

Digital Signal Processing Course Project Adaptive Filtering in Noise- Cancelling Headphones

ECCE 402

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Abstract

This project explores the design, implementation of an adaptive noise cancellation system using the FxLMS algorithms. Adaptive filtering principles are leveraged to develop a real-time active noise control system, with a focus on mitigating secondary path effects through a pre-modelled secondary path filter. The system was implemented on a Raspberry Pi Model 3, utilizing MEMS microphones and custom 3D-printed casing to create a portable solution. Key parameters such as adaptation step size were optimized through simulations across various scenarios to balance convergence speed and stability. While we could successfully obtain a model for the secondary path effects, we could not successfully implement the headphones noise cancelling system due to some errors encountered in the code. The project provides a practical understanding of adaptive signal processing and highlights areas for future improvement, including advanced algorithm tuning for optimal performance in dynamic environments.

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Introduction

Adaptive filters are dynamic systems whose parameters adjust to adapt to changes in the environment. In applications such as echo cancellation in phone calls, channel equalization, and producing a null in a specific direction for antenna characterization, the changes in the environment cannot be specified a priori, necessitating the use of adaptive filters [1]. Among the various adaptive filter algorithms, the Least-Mean-Square (LMS) algorithm is commonly employed due to its simplicity and practicality [1,2]. However, in real -world systems, processing delays and application-specific factors, referred to as secondary path effects, can degrade the performance of noise-cancelling systems using LMS [2-5].

To address these challenges, the FxLMS algorithm, an extension of LMS, that was developed to mitigate secondary path effects [2-5]. While both IIR and FIR filters can be used for adaptive filtering, FIR filters are generally preferred due to their guaranteed stability under any changing conditions [1].

This report details the design and implementation of an adaptive noise-canceling headphone system using the FxLMS algorithm. The report begins with the theoretical background of adaptive filters, focusing on LMS and FxLMS algorithms, followed by the system design, including hardware and software components. Finally, the methodology, results, and conclusions are presented.

Adaptive Filtering and Noise Cancellation Theory

This section discusses the details of the working of the LMS and the FxLMS adaptive filtering algorithms, their block diagram in adaptive noise cancellation systems.

The LMS algorithm

Adaptive filtering algorithms are based on some optimization criterion, each yielding a corresponding equation used to calculate the vector $\mathbf{w}[n+1]$ or the updated filter weights. The algorithm iteratively updates the filter coefficients until that criterion is optimized; at that stage we say convergence is reached, and the difference between $\mathbf{w}[n+1]$ and $\mathbf{w}[n]$ becomes insignificant. The LMS algorithm is based on an optimization of the mean squared error (MES), it works to minimize the MSE to reach the least squared error point [1,3-5]. The MSE is a multivariable function of the filter coefficients [4]. Figure 1 shows the MSE for an adaptive filter with two coefficients. Here, \mathbf{w}_0 and \mathbf{w}_1 are the filter coefficients. At the bottom of the bowl-shaped function, the MSE is at a minimum, and the corresponding values of the filter coefficients, \mathbf{w}_0 and \mathbf{w}_1 , are optimal.

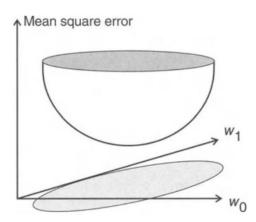


Figure 1. The MSE for an adaptive filter of two coefficients, [4]

The LMS algorithm identifies these optimal values (i.e., reach the minimum MSE point) using the gradient descent method [4,5]. Gradient descent is a method for minimizing a function (MSE) by iteratively adjusting its parameters in the direction of steepest descent [6]. Since computing the gradient of the MSE requires extensive data records and can be computationally intensive, instead, the instantaneous value of the squared error, $e^2[n]$, serves as an approximation of the mean squared error, $E(e^2[n])$ [4]. In practical systems, this would enable computationally feasible real-time processing, where filter coefficients can be updated at every iteration.

