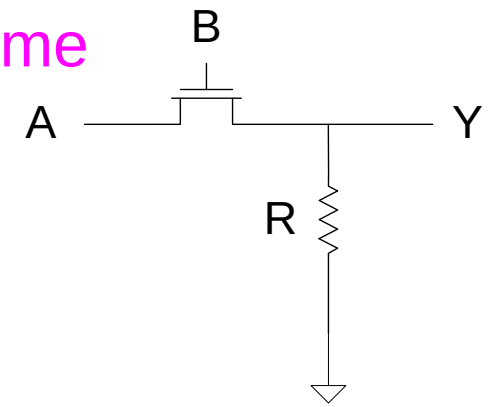
EE 330
Lecture 7

- Propagation Delay
- Stick Diagrams
- Technology Files
  - Design Rules

Pass Transistor Logic

Review from Last Time



$$Y = A \bullet B$$

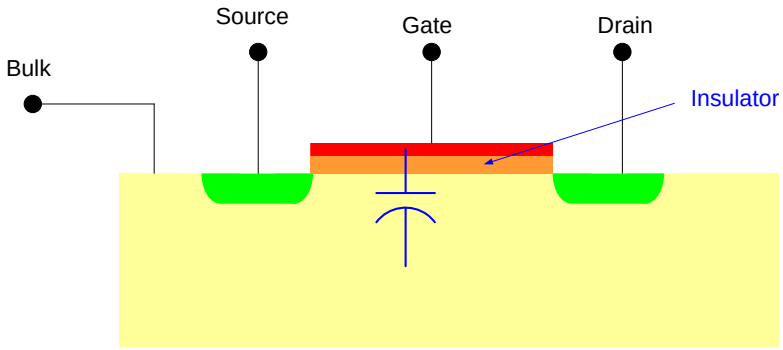
Requires only 1 transistor (and a resistor)

- **Pass transistor logic can offer significant reductions in complexity for some functions (particularly noninverting)**
- **Resistor may require more area than several hundred or even several thousand transistors**
- **Signal levels may not go to V_{DD} or to 0V**
- **Static power dissipation may not be zero**
- **Signals may degrade unacceptably if multiple gates are cascaded**
- **"resistor" often implemented with a transistor to reduce area but signal swing and power dissipation problems still persist**
- **Pass transistor logic is widely used**

Review from Last Time

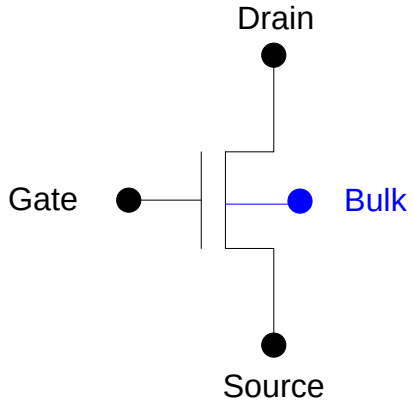
MOS Transistor

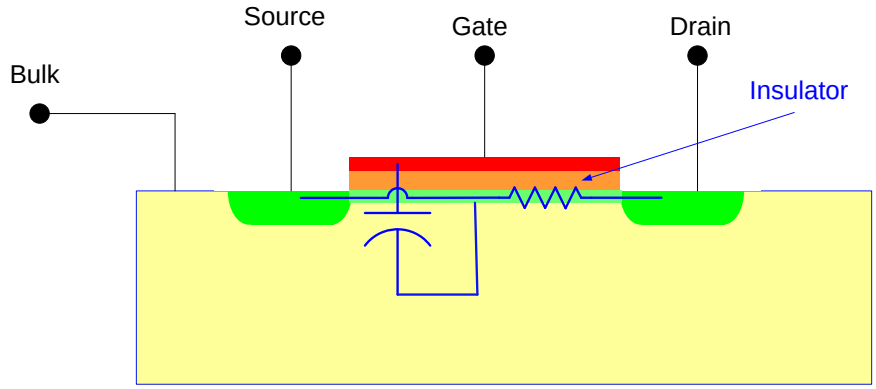
Qualitative Discussion of n-channel Operation



n-channel MOSFET

For V_{GS} small





n-channel MOSFET

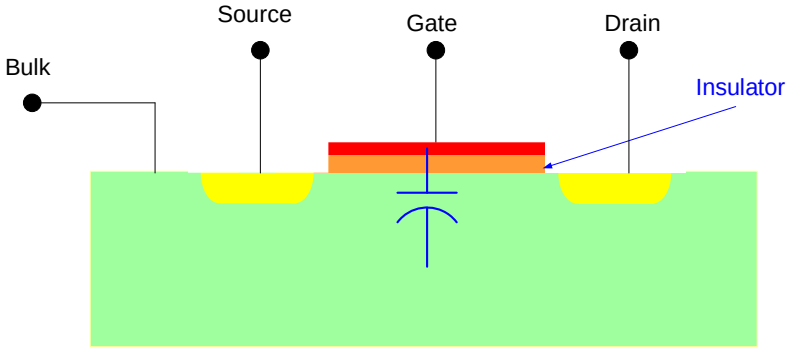
For V_{GS} large

- Electrically created inversion layer forms a “thin “film” resistor
- Capacitance from gate to channel region is distributed
- Lumped capacitance much easier to work with

Review from Last Time

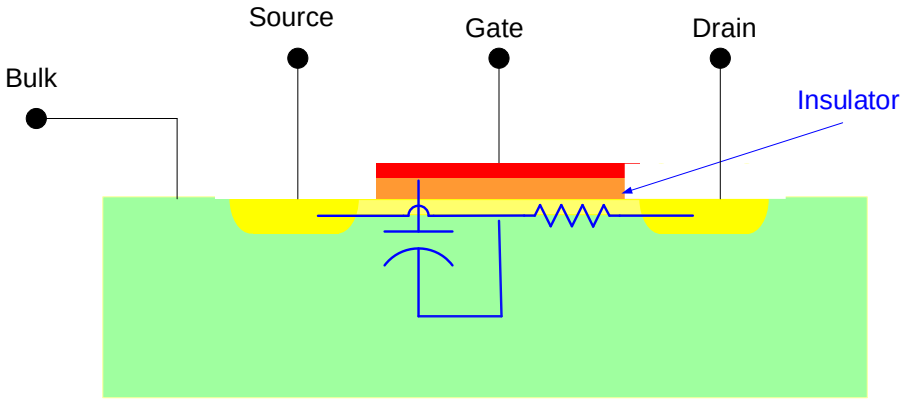
MOS Transistor

Qualitative Discussion of p-channel Operation

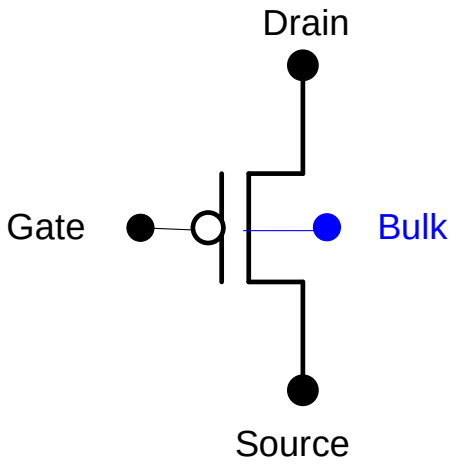


p-channel MOSFET

For $|V_{GS}|$ small



p-channel MOSFET

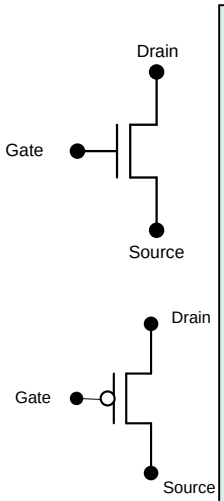


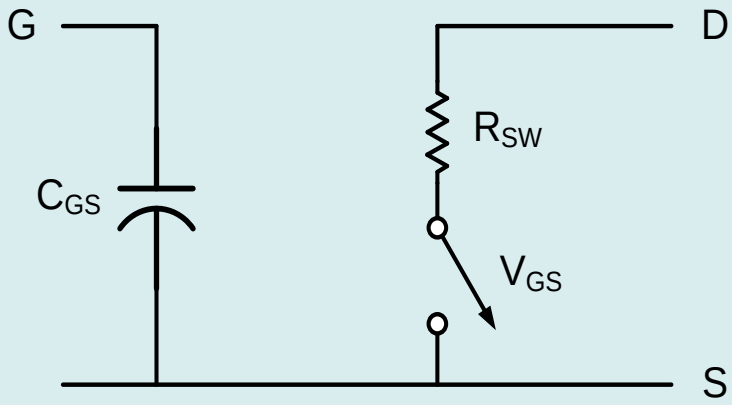
For $|V_{GS}|$ large

- Electrically created inversion layer forms a “thin “film” resistor
- Capacitance from gate to channel region is distributed
- Lumped capacitance much easier to work with

Review from Last Time

Improved Switch-Level Model





Switch closed for G =1 , n-ch
Switch closed for G =0 , p-ch

Switch-level model including gate capacitance and channel resistance

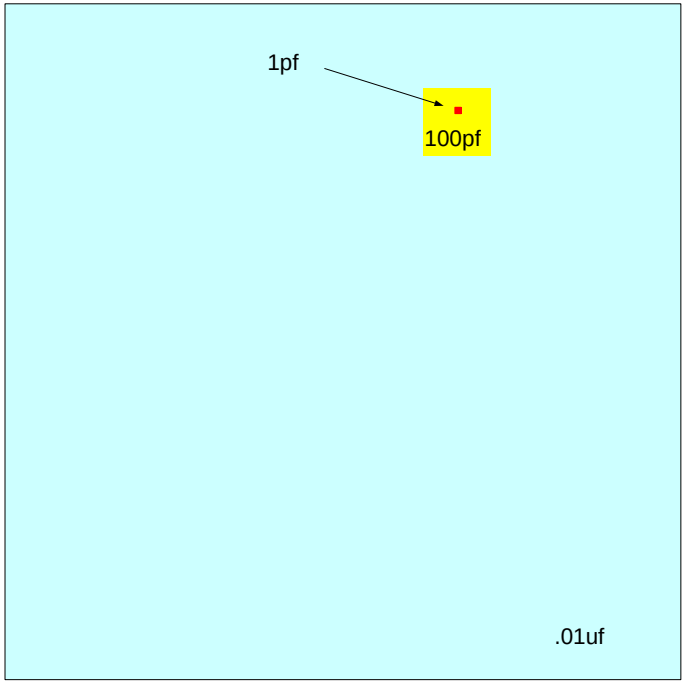
C_{GS} and R_{SW} dependent upon device sizes and process

For minimum-sized devices in a 0.5u process with V_{DD}=5V

$$C_{GS} \cong 1.5fF \qquad R_{sw} \cong \begin{cases} 2K\Omega \ \ n-channel \\ 6K\Omega \ \ p-channel \end{cases}$$

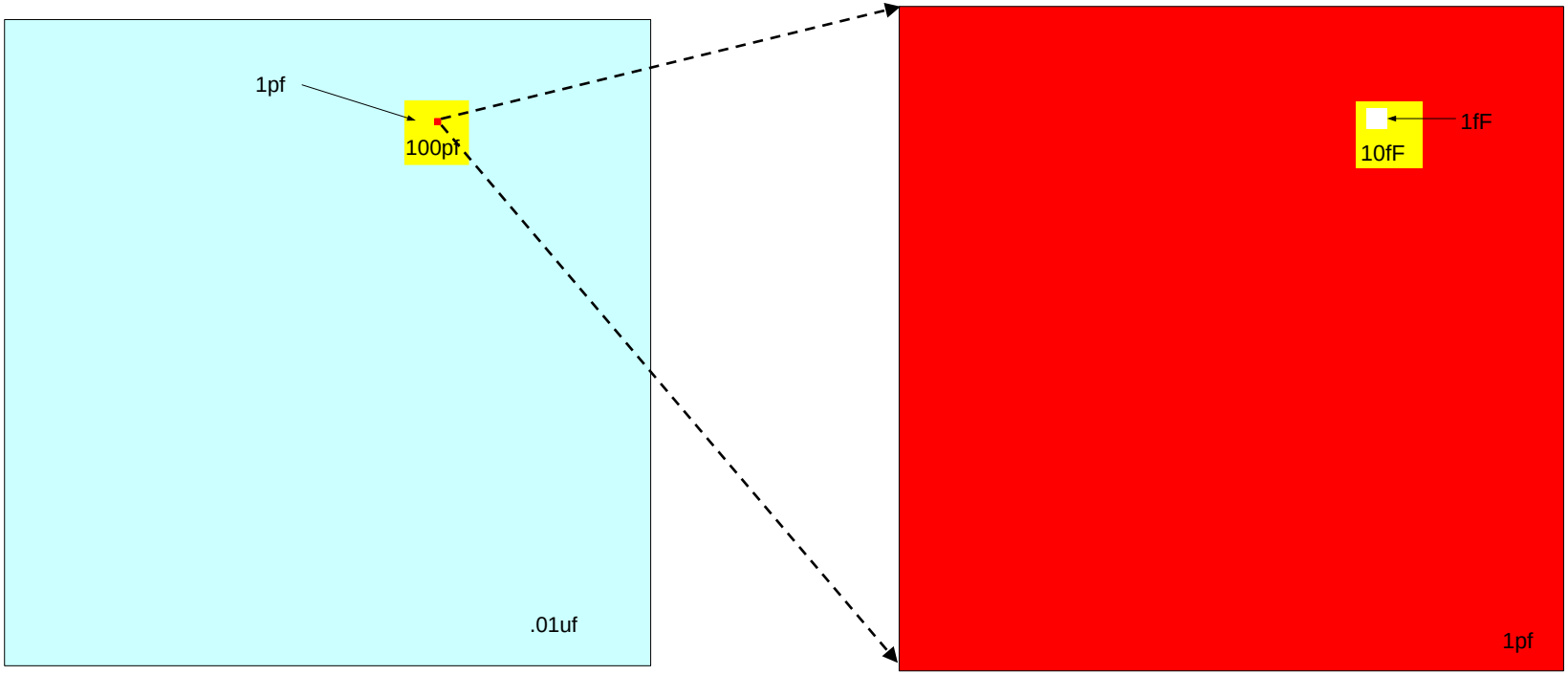
Considerable emphasis will be placed upon device sizing to manage C_{GS} and R_{SW}

Is a capacitor of 1.5fF small enough to be neglected?



