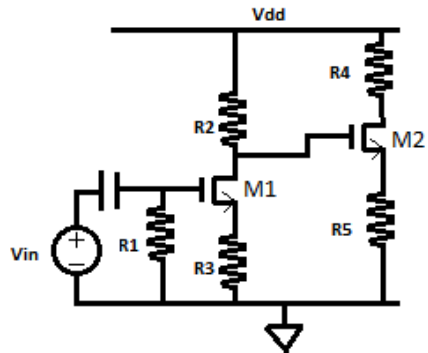




### Problem 3)

This design problem has a few parts that make it a bit difficult. First it needs an input impedance of 100k-200k $\Omega$ . This seems easy enough, but our basic amplifier structures have  $\infty$  except for the Common Gate design. I am not going to use that design because of a problem with getting the proper  $R_{in}$ , so instead I will use two CSwRS amplifiers and an extra resistor, because determining the gain of CSwRS amplifiers is the easiest.



The gain of this is fairly easy to calculate, as the second stage is  $A_{V2} = -\frac{R_4}{R_5}$  and the first stage is

$$A_{V1} = -\frac{(R_2 || R_{in2})}{R_3}, \text{ and } R_{in2} = \infty \rightarrow A_{V1} = -\frac{R_2}{R_3}$$

$$A_V = A_{V1} * A_{V2} = \frac{R_2 R_4}{R_3 R_5}$$

We can set,

$$R_2 = 5k\Omega$$

$$R_3 = 1k\Omega$$

$$R_4 = 1k\Omega$$

$$R_5 = 1k\Omega$$

$$A_V = \frac{5 * 1}{1 * 1} = 5$$

Finally we need to create the proper input impedance. The amplifier structure has an  $R_{in1} = \infty$ , but it is in parallel with  $R_1$

$R_{in} = R_{in1} || R_1 = R_1$  so we set  $R_1$  between 100K and 200K,

$$R_1 = 150k\Omega$$

### Problem 4)

$$M = \frac{W_2 + 2\Delta w}{W_1 + 2\Delta w} * \frac{L_1}{L_2}$$

$$M = \frac{10 - 0.2}{2 - 0.2} * \frac{4}{4}$$

$$M = 5.44$$

Problem 5)

a)

$$V_{01} = I_3 R_2 = \frac{A_3}{A_2} \beta I_1 R_2$$

$$V_{01} = \left( \frac{V_{DD} - 0.6}{R_1} \right) \beta \frac{A_3}{A_2} * R_2$$

$$V_{02} = I_6 R_3 = \frac{\frac{W_6}{L_6}}{\frac{W_5}{L_5}} * \frac{A_4}{A_1} * \beta I_1 R_3$$

$$V_{02} = \left( \frac{V_{DD} - 0.6}{R_1} \right) \beta * \frac{A_4}{A_1} * \frac{W_6 L_5}{W_5 L_6} R_3$$

b)

$$\left( \frac{10 - 0.6}{60k} \right) * 100 * \frac{25}{100} * R_2 = 3V$$

$$R_2 = 766.0\Omega$$

$$\left( \frac{10 - 0.6}{60k} \right) * 100 * \frac{300}{100} * \frac{16 * 1}{10 * 4} * R_3 = 6V$$

$$R_3 = 319.1\Omega$$

Problem 6

A)

$$A_{V3} = \frac{g_{m5}}{-g_{05} + g_{m6} + g_{06}} \approx -\frac{g_{m5}}{g_{m6}}$$

$$A_{V2} = -\frac{g_{m4}}{g_{m3}} \quad ; \quad A_{V1} = -\frac{g_{m1}}{g_{m2}}$$

$$A_V = A_{V1} A_{V2} A_{V3} = -\frac{g_{m1} g_{m4} g_{m5}}{g_{m2} g_{m3} g_{m6}}$$

