

Experiment 1 - Clock and Periodic Signal Generation

Abstract— This document is a student report for the 1st experiment of the Digital Logic Laboratory course (ECE 045) at University of Tehran, Department of Electrical and Computer Engineering.

Keywords— Clock Generation, Ring Oscillator, LM555, Schmitt Trigger Oscillator, FPGA Design, Frequency Divider, Baud Rate Generator, LTspice, Quartus, ModelSim

I. CLOCK GENERATION USING ICs AND ANALOG COMPONENTS

A. Ring Oscillator

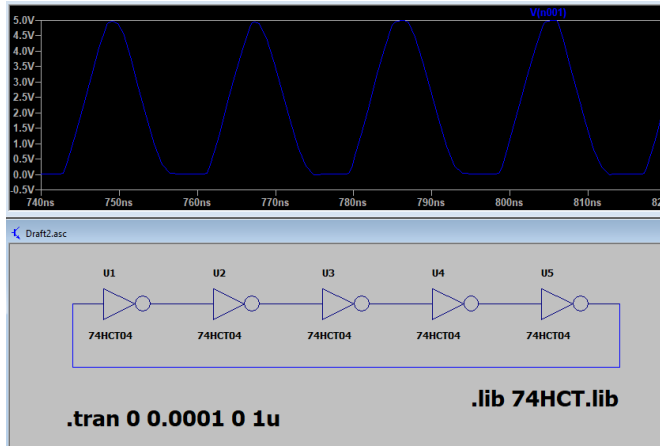


Fig. 1 Simulation waveform of a ring oscillator using 5 inverters (74HCT04)

1) *Propagation delay*: It is defined as the time from the 50% point of the input to the 50% point of the output.

Here, our input and output are connected. Therefore, the voltage waveform for all the points in the circuit are identical.

So, we will use the positive and the negative slopes of the waveform to get the propagation delay of the chain.

50% of input and output will be at +2.5v and -2.5v, referring to Fig. 1, the distance is about 8ns which is our delay.

2) *The delay of a single inverter*: $T = 2N * \text{Delay}_{\text{inv}}$

The period of the signal is about 18ns as shown in Fig. 1

$18\text{ns} = 2 * 5 * \text{Delay}_{\text{inv}} \rightarrow \text{Delay}_{\text{inv}} = 1.8\text{ns}$

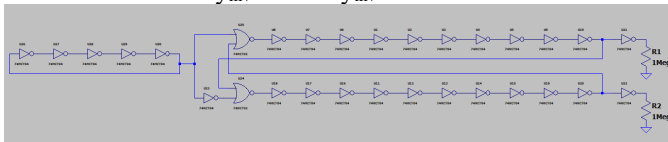


Fig. 2 Two-phase clock generator

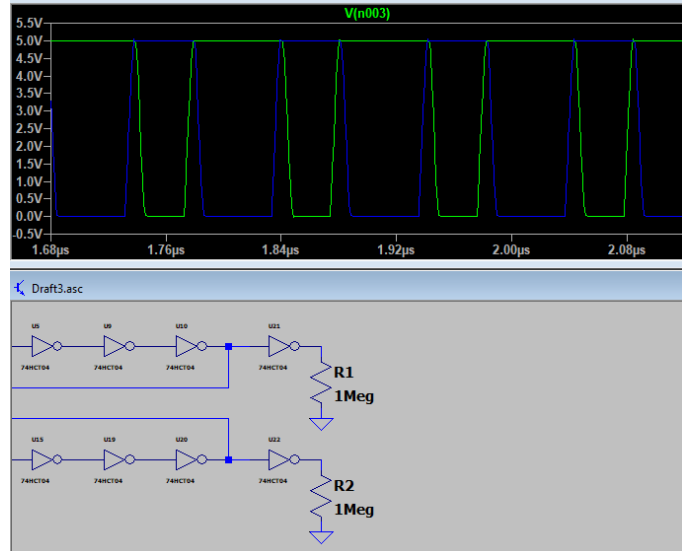


Fig. 3 Two-phase clock generator output waveform. n003 (green) is the top output (Phase1) and n005 (blue) is the bottom output (Phase2)

