Error dynamics and control of systems with constraints

Contact-aware Control, Lecture 6

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Trajectory tracking; desired trajectory

Trajectory tracking is a control problem that says:

Trajectory tracking

Find such control law $\mathbf{u} = \mathbf{u}(\mathbf{x}, t)$ that solution of the dynamical system $\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}, t)$ converges to the desired trajectory $\mathbf{x}^* = \mathbf{x}^*(t)$.

For mechanical systems specifically we can write it as:

Trajectory tracking for mechanical systems

Find such control law $\mathbf{u} = \mathbf{u}(\mathbf{q}, \dot{\mathbf{q}}, t)$ that solution of the dynamical system $\mathbf{H}\ddot{\mathbf{q}} + \mathbf{c} = \mathbf{T}\mathbf{u}$ converges to the desired trajectory $\mathbf{q}^* = \mathbf{q}^*(t)$.

Desired trajectory and constraints

Assume your system is subject to constraints $\mathbf{g}(\mathbf{q}) = 0$, and you have desired trajectory $\mathbf{q}^* = \mathbf{q}^*(t)$. Then, unless $\mathbf{g}(\mathbf{q}^*(t)) = 0$, the desired trajectory is not valid.

You can find first two time derivatives of the desired trajectory: $\dot{\mathbf{q}}^*(t)$ and $\ddot{\mathbf{q}}^*(t)$. Defining $\mathbf{F} = \frac{\partial \mathbf{g}}{\partial \mathbf{q}}$ we can write first and second derivative of the constraint as: $\dot{\mathbf{g}}(\mathbf{q}) = \mathbf{F}\dot{\mathbf{q}}$ and $\ddot{\mathbf{g}}(\mathbf{q}) = \mathbf{F}\ddot{\mathbf{q}} + \dot{\mathbf{F}}\dot{\mathbf{q}}$. Therefor we can write implied conditions on the desired trajectory:

$$\dot{\mathbf{g}}(\mathbf{q}^*) = \mathbf{F}\dot{\mathbf{q}}^* = 0 \tag{1}$$

$$\ddot{\mathbf{g}}(\mathbf{q}^*) = \mathbf{F}\ddot{\mathbf{q}}^* + \dot{\mathbf{F}}\dot{\mathbf{q}}^* = 0 \tag{2}$$

We can rewrite equations of dynamics in the normal form:

$$\ddot{\mathbf{q}} = \mathbf{H}^{-1}(\mathbf{T}\mathbf{u} - \mathbf{c}) \tag{3}$$

Let us define *control error* \mathbf{e} as follows:

$$\mathbf{e} = \mathbf{q}^* - \mathbf{q} \tag{4}$$

Then we can find its second derivative as: $\ddot{\mathbf{e}} = \ddot{\mathbf{q}}^* - \ddot{\mathbf{q}}$:

$$\ddot{\mathbf{e}} = \ddot{\mathbf{q}}^* - \mathbf{H}^{-1}(\mathbf{T}\mathbf{u} - \mathbf{c}) \tag{5}$$

If error dynamics is stable, it means the error will approach zero as the time approaches infinity. Is a good thing.

We can decide that we want error dynamics to have this form:

$$\ddot{\mathbf{e}} + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e} = 0 \tag{6}$$

where \mathbf{K}_d and \mathbf{K}_p are diagonal positive-definite matrices. Equation (6) is stable. So, if we achieve that our error dynamics takes this form, we make it stable.

Let us use (5) to re-write (6):

$$\ddot{\mathbf{q}}^* - \mathbf{H}^{-1}(\mathbf{T}\mathbf{u} - \mathbf{c}) + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e} = 0$$
 (7)

So, we have:

$$\ddot{\mathbf{q}}^* - \mathbf{H}^{-1}(\mathbf{T}\mathbf{u} - \mathbf{c}) + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e} = 0$$
 (8)

Now we can multiply it by **H** (because it is invertible, so it's null space is trivial and we do not annihilate any part of the equation):

$$\mathbf{H}(\ddot{\mathbf{q}}^* + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e}) - (\mathbf{T}\mathbf{u} - \mathbf{c}) = 0$$
(9)

...and then express \mathbf{u} out:

$$\mathbf{u} = \mathbf{T}^{+} (\mathbf{H} (\ddot{\mathbf{q}}^* + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e}) + \mathbf{c})$$
 (10)

This is called a *computed torque controller* (CTC), and it assumes that $(\mathbf{H}(\ddot{\mathbf{q}}^* + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e}) + \mathbf{c})$ is in the column space of \mathbf{T} .