B)

$$I_{m1} = I_{m4} = I_{m5} = 500\mu A$$

$$g_m = \frac{2I_{CQ}}{V_{GSQ} - V_T}$$

$$V_{GS2} = V_{TP} + \sqrt{\frac{I_{M1}}{\mu C_{OX} \left( \frac{W_2}{2L_2} \right)}} = 2.5V$$

$$V_{GS3} = V_{TP} + \sqrt{\frac{I_{M4}}{\mu C_{OX} \left( \frac{W_3}{2L_3} \right)}} = 2.5V$$

$$V_{GS6} = V_{TP} + \sqrt{\frac{I_{M5}}{\mu C_{OX} \left( \frac{W_6}{2L_6} \right)}} = 4V$$

$$g_{m1} = \frac{2 * 500\mu}{5 - 0.5} = \frac{1000\mu A}{4.5 V}$$

$$g_{m2} = g_{m3} = g_{m4} = g_{m5} = \frac{2 * 500\mu}{2.5 - 0.5} = \frac{500\mu A}{V}$$

$$g_{m6} = \frac{2 * 500\mu}{4 - 0.5} = \frac{1000\mu A}{3.5 V}$$

$$A_V = -\frac{g_{m1} g_{m4} g_{m5}}{g_{m2} g_{m3} g_{m6}} = -\frac{g_{m1}}{g_{m6}}$$

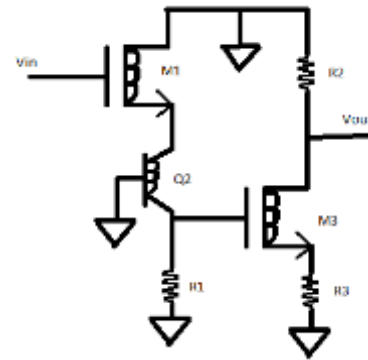
$$A_V = -\frac{3.5}{4.5} = -0.777$$

Problem 8)

$$\frac{V_{out}}{V_A} = -\frac{R_2}{R_3}$$

$$\frac{V_A}{V_{in}} = g_{m2} R_1 \left( \frac{g_{m1}}{g_{m1} g_{m2}} \right)$$

$$A_V = \frac{V_{out}}{V_{in}} = \frac{V_{out}}{V_A} * \frac{V_A}{V_{in}} = -\frac{R_2}{R_3} g_{m2} R_1 * \left( \frac{g_{m1}}{g_{m1} + g_{m2}} \right)$$



Problem 9)

$$A_{V3} = -\frac{(R_3 || R_L)}{R_2}$$

$$R_{in4} = \beta \left( \frac{V_t}{I_{CQ}} + R_2 \right)$$

$$A_{V2} = -g_{m3} R_C = -g_m \left( R_{in2} || \frac{1}{g_{m2}} \right)$$

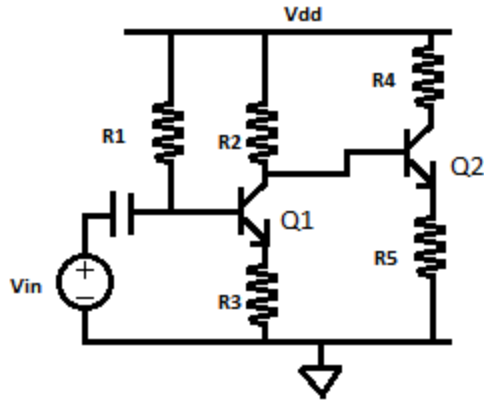
$$A_{V1} = -g_{m1} R_C = -g_{m1} R_1$$

$$A_V = A_{V1} A_{V2} A_{V3}$$

$$A_V = (-g_{m1} R_1) * g_{m3} \left( \frac{1}{g_{m2}} || R_{in4} \right) * \left( -\frac{R_3 || R_L}{R_2} \right) ; R_{in4} = r_{\pi 4} + \beta R_2$$

Problem 10)

For this problem we have BJTs to create a gain of 60 and input impedance greater than 100k. The load is 1k. The easiest amplifier to create is the CEwRE (Common Emitter with Emitter Resistor), but has the problem of creating a negative gain. So we need two stages of negative gain.



We have to work backwards for this circuit because the input impedance of Q2 ( $R_{in2}$ ) will affect the gain of Q1. First I will set  $V_{dd} = 1.8V$

$R_4$  is already determined, as it is the load resistor,  $R_4 = 1k\Omega$

We want a total gain of 60, and the gain of the above is  $-\frac{R_C}{R_E} = -\frac{R_4}{R_5}$ . Therefore if we use  $R_5 = 500\Omega$  we get a gain of  $A_{V2} = -\frac{1000}{500} = -2$  and need a gain of -30 in the first stage.

The first stage has a gain of  $-\frac{(R_2 || R_{in2})}{R_3}$ .  $R_{in2} = \beta \left( \frac{V_t}{I_{CQ2}} + R_E \right) = 100 \left( \frac{0.0259}{I_{CQ2}} + 500\Omega \right)$ .

