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# **Open Lockdown Period Practice Series for GATE**

(Also useful for ESE & Other Exams)

## EC : ELECTRONICS ENGINEERING

TEST No. - 01 | NETWORK THEORY

## Read the following instructions carefully

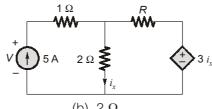
1. This question paper contains 33 MCQ's & NAQ's. Bifurcation of the questions are given below:

| Subjectwise Test Pattern |                             |                  |                     |       |                         |                     |  |
|--------------------------|-----------------------------|------------------|---------------------|-------|-------------------------|---------------------|--|
| Questions                | Question Type               |                  | No. of<br>Questions | Marks | Total<br>Marks          | Negative<br>Marking |  |
| 1 to 10                  | Multiple Choice Ques.       |                  | 10                  | 1     | 10                      | 0.33                |  |
| 11 to 16                 | Numerical Answer Type Ques. |                  | 6                   | 1     | 6                       | None                |  |
| 17 to 26                 | Multiple Choice Ques.       |                  | 10                  | 2     | 20                      | 0.66                |  |
| 27 to 33                 | Numerical Answer Type Ques. |                  | 7                   | 2     | 14                      | None                |  |
| Total Questions : 33     |                             | Total Marks : 50 |                     | T     | Total Duration : 90 min |                     |  |

2. Choose the closest numerical answer among the choices given.

## Multiple Choice Questions: Q.1 to Q.10 carry 1 mark each

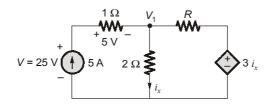
Q.1 For the circuit shown below, if the voltage (V) across the current source of 5 A is 25 V, then the value of 'R' is



- (a)  $1 \Omega$
- (c) 3 Ω

(b)  $2\Omega$  (d)  $4\Omega$ 

1. (b)



Applying Nodal analysis

$$\Rightarrow \frac{V_1 - 3i_x}{R} + i_x = 5 \text{ A}$$

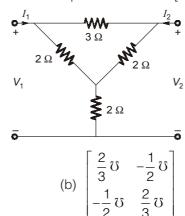
Also 
$$\begin{aligned} V_1 &= 2i_x = 20 \text{ V} \\ i_x &= 10 \text{ A} \end{aligned}$$

$$\therefore \frac{2i_x - 3i_x}{R} + i_x = 5 \text{ A}$$

$$\frac{20 - 30}{R} + 10 = 5 \text{ A}$$

$$R = 2 \Omega$$

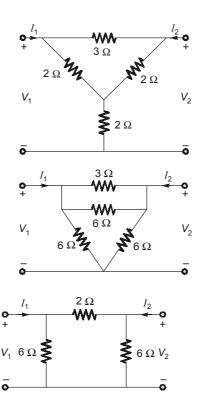
Q.2 For the circuit shown below, the admittance parameter matrix [Y] will be



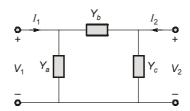
- (a)  $\begin{bmatrix} \frac{2}{3} & \frac{1}{2} & \frac{1}{2} \\ \frac{1}{2} & \frac{2}{3} \end{bmatrix}$
- (c)  $\begin{bmatrix} \frac{2}{3} & 0 & -1 \\ -1 & \frac{2}{3} & 0 \end{bmatrix}$

(d)  $\begin{bmatrix} \frac{2}{3} & 0 & 1 \\ 1 & \frac{2}{3} & 0 \end{bmatrix}$ 

#### 2. (b)



 $\text{for }\Pi\text{-network}$ 



$$Y_b = \frac{1}{2} \Im$$
$$Y_c = \frac{1}{6} \Im$$

$$\therefore \qquad [Y] = \begin{bmatrix} \frac{2}{3} \mho & -\frac{1}{2} \mho \\ -\frac{1}{2} \mho & \frac{2}{3} \mho \end{bmatrix}$$

- If  $V(t) = 2\cos(10t)(1 + 2\cos 5t)$  V, then the rms value of  $(V(t) + \sqrt{3})$  is Q.3
  - (a)  $3\sqrt{3} \text{ V}$

(b)  $3\sqrt{2} \text{ V}$ 

(c) 3 V

(d) 4 V

3. (c)

$$f(t) = V(t) + \sqrt{3}$$

$$V(t) = 2\cos 10t + 4\cos 10t \cdot \cos 5t$$
  
= 2\cos 10t + 2\cos 15t + 2\cos 5t

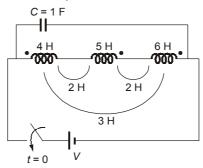
$$f(t) = \sqrt{3} + 2\cos 10t + 2\cos 15t + 2\cos 5t$$

rms 
$$\{f(t)\}\ = \sqrt{(\sqrt{3})^2 + \left(\frac{2}{\sqrt{2}}\right)^2 + \left(\frac{2}{\sqrt{2}}\right)^2 + \left(\frac{2}{\sqrt{2}}\right)^2}$$
  

$$= \sqrt{3 + \frac{4}{2} + \frac{4}{2} + \frac{4}{2}}$$
  

$$= \sqrt{3 + 6} = \sqrt{9} = 3 \text{ V}$$

**Q.4** For the tank circuit shown below, the time period of oscillations is



- (a)  $2\pi$  seconds
- (c)  $6\pi$  seconds

- (b)  $4\pi$  seconds
- (d)  $8\pi$  seconds

4. (c)

