

# A Guided Tour of CML, the Coded Modulation Library

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# \*Outline

1. CML overview
  - What is it? How to set it up and get started?
2. Uncoded Modulation
  - Simulate uncoded BPSK and QAM in AWGN and Rayleigh fading
3. Coded Modulation
  - Simulate a turbo code from UMTS 25.212
  - \*Specify a custom eIRA LDPC parity check matrix
4. Ergodic (Shannon) capacity analysis
  - Determine the modulation constrained capacity of BPSK and QAM
5. EXIT analysis
  - Determine the EXIT characteristic for BPSK in AWGN for a particular LDPC degree distribution

# Outline

## 6. Outage analysis

- Determine the outage probability over block fading channels
- Determine the outage probability of finite-length codes

## 7. Digital Network Coding

- Describe CML implementation of digital network coding
- Simulate error rate of digital network coding for non-channel-coded 4-FSK in Rayleigh fading

## 8. The internals of CML

## 9. Throughput calculation

- Convert BLER to throughput for hybrid-ARQ



# What is CML?

- CML is an open source toolbox for simulating capacity approaching codes in MATLAB.
- Available for free at the **Iterative Solutions** website:
  - [www.iterativesolutions.com](http://www.iterativesolutions.com)
- Runs in MATLAB, but uses c-mex for efficiency.
- First release was in Oct. 2005.
  - Used code that has been developed since 1996.

# Features

- Simulation of BICM (bit interleaved coded modulation)
  - Turbo, LDPC, or convolutional codes.
  - User-specified eIRA LDPC parity check matrices.
  - PSK, QAM, FSK modulation.
  - BICM-ID: Iterative demodulation and decoding.
- Generation of ergodic capacity curves
  - BICM/CM constrained modulation.
- EXIT analysis
  - Generate LDPC EXIT characteristics for all CML detector types.
- Information outage probability
  - Block fading channels.
  - Blocklength-constrained channels (AWGN or fading)

# Features

- Digital network coding
  - Error-rate of digital network coding in the two-way relay channel
  - Noncoherent FSK modulation
- Calculation of throughput of hybrid-ARQ.

# Supported Standards

- Binary turbo codes:
  - UMTS/3GPP, including HSDPA and LTE.
  - cdma2000/3GPP2.
  - CCSDS.
- Duobinary turbo codes:
  - DVB-RCS.
  - WiMAX IEEE 802.16.
- LDPC codes:
  - DVB-S2.
  - Mobile WiMAX IEEE 802.16e.

# Simulation Data is Valuable

- CML saves simulation state frequently
  - parameter called “save\_rate” can be tuned to desired value.
- CML can be stopped at any time.
  - Intentionally: Hit CTRL-C within matlab.
  - Unintentionally: Power failure, reboot, etc.
- CML automatically resumes simulation
  - If a simulation is run again, it will pickup where it left off.
  - Can reset simulation by setting “reset=1”.
  - SNR points can be added or deleted prior to restarting.
- Simulations can be made more confident by requesting additional trials prior to restarting.
  - The new results will be added to the old ones.



# Compiled Mode

- A flag called “compiled\_mode” can be used to run CML independently of matlab.
- CML must first be compiled using the matlab compiler.
- Advantages:
  - Can run on machines without matlab.
  - Can run on a grid computer.

# WebCML

- WebCML is a new initiative sponsored by NASA and NSF.
- Idea is to upload simulation parameters to a website and hit a “simulate” button.
  - Simulation begins on the webserver.
  - The webserver will divide the simulation into multiple jobs which are sent to a grid computer.
- Results can be retrieved while simulation is running and once it has completed.
- The grid is comprised of ordinary desktop computers.
  - The grid compute engine is a screen saver.
    - Kicks in only when computer is idle.
  - Users of WebCML are encouraged to donate their organizations computers to the grid.

# Getting Started with CML

## ■ Download

- [www.iterativesolutions.com/download.htm](http://www.iterativesolutions.com/download.htm)
- Decompress downloaded archive
  - Root directory will be `./cml`

## ■ About simulation databases

- A large database of previous simulation results is available.
- Unzip each database and place each extracted directory into the `./cml/output` directory

## ■ About C-mex files.

- C-mex files are compiled for PC computers.
- For unix and mac computers, must compile.
  - Within matlab, cd to `./cml/source` and type “make”.

# Starting and Interacting with CML

- Launch MATLAB
- Cd to the <CMLROOT> directory
- Type “CmlStartup”
  - This sets up paths and determines the version of MATLAB.
- To run CML, only two functions are needed:
  - CmlSimulate
    - Runs one or more simulations.
    - Simulation parameters are stored in .m scripts.
    - Input arguments tell CML which simulation(s) to run.
  - CmlPlot
    - Plots the results of one or more simulations.

# Scenario Files and the SimParam Structure

- The parameters associated with a set of simulations is stored in a scenario file.
  - Located in one of two directories
    - ./cml/scenarios for publicly available scenarios
    - ./cml/localscenarios for personal user scenarios
    - Other directories could be used if they are on the matlab path.
  - .m extension.
- Exercise
  - Edit the example scenario file: UncodedScenarios.m
- The main content of the scenario file is a structure called `sim_param`
  - `Sim_param` is an array.
  - Each element of the array is called a *record* and corresponds to a single distinct simulation.

# Common Parameters

- List of all parameters can be found in:
  - ./cml/mat/DefineStructures.m
  - ./cml/documentation/readme.pdf
- Default values are in the DefineStructures.m file
- Some parameters can be changed between runs, others cannot.
  - `sim_param_changeable`
  - `sim_param_unchangeable`

# Dissecting the SimParam Structure:

## The simulation type

### ■ `sim_param(record).sim_type =`

- ‘uncoded’
  - BER and SER of uncoded modulation
- ‘coded’
  - BER and FER of coded modulation
- ‘capacity’
  - The Shannon capacity under modulation constraints.
- ‘exit’
  - EXIT characteristic of selected detector and parameterized LDPC code
- ‘outage’
  - The information outage probability of block fading channels
  - Assumes codewords are infinite in length
- ‘bloutage’
  - Information outage probability in AWGN or ergodic/block fading channels
  - Takes into account length of the code.

# Dissecting the SimParam Structure:

## The simulation type

- `sim_param(record).sim_type =`
  - ‘throughput’
    - By using FER curves, determines throughput of hybrid ARQ
    - This is an example of an *analysis* function ... no simulation involved.



# Lesser Used Simulation Types

- `sim_param(record).sim_type =`
  - ‘bwcapacity’
    - Shannon capacity of CPFSK under bandwidth constraints.
  - ‘minSNRvsB’
    - Capacity limit of CPFSK as a function of bandwidth

# Parameters Common to All Simulations

## ■ `sim_param(record)`.

- `comment = {string}`
  - Text, can be anything.
- `legend = {string}`
  - What to put in figure caption
- `linetype = {string}`
  - Color, type, and marker of line. Uses syntax from matlab “plot”.
- `filename = {string}`
  - Where to save the results of the simulation
  - Once filename is changed, any parameter can be changed.
- `reset = {0,1}` with default of 0
  - Indication to resume “0” or restart “1” simulation when run again.
  - If reset = 1, any parameter may be changed.

# Specifying the Simulation

## ■ `sim_param(record).`

### – `SNR = {vector}`

- Vector containing SNR points in dB
- Can add or remove SNR points between runs

### – `SNR_type = {'Eb/No in dB' or 'Es/No in dB'}`

- For some simulation types, only one option is supported.
- E.g. for *capacity* simulations, it must be Es/No

### – `save_rate = {scalar integer}`

- An integer specifying how often the state of the simulation is saved
- Number of trials between saves.
- Simulation echoes a period '.' every time it saves.

# Specifying the Simulation (cont'd)

## ■ `sim_param(record)`.

### – `max_trials = {vector}`

- A vector of integers, one for each SNR point
- Tells simulation maximum number of trials to run per point.

### – `max_frame_errors = {vector}`

- Also a vector of integers, one for each SNR point.
- Tells simulation maximum number of frame errors to log per point.
- Simulation echoes a 'x' every time it logs a frame error.

### – `minBER = {scalar}`

- Simulation halts once this BER is reached

### – `minFER = {scalar}`

- Simulation halts once this FER is reached.

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# Specifying Modulation

## ■ `sim_param(record).`

### – `modulation = {string}`

- Specifies the modulation type
- May be 'BPSK', 'QPSK', 'QAM', 'PSK', 'APSK', 'HEX', or 'FSK'
- 'HSDPA' used to indicate QPSK and QAM used in HSDPA.
- All but FSK are 2 dimensional modulations
  - Uses a complex scalar value for each symbol.
- Default is 'BPSK'
- New (version 1.9 and above): Can also be set to "custom".

### – `mod_order = {integer scalar}`

- Number of points in the constellation.
- Power of 2.
- Default is 2.
- In some cases, M=0 is used to indicate an unconstrained Gaussian input.

### – `S_matrix = {complex vector}`

- Only used for "custom" modulation type.
- A vector of length "mod\_order" containing the values of the symbols in the signal set S.

# Specifying Modulation

## ■ `sim_param(record)`.

### – `mapping = {integer vector}`

- A vector of length M specifying how data bits are mapped to symbols.
- Vector contains the integers 0 through M-1 exactly once.
- ith element of vector is the set of bits associated with the ith symbol.
- Alternatively, can be a string describing the modulation, like 'gray' or 'sp'
- Default is 'gray'

### – `framesize = {integer scalar}`

- The number of symbols per Monte Carlo trial
- For coded systems, this is number of bits per codeword

### – `demod_type = {integer scalar}`

- A flag indicating how to implement the demodulator
  - 0 = log-MAP (approximated linearly)
  - 1 = max-log-MAP
  - 2 = constant-log-MAP
  - 3 and 4 other implementations of log-MAP
- Max-log-MAP is fastest.
- Does not effect the uncoded error rate.
  - However, effects coded performance

# M-ary Complex Modulation

- $\mu = \log_2 M$  bits are mapped to the symbol  $\mathbf{x}_k$ , which is chosen from the set  $S = \{\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_M\}$ 
  - The symbol is multidimensional.
  - 2-D Examples: QPSK, M-PSK, QAM, APSK, HEX
    - These 2-D signals take on complex values.
  - M-D Example: FSK
    - FSK signals are represented by the M-dimensional complex vector  $\mathbf{X}$ .
- The signal  $y = h\mathbf{x}_k + n$  is received
  - $h$  is a complex fading coefficient (scalar valued).
  - $n$  is complex-valued AWGN noise sample
  - More generally (FSK),  $\mathbf{Y} = h \mathbf{X} + \mathbf{N}$ 
    - Flat-fading: All FSK tones multiplied by the same fading coefficient  $h$ .
- Modulation implementation in CML
  - The complex signal set  $S$  is created with the **CreateConstellation** function.
  - Modulation is performed using the **Modulate** function.



# Log-likelihood of Received Symbols

- Let  $p(\mathbf{x}_k|\mathbf{y})$  denote the probability that signal  $\mathbf{x}_k \in S$  was transmitted given that  $\mathbf{y}$  was received.
- Let  $f(\mathbf{x}_k|\mathbf{y}) = K p(\mathbf{x}_k|\mathbf{y})$ , where  $K$  is any multiplicative term that is constant for all  $\mathbf{x}_k$ .
- When all symbols are equally likely,  $f(\mathbf{x}_k|\mathbf{y}) \propto f(\mathbf{y}|\mathbf{x}_k)$
- For each signal in  $S$ , the receiver computes  $f(\mathbf{y}|\mathbf{x}_k)$ 
  - This function depends on the modulation, channel, and receiver.
  - Implemented by the **Demod2D** and **DemodFSK** functions, which actually computes  $\log f(\mathbf{y}|\mathbf{x}_k)$ .
- Assuming that all symbols are equally likely, the most likely symbol  $\mathbf{x}_k$  is found by making a hard decision on  $f(\mathbf{y}|\mathbf{x}_k)$  or  $\log f(\mathbf{y}|\mathbf{x}_k)$ .

# Example: QAM over AWGN.

- Let  $y = x + n$ , where  $n$  is complex i.i.d.  $N(0, N_0/2)$  and the average energy per symbol is  $E[|x|^2] = E_s$

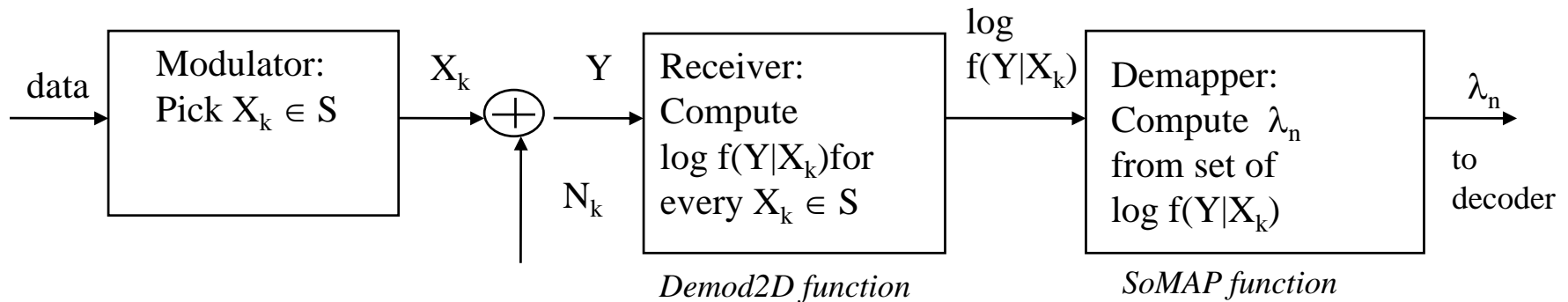
$$p(y | x_k) = \frac{1}{2\sigma^2} \exp\left\{-\frac{|y - x_k|^2}{2\sigma^2}\right\}$$

$$f(y | x_k) = \exp\left\{-\frac{|y - x_k|^2}{2\sigma^2}\right\}$$

$$\log f(y | x_k) = \frac{-|y - x_k|^2}{2\sigma^2}$$

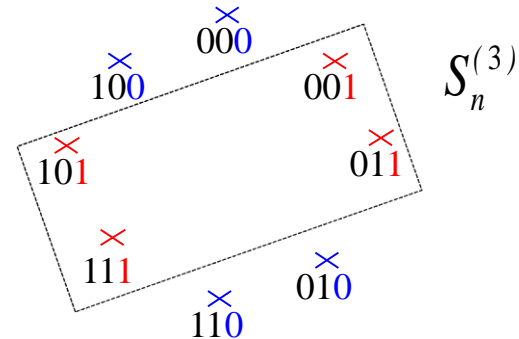
$$= \frac{-E_s |y - x_k|^2}{N_0}$$

# Converting symbol likelihoods to bit LLR



- The symbol likelihoods must be transformed into bit log-likelihood ratios (LLRs):

$$\lambda_n = \log \frac{P[d_n = 1]}{P[d_n = 0]} = \log \frac{\sum_{X_k \in S_n^{(1)}} f(Y / X_k)}{\sum_{X_k \in S_n^{(0)}} f(Y / X_k)}$$