$I_{CQ} = \beta I_{B2}$ ,  $I_{B2} = \frac{5V - 0.6 - V_{R5}}{R_2} = \frac{V_{R5}}{500}$ . We want this small so that the resistance is large.

I am going to set  $R_2 = 500k\Omega$ . This sets  $V_{R5} = 4.39mV$  and  $I_B = 8.79\mu A$ .  $I_C = 100(i_B) = 879\mu A$

$R_{in2} = 100 * \left( \frac{0.0259}{0.000879} + 500 \right) = 52.9k\Omega$

$R_2 || R_{in2} = 47.9k\Omega$ , so we want  $R_3 = \frac{47.85k}{20} = 239\Omega$

$A_{V1} = -\frac{47850}{1595} \approx -30$

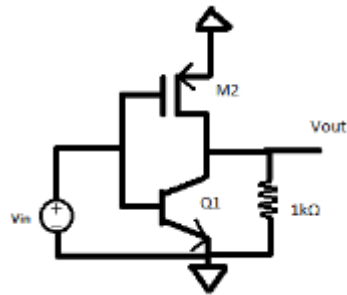
$A_V = A_{V1} * A_{V2} = -30 * -2 \rightarrow A_V = 60 \frac{V}{V}$

The input impedance of the system is  $100 * \left( \frac{0.0259}{I_{CQ1}} + 957 \right)$ .  $I_{CQ1} = \beta I_{B1}$ , we want this really small try

$R_1 = 7M\Omega \rightarrow I_{B1} = 0.628\mu A \rightarrow I_{C1} = 62.8\mu A \rightarrow R_{in1} = 200.7k\Omega$

Problem 11)

A) SS equivalent circuit:



$$A_V = -(g_{m1} + g_{m2}) * \left( 1k \parallel \frac{1}{g_{o1}} \parallel \frac{1}{g_{o2}} \right)$$

$$I_{DQ} = I_{CQ} = J_s A_E \exp\left(\frac{V_{BE}}{V_t}\right) = 10.5mA$$

$$g_{m2} = \frac{2I_{CQ}}{V_{GS} - V_T} = \frac{2 * 0.0105}{5V - 0.5V} = \frac{0.021}{4.5} = 0.0046667$$

$$g_{m1} = \frac{I_{CQ}}{V_t} = \frac{0.0105}{0.0259} = 0.4054$$

$$g_{o1} = \frac{I_{CQ}}{V_{AF}} = \frac{0.0105}{100} = 0.000105$$

$$g_{o2} = \lambda I_{DQ} = 0.01 * 0.0105 = 0.000105$$

$$\frac{1}{g_{o2}} = \frac{1}{g_{o1}} = \frac{1}{0.000105} = 9523.8$$

$$A_V = -(0.4054 + 0.00467) * \left( \frac{1}{\frac{1}{1000} + \frac{1}{9523} + \frac{1}{9523}} \right)$$

$$A_V = -338.9 \frac{V}{V}$$

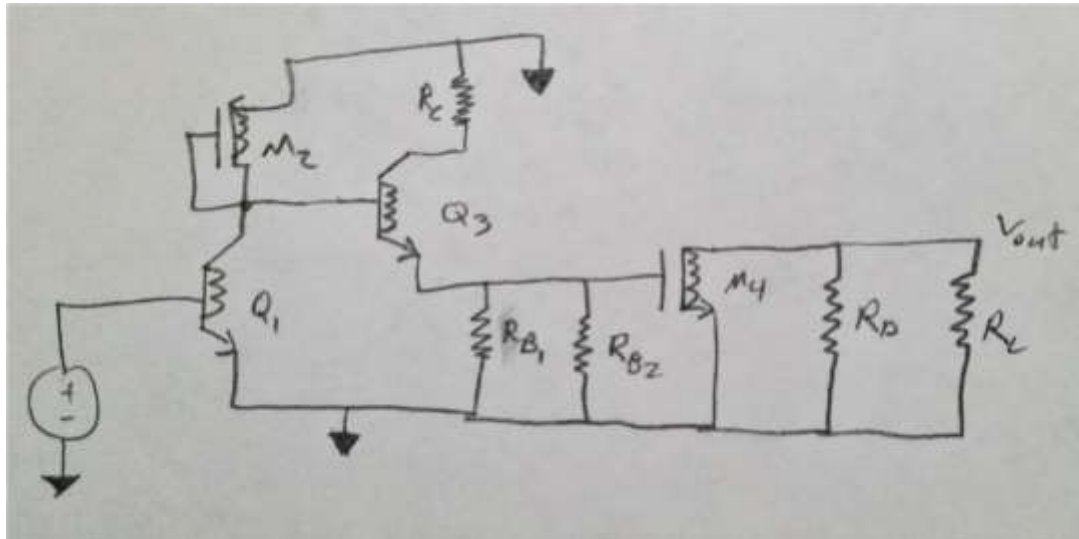
B)