Time period 
$$(T) = \frac{2\pi}{\omega}$$

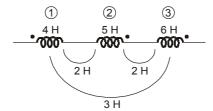
where

$$\omega = \frac{1}{\sqrt{LC}}$$

thus,

$$T = 2\pi\sqrt{LC}$$

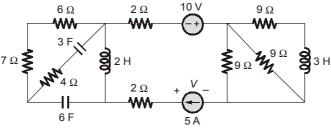
In figure



$$L_{\text{eq}} = L_1 + L_2 + L_3 - 2M_{12} + 2M_{23} - 2M_{13}$$
  
= 4 + 5 + 6 - 2(2) + 2(2) - 2(3) = 9 H  
$$C = 1 \text{ F}$$

$$T = 2\pi\sqrt{9} = 6\pi \sec \theta$$

**Q.5** For the circuit shown below, the steady state voltage 'V' across the current source is \_\_\_\_\_?

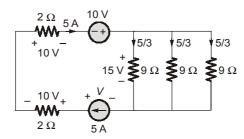


- (a) 35 V
- (c) 15 V

- (b) 25 V
- (d) none of these

5. (b)

The circuit can be redrawn by short circuiting inductor and open circuiting capacitor as DC sources are used.



Applying KVL

$$V - 10 - 10 + 10 - 15 = 0$$

$$V = 25 \text{ V}$$

## Match List-I with List-II and select the correct answer using the codes given below the lists: **Q.6**

# **A.** S = 10 kVA, pf = 0.5 (lagging)

**1**. 
$$(5+j5\sqrt{3})$$
kVA

List-II

**B.** 
$$S = 13 \text{ kVA}, Q = 5 \text{ VAR (capacitative)}$$

**2.** 
$$(5 - j5\sqrt{3})$$
kVA

**C.** 
$$P = 5$$
 kW,  $pf = 0.5$  (inductive)

## Codes:

for A: 
$$S = 10 \text{ kVA}; \cos \phi = \text{pf} = 0.5$$

$$s = 10[\cos\phi + j\sin\phi] \text{ kVA} = (5 + j5\sqrt{3}) \text{kVA}$$

for B: 
$$S = 13 \text{ kVA}$$
;  $Q = 5 \text{ VAR}$ 

$$\sin \phi = \frac{Q}{S} = \frac{5}{13} \implies \cos \phi = \frac{12}{13}$$

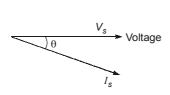
$$s = 13\left(\frac{12}{13} - j\frac{5}{13}\right) \text{kVA} = (12 - j5) \text{ kVA}$$
  
 $P = 5 \text{ kW}; \quad \cos\phi = \text{pf} = 0.5$ 

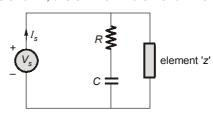
for C: 
$$P = 5 \text{ kW}$$
;  $\cos \phi = \text{pf} = 0.5$ 

$$S = \frac{P}{pf} = 10 \text{ kVA}; \quad \sin\phi = \frac{\sqrt{3}}{2}$$

$$S = 10\left(\frac{1}{2} + j\frac{\sqrt{3}}{2}\right) \text{kVA} = \left(5 + j5\sqrt{3}\right) \text{kVA}$$

For the following phasor diagram of the circuit as shown, the unknown element 'z' is Q.7





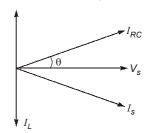
(a) R

(b) C

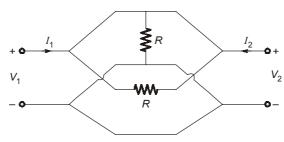
(c) L

(d) combination of RC

- 7. (c)
  - Since only *L* can result in lagging of current with respect to voltage.



**Q.8** The Y-parameter matrix of the circuit shown below is



(a)  $\begin{bmatrix} 2R & 2R \\ 2R & 2R \end{bmatrix}$ 

(c)  $\begin{bmatrix} \frac{1}{2R} & \frac{1}{2R} \\ \frac{1}{2R} & \frac{1}{2R} \end{bmatrix}$ 

(d) doesn't exist

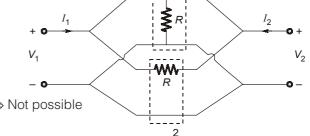
8. (d)

$$[Y] = [Y_1] + [Y_2]$$
$$[Z_1] = \begin{bmatrix} R & R \\ R & R \end{bmatrix}$$

Since determinant is zero,

therefore 'Y' parameter doesn't exist.