Figure 2 shows the block diagram of adaptive filters in noise cancellation using the LMS algorithm. P(z) represents an unknown system (plant), also called the primary path model. Both the plant system and the adaptive FIR model W(z) are excited by the same input x[n]. The FIR model's output, y[n], known as the anti-noise signal, is expressed as:

$$y[n] = \sum_{k=0}^{M-1} w[k] x[n-k]$$

Here, w[k] are the adaptive filter coefficients, and M is the number of coefficients. The error, e[n], is the difference between the plant output, d[n], and the model's output, y[n]. Once convergence is reached, the MSE is minimized, and the FIR filter model equates to the primary path model.

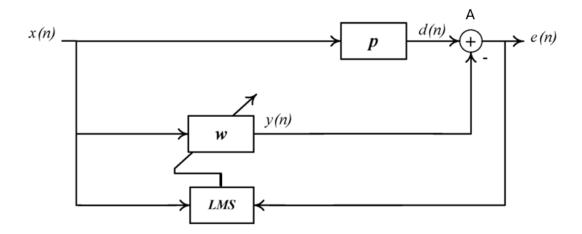


Figure 2. Adaptive filters in noise- cancellation using the LMS algorithm, [7]

In real-world systems, as stated previously, noise- cancelling systems using the LMS algorithm would yield disappointing results due to it overlooking of the secondary path effects. The next section discusses the FxLMS algorithm, which builds on the LMS algorithm to address these limitations. The FxLMS outperforms the LMS by integrating a secondary path effects model, S(z).

Secondary Path Effects and the FxLMS Algorithm

Secondary path effects describe the modifications on the anti-noise, y[n], as it travels through the various components of the system. In noise cancellation systems, the error signal, e[n] is the signal resulting from the acoustic interfaces between the noise modified by the primary path model (headphone), d[n], and the anti-noise signal streamed via a speaker, y[n].

The effectiveness of noise cancellation depends on accurately modelling secondary path effects. For instance, if a delay was introduced by the secondary path effects, this would disrupt the exact 180-degree phase shift between the noise and anti-noise signals required for complete cancellation of the noise, resulting on a comprised performance of the system.

Specifically in this project, the secondary path effects include the impact on the signal, y[n], as it travels through the acoustic path from the speaker to the noise source (e.g. phase shift), and also through the electro-acoustic path, which includes the effects of delays, quantization noise, and distortions introduced by the DAC (Digital-to-Analog Converter) and ADC (Analog-to-Digital Converter), phase shifts, delays introduced by smoothing filters, and noise generated by amplifiers and microphones [2,3].

The FxLMS algorithm is essential for real-time processing applications; it handles those effects by incorporating a secondary path effect model, S(z), cascaded with the FIR adaptive filter model to counteract those effects. The FxLMS algorithm block diagram is shown in Figure 3. The next section derives the w[n+1] or updated filter coefficients equation.

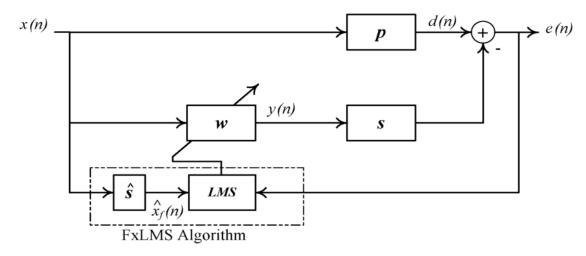


Figure 3. Adaptive filters in noise- cancellation using the FxLMS algorithm, [7]

Derivation of the Updated Adaptive Filter Coefficients Equation

Based on the gradient descent algorithm, the FIR model's updated coefficients, w[n+1], are equal to their previous weights modified by a given percentage of the estimated gradient, as given below. The coefficients iteratively become updated in the direction of the negative of the gradient [1,3-6]

$$w[n+1] = w[n] + \mu \frac{\widehat{V}}{2}$$

Here, μ is the adaptation step size, a crucial parameter that determines by how much the filter coefficients are updated each iteration (the percentage of the estimated gradient that is being added to the current filter coefficients) [1,3-6]. The gradient and the estimated gradient are defined as follows, where $\frac{\partial}{\partial w}$ is the partial derivative with respect to the filter weights.