$$V_{out} = A_V V_{in}$$

$$V_{out} = 3.38 \sin(1000t)$$

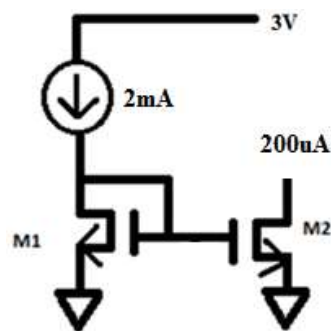
Problem 12

a)



$$b) A_v = -g_{m1} \left( \frac{1}{g_{m2}} || R_{in3} \right) * g_{m3} \left( \frac{R_c}{R_{B1} || R_{B2}} \right) * -g_{m4} (R_D || R_L)$$

13)



a) Set  $L_1 = L_2 = W_1 = 10\mu$

We want  $1/10^{\text{th}}$  the current at the output, so  $W_2 = 1\mu$ .

b)

$$I_1 = \frac{\mu_p C_{ox} W_4}{2L_4} (|V_{GS}| - |V_T|)^2 = 2mA$$

$$\frac{33 * 10^{-6} * 2}{2 * 1} (|V_{GS}| - |V_T|)^2 = 2mA$$

$$|V_{GS}| - |V_T| = \sqrt{\frac{2 * 1 * 2mA}{33 * 10^{-6} * 2}} = 7.78V$$

14)

$$A_{VLeft} = -\frac{1}{2} \left( \frac{g_{m1}}{g_{01}} \right) < A_{VMid} = -\frac{1}{2} \left( \frac{g_{m1}g_{m2}}{g_{01}g_{02}} \right) < A_{VRight} = -\left( \frac{g_{m1}g_{m2}}{g_{01}g_{02}} \right)$$

15) Current mirror and voltage across a resistor,

$$V_{out} = 8V - \frac{\left(\frac{15}{1}\right)}{\left(\frac{5}{1}\right)} 50\mu * 25000 = 4.25V$$

16) Current mirror of a BJT,  $\frac{A_2}{A_1} i_{in} = i_{out}$

$$i_{out} = \frac{50+300+1200}{300+600+1200} 50 * 10^{-6} = 36.9 * 10^{-6}$$

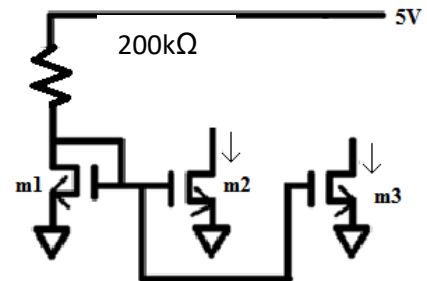
$$V_{out} = 5V - (6000 * 36.9e - 6) = 4.78$$

17) The current through the resistor is

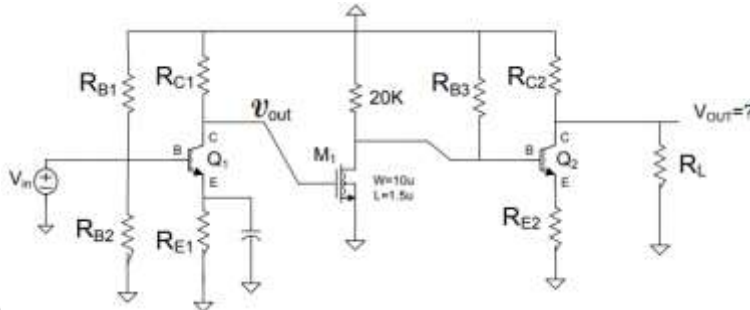
$$I_{in} = \frac{5 - 0.6}{200000} = 22\mu A$$

We want 10/22 and 50/22 gain, so set  $W_1 = L_1 = L_2 = L_3 = 22\mu m$ .