[Y] = 
$$[Z]^{-1} = \frac{1}{0} \begin{bmatrix} R & R \\ R & R \end{bmatrix} \Rightarrow \text{Not possible}$$



- as  $[Y_1]$  doesn't exist, [Y] also doesn't exist.
- The unit of  $\frac{R}{2}\sqrt{\frac{L}{C}}$  is Q.9
  - (a)  $\Omega$

(b)  $\Omega^2$ 

(c) no unit

(d)  $\Omega$ H rad

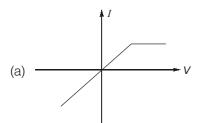
9. (b)

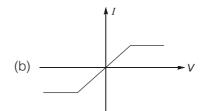
$$\frac{R}{2}\sqrt{\frac{L}{C}} \Rightarrow \frac{R}{2}\sqrt{\frac{L\omega}{C\omega}} = \frac{R}{2}\sqrt{X_L X_C}$$

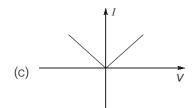
Unit of  $R_1$  is  $\Omega$ Unit of ' $X_L$ ' and ' $X_C$ ' is  $\Omega$ 

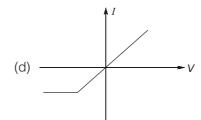
Unit of 
$$\frac{R}{2}\sqrt{\frac{L}{C}}$$
 is  $\Omega \times \sqrt{\Omega \times \Omega} \Rightarrow (\Omega)^2$ 

Q.10 Which of the following graphs, represents an ACTIVE element.









10. (c)

## Numerical Answer Type Questions: Q. 11 to Q. 16 carry 1 mark each

- **Q.11** A parallel R, L, C circuit has R=2 k $\Omega$ , L=240 mH, C=54  $\mu$ F. The quality factor of the circuit at resonance is \_\_\_\_\_.
- 11. (30)

For parallel resonant circuit

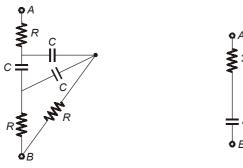
$$Q_0 = R\sqrt{\frac{C}{L}}$$

$$Q_0 = 2000\sqrt{\frac{54 \times 10^{-6}}{240 \times 10^{-3}}}$$

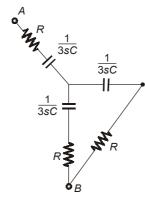
$$Q_0 = 2000\sqrt{\frac{9}{4} \times 10^{-4}}$$

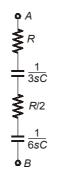
$$Q_0 = \frac{2000}{100} \times \frac{3}{2}$$

Q.12 If both the circuits shown below are equivalent, then the value of RC is \_\_\_\_\_ sec.



12. (4)



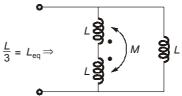




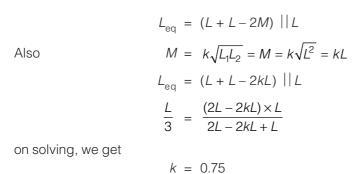
compairing with given diagram

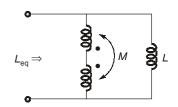
$$R_{\rm eq} = 3\,\Omega = \frac{3R}{2} \ \ {\rm and} \ \ C_{\rm eq} = 4\,{\rm F} = 2\,{\rm C}$$
 so, 
$$R = 2\,\Omega$$
 
$$C = 2\,{\rm F}$$
 
$$RC = 4\,{\rm sec}$$

**Q.13** If  $L_{\text{eq}}$  of the circuit shown below is  $\left(\frac{L}{3}\right)$  H, then the coupling coefficient (k) is \_\_\_\_\_\_.

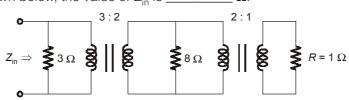


13. 0.75 (0.70 to 0.80)



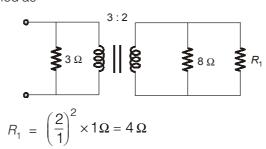


**Q.14** For the circuit shown below, the value of  $Z_{in}$  is \_\_\_\_\_



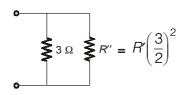
#### 14. (2)

The circuit can be modified as



$$R_1 = \left(\frac{2}{1}\right)^2 \times 1\Omega = 4\Omega$$

$$R' = 4 | 8 = \frac{4 \times 8}{12} = \frac{8}{3} \Omega$$



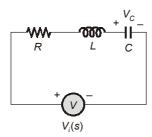
$$R'' = \frac{8}{3} \times \frac{9}{4} = \frac{72}{12} = 6 \Omega$$

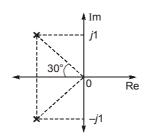
$$Z_{\text{in}} = R'' || 3 \Omega = 6 \Omega || 3 \Omega$$
$$Z_{\text{in}} = 2 \Omega$$

$$Z_{\rm in} = 2 \, \Omega$$

Q.15 For a series RLC circuit shown below, the poles of the transfer function  $\left(\frac{V_C(s)}{V_i(s)}\right)$  are plotted as shown. If

 $\frac{R}{C} = 2\sqrt{3}$ , then the value of *R* is \_\_\_\_\_ ohms.

