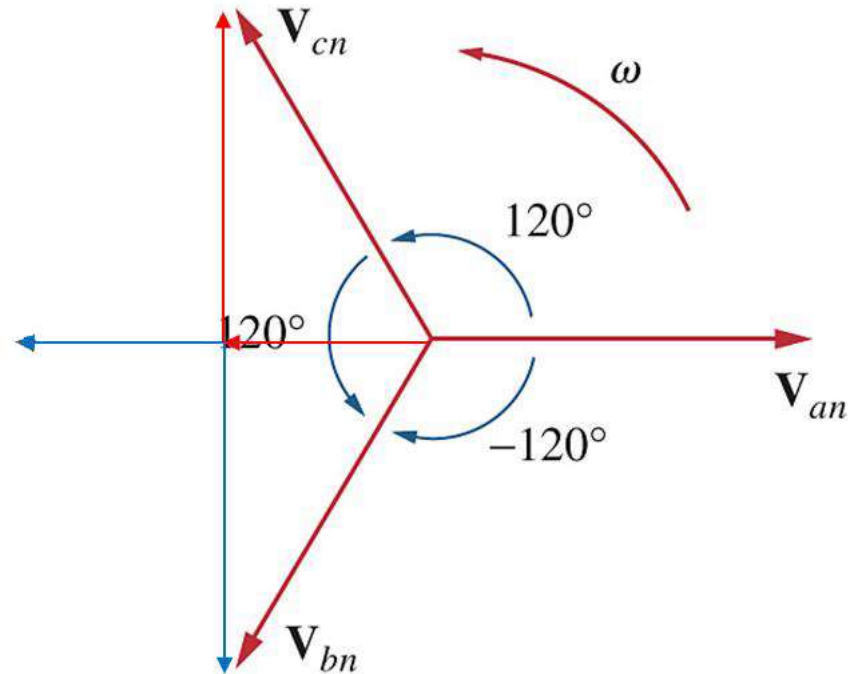


# VE215 RC 6

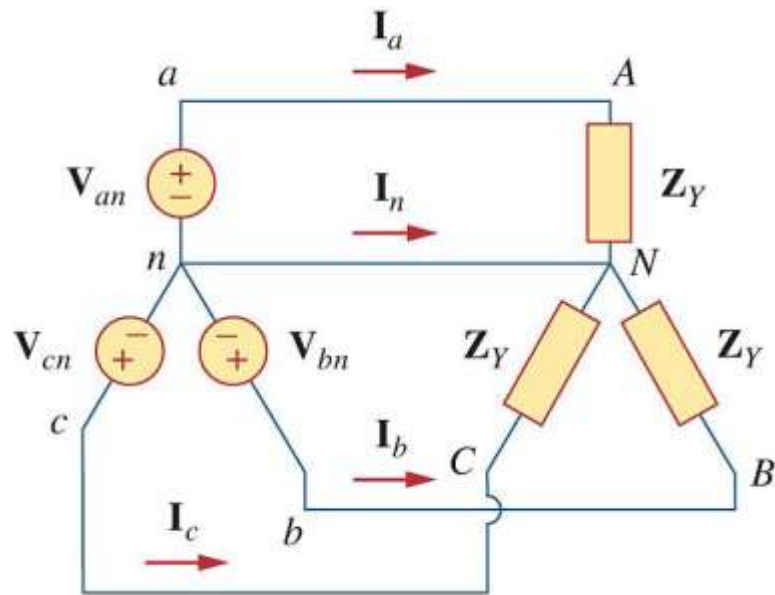
Three-phase circuits and magnetically coupled circuits

# Physical illustration of three-phase



# Balanced Y-Y

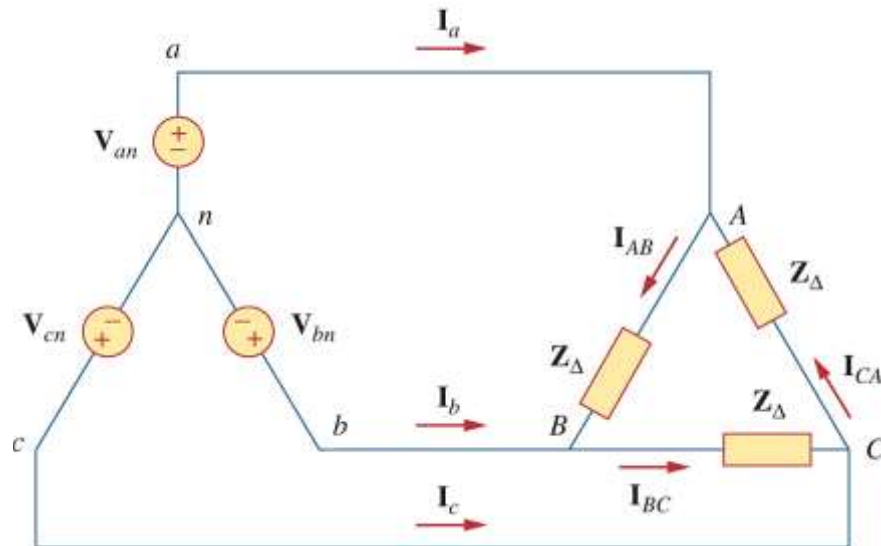
$V_p$  is RMS here  
copy these 4 pages onto cheating paper



Summary of phase and line voltages/currents for balanced three-phase systems.<sup>1</sup>