Area allocations shown to relative scale:

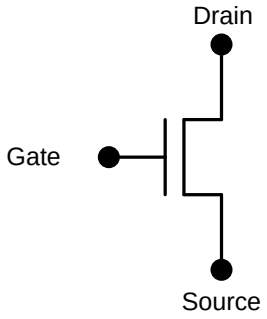
Is a capacitor of 1.5fF small enough to be neglected?



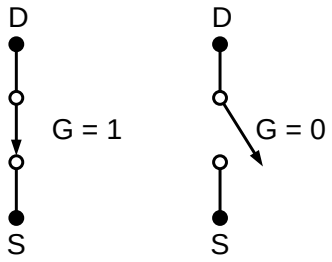
Area allocations shown to relative scale:

- **Not enough information at this point to determine whether this very small capacitance can be neglected**
- **Will answer this important question later**

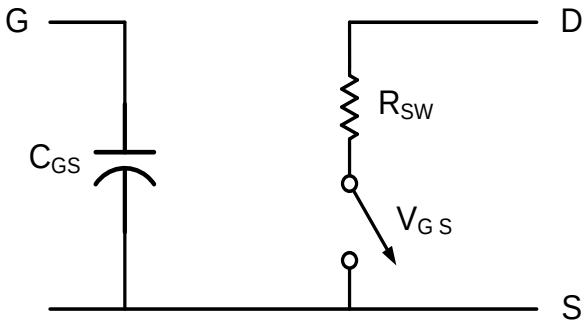
Model Summary (for n-channel)



1. Switch-Level model



2. Improved switch-level model

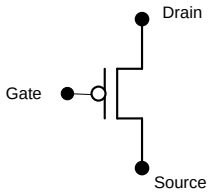


Switch closed for V_{GS}= large

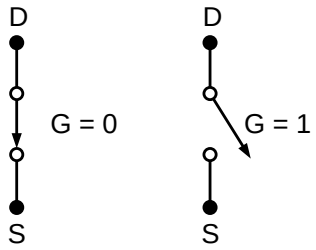
Switch open for V_{GS}= small

Other models will be developed later

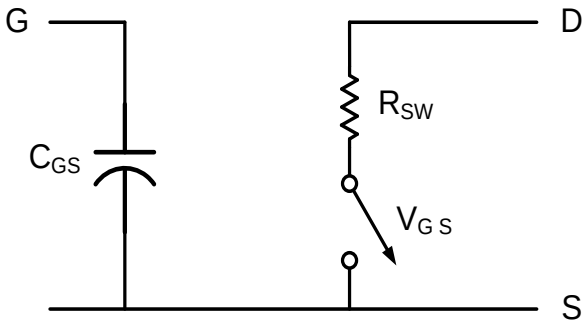
Model Summary (for p-channel)



1. Switch-Level model



2. Improved switch-level model



Switch closed for $|V_{GS}|$= large

Switch open for $|V_{GS}|$= small

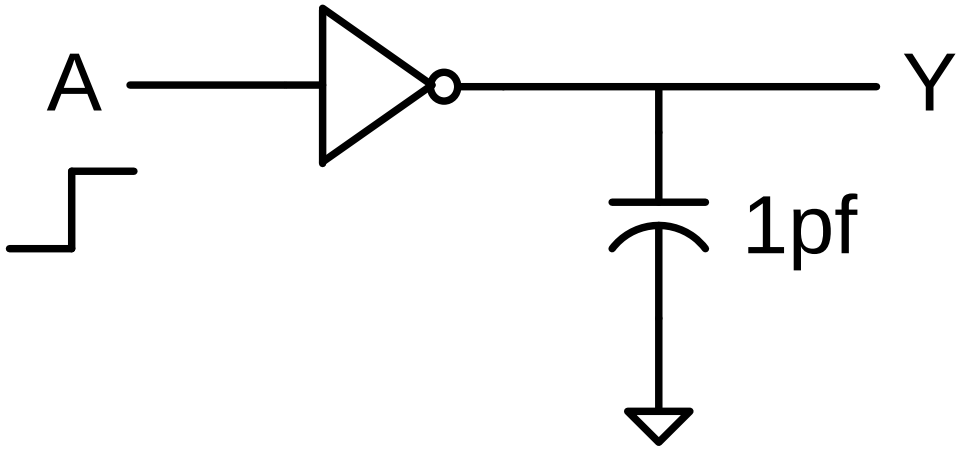
Other models will be developed later

- Pass Transistor Logic
- Improved Switch-Level Model
- Propagation Delay
- Stick Diagrams
- Technology Files

Example

What are t_{HL} and t_{LH}?

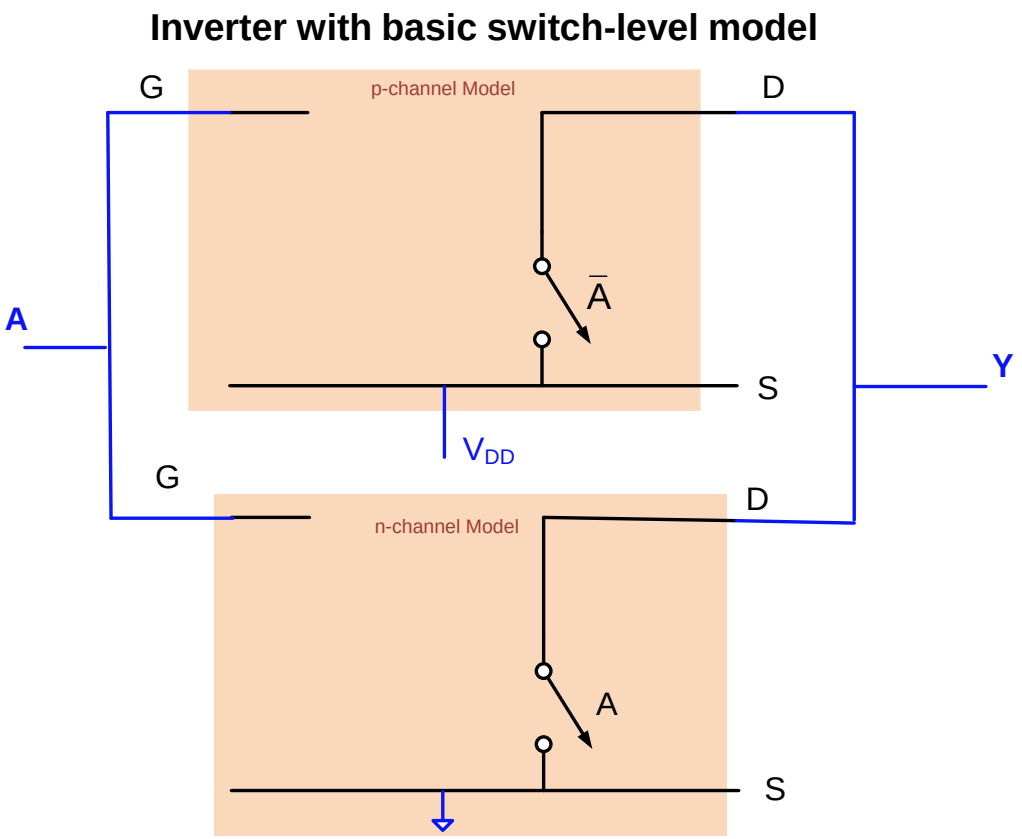
Assume V_{DD}=5V



With basic switch level model ?

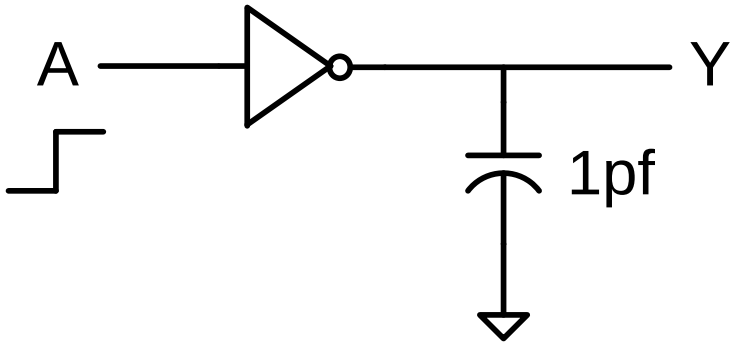
With improved switch level model ?

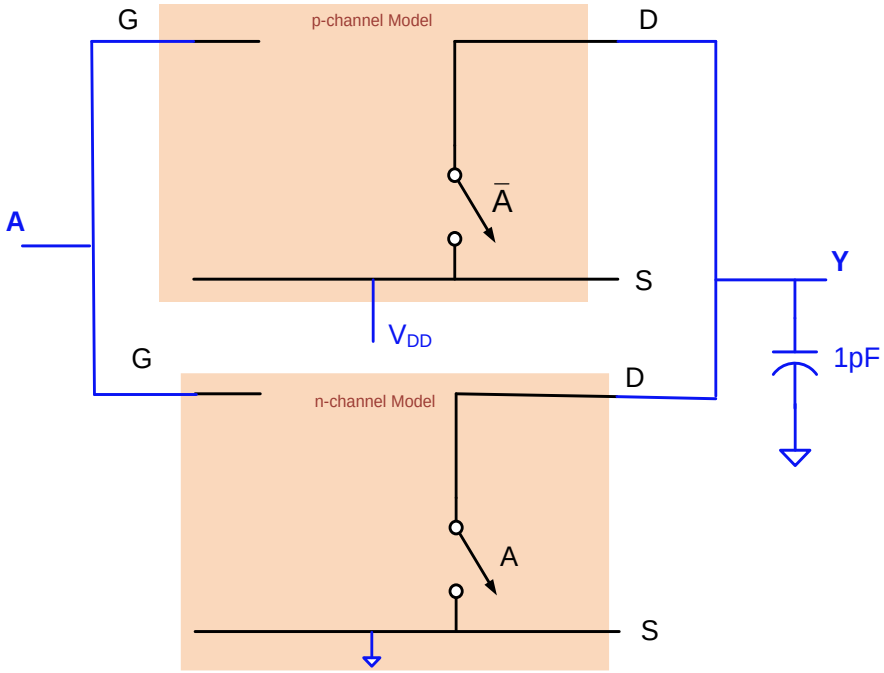
Example


Inverter with basic switch-level model
G
p-channel Model
D
$\overline{A}$
A
S
Y
V_{DD}
G
D
n-channel Model
A
S

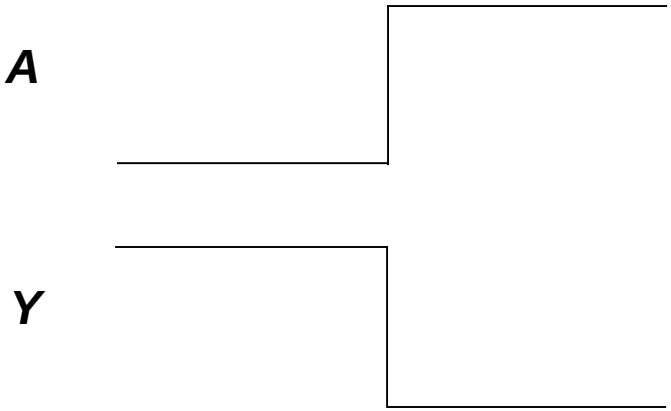
Example

What are t_{HL} and t_{LH}?





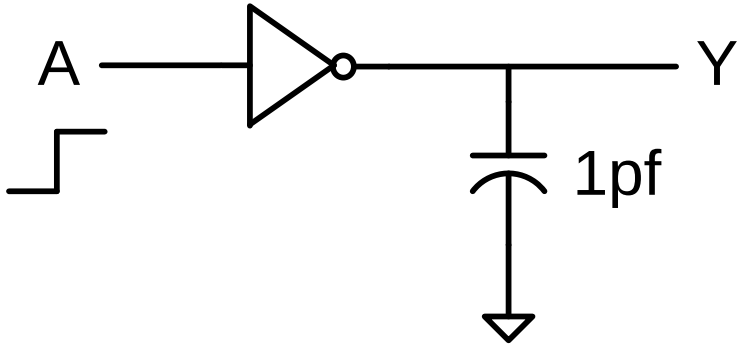
With basic switch level model



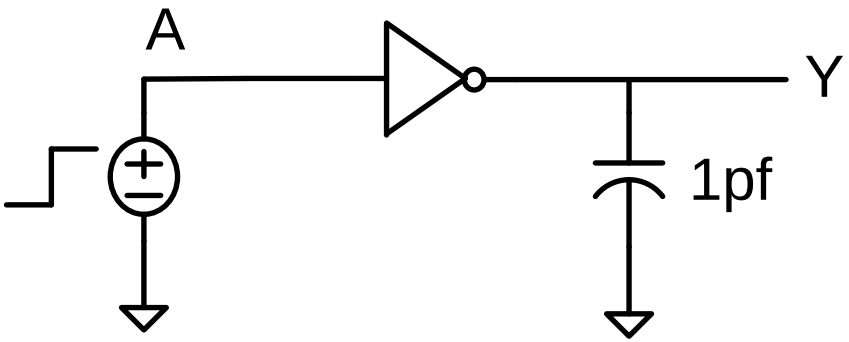
$t_{HL}=t_{LH}=0$

Example (cont)

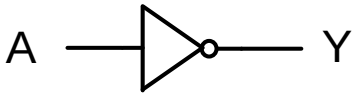
With simple switch-level model $t_{HL}=t_{LH}=0$



With improved model ?



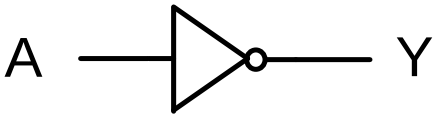
Inverter Model?