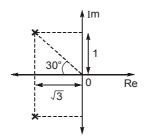




15. 1.73 (1.60 to 1.80)

$$\frac{V_C(s)}{V_i(s)} = \frac{\frac{1}{sC}}{sL + R + \frac{1}{sC}} = \frac{\frac{1}{sC}}{\frac{s^2LC + sRC + 1}{sC}} = \frac{\frac{1}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

The characteristic equation :  $s^2 + \frac{R}{L}s + \frac{1}{LC}$  ...(i) as per graph



$$C(s) = (s + \sqrt{3} + j1)(s + \sqrt{3} - j1)$$

$$C(s) = s^2 + 2\sqrt{3}s + 3 + 1$$
  
=  $s^2 + 2\sqrt{3}s + 4$  ...(ii)

On comparison

$$\frac{R}{L} = 2\sqrt{3} \qquad \dots(iii)$$

$$\frac{1}{LC} = 4 \qquad \dots (iv)$$

$$\frac{R}{C} = 2\sqrt{3}$$
 given ...(v)

from (ii) and (iii)

$$\frac{\frac{R}{L}}{\frac{1}{LC}} = RC = \frac{\sqrt{3}}{2}$$

$$\frac{R}{C} = 2\sqrt{3}$$

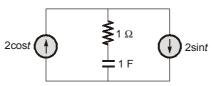
multiplying

$$R^2 = \frac{\sqrt{3}}{2} \times 2\sqrt{3} = 3$$

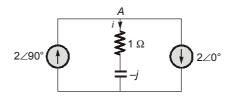
$$R = \sqrt{3}$$

$$R = 1.732 \Omega$$

**Q.16** In the circuit shown below, the average power dissipated by 1  $\Omega$  resistor is \_\_\_\_\_\_ Watts.



16. (4)



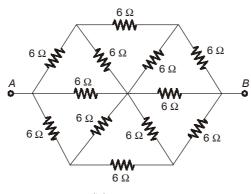
## Applying KCL at node A

$$i = 2\angle 90^{\circ} - 2\angle 0^{\circ}$$
  
 $i = 2j - 2 = -2 + 2j$   
 $i = 2\cos t - 2\sin t$   
 $p(t) = i^{2}R$   
 $p(t) = (2\cos t - 2\sin t)^{2} \times 1$   
 $p(t) = 4\cos^{2}t + 4\sin^{2}t - 8\sin t\cos t$   
 $p(t) = 4 - 4\sin 2t$   
 $p_{avg} = 4$  Watts

## Multiple Choice Questions: Q.17 to Q.26 carry 2 marks each

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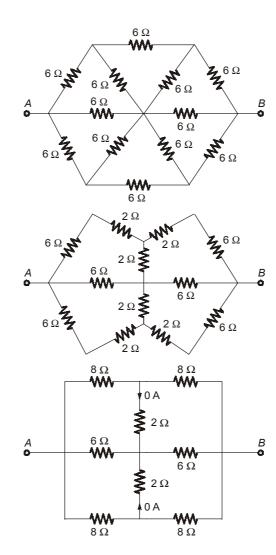
**Q.17** The equivalent resistance between the terminals *A* and *B* is



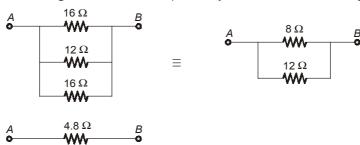
- (a)  $1.2 \Omega$
- (c)  $4.8 \Omega$

- (b)  $3.6 \Omega$
- (d)  $6 \Omega$

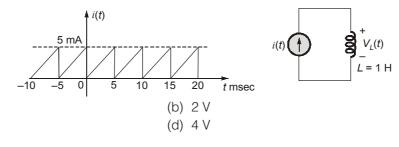
17. (c)



Since two Wheatstone bridges are connected parallelly, no current flows through 2  $\Omega$  resistors.



