# A Term Project Report

in

Control Engineering for

**Robotics, RA-602** 

offered by



# Department of CICPS

Submitted to

# Dr. Parijat Bhowmick

Department of Electrical and electronics Engineering

# Indian Institute of Technology Guwahati

Submitted by

Tabrej Alam (224102113), Shivraj Vishwakarma (224102112), Ram Gopal (224102111), Mahendra Patel (224102107) & Ghanshyam Das Gupta (224102109)

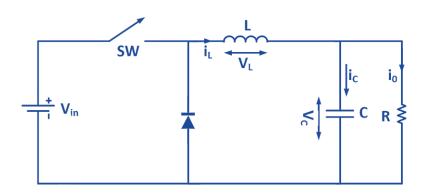
### **OBJECTIVE**

In the dc-dc(buck) converter with L=10 mH, C=1 mF, R=100ohm, a 50 percent PWM duty cycle, and assuming the system's output is the voltage across the capacitor.

- a) To write the converter's equation in the form of State Space model.
- **b)** To Find the system's transfer function.
- c) To Express the system's state equation in phase variable form.
- d) To Design a state feedback gain for the system with given closed loop poles  $s = -2+j2\sqrt{3}$ ,  $s = -2-j2\sqrt{3}$ ,
- e) To Design an observer for the dc-dc converter with the given specification: overshoot = 40 percent, settling time = 0.3 sec.
- f) To Simulate the system and observer for a unit step input using Simulink. Assuming that the initial conditions for the original system are x(0) = (2|1). The observer should have initial conditions x'(0) = (0|0).

> **SOFTWARE:** MATLAB R2022a

# a) State Space model:



From the circuit of Buck converter,

When switch is ON:

$$V_{L} = V_{in} - V_{c} \Rightarrow L \frac{di_{L}}{dt} = V_{in} - V_{c} \Rightarrow \frac{di_{L}}{dt} = \frac{V_{in}}{L} - \frac{V_{c}}{L}$$

$$(1)$$

$$i_c = i_L - \frac{V_c}{R} \Rightarrow C \frac{dV_c}{dt} = i_L - \frac{V_c}{R} \Rightarrow \frac{dV_c}{dt} = \frac{i_L}{C} - \frac{V_c}{RC}$$
 (2)

From equation 1 & 2

$$\begin{bmatrix} i_L' \\ V_c' \end{bmatrix} = \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix} \begin{bmatrix} i_L \\ V_c \end{bmatrix} + \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} V_{in}$$
(3)

When switch is off:

$$V_L = -V_c \Rightarrow L \frac{di_L}{dt} = -V_c \Rightarrow \frac{di_L}{dt} = -\frac{V_c}{L}$$
 (4)

$$i_c = i_L - \frac{V_c}{R} \Rightarrow C \frac{dV_c}{dt} = i_L - \frac{V_c}{R} \Rightarrow \frac{dV_c}{dt} = \frac{i_L}{C} - \frac{V_c}{RC}$$
 (5)

Combining equation 4 & 5

$$\begin{bmatrix} i_L' \\ V_c' \end{bmatrix} = \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix} \begin{bmatrix} i_L \\ V_c \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \end{bmatrix} V_{in}$$
(6)

After deriving the buck converter state space, A and B matrix for its 'ON' and 'OFF' state. It is required to find its average A and B matrix with the account of switching duty cycle D.

So as,

$$A = D * A_{(ON)} + (1 - D) * A_{(OFF)}$$
(7)

$$A = D * \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix} + (1 - D) * \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix} \Rightarrow A = \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix}$$
(8)

And

$$B = D * B_{(ON)} + (1 - D) * B_{(OFF)}$$
(9)

$$B = D * \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} + (1 - D) * \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow B = \begin{bmatrix} \frac{D}{L} \\ 0 \end{bmatrix}$$
 (10)

Finally, we have:

$$\begin{bmatrix} i_L' \\ V_c' \end{bmatrix} = \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix} \begin{bmatrix} i_L \\ V_c \end{bmatrix} + \begin{bmatrix} \frac{D}{L} \\ 0 \end{bmatrix} V_{in}$$
(11)

To obtain the output of system as the voltage across capacitor  $(V_c)$ 

$$Y = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} i_L \\ V_c \end{bmatrix} \tag{12}$$