Thus CTC has the form: $\mathbf{u} = \mathbf{T}^+ (\mathbf{H}(\ddot{\mathbf{q}}^* + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e}) + \mathbf{c})$

We can separate feedback part \mathbf{u}_{FB} and feedforward part \mathbf{u}_{FF} :

$$\mathbf{u} = \mathbf{u}_{FB} + \mathbf{u}_{FF} \tag{11}$$

$$\mathbf{u}_{FB} = \mathbf{T}^{+} \mathbf{H} (\mathbf{K}_{d} \dot{\mathbf{e}} + \mathbf{K}_{p} \mathbf{e}) \tag{12}$$

$$\mathbf{u}_{FF} = \mathbf{T}^{+}(\mathbf{H}\ddot{\mathbf{q}}^{*} + \mathbf{c}) \tag{13}$$

Notice that feedback part is just a PD (proportional-derivative) controller with varying gains, while the feedforward part is just \mathbf{u} expressed out of the robot's dynamics $\mathbf{H\ddot{q}} + \mathbf{c} = \mathbf{Tu}$; the latter - finding \mathbf{u} directly from the dynamics - is called *inverse dynamics*.

Dynamical system with constraints can be written as:

$$\begin{cases} \mathbf{H}\ddot{\mathbf{q}} + \mathbf{c} = \mathbf{T}\mathbf{u} + \mathbf{F}^{\top}\lambda \\ \mathbf{F}\ddot{\mathbf{q}} + \dot{\mathbf{F}}\dot{\mathbf{q}} = 0 \end{cases}$$
(14)

How do we apply the ideas about stable error dynamics here?

One naive approach is to define a new variable $\mathbf{v} = \mathbf{T}\mathbf{u} + \mathbf{F}^{\top}\lambda$ and rewrite the first equation in the system (14) as:

$$\mathbf{H\ddot{q}} + \mathbf{c} = \mathbf{v} \tag{15}$$

Here it seems we can apply CTC directly. But we need to study how (and when) it will work. We are considering equation $\mathbf{H}\ddot{\mathbf{q}} + \mathbf{c} = \mathbf{v}$. CTC for this case will take form:

$$\mathbf{v} = \mathbf{H}(\ddot{\mathbf{q}}^* + \mathbf{K}_d \dot{\mathbf{e}} + \mathbf{K}_p \mathbf{e}) + \mathbf{c}$$
 (16)

But remember that $\mathbf{v} = \mathbf{T}\mathbf{u} + \mathbf{F}^{\top}\lambda$. We can always try to do our best to find such \mathbf{u} that it holds, but what about λ ?

On the value of λ

We should always keep in mind that λ is uniquely determined for any given **u**. So, while we can't assign λ arbitrarily, as long as we assigned **u**, we did in fact determined what value λ will take.

Part 3: calculating control input expressing reaction forces out

We remember from the previous lecture, that if we define $\mathbf{M} = \begin{bmatrix} \mathbf{H} & -\mathbf{F}^{\top} \\ \mathbf{F} & \mathbf{0} \end{bmatrix}$, then assuming \mathbf{L} is a left inverse of \mathbf{M} , then we can write expressions for both $\ddot{\mathbf{q}}$ and λ in terms of its components:

$$\begin{cases}
\ddot{\mathbf{q}} = \mathbf{L}_{11}(\mathbf{T}\mathbf{u} - \mathbf{c}) - \mathbf{L}_{12}\dot{\mathbf{F}}\dot{\mathbf{q}} \\
\lambda = \mathbf{L}_{21}(\mathbf{T}\mathbf{u} - \mathbf{c}) - \mathbf{L}_{22}\dot{\mathbf{F}}\dot{\mathbf{q}}
\end{cases}$$
(17)

We can try to use this naive way of finding relation between λ and the control input \mathbf{u} , or one of the many more sophisticated ones. The idea would be to substitute the expression into the equation $\mathbf{v} = \mathbf{T}\mathbf{u} + \mathbf{F}^{\top}\lambda$, giving, in this case:

$$\mathbf{v} = \mathbf{T}\mathbf{u} + \mathbf{F}^{\top} (\mathbf{L}_{21}(\mathbf{T}\mathbf{u} - \mathbf{c}) - \mathbf{L}_{22}\dot{\mathbf{F}}\dot{\mathbf{q}})$$
(18)

Part 4: calculating control input, accelerations and reaction forces simultaneously