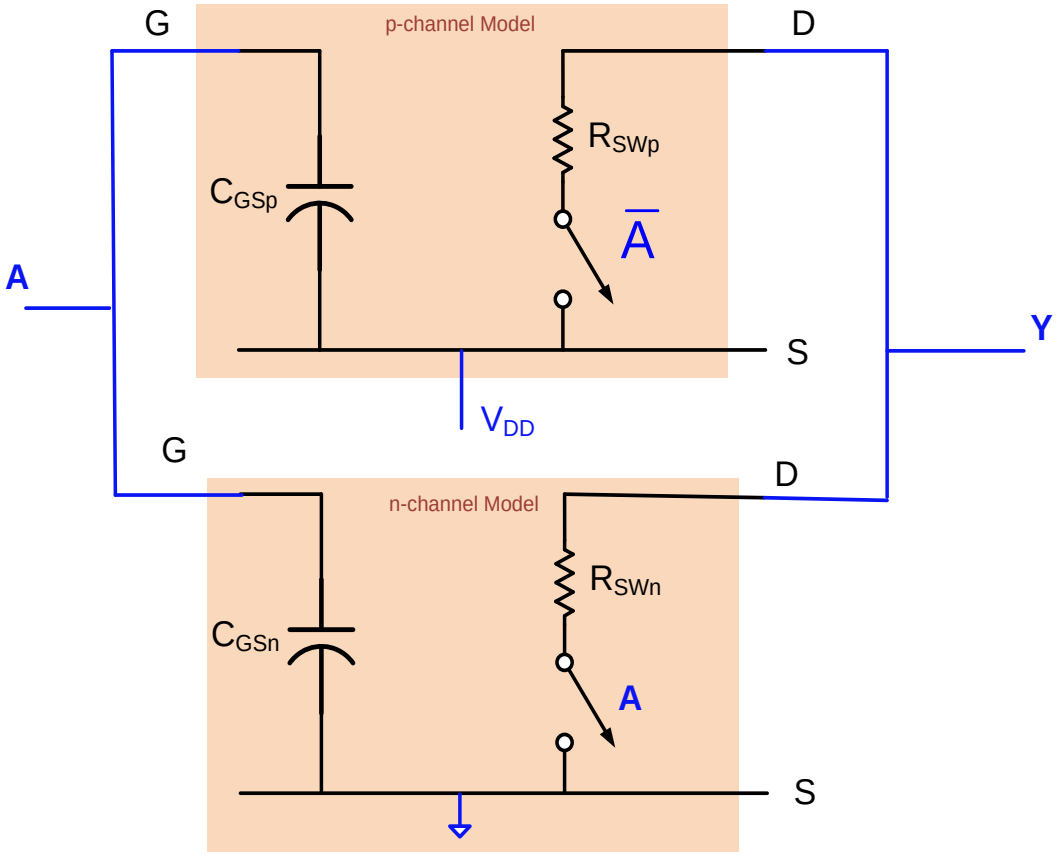
Example (cont)

Inverter with improved model

Inverter Model

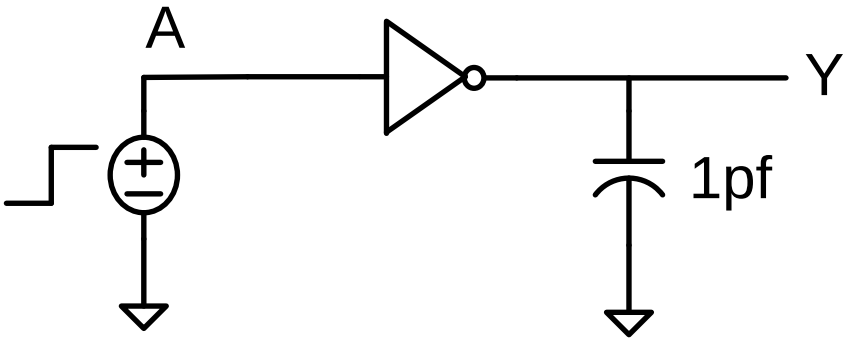


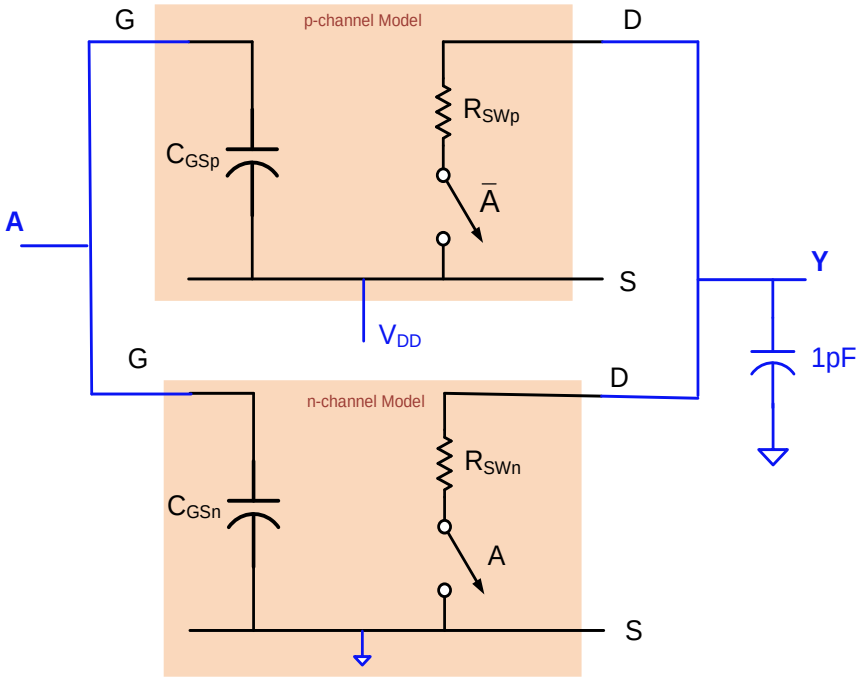
Inverter with Improved Model



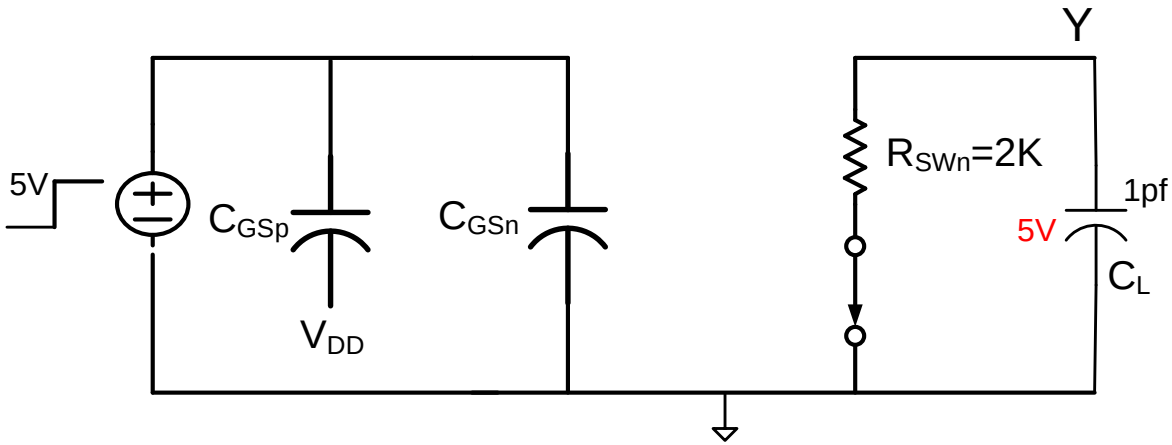
Example (cont)

With improved model $t_{HL}=?$





To initiate a HL output transisition, assume Y has been in the high state for a long time and lower switch closes at time t=0

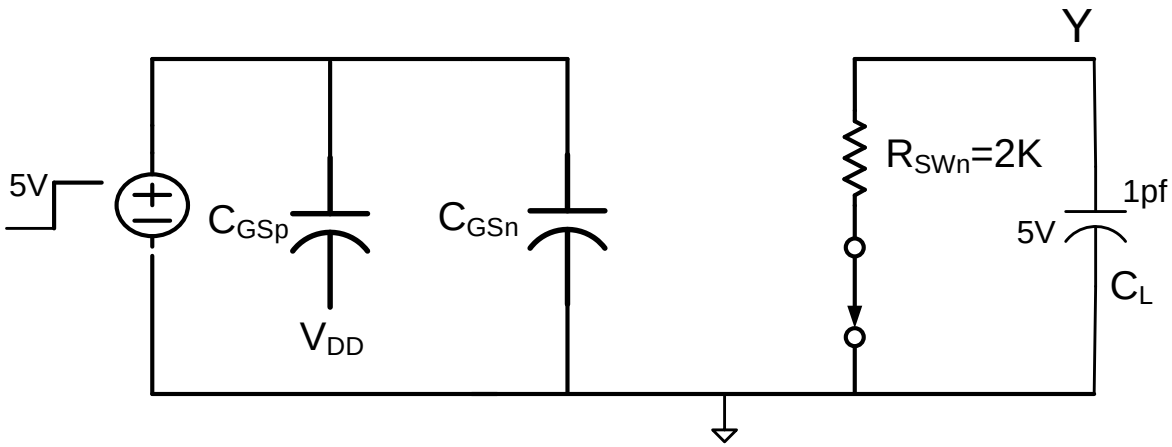


5V is the initial condition on C_L

Example (cont)

With improved model

t_{HL}=?



Recognize circuit as a first-order RC network

Recall: Step response of any first-order network with LHP pole can be written as

$$y(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

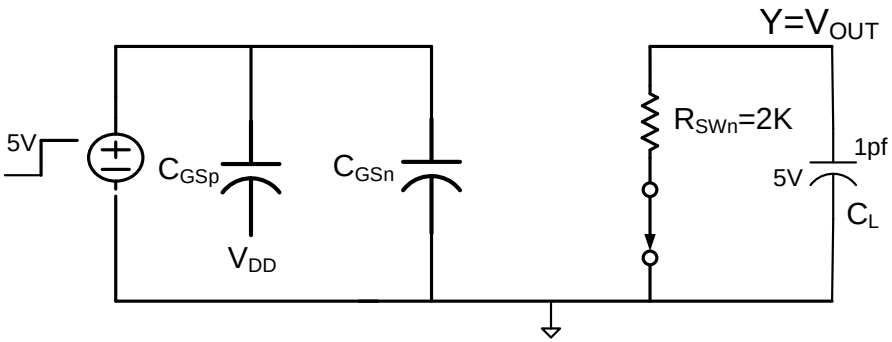
where F is the final value, I is the initial value and τ is the time constant of the circuit

(from Chapter 7 of Nilsson and Riedel)

For the circuit above, F=0, I=5 and $\tau = R_{SWn}C_L$

Example (cont)

With improved model

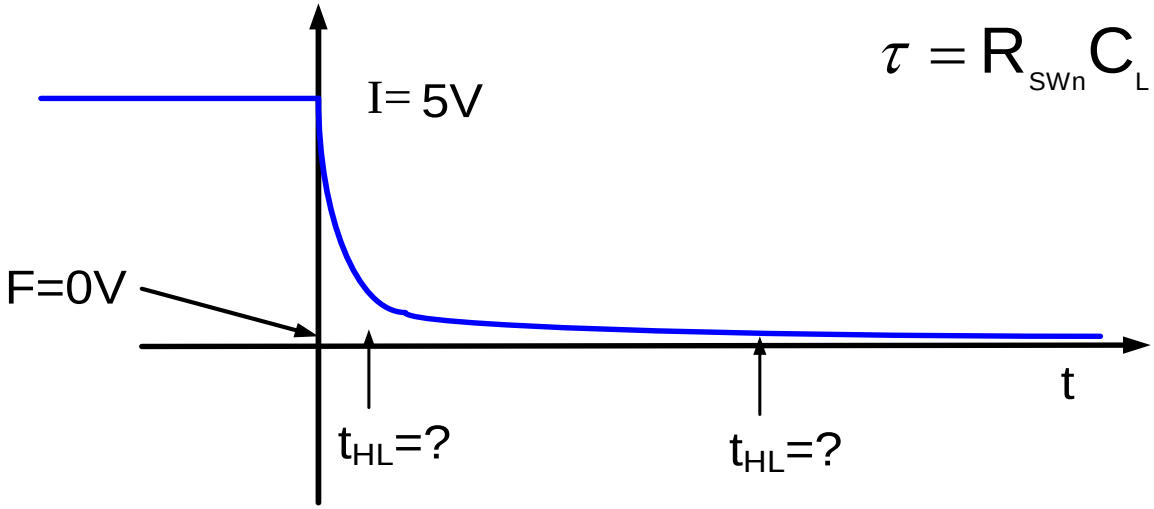


$t_{HL}=?$

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

$$V_{OUT}(t) = 5e^{-\frac{t}{\tau}}$$

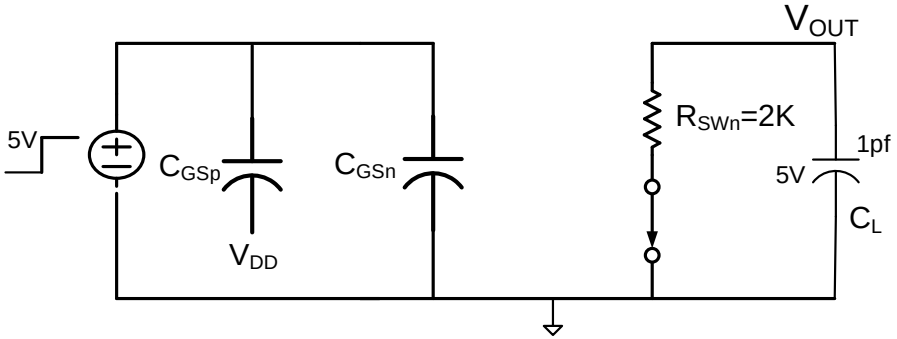
$$\tau = R_{SWn}C_L$$

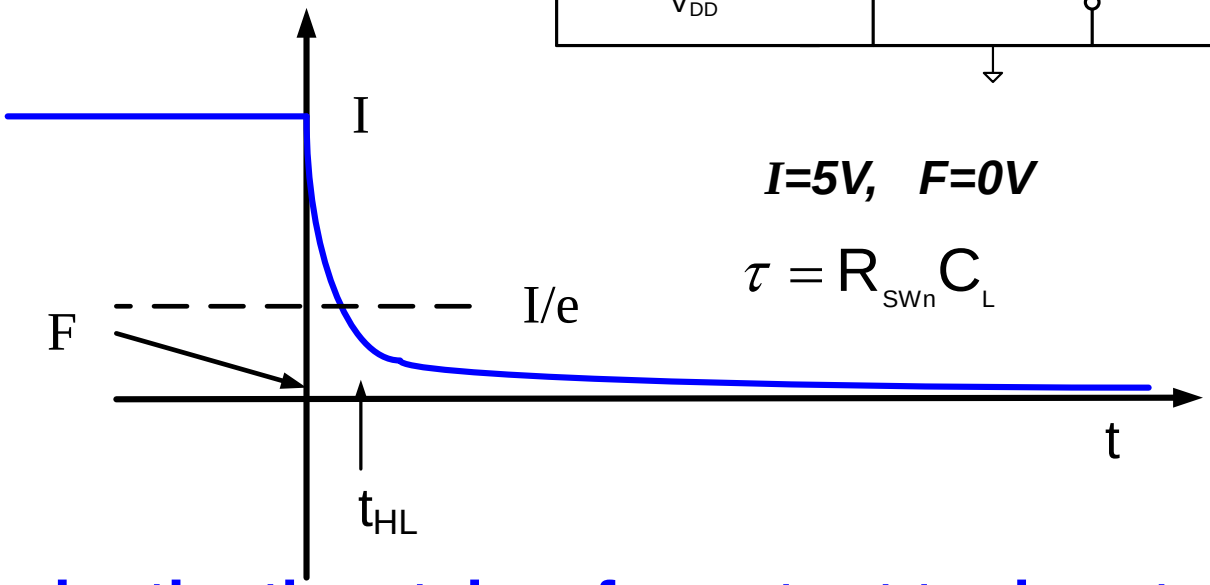


how is t_{HL} defined?