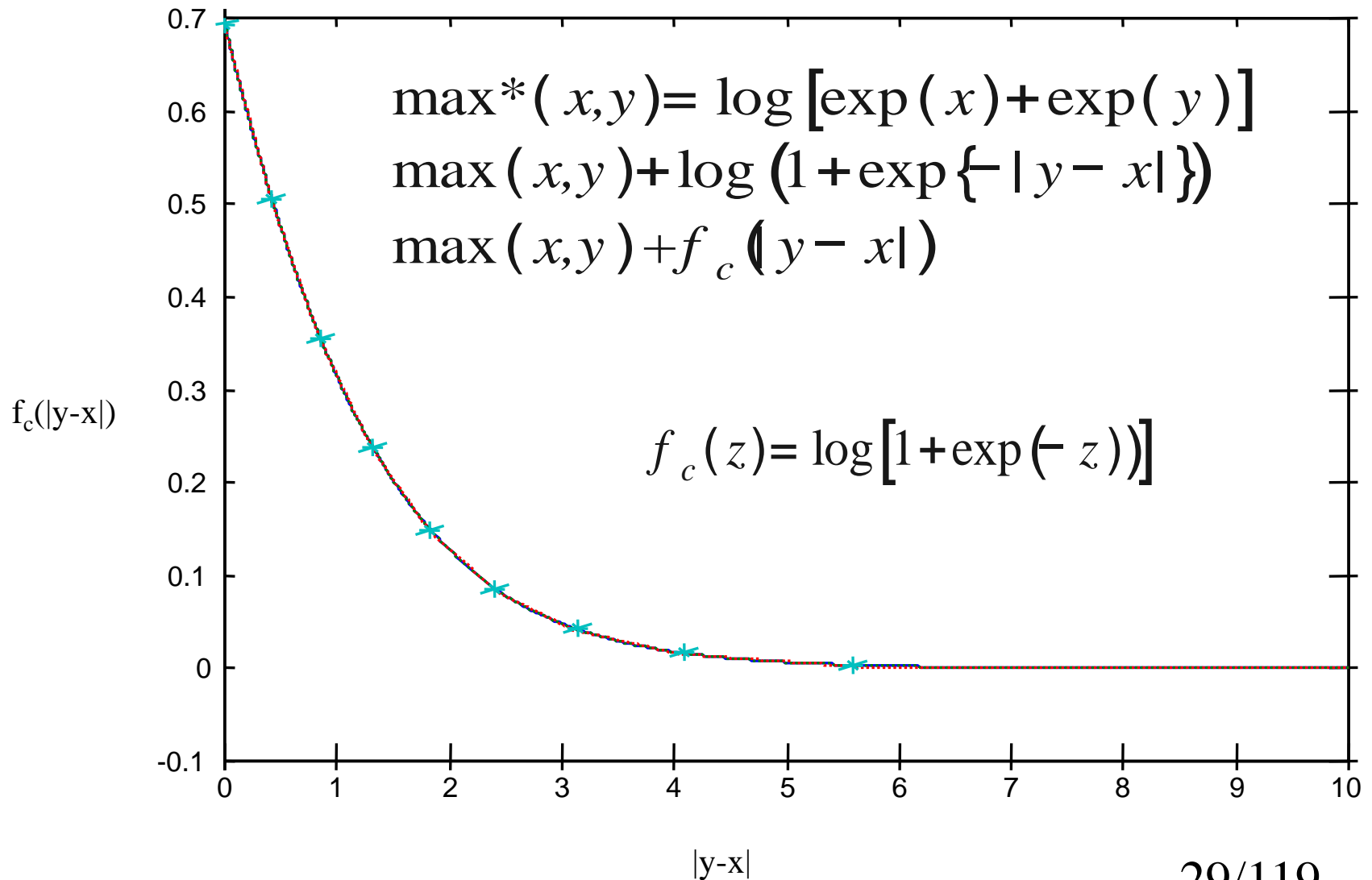


- Where  $S_n^{(1)}$  represents the set of symbols whose nth bit is a 1.
- and  $S_n^{(0)}$  is the set of symbols whose nth bit is a 0.

# Log-domain Implementation

$$\begin{aligned}\lambda_n &= \log \frac{\sum_{X_k \in S_n^{(1)}} f(Y | X_k)}{\sum_{X_k \in S_n^{(0)}} f(Y | X_k)} \\&= \log \sum_{X_k \in S_n^{(1)}} f(Y | X_k) - \log \sum_{X_k \in S_n^{(0)}} f(Y | X_k) \\&= \max_{X_k \in S_n^{(1)}} * \{\log f(Y | X_k)\} - \max_{X_k \in S_n^{(0)}} * \{\log f(Y | X_k)\} \quad \begin{array}{l} \text{log-MAP} \\ \text{demod\_type} = 0 \end{array} \\&\approx \max_{X_k \in S_n^{(1)}} \{\log f(Y | X_k)\} - \max_{X_k \in S_n^{(0)}} \{\log f(Y | X_k)\} \quad \begin{array}{l} \text{max-log-MAP} \\ \text{demod\_type} = 1 \end{array}\end{aligned}$$

# The max\* function



# FSK-Specific Parameters

## ■ `sim_param(record).`

### – `h = {scalar}`

- The modulation index
- `h=1` is orthogonal

### – `csi_flag = {integer scalar}`

0 = coherent (only available when `h=1`)

1 = noncoherent w/ perfect amplitudes

2 = noncoherent without amplitude estimates

# Specifying the Channel

- `sim_param(record).`
  - `channel = {'AWGN', 'Rayleigh', 'block'}`
    - 'Rayleigh' is “fully-interleaved” Rayleigh fading
    - 'block' is for coded simulation type only
  - `blocks_per_frame = {scalar integer}`
    - For block channel only.
    - Number of independent blocks per frame.
    - Block length is `framesize/blocks_per_frame`
  - `bicm = {integer scalar}`
    - 0 do not interleave bits prior to modulation
    - 1 interleave bits prior to modulation (default)
    - 2 interleave and perform iterative demodulation/decoding
    - This option is irrelevant unless a channel code is used

# Exercises

- Create and run the following simulations:
  - BPSK in AWGN
  - 64QAM with gray labeling in AWGN
  - 64QAM with gray labeling in Rayleigh fading
- Choices that need to be made?
  - Framesize?
  - Save\_rate?
  - Min\_BER?
  - Min\_frame\_errors?
  - Demod\_type?
- Plot all the results on the same figure.



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- **Simulate a turbo code from UMTS 25.212**
- **\*Specify a custom eIRA LDPC parity check matrix**

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- Determine the modulation constrained capacity of BPSK and QAM

## 5. EXIT analysis

- Determine the EXIT characteristic for BPSK in AWGN for a particular LDPC degree distribution

# Coded Systems: Code Configuration

- Only for `sim_param(record).sim_type = 'coded'`
- `sim_param(record).code_configuration = {scalar int}`
  - 0 = Convolutional
  - 1 = binary turbo code (PCCC)
  - 2 = LDPC
  - 3 = HSDPA turbo code
  - 4 = UMTS turbo code with rate matching
  - 5 = WiMAX duobinary tailbiting turbo code (CTC)
  - 6 = DVB-RCS duobinary tailbiting turbo code

# Convolutional Codes

- Only rate  $1/n$  mother codes supported.
  - Can puncture to higher rate.
- Code is always terminated by a tail.
  - Can puncture out the tail.
- `sim_param(record)`.
  - `g1 = {binary matrix}`
    - Example: (133,171) code from Proakis
      - $g1 = \begin{bmatrix} 1 & 0 & 1 & 1 & 0 & 1 & 1 \\ & 1 & 1 & 1 & 1 & 0 & 0 & 1 \end{bmatrix}$ ;
    - Constraint length = number of columns
    - Rate  $1/n$  where  $n$  is number of rows.
  - `nsc_flag1 = {scalar integer}`
    - 0 for RSC
    - 1 for NSC
- Can handle cyclic block codes as a rate 1 terminated RSC code

# Convolutional Codes: Decoding Algorithms

- `sim_param(record).decoder_type = {integer scalar}`
  - negative value for Viterbi algorithm
  - 0 = log-MAP (approximated linearly)
  - 1 = max-log-MAP
  - 2 = constant-log-MAP
  - 3 and 4 other implementations of log-MAP
- Decodes over entire trellis (no sliding window traceback)

# Punctured Convolutional Codes

## ■ sim\_param(record).

### – pun\_pattern1 = {binary matrix}

- Puncturing pattern
- n rows
- arbitrary number of columns (depends on puncture period)
- 1 means keep bit, 0 puncture it.
- number greater than 1 is number of times to repeat bit.

### – tail\_pattern1 = {binary matrix}

- tail can have its own puncturing pattern.

# Turbo Codes

- `sim_param(record).`
  - Parameters for first constituent code
    - `g1`
    - `nsc_flag1`
    - `pun_pattern1`
    - `tail_pattern1`
  - Parameters for second constituent code
    - `g2`
    - `nsc_flag2`
    - `pun_pattern2`
    - `tail_pattern2`

# Turbo Codes (cont'd)

## ■ `sim_param(record).`

### – `code_interleaver = {string}`

- A string containing the command used to generate the interleaver.
- Examples include:
  - “CreateUmtsInterleaver(5114)” % UMTS interleaver.
  - “CreateLTEInterleaver(6144)” % LTS interleaver.
  - “CreateCCSDSInterleaver(8920)” % CCSDS interleaver.
  - “randperm(40)-1” % a random interleaver of length 40.
  - Can replace above lengths with other valid lengths.

### – `decoder_type = {integer scalar}`

- Same options as for convolutional codes (except no Viterbi allowed).

### – `max_iterations = {integer scalar}`

- Number of decoder iterations.
- Decoder will automatically halt once codeword is correct.

### – `plot_iterations = {integer scalar}`

- Which iterations to plot, in addition to `max_iterations`

# UMTS Rate Matching

- `sim_param(record)`
  - `framesize = {integer scalar}`
    - number of data bits
  - `code_bits_per_frame = {integer scalar}`
    - number of code bits
- When `code_configuration = 4`, automatically determines rate matching parameters according to UMTS (25.212)



# HSDPA Specific Parameters

- `sim_param(record).`
  - `N_IR = {integer scalar}`
    - Size of the virtual IR buffer
  - `X_set = {integer vector}`
    - Sequence of redundancy versions (one value per ARQ transmission)
  - `P = {integer scalar}`
    - Number of physical channels per turbo codeword
- Examples from HSET-6 TS 25.101
  - `N_IR = 9600`
  - QPSK
    - `framesize = 6438`
    - `X_set = [0 2 5 6]`
    - `P = 5` (i.e. 10 physical channels used for 2 turbo codewords)
  - 16-QAM
    - `framesize = 9377`
    - `X_set = [6 2 1 5]`
    - `P = 4` (i.e. 8 physical channels used for 2 turbo codewords)

# LDPC

## ■ sim\_param(record).

### – parity\_check\_matrix = {string}

- A string used to generate the parity check matrix

### – decoder\_type

- 0 Sum-product (default)
- 1 Min-sum
- 2 Approximate-min-star

### – max\_iterations

- Number of decoder iterations.
- Decoder will automatically halt once codeword is correct.

### – plot\_iterations

- Which iterations to plot, in addition to max\_iterations

# LDPC

## ■ Specifying a Parity Check Matrix

- Several methods available
  - DVB-S2 type via CML function '`InitializeDVBS2()`'
  - Flat-text `alist` file [1]
  - CML-native `H_rows`, `H_cols` format stored as `.mat` file
- All parity check matrix data files must be stored in `<CMLROOT>/data/ldpc`

Example:

`<CMLROOT>/data/ldpc/test_ldpc_hmat.alist`

[1] <http://www.inference.phy.cam.ac.uk/mackay/codes/alist.html>

# LDPC

## ■ Specifying a Parity Check Matrix

### Function InitializeDVBS2()

- Generates parity check matrix conforming to DVB-S2 standard

- Example

CML Scenario: Ldpchmat

Record: 1

# LDPC

## ■ Specifying a Parity Check Matrix

# File-based specification

- Consider an eIRA LDPC code having rate  $K/N$
- The parity check matrix has form

[illegible]

- User-specified parity check matrices define the  $N-K \times K$  sub-matrix denoted by (1)

# LDPC

## ■ Specifying a Parity Check Matrix

### H\_rows, H\_cols format

- Parity check matrix stored as .mat file
  - File contains two variables, H\_rows and H\_cols
    - H\_rows:  $N-K \times R$  double matrix
    - H\_cols:  $K \times C$  double matrix
- where  $R$  denotes the maximum weight of all rows and  $C$  denotes the maximum weight of all columns
- First dimension of H\_rows and H\_cols denotes a particular parity check matrix row or column, respectively.
  - Each entry of H\_rows and H\_cols denotes the location of a “1” for row or column specified by the first dimension with entry “0” denoting the absence of a “1”.

# LDPC

## ■ Specifying a Parity Check Matrix

### H\_rows, H\_cols format

#### – Trivial Example

$$H = \left[ \begin{array}{cccc|cccc} 1 & 1 & & 1 & 1 & 1 & & \\ 1 & & & 1 & & 1 & 1 & \\ & 1 & 1 & & & & 1 & 1 \\ & & 1 & & & & & 1 \end{array} \right]$$

$$H\_rows = \left[ \begin{array}{ccc} 1 & 2 & 4 \\ 1 & 4 & 0 \\ 2 & 3 & 0 \\ 3 & 0 & 0 \end{array} \right]$$

$$H\_cols = \left[ \begin{array}{cc} 1 & 2 \\ 1 & 3 \\ 3 & 4 \\ 1 & 2 \end{array} \right]$$

# LDPC

## ■ Specifying a Parity Check Matrix alist format

- Parity check matrix stored as .alist file in
  - <CMLROOT>/data/ldpc
- Parity check matrix must obey eIRA constraint
- For a full description of the alist format, see  
<http://www.inference.phy.cam.ac.uk/mackay/codes/alist.html>



# LDPC

## ■ Specifying a Parity Check Matrix

### Examples

#### CML Scenario: LdpcHmat

- Record 1: InitializeDVBS2()
- Record 2: alist
  - Data file: <CMLROOT>/data/ldpc/test\_ldpc\_hmat.alist
- Record 3: H\_rows, H\_cols
  - Data file: <CMLROOT>/data/ldpc/test\_ldpc\_hmat.mat

# Block Fading

- For coded simulations, block fading is supported.
- `Sim_param(record).channel = 'block'`
- `Sim_param(record).blocks_per_frame`
  - The number of independent blocks per frame
- Example, HSDPA with independent retransmissions
  - `blocks_per_frame = length(X_set );`

# Exercises

## ■ Simulate

- A convolutional code with  $g=(7,5)$  over AWGN with BPSK
- The same convolutional code punctured to rate  $3/4$ .
- The UMTS turbo code with 16-QAM
  - Unpunctured w/ 640 input bits
  - Punctured to force the rate to be  $1/2$ .
  - Compare log-MAP and max-log-MAP
- HSDPA
  - HSET-6
  - Quasi-static block fading

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## 4. Ergodic (Shannon) capacity analysis

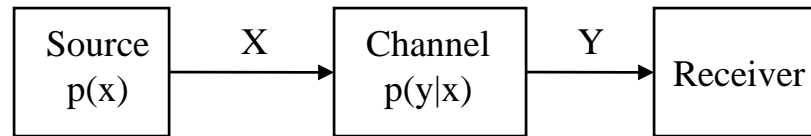
- **Determine the modulation constrained capacity of BPSK and QAM**

## 5. EXIT analysis

- Determine the EXIT characteristic for BPSK in AWGN for a particular LDPC degree distribution

# Noisy Channel Coding Theorem (Shannon 1948)

- Consider a memoryless channel with input  $X$  and output  $Y$



- The channel is completely characterized by  $p(x,y)$

- The **capacity**  $C$  of the channel is

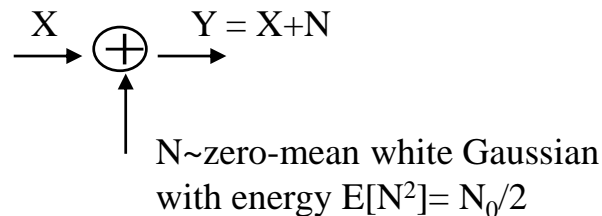
$$C = \max_{p(x)} I(X; Y) = \max_{p(x)} \left\{ \iint p(x, y) \log \frac{p(x, y)}{p(x)p(y)} dx dy \right\}$$

- where  $I(X,Y)$  is the (average) **mutual information** between  $X$  and  $Y$ .
- The channel capacity is an upper bound on **information rate**  $r$ .
  - There exists a code of rate  $r < C$  that achieves reliable communications.
  - “Reliable” means an arbitrarily small error probability.