So set  $W_2 = 10, W_3 = 50$ .



18)



a)

b)

$$A_V = -\frac{(R_{C2} || R_L)}{R_{E2}} * -g_m(R_{B3} || 20k\Omega || r_{\pi 2} + \beta R_{E2}) * -g_m R_{C1}$$

c)

$$A_V = -\frac{(R_{C2} || R_L)}{R_{E2}} * -\frac{2I_{DQ1}}{V_{EB}} (R_{B3} || 20k\Omega || r_{\pi 2} + \beta R_{E2}) * -\frac{I_{CQ2} R_{C1}}{V_t}$$



Fall 2018 – Homework 11 – TA: Robert Buckley

19)

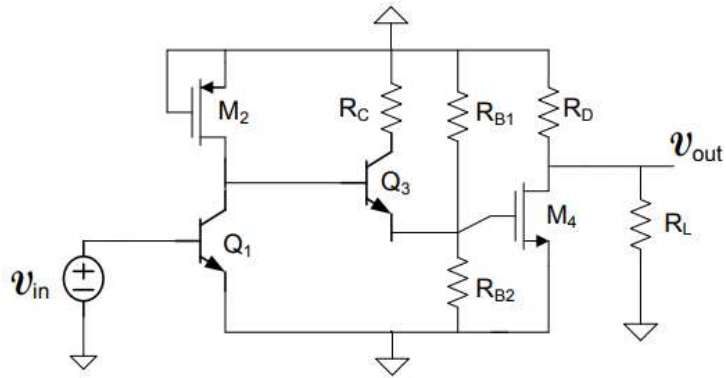
20)

21)

22)

23.

a)

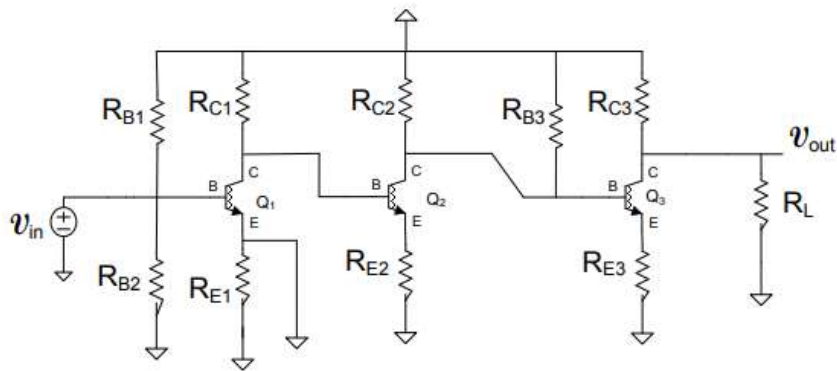


b)

$$A_V = -g_{m1} \left( \frac{1}{g_{m2}} \right) * - \frac{R_C}{(R_{B1} || R_{B2} || \infty)} * -g_{m4}(R_D || R_L)$$

24.

a)

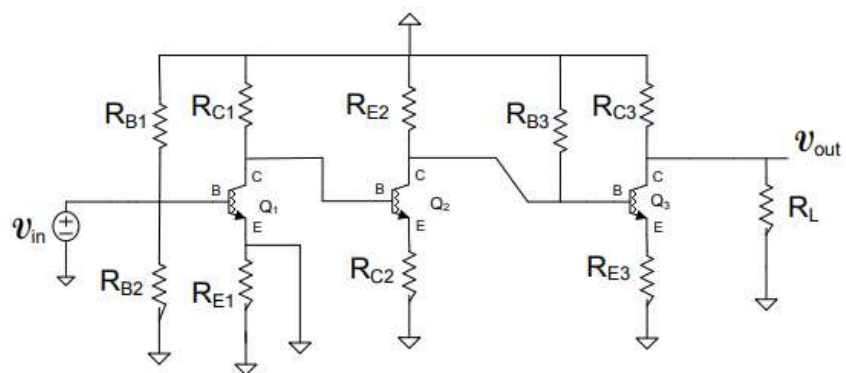


b)

$$A_v = \frac{(R_{C1} || r_{\pi 2} + \beta R_{E2})}{R_{E1}} * - \frac{(R_{C2} || R_{B3} || r_{\pi 3} + \beta R_{E3})}{R_{E2}} * - \frac{(R_{C3} || R_L)}{R_{E3}}$$