Gradient:
$$\nabla = \frac{\partial}{\partial w} [E(e^2[n])]$$

Estimated Gradient:
$$\hat{\nabla} = \frac{\partial}{\partial w} [e^2[n]] = 2 \frac{\partial}{\partial w} [e[n]]$$

For the FxLMS algorithm, as derived in [6], the factor $\frac{\partial}{\partial w}[e[n]]$ is given by $-x_s[n]$, which is the filtered x (Fx) or the reference input passed through S(z). This explains the addition of a copy of the secondary path block at the input signal (Figure 3). The FxLMS calculates the updated filter coefficients as a function of $x_s[n]$, and the feedback error signal, e[n]. The simplified updated coefficients of the FIR model equation then become:

$$w[n+1] = w[n] - \mu e[n] x_s[n]$$

The product of the error and the filtered input signal can be interpreted as the cross-correlation between the two signals [5,6]. Cross-correlation is a measure of the similarity between two signals as a function of their time shift. Therefore, the filter coefficients are updated such that the component of $x_s[n]$ that is present on (or contributing to) the error signal is removed, reducing the cross-correlation between the two signals over time.

Having gained an understanding of the theoretical foundations on adaptive filtering and the noise cancellation theory, next we apply those concepts in the design and implementation of a practical noise - cancelling headphones system.

System Design and Overview

The system block diagram is shown in figure 4. The reference microphone captures the noise signal, s[n], this signal is then filtered through an anti-aliasing filter and digitized by an ADC. The digital signal is processed by the processing unit to generate the anti-noise, y[n]. The anti-noise then goes through a smoothing filter, and a DAC before being streamed through the headphones. The error microphone captures the resulting sound signal from the destructive interference of s'[n] and the streamed anti-noise signal. The feedback error signal together with the reference input signal filtered through the S(z) model are used by the LMS to update the FIR model coefficients, as was shown by the equation given above. This process repeats until convergence is reached, and the squared error is at a minimum.

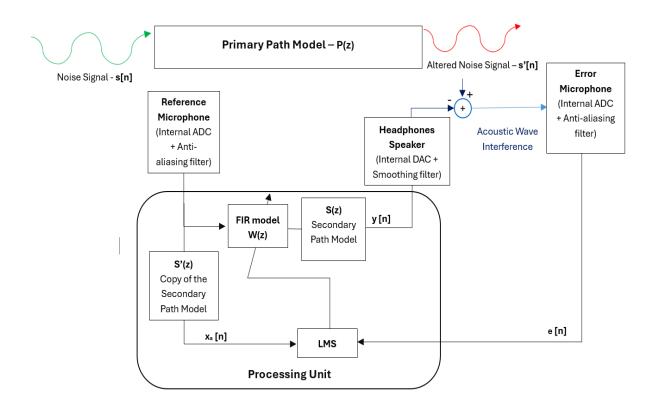


Figure 4. Block Diagram of Noise- Cancelling Headphones System

Figure 5 shows the system design, illustrating the position of every component. Figure 6 shows the 3D model of the casing. The error microphone is placed inside the earcup near the headphone speaker, while the reference microphone is mounted on the processing unit's casing such that its closest to the noise source. Since the reference microphone is in close proximity to the processing unit, it was decided to apply an aluminium tape to the casing to mitigate electromagnetic interference from the processing unit and to enhance system performance.

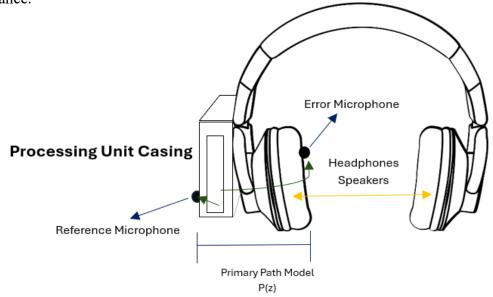


Figure 5. The Design of The Noise- Cancelling Headphones System



Figure 6. 3D Model Design of the Processing Unit Casin.