Connection	Phase voltages/currents	Line voltages/currents
Y-Y	$V_{an} = V_p \angle 0^\circ$ $V_{bn} = V_p \angle -120^\circ$ $V_{cn} = V_p \angle +120^\circ$ Same as line currents	$V_{ab} = \sqrt{3} V_p \angle 30^\circ$ $V_{bc} = V_{ab} \angle -120^\circ$ $V_{ca} = V_{ab} \angle +120^\circ$ $I_a = V_{an} / Z_Y$ $I_b = I_a \angle -120^\circ$ $I_c = I_a \angle +120^\circ$

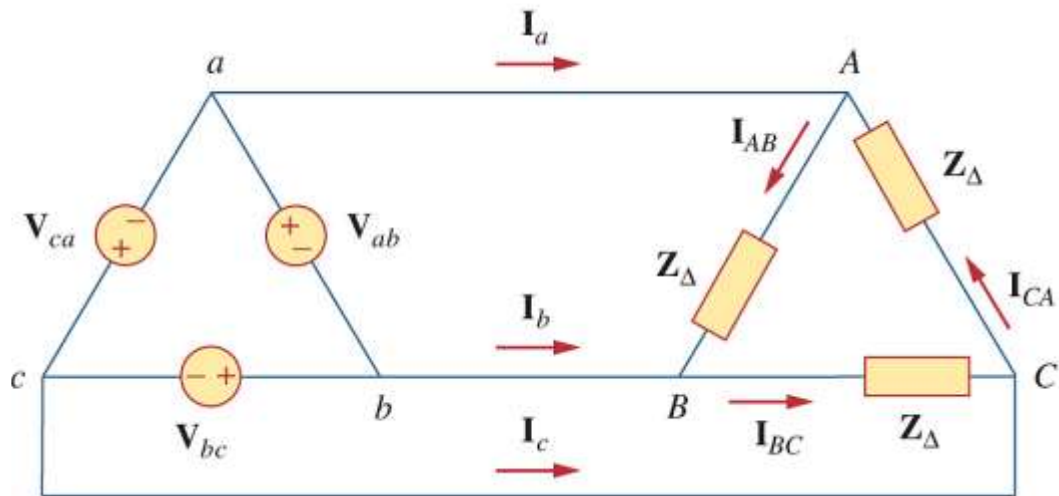
# Balanced Y-Δ



Summary of phase and line voltages/currents for balanced three-phase systems.<sup>1</sup>

Connection	Phase voltages/currents	Line voltages/currents
Y-Δ	$V_{an} = V_p \angle 0^\circ$ $V_{bn} = V_p \angle -120^\circ$ $V_{cn} = V_p \angle +120^\circ$ $I_{AB} = V_{AB} / Z_{\Delta}$ $I_{BC} = V_{BC} / Z_{\Delta}$ $I_{CA} = V_{CA} / Z_{\Delta}$	$\tilde{V}_{ab} = \tilde{V}_{AB} = \sqrt{3} V_p \angle 30^\circ$ $\tilde{V}_{bc} = \tilde{V}_{BC} = \tilde{V}_{ab} \angle -120^\circ$ $\tilde{V}_{ca} = \tilde{V}_{CA} = \tilde{V}_{ab} \angle +120^\circ$ $I_a = I_{AB} \sqrt{3} \angle -30^\circ$ $I_b = I_a \angle -120^\circ$ $I_c = I_a \angle +120^\circ$

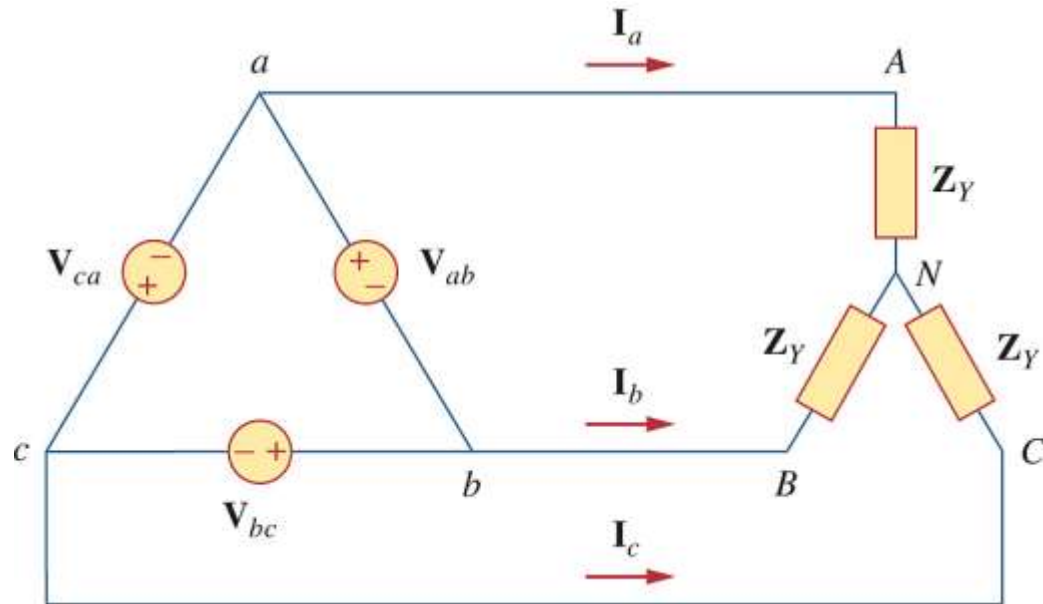
# Balanced $\Delta$ - $\Delta$



Summary of phase and line voltages/currents for balanced three-phase systems.<sup>1</sup>

