega_o}{2Q} = \frac{\int \int \int \Delta \sigma |E_o|^2 d\tau}{2 \int \int \int E |E_o|^2 d\tau}$

: Q = W. [[[& | E |] d Z

7-8 From equation (7-18) $\omega-\omega_0 = -\frac{115}{2} \Delta \epsilon = \frac{1}{12} \Delta \epsilon = \frac{1}{12}$ $\frac{1}{12} \Delta \epsilon = \frac{1}{12} \Delta \epsilon =$

By the terms of the problem

Eint = East and the dominant

mode:

Eest = [sin # y sin # 3. ûx

 $\frac{\omega - \omega_0}{\omega} \approx -\frac{A_{va} \cdot \alpha \cdot \underline{c}^2 (\varepsilon_0 \varepsilon_R - \varepsilon_0)}{2 \varepsilon_0 \alpha \underline{c}^2 \int_0^{\varepsilon} \frac{1}{2} \sin^2 \frac{\pi y}{c} \sin^2 \frac{\pi y}{c} dy dy}$

 $\approx \frac{2 A_{res} (i - \varepsilon_r)}{bc}$

 $\frac{\omega_{n}-\omega_{o}}{\omega_{o}} + \frac{j}{2Q} = -\frac{\int \int \int (\epsilon'-\epsilon_{o}-j\epsilon'')|\epsilon_{o}|d\epsilon}{Q|\int \int \epsilon''|\epsilon_{o}|d\epsilon}$ and by equating the real parts
of the above equation $\frac{\omega_{o}-\omega_{T}}{\omega_{o}} = \frac{\int \int (\epsilon'-\epsilon_{o})|\epsilon_{o}|'d\epsilon}{Q|\int \int \epsilon''|\epsilon_{o}|'d\epsilon}$ Hence $Q \cdot \frac{\omega_{o}-\omega_{T}}{\omega_{n}} \approx \frac{\epsilon'-\epsilon_{o}}{\epsilon''}$

 $\frac{7-9}{\text{Same as in } 7-8} \text{ except}$ $\frac{3}{2+E_R} = \frac{3}{2+E_R} = \text{ext}$ $\frac{4\pi d^3}{3} = \frac{3}{2+E_R} = \frac{(E_N-1)}{2} \cdot \frac{E_0}{6} = \frac{3\pi}{2} \cdot \frac{1}{2}$ $= -\frac{\pi d^3}{abc} = \frac{E_V-1}{E_V+2}$



$$\frac{\Delta \omega_e}{\omega_e} = -\frac{\Delta W_e}{W}$$

$$H_{p} = \frac{1}{2} k_{p} (-jk_{3}) J_{1}^{\prime}(k_{p} p) \cos p e^{-jk_{3} 3}$$

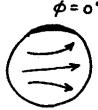
$$\Delta W_e = \frac{1}{2} \mathcal{E} \left| \mathcal{E}_f \right|^2 A_{re} \Big|_{f=a} = \frac{A \mathcal{E}}{2} \frac{J_1(k_a)}{a^2}$$

$$\phi = g\sigma$$

$$= -\frac{2AE}{2a^2 E\pi \left[k_1^2a^2-1\right]}$$

but & a = 1.841 and & a = 3.389

$$\frac{\Delta \omega_{c}}{\omega_{c}} = -\frac{A}{\pi a^{2}} \frac{1}{z.389}$$
$$= -0.418 A/(\pi a^{2})$$



for the y-polarization

$$= \frac{\mu A}{2} \frac{k_{p}^{4}}{2^{2}} J_{1}^{2}(k_{p}a)$$

$$W_{m} = \frac{\xi \pi}{2}$$
, 2.389, $J_{i}^{2}(k_{i}a)$

=
$$\frac{\mu A}{2} \frac{k_p^4 \cdot 2}{\omega \mu \epsilon \pi \cdot (2.389)}$$

and kpa = 1.841

$$= \frac{A}{\pi a^2} \frac{(1.841)^2}{2.389}$$

$$= 1.42 \frac{A}{\pi a^2}$$

Δω_e =

$$\frac{\int \Delta \omega_{e}}{\omega_{e}} = \frac{\int \int \Delta \varepsilon \ \tilde{E}_{int} \cdot \tilde{E}_{o} ds}{2 \int \int \varepsilon |\tilde{E}_{o}|^{2} ds}$$

By the terms of the problem

dominant mode.

$$\frac{\Delta \omega_{c}}{\omega_{c}} = -\frac{\left(\varepsilon_{v}-1\right) \frac{\pi d^{2}}{4} \varepsilon_{o}^{4} \left|\frac{2}{b/2} + \varepsilon_{o}\right|_{b/2}}{2 \int_{0}^{a} \int_{0}^{b} \varepsilon \varepsilon_{o}^{2} \frac{\sin \pi y}{b} dx dy}$$

$$= - (\xi_r - 1) \frac{\pi d^2}{4} \frac{2}{1 + \xi_r} \frac{\xi_0^2}{ab \xi_0^2}$$

$$= -\frac{\mathcal{E}_{r}-1}{\mathcal{E}_{r}+1} \frac{\pi d^{2}}{2ab}$$

$$7-13$$

$$\begin{array}{c}
\varepsilon_{2}, \mu_{2} \\
77/\varepsilon_{1}, \mu_{1}/1/1/1 \\
\end{array}$$

Alm p-p. = w spre E. Eint Z. ds

Since Eo = 2 sin my for tendominant

mode and Eint = Ez Eaxt.

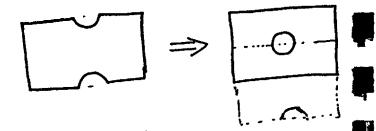
$$\omega_{\beta} - \beta_{0} = \frac{\omega(\frac{\epsilon_{1} - \epsilon_{2}}{\epsilon_{1}}) \int_{0}^{\beta} \int_{0}^{\epsilon_{1}} \sin^{2} \frac{\pi y}{b} dy dx}{2 \int_{0}^{\epsilon_{1}} \int_{0}^{\epsilon_{2}} \sin^{2} \frac{\pi y}{b} dy dx}.$$

 $= \frac{\varepsilon_{1}}{\varepsilon_{1}} \frac{\omega(\varepsilon_{1} - \varepsilon_{2})}{2} \frac{d}{a} \cdot z_{0}.$ Since

$$\omega \eta = \frac{\omega^2 \mu_0}{k_2}$$
, Hence

$$\beta - \beta_0 = \frac{\epsilon_1}{2} \frac{(k_1^2 - k_2^2)}{2\alpha k_1 \left[1 - (f_c/f)^2\right]^{1/2}}$$

[7-12] The problem of a half of a metallic of



A material perturbation is assumed inside the cylinder and let $\xi \to \infty$ and $\mu \to 0$ in the region $\Delta \tau$.

Hense as
$$\xi_{R} \rightarrow \infty$$
 in $7-11$

Hense as $\xi_{R} \rightarrow \infty$ in $4-11$

Hen $\Delta w_{c} = -\frac{\pi d^{2}}{2ab}$

For TEO2 mode

$$\frac{\Delta \omega_c}{\omega_c} = -\frac{\iint (\Delta \epsilon \, \bar{E} \cdot \bar{E}_0^* + \Delta \mu \, \bar{H} \cdot \bar{H}_0^*) dS}{\iint (\epsilon \, |\epsilon_0|^2 + \mu \, |H_0|^2) dS}$$

and $Ey^{\circ} = -\frac{j\omega\mu b}{2\pi}C \sin \frac{2\pi y}{b}$ $H_{x}^{\circ} = \frac{j\beta b}{2\pi}C \sin \frac{2\pi y}{b}$ $H_{z}^{\circ} = C \cos \frac{2\pi y}{b}$

$$\frac{1. \Delta \omega_{c}}{\omega_{c}} = \frac{\mu_{x} H_{z} |_{b/2} \pi d^{2}/4}{\mu_{x} \int_{0}^{1} \left[\frac{\omega' \mu_{x} \epsilon \mu_{x} b^{2} c^{2} \sin^{2}(\pi y^{2})}{4 \pi^{2}} + c^{2} \cos^{2}(\pi y^{2}) + c^$$

$$= \frac{\mu_v \pi d^2/4 C^2}{C^2 ab} = \frac{\pi d^2}{4ab}$$

7-14 By the terms of the problem, $\beta - \beta_s = \omega \frac{\int (\Delta \epsilon \hat{\epsilon}, \epsilon_s^*) ds}{s}$ 2 | (E. + x H). uzds Since for the dominant made E = ûx E, sin Ty and East = 2 Eint . Hence $\beta - \beta \approx \frac{\omega(\epsilon_{n-1}) \epsilon^{2} \pi d_{A}^{2} \cdot 2 \cdot \epsilon_{o}}{2 \cdot \epsilon_{o}}$ 2. ab . Eo2 (1+Ex) Also, $Z_0^{TE} = \frac{\sqrt{1 - (\omega_c)^2}}{\sqrt{1 - (\omega_c)^2}}$ and hence $\beta - \beta_0 = \frac{\pi d^2}{2ab} \cdot \frac{\mathcal{E}_{n-1}}{\mathcal{E}_{n+1}} \frac{\omega \eta \, \mathcal{E}_0}{\left[1 - (\omega_c/\omega)^2\right]^{\frac{1}{2}}}$ and to = WE. M and therefore $\frac{\beta - \beta_0}{+ k_0} = \frac{\pi d^2}{2ab} \frac{\xi_r - 1}{\xi_r + 1} \frac{1}{\left[1 - \left(\frac{\omega_c}{\omega}\right)^2\right]^{\frac{r}{2}}}$ - (6+4 c) F.E. Subtracting one equation from other $\#(\vec{\epsilon_0} \times \vec{H} - \vec{\epsilon} \times \vec{H_0}) = \iiint \left[\nabla \cdot (\vec{\epsilon_0} \times \vec{H}) - \nabla \cdot (\vec{\epsilon_0} \times \vec{H_0}) dt \right]$ = -j []] { [w. AE- j AO] E. E. - w. AM H. H. }d?