As shown in Fig. 3, the outputs are two non-overlapping signals which are generated from a single clock signal (the ring oscillator)

B. LM555 Timer

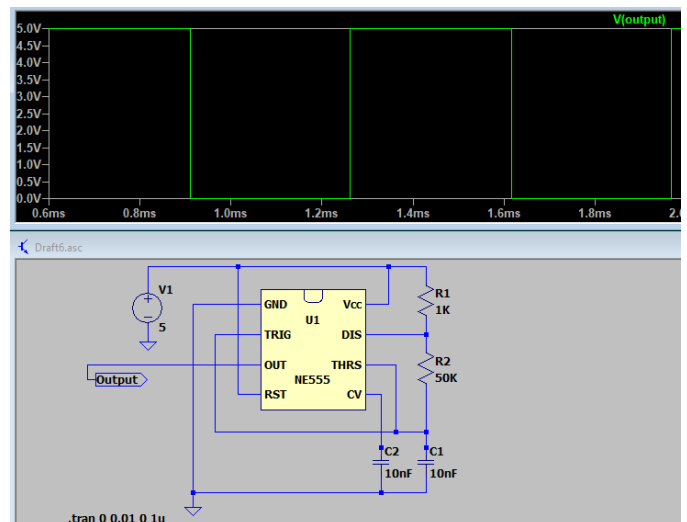


Fig. 4 LM555 in astable mode

1) From Fig. 1, we can see that the total time period of the square wave created by the LM555 IC is about 0.7ms. Therefore, the frequency ($f = 1/T$) is 1428.57Hz.

In each period, the signal changes at about the mid-way of the period, so the duty cycle is around 50%

We will now confirm the observations with the theoretical values:

$$T = T_{\text{charge}} + T_{\text{discharge}} = 0.693 * (R_1 + 2R_2) * C$$

$$= 0.693 * (1 + 2*50)\text{e}3 * 10\text{e-}9 = 0.69993\text{ms}$$

$$\text{Duty Cycle} = (R_1 + R_2) / (R_1 + 2R_2)$$

$$= (1 + 50) / (1 + 100) = 50.5\%$$

We can see that the theoretical values and our observation have the same results.

Now we will do the equations for different R2 values:

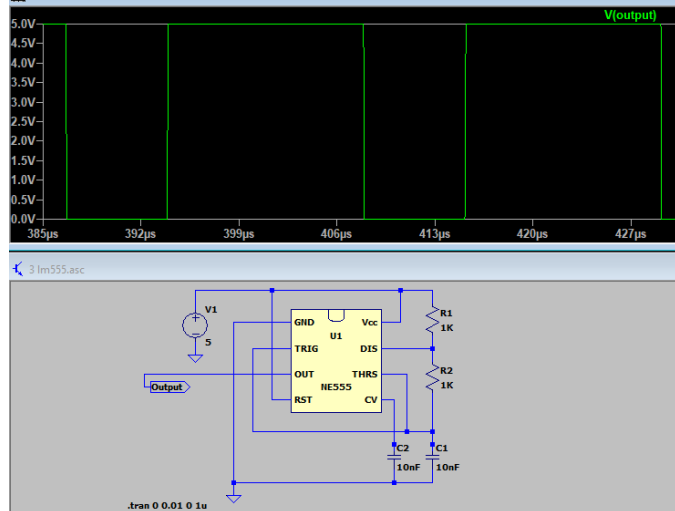


Fig. 5 R2 = 1KΩ

Observations:

$$T = 21\mu\text{s}, F = 47.619\text{KHz}$$

$$\text{Duty Cycle} = 14/21 = 66.6\%$$

Theoretical:

$$T = 0.693 * (1 + 2)\text{e}3 * 10\text{e-}9 = 20.79\mu\text{s}$$

$$\text{Duty Cycle} = 2/3 = 66.6\%$$

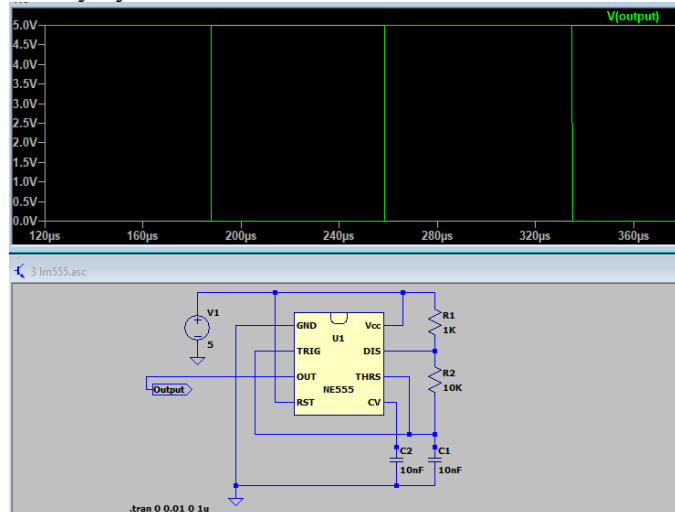


Fig. 6 R2 = 10KΩ

Observations:

$$T = 147\mu\text{s}, F = 6.8\text{KHz}$$

$$\text{Duty Cycle} = 77/147 = 52.3\%$$

Theoretical:

$$T = 0.693 * (1 + 20)\text{e}3 * 10\text{e-}9 = 145.5\mu\text{s}$$

$$\text{Duty Cycle} = 11/21 = 52.3\%$$

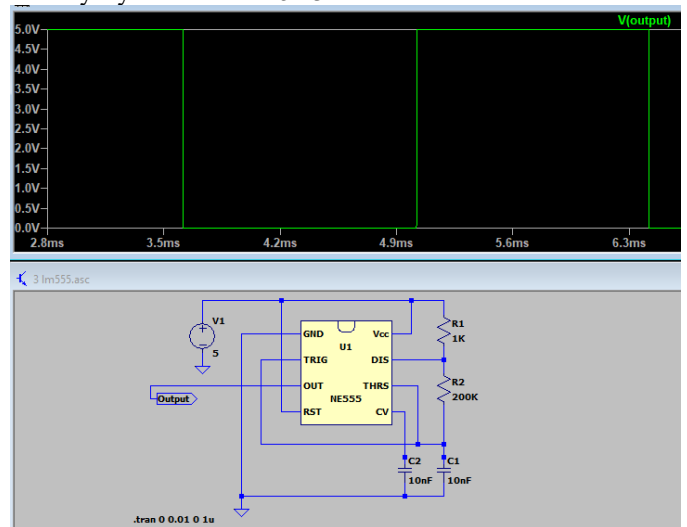


Fig. 7 R2 = 200KΩ

Observations:

$$T = 2.8\text{ms}, F = 375.1\text{Hz}$$

$$\text{Duty Cycle} = 1.4/2.8 = 50\%$$

Theoretical:

$$T = 0.693 * (1 + 400)\text{e}3 * 10\text{e-}9 = 2.78\text{ms}$$

$$\text{Duty Cycle} = 201/401 = 50.1\%$$

C. Schmitt Trigger Oscillator

$$f = \alpha / RC$$

We will now try the circuit with different R values:

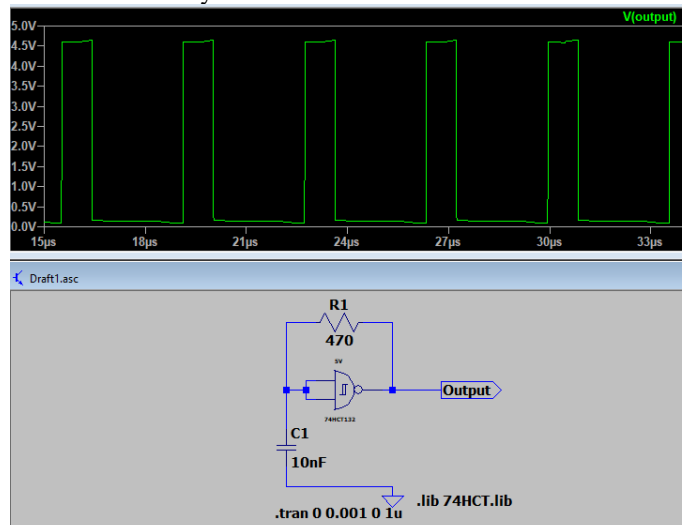


Fig. 8 R = 470Ω

$$T = 3.6\mu\text{s}, F = 277.8\text{KHz}$$

$$F = \alpha / RC, 277.8\text{e}3 = \alpha / 470 * 10\text{e-}9 \rightarrow \alpha = 1.306$$