**Q.18** A periodic sawtooth current wave is passed through an inductor of 1 H as shown below. The value of  $V_L(t)$  at t=3 msec is



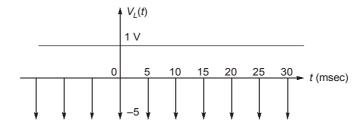
18. (a)

(a) 1 V

(c) 3 V

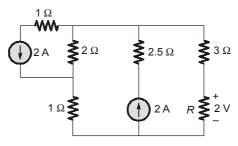
$$V_L(t) = \frac{Ldi(t)}{dt} = \frac{di(t)}{dt}$$
  $\therefore L = 1 \vdash$ 

Differentiating i(t), we get



 $V_L(t)$  at t = 3 msec = 1 V

## **Q.19** For the circuit shown below, the value of R (in ohms) is

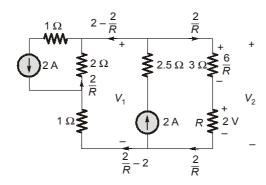


- (a) 2
- (c) 6

 $\Rightarrow$ 

- (b) 4
- (d) none of these

#### 19. (d)



$$V_1 = -\frac{4}{R} - 1\left(\frac{2}{R} - 2\right)$$

$$V_1 = -\left(\frac{4}{R} + \frac{2}{R} - 2\right) = -\left(\frac{6}{R} - 2\right) = 2 - \frac{6}{R}$$

$$V_2 = \frac{6}{R} + 2$$

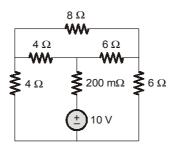
 $V_2 = \frac{6}{R} + 2$   $V_1 = V_2$  (: for parallel branches voltage is same)

$$2 - \frac{6}{R} = \frac{6}{R} + 2$$

$$\frac{12}{R} = 0$$

$$R = \infty$$
 (open circuit)

**Q.20** The power absorbed by 200 m $\Omega$  resistance in the given circuit is

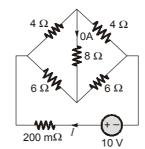


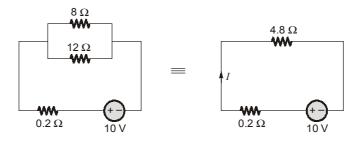
- (a) 0.2 Watts
- (c) 0.6 Watts

(b) 0.4 Watts(d) 0.8 Watts

20. (d)

The circuit can be redrawn as

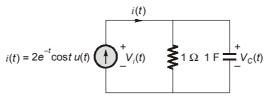




$$I = \frac{10}{5} = 2 \text{ A}$$

$$P_{\text{absorbed}} = (0.2)(2)^2$$
  
= 0.2 \times 4 = 0.8 Watts

**Q.21** For the circuit shown below, if  $i(t) = 2e^{-t}\cos t \, u(t)$  A, then the voltage across the capacitor  $V_C(t)$  will be



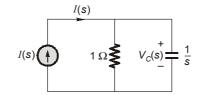
(a)  $2e^{-t}\cos t \, u(t) \, V$ 

(b)  $e^{-t} \cos t u(t)$ 

(c)  $e^{-t} \sin t u(t) V$ 

(d)  $2e^{-t} \sin t \, u(t) \, V$ 

21. (d)



$$V_C(s) = I(s) \times \frac{1}{1 + \frac{1}{s}} \times \frac{1}{s} = I(s) \times \frac{1}{s+1}$$

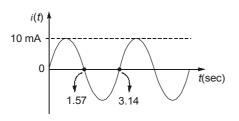
$$I(s) = \frac{2(s+1)}{(s+1)^2+1}$$

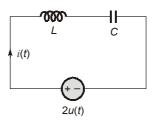
$$V_C(s) = \frac{2(s+1)}{(s+1)^2+1} \times \frac{1}{1+s} = \frac{2}{(s+1)^2+1}$$

$$V_C(t) = 2e^{-t} \sin t \, u(t) \, V$$

**GTOEC17** 

**Q.22** The current flowing through a series *LC* circuit excited by a step voltage is sinusoidal as shown below. Find the values of *L* and *C* 





- (a) L = 10 H, C = 25 mF
- (b) L = 100 H, C = 2.5 mF
- (c) L = 100 H, C = 25 mF
- (d) L = 10 H, C = 2.5 mF

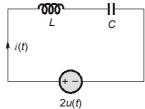
22. (b)

from the given diagram

$$i(t) = 10\sin 2t \, u(t) \, \text{mA}$$



...(iii)



 $\Rightarrow$ 

$$I(s) = \frac{\frac{2}{s}}{sL + \frac{1}{sC}}$$

$$I(s) = \frac{\frac{2}{s}}{\frac{s^2LC+1}{sC}} = \frac{2C}{s^2LC+1}$$

$$I(s) = \frac{\frac{2}{L}}{s^2 + \frac{1}{LC}} \qquad \dots (ii)$$

From equation (i)

$$I(s) = 10 \times 10^{-3} \left( \frac{2}{s^2 + 4} \right)$$

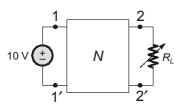
*:*.

$$\frac{1}{LC} = 4$$
 and  $\frac{2}{100} = \frac{2}{L}$   
 $L = 100 \text{ H}$   
 $C = 2.5 \text{ mF}$ 



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## Q.23 Consider the network shown below.