7-16 By the terms of the problem ¬.(Ε, ×+) = -jω, μ H, H, - jω, (ε+Δε) Ε. Ε. ₹.(Ē×Ho) = -jω. (μ+Δμ)H.Ho-jω.εĒ.Ē. and taking the volume integral. # = # side walls + # lot + # bottom = \$ top + \$ bottom & hance from 7.15 and after simplification [(ê,xA-êxA.).û3ds

103 7-15 et is gwen and E=fetz Ē. = Ê. e-1.2 H=Aetz Ho = Ho e - 407 Therefore, Ţ. (Ē.×Ĥ) - Ţ. (Ē×Ħ.) = H. V x E. - E. . V X H - H. . V X E + E. V X H. = -jwm H. Ho -jwE E. Eo +jwm H. Ho + jwe E.E. = 0Taking the surface integral $\#[(\bar{\epsilon}_0 \times \bar{H}) - (\bar{\epsilon}_0 \times \bar{H}_0)].d\bar{s} = 0.$ $= \Delta Z \stackrel{2}{=} \iint (E_0 \times \overline{H}) \cdot d\overline{S} - \iint (E_0 \times \overline{H}) dS$ (de) - de = nde dz = (7-80) D3 S (E0 × H) J. S- S (E, x H) 45 Similarly \$ (E × Ho). d\$ = (7-70) DZ ((E × Ho) : (γ-Y0) Δ3 [(£0×H - E × H0) . LS . - SS (E. x H) ds =0 Since SS (E.xH) ds = A3 \$ (E.xH). Tidl

$$\frac{-17}{1-8} \xrightarrow{\text{from}} \frac{7-18}{5} = \frac{17}{5} = \frac{17}{$$

Zet $\hat{E} \approx \hat{E}_{o}^{*}$ and $\hat{H} \approx -\hat{H}_{o}^{*}$ and $Y = \alpha + j\beta$ and substitution in the above equation and refaration of the real and imaginary parts lead to

$$d = \frac{\iint_{S} \Delta \sigma |\hat{E}_{o}|^{2} dS}{\iint_{S} (\hat{E}_{o} \times \hat{H}_{o}^{*} + \hat{E}_{o}^{*} \times \hat{H}_{o}) \cdot \hat{u}_{z} dS}$$

and
$$\beta - \beta_0 = \frac{\int \omega \left[\Delta \varepsilon \left| \hat{E}_0 \right|^2 - \Delta \mu \left| \hat{H}_0 \right|^2 \right]}{\int \left(\hat{E}_0^* \times \hat{H}_0 + \hat{E}_0 \times \hat{H}_0^* \right) \cdot \hat{U}_z dS}$$

7-20 Here.

$$\omega_{\pi} = \frac{\iint \bar{\mathbf{E}} \cdot \bar{\nabla} \times \mu^{-1} (\bar{\nabla} \times \hat{\mathbf{E}}) d\tau}{\iint \mathbf{E} |\bar{\mathbf{E}}|^{2} d\tau}$$

ح. [(بَدُ عَ× فَ) × فَ] = قَ. عَ× (بَدُ عَ× فَ) -(~, \$\vec{\pi} \). (\$\vec{\pi} \vec{\pi} \)

Taking the volume integral of both sides [[[v.[(~~ē)×ē]. dV = \$[~[~[√×ē)×ē]. āds

Since TXE = 0 them.

$$\omega_{\pi}^{2} = \frac{\int \int d^{2} d^{2} d^{2} d^{2}}{\int \int d^{2} d^{2} d^{2}}$$

7-18 nxE=ZH and from (7-15) we obtain y-y = \$\frac{\phi}{e} \times \hat{h} \. \tilde{\pi} \ dl ∬(Ê,×h-Ê×h.).ū, ds

Since
$$\hat{E}_0 \times \hat{H} \cdot \hat{\pi}$$

$$= \hat{\pi} \times \vec{E}_0 \cdot \hat{H}$$

$$= \cancel{\cancel{\xi}} \cdot \hat{H} \cdot \hat{H}_0$$

$$= \cancel{\cancel{\xi}} \cdot \hat{H} \cdot \hat{H}_0 \cdot \hat{H}_0 \cdot \hat{H}_0$$

$$= \cancel{\cancel{\xi}} \cdot \hat{H} \cdot \hat{H}_0 \cdot \hat{H}_0 \cdot \hat{H}_0 \cdot \hat{H}_0$$

$$= \cancel{\cancel{\xi}} \cdot \hat{H} \cdot \hat{H}_0 \cdot \hat{H}_0 \cdot \hat{H}_0 \cdot \hat{H}_0 \cdot \hat{H}_0$$

7-19 Let to = j A Z = R+jx $Y = \alpha + j\beta$

Hence from 7-18 φ(R+jx) | Ĥ. | dl d+ i(B-B) = . ∬(Ê. x Ĥ.* + Ê.* xĤ.).Û2ds φ(R+jx) | ĥ.|2 de 2 Re S (Ê. × Ĥ.*). Ûeds

β-β =
$$\frac{\int x |\hat{H}_{\bullet}|^2 dl}{2 Re \int (\hat{E}_{\bullet} x \hat{H}_{\bullet}^*).\hat{u}_z ds}$$

$$\begin{array}{ll}
\overline{7.21} & \omega_{\pi}^{2} = \frac{\int \left[\overline{\nabla} \times \overline{H} \right]^{2} d\tau}{\int \left[\overline{\nabla} \times \overline{H} \right]^{2} d\tau} = \frac{N}{D} \\
\delta \omega_{\pi}^{2} = \frac{1}{D^{2}} \left[D \delta N - N \delta D \right] \\
&= \frac{1}{D} \left[\delta N - \omega_{\pi}^{2} \delta D \right] \\
&= \frac{1}{D} \left[2 \iint \overline{\varepsilon}^{1} (\overline{\nabla} \times \overline{H}) \cdot (\overline{\nabla} \times \overline{\delta} \overline{H}) d\tau \\
&- \omega_{\pi}^{2} 2 \iint \overline{\mu} \, \overline{H} \cdot \delta \overline{H} \, \delta \tau \right] \\
&= \frac{1}{D} \left\{ 2 \iint \overline{\delta} \, \overline{H} \cdot \overline{\nabla} \times \overline{\varepsilon}^{1} (\overline{\nabla} \times \overline{H}) d\tau \\
&- \omega_{\pi}^{2} 2 \iint \overline{\mu} \, \overline{H} \cdot \delta \overline{H} \, \delta \tau \right] \\
&= \frac{2}{D} \iint \overline{\delta} \, \overline{H} \cdot \left[\overline{\nabla} \times \overline{\varepsilon}^{1} (\overline{\nabla} \times \overline{H}) - \omega_{\pi}^{2} \mu \overline{H} \right] d\tau \\
&= 0 \text{ for abbitrary } \delta H
\end{array}$$

$$\begin{array}{l}
\overline{7-23} \quad \text{Since}
\end{array}$$

$$\omega_{x^{\frac{1}{2}}} = \frac{\int \left(\vec{\nabla} \times \vec{E} \right)^{2} dx + 2 \left(\vec{\mu} \cdot \vec{\nabla} \times \vec{E} \right) \times E \right) \cdot ds}{\int \left(\vec{\nabla} \times \vec{E} \right)^{2} dx + 2 \left(\vec{\mu} \cdot \vec{\nabla} \times \vec{E} \right) \times E \right) \cdot ds}$$

But $\vec{E} \cdot \vec{\nabla} \times (-j\omega \vec{H}) = \vec{E} \cdot (\omega^{\epsilon}) \vec{E} = \omega^{\epsilon} |\vec{E}|^2$ and assuming $\vec{\nabla} = \vec{\nabla} \times (-j\omega \vec{H}) = \vec{E} \cdot (\omega^{\epsilon}) \vec{E} = \omega^{\epsilon} |\vec{E}|^2$ and assuming $\vec{\nabla} = \vec{\nabla} \times (-j\omega \vec{H}) = \vec{E} \cdot (\omega^{\epsilon}) \times (-j\omega \vec{H$

Since $\omega \to \omega_0$ and $E \to E_0$ hence $\omega_x^2 - \omega_0^2 = \frac{\iint \left[\mu^{-1} \left(\nabla \times \widetilde{E}_0 \right) \times \widetilde{E}_0 \right] \cdot dS}{\iint S \left[E \right]^2 dS}$

Also

[[[(\(\bar{v} \) \(\ba

$$= -j\omega_0 \iiint (\vec{E}_0 \cdot \nabla \times \vec{H}_0 - \vec{H}_0 \cdot \nabla \times \vec{E}_0) d\tau = \omega_0^2 \iiint [\vec{E}_0 | -\mu | H_0|^2] d\tau$$

$$\frac{\omega_{n}^{2} - \omega_{0}^{2}}{\omega_{0}^{2}} = \frac{\int \int \left(\epsilon |\epsilon_{0}|^{2} - \mu |H_{0}|^{2} \right) d\epsilon}{\int \int \left(\epsilon |\epsilon_{0}|^{2} - \mu |H_{0}|^{2} \right) d\epsilon}$$

$$\bar{H} = \hat{u}_{p} J_{1} (2.405 \frac{f}{a})$$
 and from (7.46)

in
$$\nabla \times \vec{H} = \hat{u}_3 \left[\frac{J_1(2.405 f/a)}{f} + \frac{2.405}{a} J_1'(2.405 f/a) \right]$$