The next section details the hardware and software components used for the system implementation and the discuses the selection considerations.

Software and Hardware Specifications

The system uses a Raspberry Pi Model 3 as the processing unit, for it can handle complex and computationally- intensive programs required for this implementation. The microphones used are Micro-Electro-Mechanical-System (MEMS), featuring an internal anti-aliasing filter and an ADC, and outputting a 24- bit sample in the pulse width modulation (PCM) [2].

Python programming language was chosen for this implementation. Although C++ is more widely used because of its memory efficiency and speed [8], Python was preferred because to its ease of use, and extensive libraries. The main libraries used in this implementation include numpy, sounddevice, and adafilt. The real- time processing code used is given in <u>Appendix A</u>. Moreover, summarized below are the hardware components of the system:

- ➤ Headphones;
- Raspberry Pi Model 3;
- > 3D printed casing of the processing unit;
- ➤ (2) MEMS microphones (SPH0645LM4H);
- > Jumper wires (male to male and female to male);
- ➤ 4-Pole Male to Male Audio Cable
- Adhesive suction holder (to mount the casing to the headphones);
- > Aluminium tape;
- Plastic nails (to anchor the Raspberry Pi on the casing); and
- Double sided tape.

Methodology

This project implemented a systematic approach to design, optimize, and test an adaptive noise-cancellation system using the FxLMS algorithm. Key steps included identifying critical parameters, running simulations to determine optimal configurations, secondary path modelling, and testing the system in real-time conditions.

Parameter Optimization

The parameters considered for optimization are discussed below. The simulation was used to determine the optimal value of the adaptation step size (μ) . The number of the adaptive filter coefficients and the buffer size parameters were selected based on the computational load capabilities of the processing unit used.

Adaptation Step Size: The step size was a primary focus of optimization, as it directly affects the system's balance between convergence speed and stability.

Number of Adaptive Filter Coefficients (Taps): The number of filter coefficients was another critical parameter. An 8192 - adaptive filter was chosen for the real – time processing implementation to accurately model the dynamic behavior of both primary paths, while ensuring computational efficiency.

Buffer Size: The buffer size is the number of data samples processed simultaneously during each iteration. The buffer size was set to 2048. This ensured maintaining a balance between processing speed and accuracy.

The simulation implementation discussed in the next section tested the performance of the system and enabled us to determine the optimal adaptation step size for the real-time processing implantation.

Simulation Implementation

The simulation implementation was conducted using predefined impulse response values for the primary (h_pri) and secondary (h_sec) paths in each scenario (Figure 3). These configurations reflect the complexity and characteristics of the system's environment, each presenting unique challenges. The system scenario specifications used in the simulation are summarized below. Detailed simulation code and the setup for testing various scenarios are provided in Appendix B.

Scenario 1 (Simple Path Configurations):

Primary Path (h pri):

A single dominant impulse response was placed at index 50, with a value of 0.8.

Secondary Path (h sec):

A smaller impulse response was placed at index 10, with a value of 0.5.

Scenario 2 (Complex Configuration):

Primary Path (h pri):

Two impulse responses were placed at indices 30 and 40, with values of 0.5 and - 0.3, respectively.

Secondary Path (h sec):

A single impulse response was placed at index 20, with a value of -0.4.

Scenario 3 (Moderately *Complex Configuration*):

Primary Path (h pri):

A single impulse response was placed at index 25, with a value of 0.7.

Secondary Path (h sec):

A single impulse response was placed at index 15, with a value of 0.3.