II. FPGA DESIGN

A. Ring Oscillator

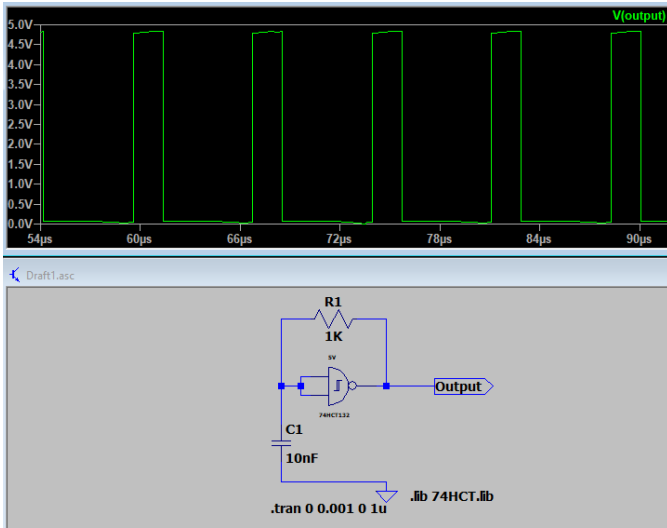


Fig. 9 $R = 1K\Omega$

$$T = 7.2\mu s, F = 138.8KHz$$

$$F = \alpha / RC, 138.8e3 = \alpha / e3 * 10e-9 \rightarrow \alpha = 1.388$$

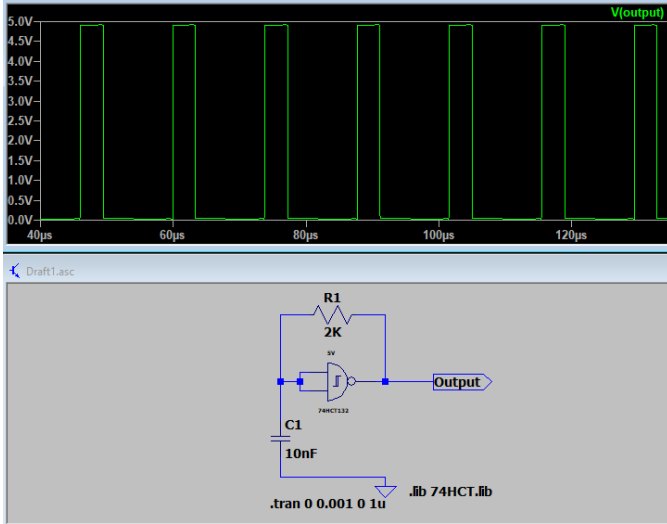


Fig. 10 $R = 2K\Omega$

$$T = 14\mu s, F = 71.4KHz$$

$$F = \alpha / RC, 71.4e3 = \alpha / 2e3 * 10e-9 \rightarrow \alpha = 1.428$$

So α has an average of 1.374

```

1  `timescale 1ns/1ps
2
3  module ring_osc #(
4      parameter N = 5,
5      parameter Delay = 2
6  )
7      output out,
8      input rst
9  );
10     wire [N:0] w;
11     assign w[0] = rst ? 1'b0 : w[N];
12
13     generate
14         genvar i;
15         for (i = 0; i < N; i = i+1) begin: Inverters
16             not #(Delay) inv(w[i+1], w[i]);
17         end
18     endgenerate
19
20     assign out = w[0];
21 endmodule

```

Fig. 11 Ring oscillator Verilog description

```

1  ring_osc_tb.v > ...
2  `timescale 1ns/1ps
3  `include "ring_osc.v"
4
5  module ring_osc_tb();
6      wire out;
7      reg rst = 1'b1;
8
9      ring_osc #(N(5), .Delay(1.8)) osc(out, rst);
10
11     initial begin
12         #20 rst = 1'b0;
13         #200 $stop;
14     end
15 endmodule

```

Fig. 12 Ring oscillator Verilog test-bench

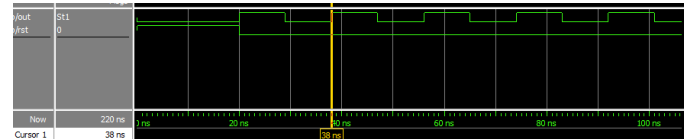


Fig. 13 Ring oscillator simulation output

As we can see, the period of the oscillator is, as expected, 18ns which will equate to 55.56MHz.

B. Synchronous Counter as a Frequency Divider

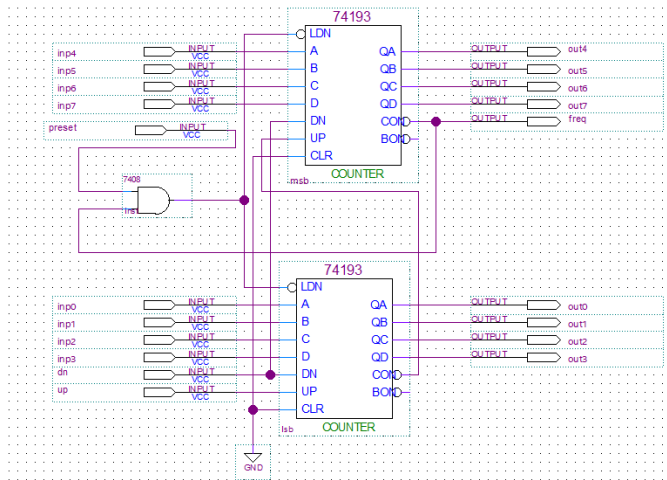


Fig. 14 Frequency divider

```

1  freq_divider.tbv > ...
2  *timescale 1ns/1ps
3
4  'include "freq_divider.vo"
5
6  module freq_divider.tbv();
7      reg preset, rst;
8      wire up, freq;
9      wire [7:0] i, q;
10     assign i = 8'd153;
11
12     ring_osc #(N(5), .Delay(1.8)) osc(up, rst);
13
14     freq_divider div(.inp7(i[7]), .inp6(i[6]), .inp5(i[5]), .inp4(i[4]), .inp3(i[3]), .inp2(i[2]), .inp1(i[1]),
15     .out7(q[7]), .out6(q[6]), .out5(q[5]), .out4(q[4]), .out3(q[3]), .out2(q[2]), .out1(q[1]),
16     .up(up), .dn(1'b1), .preset(preset), .freq(freq));
17
18     initial begin
19         #0 {rst, preset} = 2'b10;
20         #200ns {rst, preset} = 2'b01;
21         #400ns $stop;
22     end
23 endmodule