Connection	Phase voltages/currents	Line voltages/currents
$\Delta$ - $\Delta$	$V_{ab} = V_p \angle 0^\circ$ $V_{bc} = V_p \angle -120^\circ$ $V_{ca} = V_p \angle +120^\circ$ $I_{AB} = V_{ab} / Z_{\Delta}$ $I_{BC} = V_{bc} / Z_{\Delta}$ $I_{CA} = V_{ca} / Z_{\Delta}$	<p>Same as phase voltages</p> $I_a = I_{AB} \sqrt{3} \angle -30^\circ$ $I_b = I_{AB} \angle -120^\circ$ $I_c = I_{AB} \angle +120^\circ$

# Balanced $\Delta$ -Y

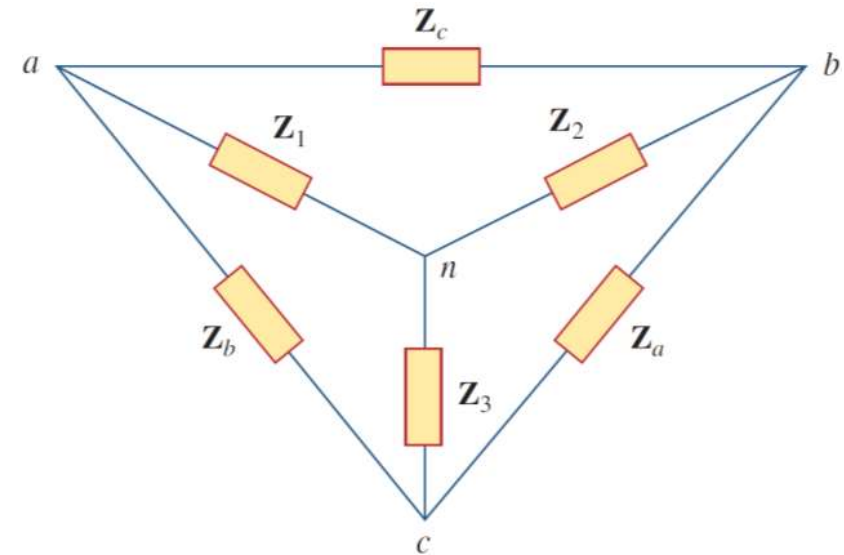
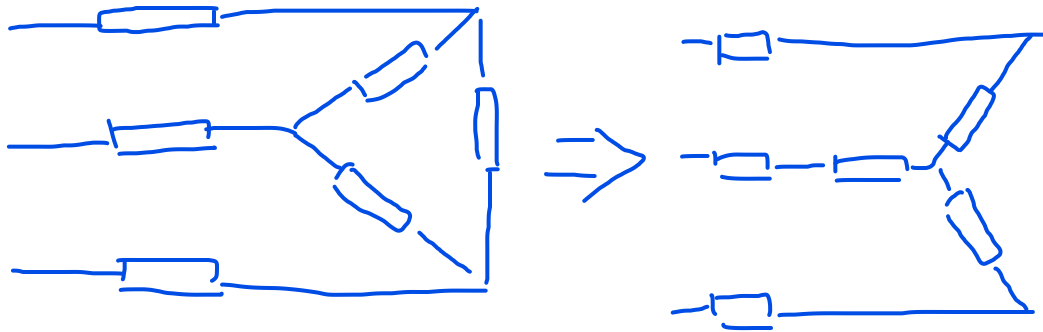


Summary of phase and line voltages/currents for balanced three-phase systems.<sup>1</sup>

Connection	Phase voltages/currents	Line voltages/currents
$\Delta$ -Y	$V_{ab} = V_p \angle 0^\circ$ $V_{bc} = V_p \angle -120^\circ$ $V_{ca} = V_p \angle +120^\circ$  Same as line currents	Same as phase voltages   $I_a = \frac{V_p \angle -30^\circ}{\sqrt{3}Z_Y}$ $I_b = I_a \angle -120^\circ$ $I_c = I_a \angle +120^\circ$

# Balanced Y- $\Delta$ transformation

- $Z_{\Delta} = 3Z_Y$
- A common trick when the lines also consume power in a  $\Delta$  load circuit is to convert the  $\Delta$  load into Y load.



# Three-phase power analysis

- Total instantaneous/average power
- $P = 3V_{Prms}I_{Prms} \cos(\angle Z)$
- Total reactive power
- $Q = 3V_{Prms}I_{Prms} \sin(\angle Z)$
- Total complex power
- $S = 3\widetilde{V_{Prms}}\widetilde{I_{Prms}}^*$