=
$$2\pi d \left(\frac{\alpha}{2.405}\right)^{2} \left[\left(\frac{2.405}{2}\right)^{2} \left\{ \frac{1}{(2.405)^{2}} \right\} J_{1}^{2} \left(2.405\right) + J_{1}^{2} \left(2.405\right) \right]$$

$$\int_{0}^{\infty} (\nabla \times H)^{2} d\tau = \int_{0}^{\infty} \left\{ \frac{J_{1}^{2}(2.405)}{2.405} + P(\frac{2.405}{a}) \left[J_{1}^{2}(2.405) \right]^{2} \right\} dP(\frac{2.405}{a})$$

$$= \frac{1}{2} (2.405)^{2} \left\{ (1 - \frac{1}{(2.405)^{2}}) J_{1}^{2} (2.405) \right\} + \left\{ J_{1}^{2} (2.405) \right\}^{2} + 2.405 J_{1} (2.405) J_{1}^{2} (2.405) .$$

$$_{32}^{2} \approx \frac{2\pi b/(\epsilon_0 \epsilon_r) + 2\pi (d-b)/\epsilon_0}{2\pi \mu d (\alpha/2.405)^2}$$
 { reglecting the second term in the expansion}

$$\approx \left(\frac{2.405}{a}\right)^2 \frac{1}{\mu_0 \, \epsilon_0} \left[1 - \frac{b}{d} \left(1 - \frac{1}{\epsilon_A}\right)\right]$$

[7-25] The E-field formula is given by equation (7.80) and $\omega_c^{\perp} = \frac{\int \int \mu^{-1} (\nabla \times \vec{E})^{\perp} dS + 2 \Phi \left[\mu^{-1} (\nabla \times \vec{E}) \times \vec{E} \right] \cdot \vec{n} dl}{\left(\int \int |\vec{E}|^2 dS \right)}$

Let
$$\overline{E} = 2 \sin \frac{\pi y}{\alpha}$$
, then

$$\frac{1}{2}$$

Now
$$\nabla \times \vec{E} = -\frac{\hat{3}}{3} \frac{\pi}{a} \cos \frac{\pi y}{a}$$
 and so

$$\| \iint \vec{\mu}' (\vec{\nabla} \times \vec{E})^T dS = 2 \frac{\pi^2}{\mu a^2} \int_0^1 y \cos^2 \frac{\pi y}{a} dy = \pi^2/(8\mu)$$

$$\bar{n}_2 = \frac{\hat{x} - \hat{y}}{\sqrt{2}}$$
 hence

$$\oint \left[(\bar{\nabla} \times \bar{E}) \times \bar{E} \right] . \bar{\nabla} dl = \int_{2\sqrt{2}a}^{a/2} \sin \frac{2\pi y}{a} dy - \int_{\sqrt{2}\sqrt{2}a}^{\pi} \sin \frac{2\pi y}{a} dy = \frac{\pi}{2\sqrt{2}a} . 4 . \frac{\alpha}{2\pi}$$

$$: \omega_c^2 = \frac{\frac{\pi^2}{8\mu} + \frac{\sqrt{2}}{\mu}}{\frac{\alpha^2 \epsilon}{2}} = \frac{\pi^2 + 8\sqrt{2}}{\mu \epsilon \alpha^2} = \frac{9.87 + 11.3}{\mu \epsilon \alpha^2} = \frac{21.17}{\mu \epsilon \alpha^2}$$

[7-26] By the terms of the problem

$$(\vec{E} \times \vec{H}) \cdot d\vec{S} = \frac{A_1}{3} \sin^2 \frac{\pi y}{b} \sin \frac{2\pi z}{c}$$
. Hence

$$\iiint \left[\bar{\epsilon}, \bar{\sigma} \times \bar{H} + \bar{H}, \bar{\sigma} \times \bar{\epsilon}\right] \Lambda \tau = \frac{\alpha b c}{4} \left(-\frac{A c}{b}^{T} + \frac{A \bar{H}}{c}\right) \Lambda \iiint \left[\mu H^{2} - \epsilon \epsilon^{2}\right] \Lambda \tau$$

$$= \mu \left[\Lambda^{2} \frac{d c}{d} + \Lambda^{2} \frac{d c}{d}\right] \frac{d c}{d}$$

$$\therefore \omega = j \frac{\left(\frac{A_1}{c} - \frac{A_2}{b}\right)\pi}{\mu A_1^2 + \mu A_2^2 - \epsilon} \quad \text{Since } \frac{\partial \omega}{\partial A_1} = 0 = \frac{\partial \omega}{\partial A_2} \quad \therefore \frac{A_1}{A_2} = -\frac{b}{c} \quad \text{and}$$

$$A_1 = \frac{jb}{\gamma \sqrt{b^2+c^2}}$$
; $A_2 = \frac{-jc}{\gamma \sqrt{b^2+c^2}}$; hence $\omega = \frac{\pi}{bc} \sqrt{\frac{b^2+c^2}{E\mu}}$

>8 [7-27] An assumed E-field can be supported by the electric currents $\overline{J} = -j\omega\varepsilon \overline{E} - \frac{1}{j\omega} \overline{\nabla} \times (\mu^{-1} \overline{\nabla} \times \overline{E}) - - [A]$ Here no magnetic surface currents are required on S because $\vec{n} \times \vec{E} = 0$. But an additional magnetic surface current My = nx(E2-E,) is required on I since nx = ±0. Hence from Harrington (7.67) and equation @ and B O = -jusse E.Edr+jsse E. Tx(mi TxE)dr -js ff[Tx(Ez-Ei)].(midxE)ds Also [[Ē.ōx(rī ōxĒ)dr=]]] ō. (m-1ō xĒ) xĒdr+ [] (m-1ōxĒ). (ōxĒ)dr = \((m \ \varphi \x \varepsilon \) \x \varepsilon \delta \x \varepsilon \) \(\varphi \x \varepsilon \) \delta \x \varepsilon \) since $\vec{n} \times (\vec{\mu}^{T} \vec{\nabla} \times \vec{E})$ is continuous across $\vec{\mathcal{B}}$ and $\vec{n} \times \vec{E}$ is discontinuous = \[(\n^1 \overline{\rightarrow} \rightarrow (\overline{\rightarrow} - \overline{\rightarrow} \rightarrow \overline{\rightarrow} \rightarrow (\overline{\rightarrow} - \overline{\rightarrow} \rightarrow \overline{\rightarrow} \rightarrow \overline{\rightarrow} \rightarrow (\overline{\rightarrow} - \overline{\rightarrow} \rightarrow \overli jussset dt = issser (r v x E). (v x E) dt + is \$2(r v x E) x (ēz-Ēi) .: W= (\(\varepsilon_{\varepsilon'}\varepsilon_{\varepsilon}\varepsilo ∭εE² de By the terms of the problem III Ē· ▽ ×(m' ▽×Ē) dt = III ▽·(m' ▽×Ē) ×Ē dt + III(m' ▽×Ē). (▽×Ē) dt Since Tix E is continuous at & and Tix = = 0 on S = []] M-1 (AXE)_ 45 를 주×(Ē,-Ē,). μ'(주×Ē)ds=0 Hence 0= -ju ||| EEt de + ju ||| pt (\$x \in) de and so m= [[] m1 (4x E) de

SSEZYZ

7-29 From (7-72)

 $\omega \iiint (\mu H^2 - \epsilon E^2) d\tau = j \iiint (\bar{E} \cdot \nabla \times \bar{H} + \bar{H} \cdot \nabla \times \bar{E}) d\tau - j \not D \bar{E} \times \bar{H} \cdot ds$ or, by terms of the problem: $\omega \iiint (\mu H^2 - \epsilon E^2) d\tau + j \not D \bar{E} \times \bar{H} \cdot dS = j \iiint [j\omega (\epsilon + \Delta \epsilon)|\epsilon|^2$ $\omega \iiint (\mu H^2 - \epsilon E^2) d\tau + j \not D \bar{E} \times \bar{H} \cdot dS = j \iiint [j\omega (\epsilon + \Delta \epsilon)|\epsilon|^2$ $-j\omega(\mu + \Delta \mu) |H|^2] d\tau$

similarly for the importurbed county

woll (MH==EE=) dT+jffE.×Ho.ds

=j[[[jw_E |Eo|^-jw_po |Ho|^] dT

Since mxE=0 on S and mxE=0 on S, and in the limit, as $\Delta E \rightarrow 0$, and $\Delta \mu \rightarrow 0$, we can approximate E, H, ω by E, H, ω and obtain

<u>ω-ω.</u> = - <u>[[[(ε|Ε.)] + Δμ (H.)]) ατ</u> [[(ε |Ε.)] + μ | H.)] ατ

It is assumed in deriving the above formula that E and H are 90° out of phase in the loss free case.

7-30 chapter 9 -Field Computation by moment methods by R.F. Harrington.

[7-31] By the terms of the problem,

$$\omega_{e}^{2} = \iint \mu^{-1}(\bar{\nabla} \times \bar{E})^{-1} ds = \bar{\mu}^{-1} ab \text{ and } 6[$$

$$\lim_{x \to \infty} |\mu^{-1}(\bar{\nabla} \times \bar{E})^{-1} ds = \bar{\mu}^{-1} ab \text{ and } 6[$$

and
$$\omega_{e}^{2} = \frac{\int \int \mu'(\bar{\nabla} \times \bar{E})^{2} ds + 2 \oint [(\mu'\bar{\nabla} \times \bar{E}) \times \bar{E}].\bar{n} dl}{\{\int \int \mathcal{E}^{2} ds\}}$$

Since
$$\nabla \times \vec{E} = \vec{u}_3 \vec{\pi} \cos \vec{\pi}_2$$
 and $(\nabla \times \vec{E}) \times \vec{e} = -\hat{U}_2 \vec{\pi} \sin^2 \vec{\pi}$

E = Uy Suh TX

Also
$$\iint \mathcal{E} \mathcal{E}^{-} dS = \int_{0}^{b} dx \left[\int_{0}^{d} \sin 2\pi x \mathcal{E}_{+} dx + \int_{0}^{a} \sin^{2} \pi x dx \right]$$

$$= \frac{ab\mathcal{E}}{2} \left[1 + (\mathcal{E}_{n} - 1) \left\{ \frac{d}{a} - \frac{1}{2\pi} \sin^{2} \frac{2\pi d}{a} \right\} \right]$$

:.
$$\omega_c^2 = \frac{\pi^2}{a^2 \mu^2} \frac{1}{\left[1 + (\Sigma_{r-1})\left\{\frac{d}{a} - \frac{1}{2\pi} \sin \frac{2\pi d}{a}\right\}\right]}$$

7-32 By the terms of the problem.