Let's assume state variable as

$$i_L = x_1$$
,  $V_c = x_2$  and output of system =Y

The Buck converter state space equation will be given as:

$$\begin{bmatrix} x_1' \\ x_2' \end{bmatrix} = \begin{bmatrix} 0 & \frac{-1}{L} \\ \frac{1}{C} & \frac{-1}{RC} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \frac{D}{L} \\ 0 \end{bmatrix} V_{in}$$
(13)

$$Y = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \tag{14}$$

For the above system i.e. (13) & (14) using the given values as

state space equation can be obtained as:

$$\begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 0 & -100 \\ 1000 & -10 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 50 \\ 0 \end{bmatrix} V_{in}$$
 (15)

$$Y = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \tag{16}$$

#### b) Transfer Function:

Now the Transfer function of this system is can be obtained using:

$$\frac{Y(s)}{U(s)} = C(sI - A)^{-1}B + D \tag{17}$$

Where

$$A = \begin{bmatrix} 0 & -100 \\ 1000 & -10 \end{bmatrix}, \quad B = \begin{bmatrix} 50 \\ 0 \end{bmatrix}, \quad C = \begin{bmatrix} 0 & 1 \end{bmatrix} & \& \quad D = 0$$

So now,

$$(SI - A)^{-1} = \left[ s * \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} - \begin{pmatrix} 0 & -100 \\ 1000 & -10 \end{pmatrix} \right]^{-1} = \left[ \begin{matrix} s & 100 \\ -1000 & s + 10 \end{matrix} \right]^{-1}$$

$$= \frac{1}{s(s+10)+10^5} \left[ \begin{matrix} s+10 & -100 \\ 1000 & s \end{matrix} \right]$$

And 
$$C(sI - A)^{-1}B = \begin{bmatrix} 0 & 1 \end{bmatrix} * \frac{1}{s(s+10)+10^5} \begin{bmatrix} s+10 & -100 \\ 1000 & s \end{bmatrix} * \begin{bmatrix} 50 \\ 0 \end{bmatrix}$$
$$= \frac{1}{s(s+10)+10^5} * \begin{bmatrix} 0 & 1 \end{bmatrix} * \begin{bmatrix} 50(s+10) \\ 50000 \end{bmatrix}$$

$$=\frac{50000}{s(s+10)+10^5}$$

So, the Transfer function of the system is

$$T(s) = \frac{50000}{s(s+10)+10^5} \tag{18}$$

#### c) Phase Variable Canonical Form:

$$T(s) = \frac{Y(s)}{U(s)} = \frac{50000}{s^2 + 10s + 10^5} \Rightarrow \frac{V(s)}{U(s)} * \frac{Y(s)}{V(s)} = \frac{1}{s^2 + 10s + 10^5} * 50000$$

U(s) 
$$\frac{1}{s^2 + 10s + 10^5}$$
  $\frac{V(s)}{50000}$ 

$$\frac{V(s)}{U(s)} = \frac{1}{s^2 + 10s + 10^5} \tag{19}$$

$$(s^{2} + 10s + 10^{5})V(s) = U(s) \Rightarrow s^{2}V(s) + 10sV(s) + 10^{5}V(s) = U(s)$$
$$\Rightarrow \ddot{V}(t) + 10sV(t) + 10^{5}V(t) = U(s)$$

Let  $x_1 = V(t)$ ,  $x_2 = V(t)$ ,  $\dot{x}_2 = V(t)$ , now we'll get:

$$\dot{x_1} = x_2 \tag{20}$$

$$\dot{x}_2 = -10^5 x_1 - 10 x_2 + U(s) \tag{21}$$

$$\frac{Y(s)}{V(s)} = 50000 \implies Y(s) = 50000V(s) \implies Y(t) = 50000V(t)$$

$$\Rightarrow Y(t) = 50000x_1 \tag{22}$$

Using equation 20, 21 & 22 we can write the phase variable form representation as:

$$\begin{bmatrix} \dot{x_1} \\ \dot{x_2} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -10^5 & -10 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$
 (23)

$$Y(t) = \begin{bmatrix} 50000 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$
 (24)

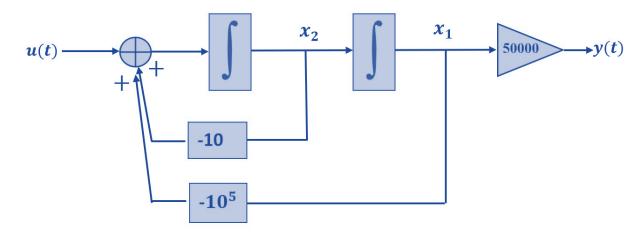


Fig:2 Block diagram representation of Phase variable form

# d) Design of state feedback Gain:

#### > MATLAB CODE:

```
clc;
clear all;
close all;
% Given state model
A = [0 -100; 1000 -10]; B = [50; 0]; C = [0 1]; D = 0;
%% Close loop pole to design state feedback model
J = [-2+1i*3.464 -2-1i*3.464];
% Calculating State feedback gain
K = acker (A, B, J)
```

#### > OUTPUT:

#### e) Observer Design:

For the given system we have,

$$^{\text{GP}}$$
 Overshoot,  $M_p = 40 \% \&$ 

 $rac{1}{2}$  settling time  $t_s = 0.3$  sec.