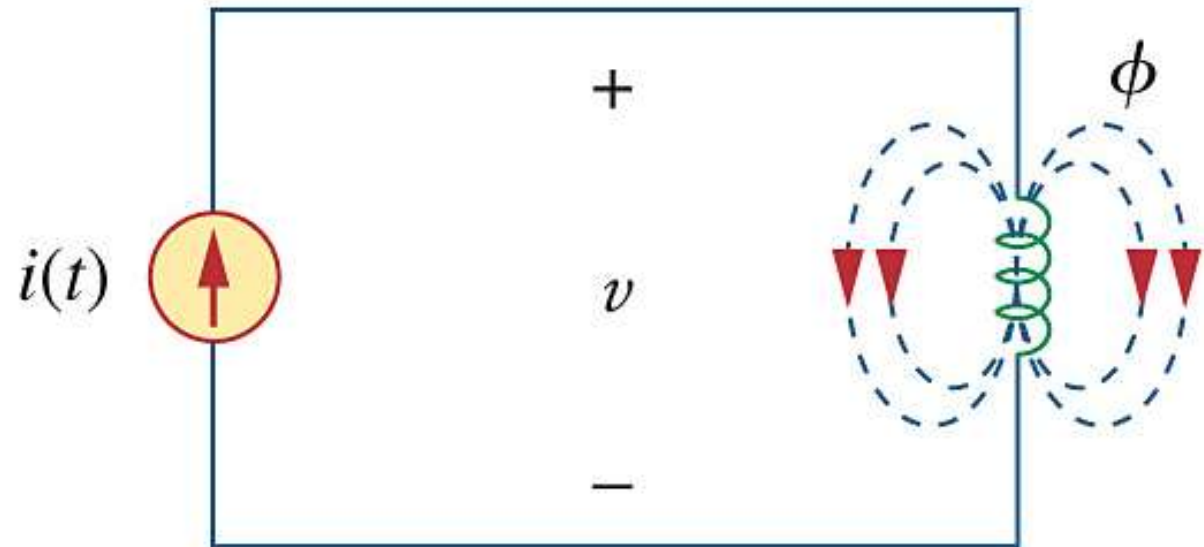


# Three-phase power analysis

- In particular, for Y-Y and  $\Delta$ - $\Delta$  structures:
- $P = 3V_{Prms}I_{Prms} \cos(\angle Z) = \sqrt{3}V_{Lrms}I_{Lrms} \cos(\angle Z)$
- The material to deliver the same power and tolerate the same loss for a three-phase system is 75% of that for a single-phase system.

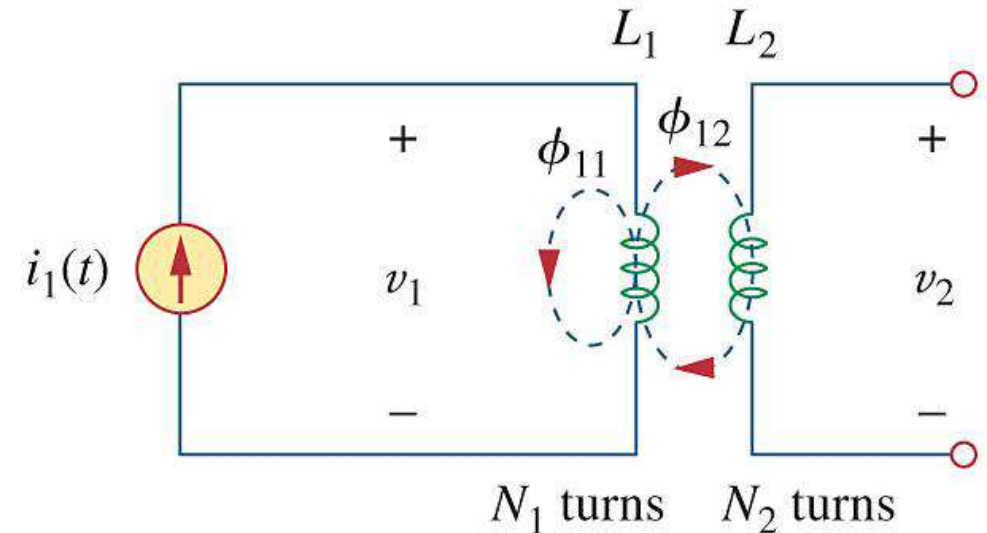
# Self-inductance

- A single coil with  $N$  turns
- Current causes magnetic flux
- $v = N \frac{d\phi}{dt}$
- $L = N \frac{d\phi}{di}$



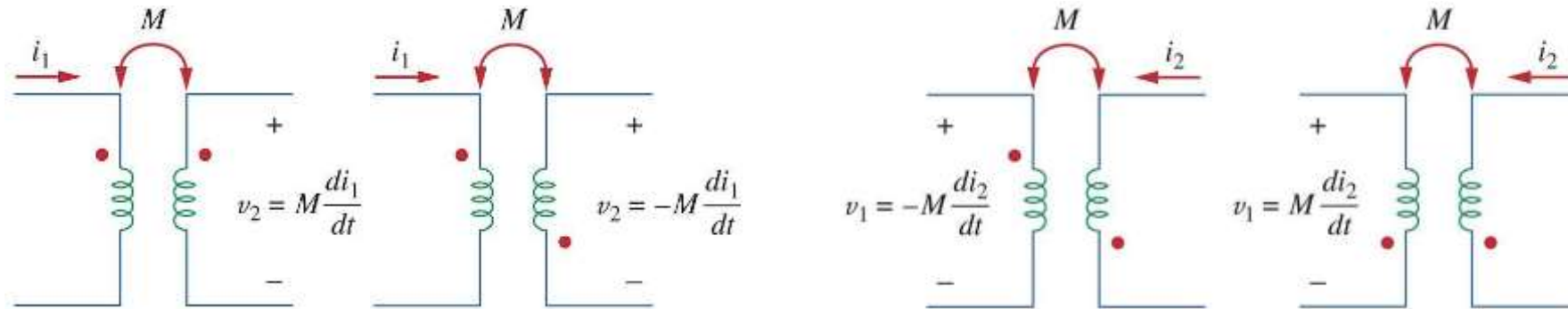
# Mutual inductance

- $v_1 = N_1 \frac{d\phi_1}{dt} = \overset{=L_1}{N_1 \frac{d\phi_1}{di_1} \frac{di_1}{dt}} = L_1 \frac{di_1}{dt}$
- $v_2 = N_2 \frac{d\phi_{12}}{dt} = \overset{=M}{N_2 \frac{d\phi_{12}}{di_1} \frac{di_1}{dt}} = M \frac{di_1}{dt}$
- $M = N_2 \frac{d\phi_{12}}{di_1}$



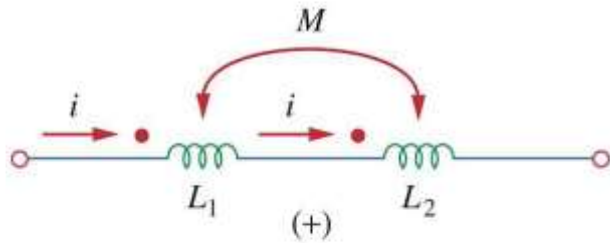
# Dot convention

- If dots are on the **same** side, the direction of the induced voltage difference (from high to low) is the **parallel** to that of the current.
- If dots are on the **opposite** side, the direction of the induced voltage difference (from high to low) is **antiparallel** to that of the current.