Example (cont)

t_{HL}=?





I=5V, F=0V

$$\tau = R_{SWn} C_L$$

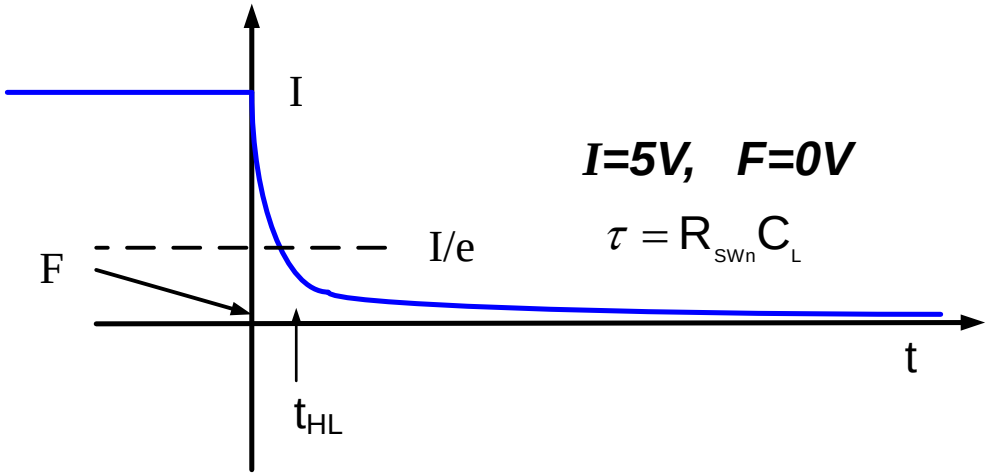
Define t_{HL} to be the time taken for output to drop to I/e

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}} \quad \Rightarrow \quad \frac{I}{e} = F + (I - F)e^{-\frac{t_{HL}}{\tau}}$$

t_{HL} as defined here has proved useful at analytically predicting response time of circuits

Example (cont)

With improved model



$$\frac{I}{e} = F + (I - F) e^{-\frac{t_{HL}}{\tau}}$$

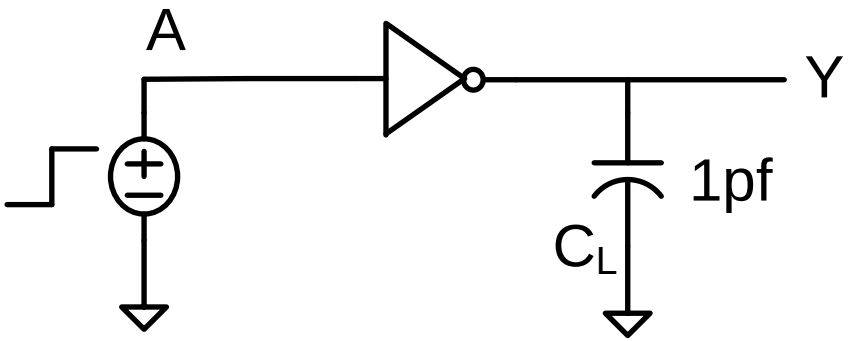
$$\frac{I}{e} = I e^{-\frac{t_{HL}}{\tau}}$$

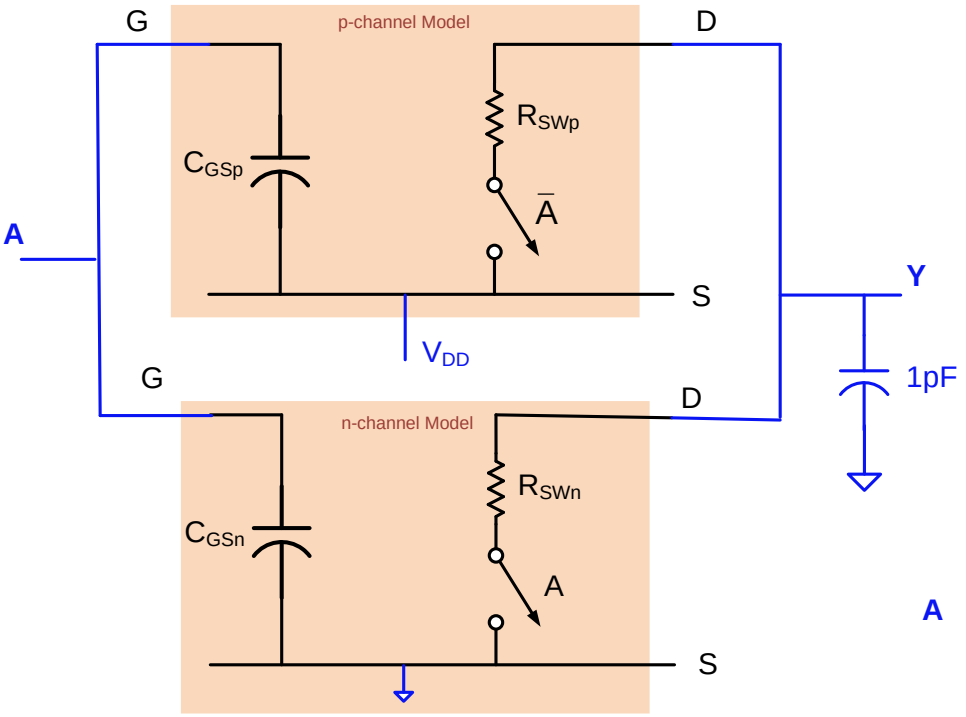
$$\frac{1}{e} = e^{-\frac{t_{HL}}{\tau}}$$

$$t_{HL} = \tau \quad \Rightarrow \quad t_{HL} = R_{SWn} C_L$$

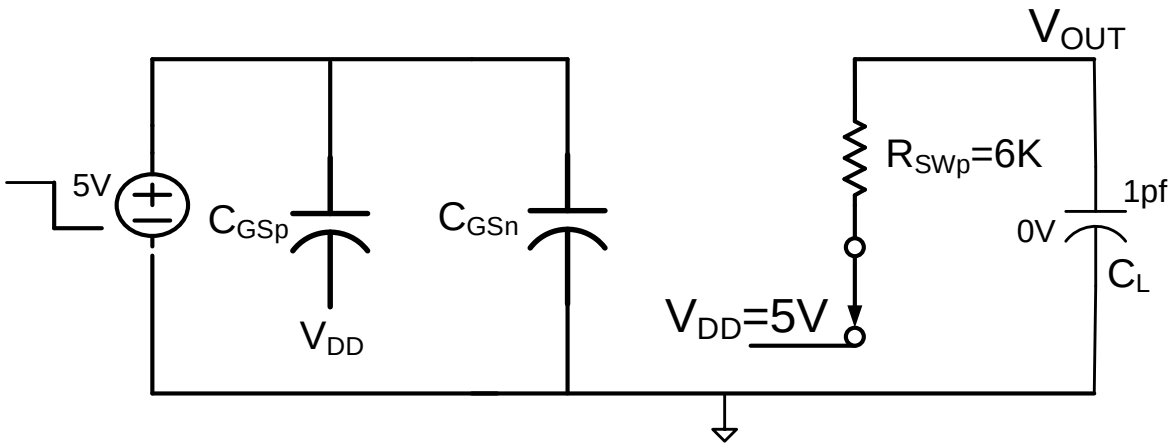
Example (cont)

With improved model $t_{LH}=?$





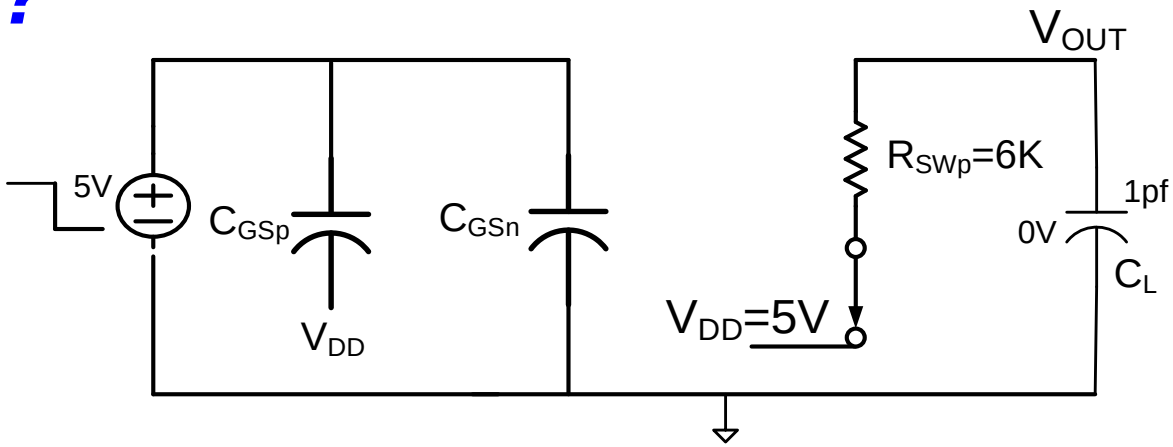
Assume output in low state for a long time and upper switch closes at time t=0



0V is the initial condition on C_L

Example (cont)

With improved model $t_{LH}=?$

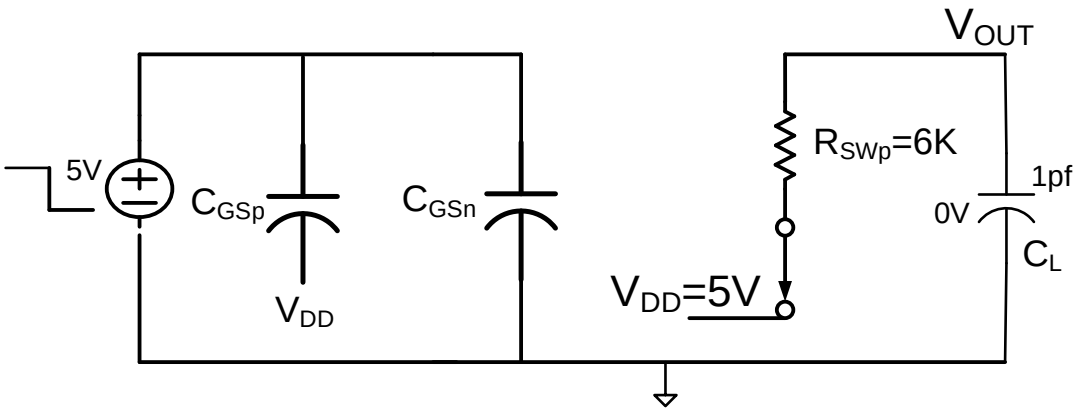


$$y(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

For this circuit, F=5, I=0 and $\tau = R_{SWp}C_L$

Example (cont)

With improved model

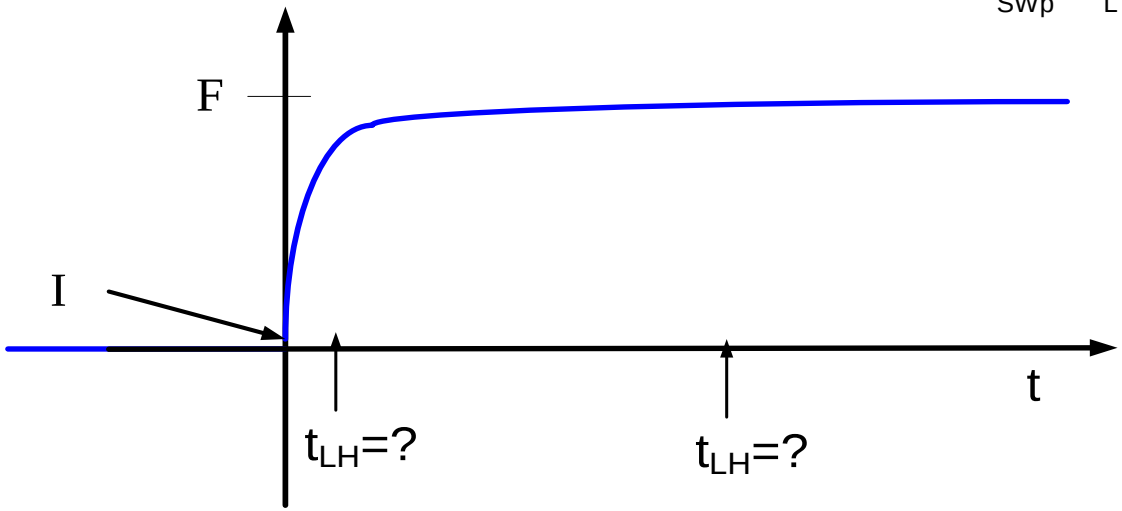


t_{LH}=?

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}}$$

$$V_{OUT}(t) = 5\left(1 - e^{-\frac{t}{\tau}}\right)$$

$$\tau = R_{SWp}C_L$$

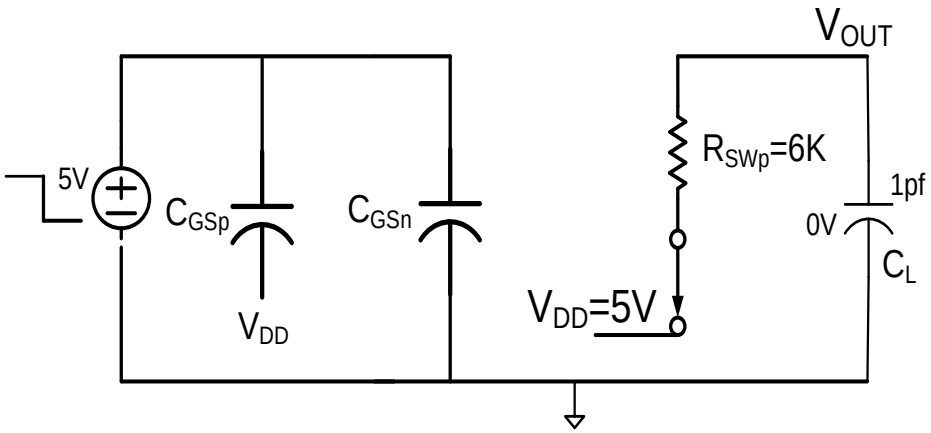


how is t_{LH} defined?

