

**Full Name:** \_\_\_\_\_  
**EEL 4750 / EEE 5502 (Fall 2021) – HW #8** **Due: 4:00 PM ET, Nov. 02, 2021**

### **Concept Questions 8**

**Question #1:** Complete the Canvas questions here: <https://ufl.instructure.com/courses/437179/assignments/4812584>

## Theory Questions 8

**Question #1:** Consider the analog biquad filter, an analog filter design that is very common because it is flexible (can be low-pass, high-pass, band-pass, or stop-band) and easy to manufacture. A biquad filter is most generally defined by:

$$H(s) = \frac{b_2 s^2 + b_1 s + b_0}{s^2 + a_1 s + a_0} = \frac{K (s^2 + 2(\Omega_z/Q_z)s + \Omega_z^2)}{s^2 + 2(\Omega_p/Q_p)s + \Omega_p^2}$$

- (a) Show that the system

$$\text{poles are equal to: } s = \Omega_p \left( -1/Q_p \pm \sqrt{1/Q_p^2 - 1} \right)$$

$$\text{zeros are equal to: } s = \Omega_z \left( -1/Q_z \pm \sqrt{1/Q_z^2 - 1} \right)$$

Note that with the Laplace transform,  $s = j\omega + \sigma$  whereas the z-transform  $z = e^{j\omega + \sigma}$ . In both cases, the  $\omega$  values correspond to angular frequency and the  $\sigma$  values correspond to the impulse response decay rate.

- (b) When  $\Omega_z = 0$ ,  $Q_p = \infty$ , and  $\Omega_p > 0$ , at what frequencies (as a function of  $\Omega_p$ ) are approximately in the analog filter's passband (in radians/s)?
- (c) When  $\Omega_z = 0$ ,  $0 \leq Q_p \leq 1$ , and  $\Omega_p > 0$ , at what frequencies (as a function of  $\Omega_p$ ) are approximately in the analog filter's passband (in radians/s)?
- (d) When  $\Omega_z = 0$ ,  $Q_p = \sqrt{2}$ , and  $\Omega_p > 0$ , at what frequencies (as a function of  $\Omega_p$ ) are approximately in the analog filter's passband (in radians/s)?
- (e) Use the difference approximation method with a sampling rate of  $f_s = 1$  to design a digital IIR filter based on the analog biquad filter. Under this approximation, show that

$$\text{poles are equal to: } z = \frac{1}{1 + \Omega_p \left( 1/Q_p \pm \sqrt{1/Q_p^2 - 1} \right)}$$

$$\text{zeros are equal to: } z = \frac{1}{1 + \Omega_z \left( 1/Q_z \pm \sqrt{1/Q_z^2 - 1} \right)}$$

- (f) When  $\Omega_z = 0$ ,  $Q_p = \infty$ , and  $\Omega_p$  approaches infinity (i.e.,  $\Omega_p \gg 1$  <sup>Footnote:1</sup>), frequencies (as a function of  $\Omega_p$ ) are approximately in the digital filter's passband (in radians/s)?
- (g) When  $\Omega_z = 0$ ,  $0 \leq Q_p \leq 1$ , and  $\Omega_p$  approaches infinity (i.e.,  $\Omega_p \gg 1$ ), frequencies (as a function of  $\Omega_p$ ) are approximately in the digital filter's passband (in radians/s)?
- (h) When  $\Omega_z = 0$  and  $\Omega_p$  approaches zero (i.e.,  $\Omega_p \ll 1$  <sup>Footnote: 2</sup>), what frequencies (as a function of  $\Omega_p$ ) are approximately in the analog filter's passband (in radians/s)?
- (i) Based on these results, what type(s) of filter(s) do you think the first difference method here would be most effective at approximating? Why?

---

<sup>1</sup>Hence,  $1 + \Omega_p \approx \Omega_p$

<sup>2</sup>Hence,  $1 + \Omega_p \approx 1$

## Implementation Questions 8

**Question #1:** Consider a **negative** ideal low-pass, continuous-time filter defined by

$$H(\Omega) = u(\Omega + \Omega_c) - u(\Omega - \Omega_c)$$

and its DTFT equivalent frequency response

$$H(\omega) = u(\omega + \Omega_c/f_s) - u(\omega - \Omega_c/f_s) \quad (\text{this repeats for every } 2\pi \text{ shift})$$

For all of the components below, use a sampling rate of  $f_s = 1$  Hz and a cut-off frequency of  $\Omega_c = \pi/4$  rad/s. For every designed filter, ensure that the DC gain (i.e., at  $\omega = 0$ ) is equal to 1.

- (a) Use the windowing method with a Hann window<sup>3</sup> to design an FIR, linear phase low-pass filter of length  $N = 50$  from  $H(\Omega)$ .<sup>4</sup> Plot the pole-zero plot for the filter and plot the magnitude and phase response.
- (b) Use the frequency sampling method to design an FIR, linear phase low-pass filter of length  $N = 50$  from  $H(\Omega)$ . Plot the pole-zero plot for the filter and plot the magnitude and phase response.
- (c) Use the difference approximation method to design an IIR filter with  $K = 1$ ,  $\Omega_z = 20$ , and  $Q_p = 5$ ,  $Q_z = 2$ , and  $\Omega_p = 0.8$  for the bi-quad filter from theory questions. Plot the pole-zero plot for the filter and plot the magnitude and phase response.
- (d) Using filters in (a), (b) and (c) above, filter the following square waves defined in the template:  

```
sqwaves{1} = square(2*pi*0.4*t);  
sqwaves{2} = square(2*pi*1*t);
```

---

<sup>3</sup>A Hann window is defined by  $w[n] = 0.5(1 - \cos(2\pi n/(N-1)))$  for  $0 \leq n \leq N-1$ , where  $N$  is the window length

<sup>4</sup>Note: You may need to use a `sinc` function. I am providing a `sinc_func` function in the template since MATLAB's `sinc` function uses a different definition, as discussed in class.