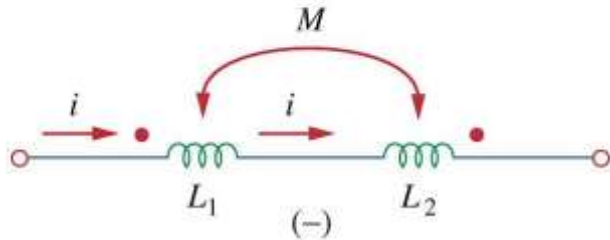


# Dot convention

- $L = L_1 + L_2 + 2M$

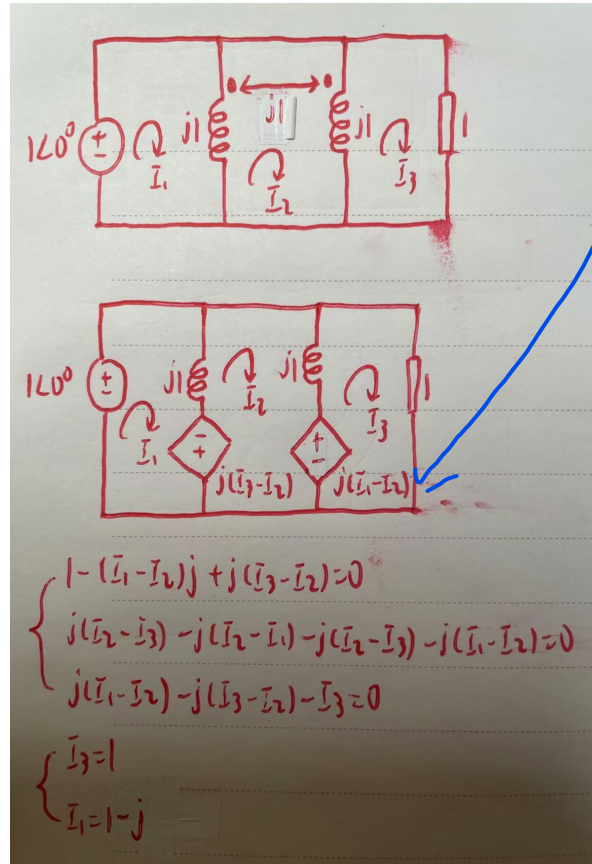
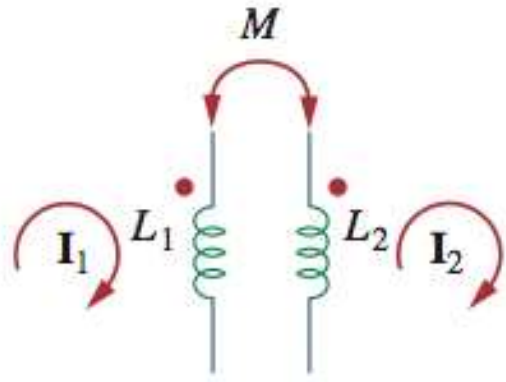


- $L = L_1 + L_2 - 2M$

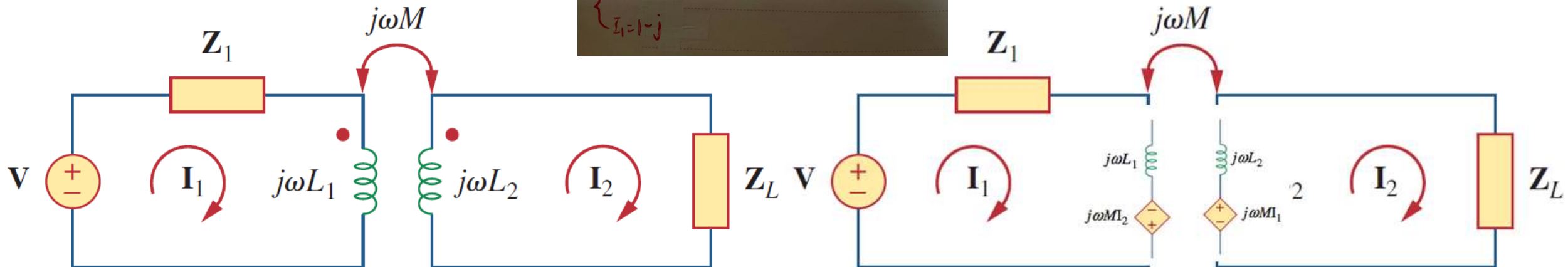
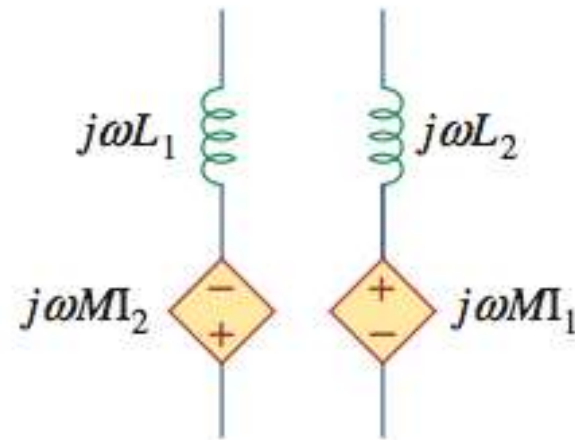