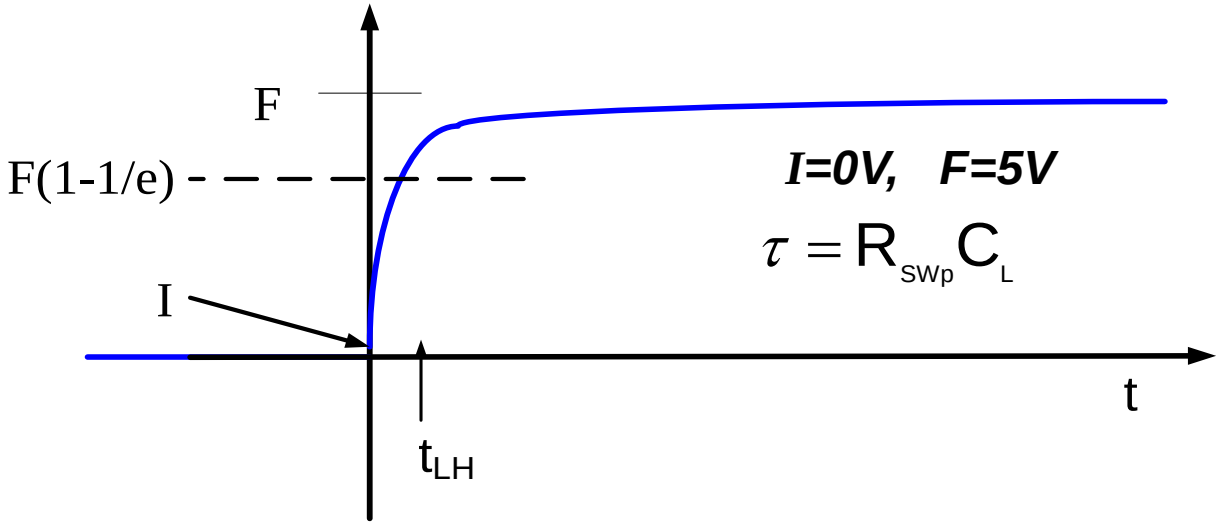
Example (cont)

With improved model

t_{LH}=?



Define t_{LH} as shown on figure



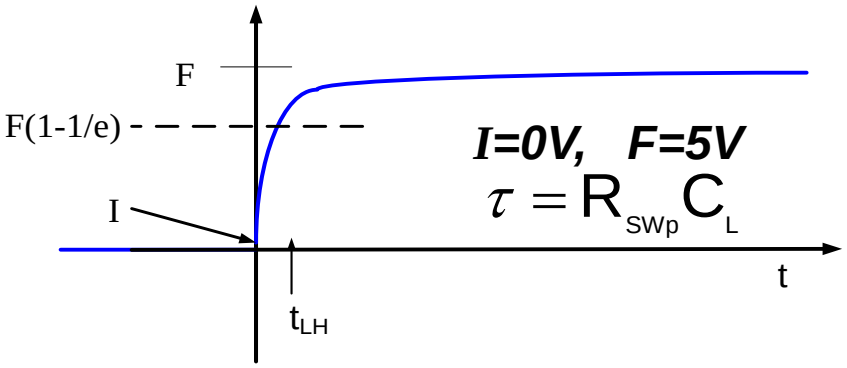
t_{LH} as defined has proven useful for analytically predicting response time of circuits

$$V_{OUT}(t) = F + (I - F)e^{-\frac{t}{\tau}} \quad \Rightarrow \quad F\left(1 - \frac{1}{e}\right) = F + (I - F)e^{-\frac{t_{LH}}{\tau}}$$

Example (cont)

With improved model

t_{LH}=?



$$F\left(1-\frac{1}{e}\right)=F+\left(I-F\right)e^{-\frac{t_{LH}}{\tau}}$$

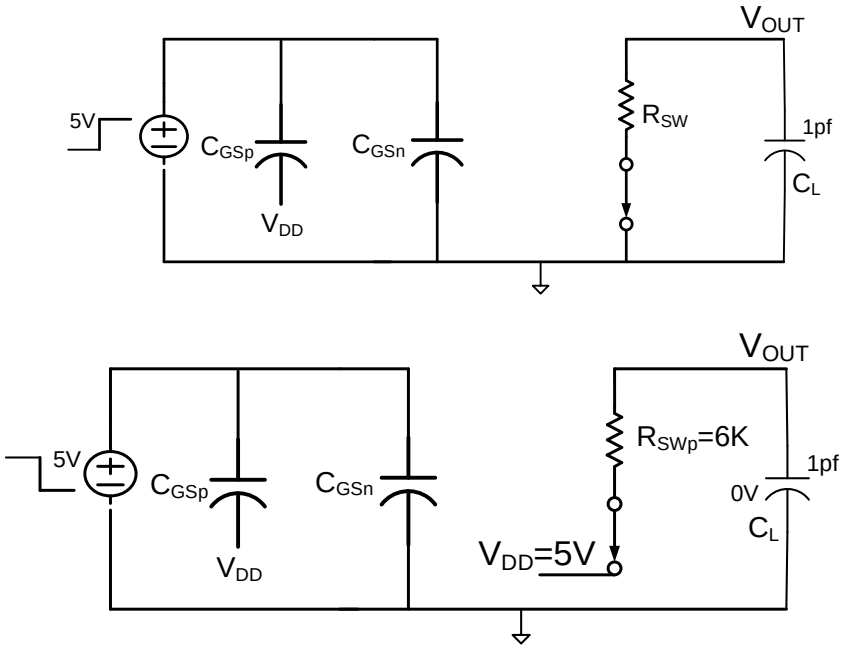
$$F\left(1-\frac{1}{e}\right)=F+\left(F\right)e^{-\frac{t_{LH}}{\tau}}$$

$$1-\frac{1}{e}=1+e^{-\frac{t_{LH}}{\tau}}$$

$$t_{LH}=\tau \quad \Rightarrow \quad t_{LH}=R_{SWp}C_L$$

Example (cont)

With improved model



In the ON 0.5u process

$$t_{HL} \cong R_{SWn} C_L = 2K \bullet 1pF = 2n\,\text{sec}$$

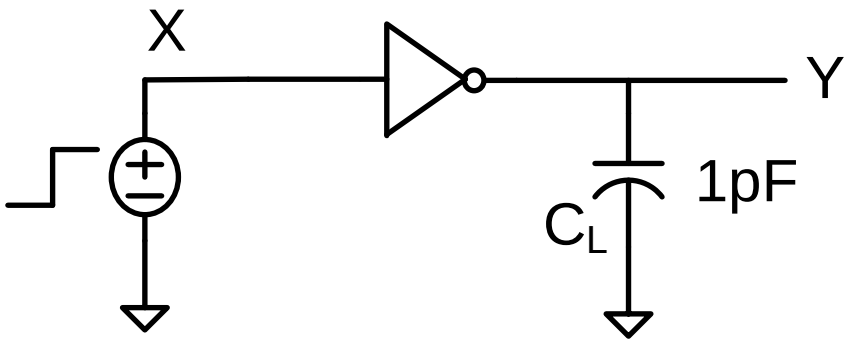
$$t_{LH} \cong R_{SWp} C_L = 6K \bullet 1pF = 6n\,\text{sec}$$

Note this circuit is quite fast !

Note that t_{HL} is much shorter than t_{LH}

Often C_L will be even smaller and the circuit will be much faster !!

Summary: What is the delay of a minimum-sized inverter driving a 1pF load?

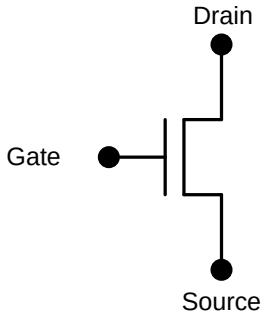


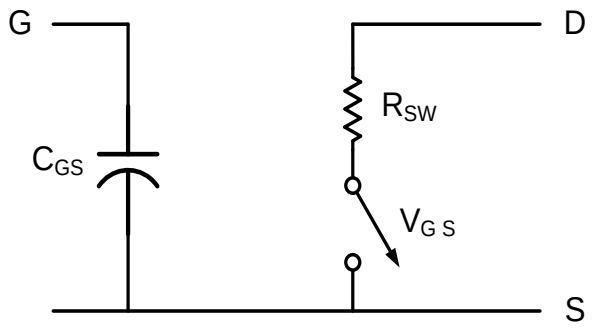
In the ON 0.5µ process

$$t_{HL} \cong R_{SWn} C_L = 2K \bullet 1pF = 2n\,\text{sec}$$

$$t_{LH} \cong R_{SWp} C_L = 6K \bullet 1pF = 6n\,\text{sec}$$

Improved switch-level model





Switch closed for V_{GS}= large

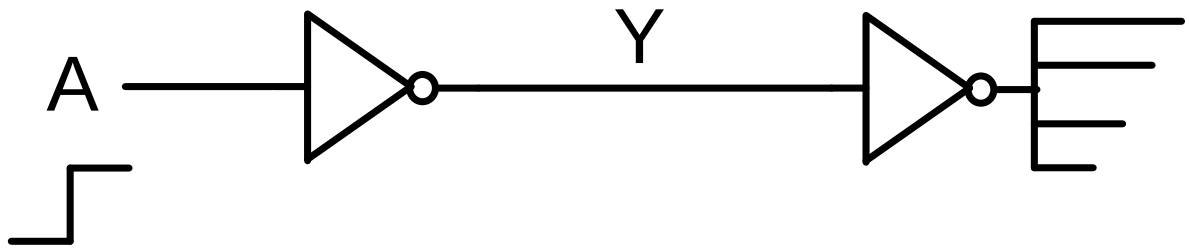
Switch open for V_{GS}= small

- **Previous example showed why R_{SW} in the model was important**
- **But of what use is the C_{GS} which did not enter the previous calculations?**

For minimum-sized devices in a 0.5μ process

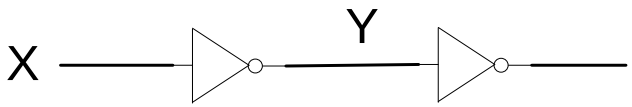
$$C_{GS} \cong 1.5fF \qquad R_{sw} \cong \begin{cases} 2K\Omega \ \ n-channel \\ 6K\Omega \ \ p-channel \end{cases}$$

One gate often drives one or more other gates !



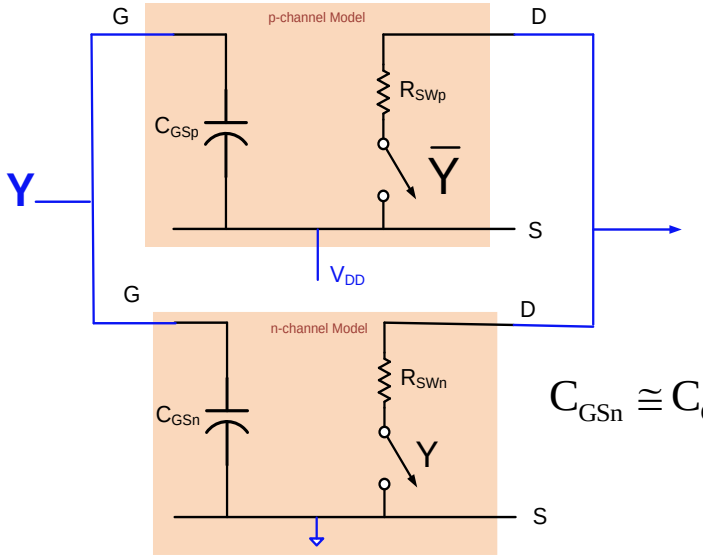
What are t_{HL} and t_{LH}?

Example: What is the delay of a minimum-sized inverter driving another identical device?