The output of the simulation showing the moving rms value of the error signal for each scenario as a function of the iterations of the system are given in Figures 7-9, for a step size of 0.01, 0.05, and 0.1 respectively. The simulations provided a nuanced understanding of how step size influences error curves and convergence rates. The results confirmed that a step size of around **0.05** delivered the best balance between speed and stability. The analysis of the results obtained is given below:

Scenario 1: Showed faster convergence but experienced higher oscillations, making it less stable for real-world applications.

Scenario 2: Offered a balanced performance, achieving both quick convergence and minimal oscillations, making it the most optimal choice.

Scenario 3: Demonstrated moderate performance, with slower convergence and increased residual error compared to Scenario 2

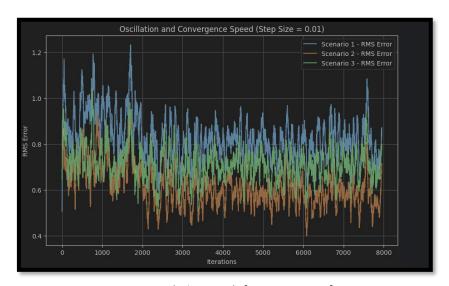


Figure 7 Simulation Result for a Step Size of 0.01

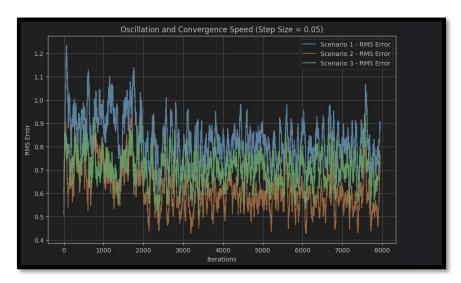


Figure 8 Simulation Result for a Step Size of 0.05

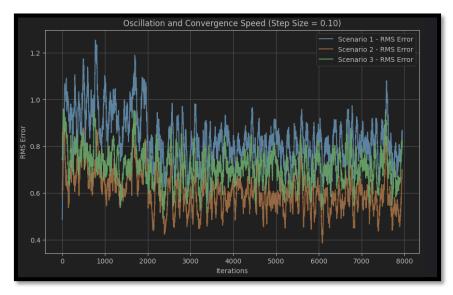


Figure 9 Simulation Result for a Step Size of 0.1

Real-Time Processing Implementation

This section discusses the circuit connections of the microphones and headphones to the Raspberry pi and presents the final prototype design. The details of setting up the Raspberry Pi and running the code is given in Appendix D.

In order to connect the I2S microphones to the Raspberry Pi the following connection (Figure 10) was used as reference. The headphone was connected to the Rassberry Pi's audio jack via a 4-pole male to male wire.

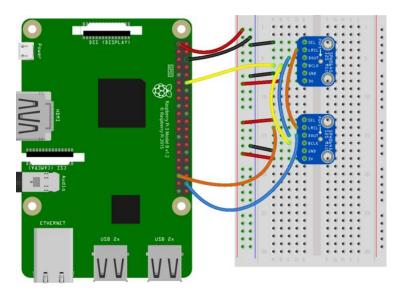


Figure 10. Circuit Connection of the microphones to the Raspberry Pi

The final prototype is shown in Figure 11. The Next section discusses secondary path effects modelling, the experimental setup used to obtain the model.

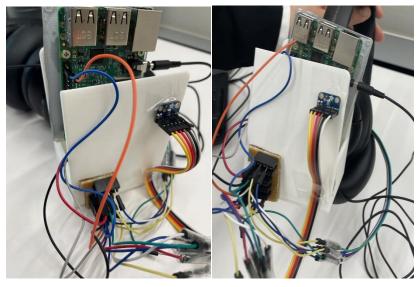


Figure 11. Prototype

Secondary Path Effects Modelling

The model S(z) can be obtained using adaptive filters in system identification as shown in figure 12. S(z) can be modelled either offline (separately) or simultaneously with the primary path model. Offline modelling removes the primary path during the identification of S(z). The simultaneous modelling approach yields a more effective noise-cancelling system, nevertheless, it requires a high-performance processing unit, as any processing delays during can diminish the system's effectiveness [3,4]. In this project, S(z) was modelled separately due to limited capabilities of the processing unit used.