Example: Find  $I_1$  and  $I_3$

# KVL application

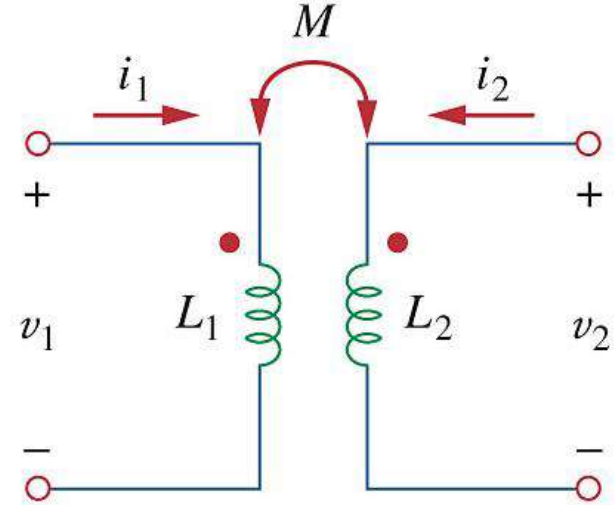


Do not forget to consider currents on both sides !!!

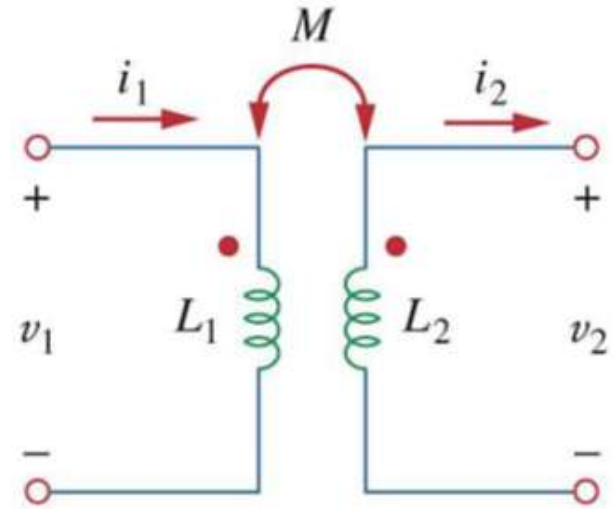


# Energy

- $w = \frac{1}{2}L_1I_1^2 + \frac{1}{2}L_2I_2^2 + MI_1I_2$



- $w = \frac{1}{2}L_1I_1^2 + \frac{1}{2}L_2I_2^2 - MI_1I_2$



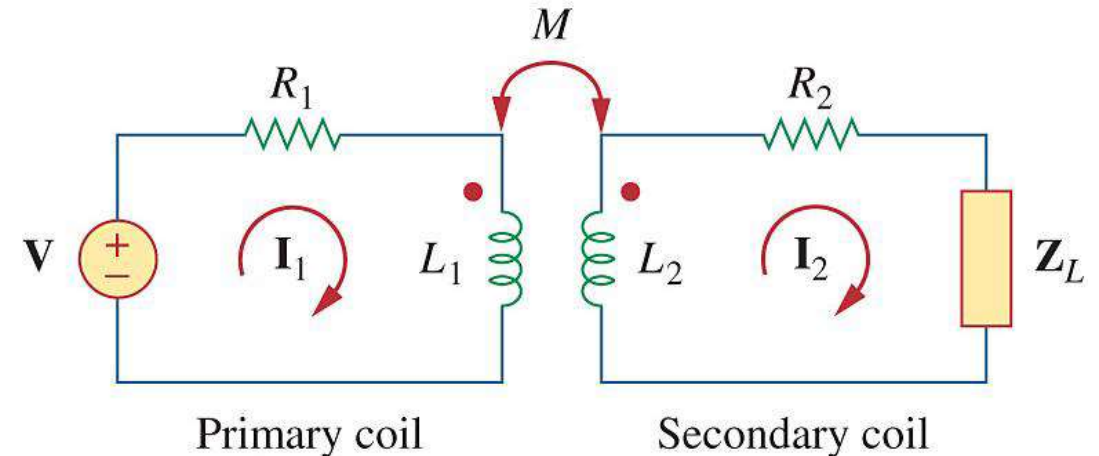
# Coupling coefficient

- $k = \frac{M}{\sqrt{L_1 L_2}} \in [0,1]$
- $k < 0.5$ , loosely coupled
- $k > 0.5$ , tightly coupled