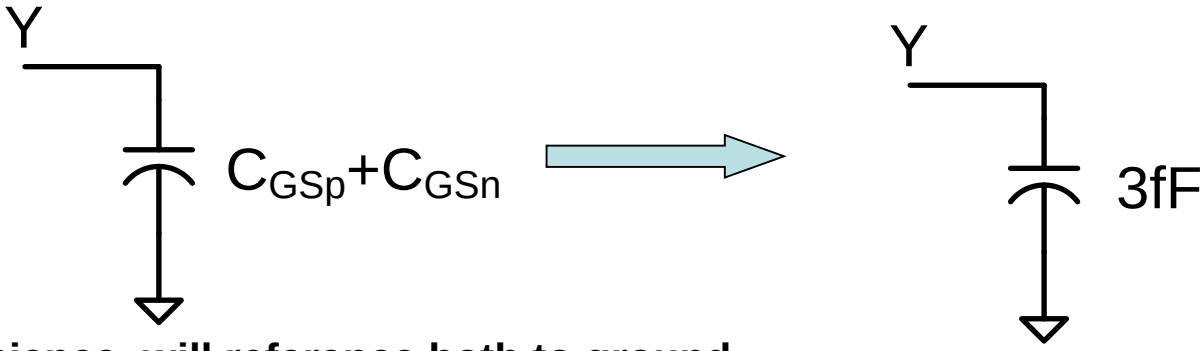
Load on first inverter

C_{GSn} and C_{GSp} both 1.5fF



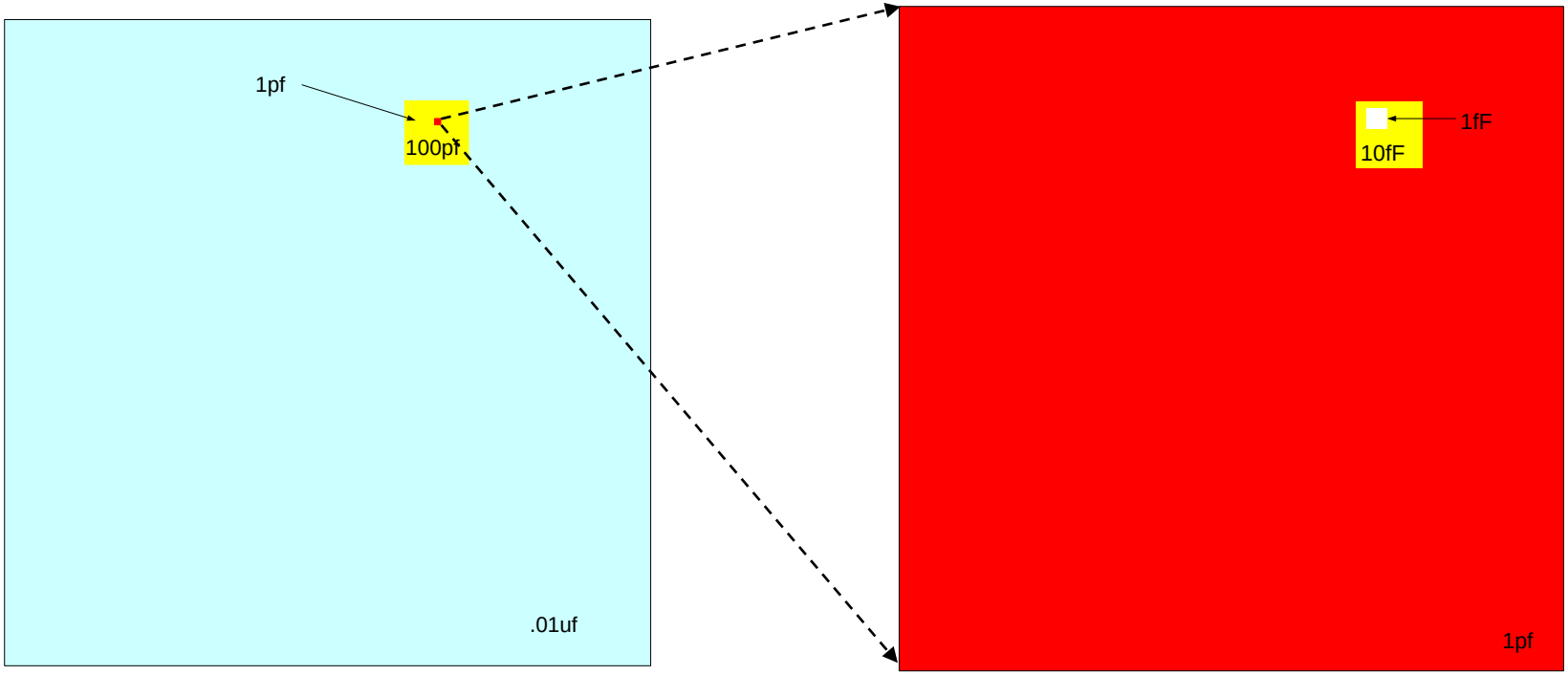
$C_{GSn} \cong C_{GSp} \cong 1.5\text{fF}$

Loading effects same whether C_{GSp} and/or C_{GSn} connected to V_{DD} or GND



For convenience, will reference both to ground

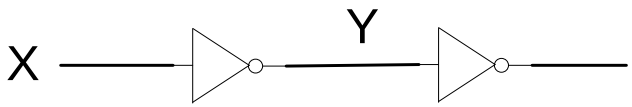
Is a capacitor of 1.5fF small enough to be neglected?

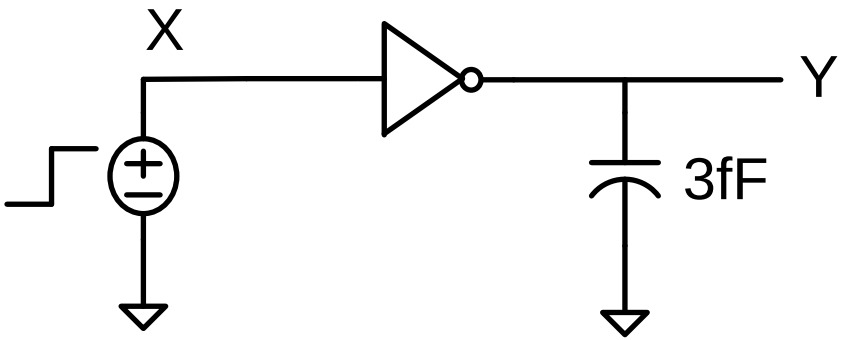


Area allocations shown to relative scale:

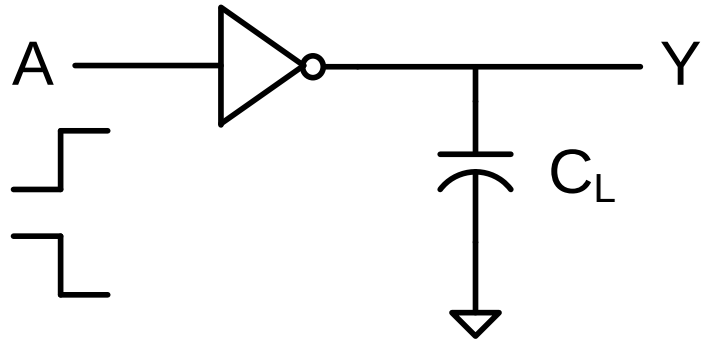
- **This example will provide insight into the answer of the question**

Example: What is the delay of a minimum-sized inverter driving another identical device? Assume V_{DD}=5V





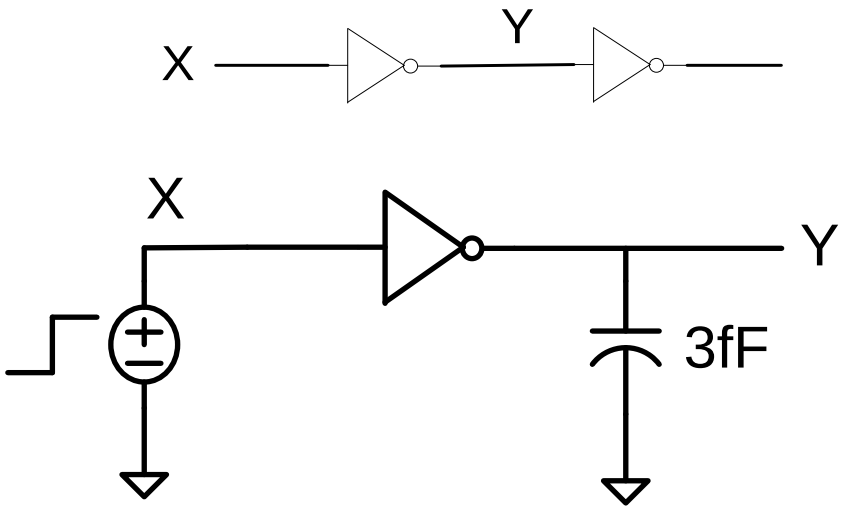
Generalizing the Previous Analysis to Arbitrary Load



$$t_{HL} \cong R_{SWn} C_L$$

$$t_{LH} \cong R_{SWp} C_L$$

Example: What is the delay of a minimum-sized inverter driving another identical device?



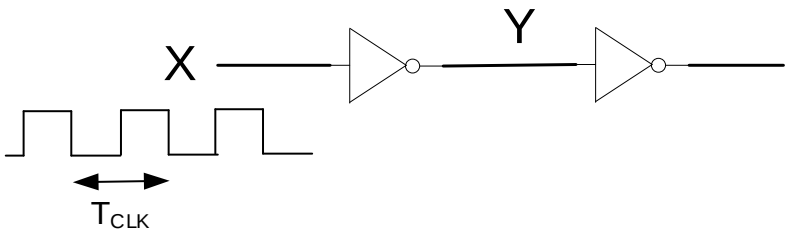
$$t_{HL} \cong R_{SWn} C_L = 2K \bullet 3fF = 6p\,\text{sec}$$

$$t_{LH} \cong R_{SWp} C_L = 6K \bullet 3fF = 18p\,\text{sec}$$

Do gates really operate this fast?

What would be the maximum clock rate for acceptable operation?

Example: What is the delay of a minimum-sized inverter driving another identical device?



$$t_{HL} \cong R_{SWn} C_L = 6p\,\text{sec}$$

$$t_{LH} \cong R_{SWp} C_L = 18p\,\text{sec}$$

What would be the maximum clock rate for acceptable operation?

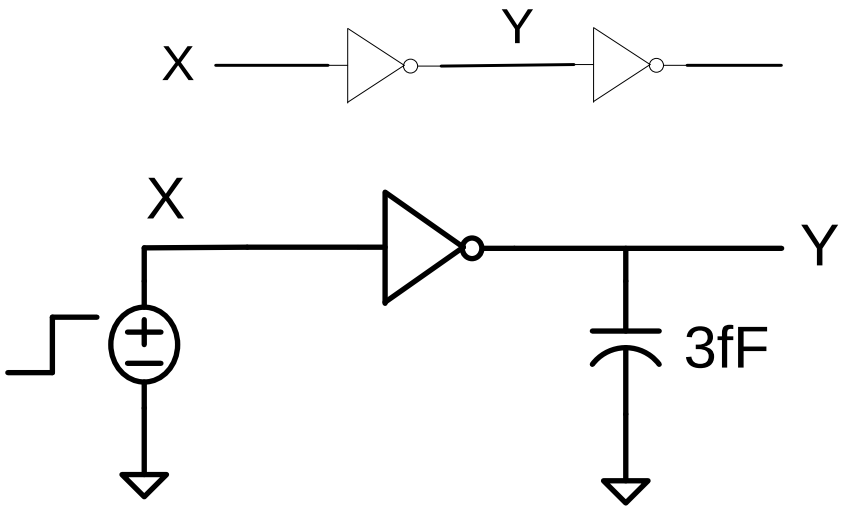
$$T_{CLK\text{-}min} = t_{HL} + t_{LH}$$

$$f_{CLK\text{-}max} = \frac{1}{T_{CLK\text{-}min}} = \frac{1}{24\text{psec}} = 40\text{GHz}$$

And much faster in a finer feature process !! **??????**

What would be the implications of allowing for 10 levels of logic and 10 loads (FanOut=10)?

Example: What is the delay of a minimum-sized inverter driving another identical device? SUMMARY

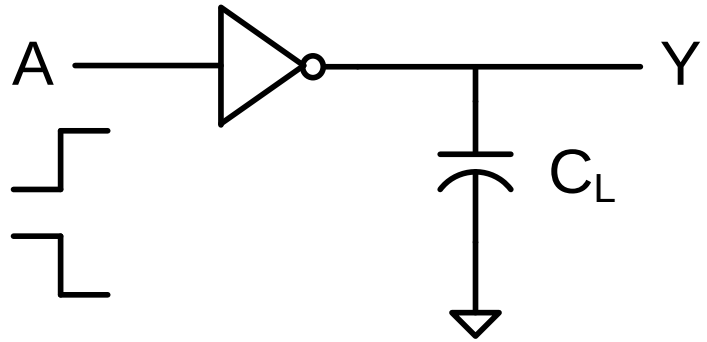


$$t_{HL} \cong R_{SWn} C_L \qquad = 2K \bullet 3\,fF = 6\,p\,\text{sec}$$

$$t_{LH} \cong R_{SWp} C_L \qquad = 6K \bullet 3\,fF = 18\,p\,\text{sec}$$

Note this is very fast but even the small 1.5fF capacitors are not negligable !

Response time of logic gates



$$t_{HL} \cong R_{SWn} C_L$$

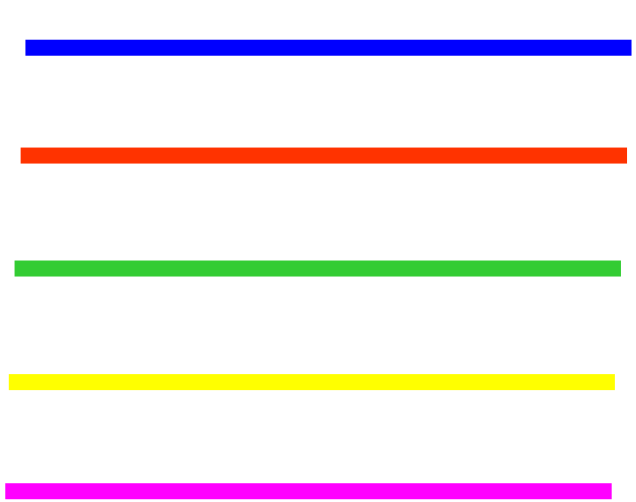
$$t_{LH} \cong R_{SWp} C_L$$

- **Logic Circuits can operate very fast**
- **Extremely small parasitic capacitances play key role in speed of a circuit**

Stick Diagrams

- It is often necessary to obtain information about placement, interconnect and physical-layer structure
- Stick diagrams are often used for small component-count blocks
- Approximate placement, routing, and area information can be obtained rather quickly with the use of stick diagrams