So now,

$$t_s = \frac{4}{\zeta \omega_n} = 0.3 \implies \zeta \omega_n = 13.33$$

$$M_p = e^{\{\frac{-\pi\zeta}{\sqrt{1-\zeta^2}}\}} = 0.4 \implies \zeta = 0.28087$$

$$\omega_n = \frac{13.33}{0.28087} = 47.4596$$

Now using the standard characteristic equation for second order system as

$$s^{2} + 2\zeta \omega_{n} s + \omega_{n}^{2} = 0$$

$$\Rightarrow s^{2} + 26.6599 s + 2252.4136 = 0$$

$$\Rightarrow s = -13.32995 \pm j45.5491$$
(25)

From (25) we have got the closed loop poles as:

$$(-13.32995 + j45.5491) & (-13.32995 - j45.5491)$$

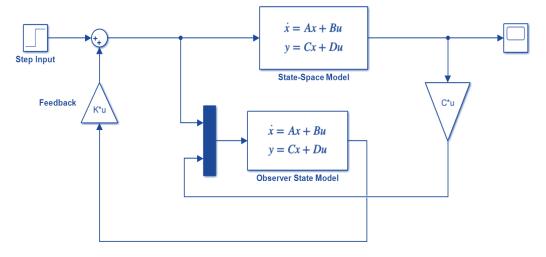
Now with these closed loops pole we will find the observer gain L using acker () command.

#### f) Simulation of Observer:

#### > MATLAB CODE:

```
clc;
clear all;
close all;
% Given state model
A = [0 -100; 1000 -10]; B = [50; 0]; C = [0 1]; D = 0;
%% Close loop pole to design state feedback model
J = [-2+1i*3.464 -2-1i*3.464]
% Calculating State feedback gain
K = acker(A, B, J)
%% Calculated close loop pole to design observer
P = [-13.3298+1i*45.5489 -13.3298-1i*45.5489];
% calculating observer gain matrix
1 = acker (A', C', P);
L = transpose(1)
%% calling Simulink file from .m file
sim('Observer.slx')
% ploting
figure (1)
step (ss (A-L*C, B, C, D))
```

# ➤ Simulink file used in MATLAB code: 'Observer.slx'



#### > OUTPUT

# > Output Response:

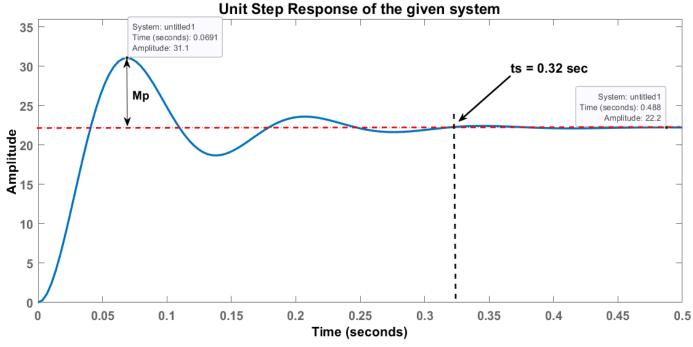


Fig1: step response

#### > System states:

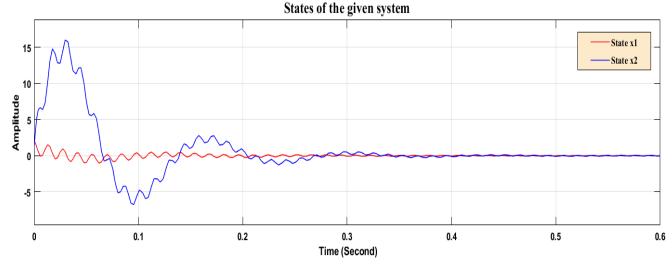


Fig2: System states

#### > CALCULATIONS:

#### Peak Overshoot:

From the obtained output response, we found that the peak value of the response is:

$$C(peak) = 31.1$$

And the steady state value is

$$C(\infty) = 22.2$$

So now we can calculate the peak overshoot as

$$M_p = \frac{C(peak) - C(\infty)}{C(\infty)} * 100\%$$

$$= 40.09\%$$

#### **Settling time:**

As we can see from the step response, the settling times comes out to be

$$t_s = 0.32s$$

#### > CONCLUSION:

We have simulated the system & observer with given specifications and we obtained the step response which has the similar overshoot (40.09%) and settling time (0.32s) as it was given 40% & 0.3s respectively.