If the hybrid parameter matrix of the Network 'N' is  $[h] = \begin{bmatrix} 3 & 5 \\ 1 & 2 \end{bmatrix}$ , then the maximum power that can be transferred to  $R_L$  is \_\_\_\_\_.

(a) 8.33 W

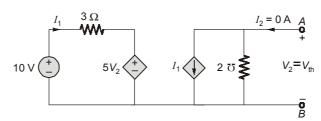
(b) 16.66 W

(c) 25 W

(d) 50 W

### 23. (a)

to determine  $V_{\rm th}$ :



$$I_1 = \frac{10 - 5V_2}{3} = \frac{10 - 5V_{th}}{3}$$

$$V_{\text{th}} = -\frac{I_1}{2} = \frac{5V_{th} - 10}{6}$$

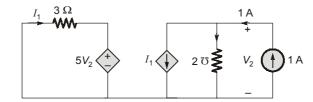
$$6V_{\text{th}} = 5V_{\text{th}} - 10$$
  
 $V_{\text{th}} = -10 \text{ V}$ 

$$V_{th} = -10 \text{ V}$$

to determine  $R_{th}$ :

$$I_2 = 2V_2 + I_1$$
  
1 A =  $2V_2 + I_1$   
0 =  $3I_1 + 5V_2$ 

$$I_1 = -\frac{5}{3}V_2$$



$$1 A = 2V_2 - \frac{5}{3}V_2$$

$$V_2 = 3 \text{ V}$$

$$R_{\rm th} = \frac{V_2}{1 \, \text{A}} = 3 \, \Omega$$

$$P_{L \text{ max}} = \frac{V_{th}^2}{4R_{th}} = \frac{100}{12} \text{ W} = 8.33 \text{ W}$$

GTOEC17

Q.24 If the maximum energy stored in a capacitor shown below is finite, then the possible current source charging the capacitor is



- (a) Impulse
- (c) Unit step

- (d) Parabolic

### 24. (a)

Maximum energy stored in a capacitor is given as

$$E_{\text{max}} = \frac{1}{2}CV_{\text{max}}^2$$

option (a) Impulse current

$$i(t) = \frac{Cdv}{dt}$$

$$V_C(t) = \frac{1}{C} \int_{-\infty}^{t} i(t) dt$$

$$i(t) = \delta(t)$$

$$V_C(t) = \frac{1}{C} u(t)$$

$$V_{\text{max}} = \frac{1}{C}$$

$$E_{\text{max}} = \frac{1}{2} C \left(\frac{1}{C}\right)^2 = \frac{1}{2C}$$

option (b) Ramp

$$V_C(t) = \frac{1}{C} \int \text{Ramp} = \text{parabolic}$$
  
 $V_{\text{max}} = \infty$ 

option (c) unit step

$$V_{C}(t) = \frac{1}{C} \int_{-\infty}^{t} u(t) dt = \frac{1}{C} r(t) = \frac{1}{C} t u(t)$$

$$V_{\text{max}} = \infty$$

$$V_{\rm max} = \infty$$

option (d) parabolic

$$V_{\rm max} = \infty$$

Thus option (a) is the correct choice.

**GTOEC17** 

**Q.25** If  $V_{\text{rms}} = 120 \angle 60^{\circ}$  and Z = 60 - j80, then the complex power 'S' is

(a) (90 - i120) VA

(b) (86.4 + j115.2) VA

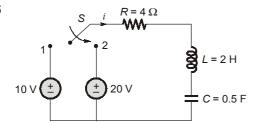
(c) (86.4 – j115.2) VA

(d) (90 + j120) VA

25. (c)

Complex power 'S' = 
$$\frac{|V_{\text{rms}}|^2}{Z^*} = \frac{|120|^2}{60 + j80} = \frac{14400 \times (60 - j80)}{100 \times 100} = 1.44(60 - j80)$$
  
=  $(86.4 - j115.2) \text{ VA}$ 

Q.26



If the switch is at position '1' for a long time and at (t = 0) it is moved to position '2', then the current i(t) for t > 0 will be

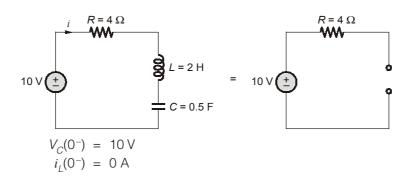
(a)  $(5e^{-t} + 25e^t)$  A

(b)  $(5 + 25te^{-t})$  A

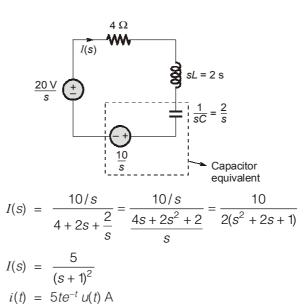
(c)  $(5te^{-t})$  A

(d)  $(25te^{-t} + 5) A$ 

26. (c) At  $(t = 0^{-})$ 



At  $(t = 0^+)$ 

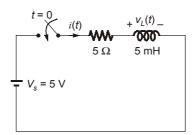


## Numerical Answer Type Questions: Q.27 to Q.33 carry 2 marks each

**Q.27** For the circuit shown below, the current i(t) and the voltage  $v_i(t)$  are given as respectively

$$i(t) = A(1 - e^{-Bt}) u(t) A$$
  
$$v_L(t) = Ce^{-Bt} u(t) V$$

The value of expression  $\frac{AB}{C}$  is \_\_\_\_\_.