```

Fig. 15 Test-bench

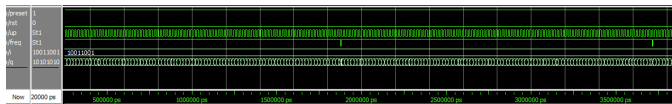


Fig. 16 Simulation results

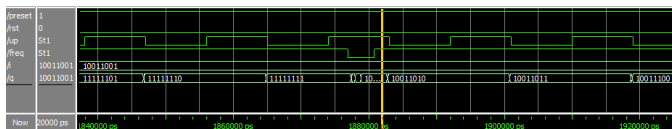


Fig. 17 When carryout becomes 0 (active-low)

The clock frequency is, as calculated before, 55.56MHz.

The counter carryout is set after 103 clocks, basically dividing the signal by 103 meaning that the frequency should be around 539.4KHz.

Observing the simulation results, the frequency is about 542.2KHz which almost matches with the theoretical value.

C. T Flip-Flop

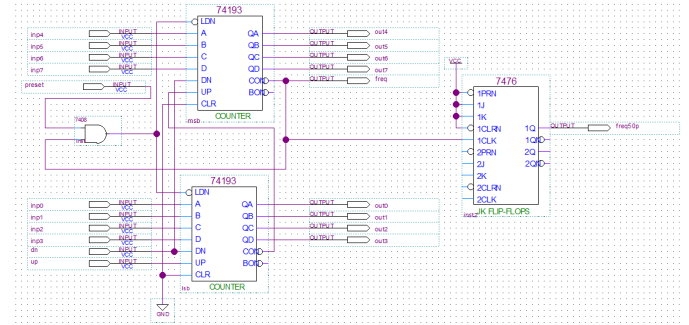


Fig. 18 Frequency divider with 50% duty cycle

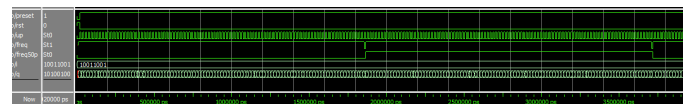


Fig. 19 Simulation results

The JK flip-flop is turned into a T flip-flop by connecting its J and K inputs.

To reach 50% duty cycle (as in Fig. 19), the TFF toggles whenever the MSB carryout is changed.