# Capacity of the AWGN Channel with Unconstrained Input

- Consider the one-dimensional AWGN channel

The input  $X$  is drawn  
from *any* distribution  
with average energy  
 $E[X^2] = E_s$



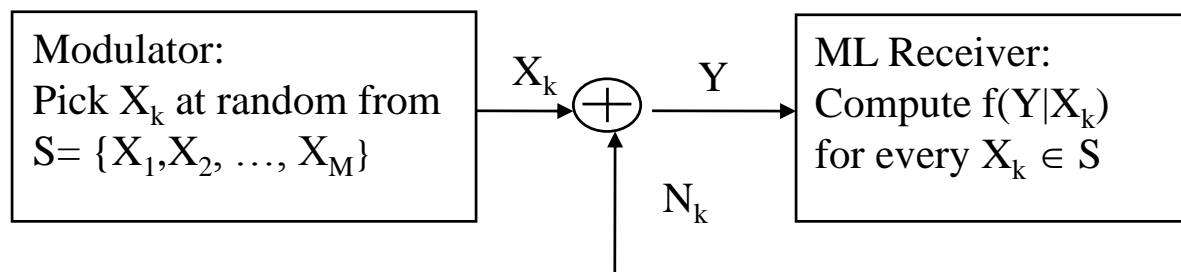
- The capacity is

$$I(X; Y) = \frac{1}{2} \log_2 \left( \frac{2E_s}{N_o} + 1 \right) \quad \text{bits per channel use}$$

- The  $X$  that attains capacity is Gaussian distributed.
  - Strictly speaking, Gaussian  $X$  is not practical.

# Capacity of the AWGN Channel with a Modulation-Constrained Input

- Suppose  $X$  is drawn with equal probability from the finite set  $S = \{X_1, X_2, \dots, X_M\}$



- where  $f(Y|X_k) = \kappa p(Y|X_k)$  for any  $\kappa$  common to all  $X_k$
- Since  $p(x)$  is now fixed

$$\{I(X;Y)\} = I(X;Y)$$

- i.e. calculating capacity boils down to calculating mutual info.

# Entropy and Conditional Entropy

- Mutual information can be expressed as:

$$I(X; Y) = H(X) - H(X / Y)$$

- Where the **entropy** of X is

$$H(X) = E[h(X)] = \int p(x)h(x)dx$$

$$\text{where } h(x) = \log \frac{1}{p(x)} = -\log p(x)$$

self-information

- And the **conditional entropy** of X given Y is

$$H(X / Y) = E[h(X / Y)] = \iint p(x, y)h(x / y)dx dy$$

$$\text{where } h(x | y) = -\log p(x | y)$$



# Calculating Modulation-Constrained Capacity

- To calculate:

$$I(X; Y) = H(X) - H(X | Y)$$

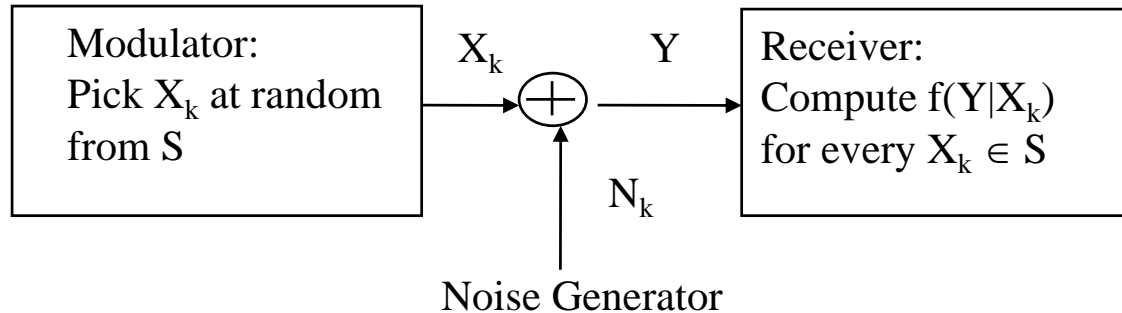
- We first need to compute  $H(X)$

$$H(X) = E[h(X)]$$

$$\begin{aligned} &= E\left[\log \frac{1}{p(X)}\right] \\ &= E[\log M] \quad \text{---} \quad p(X) = \frac{1}{M} \\ &= \log M \end{aligned}$$

- Next, we need to compute  $H(X|Y) = E[h(X|Y)]$ 
  - This is the “hard” part.
  - In some cases, it can be done through numerical integration.
  - Instead, let's use Monte Carlo simulation to compute it.

# Step 1: Obtain $p(x|y)$ from $f(y|x)$



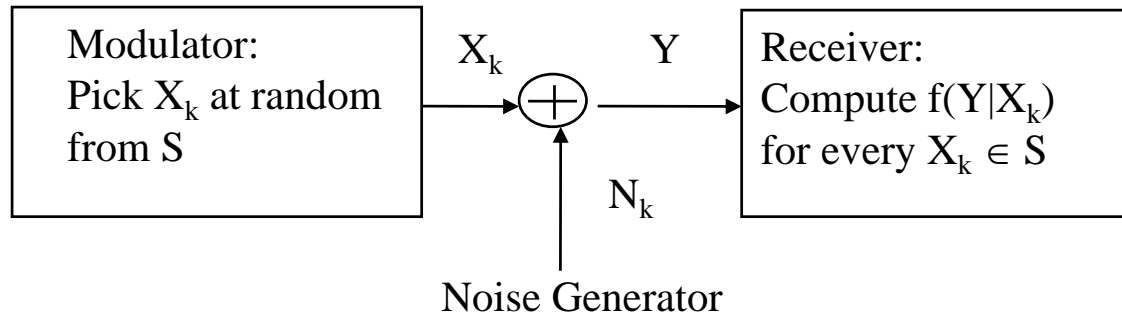
■ Since

$$\sum_{x' \in S} p(x'|y) = 1$$

■ We can get  $p(x|y)$  from

$$p(x|y) = \frac{p(x|y)}{\sum_{x' \in S} p(x'|y)} = \frac{\frac{p(y|x)p(x)}{p(y)}}{\sum_{x' \in S} \frac{p(y|x')p(x')}{p(y)}} = \frac{f(y|x)}{\sum_{x' \in S} f(y|x')}$$

## Step 2: Calculate $h(x|y)$



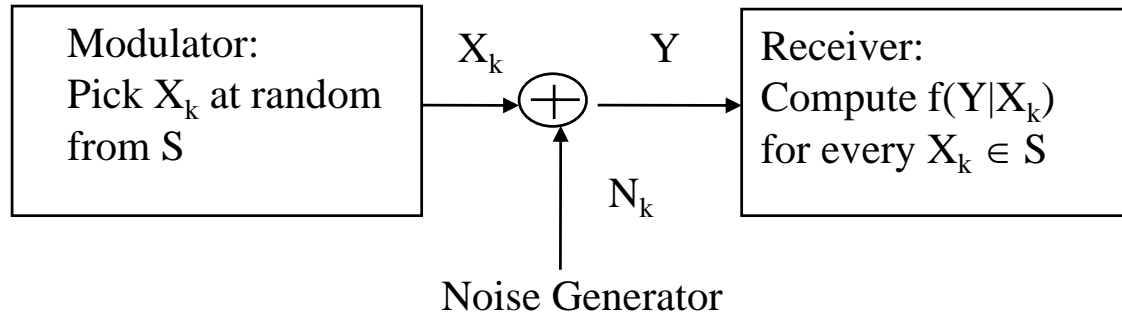
- Given a value of  $x$  and  $y$  (from the simulation) compute

$$p(x|y) = \frac{f(y|x)}{\sum_{x' \in S} f(y|x')}$$

- Then compute

$$h(x/y) = -\log p(x/y) = -\log f(y/x) + \log \sum_{x' \in S} f(y/x')$$

## Step 3: Calculating $H(X|Y)$

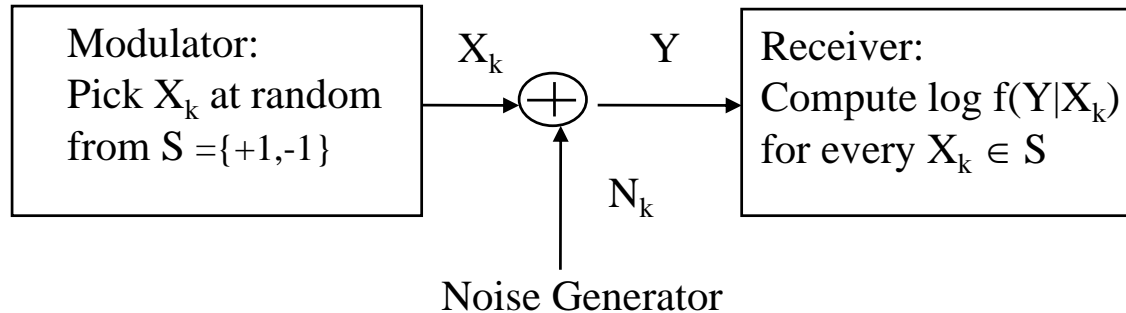


- Since: 
$$H(X / Y) = E[h(X / Y)] = \iint p(x, y) h(x / y) dx dy$$
- Because the simulation is ergodic,  $H(X|Y)$  can be found by taking the sample mean:

$$\frac{1}{N} \sum_{n=1}^N h(X^{(n)} / Y^{(n)})$$

- where  $(X^{(n)}, Y^{(n)})$  is the  $n^{\text{th}}$  realization of the random pair  $(X, Y)$ .
  - i.e. the result of the  $n^{\text{th}}$  simulation trial.

# Example: BPSK

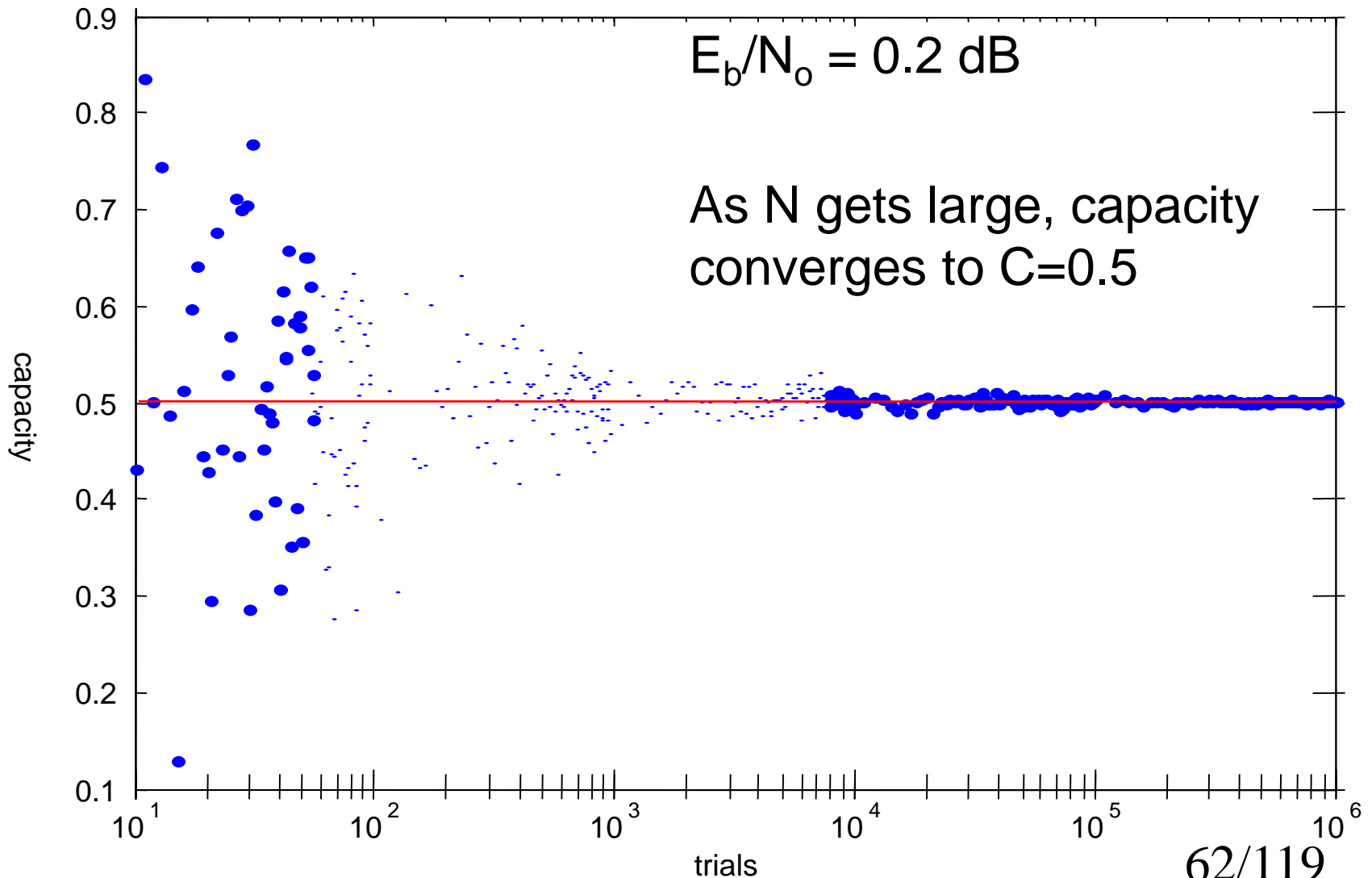


- Suppose that  $S = \{+1, -1\}$  and  $N$  has variance  $N_0/2E_s$

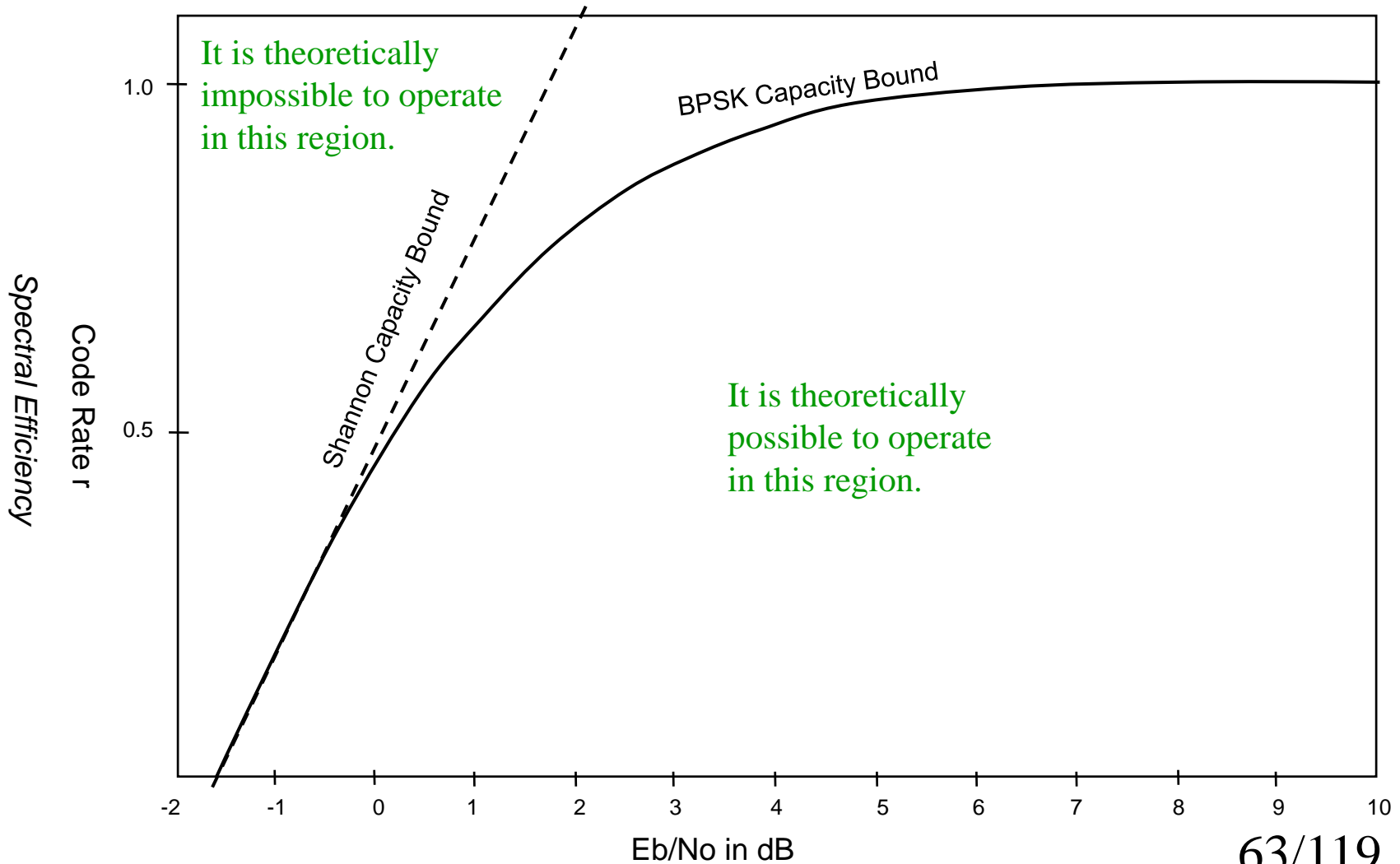
- Then:

$$\log f(y|x) = -\frac{E_s}{N_o} \|y - x\|^2$$

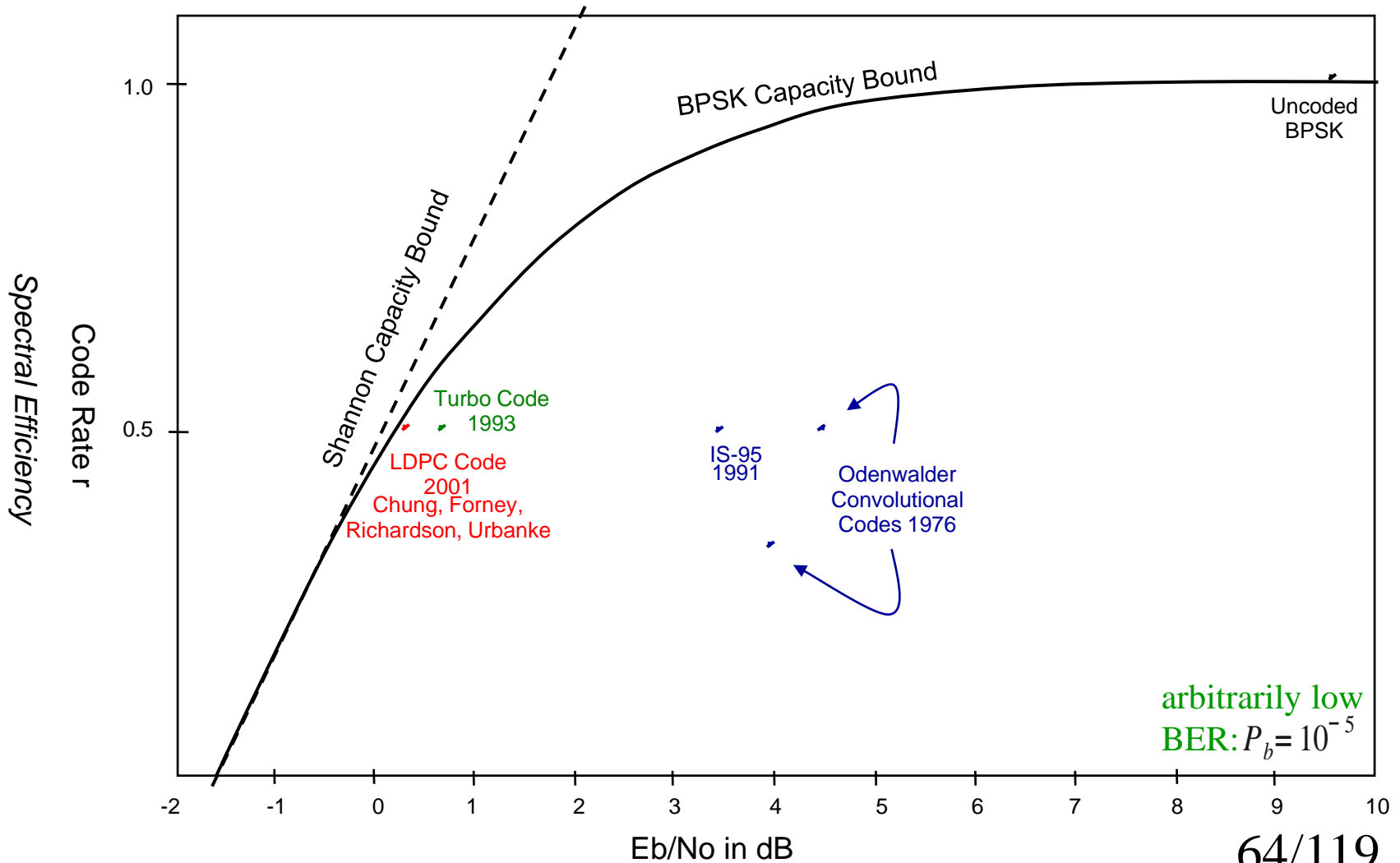
# BPSK Capacity as a Function of Number of Simulation Trials



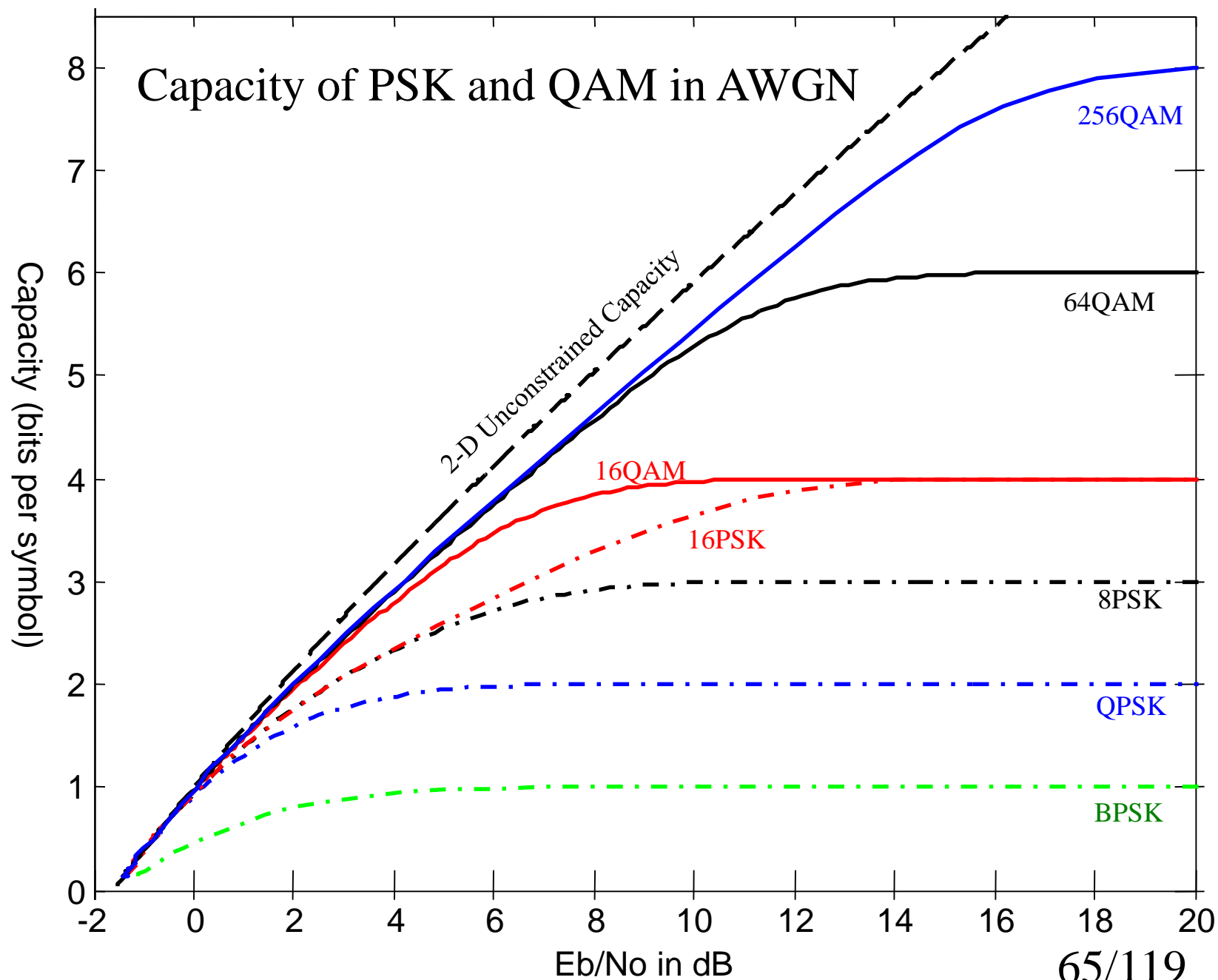
# Unconstrained vs. BPSK Constrained Capacity



# Power Efficiency of Standard Binary Channel Codes







# Capacity of Noncoherent Orthogonal FSK in AWGN

W. E. Stark, "Capacity and cutoff rate of noncoherent FSK with nonselective Rician fading," *IEEE Trans. Commun.*, Nov. 1985.

M.C. Valenti and S. Cheng, "Iterative demodulation and decoding of turbo coded M-ary noncoherent orthogonal modulation," *IEEE JSAC*, 2005.

Minimum  $E_b/N_0$  (in dB)

Noncoherent combining penalty

M=2

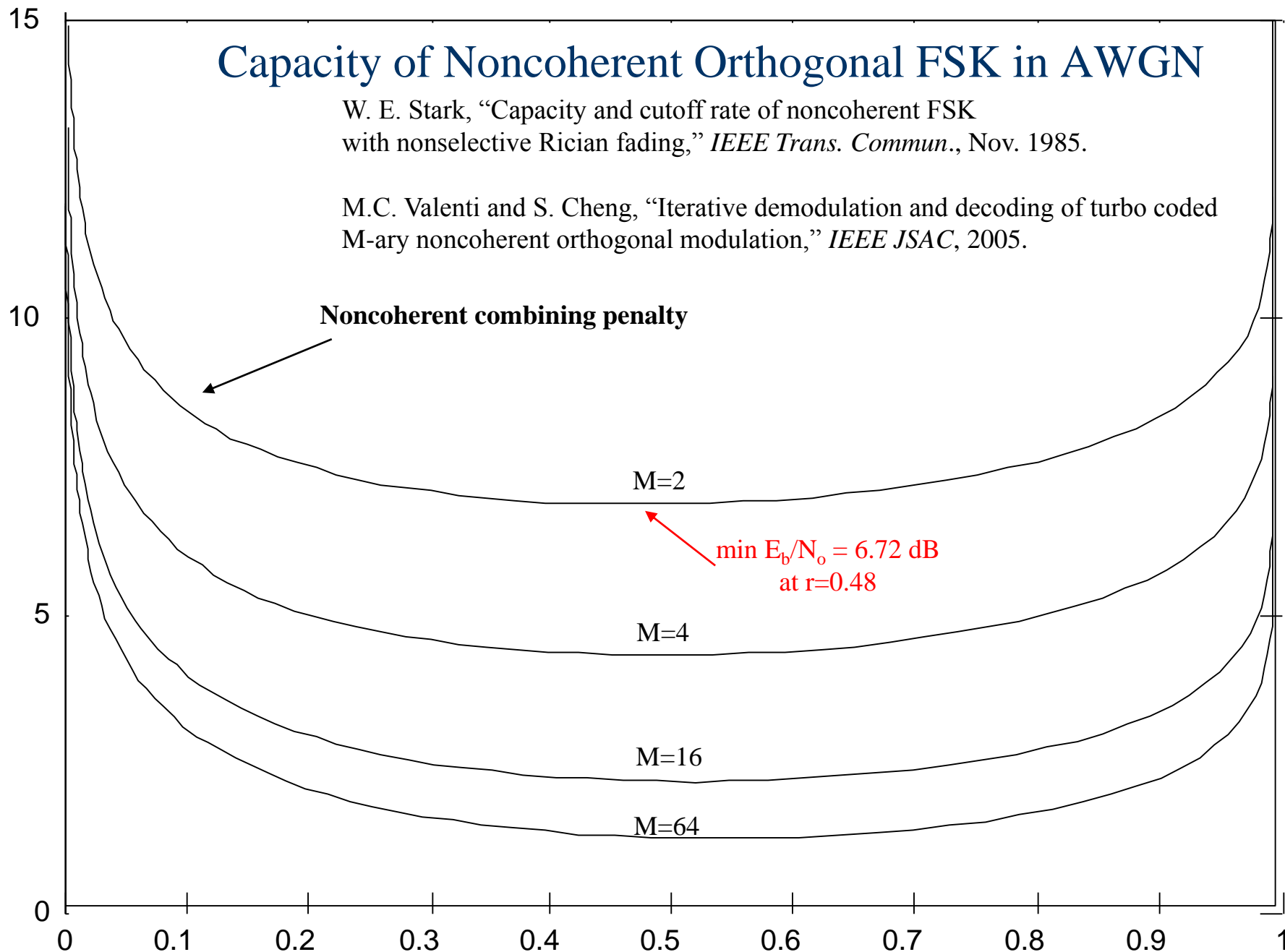
$\min E_b/N_0 = 6.72 \text{ dB}$   
at  $r=0.48$

M=4

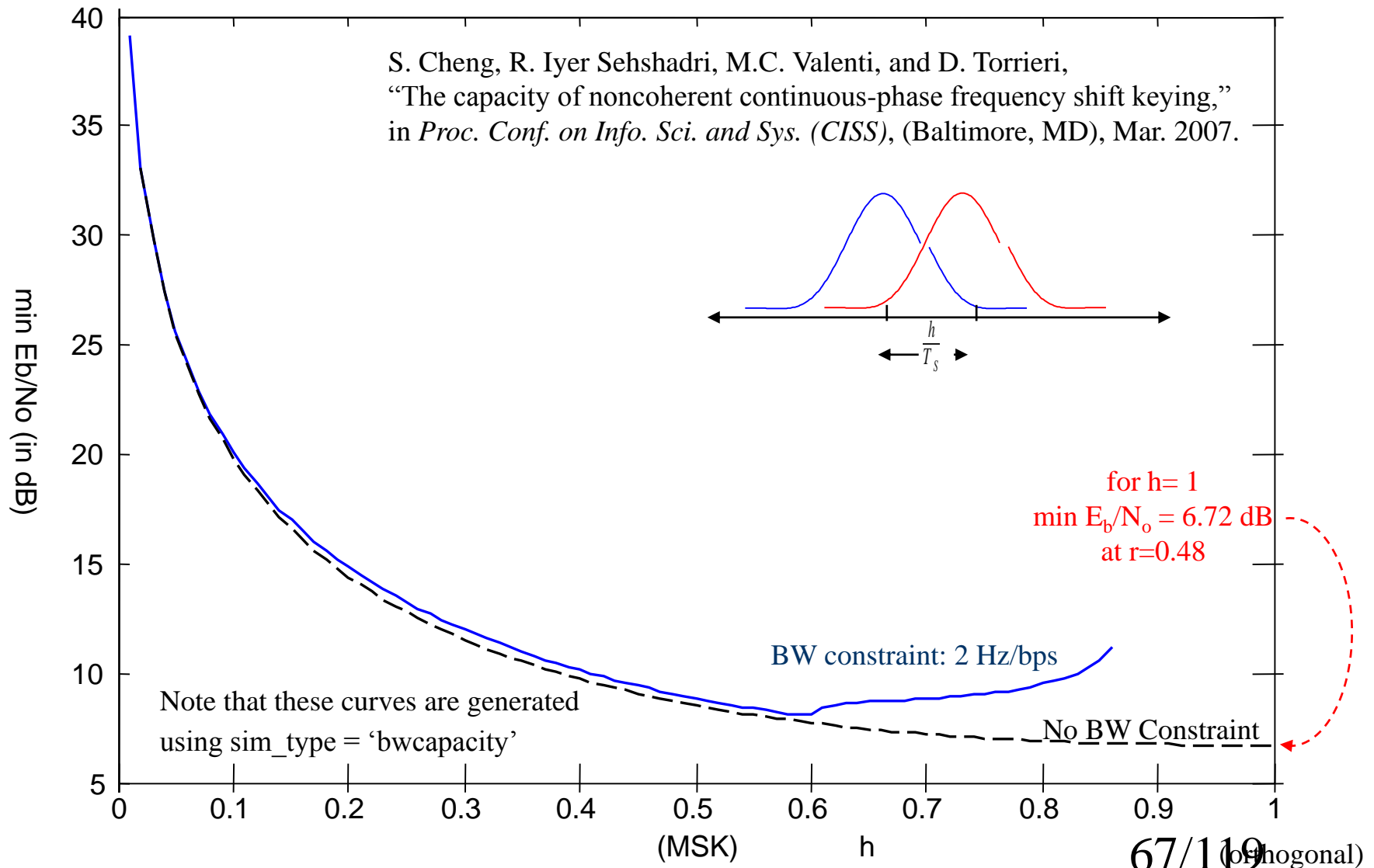
M=16

M=64

Rate R (symbol per channel use)