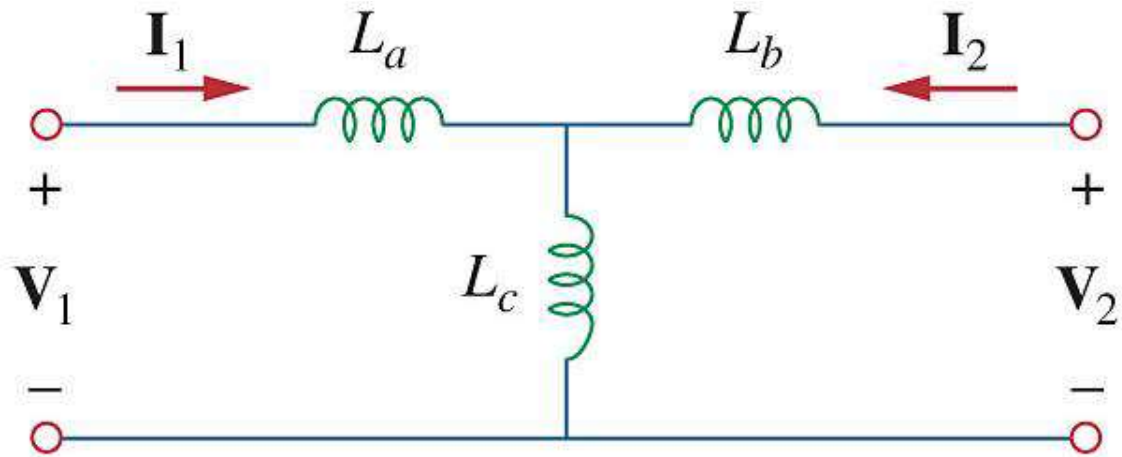
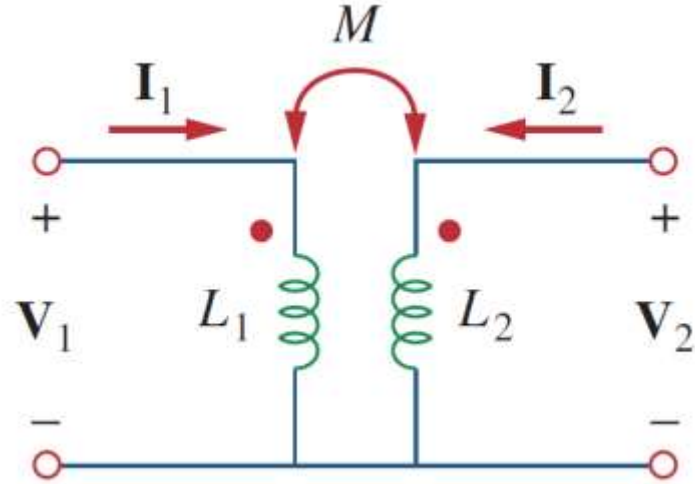
# Linear transformer

- $V = (R_1 + j\omega L_1)I_1 - j\omega M I_2$
- $0 = -j\omega M I_1 + (R_2 + j\omega L_2 + Z_L)I_2$
- $I_2 = \frac{j\omega M I_1}{R_2 + j\omega L_2 + Z_L}$
- $Z_{in} = \frac{V}{I_1} = R_1 + j\omega L_1 + \frac{\omega^2 M^2}{R_2 + j\omega L_2 + Z_L}$



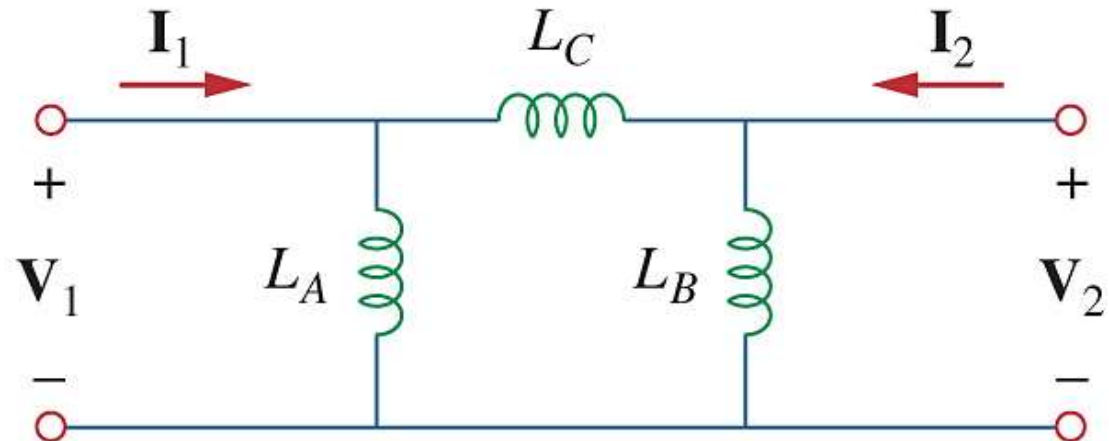
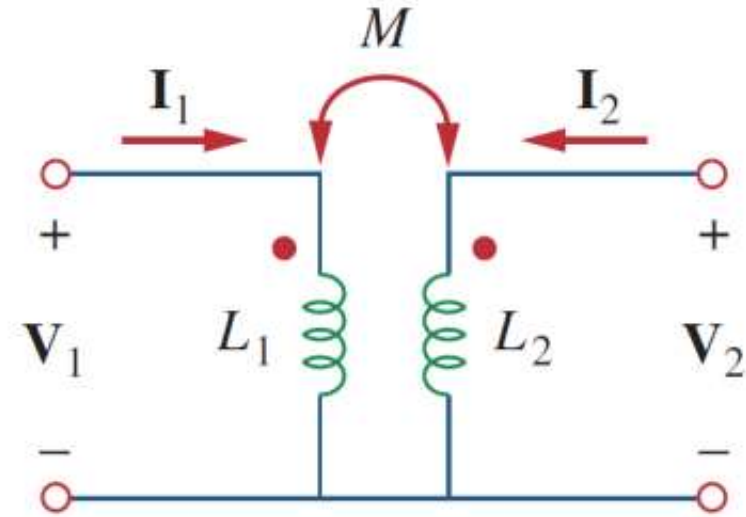
# Linear transformer

- $V_1 = j\omega L_1 I_1 + j\omega M I_2$
- $V_2 = j\omega M I_1 + j\omega L_2 I_2$
- Equivalent T circuit
  - $L_a = L_1 - M$
  - $L_b = L_2 - M$
  - $L_c = M$