Multiplying Aby F and Bby E D×Ē + jωμ H̄ = jβ ŪexĒ TXH = jue = jpuexn - ®

and subtracting one from the other we obtain

Q. (EXH) + j ω (μη²+ε ε²) = j 2β ūz. Ex H After taking the integral over the whole cross-section of the

B= WSS(MHZ+EEZ)ds-j & EXH. Adl vane guide 2 MEXH. Uz ds

$$\beta = \frac{\omega \int_{S}^{S} (\mu H^{2} + \epsilon E^{2}) ds - j \oint_{S}^{E} E \times \overline{H} \cdot \overline{n} dl}{2 \int_{S}^{S} (\hat{E} \times \hat{H} \cdot \hat{U}_{2}) ds}$$

$$\beta = \frac{\omega \int_{S}^{S} (\mu H_{0}^{2} + \epsilon E_{0}^{2}) ds - j \oint_{S}^{E} (\overline{E}_{0} \times \overline{H}_{0}) \cdot \overline{n} dl}{2 \int_{S}^{S} (\overline{E}_{0} \times \overline{H}_{0}) \cdot \overline{n} dl}$$

The two denominators can be assumed to be the came, because they represent the time average power flow. Hence if $E \approx E_0 \& \bar{H} \approx \bar{H}_0$ li (MHg-ElEO) de

{Application of } 1-62 leads to } 2 [] (E, XH,). ú, ds f^{-34} If we assume the material is homogeneous than f^{+} and f^{-} of equation (7.86) becomes $\hat{f}^{+} = \hat{f}^{-} = \hat{f}^{-}$ and similarly $\hat{f}^{+} = \hat{f}^{-} = \hat{f}^{-}$. Since the Lenominator is twice the average power flows in the guide, it may be assumed to be the same both in the perturbed in the guide, it may be assumed to be the same both in the perturbed and upperturbed guide for Shellow, smooth deformations of waveguide walls. Hence

 $P - P_{\bullet} = \frac{\iint \left[\omega \Delta E\left(\overline{E}_{\bullet}\right)^{2} - \omega \Delta \mu \left(\overline{H}_{\bullet}\right)^{2}\right] dS}{2\iint \overline{E}_{\bullet} \times \overline{H}_{\bullet} \cdot \overline{U}_{E} dS}$ $Since in the loss free care
<math display="block">\overline{E}_{0} = 2E_{0}$ $\overline{H}_{0} = 2(-i)H_{0}$

Hence $\beta - \beta_0 = \frac{\omega \iint \left[\Delta E E_0^2 + \Delta \mu H_0^2\right] dS}{2 \iint \overline{E}_0 \times \overline{H}_0 \cdot \overline{u}_0 dS}$

[7-35] Issume a constant current Iz 3. Ilen

$$Z_{ii} = -\frac{\langle u, u \rangle}{T^2} = \frac{-\iint E_a \cdot \overline{I}_a ds}{T d T^2}$$
. Now

$$E_3^{\alpha} = \frac{-k^2 I_S^{\alpha} H_0^{(2)}(k_f)}{4\omega E}$$
 and $H_{\phi} = -\frac{k I_S^{\alpha}}{4j} H_0^{(2)}(k_f)$ (5.84)

Expanding
$$H_1^{(2)}(AP) \xrightarrow{\frac{1}{kP} \to 0} \left(\frac{kP}{k}\right) - \frac{j}{\pi} \left(\frac{2}{kP}\right) - \cdots \left(0.10\right)$$

$$I = \left(\frac{k\rho}{4j}\right) I_s^{\alpha} 2\pi \left[\frac{k\rho}{2} - \frac{j}{\pi} \frac{2}{k\rho}\right] \left|_{\rho = d/2} = \frac{I_s^{\alpha}}{4j} \left[\pi \left(\frac{k^2 d^2}{4}\right) - j4\right]$$

Hence
$$E_3^{\alpha}$$
 [1 - $\frac{j_2 \log \frac{y + y}{2}}{4\omega \epsilon}$ [1 - $\frac{j_2 \log \frac{y + y}{2}}{4\omega \epsilon}$] -- (D.9)
for Hankel for.

$$Z_{in} = \frac{k^2 a}{4\pi\epsilon} \left[1 - j \frac{2i \log \frac{7kd}{4}}{\pi} \right] \quad \text{for } a << \lambda$$

$$= \frac{k \pi a}{4} \left(1 - j \frac{2}{\pi} \log \frac{y k d}{4}\right)$$

2
$$\frac{7-36}{7-36}$$
 Since $Z_{ii} = -\frac{\langle a,a\rangle}{I^2} = -\frac{\int \int E^a \cdot I_s^a dS}{\pi d I^2}$ knowing E^a , $I_s^a RI$ will be

Eight do exclusite
$$Z_{ii}$$
.

$$E_{3}^{\alpha} = \frac{-k^{2} I_{s}^{\alpha}}{4 \omega \epsilon} H_{0}^{(2)}(kp) \quad \text{and} \quad H_{p}^{\alpha} = -\frac{k}{4j} I_{0}^{(2)}(kp) = \frac{k I_{s}^{\alpha}}{4j} H_{1}^{(2)}(kp)$$

Tok I_{s}^{α} $I_{s}^{(2)}$

$$N_{\text{ow}} I = \int_{0}^{2\pi} H_{p}^{\alpha} \cdot \rho \, d\rho \Big|_{\rho = \frac{d}{2}} = \frac{2\pi \rho \, k \, T_{s}^{\alpha} \, H_{s}^{(2)} \left(\frac{kd}{2} \right)}{4j}$$

$$\langle E_3^{\alpha}, I_5^{\alpha} \rangle = -\frac{k^2}{4\omega^2} H_0^{(2)} (\underline{k}\underline{k}) \pi d \int_0^{\infty} \cos^2 k(\alpha-3) d3$$

$$-\frac{k^{2}}{4\omega\epsilon}H_{o}^{(2)}(\frac{kd}{2})\frac{\pi d}{2}\left[\alpha+\frac{\sin 2k\alpha}{2k}\right]$$

Hence
$$Z_{ii} = -\frac{k^2}{4\pi\epsilon} \frac{H_o^{(2)}(\frac{kd}{2})}{\pi d} \frac{\pi d}{2} \frac{\left[\alpha + \frac{\sin 2k\alpha}{2k}\right]}{\pi^2 d^2 k^2 \cos^2 k\alpha \left\{H_o^{(2)}(\frac{kd}{2})\right\}^2}$$

$$= -\frac{2}{\omega \epsilon} \frac{\left(\alpha + \frac{\sin 2k\alpha}{2k}\right) + H_0^{(2)}(\frac{kd}{2})}{\left[\pi d \cos k\alpha + H_1^{(2)}(\frac{kd}{2})\right]^2}$$

$$\frac{7-37}{I_{v}^{2}\langle u,u\rangle - 2I_{u}I_{v}\langle u,v\rangle + I_{u}^{2}\langle v,v\rangle}$$
 from (7.99)

Here
$$I'' = \cos k(a-3)$$
; $I'' = 1$; hence

$$\langle u, u \rangle = -\frac{k^2}{4\omega\epsilon} H_0^{(2)} \left(\frac{kd}{2}\right) \frac{1}{2} \left(a + \frac{\sin 2ka}{2k}\right)$$
 from $\frac{7-36}{2}$

$$\langle v, v \rangle = -\frac{k^2}{4v^2} \alpha H_0^{(2)} (\frac{kd}{2})$$
 from $\frac{7-35}{4v^2}$

$$\langle u,v\rangle = \iint E^{u} \cdot J^{v} ds = \int_{0}^{\infty} \frac{-k^{2}}{4wE} H_{0}^{(e)}(\frac{kd}{2}) \cos k(a-3) d3$$

$$= -\frac{k^2}{4\omega\epsilon} H_0^{(2)} \left(\frac{kd}{\epsilon}\right) \frac{\sin(4\alpha)}{k}$$

$$\frac{1}{2} \frac{1}{4\omega \epsilon} = \frac{\left[\frac{k^2}{4\omega \epsilon} + \frac{(2)}{6} \left(\frac{kd}{2}\right) \frac{\sin k\alpha}{k}\right]^2 - \left[\frac{k^2}{4\omega \epsilon} + \frac{(2)}{6} \left(\frac{kd}{2}\right)\right]^2 \frac{\alpha}{2} \left(\alpha + \frac{\sin 2k\alpha}{2k}\right)}{k}$$

$$-\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kk}{2})\left[\frac{1}{2}(\alpha+\frac{\sin k2\alpha}{2k})-\frac{2\sin k\alpha\cos k\alpha}{k}+\cos^{2}k\alpha\right]$$

$$-\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{1}{2}(\alpha+\frac{\kappa_{0}R_{2}R_{1}}{2k})-\frac{\alpha^{2}}{4}-\frac{\alpha \sin^{2}k\alpha}{4k}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha \sin^{2}k\alpha}{4k}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha \sin^{2}k\alpha}{4k}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha \sin^{2}k\alpha}{4k}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\sin^{2}k\alpha}{k^{2}}+\cos^{2}k\alpha\cdot\alpha\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\sin^{2}k\alpha}{k^{2}}+\cos^{2}k\alpha\cdot\alpha\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\sin^{2}k\alpha}{k^{2}}+\cos^{2}k\alpha\cdot\alpha\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\sin^{2}k\alpha}{k^{2}}+\cos^{2}k\alpha\cdot\alpha\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha^{2}}{2}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\alpha\sin^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}+\frac{\cos^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\cos^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\cos^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\cos^{2}k\alpha}{k^{2}}\right]$$

$$=\frac{k^{2}}{4\omega\epsilon}H_{0}^{(2)}(\frac{kd}{2})\left[\frac{\sin^{2}k\alpha}{k^{2}}-\frac{\cos^{2}k\alpha}{k^{2}}\right]$$

$$= -\frac{k\eta}{4a} + \frac{(2)(\frac{kd}{2})[\frac{1}{k^2} - \frac{\chi^2}{32}]}{4a} = \eta + \frac{(2)(\frac{kd}{2}) \times (0.0386)}{(0.0386)}$$