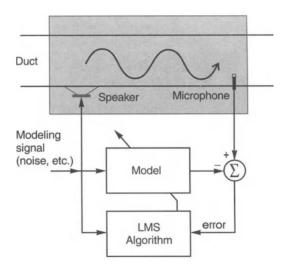


Figure 12. Adaptive filters in system identification, modelling the secondary path effects, S(z) [4].

As figure 4 illustrated, both the actual and adaptive FIR models are excited using a known signal, s[n]. (therefore, requiring no reference microphone to capture it). Ideally, the medium between the noise source and the microphone is a large duct, eliminating any environmental interferences in the modelling. Eventually, the converged model would account for delays, distortion and noise contributions by the non-idealizations of the system and the phase shift experienced by the sound signal as travels from the noise source to the microphone, which is P(z). Figure 13 shows the experimental setup used in modelling S(z). The code used in modelling the secondary path effects is given in Appendix C.

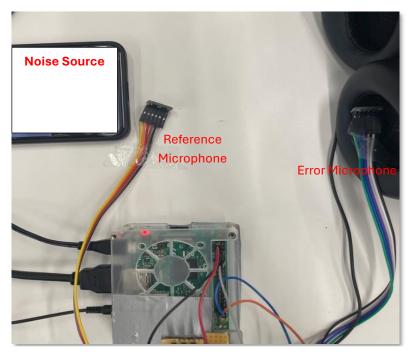


Figure 13. Experimental Setup to obtain the S(z) Model

Figure 14 shows the terminal output results. The obtained FIR coefficients of S(z) were saved in a txt file and were used as a fixed filter model in the main adaptive filter code. Due to an unresolved error in the main code, we could not verify the working of the adaptive filtering headphones. Since the codes for both the secondary path effects modelling and the main code are very similar, if given enough time, we would have been able to troubleshoot the issue easily.

```
Estimated Secondary Path Model (FIR Coefficients):

Estimated Secondary Path Model (FIR Coefficients):

Estimated Secondary Path Model (FIR Coefficients):

B. 38373486-09 B. 38373678-09]

Goofficients saved to secondary_path_coefficients.txt

Expression 'phtread_join( self-othread, Spret )' failed in 'src/os/unix/pa_unix_util.c', line: 441

Expression 'phtread_join( self-othread, Spret )' failed in 'src/os/unix/pa_unix_util.c', line: 442

Expression 'phtread_join( self-othread, Spret )' failed in 'src/os/unix/pa_unix_util.c', line: 442

Expression 'phtread_join( self-othread, Spret )' failed in 'src/os/unix/pa_unix_util.c', line: 442

Expression 'phtread_join( self-othread, Spret )' failed in 'src/os/unix/pa_unix_util.c', line: 3102

Stream terminated after convergence.

Program endod.

(gyvenv) dsp8raspberrypi:~ $
```

Figure 14. Terminal Output Showing the FIR Coefficient of the Secondary Path Effects model, S(z)

Conclusion and Future Work

Although we were unable to resolve the issue in the main code, this project provided a valuable journey into adaptive noise cancellation systems. In this project, we explored the concept of adaptive noise cancellation systems, and their practical implementation through the FxLMS algorithm. Addressing realistic problems and testing real-world systems taught us a lot about noise-cancellation technology. The project gave us the opportunity to take theoretical knowledge in adaptive signal processing and apply it to a practical application. Simulation, optimization of parameters, and testing of real-time systems are some of the critical skills that we acquired, which will be helpful in future engineering projects. The errors encountered in both hardware and software integration improved our problem-solving ability and teamwork.

Future work may give a deeper focus on optimizing the three key parameters, filter tap size, buffer size, and step size, which can yield better noise cancellation performance. Additionally, more complex and efficient methods found in the literature may be investigated, including adaptive gain algorithms and FPGA-based implementations for possible inclusions into the design [9,10]. This will further enhance system performance, lessen computational loads, and enrich real-time responsiveness.