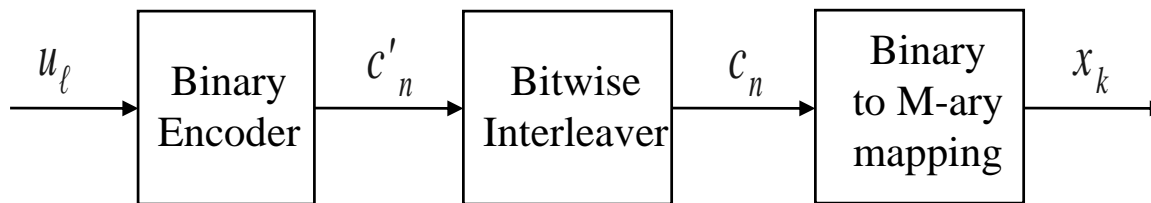
# Capacity of Nonorthogonal CPFSK



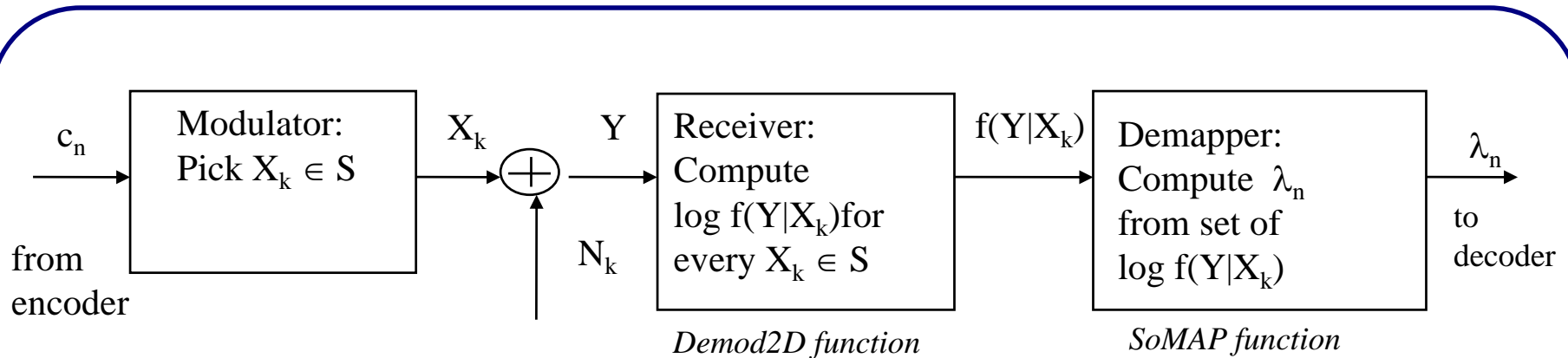
# BICM

(Caire 1998)

- Coded modulation (CM) is required to attain the aforementioned capacity.
  - Channel coding and modulation handled jointly.
  - Alphabets of code and modulation are matched.
  - e.g. trellis coded modulation (Ungerboeck); coset codes (Forney)
- Most off-the-shelf capacity approaching codes are binary.
- A pragmatic system would use a binary code followed by a bitwise interleaver and an M-ary modulator.
  - Bit Interleaved Coded Modulation (BICM).



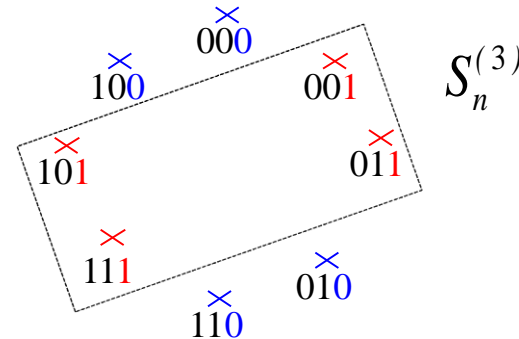
# BICM Receiver



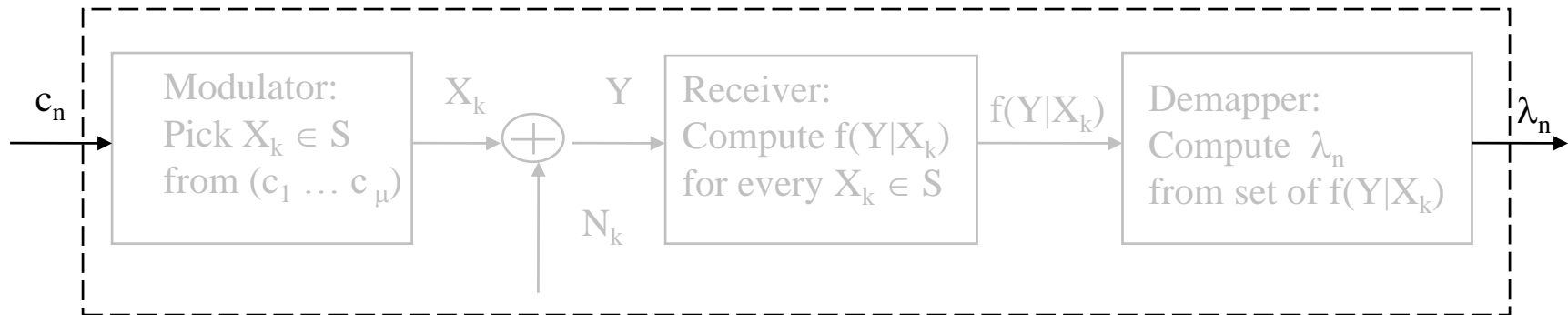
- The symbol likelihoods must be transformed into bit log-likelihood ratios (LLRs):

$$\lambda_n = \log \frac{\sum_{X_k \in S_n^{(1)}} f(Y | X_k)}{\sum_{X_k \in S_n^{(0)}} f(Y | X_k)}$$

- where  $S_n^{(1)}$  represents the set of symbols whose  $n$ th bit is a 1.
- and  $S_n^{(0)}$  is the set of symbols whose  $n$ th bit is a 0.



# BICM Capacity



- Can be viewed as  $\mu = \log_2 M$  binary parallel channels, each with capacity

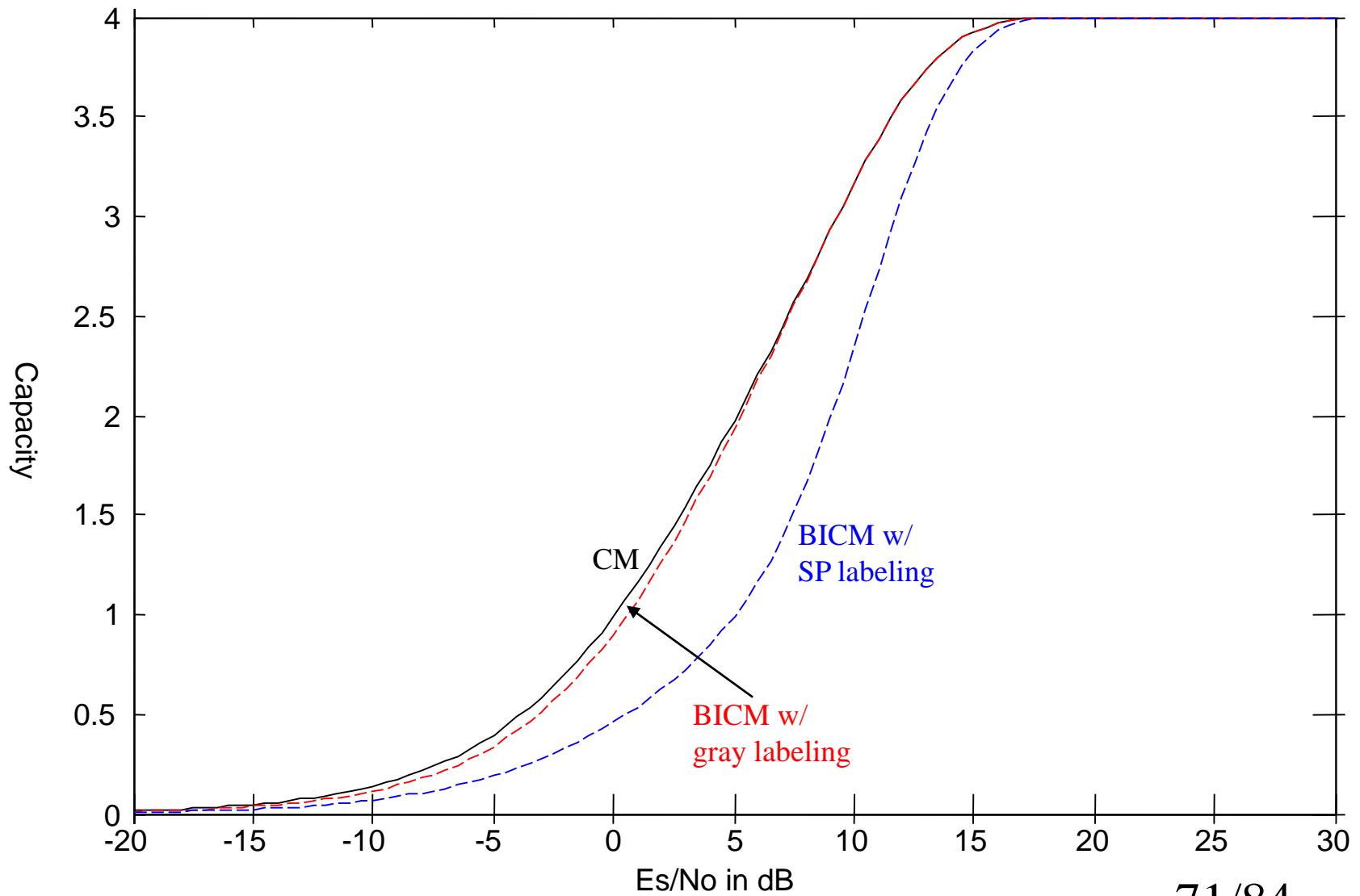
$$C_n = I(c_n, \lambda_n)$$

- Capacity over parallel channels adds:

$$C = \sum_{n=1}^{\mu} C_n$$

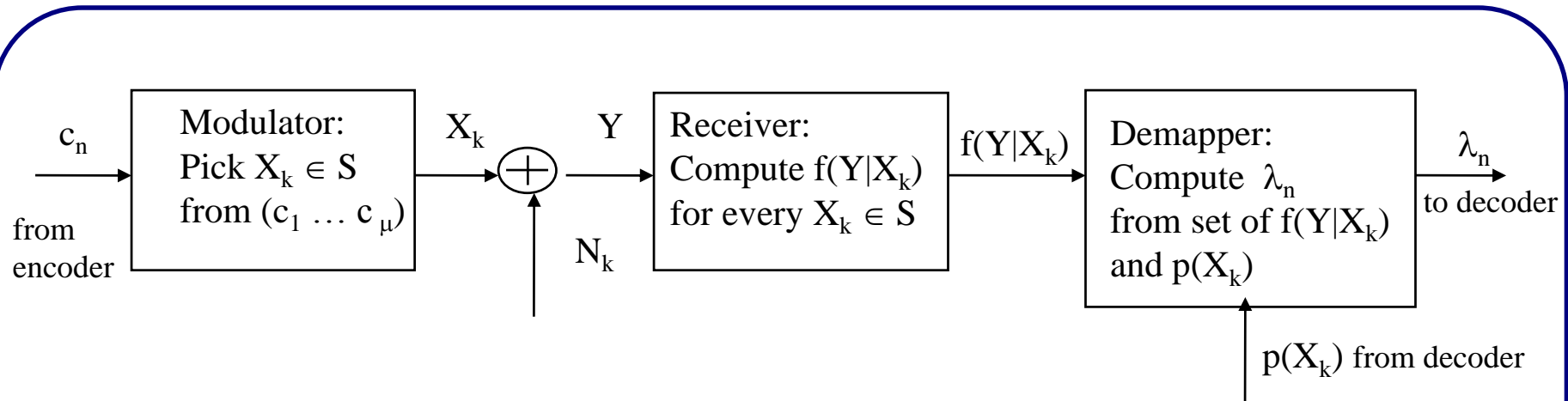
- As with the CM case, Monte Carlo integration may be used.

# CM vs. BICM Capacity for 16QAM



# BICM-ID

(Li & Ritcey 1997)



- A SISO decoder can provide side information to the demapper in the form of a priori symbol likelihoods.
  - BICM with Iterative Detection The demapper's output then becomes

$$\lambda_n = \log \frac{\sum_{X_k \in S_n^{(1)}} f(Y / X_k) p(X_k)}{\sum_{X_k \in S_n^{(0)}} f(Y / X_k) p(X_k)}$$



# Capacity Simulations in CML

- `sim_param(record).sim_type = 'capacity'`
- Exact same parameters as for uncoded simulations
  - SNR
  - SNR\_type = 'Es/No in dB'
  - framesize
  - modulation
  - mod\_order
  - channel
  - bicm
  - demod\_type
  - max\_trials

# Exercises

- Determine the capacity for
  - BPSK in AWGN
  - 64QAM with gray labeling in AWGN
  - 64QAM with gray labeling in Rayleigh fading
- Setup BICM-ID for
  - 16-QAM with SP mapping in AWGN and (7,5) CC.

# Outline

## 1. CML overview

- What is it? How to set it up and get started?