[7-38] egwin
$$J_y = \hat{u}_y \cos \frac{\pi y}{a}$$
 and

$$L_{e} = \frac{\pi}{2\lambda} \left| \eta \left(\frac{\phi J_{3}^{a} e^{jkx} dl}{\phi \overline{\epsilon}_{a}^{a} \cdot \overline{J}^{a} dl} \right)^{2} \right|^{2}$$

- (7. 124) Harrington

and
$$\oint \vec{E}^{\alpha} \cdot \vec{J}^{\alpha} dl = \int_{\alpha/2}^{\alpha/2} \vec{J}_{j}^{\alpha} dy$$

$$= \int_{\alpha/2}^{\alpha/2} \vec{J}_{j}^{\alpha} dy = -P, \text{ where } P$$

is the complexe power permit length supplied by Jya. But the ribbon

have already been analyzed in

Sec 4.12 and Zale rib = $\frac{\chi^2}{2}$ Yapert.

In defining Yapers = $\frac{P^*}{|V|^2}$ (p. 185)

of the apenture. Hence

P= 15/ Zelec xib where I is the

current per unit length & to

P =
$$\frac{\gamma^2}{2}$$
 Yapart = Zelee vib (Eince | []=1)

$$\therefore L_e = \frac{\pi}{2\lambda} \left| \frac{4\alpha^2}{\pi^2} \frac{2}{\eta^2} \frac{2}{\gamma^2 \eta^4} \right|^2$$

$$= \frac{32\alpha^4}{\pi^3 \lambda} \left| \frac{1}{\eta \text{ Yapert}} \right|^2$$

7-39 sere

$$A_e = \pi \left| \frac{\eta}{\lambda} \frac{\left(\iint J_e^a e^{jkx} ds \right)^2}{\iint \widehat{E}^* . \widehat{J}^a ds} \right|$$

- (7.115) Harrington

Since
$$\int_{-L/2}^{L/2} \cos kz = \frac{2}{k} \Big|_{L=3/2} = \frac{\lambda}{11}$$

and (a,a) = 73 = \$ = . Jads

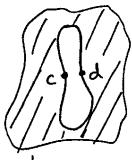
7-40

Same as [7-41] with 0=0

[7-41] Let the E-field be of the form E = E. (uz sin 0 + up cos 0) e jez cos 0 The current induced on the plate then has a form $J_2 = \sin \theta e^{ikz\cos \theta}$ Hence \$\overline{\pi_3}. \overline{\pi_2} transmitter $= \frac{2}{k} \frac{\cos(\frac{\pi}{2}\cos\theta)}{\sin\theta} = \frac{\lambda}{\pi} \frac{\cos(\frac{\pi}{2}\cos\theta)}{\sin\theta}$ Similarly \$ uz. Freceiver ds = $\frac{\lambda}{\pi} \frac{\cos(\frac{\pi}{2}\cos\theta')}{\sin \theta'}$. Hence application of (7.135) yields $He = \frac{\pi}{\lambda} \frac{\frac{\gamma}{\pi^2} \frac{\lambda^2}{73} \frac{1}{\frac{\cos(\frac{\pi}{2} \cos \theta)}{\sin \theta}} \frac{\cos(\frac{\pi}{2} \cos \theta')}{\frac{\sin \theta'}{\sin \theta'}}$ $\approx 0.86 \, \lambda^2 \left[\frac{\cos(\frac{\pi}{2}\cos\theta)}{\sin\theta} \cdot \frac{\cos(\frac{\pi}{2}\cos\theta')}{\sin\theta'} \right]^2$

[7-42] By the terms of the problem, the currents induced on the obstacle is given by Jd = jw(ε-ε.) E = Ke E, and Md = jw(~~ Mo) H = KmH (from 7 - 136). Hence (7-137) can be replaced by $-V_{n}^{S} = \int \int \left[(\bar{E}^{i})^{n} \cdot (\bar{J}^{d})^{t} - (\bar{H}^{i})^{n} \cdot (\bar{M}^{d})^{t} \right] dv$ and (7-138) by - V= = \$ [[K= (Jd) (Jd)t - (Es)2. (34) + K (M4)2 (M4)2 - (Hs)2. (Md)+]d2 = F (Cx,Ct) - < Cx,Ct) Hence (7.142) can be expressed as - Vn = [[[[[[[[[i]]], (]]], ([]]), ([])], ([])] [[[K=1 (Ja)2. (Ja)t-K=1 (Ma)2. (Ma)t]dt × [[][{(£i)t.(ī~)*_ (Hi)t.(Ma)*} dz] - [[[(Ea)2.(3a)4-(Ha)2. (Ha)4 } dt when the fransmitter and receiver are represented by the same Source (7.143) would then be replaced by [[[{ K=1 (Ja) = K=1 (Ma) } d]

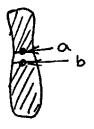
- [] [E , Ja _ Ha, Ma} de



$$I = 2 \int_{a}^{b} \overline{H}_{s} . d\overline{s}$$

$$V = \int_{e}^{d} \overline{E}_{s} . d\overline{s}$$

$$Y_{s} = \frac{2 \int_{a}^{b} \overline{H}_{s} . d\overline{s}}{\int_{e}^{d} \overline{E}_{s} . d\overline{s}}$$



$$V = -\int_{0}^{a} \overline{E}_{d} \cdot d\overline{S}$$

$$T = \oint_{0}^{a} \overline{H}_{d} \cdot d\overline{S} = 2 \int_{0}^{a} \overline{H}_{d} \cdot d\overline{S}$$

$$Z_{d} = -\int_{0}^{a} \overline{E}_{d} \cdot d\overline{S}$$

$$Z_{d} = -\int_{0}^{a} \overline{E}_{d} \cdot d\overline{S}$$

$$Z_{d} = -\int_{0}^{a} \overline{E}_{d} \cdot d\overline{S}$$

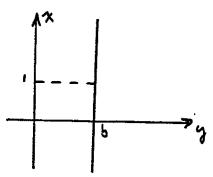
At points, sufficiently distant away the fields are essentially plane waves hence $E_S = \gamma H_S$ and $E_A = -\gamma H_A$. Also it is apparent that, mathematically these two problems are identical. It is necessary only to interchange E and H to pass from one problem to the other. Therefore, except for a constant, the solution obtained for E for the slot will be the same as the solution for H for the dipole and it is possible to write for the fields at any corresponding points $E_S = k$, Hd where the subscripts E and E respectively. Similarly, the magnetic field of the slot and the dipole respectively. Similarly, the magnetic field of the slot and the dipole E and E the dipole are related by E and E and E there E and E are related by E and E and the dipole E and E are related by E and E and the dipole E and E are related by E and E and E are related by E and E are related by E and E are related by E are related by E and E are related by E and E are related by E and E are related by E are related by E and E are related by E and E are related by E and E are related by E are related by E and E are related by E and E are related by E and E are related by E are related by E and E are related by E and

 $\frac{7-44}{30-3} = \frac{1}{2} \times \frac{1}{2}$

$$T = \frac{2\eta}{\omega_{\lambda}} \frac{\omega_{\lambda}^{2}}{4\eta^{2}\omega^{2}} \frac{\omega_{\lambda}^{2}}{4\eta^{2}\omega^{2}} = \frac{\lambda}{2\omega_{\eta}} \frac{\partial \omega_{\eta}}{\partial \omega_{\eta}} \frac{\partial \omega_{\eta}}{\partial \omega_{\eta}}$$

$$T = \frac{\lambda}{\omega} \frac{120\pi 2}{21 \times 4 \times 73} = 0.645 \frac{\lambda}{\omega}$$

From prob. 2-28:



For TM case:
$$e^{m} = -\nabla_t \gamma_n^{m} = -\frac{n\pi}{b} \cos \frac{n\pi y}{b} \overline{u}_y$$

$$\iint_0^b e^{m} e^{m} dxdy = \left(\frac{n\pi}{b}\right)^2 \frac{b}{b}$$

:. Normalized eigenfunction is
$$\overline{E}_{n}^{m} = \frac{\sqrt{26}}{n\pi} \sin \frac{n\pi y}{6}, n=1,2,...$$

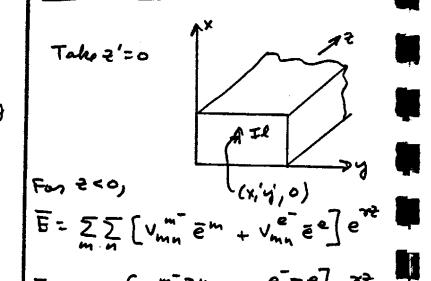
For TE case,

$$e^{e} = \overline{u}_{2} \times \overline{v}_{2} \gamma_{n}^{e} = \frac{n\pi}{b} \sin \frac{n\pi}{b} \sqrt{4z}$$

$$\int_{0}^{b} e^{m} \cdot e^{m} dx dy = \left(\frac{n\pi}{b}\right)^{2} \frac{b}{2}$$

- 2 an arbitrary field mariolo a waveguide can be expressed as a sum over all possible