Moreover, better and a wider range of testing methodologies shall be developed for the assessment of the performance of the system based on approaches considered in earlier works. Such advanced metrics would give further insights into how noise cancellation would perform in field conditions under variable environments.

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Appendices

Appendix A: Real – Time Processing Code

```
import threading
import numpy as np
import sounddevice as sd
from adafilt import MultiChannelBlockLMS
from vispy import app, scene
from vispy.scene.visuals import Text
from vispy.scene.widgets import Grid, ViewBox
print(sd.query devices())
input_device = 6 # Input device (microphones)
output device = 6 # Output device (speaker via headphone jack)
device = sd.query devices(input device)
print(device)
samplerate = device["default samplerate"]
print(samplerate)
latency = "low"
dtype = "float32"
channels = inch, outch = (2, 1) # 2 input channels (reference and error),
  = np.zeros((blocksize, 1)) # Error signal (difference between desired
```

```
y = np.zeros((blocksize, 1)) # Output signal (anti-noise generated by
       self.pos1 = np.zeros((N, 2)) # Positions for plotting filter
       self.pos1[:, 0] = np.arange(N) / samplerate # Time axis
       self.pos2[:, 0] = np.arange(N) / samplerate # Time axis
       main grid = canvas.central widget.add grid()
       info = scene.widgets.ViewBox(border color="b")  # Viewbox for
       x axis = scene.AxisWidget(orientation="bottom") # Bottom axis for
       y axis = scene. Axis Widget (orientation="left") # Left axis for
       grid.add widget(x axis, row=1, col=1)
       grid.add widget(y axis, row=0, col=0)
        self.line1 = scene.Line(self.pos1, self.color1,
```

```
t=viewbox.scene)
     self.pos1[:, 1] = w[:, 0, 0] # Update first filter weight plot
self.pos2[:, 1] = w[:, 0, 1] # Update second filter weight plot
self.line1.set_data(pos=self.pos1, color=self.color1)
self.line2.set_data(pos=self.pos2, color=self.color2)
self.text.text = f"CPU load: {load * 100:.1f}%" # Update CPU load
      y[:] = filt.filt(fx[:, None])
      filt.adapt(fx[:, None], e)
      filter monitor.update(filt.w, stream.cpu load)
      print(type(e).
```

```
# Event for thread synchronization (ensure stream has finished processing)
callback_finished_event = threading.Event()

# Stream setup for real-time audio processing (using the selected
input/output devices)
stream = sd.Stream(
    device=(input_device, output_device), # Set input and output devices
    samplerate=samplerate, # Set the sample rate
    blocksize=blocksize, # Set the block size (number of samples per
block)
    dtype=dtype, # Set the data type (32-bit float)
    latency=latency, # Set the latency for real-time processing
    channels=channels, # Set the number of input and output channels
    callback=callback, # Set the callback function for processing audio
    finished_callback=callback_finished_event.set, # Event when the stream
finishes
)

try:
    # Run the audio stream and start the real-time FxLMS process
    with stream:
        app.run() # Run the Vispy app for visualization
except KeyboardInterrupt:
    sys.exit(0) # Exit gracefully on keyboard interrupt
    plt.plot(filt.w)
plt.show()
```

Appendix B: Real – Simulation Code

```
import numpy as np
import matplotlib.pyplot as plt
from adafilt.utils import wgn
def setup paths(scenario):
    if scenario == "Scenario 1":
    elif scenario == "Scenario 2":
    elif scenario == "Scenario 3":
    return h pri, h sec
length = 64
blocklength = 4
n buffers = 2000
estimation phase = 500
blocklength)
all results = {}
        h pri, h sec = setup paths(scenario)
            blocklength, predefined_noise_signal, h_pri=h_pri, h_sec=h_sec,
noise=wgn(predefined noise signal, 20, "dB")
        plant model = FIRFilter(np.zeros(blocklength + length))
        adaptive_plant_model = FastBlockLMSFilter(length, blocklength,
```

```
if i < estimation phase:</pre>
                 v = np.random.normal(0, 1, blocklength) # Estimation noise
                 v = np.random.normal(0, 0.01, blocklength) # Operation
                 adaptive plant model.stepsize = step size
            x, e, u, d = sim.playrec(-y + v)
plant_model.w[blocklength:] = adaptive_plant_model.w
                 filt.adapt(fx, e)
             elog.append(np.abs(e))
        scenario results[scenario] = np.array(elog)
for step size, results in all results.items():
        rms error = np.sqrt(np.convolve(elog.flatten() ** 2, np.ones(50) /
    plt.legend()
    plt.show()
```