25.

a)

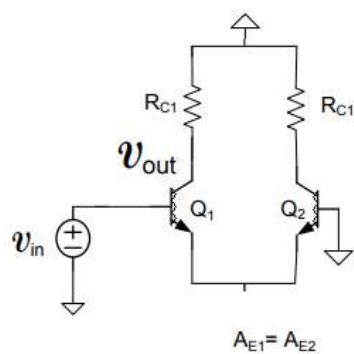


b)

$$A_V = \frac{(R_{C1} || r_{\pi 2} + \beta R_{E2})}{R_{E1}} * - \frac{(R_{E2} || R_{B3} || r_{\pi 3} + \beta R_{E3})}{R_{E2}} * - \frac{(R_{C3} || R_L)}{R_{E3}}$$

26.

a)

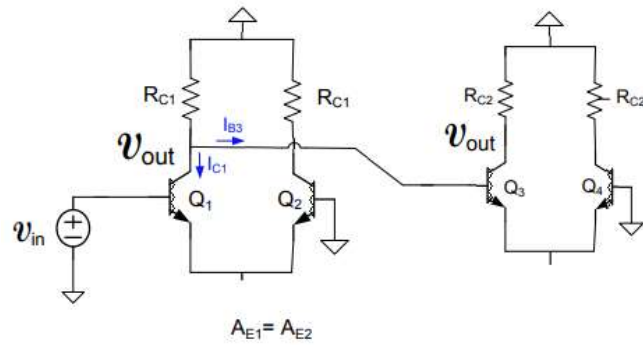


b)

$$A_V = - \frac{R_C}{R_E} = - \frac{R_{C1}}{\left( R_{C1} + \frac{v_t}{I_{CQ}} \right)}$$

27)

a)

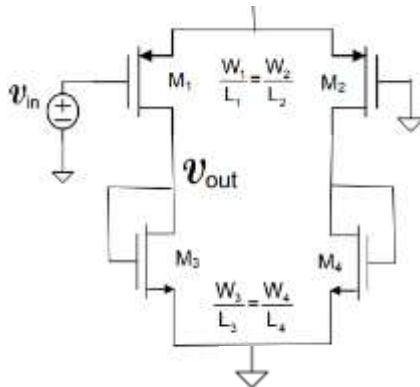


b)

$$A_V = -\frac{R_{C1}}{R_{E1}} * \frac{R_{C2}}{R_{E2}} = \frac{-R_{C1}}{\left(R_{C1} + \frac{v_t}{I_{CQ1}}\right)} * \frac{-R_{C2}}{\left(R_{C2} + \frac{v_t}{I_{CQ2}}\right)}$$

28)

a)

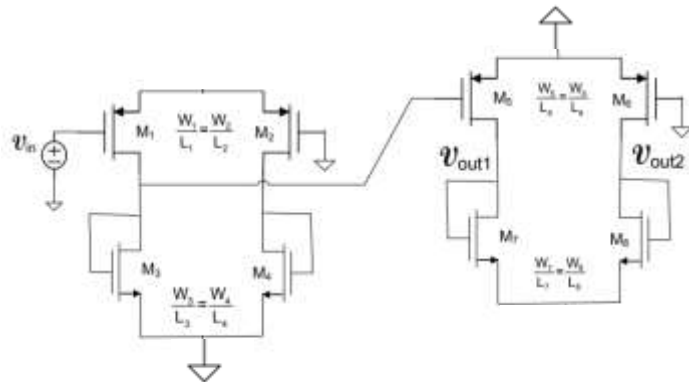


b)

$$A_V = -\frac{R_D}{R_B} = \frac{-\left(\frac{1}{g_{m3}}\right)}{\left(\frac{1}{g_{m2}} + \frac{1}{g_{m4}}\right)}$$

29)

a)



b)

$$A_V = -\frac{R_{D1}}{R_{E1}} * -\frac{R_{D2}}{R_{E2}} = \frac{-\left(\frac{1}{g_{m3}}\right)}{\left(\frac{1}{g_{m2}} + \frac{1}{g_{m4}}\right)} \frac{-\left(\frac{1}{g_{m7}}\right)}{\left(\frac{1}{g_{m6}} + \frac{1}{g_{m8}}\right)}$$