27. (200)

For a series RL circuit with DC excitation,

$$i(t) = \frac{V_s}{R} \left( 1 - e^{-\frac{Rt}{L}} \right) u(t) A$$

$$v(t) = V_s \left(e^{-\frac{Rt}{L}}\right) u(t) A$$

$$A = \frac{V_s}{R}$$

$$B = \frac{R}{L}$$
$$C = V_s$$

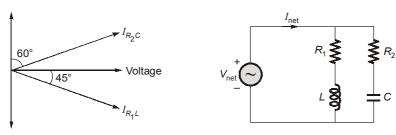
$$C = V_s$$

$$\frac{AB}{C} = \frac{\frac{V_s}{R} \times \frac{R}{L}}{V_s} = \frac{1}{L}$$

$$= \frac{1}{5 \times 10^{-3}} = 200$$

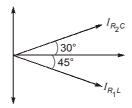
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## Q.28 Consider the circuit shown below.



If L = C, then the minimum value of  $R_1 + R_2$  is \_\_\_\_\_  $\Omega$ 

## 28. 2.63 (2.50 to 2.80)



From phasor, we can write

$$\tan 30^{\circ} = \frac{X_C}{R_2}$$

$$\Rightarrow R_2 = X_C \sqrt{3} = \frac{\sqrt{3}}{\omega C}$$

$$\tan 45^{\circ} = \frac{X_L}{R_1}$$

$$\Rightarrow R_1 = X_C = \omega L$$

$$R_1 R_2 = \frac{\sqrt{3}}{\omega C} \times \omega L = \frac{L}{C} \sqrt{3}$$

$$R_1 R_2 = \sqrt{3} = 1.732$$

we know

$$\frac{R_1 + R_2}{2} \geq \sqrt{R_1 R_2}$$

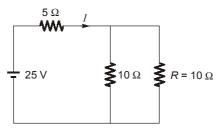
as arithmetic mean  $\geq$  geometric mean; (for non-negative real numbers)

$$R_1 + R_2 \ge 2\sqrt{\sqrt{3}}$$

$$R_1 + R_2 \ge 2(3)^{1/4}$$

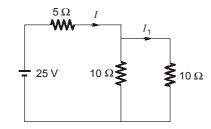
Minimum value of  $R_1 + R_2 = 2.63 \Omega$ 

**Q.29** For the circuit shown below, if *R* is increased by 10%, then the change in current *I* will be \_\_\_\_\_ mA.

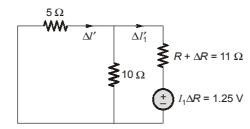


29. -58.14 (-58.50 to -57.50)

$$I_{\text{initial}} = 2.5 \text{ A}$$
  
 $I_{1} = 1.25 \text{ A}$ 



Using compansation theorem

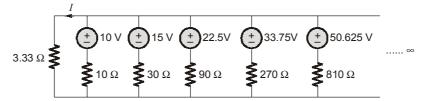


$$\Delta R = 1 \Omega$$

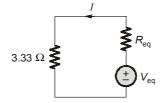
$$-\Delta I_1' = \frac{1.25}{(10||5)+11} = \frac{1.25}{14.33} = 0.08721 \,\text{A}$$

$$\Delta I' = -\Delta I_1' \times \frac{10}{15} = \frac{-2}{3} \Delta I_1' = -0.05814 \text{ A} = -58.14 \text{ mA}$$

**Q.30** For the circuit shown below, the current *I* is \_\_\_\_\_ Amp.



## 30. 1.33 (1.30 to 1.40)



Applying Millman's Theorem

$$\frac{1}{R_{eq}} = \frac{1}{10} + \frac{1}{30} + \frac{1}{90} + \dots$$

$$\frac{1}{R_{eq}} = \frac{1}{10} \left( 1 + \frac{1}{3} + \frac{1}{9} + \dots \right)$$

$$\frac{1}{R_{eq}} = \frac{1}{10} \left( \frac{1}{1 - \frac{1}{3}} \right) = \frac{3}{10 \times 2} = \frac{3}{20}$$

$$R_{eq} = \frac{20}{3} \Omega$$

$$V_{eq} = \frac{\frac{E_1}{R_1} + \frac{E_2}{R_2} + \frac{E_3}{R_3} + \dots}{\frac{1}{R_{eq}}}$$