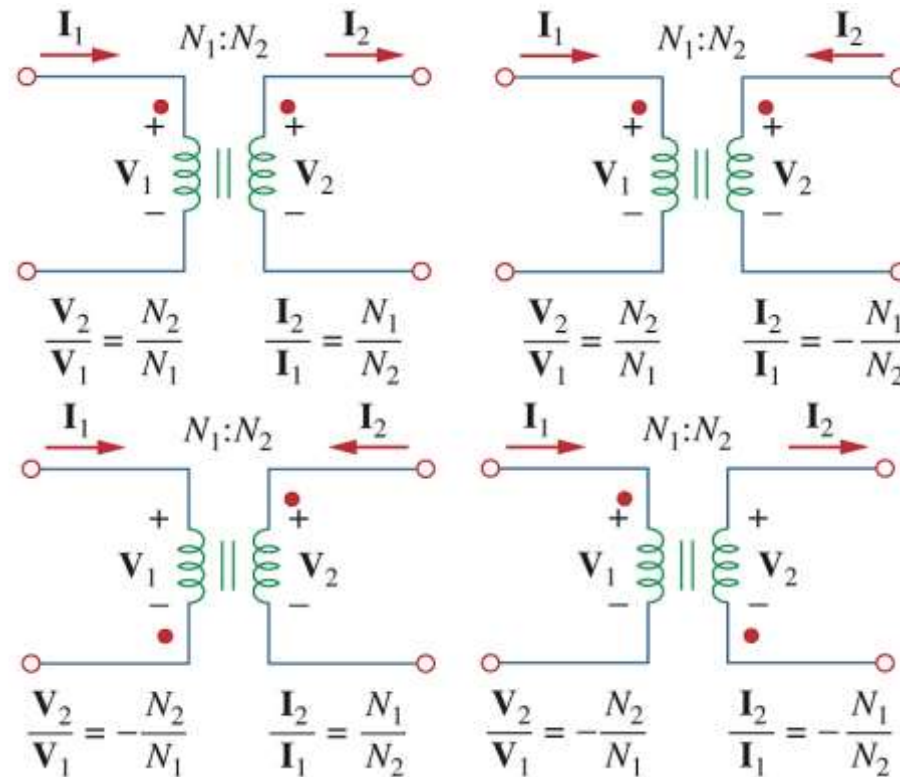
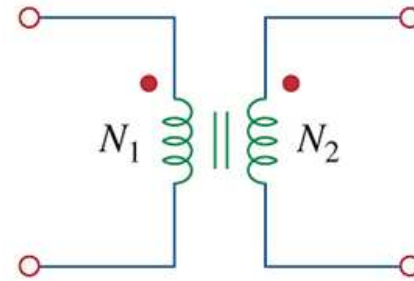
# Linear transformer

- $V_1 = j\omega L_1 I_1 + j\omega M I_2$
- $V_2 = j\omega M I_1 + j\omega L_2 I_2$
- Equivalent  $\pi$  circuit
- $L_A = \frac{L_1 L_2 - M^2}{L_2 - M}$
- $L_B = \frac{L_1 L_2 - M^2}{L_1 - M}$
- $L_C = \frac{L_1 L_2 - M^2}{M}$



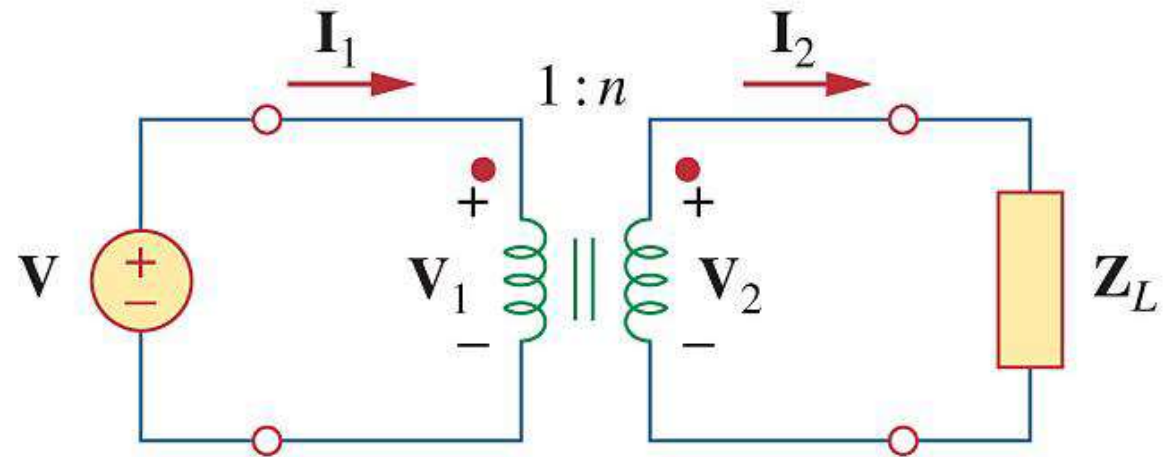
# Ideal transformer

- $k = 1$
- $R_1 = R_2 = 0$
- $L_1 \rightarrow \infty$
- $L_2 \rightarrow \infty$
- $M \rightarrow \infty$



# Ideal transformer

- $S_1 = \tilde{V}_1 \tilde{I}_1^* = \frac{\tilde{V}_2}{n} (n \tilde{I}_2)^* = \tilde{V}_2 \tilde{I}_2^* = S_2$
- $Z_{in} = \frac{V_1}{I_1} = \frac{V_2}{n^2 I_2} = \frac{Z_L}{n^2}$



# Ideal transformer