Stick Diagrams



Metal 1

poly

n-diffusion

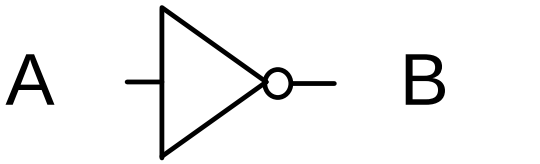
p-diffusion

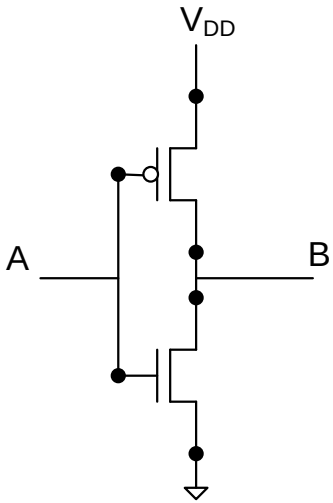
Metal 2

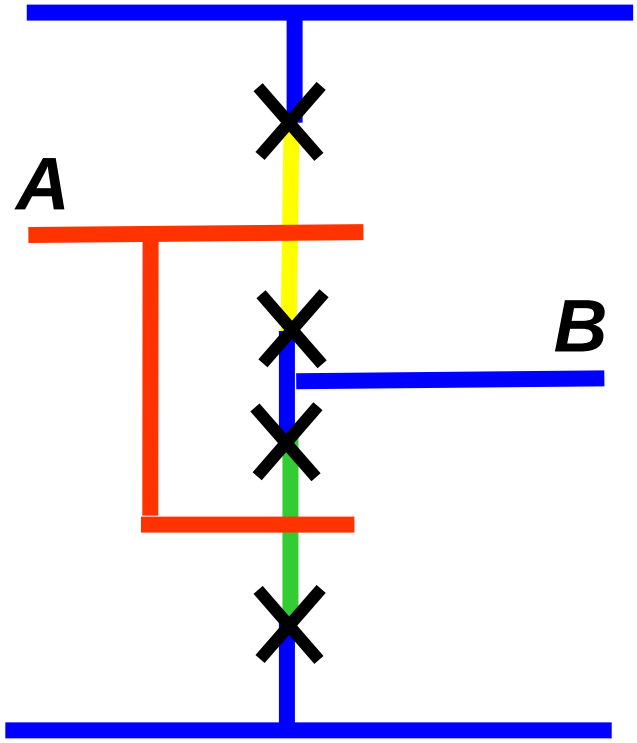
Contact

Additional layers can be added and color conventions are peronal

Stick Diagram





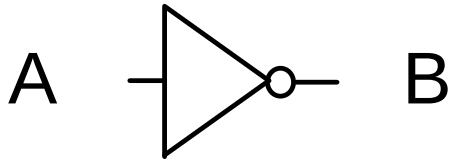


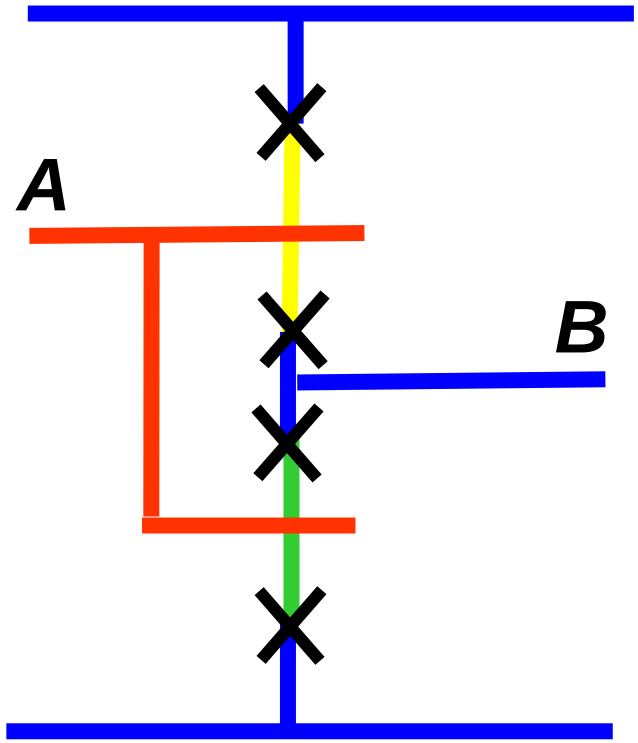
A stick diagram is not a layout but gives the basic structure (including location,, orientation and interconnects) that will be instantiated in the actual layout itself

Modifications can be made much more quickly on a stick diagram than on a layout

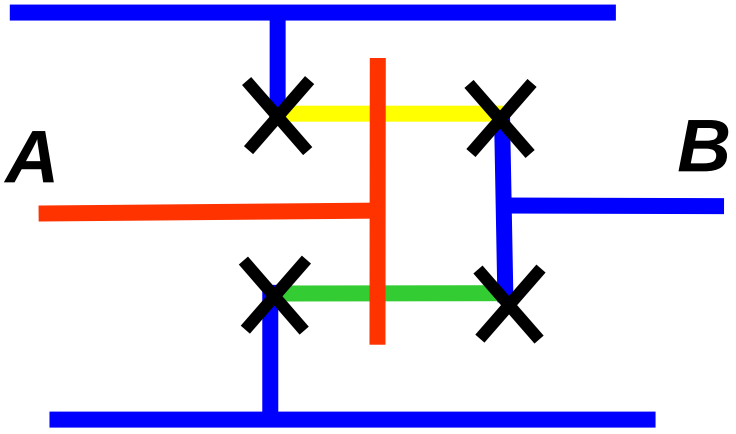
Iteration may be needed to come up with a good layout structure

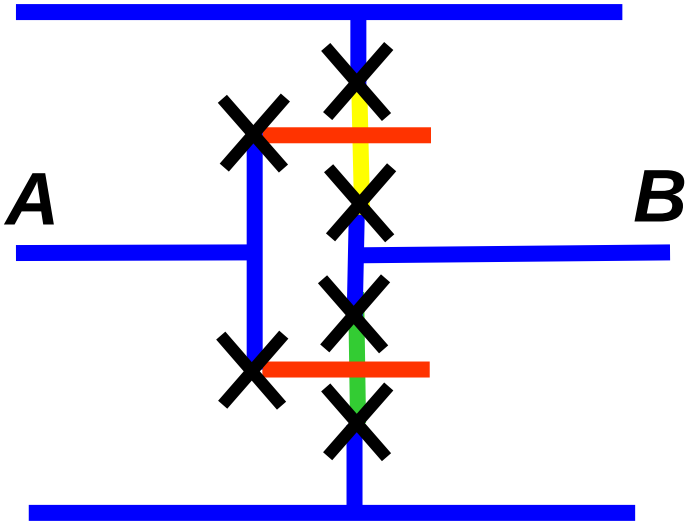
Stick Diagram

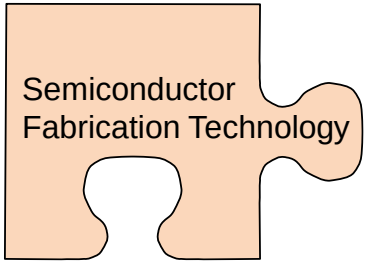




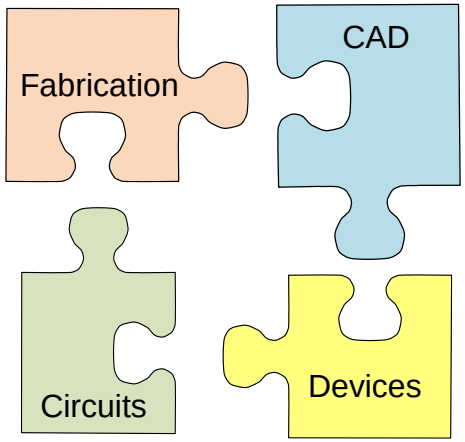
Alternate Representations







Technology Files



- Provide Information About Process
  - Process Flow (Fabrication Technology)
  - Model Parameters
  - Design Rules
- Serve as Interface Between Design Engineer and Process Engineer
- Insist on getting information that is deemed important for a design
  - Limited information available in academia
  - Foundries often sensitive to who gets access to information
  - Customer success and satisfaction is critical to foundries

Technology Files

- Design Rules
- Process Flow (Fabrication Technology) (will discuss next)
- Model Parameters (will discuss in substantially more detail after device operation and more advanced models are introduced)

First – A preview of what the technology files look like !

Typical Design Rules

TABLE 2B.2
Design rules for a typical p-well CMOS process
(See Table 2B.3 in color plates for graphical interpretation)

		Dimensions	
		Microns	**Scalable**
1.	p-well (CIF Brown, Mask #1[a])		
1.1	Width	5	4λ
1.2	Spacing (different potential)	15	10λ
1.3	Spacing (same potential)	9	6λ
2.	Active (CIF Green, Mask #2)		
2.1	Width	4	2λ
2.2	Spacing	4	2λ
2.3	p^+ active in n-subs to p-well edge	8	6λ
2.4	n^+ active in n-subs to p-well edge	7	5λ
2.5	n^+ active in p-well to p-well edge	4	2λ
2.6	p^+ active in p-well to p-well edge	1	λ
3.	Poly (POLY I) (CIF Red, Mask #3)		
3.1	Width	3	2λ
3.2	Spacing	3	2λ
3.3	Field poly to active	2	λ
3.4	Poly overlap of active	3	2λ
3.5	Active overlap of poly	4	2λ
4.	p^+ select (CIF Orange, Mask #4)		
4.1	Overlap of active	2	λ
4.2	Space to n^+ active	2	λ
4.3	Overlap of channel[b]	3.5	2λ
4.4	Space to channel[b]	3.5	2λ
4.5	Space to p^+ select	3	2λ
4.6	Width	3	2λ

Typical Design Rules (cont)

5.	Contact[c] (CIF Purple, Mask #6)			
	5.1	Square contact, exactly	3 × 3	2λ × 2λ
	5.2	Rectangular contact, exactly	3 × 8	2λ × 6λ
	5.3	Space to different contact	3	2λ
	5.4	Poly overlap of contact	2	λ
	5.5	Poly overlap in direction of metal 1	2.5	2λ
	5.6	Space to channel	3	2λ
	5.7	Metal 1 overlap of contact	2	λ
	5.8	Active overlap of contact	2	λ
	5.9	p^+ select overlap of contact	3	2λ
	5.10	Subs./well shorting contact, exactly	3 × 8	2λ × 6λ
6.	Metal 1[d] (CIF Blue, Mask #7)			
	6.1	Width	3	2λ
	6.2	Spacing	4	3λ
	6.3	Maximum current density	0.8 mA/μ	0.8 mA/μ

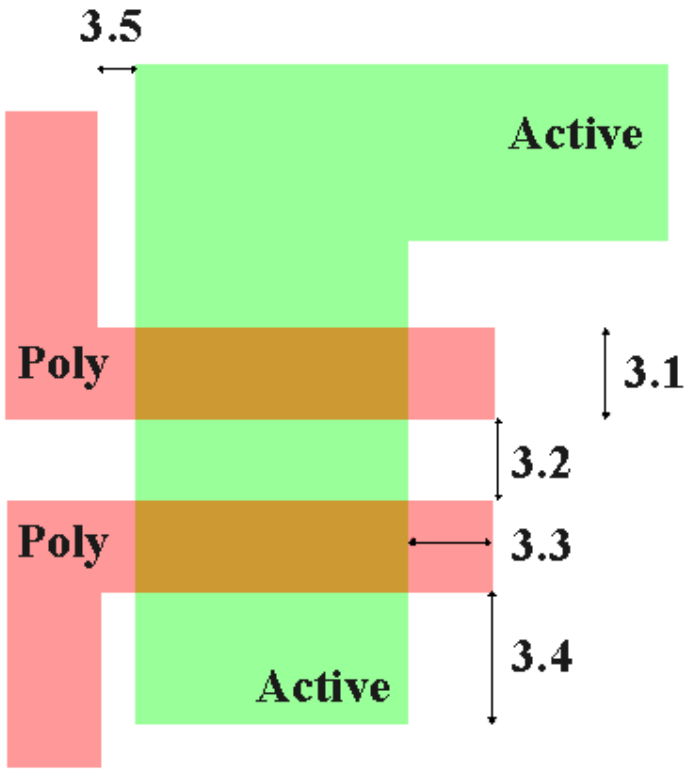
Typical Design Rules (cont)

7.	Via [e] (CIF Purple Hatched, Mask #C1)			
	7.1	Size, exactly	3 × 3	2λ × 2λ
	7.2	Separation	3	2λ
	7.3	Space to poly edge	4	2λ
	7.4	Space to contact	3	2λ
	7.5	Overlap by metal 1	2	λ
	7.6	Overlap by metal 2	2	λ
	7.7	Space to active edge	3	2λ
8.	Metal 2 (CIF Orange Hatched, Mask #C2)			
	8.1	Width	5	3λ
	8.2	Spacing	5	3λ
	8.3	Bonding pad size	100 × 100	100 μ × 100 μ
	8.4	Probe pad size	75 × 75	75 μ × 75 μ
	8.5	Bonding pad separation	50	50 μ
	8.6	Bonding to probe pad	30	30 μ
	8.7	Probe pad separation	30	30 μ
	8.8	Pad to circuitry	40	40 μ
	8.9	Maximum current density	0.8 mA/μ	0.8 mA/μ
9.	Passivation[f] (CIF Purple Dashed, Mask #8)			
	9.1	Bonding pad opening	90 × 90	90 μ × 90 μ
	9.2	Probe pad opening	65 × 65	65 μ × 65 μ
10.	Metal 2 crossing coincident metal 1 and poly[g]			
	10.1	Metal 1 to poly edge spacing when crossing metal 2	2	λ
	10.2	Rule domain	2	λ
11.	Electrode (POLY II)[h] (CIF Purple Hatched, Mask #A1)			
	11.1	Width	3	2λ
	11.2	Spacing	3	2λ
	11.3	POLY I overlap of POLY II	2	λ
	11.4	Space to contact	3	2λ

Typical Design Rules (cont)

SCMOS Layout Rules - Poly

Rule	Description	Lambda		
		SCMOS	SUBM	DEEP
3.1	Minimum width	2	2	2
3.2	Minimum spacing over field	2	3	3
3.2.a	Minimum spacing over active	2	3	4
3.3	Minimum gate extension of active	2	2	2.5
3.4	Minimum active extension of poly	3	3	4
3.5	Minimum field poly to active	1	1	1



Typical Process Description

Process scenario of major process steps in typical p-well CMOS process[a]

1.	Clean wafer	
2.	GROW THIN OXIDE	
3.	Apply photoresist	
4.	PATTERN P-WELL	(MASK #1)
5.	Develop photoresist	
6.	Deposit and diffuse p-type impurities	
7.	Strip photoresist	
8.	Strip thin oxide	
9.	Grow thin oxide	
10.	Apply layer of Si_3N_4	
11.	Apply photoresist	
12.	PATTERN Si_3N_4 (active area definition)	(MASK #2)
13.	Develop photoresist	
14.	Etch Si_3N_4	
15.	Strip photoresist	
	Optional field threshold voltage adjust	
	A.1 Apply photoresist	
	A.2 PATTERN ANTIMOAT IN SUBSTRATE	(MASK #A1)
	A.3 Develop photoresist	
	A.4 FIELD IMPLANT (n-type)	
	A.5 Strip photoresist	
16.	GROW FIELD OXIDE	
17.	Strip Si_3N_4	
18.	Strip thin oxide	
19.	GROW GATE OXIDE	
20.	POLYSILICON DEPOSITION (POLY I)	
21.	Apply photoresist	
22.	PATTERN POLYSILICON	(MASK #3)
23.	Develop photoresist	
24.	ETCH POLYSILICON	

Typical Process Description (cont)