Appendix C: Secondary Path Effects Modelling Code

```
import numpy as np
import sounddevice as sd
print(sd.query devices()) # Query available sound devices
input device a = 6 # Input device (reference microphones)
input device b = 6 # Input device (error microphone)
device = sd.query devices(input device a)
print(device)
samplerate = device["default samplerate"]
print(f"Samplerate: {samplerate}")
blocksize = 2048  # Block size for real-time processing
filter_length = blocksize * 4  # Length of the adaptive filter (number of
latency = "low" # Low latency for real-time processing
channels = 2 # Two channels: reference and error microphones
filt = FastBlockLMSFilter(length=filter length, blocklength=blocksize)
h sec estimate = np.zeros(filter length)
previous weights = np.zeros(filter length)
    global h sec estimate, filt, previous weights, convergence counter
    if status:
```

```
# Compute the error signal (difference between desired and predicted)
filt.adapt(x, e)
if weight change < threshold:</pre>
   convergence counter += 1
   convergence counter = 0 # Reset counter if the change is too large
previous weights = filt.w.copy()
if convergence_counter >= stable_iterations:
device=(input device a, input device b),
blocksize=blocksize,
callback=callback,
with stream:
    sd.sleep(10000) # Run for 10 seconds (or however long needed for
```

Appendix D: Steps for Setting up the Raspberry Pi and Running the Code

Setting Up the Raspberry Pi

- Using the Raspberry Pi Imager application the operating system was downloaded onto a SD card. The RASPBERRY PI OS (32-BIT) was chosen for this project as it a better fit for Raspberry Pi 3 that is used for this project.
- The Microphone was configured on to the Raspberry Pi by going to the configuration file of the Raspberry pi.
 - On the terminal the following needs to types to access that file /boot/firmware/config.txt
 - Once in the file the following text was added to ended of the document to configure the microphones.

dtoverlay=googlevoicehat-soundcard

• The headphones did not require configuring.

Running the Program

- In order to run the program on Raspberry Pi a Python virtual in environment was required and it had to created on the terminal.
 - 1. The package required to use virtual environment is first downloaded.

 pip install virtualenv
 - 2. Then a virtual environment is created named *myenv*.

python -m venv myenv

- 3. To access the environment the through the terminal the following is typed. source myenv/bin/activate
- 4. Before running the code all the necessary libraries need to downloaded within the environment. The ones needed for running the system_identification.py and real-time-visuals-2-channel.py were the following.

```
pip install numpy
pip install matplotlib
pip install sounddevice
pip install vispy
```

- 5. To run a code within the environment the terminal needs to run the program.

 *python /home/dsp/projectdsp/system identification.py
- The system_identification.py program was then run using the environment. This code estimates the secondary path model in real-time for adaptive noise cancellation by using an adaptive filter to process reference and error microphone signals, identifying filter coefficients when convergence is achieved.

python/home/dsp/projectdsp/system identification.py

• The real-time-visuals-2-channel.py which is meant to implement the real-time adaptive noise cancellation system using the FxLMS algorithm, processing audio input to generate anti-noise output while visualizing filter weights and CPU load. It uses the coefficients found in the system_idenfication.py program which is stored in on a txt file to do the calculations.

python /home/dsp/projectdsp/ real-time-visuals-2-channel.py