III. BAUD RATE GENERATOR FOR UART SERIAL COMMUNICATION

A. Automatic Baud Rate Calculator

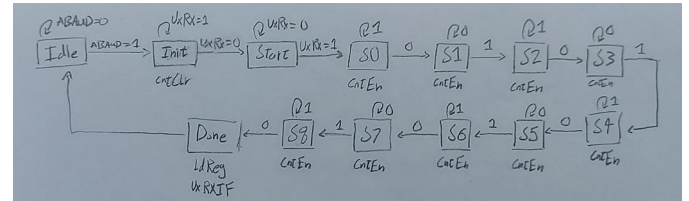


Fig. 20 Auto baud rate controller state diagram

```

1  `define Idle    4'b0000
2  `define Init    4'b0001
3  `define Start   4'b0010
4  `define S0      4'b0011
5  `define S1      4'b0100
6  `define S2      4'b0101
7  `define S3      4'b0110
8  `define S4      4'b0111
9  `define S5      4'b1000
10 `define S6      4'b1001
11 `define S7      4'b1010
12 `define S8      4'b1011
13 `define Done    4'b1100
14
15 module ABRCKT_controller(
16     input ABAUD, UxRX, clk,
17     output reg cntClr, cntEn, UxRXIF, ldReg
18 );
19     reg [3:0] ps, ns;
20
21     always @(posedge clk) begin
22         ps <= ns;
23     end
24
25     always @(ps, ABAUD, UxRX) begin
26         case (ps)
27             `Idle: ns = ABAUD ? `Init : `Idle;
28             `Init: ns = UxRX ? `Init : `Start;
29             `Start: ns = UxRX ? `S0 : `Start;
30             `S0: ns = UxRX ? `S0 : `S1;
31             `S1: ns = UxRX ? `S2 : `S1;
32             `S2: ns = UxRX ? `S2 : `S3;
33             `S3: ns = UxRX ? `S4 : `S3;
34             `S4: ns = UxRX ? `S4 : `S5;
35             `S5: ns = UxRX ? `S6 : `S5;
36             `S6: ns = UxRX ? `S6 : `S7;
37             `S7: ns = UxRX ? `S8 : `S7;
38             `S8: ns = UxRX ? `Done : `S8;
39             `Done: ns = `Idle;
40             default: ns = `Idle;
41         endcase
42     end
43
44     always @(ps) begin
45         {cntClr, cntEn, ldReg, UxRXIF} = 4'd0;
46         case (ps)
47             `Idle:;
48             `Init: cntClr = 1'b1;
49             `Start:;
50             `S0: cntEn = 1'b1;
51             `S1: cntEn = 1'b1;
52             `S2: cntEn = 1'b1;
53             `S3: cntEn = 1'b1;
54             `S4: cntEn = 1'b1;
55             `S5: cntEn = 1'b1;
56             `S6: cntEn = 1'b1;
57             `S7: cntEn = 1'b1;
58             `S8: cntEn = 1'b1;
59             `Done: {ldReg, UxRXIF} = 2'b11;
60             default:;
61         endcase
62     end
63 endmodule

```

Fig. 21 Controller Verilog description

```

ABRCKT_datapath.v > ...
1  module ABRCKT_datapath(
2      input cntClr, cntEn, ldReg, clk,
3      output [7:0] regOut
4  );
5      wire [7:0] bReg, cnt;
6      wire co;
7
8      register #(N(8)) brgReg(.load(ldReg), .ldData(cnt), .out(bReg), .clr(1'b0), .clk(clk));
9      counter #(N(8)) brgCnt(.en(cntEn), .clr(cntClr), .clk(clk), .co(co), .cnt(cnt));
10
11      assign regOut = bReg;
12  endmodule

```

Fig. 22 Datapath Verilog description

```

ABRCKT.v > ...
1  module ABRCKT(
2      input ABAUD, UxRX, clk,
3      output UxRXIF,
4      output [7:0] baudVal
5  );
6      wire cntClr, cntEn, ldReg;
7      wire [7:0] bReg;
8
9      ABRCKT_datapath dp(cntClr, cntEn, ldReg, clk, bReg);
10     ABRCKT_controller cu(ABAUD, UxRX, clk, cntClr, cntEn, UxRXIF, ldReg);
11
12     assign baudVal = bReg;
13 endmodule

```

Fig. 23 Automatic baud rate top-level module

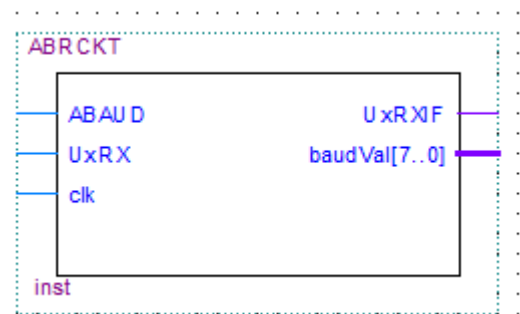


Fig. 24 Module symbol

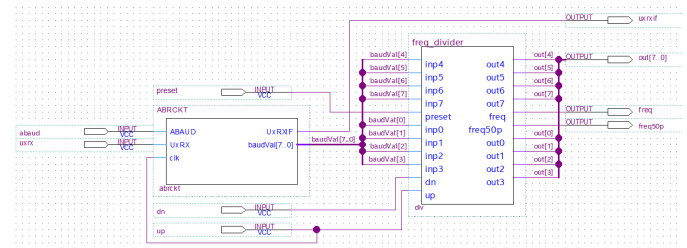


Fig. 25 BRGCKT