## 2. Uncoded Modulation

- Simulate uncoded BPSK and QAM in AWGN and Rayleigh fading

## 3. Coded Modulation

- Simulate a turbo code from UMTS 25.212
- \*Specify a custom eIRA LDPC parity check matrix

## 4. Ergodic (Shannon) capacity analysis

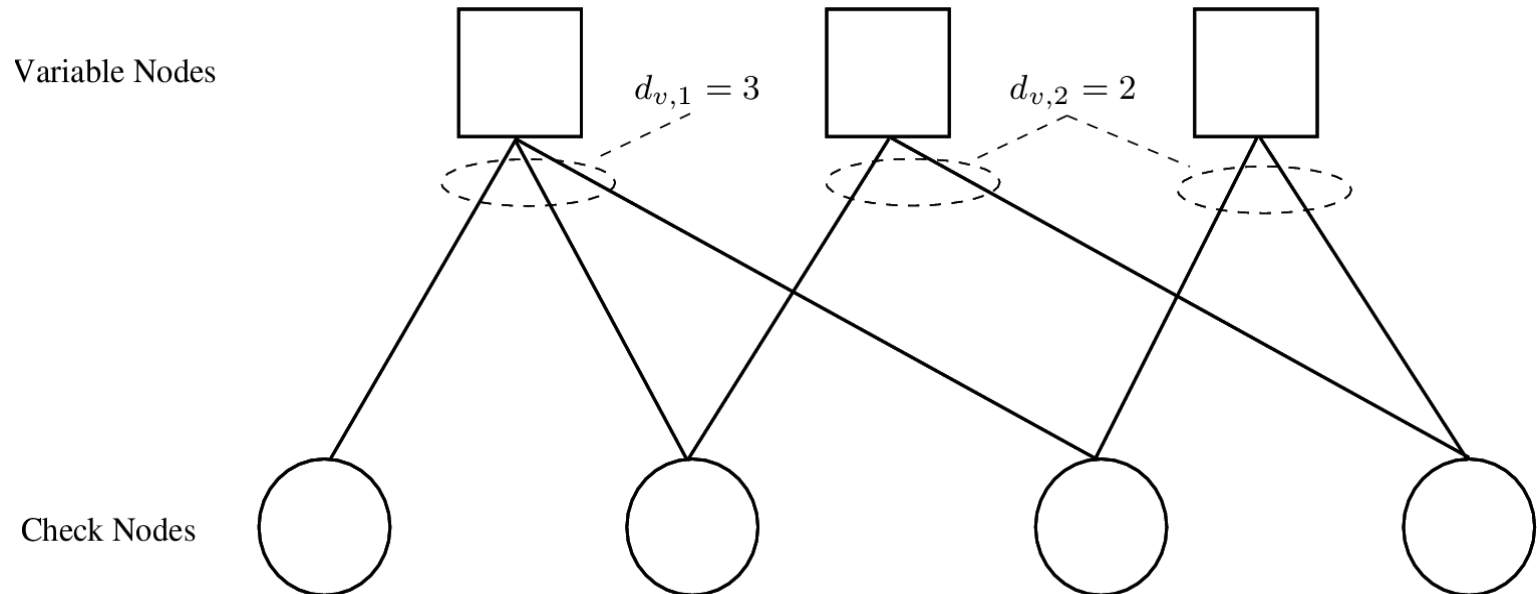
- Determine the modulation constrained capacity of BPSK and QAM

## 5. EXIT analysis

- **Determine the EXIT characteristic for BPSK in AWGN for a particular LDPC degree distribution**

# EXIT Analysis

- EXIT analysis may be applied to design LDPC code degree distributions which optimize decoding performance



# EXIT Analysis

- Design goal: match mutual information transfer characteristics of LDPC decoders through selection of variable node degrees
  - VND - variable node decoder
  - CND - check node decoder
- VND decoder captures characteristic of detector and variable node decoder
- CND curve depends only on channel code parameters

# EXIT Analysis

- First step of generating VND transfer characteristic is to compute the mutual information at the input and output of the detector

$$I_{E,DET}(I_A, E_S / N_0)$$

- where

$I_A$  mutual information of a-prior LLRs at detector input

- Any detector supported by CML may be analyzed.

# EXIT Analysis

- The extrinsic information at the output of the combined variable node decoder/detector is

$$I_{E,VND}(I_A, d_v, E_S / N_0) = \dots$$

$$J\left(\sqrt{(d_v - 1)[J^{-1}(I_A)]^2 + [J^{-1}(I_{E,DET}(I_A, E_S / N_0))]}^2\right)$$

- where

$d_v$  variable node degree

$J()$  function computing the mutual information  $I(X, \Lambda(Y))$  [1]

and  $Y = X + N$  is an AWGN channel where  $X$  is drawn from a BPSK constellation, and  $\Lambda(Y)$  is the LLR of the channel output.

[1] S. ten Brink, G. Kramer, and A. Ashikhmin, "Design of low-density parity-check codes for modulation and detection," IEEE Trans. Commun., vol. 52, pp. 670–678, Apr. 2004.

# EXIT Analysis

- Considering an LDPC decoder having multiple variable node degrees (irregular)

$$I_{E, VND}(I_A, E_S / N_0) = \sum b_i \cdot I_{E, VND}(I_A, d_{v,i}, E_S / N_0)$$

- where

$D$  number of distinct variable node degrees

$b_i$  fraction of edges incident on variable nodes of degree  $d_{v,i}$

$d_{v,i}$  i-th variable node degree



# EXIT Analysis - LDPC

- The extrinsic information at the output of the CND is

$$I_{E,CND}(I_A, d_c) = 1 - J(\sqrt{d_c - 1} \cdot J^{-1}(1 - I_A))$$

- where

$d_c$  check node degree (check regular)

- Note that the CND transfer characteristic depends only on the code parameters.

# EXIT Simulation Example

- First step: Simulate **detector** characteristic  
CML Scenario: ExitP2P

Record: 1

- `sim_param(record).sim_type = 'exit'`
- `sim_param(record).exit_param.exit_phase = 'detector'`
- `sim_param(record).exit_param.exit_type = 'ldpc'`
- `sim_param(record).exit_param.requested_IA = 0 : 0.01 : 0.9`
- `sim_param(record).channel = 'awgn'`
- `sim_param(record).SNR = [ 0 : 0.5 : 10 ]`
- `sim_param(record).modulation = 'psk'`
- `sim_param(record).mod_order = 4`

# EXIT Simulation Example

- Second step: Simulate **decoder** characteristic  
CML Scenario: ExitP2P

Record: 2

- `sim_param(record).sim_type = 'exit'`
- `sim_param(record).exit_param.exit_phase = 'decoder'`
- `sim_param(record).exit_param.exit_type = 'ldpc'`
- `sim_param(record).exit_param.requested_IA = 0 : 0.01 : 0.9`
- `sim_param(record).channel = 'awgn'`
- `sim_param(record).SNR = [ 0 : 0.5 : 10 ]`
- `sim_param(record).modulation = 'psk'`
- `sim_param(record).mod_order = 4`

# EXIT Simulation Example

- Second step: Simulate **decoder** characteristic  
CML Scenario: ExitP2P

Record: 2

- `sim_param(record).exit_param.rate = 0.6`
- `sim_param(record).exit_param.dv = [ 2 4 19 ]`
- `sim_param(record).exit_param.dv_dist = [ 0.4 0.52 0.08 ]`
- `sim_param(record).exit_param.dc = 11`
- `sim_param(record).exit_param.det_scenario = 'ExitP2P'`
- `sim_param(record).exit_param.det_record = 1`

# EXIT Simulation Example

## User Simulation Steps

Start MATLAB and initialize CML

```
>> CmlStartup
```

Execute detector simulation

```
>> CmlSimulate('ExitP2P', 1)
```

Execute decoder simulation

```
>> CmlSimulate('ExitP2P', 2)
```

Plot results

```
>> CmlPlot('ExitP2P', 2, 1) % Red argument: SNR to plot
```

# Outline

## 6. Outage analysis

- **Determine the outage probability over block fading channels**
- **Determine the outage probability of finite-length codes**

## 7. Digital Network Coding

- Describe CML implementation of digital network coding
- Simulate error rate of digital network coding for non-channel-coded 4-FSK in Rayleigh fading

## 8. The internals of CML

## 9. Throughput calculation

- Convert BLER to throughput for hybrid-ARQ

# Ergodicity vs. Block Fading

- Up until now, we have assumed that the channel is **ergodic**.
  - The observation window is large enough that the time-average converges to the statistical average.
- Often, the system might be **nonergodic**.
- Example: **Block fading**

b=1	b=2	b=3	b=4	b=5
$\gamma_1$	$\gamma_2$	$\gamma_3$	$\gamma_4$	$\gamma_5$



The codeword is broken into B equal length blocks  
The SNR changes randomly from block-to-block  
The channel is conditionally Gaussian  
The instantaneous  $E_s/N_0$  for block b is  $\gamma_b$

# Accumulating Mutual Information

- The SNR  $\gamma_b$  of block  $b$  is a random.
- Therefore, the mutual information  $I_b$  for the block is also random.
  - With a complex Gaussian input,  $I_b = \log(1 + \gamma_b)$
  - Otherwise the modulation constrained capacity can be used for  $I_b$

b=1	b=2	b=3	b=4	b=5
$I_1 = \log(1 + \gamma_1)$	$I_2$	$I_3$	$I_4$	$I_5$

The mutual information of each block is  $I_b = \log(1 + \gamma_b)$

Blocks are conditionally Gaussian

The entire codeword's mutual info is the sum of the blocks'

$$I_1^B = \sum_{b=1}^B I_b \quad (\text{Code combining})$$



# Information Outage

- An **information outage** occurs after B blocks if

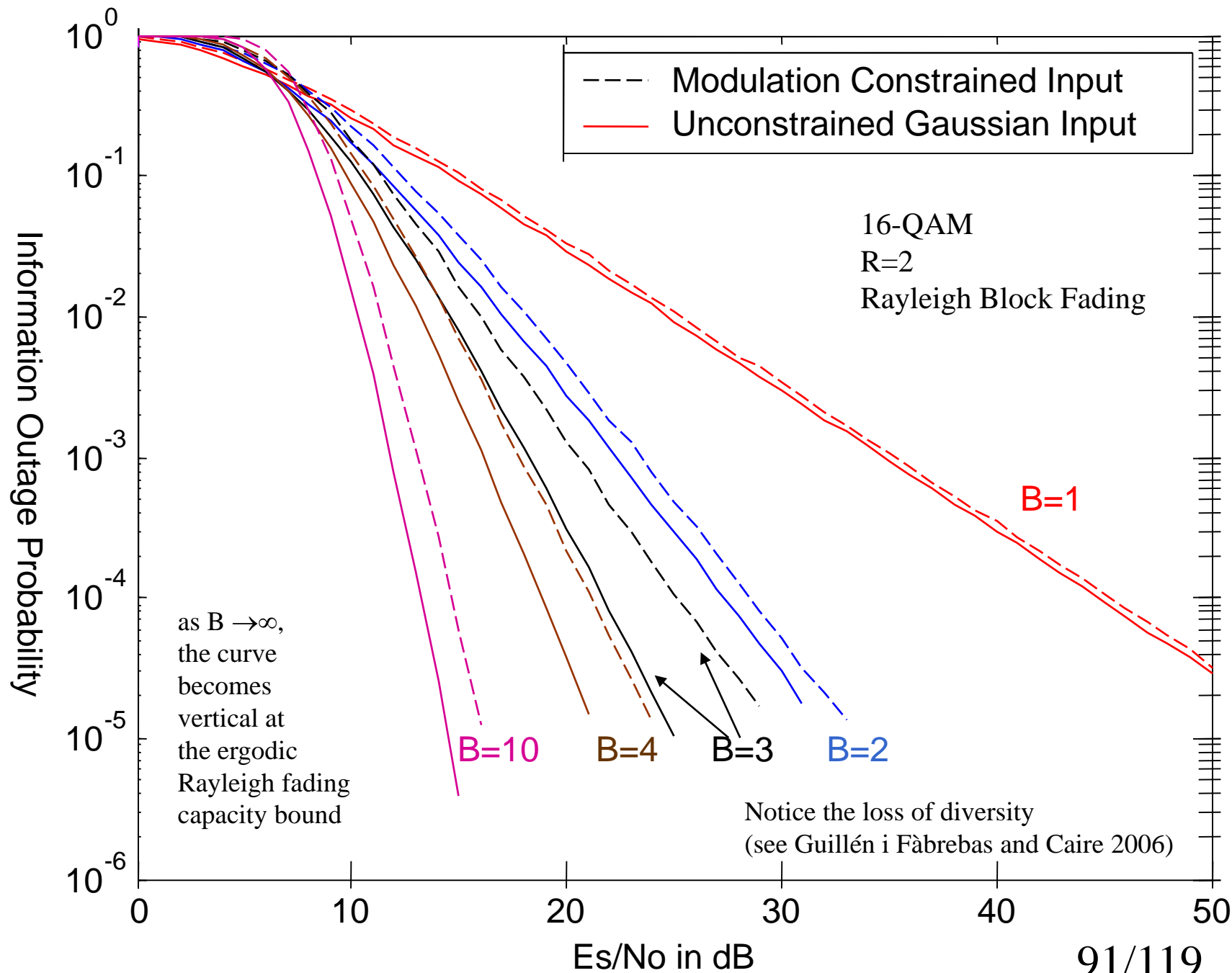
$$I_1^B < R$$

- where  $R \leq \log_2 M$  is the rate of the coded modulation

- An outage implies that no code can be reliable for the particular channel instantiation
- The information outage probability is

$$P_0 = P[I_1^B < R]$$

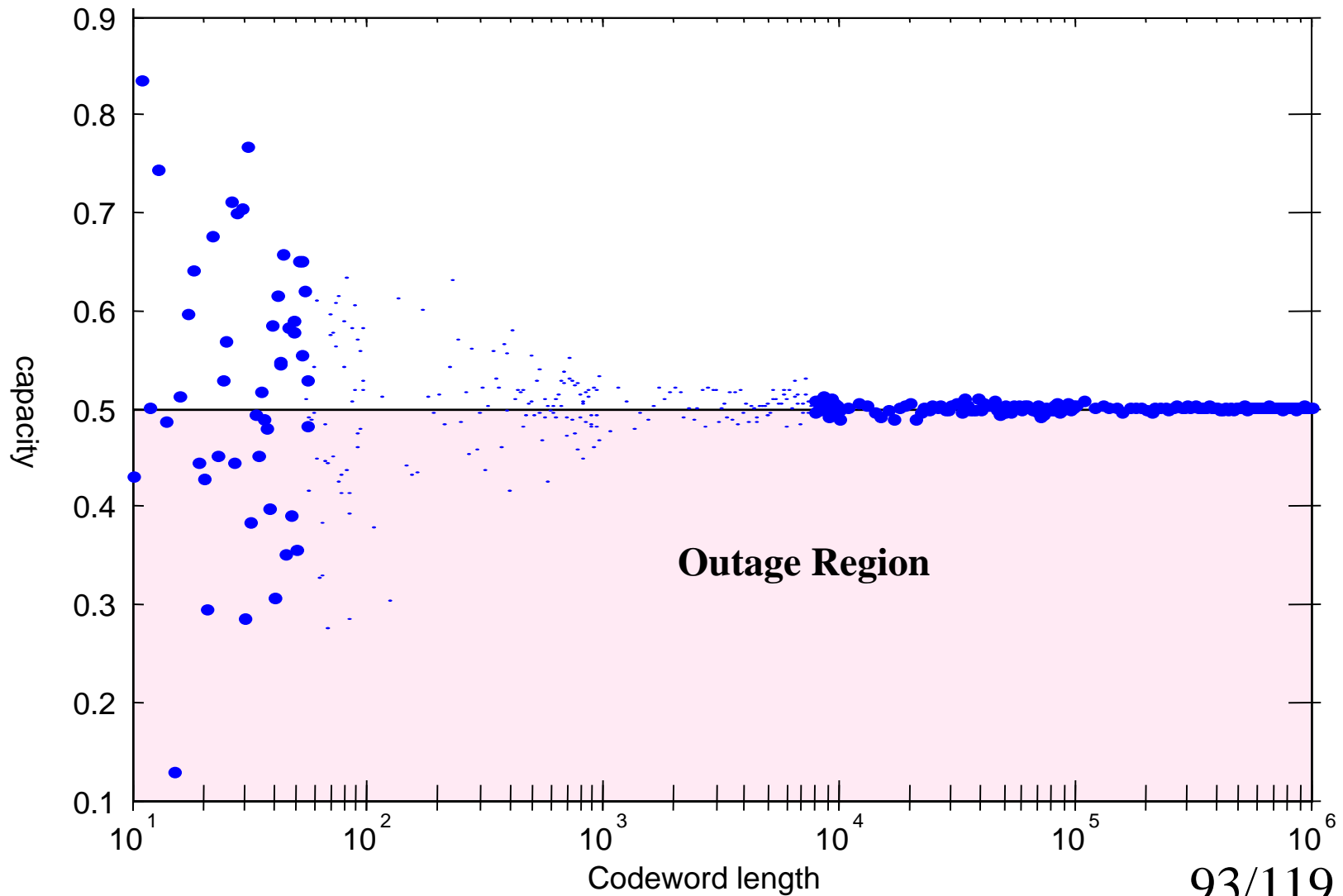
- This is a practical bound on FER for the actual system.