8-2 (cont)



and for 270,

The E field must be continuous

$$\mathcal{H}_{mn}^{e} = \frac{1}{\pi \sqrt{\frac{ab \, \mathcal{E}_{n} \, \mathcal{E}_{m}}{(mb)^{2} + (na)^{2}}} \cos \frac{m\pi \times \cos n\pi \times$$

$$A_{mn}^{m} = \frac{2}{\pi} \sqrt{\frac{ab}{(mb)^2 + (nq)^2}} \sin \frac{m\pi}{a} \times \sin \frac{n\pi y}{b}$$

$$\gamma_{on}^{m} = 0$$
 , $n = 1, 2, \dots$

$$H_{X}^{+} = \sum_{m} \sum_{n} \left(\frac{n \sqrt{nn}}{2 \sqrt{n6}} + \frac{m \sqrt{nn}}{a 2 \sqrt{n6}} \right) \sqrt{\frac{a 6}{(m6)^{2} + (na)^{2}}} \sum_{m} \frac{m \pi x}{a} \cos \frac{m \pi x}{6} e^{-32}$$

$$H_{y}^{2} = \sum_{n} \sum_{h} \left[\frac{n V_{mn}}{b z_{0}^{e}} - \frac{m V_{mn}}{a z_{0}^{m}} \right] \left[\frac{a b}{(m b)^{2} + (n a)^{2}} \cos \frac{m \pi x}{a} \sin \frac{n \pi y}{b} e^{-7 z} \right]$$

This give
$$4 \frac{Z}{a} \frac{Z}{a} \left[\frac{m v_{mn}}{a \stackrel{2}{\epsilon}_{b}^{m}} - \frac{n v_{mn}}{b \stackrel{2}{\epsilon}_{b}^{e}} \right] \sqrt{\frac{ab}{(mb)^{2} + (na)^{2}}} \cos \frac{m \pi x}{a} \sin \frac{n \pi y}{b}$$

multiply through by cosmit sin mity and integrate from o to a and o to 6.

$$V_{mn} = m \frac{2}{a} \sqrt{\frac{b}{a}} \frac{I l}{\sqrt{(mb)^2 + (nq)^2}} \cos \frac{m\pi x}{a} \sin \frac{n\pi y}{b} = \frac{72}{6}$$

$$V_{mn}^{e} = -n \frac{2e}{\sqrt{b}} \int_{b}^{a} f_{mn}$$

For m =0, 4

Egn. to be satisfied is

weetigh through by sin noty and integrate from ote 6 and oto a, Von = -Il 20 √206 sin my' e-72

$$\sin = -IR \approx \sqrt{2ab}$$

$$w_e = \frac{1}{2} \iint E |E|^2 dS$$

$$u_{\ell} = \frac{1}{2} \iint E |E|^2 dS$$

For each mode,

and |e|= |em|=1

Because em = -0+ xm.

Thus for all modes,

$$I_{1} = \{ + (I_{1} - I_{2}) \ge a = V_{1} \}$$
 $I_{2} = \{ + (I_{2} - I_{1}) \ge a = V_{2} \}$
 $I_{1} = \{ + \{ + \}_{2} \}$

From analogy with transmission, line formulas,

$$\frac{dt_e^*}{dt} = jk_t A e^{jk_t t}$$

$$= R \left[\int \frac{k^2}{w_{jl}^2} \left(\frac{\partial x^e}{\partial x} \right) dx + \int \frac{k_c^4}{w_{jl}^2} \left(\frac{\partial x^e}{\partial x} \right) dx \right]$$

$$\alpha_{c} = \frac{P_{d}}{2P} = \frac{R k^{2}}{2w_{R}} \left[\frac{g(\frac{\partial A_{e}}{\partial R})^{2}}{2k} \right]^{2}$$

=
$$\frac{Rke}{27k} \int \left[\left(\frac{\partial \gamma_e}{\partial \ell} \right)^2 + \frac{kc}{k^2} \left(\gamma_e \right)^2 \right] d\ell$$

For The modes

$$\overline{H}_{tom} = \overline{\ell} \cdot \left(-\overline{u}_2 \times \nabla_{\ell} \gamma^{m} \right)$$

$$= -\frac{\partial \gamma^{m}}{\partial n}$$

$$4^{e} = \frac{1}{\pi \sqrt{(m6)^{2} + (na)^{2}}} \cos \frac{m\pi x}{a} \cos \frac{n\pi y}{6}$$

$$\gamma^{m} = \frac{2}{\pi \sqrt{(m6)^{2} + (nq)}} \sin \frac{m\pi}{a} \times \sin \frac{n\pi\eta}{6}$$

$$\propto_{c} = \frac{Rk}{2\eta\beta} \left(2 \int_{0}^{A} A^{2} \left(\frac{m\pi}{6} \right)^{2} \sin^{2} \frac{m\pi}{a} \times dy \right)$$

$$+2\int_{0}^{b}A^{2}(\underline{m\pi})^{2}\sin^{2}\underline{n\pi}\times dx$$

$$= \frac{RkA^{2}\left[a\left(\frac{n\pi}{b}\right)^{2} + b\left(\frac{m\pi}{a}\right)^{2}\right]}{2\pi\beta}$$

$$\alpha_{c} = \frac{Rk}{27\beta} \frac{4}{\pi^{2}} \frac{ab}{(mb)^{2} + (na)^{2}} \left[\frac{n^{2}a^{3} + m^{2}b^{3}}{a^{2}b^{2}} \right]$$

$$\beta = k \sqrt{1 - (f_{c}/f)^{2}}$$

$$K_{c} = \frac{2R}{7ab\sqrt{1-(4c/4)^{2}}} \left[\frac{m^{2}b^{3}+n^{2}a^{3}}{m^{2}b^{2}+n^{2}a^{2}} \right]$$

For Tomme modes.

For TEmm modes :

at 3
$$\frac{\partial f}{\partial R} = \frac{\partial f}{\partial x} = \frac{\partial f}{\partial$$

$$=-\frac{\partial P}{\partial l}$$
 at 0

$$X_{c} = \frac{RB}{27k} \left(\frac{2}{3} \right)^{2} \frac{B^{2} \left(\frac{m\pi}{a} \right)^{2} \sin^{2} \frac{m\pi\pi}{a} dx}{a} dx$$

$$+ 2 \int B^{2} \left(\frac{n\pi}{b} \right)^{2} \sin^{2} \frac{n\pi\pi}{b} dy$$

$$K_{c} = \frac{1}{\pi^{2}} \frac{R\beta}{2\eta k} \left[\frac{4ab}{(hb)^{2} + (ha)^{2}} \right] \times$$

$$\left[\frac{a(\frac{m\pi}{a})^{2} + (\frac{n\pi}{b})^{2}b + \frac{kc^{4}}{\beta^{2}}(a+b)}{k}\right]$$

$$\frac{\beta}{k} = \sqrt{1 - (\frac{kc}{\beta})^{2}}$$

$$K_{c} = \frac{R\beta}{27k} \left[\frac{k_{c}^{\dagger}(a+b)}{\beta^{2}} + \frac{bm^{2}+an^{2}}{ab} \right] \times \frac{1}{2}$$

$$\left|\frac{\partial x^e}{\partial \ell}\right| = C \frac{\eta \pi}{6} \sin \frac{\eta \pi y}{6}$$

$$\alpha_{c} = \frac{R\beta}{2\pi k} \left[2 \int_{0}^{6} \left(\frac{h\pi}{b} \right)^{2} \sin^{2} \frac{n\pi y}{b} dy \right]$$

$$+\frac{k_{e}^{4}}{\beta^{2}}\left\{2\int_{C}^{2}\cos^{2}n\pi ydx\right.$$

$$+2\int_{D}^{C}\left(\frac{y_{n}\pi}{b}\right)^{2}\cos^{2}n\pi ydy$$

$$\kappa_{c} = \frac{R}{a \gamma \sqrt{1 - \left(\frac{f_{c}}{4}\right)^{2}}} \left[1 + \frac{2a}{b} \left(\frac{f_{c}}{4}\right)^{2}\right]$$

Now for cylindrical case;

TM case,

$$\frac{\partial \gamma_m}{\partial e} = \frac{2}{2e} \left[\int_{TI}^{E} \frac{J_n(x_{np}e/a)}{x_{np}J_{n+1}(x_{np})} \cos n\phi \right]$$

=
$$\frac{1}{a} \int_{\overline{\Pi}}^{E_n} \frac{J_n'(n_n e/a)}{J_{nn}(n_p)} \cos n\phi$$

$$K_{c} = \frac{1}{2} \frac{R}{7} \frac{k}{B} \int \frac{\varepsilon_{m}}{\tau_{T}} \left(\frac{J_{n}'(x_{np})_{A}}{J_{nm}(x_{np})} \right)^{2} \cos n\phi d\phi$$

$$\frac{J_{n}(x_{np})}{J_{n+1}(x_{np})} = \frac{-J_{n+1}(x_{np}) + \frac{N}{X_{np}}J_{n}(x_{np})}{J_{n+1}(x_{np})} = -1$$

$$\int_{0}^{2\pi} \frac{E_{n}}{2\pi T} \cos^{2}n \phi = 1 , n \neq 0$$

$$\frac{\partial f^e}{\partial l} = \frac{1}{R} \frac{\partial f^e}{\partial \phi} \Big|_{e=q}$$

$$= \frac{n}{a \sqrt{\pi \left[x_i \right]^2 - n^2 \right]}} con \phi$$

$$\frac{8-G(cont.)}{\frac{2\pi}{2}} \frac{1}{\frac{2\pi}{2}} \frac{1}{\frac{2\pi}$$

$$\Gamma = \frac{v^{r}}{v^{\lambda}}, \quad V = v^{\lambda} + \rho v^{\lambda}$$
$$= v^{\lambda}(1+\Gamma)$$

$$I = \frac{2}{\sqrt{2}} = \frac{2}{\sqrt{2}} \left(1 + \frac{1}{2} \right)$$

E quating real and imaginary parts: Pd = 11/2 (1-11/2)

$$\frac{\left[5\right]}{\left[5\right]} = \begin{bmatrix} 5^{1/2} & 5^{5/2} \\ 5^{1/2} & 5^{1/2} \end{bmatrix} \quad \Gamma = \begin{bmatrix} 1^{5/2} \\ 1^{1/2} \end{bmatrix}$$

From Egn. 8-41,

Take real part of both sides;

$$Re(2_1)I_1I_1^* + Re(2_1)I_1I_2^* +$$

Re (221) IZI, + Re (222) IZI2 = Pd H I, as I 2 ≥0 , Since Pd ≥0, Re(Z11) 20 R(Zzz) 20 } The coefficient of Ro (tiz) and Re (221) is a minimum when:

$$I_1 = I_1 e^{j\Theta}$$
, $I_2 = I_2 e^{j\Phi}$
 $I_1 I_2 e^{j(\Theta-\Phi)}$
 $\Theta-\Phi = O$ on T .