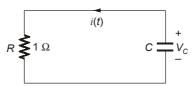
$$= \frac{1 + \frac{1}{2} + \frac{1}{4} + \frac{1}{8} + \dots}{\frac{3}{20}} = \frac{\frac{1}{1 - \frac{1}{2}}}{\frac{3}{20}} = \frac{20}{3} \times 2$$

$$V_{eq} = \frac{40}{3} \text{ V}$$

$$I = \frac{\frac{40}{3}}{\frac{20}{3} + \frac{10}{3}} = \frac{\frac{40}{3}}{\frac{30}{3}} = \frac{4}{3} A$$

$$I = 1.33 \, A$$

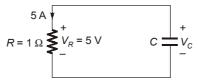
## Q.31 For the circuit shown below,



If  $i(0^-) = 5$  A and  $i(t)|_{t=3 \text{ msec}} = 2.5$  A, then the value of C is \_\_\_\_\_ mF.

#### 31. 4.328 (4.20 to 4.50)

at  $t = 0^-$ :



$$\Rightarrow$$

$$V_R - V_C = 0$$
  
 $V_R = V_C = 5 \text{ V}$   
 $V_C(0^-) = 5 \text{ V} = V_C(0^+)$ 

thus Taking Laplace domain

$$I(s) = \frac{\frac{5}{s}}{1 + \frac{1}{sC}}$$

$$I(s) = \frac{\frac{5}{s}}{\frac{sC+1}{sC}}$$

$$I(s) = \frac{5C}{1+sC} = \frac{5}{s+\frac{1}{C}}$$

$$i(t) = 5e^{-t/C} u(t)$$
  
 $i(t)|_{3 \text{ msec}} = 2.5$ 

$$2.5 = 5e^{-\frac{3 \times 10^{-3}}{C}}$$

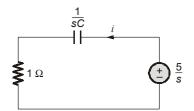
$$\frac{1}{2} = e^{-\frac{3 \times 10^{-3}}{C}}$$

$$-ln2 = \frac{-3 \times 10^{-3}}{C}$$

$$C = \frac{3 \times 10^{-3}}{ln2}$$

$$C = \frac{3}{ln2} \, \text{mF}$$

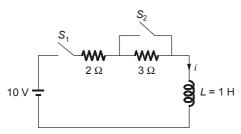
$$C = 4.328 \, \text{mF}$$





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**Q.32** For the circuit shown below, the switch  $S_1$  and  $S_2$  are open for a long time. At t = 0,  $S_1$  is closed, while  $S_2$ is open. At  $t = 2 \sec$ ,  $S_2$  is also closed.



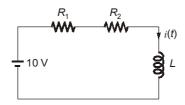
The value of current i at t = 3 sec is \_\_ (Take appropriate approximations).

#### 32. 4.594 (4.40 to 4.70)

At  $(t = 0^-)$ , both the switches are opened.

L is initially uncharged  $i_1(0^-) = 0$ 

At  $(t = 0^+)$ 



$$i(t) = i(\infty) + (i(0^{+}) - i(\infty)) e^{-\frac{R_{eq} t}{L_{eq}}}$$

$$R_{eq} = 5 \Omega$$

$$L_{eq} = 1 H$$

$$i(0^{+}) = 0 A$$

$$i(0^+) = 0 A$$

$$i(t) = 2 + (0-2)e^{-\frac{5t}{1}} A$$
; for  $t > 0$ 

At  $(t = 2^{-})$ 

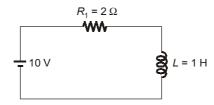
$$i(2^{-}) = 2 - 2e^{-\frac{10}{1}} A$$

$$i(2^{-}) \approx 2 A$$

At  $(t = 2^+)$ 

$$i(2^{-}) = i(2^{+}) = 2 A$$

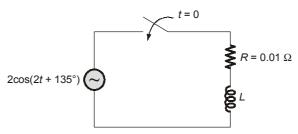
for t > 2 sec



$$i(t) = i(\infty) + (i(2^{+}) - i(\infty)) e^{-\frac{1}{L}}$$
; for  $t > 2$   
 $i(t)|_{t=3s} = 5 + (2-5)e^{-\frac{2}{1}(3-2)} = 5 - 3e^{-2}$   
 $= 4.594 \text{ A}$ 

**GTOEC17** 

**Q.33** If the circuit shown below has zero transient response, then the value of *L* is \_\_\_\_\_ mH.



33. (5)

$$V = 2\cos(2t + 135^\circ)$$

In AC transients, for transient free response

$$\theta = \tan^{-1}\left(\frac{\omega L}{R}\right) + \frac{\pi}{2}$$

$$135^{\circ} = \tan^{-1}\left(\frac{\omega L}{R}\right) + 90^{\circ}$$

$$\tan(45^{\circ}) = \frac{\omega L}{R}$$

$$\omega L = R$$

$$R = 2L$$

$$\frac{R}{2} = L = \frac{0.01}{2}$$

$$L = 5 \text{ mH}$$

0000