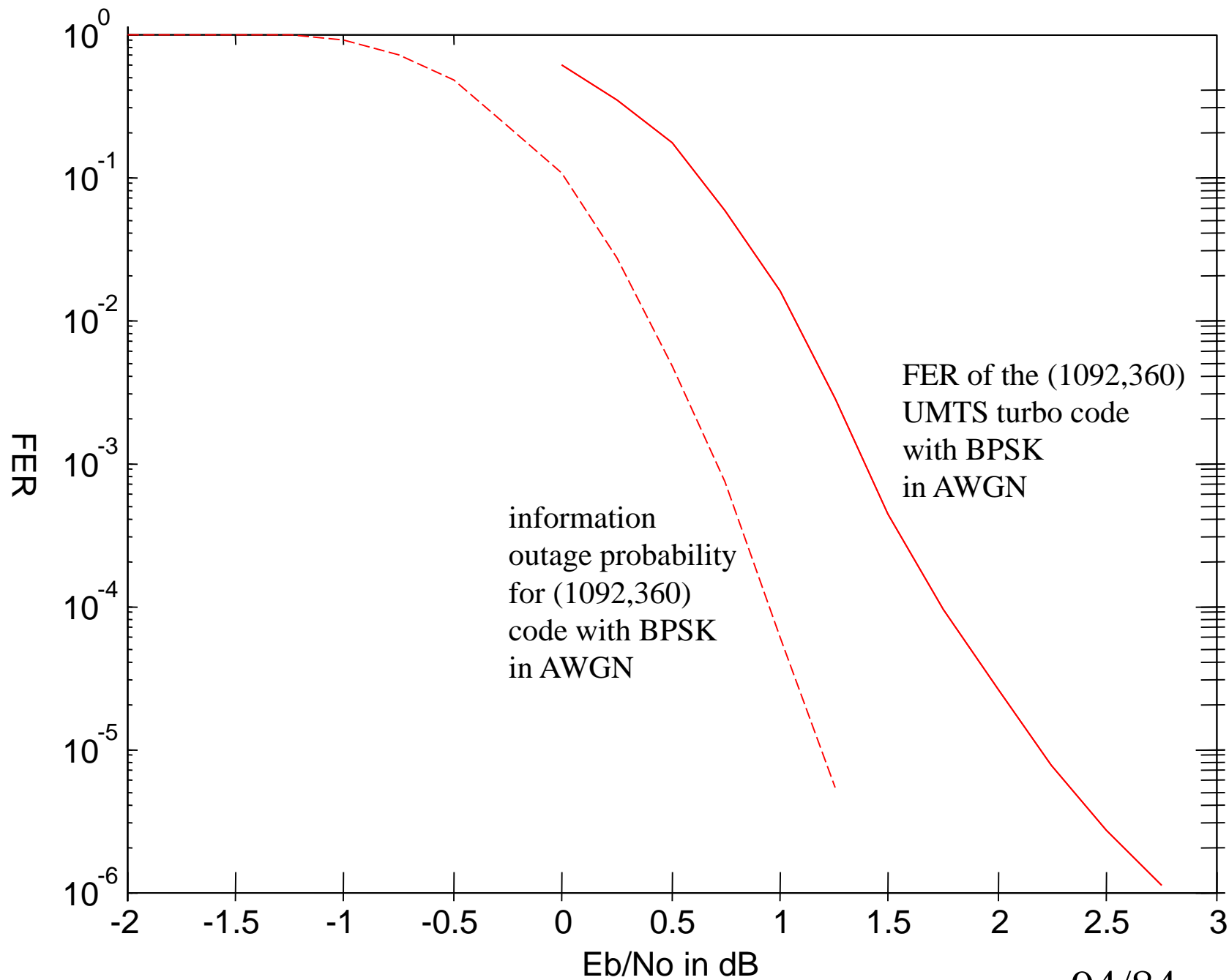


# Outage Simulation Type

- `sim_param(record).`
  - `blocks_per_frame`
    - Assumes block fading channel
  - `mod_order`
    - 0 for Gaussian input case
  - `rate`
    - Code rate.
    - Outage whenever  $MI < \text{rate}$
  - `combining_type = {'code', 'diversity'}`
  - `input_filename`
    - Required if `mod_order > 0`
    - Contains results of a capacity simulation.
    - Used for a table look-up operation

# Finite Length Codeword Effects





# Bloutage Simulation Type

- Set up like an uncoded simulation
  - framesize
  - specify the modulation
    - Set `mod_order = 0` for unconstrained Gaussian input
  - specify the channel (AWGN, Rayleigh, etc.)
- Also requires the rate
- Saves FER, not BER

# Outline

## 6. Outage analysis

- Determine the outage probability over block fading channels
- Determine the outage probability of finite-length codes

## 7. Digital Network Coding

- **Describe CML implementation of digital network coding**
- **Simulate error rate of digital network coding for non-channel-coded 4-FSK in Rayleigh fading**

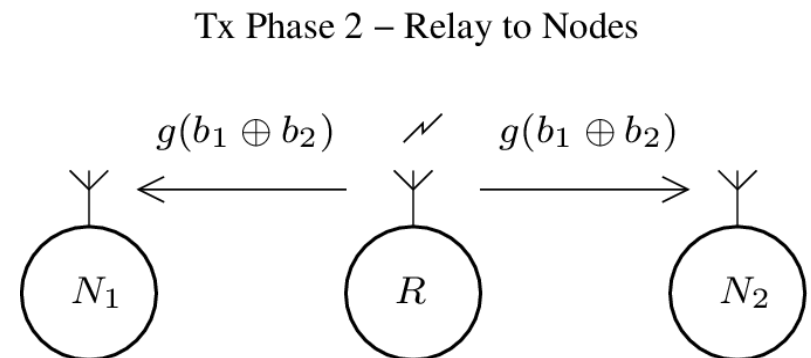
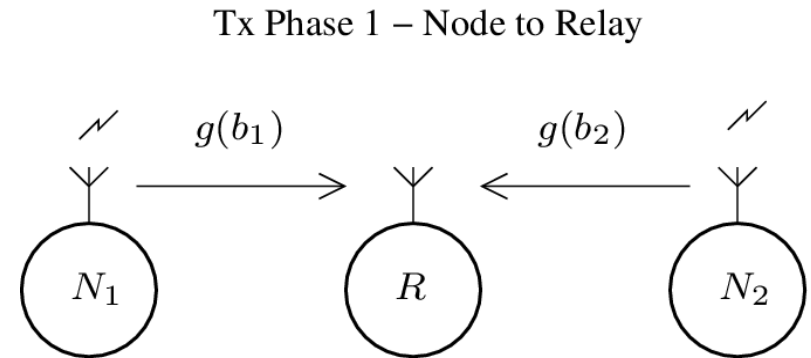
## 8. The internals of CML

## 9. Throughput calculation

- Convert BLER to throughput for hybrid-ARQ

# Digital Network Coding

- Digital network coding (DNC) is a relaying scheme which improves the capacity of multiple-access relay channels.
- Two or more nodes transmit in the same time and band to a relay, deliberately interfering.
- Relay receives interfered signal, encodes and broadcasts such that nodes may decode desired information.



Digital Network Coding Example  
Two-way Relay Channel

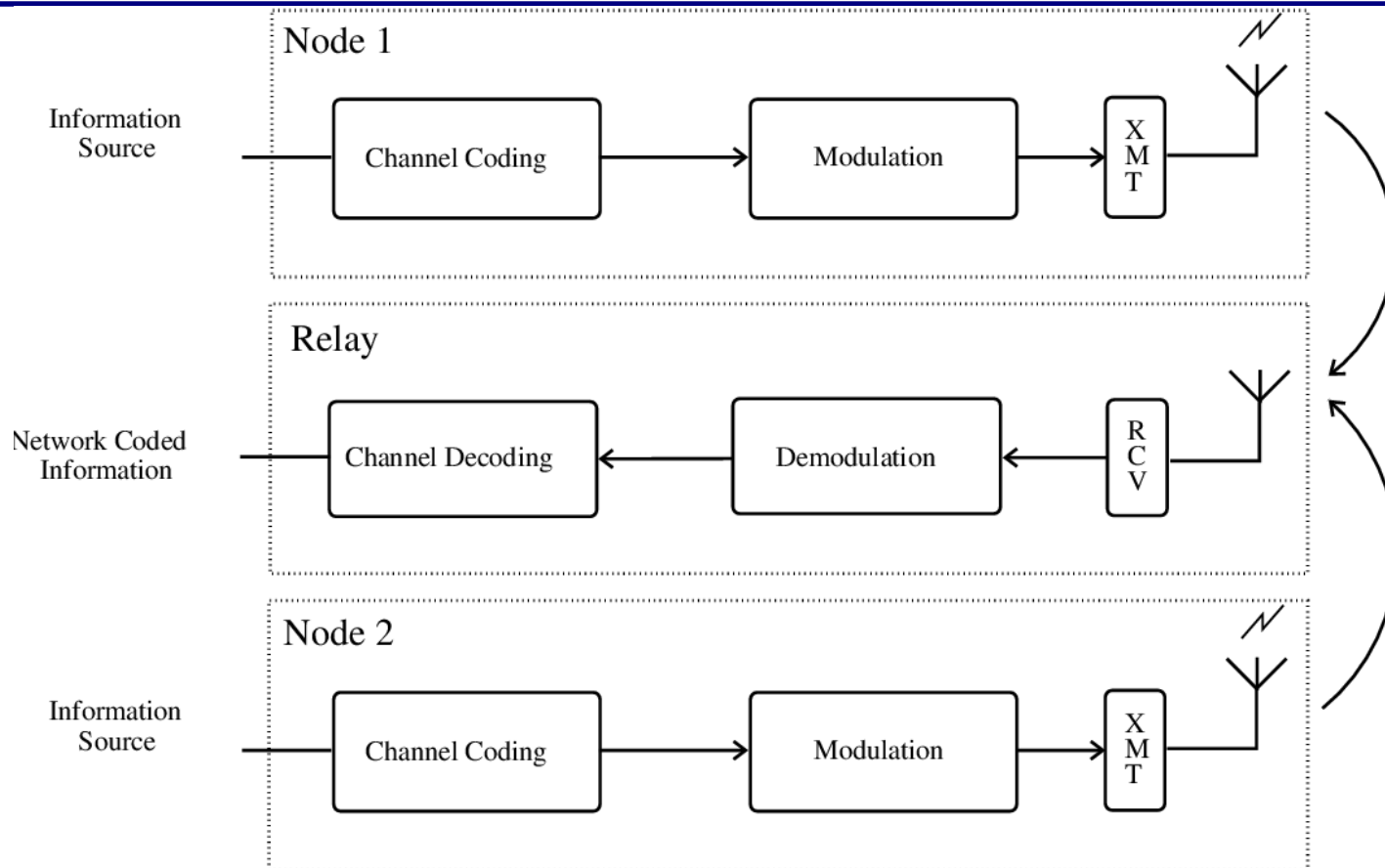


# Digital Network Coding

## Technical Requirements

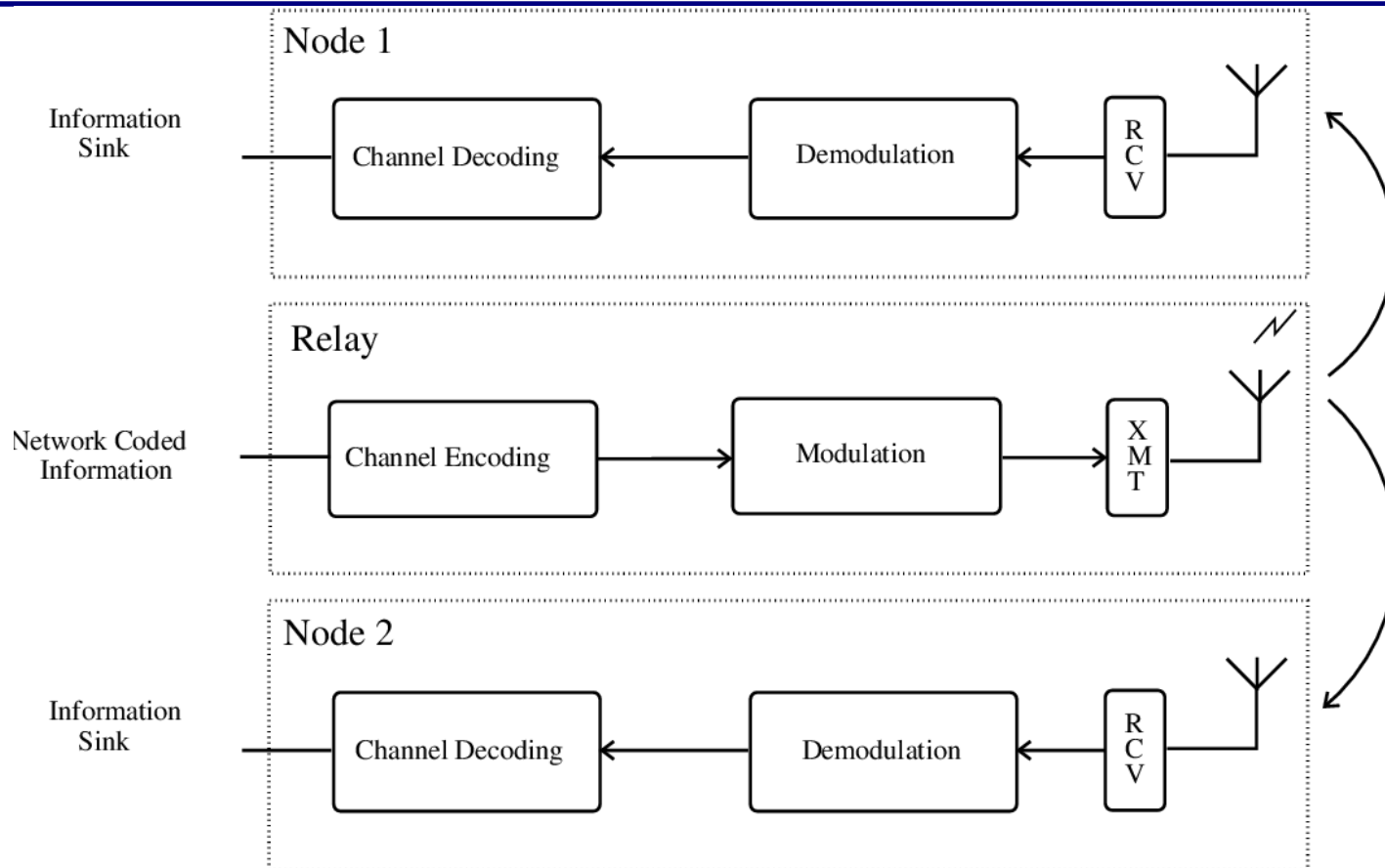
- Network encoding scheme at relay.
- Demodulation scheme at relay which performs network encoding.
- Demodulator producing soft-output channel observations suitable for capacity-approaching iterative decoding.

# Digital Network Coding



Tx Phase 1 - Block Diagram

# Digital Network Coding



Tx Phase 2 - Block Diagram

# Digital Network Coding

## CML DNC Implementation [2]

- Topology: Two-way Relay Channel
- Network encoding scheme at relay.
  - Exclusive-or  $b = b_1 \oplus b_2$
- Demodulation scheme at relay which performs network encoding.
  - Noncoherent FSK
- Demodulator producing soft-output channel observations suitable for capacity-approaching iterative decoding.
  - Codes supported: Turbo and LDPC

[2] M. C. Valenti, D. Torrieri, and T. Ferrett, "Noncoherent physical-layer network coding using binary CPFSK modulation," Proc. IEEE Military Commun. Conf., Oct. 2009.