 $\frac{I_1}{I_2} = \frac{I_1}{I_2} e^{\beta(\Delta - \beta)} = \text{real NO}.$ Dividing through by IzIz, $Re(\frac{2\pi}{I_2})$ + $Re(\frac{2\pi}{I_2})$ I_1 $+ \operatorname{Re}(\tilde{\epsilon}_{i}) \underline{I}_{i} + \operatorname{Re}(\tilde{\epsilon}_{i}) \ge 0$ which is satisfied $\mathcal{E}(\frac{\Sigma_1}{\Xi_2})$ is not

a single root.

:. Re(Z12) Re(Z2) - Re(Z11) Re(Z22) 60 on Re(Z11) Re(Zz2) - Re(Z12) Re(Zu) 20 which is one of Egns. 8-72,

The Egn. for [4] can be derived by using :

$$Z_{01} = Z_{02} = 1$$

$$V_{1} = V_{1}^{1} + V_{1}^{1}$$

$$T_{1} = T_{1}^{1} + T_{1}^{1} = V_{1}^{1} - V_{1}^{1}$$

$$V_{2} = V_{2}^{1} + V_{2}^{1}$$

$$T_{2} = T_{2}^{1} + T_{2}^{1} = V_{2}^{1} - V_{2}^{1}$$

$$V_{i}^{\lambda} + V_{i}^{r} = E_{ii} (V_{i}^{\lambda} - V_{i}^{r}) + E_{i2} (V_{i}^{\lambda} - V_{i}^{r})$$
 $V_{2}^{\lambda} + V_{2}^{r} = E_{2i} (V_{i}^{\lambda} - V_{i}^{r}) + E_{22} (V_{2}^{\lambda} - V_{2}^{r})$

$$2 v_{\lambda}^{2} = \frac{v_{i}^{2} + v_{i}^{2}}{2i2} + \frac{2i1}{2i2} (v_{i}^{2} - v_{i}^{2})$$

also,

$$2 v_{k}^{2} = -\frac{\left(v_{i}^{2} + v_{i}^{2}\right)}{2 \cdot 2} - \frac{2 \cdot 2}{2 \cdot 2} \left(v_{i}^{2} - v_{i}^{2}\right) + \frac{2 \cdot 2}{2 \cdot 2} \left(v_{i}^{2} + v_{i}^{2}\right)$$

$$Naw \begin{pmatrix} v_2^{r} \\ v_2^{i} \end{pmatrix} = \begin{pmatrix} T_{11} & T_{12} \\ T_{21} & T_{22} \end{pmatrix} \begin{bmatrix} v_1^{i} \\ v_1^{r} \end{bmatrix}$$

$$V_{2}^{r} = T_{11}V_{1}^{r} + T_{12}V_{1}^{r}$$
 $V_{2}^{r} = T_{21}V_{1}^{r} + T_{22}V_{1}^{r}$

8-9 (conti)

$$= \frac{21}{5} + \frac{21}{5} \left[\frac{5}{5} + \frac{21}{5} + \frac{21}{5} + \frac{21}{5} + \frac{21}{5} \right]$$

$$= \frac{21}{5} + \frac{21}{5} \left[\frac{5}{5} + \frac{21}{5} + \frac{21}{5} + \frac{21}{5} + \frac{21}{5} + \frac{21}{5} \right]$$

$$2T_{12} = \frac{-1}{2_{12}} + \frac{2_{11}}{2_{12}} - \frac{2_{21}}{2_{12}} + \frac{2_{22}}{2_{12}} - \frac{2_{11}2_{22}}{2_{12}}$$

$$5L^{51} = \frac{515}{7} + \frac{515}{511} + \frac{515}{515} + \frac{515}{515} + \frac{515}{515}$$

For the lossless case, [2]

is pure imaginary and far

any compar no. 3, (-3) = 3

$$T_{21}^{*} = \frac{1}{21} + \frac{1}{21} (1 - 21) (22 + 1)$$

 $= -\frac{1}{2} \left(1 + \frac{1}{2} \left(1 + \frac{1}{2} \right) \left(-\frac{1}{2} + 1 \right) \right)$ $= -\frac{1}{2} \left(1 + \frac{1}{2} \left(1 + \frac{1}{2} \right) \left(\frac{1}{2} + \frac{1}{2} - 1 \right) \right)$ $= 2 T_{12}$

 $\frac{8-10}{E_t} = A(e^{-i\beta^2} + e^{i\beta^2}) = \frac{2}{2}A \cos \beta = \frac{1}{2}$ $= \frac{2}{2}A \cos \beta = \frac{1}{2}A \cos$

Boundary conditions at ==-l,

 $\overline{u_e} \times [H_t'] = \overline{J_s}$ $\overline{E_t'} \times \overline{u_e} = \overline{M_s}$

 $\overline{J}_{S} = \frac{2A}{j^{\frac{2}{6}}} \sin \beta l \, \overline{e} \quad (\overline{u}_{2} \times \overline{h} = -\overline{e})$

Ms = - ZA GODL h (ExTz) =- h

<s, 5> = \(\int_{\text{E}} \cdot \vec{T}_{\text{S}} + \vec{H} \cdot \vec{m}_{\text{S}} \ ds

= 4A2 cospl singl + 4A2 singlospl

= $\frac{4A^2}{j^{\frac{2}{6}}}$ sin $2\beta l = Self-reaction of unidorectional source$

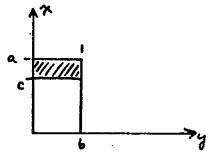
$$\frac{8-13}{\int_{0}^{C} \sin \frac{\pi x}{c} \sin \frac{\pi \pi x}{c}} dx$$

$$= \int_{0}^{C} \cos \pi x \left(\frac{1}{c} - \frac{x}{a} \right) - \cot \pi \left(\frac{1}{c} + \frac{x}{a} \right) dx$$

$$= \frac{1}{\pi \left(\frac{1}{c} - \frac{x}{a} \right)} \times \int_{0}^{C} - \frac{1}{\pi \left(\frac{1}{c} + \frac{x}{a} \right)} dx$$

$$= \frac{1}{\pi \left(\frac{1}{c} - \frac{x}{a} \right)} \times \frac{1}{\pi \left(\frac{1}{c} + \frac{x}{a}$$

$$\frac{1}{2} \frac{1}{4} + \frac{1}{2} \frac{1}{4} \frac$$



On the window,

$$E_{y}|_{z=0} = -2^{TE^{\times}} \cdot H_{x}|_{z=0}$$

$$= 2^{TE^{\times}} \cdot J_{y}|_{z=0}$$

$$= 2^{TE_{\times}} \cdot I_{y} \cdot q(x)$$

Honce Eno are given by Egn 4-73.

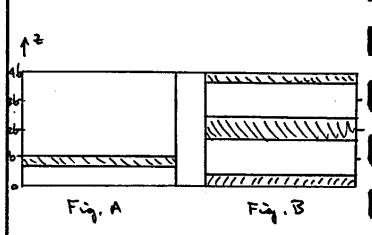
$$E_{no} = \frac{2}{a} \int_{c}^{a} f(x) \sin \frac{n\pi x}{a} dx$$

Then as on p. 416,

and
$$(Y_0)_{NO} = \frac{j 2 q(Y_0)_{10}}{\lambda_{\overline{q}} \int_{N^2} \frac{(2q)^2}{\lambda_{\overline{q}}}}$$

i. Egn 8-107 becomes :

$$\frac{Y_0}{B} = -\frac{\sum_{k=2}^{\infty} \frac{a}{\lambda_{q} \int_{N} \frac{a}{\lambda_{q}} \left(\sum_{k=1}^{\infty} \frac{a}{\lambda_{q}} \right)^{2} \left(\sum_{k=1}^{\infty} \frac{a}{\lambda_{q}} \frac{a}{\lambda_{q}} \left(\sum_{k=1}^{\infty} \frac{a}{\lambda_{q}} \frac{a}{\lambda_{q}} \right)^{2} \left(\sum_{k=1}^{\infty} \frac{a}{\lambda_{q}} \frac{a}{\lambda_{q}} \frac{a}{\lambda_{q}} \right)^{2}}{\left(\sum_{k=1}^{\infty} \frac{a}{\lambda_{q}} \frac{a}{\lambda_{q}} \frac{a}{\lambda_{q}} \right)^{2}}$$



Extend Fig. 8-22a into Fig A above and using the same techniques as in obtaining Figs. 8-22 b & C conducting plates are placed at $z = \pm nb$, n = 1, 4, ... to form a system of images. The two configurations are identical and Thus have the same equivalent arenit and

8-17 Using equivalent circuit

B Fig. 8-206 from which Egn 8-12;
was derived and changing
slightly me have:

1: N 18 + 1 18+ %

shunt capacitanse.

Replace j & in Fig 8-20 with two susceptances in parallel with ideal transformer.