Alternatively, we can try to simultaneously calculate generalized accelerations $\ddot{\mathbf{q}}$, control inputs \mathbf{u} and reaction forces λ . This allows us to bring in the constraint equation $\mathbf{F}\ddot{\mathbf{q}} + \dot{\mathbf{F}}\dot{\mathbf{q}} = 0$:

$$\begin{cases}
\mathbf{H}\ddot{\mathbf{q}} + \mathbf{c} = \mathbf{v} \\
\mathbf{T}\mathbf{u} + \mathbf{F}^{\top}\lambda = \mathbf{v} \\
\mathbf{F}\ddot{\mathbf{q}} + \dot{\mathbf{F}}\dot{\mathbf{q}} = 0
\end{cases}$$
(19)

where unknowns are $\ddot{\mathbf{q}}$, \mathbf{u} and λ . This is solved as a simple linear system:

$$\begin{bmatrix} \ddot{\mathbf{q}} \\ \mathbf{u} \\ \lambda \end{bmatrix} = \begin{bmatrix} \mathbf{H} & 0 & 0 \\ 0 & \mathbf{T} & \mathbf{F}^{\top} \\ \mathbf{F} & 0 & 0 \end{bmatrix}^{+} \begin{bmatrix} \mathbf{v} - \mathbf{c} \\ \mathbf{v} \\ -\dot{\mathbf{F}}\dot{\mathbf{q}} \end{bmatrix}$$
(20)

Algebra recap

Fundamental subspaces

Column space

All possible outputs of a linear operator \mathbf{A} are called *column* space of \mathbf{A} .

Null space

Null space of \mathbf{A} is the set of all vectors \mathbf{x} that \mathbf{A} maps to 0

Now we can find all solutions to the system of equations $\mathbf{A}\mathbf{x} = \mathbf{0}$ by using functions that generate an orthonormal *basis* in the null space of \mathbf{A} . In MATLAB it is function null().

In MATLAB it can be constructed by calling function orth(). Both orth() and null() (as well as rank() and pinv()) simply call svd() and perform minimal computations on the resulting decomposition. You can check it by typing open orth in MATLAB command window.

Algebra recap

We can project any vector onto a subspace using a projector.

Definition 1

For linear space $\mathcal{L} \subset \mathbb{R}^n$, an orthogonal projector **P** onto it has properties:

- $\forall \mathbf{x} \in \mathbb{R}^n, \, \mathbf{P}\mathbf{x} \in \mathcal{L}$
- $\forall \mathbf{x} \in \mathcal{L}, \ \mathbf{P}\mathbf{x} = \mathbf{x}$
- $\forall \mathbf{y} \in \mathcal{L}, \, \mathbf{y}^{\top} (\mathbf{I} \mathbf{P}) \mathbf{x} = 0$

Projector \mathbf{P}_c onto the column space of an operator \mathbf{A} can be found as:

$$\mathbf{P}_c = \mathbf{A}\mathbf{A}^+ \tag{21}$$

Part 5: feasibility conditions

Last expression suggests a simple feasibility condition for the existence of the control input that will generate the desired \mathbf{v} , namely that the left-hand-side vector of the linear system should lie in the column space of the matrix of the linear system:

$$\left(\mathbf{I} - \begin{bmatrix} \mathbf{H} & 0 & 0 \\ 0 & \mathbf{T} & \mathbf{F}^{\top} \\ \mathbf{F} & 0 & 0 \end{bmatrix} \begin{bmatrix} \mathbf{H} & 0 & 0 \\ 0 & \mathbf{T} & \mathbf{F}^{\top} \\ \mathbf{F} & 0 & 0 \end{bmatrix}^{+} \right) \begin{bmatrix} \mathbf{v} - \mathbf{c} \\ \mathbf{v} \\ -\dot{\mathbf{F}}\dot{\mathbf{q}} \end{bmatrix} = 0 \qquad (22)$$

Part 6: feasibility conditions, simpler

We can make a simple set of necessary conditions. Remember that $\mathbf{v} = \mathbf{T}\mathbf{u} + \mathbf{F}^{\top}\lambda$. Therefore, vector \mathbf{v} should lie in the column space of the matrix $\begin{bmatrix} \mathbf{T} & \mathbf{F}^{\top} \end{bmatrix}$:

$$\left(\mathbf{I} - \begin{bmatrix} \mathbf{T} & \mathbf{F}^{\top} \end{bmatrix} \begin{bmatrix} \mathbf{T} & \mathbf{F}^{\top} \end{bmatrix}^{+} \right) \mathbf{v} = 0$$
 (23)

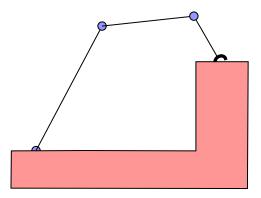
Read more

You can read more at:

■ Slotine, J.J.E. and Li, W., 1987. On the adaptive control of robot manipulators. The international journal of robotics research, 6(3), pp.49-59 - learn more about error dynamics and similar techniques, is very interesting!

Homework

Write a tracking controller for this robot (description of the robot is given in the previous lectures).



Lecture slides are available via Moodle.

 $You\ can\ help\ improve\ these\ slides\ at:$ github.com/SergeiSa/Contact-Aware-Control-Slides-Fall-2020

Check Moodle for additional links, videos, textbook suggestions.