25.	Strip photoresist	
	Optional steps for double polysilicon process	
	B.1 Strip thin oxide	
	B.2 GROW THIN OXIDE	
	B.3 POLYSILICON DEPOSITION (POLY II)	
	B.4 Apply photoresist	
	B.5 PATTERN POLYSILICON	(MASK #B1)
	B.6 Develop photoresist	
	B.7 ETCH POLYSILICON	
	B.8 Strip photoresist	
	B.9 Strip thin oxide	
26.	Apply photoresist	
27.	PATTERN P-CHANNEL DRAINS AND SOURCES AND P^+ GUARD RINGS (p-well ohmic contacts)	(MASK #4)
28.	Develop photoresist	
29.	p^+ IMPLANT	
30.	Strip photoresist	
31.	Apply photoresist	
32.	PATTERN N-CHANNEL DRAINS AND SOURCES AND N^+ GUARD RINGS (top ohmic contact to substrate)	(MASK #5)
33.	Develop photoresist	
34.	n^+ IMPLANT	
35.	Strip photoresist	
36.	Strip thin oxide	
37.	Grow oxide	
38.	Apply photoresist	
39.	PATTERN CONTACT OPENINGS	(MASK #6)
40.	Develop photoresist	
41.	Etch oxide	
42.	Strip photoresist	
43.	APPLY METAL	
44.	Apply photoresist	
45.	PATTERN METAL	(MASK #7)
46.	Develop photoresist	
47.	Etch metal	

Typical Process Description (cont)

48.	Strip photoresist	
	Optional steps for double metal process	
	C.1 Strip thin oxide	
	C.2 DEPOSIT INTERMETAL OXIDE	
	C.3 Apply photoresist	
	C.4 PATTERN VIAS	(MASK #C1)
	C.5 Develop photoresist	
	C.6 Etch oxide	
	C.7 Strip photoresist	
	C.8 APPLY METAL (Metal 2)	
	C.9 Apply photoresist	
	C.10 PATTERN METAL	(MASK #C2)
	C.11 Develop photoresist	
	C.12 Etch metal	
	C.13 Strip photoresist	
49.	APPLY PASSIVATION	
50.	Apply photoresist	
51.	PATTERN PAD OPENINGS	(MASK #8)
52.	Develop photoresist	
53.	Etch passivation	
54.	Strip photoresist	
55.	ASSEMBLE, PACKAGE AND TEST	

Typical Model Parameters

Process parameters for a typical[a] p-well CMOS process

	Typical	Tolerance[b]	Units
Square law model parameters			
V_{T0} (threshold voltage)			
n-channel (V_{TN0})	0.75	± 0.25	V
p-channel (V_{TP0})	−0.75	± 0.25	V
K'(conduction factor)			
n-channel	24	± 6	$\mu A/V^2$
p-channel	8	± 1.5	$\mu A/V^2$
γ(body effect)			
n-channel	0.8	± 0.4	$V^{1/2}$
p-channel	0.4	± 0.2	$V^{1/2}$
λ(channel length modulation)			
n-channel	0.01	± 50%	V^{-1}
p-channel	0.02	± 50%	V^{-1}
ϕ(surface potential)			
n- and p-channel	0.6	± 0.1	V
Process parameters			
μ (channel mobility)			
n-channel	710		$cm^2/(V \cdot s)$
p-channel	230		$cm^2/(V \cdot s)$
Doping[c]			
n^+ active	5	±4	$10^{18}/cm^3$
p^+ active	5	±4	$10^{17}/cm^3$
p-well	5	±2	$10^{16}/cm^3$
n-substrate	1	±0.1	$10^{16}/cm^3$

Typical Model Parameters (cont)

Physical feature sizes			
T_{ox} (gate oxide thickness)	500	± 100	Å
Total lateral diffusion			
n-channel	0.45	± 0.15	μ
p-channel	0.6	± 0.3	μ
Diffusion depth			
n^+ diffusion	0.45	± 0.15	μ
p^+ diffusion	0.6	± 0.3	μ
p-well	3.0	± 30%	μ
Insulating layer separation			
POLY I to POLY II	800	± 100	Å
Metal 1 to Substrate	1.55	± 0.15	μ
Metal 1 to Diffusion	0.925	± 0.25	μ
POLY I to Substrate (POLY I on field oxide)	0.75	± 0.1	μ
Metal 1 to POLY I	0.87	± 0.7	μ
Metal 2 to Substrate	2.7	± 0.25	μ
Metal 2 to Metal I	1.2	± 0.1	μ
Metal 2 to POLY I	2.0	± 0.07	μ

Typical Model Parameters (cont)

Capacitances[d]			
C_{OX} (gate oxide capacitance, n- and p-channel)	0.7	±0.1	fF/μ^2
POLY I to substrate, poly in field	0.045	±0.01	fF/μ^2
POLY II to substrate, poly in field	0.045	±0.01	fF/μ^2
Metal 1 to substrate, metal in field	0.025	±0.005	fF/μ^2
Metal 2 to substrate, metal in field	0.014	±0.002	fF/μ^2
POLY I to POLY II	0.44	±0.05	fF/μ^2
POLY I to Metal 1	0.04	±0.01	fF/μ^2
POLY I to Metal 2	0.039	±0.003	fF/μ^2
Metal 1 to Metal 2	0.035	±0.01	fF/μ^2
Metal 1 to diffusion	0.04	±0.01	fF/μ^2
Metal 2 to diffusion	0.02	±0.005	fF/μ^2
n^+ diffusion to p-well (junction, bottom)	0.33	±0.17	fF/μ^2
n^+ diffusion sidewall (junction, sidewall)	2.6	±0.6	fF/μ
p^+ diffusion to substrate (junction, bottom)	0.38	±0.12	fF/μ^2
p^+ diffusion sidewall (junction, sidewall)	3.5	±2.0	fF/μ
p-well to substrate (junction, bottom)	0.2	±0.1	fF/μ^2
p-well sidewall (junction, sidewall)	1.6	±1.0	fF/μ
Resistances			
Substrate	25	±20%	Ω-cm
p-well	5000	±2500	Ω/□
n^+ diffusion	35	±25	Ω/□
p^+ diffusion	80	±55	Ω/□
Metal	0.003	±25%	Ω/□
Poly	25	±25%	Ω/□
Metal 1–Metal 2 via (3 μ × 3 μ contact)	<0.1		Ω
Metal 1 contact to POLY I (3 μ × 3 μ contact)	<10		Ω
Metal 1 contact to n^+ or p^+ diffusion (3 μ × 3 μ contact)	<5		Ω

Typical Model Parameters (cont)

Breakdown voltages, leakage currents, migration currents and operating conditions

Punchthrough voltages (Gate oxide, POLY I to POLY II)	>10	V
Diffusion reverse breakdown voltage	>10	V
p-well to substrate reverse breakdown voltage	>20	V
Metal 1 in field threshold voltage	>10	V
Metal 2 in field threshold voltage	>10	V
Poly-field threshold voltage	>10	V
Maximum operating voltage	7.0	V
n^+ diffusion to p-well leakage current	0.25	fA/μ^2
p^+ diffusion to substrate leakage current	0.25	fA/μ^2
p-well leakage current	0.25	fA/μ^2
Maximum metal current density	0.8	mA/μ width
Maximum device operating temperature	200	°C

Typical Model Parameters (cont)

SPICE MOSFET model parameters of a typical p-well CMOS process (MOSIS[a])

Parameter (Level 2 model)	n-channel	p-channel	Units
VTO	0.827	−0.895	V
KP	32.87	15.26	$\mu A/V^2$
GAMMA	1.36	0.879	$V^{1/2}$
PHI	0.6	0.6	V
LAMBDA	1.605E−2	4.709E−2	V^{-1}
CGSO	5.2E−4	4.0E−4	fF/μ width
CGDO	5.2E−4	4.0E−4	fF/μ width
RSH	25	95	$\Omega/\square$
CJ	3.2E−4	2.0E−4	fF/μ^2
MJ	0.5	0.5	
CJSW	9.0E−4	4.5E−4	fF/μ perimeter
MJSW	0.33	0.33	
TOX	500	500	Å
NSUB	1.0E16	1.12E14	$1/cm^3$
NSS	0	0	$1/cm^2$
NFS	1.235E12	8.79E11	$1/cm^2$
TPG	1	−1	
XJ	0.4	0.4	μ
LD	0.28	0.28	μ
UO	200	100	$cm^2/(V \cdot s)$
UCRIT	9.99E5	1.64E4	V/cm
UEXP	1.001E−3	0.1534	
VMAX	1.0E5	1.0E5	m/s
NEFF	1.001E−2	1.001E−2	
DELTA	1.2405	1.938	

Typical Model Parameters (cont)

```
.MODEL CMOSN NMOS (                                                 LEVEL    = 49
+VERSION = 3.1            TNOM     = 27             TOX      = 1.4E-8
+XJ      = 1.5E-7         NCH      = 1.7E17         VTHO     = 0.6656437
+K1      = 0.875093       K2       = -0.0943223     K3       = 25.0562441
+K3B     = -8.5140476     WO       = 1.01582E-8     NLX      = 1E-9
+DVTOW   = 0              DVT1W    = 0              DVT2W    = 0
+DVTO    = 2.670658       DVT1     = 0.4282172      DVT2     = -0.1373089
+UO      = 452.3081836    UA       = 3.061716E-13   UB       = 1.515137E-18
+UC      = 1.166279E-11   VSAT     = 1.682414E5     A0       = 0.6297744
+AGS     = 0.1384489      B0       = 2.579158E-6    B1       = 5E-6
+KETA    = -3.615287E-3   A1       = 1.054571E-6    A2       = 0.3379035
+RDSW    = 1.380341E3     PRWG     = 0.0301426      PRWB     = 0.0106493
+WR      = 1              WINT     = 2.594349E-7    LINT     = 7.489566E-8
+XL      = 1E-7           XW       = 0              DWG      = -9.471353E-9
+DWB     = 3.537786E-8    VOFF     = 0              NFACTOR  = 1.0754804
+CIT     = 0              CDSC     = 2.4E-4         CDSCD    = 0
+CDSCB   = 0              ETA0     = 2.332015E-3    ETAB     = -1.531255E-4
+DSUB    = 0.076309       PCLM     = 2.6209353      PDIBLC1  = 1
+PDIBLC2 = 2.23243E-3     PDIBLCB  = -0.0436947     DROUT    = 1.0300278
+PSCBE1  = 6.619472E8     PSCBE2   = 2.968801E-4    PVAG     = 9.970995E-3
+DELTA   = 0.01           RSH      = 80.9           MOBMOD   = 1
+PRT     = 0              UTE      = -1.5           KT1      = -0.11
+KT1L    = 0              KT2      = 0.022          UA1      = 4.31E-9
+UB1     = -7.61E-18      UC1      = -5.6E-11       AT       = 3.3E4
+WL      = 0              WLN      = 1              WW       = 0
+WWN     = 1              WWL      = 0              LL       = 0
+LLN     = 1              LW       = 0              LWN      = 1
+LWL     = 0              CAPMOD   = 2              XPART    = 0.5
+CGDO    = 2.34E-10       CGSO     = 2.34E-10       CGBO     = 1E-9
+CJ      = 4.240724E-4    PB       = 0.9148626      MJ       = 0.4416777
+CJSW    = 3.007134E-10   PBSW     = 0.8            MJSW     = 0.2025106
+CJSWG   = 1.64E-10       PBSWG    = 0.8            MJSWG    = 0.2025106
+CF      = 0              PVTH0    = 0.0526696      PRDSW    = 110.1539295
+PK2     = -0.0283027     WKETA    = -0.0191754     LKETA    = 8.469064E-4
```

98 parameters in this BSIM Model !

Typical Model Parameters (cont)

```
.MODEL CMOSP PMOS (
+VERSION = 3.1            TNOM     = 27             LEVEL    = 49
+XJ      = 1.5E-7         NCH      = 1.7E17         TOX      = 1.4E-8
+K1      = 0.5600277      K2       = 9.302429E-3    K3       = 7.2192028
+K3B     = -1.0103515     WO       = 1.010628E-8    NLX      = 5.826683E-8
+DVTOW   = 0              DVT1W    = 0              DVT2W    = 0
+DVTO    = 2.2199372      DVT1     = 0.5378964      DVT2     = -0.1158128
+UO      = 220.5729225    UA       = 3.141811E-9    UB       = 1.085892E-21
+UC      = -5.76898E-11   VSAT     = 1.342779E5     AO       = 0.9333822
+AGS     = 0.157364       BO       = 9.735259E-7    B1       = 5E-6
+KETA    = -2.42686E-3    A1       = 3.447019E-4    A2       = 0.3701317
+RDSW    = 3E3            PRWG     = -0.0418484     PRWB     = -0.0212357
+WR      = 1              WINT     = 3.097872E-7    LINT     = 1.040878E-7
+XL      = 1E-7           XW       = 0              DWG      = -1.983686E-8
+DWB     = 1.629532E-8    VOFF     = -0.0823738     NFACTOR  = 0.969384
+CIT     = 0              CDSC     = 2.4E-4         CDSCD    = 0
+CDSCB   = 0              ETA0     = 0.4985496      ETAB     = -0.0653358
+DSUB    = 1              PCLM     = 2.1142057      PDIBLC1  = 0.0256688
+PDIBLC2 = 3.172604E-3    PDIBLCB  = -0.0511673     DROUT    = 0.1695622
+PSCBE1  = 1.851867E10    PSCBE2   = 1.697939E-9    PVAG     = 0
+DELTA   = 0.01           RSH      = 103.6          MOBMOD   = 1
+PRT     = 0              UTE      = -1.5           KT1      = -0.11
+KT1L    = 0              KT2      = 0.022          UA1      = 4.31E-9
+UB1     = -7.61E-18      UC1      = -5.6E-11       AT       = 3.3E4
+WL      = 0              WLN      = 1              WW       = 0
+WWN     = 1              WWL      = 0              LL       = 0
+LLN     = 1              LW       = 0              LWN      = 1
+LWL     = 0              CAPMOD   = 2              XPART    = 0.5
+CGDO    = 3.09E-10       CGSO     = 3.09E-10       CGBO     = 1E-9
+CJ      = 7.410008E-4    PB       = 0.9665307      MJ       = 0.4978642
+CJSW    = 2.487127E-10   PBSW     = 0.99           MJSW     = 0.3877813
+CJSWG   = 6.4E-11        PBSWG    = 0.99           MJSWG    = 0.3877813
+CF      = 0              PVTH0    = 5.98016E-3     PRDSW    = 14.8598424
+PK2     = 3.73981E-3     WKETA    = 2.870507E-3    LKETA    = -4.823171E-3
```

End of Lecture 7