Since capacitors in parrellel word, the susceptances in the above event are + That wend in

Egn. 8-120 where the signs used depend upon which way one looks into the circuit.

$$\frac{8^+}{Y_0} \approx \frac{4b^+}{\lambda y} \log \csc \frac{\pi c}{2b^+}$$

$$\frac{8}{7_0} \approx \frac{46}{\lambda_0} \log \csc \frac{\pi c}{26}$$

using transformer relationships

to find turns ratio:

$$\frac{1}{n^2} = \frac{R^+}{B^-} \Rightarrow n^2 = \frac{5}{6^+}$$

8-18 Replacing quantities in Egn. 8-127 with their duals,

$$I_{m}\left[\frac{1-\Gamma}{1+\Gamma}\right] = \frac{2B}{\sqrt{2}} \Rightarrow I_{m}\left[\frac{1+\Gamma}{1-\Gamma}\right] = \frac{\sqrt{2}}{2B}$$

to are real and ti, i to are imaginary for wal Ii, Ii Thus specializing above Egn. for B+ and B, me have ;

$$\frac{y_0^{-}}{iB^{-}} = \frac{z_0^{-} I_0^{-2}}{z_0^{-} I_0^{-2}} \frac{y_0^{+}}{iB^{+}} = \frac{z_0^{-} \hat{I}_0^{-2}}{z_0^{+} \hat{I}_0^{-2}}$$

From the dual of equivalent cet:

$$\frac{8-18 \ (contr)}{\frac{1}{6-2s}} = \frac{1}{n^2} \frac{2s^{\frac{1}{2}}}{2s^{\frac{1}{2}}}$$
 from above, $6z_0^2 = \frac{2s}{2s} \hat{T}_0^2$ $5s_0 + \frac{1}{2} = \frac{T_0}{\hat{T}_0^2}$

8-19 Since we are in a some free region, the angle operator D behaves like a true angle in the sense that $(\cos^2 0 + \sin^2 0) \mathcal{T} = \mathcal{T}$

Expressing of in the form of an integral over plane waves travelling in all directions; 7 (x, y) = \ \ A(4) e \ h(x cood + y sin 4)

where A (4) = plane wave function,

where \$ "= \$ -\$' - !! Thus they eind = in Ju(ke) J'do'e

Now, $e^{\sin D}\gamma(x,y) =$ $= \left(\frac{1}{jk\partial x} + \frac{1}{jk\partial y}\right) \int_{0}^{\infty} A(\phi') e^{ik(x\cos \phi' + y\sin \phi)} d\phi'$

$$=\int_{A(\phi')}^{2\pi} e^{ik(\times\cos\phi'+y\sin\phi')} (\cos\phi'+i\sin\phi')d\phi$$

 $\frac{1}{2\pi i} \int f(a, \phi) e^{in\phi} d\phi = \int_{a}^{b} J_{h}(ka) \left[e^{in\phi} H(a) \right]$

$$n^2 = \frac{2a}{b} \left(\frac{\tan ka}{ka} \right)^2 \sin^2 \frac{\pi}{b}$$

since a cch,

$$\therefore a^2 = \frac{29}{6} \sin^2 \frac{\pi c}{6}$$

Using the results of sec. 8-7 for

a post in a wanequille,

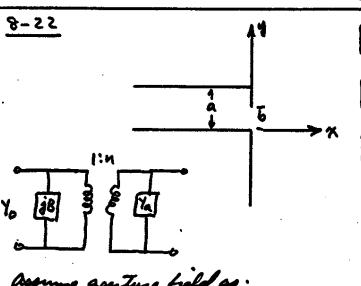
 $J_{S}^{A} = \overline{u}_{x} \frac{I}{\pi d}$

8-21 From results of prob. 7-43,

$$Y_{S} = \frac{4}{2^{2}} \geq w$$

 $7w = \langle a, a \rangle \frac{b}{a}$ because of impedance transformation by ideal transformer.

From results of pro6, 7-39, $Y_5 = \frac{4}{3^2} (73)(2) \frac{b}{2}, \quad a = \frac{\lambda}{2}$



Bosume aparture field as:

en = -1 4

ving Egn. 8-155 ;

$$V_0 = \int \int E_y^a \cdot e_0^m dy dz = -\frac{6}{\sqrt{a}}$$

Reference voltage V = Jb

$$\therefore N^2 = \frac{V^2}{V_0^2} = \frac{A}{6}$$

The relationship for the islead transformer holds as :

Ya = Ga + jBq where Ya is

altained from Fig. 4-22

and Yo is the characteristic admittance of the waveguish = WE

8-23 See "Radiation and Scattering

& Waves" by Felsen & marcurity, Prentice - Hall, 1973. 11. 237. $\frac{8-24}{4}$ The normalized mode weetos of the dominant mode is $\overline{E}_{b} = \overline{U}_{x} \frac{2}{4bc} \sin \frac{\pi y}{c} \sin \frac{\pi z}{c}$

From Egn, 8-162,

The current on the probe is assumed to be:

$$J_{x}^{a} = I \delta(y-6') \delta(y-c') \times cd$$

$$a_{o} = \iiint \overline{E}_{o} \cdot \overline{J}^{a} d\gamma$$

$$= \int_{0}^{d} \frac{2T}{\sqrt{\epsilon a 6c}} \sin \frac{\pi c}{6} \sin \frac{\pi c}{c} d\gamma$$

$$\frac{a_0}{T} = \frac{2d}{\sqrt{\epsilon abc}} \sin \frac{\pi b}{b} \sin \frac{\pi c}{c}$$

when c' << c', $sin \frac{\pi c'}{c} \approx \frac{\pi c'}{c}$

$$\approx \frac{2\pi A}{\sqrt{\epsilon abc}} \sin(\frac{\pi b}{b})$$
 where $A = c'd$

and R, L, and C B the agricult are given by Egn. 8-181.

 $\frac{8-25}{60}$ The dominant mode for the circular cavity is a TMOID mode and is given by: $E_2 = \frac{k^2}{6WS} J_o\left(\frac{x_0!}{a}\right)$

The normalization constant is obtained from.

$$\frac{8-25(cont.)}{2}$$

$$= \frac{\pi k^4 b a^2 J_1^2(x_{01})}{\omega^2 \xi}$$

Where $x_{01} = 2.405$.

The assumed current distribution on the probe is:

$$J_{2} = I \frac{\sin k(d-e)}{\sin kd} \delta(e-c) = \epsilon cd$$

27d

Es is given by Egs. Prob. 8-25.

$$\int_{0}^{d} \frac{\sin k(d-2)}{\sin kd} dz = \frac{\tan \frac{kd}{2}}{k}$$

$$\frac{a_o}{I} = \frac{J_o(x_{oi}\frac{c}{a})}{a\sqrt{\epsilon\pi b}} \frac{t_{oi}\frac{kd}{\epsilon}}{J_i(x_{oi})}$$

8-27 The assumed current distribution is :

$$J_{2} = IS(e-c) = c d$$

$$E > d$$

Es is given by Pads, 8-25.

$$\frac{a_0}{T} = \frac{J_0(x_{0i} \frac{c}{a}) d}{a \sqrt{\pi \epsilon b} J_i(x_{0i})}$$

when c = a,

$$J_{o}\left(x_{o_{1}}\frac{c}{a}\right)=J_{o}\left[x_{o_{1}}\left(1+\frac{c-a}{a}\right)\right]$$

Naw expanding in a Taylors series when c=a,

$$J_{o}\left(x_{o_{1}}\frac{c}{a}\right)=J_{o}\left(x_{o_{1}}\right)+x_{o_{1}}\frac{\left(c-a\right)}{a}J_{o}^{\prime}\left(x_{o_{1}}\right)$$

$$= x_{01} \frac{(c-a)}{a} J_{\bullet}'(x_{01})$$

$$\frac{1}{1} = \frac{d(a-c)x_{01}}{a^{2}\sqrt{\pi\epsilon b}} \approx \frac{Ax_{01}}{a^{2}\sqrt{\pi\epsilon b}}$$

8-28 Replace loop with magnetic current element

KI = - j w AId(x-d) S(y-b') S(z-c') Ty

where A is the area of loop = c'd

From Egn. z-96,

Hy = 16 Eo Sin Ty coo TE
7 162+C2 Sin Ty coo TE

to find Eo for normalization:

W = ZWm = M SSIHIZdr = 1

an E = 27/62+c2

then normalized His;

Hy = 27

Now R = 2 WMA Sin TT6 COOTTC'

in the cavity, $k=\beta=\omega\sqrt{\mu\xi}$.

: WH - B

but $\beta = \frac{\pi}{c}$ as in section 2-8.

also as c'ecc, cos Tre! -> 1

:. R = ZTTA sin Tb'

when the I has been suppressed the Lang Calculations

 $H_{\phi} = \frac{x_{ol}}{a} J_{i} \left(\frac{x_{ol}}{a} \right)$

Hornalizing Hop we have,

Ho = Xul \ \frac{w^2 \xi}{a \sqrt{\pi k^4 ba^2}} \frac{\J_1(\times_0 \end{a})}{\J_1(\times_0)}

Ke = - j wy A S(e-c) up

:. R = XulA J(Xola)

allower Ti (Xol)

for c≈a,

 $R = \frac{x_{0}, A}{a^{2} \sqrt{\pi b s}}$

8-30 The magnetic field for dominant mode is given by: $H_4 = \frac{1}{r} \hat{S}_1(2.744 \frac{r}{a}) \sin \theta$

For normalization constant,

W = MSSIHIZAT = 1 = 8TTH (1.14)

H4 = \frac{3k}{811/4 (1.14)} \frac{\hat{3}_1 (2.744 \frac{r}{a}) \sin \text{6}}{r}

where ka = 2,744

 $\therefore H_{\phi} = \frac{.535}{r J_{\alpha} \mu} \widehat{J}_{1}(2.744 \frac{r}{\alpha}) \sin \alpha$