# Digital Network Coding

## CML DNC Performance Simulation

- Compute relay error rate as a function of
  - Modulation order {2, 4, 8, 16 ...}
  - Channel Type {Rayleigh, AWGN}
  - Channel State Information {full, partial, none}
  - Channel code {Turbo, LDPC}

# DNC Simulation Example

## ■ CML Scenario: DncTwrc

Record: 1

- `sim_type = 'uncoded'`
- `topology = 'twrc'`
- `twrc_param.protocol = 'dnc'`
- `twrc_param.energy_ratio = 1`    % sources use same Tx energy
- `modulation = 'FSK'`
- `mod_order = 4`
- `channel = 'Rayleigh'`
- `csi_flag = 1`    % Partial CSI
- `SNR = [ 0 : 0.5 : 50 ]`

# DNC Simulation Example

## User Simulation Steps

Start MATLAB and initialize CML

```
>> CmlStartup
```

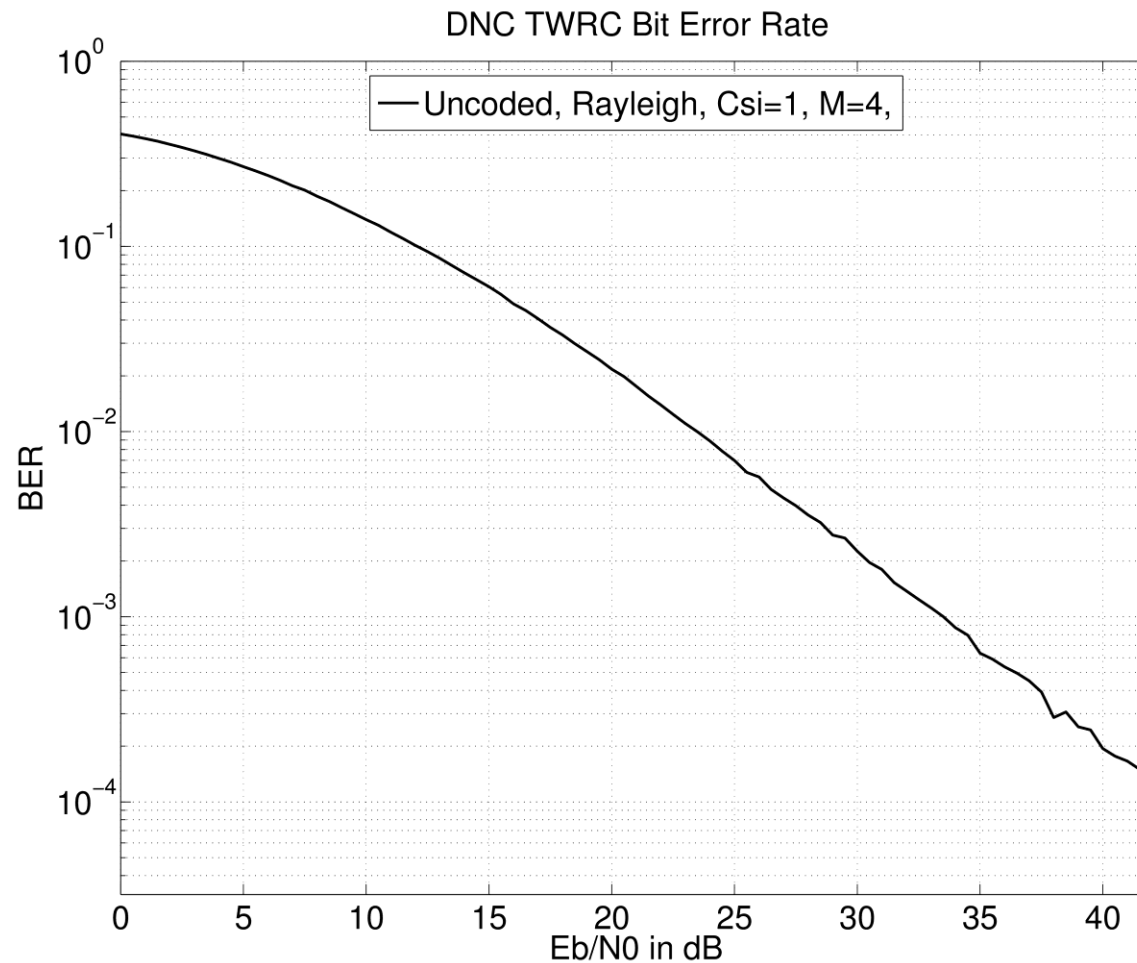
Execute DNC simulation

```
>> CmlSimulate('DncTwrc', 1)
```

Plot results

```
>> CmlPlot('DncTwrc', 1)
```

# DNC Simulation Example





# Outline

## 6. Outage analysis

- Determine the outage probability over block fading channels
- Determine the outage probability of finite-length codes

## 7. Digital Network Coding

- Describe CML implementation of digital network coding
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## 8. The internals of CML

## 9. Throughput calculation

- Convert BLER to throughput for hybrid-ARQ

# Main Program Flow

## ■ CmlSimulate

- ReadScenario
  - Runs SingleRead for each record
  - Performs sanity check on sim\_param structure
  - Initializes or restores the sim\_state structure
- For each record
  - SingleSimulate if a simulation
  - Otherwise, runs one of the analysis functions:
    - CalculateThroughput
    - CalculateMinSNR
    - CalculateMinSNRvsB

# SingleSimulate

- Seeds random number generator
- Branches into
  - SimulateMod
    - For uncoded, coded, and bloutage point-to-point channel
  - SimulateTwrc
    - For uncoded, coded, two-way relay channel
  - SimulateUGI
    - For a blocklength-constrained outage simulation with unconstrained Gaussian input.
  - SimulateCapacity
  - SimulateCapacityTwrc
  - SimulateOutage

# SimulateMod

- Main subfunctions (coded/uncoded) cases:
  - CmlEncode
  - CmlChannel
  - CmlDecode
- For bloutage, replace CmlDecode with
  - Somap
  - capacity

# SimulateTwrc

- Main subfunctions (coded/uncoded) cases:
  - CmlEncode
  - CmlTwrcRelayChannel
  - CmlTwrcRelayComputeSymbolLh
  - CmlInitSomap
  - CmlTwrcRelaySomap
  - CmlTwrcRelayDecode

# SimulateCapacity

- Operates like SimulateMod with `sim_type = 'bloutage'`
  - However, instead of comparing MI of each codeword against the rate, keeps a running average of MI.
- In case of exit simulation, computes detector and decoder mutual information characteristics

# SimulateCapacityTwrc

- Similar to SimulateTwrc but computes mutual information at output of detector and Somap rather than error rates
- In case of exit simulation, computes detector and decoder mutual information characteristic

# SimulateOutage

- Randomly generates SNR for each block
- Performs table lookup to get MI from SNR
- Compares MI against threshold



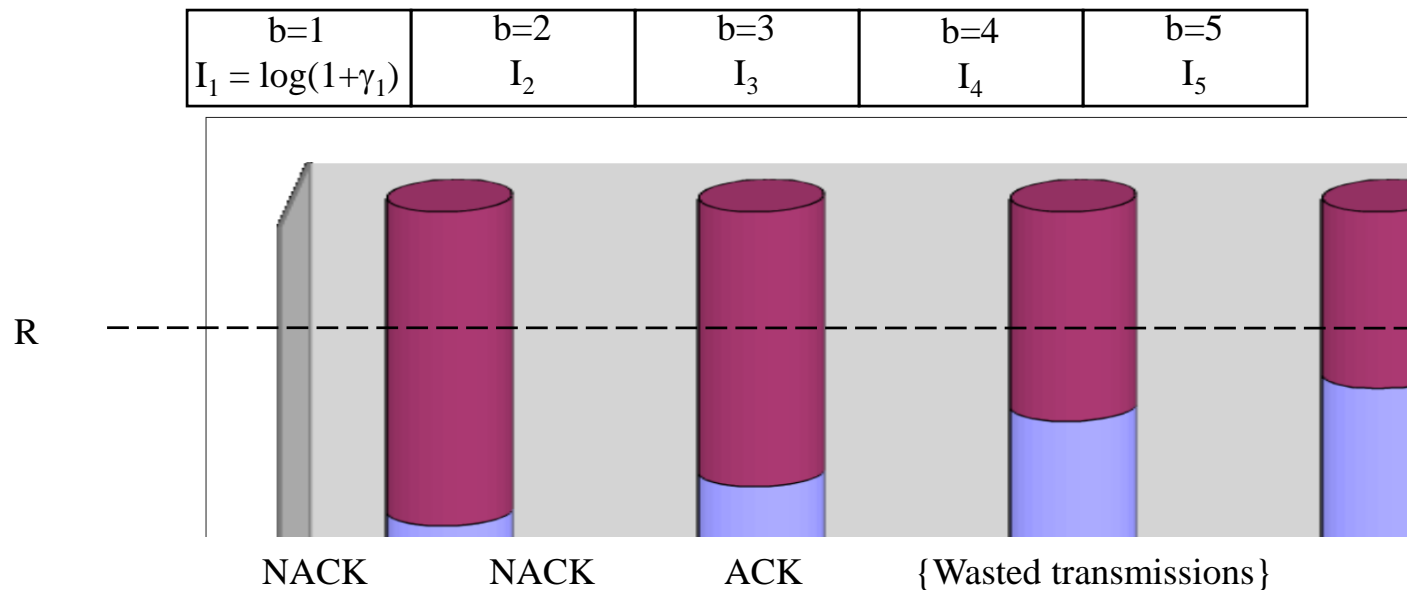
# Outline

1. CML overview
  - What is it? How to set it up and get started?
2. Uncoded modulation
  - Simulate uncoded BPSK and QAM in AWGN and Rayleigh fading
3. Coded modulation
  - Simulate a turbo code from UMTS 25.212
4. Ergodic (Shannon) capacity analysis
  - Determine the modulation constrained capacity of BPSK and QAM
5. Outage analysis
  - Determine the outage probability over block fading channels.
  - Determine the outage probability of finite-length codes
6. The internals of CML
7. Throughput calculation
  - Convert BLER to throughput for hybrid-ARQ

# Hybrid-ARQ

## (Caire and Tunninetti 2001)

- Once  $I_1^B > R$  the codeword can be decoded with high reliability.
- Therefore, why continue to transmit any more blocks?
- With hybrid-ARQ, the idea is to request retransmissions until  $I_1^B > R$ 
  - With hybrid-ARQ, outages can be avoided.
  - The issue then becomes one of latency and throughput.



# Outline

## 6. Outage analysis

- Determine the outage probability over block fading channels
- Determine the outage probability of finite-length codes

## 7. Digital Network Coding

- Describe CML implementation of digital network coding
- Simulate error rate of digital network coding for non-channel-coded 4-FSK in Rayleigh fading

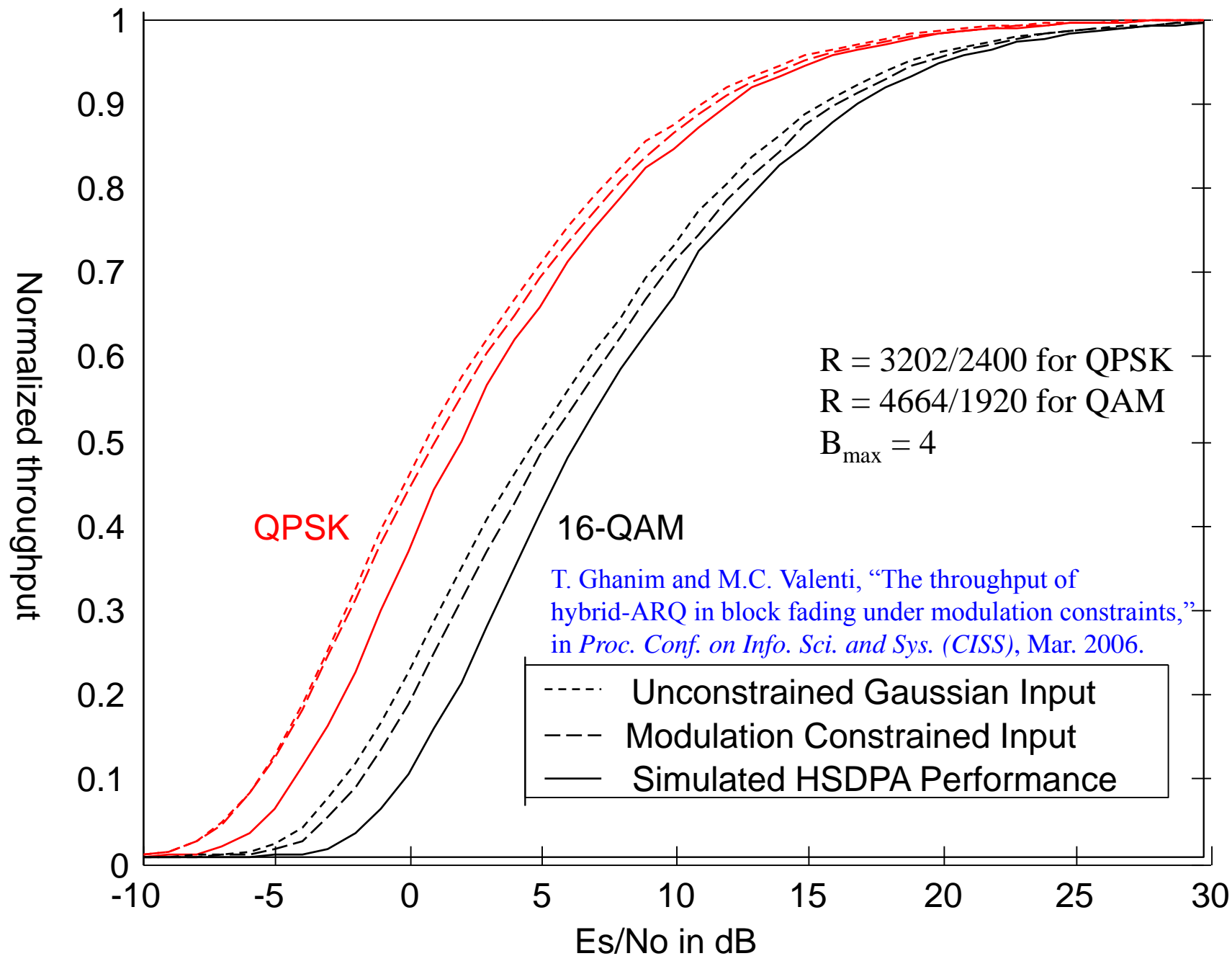
## 8. The internals of CML

## 9. Throughput calculation

- **Convert BLER to throughput for hybrid-ARQ**

# Latency and Throughput of Hybrid-ARQ

- With hybrid-ARQ  $B$  is now a random variable.
  - The average **latency** is proportional to  $E[B]$ .
  - The average **throughput** is inversely proportional to  $E[B]$ .
- Often, there is a practical upper limit on  $B$ 
  - Rateless coding (e.g. Raptor codes) can allow  $B_{\max} \rightarrow \infty$
- An example
  - HSDPA: High-speed downlink packet access
  - 16-QAM and QPSK modulation
  - UMTS turbo code
  - HSET-1/2/3 from TS 25.101
  - $B_{\max} = 4$



# Conclusions: Design Flow with CML

- When designing a system, first determine its capacity.
  - Only requires a slight modification of the modulation simulation.
  - Does not require the code to be simulated.
  - Allows for optimization with respect to free parameters.
- After optimizing with respect to capacity, design the code.
  - BICM with a good off-the-shelf code.
  - Optimize code with respect to the EXIT curve of the modulation.
- Information outage analysis can be used to characterize:
  - Performance in slow fading channels.
  - Delay and throughput of hybrid-ARQ retransmission protocols.
  - Finite codeword lengths.