

## RESEARCH ARTICLE

WILEY

# Global adaptive prescribed-time stabilization for high-order nonlinear systems

Changchun Hua<sup>1</sup> | Hao Li<sup>1</sup> | Kuo Li<sup>2</sup>

<sup>1</sup>Institute of Electrical Engineering,  
Yanshan University, Qinhuangdao, China

<sup>2</sup>Institute of Automatic Control and  
Complex Systems, University of  
Duisburg-Essen, Duisburg, Germany

## Correspondence

Changchun Hua, Institute of Electrical  
Engineering, Yanshan University, 066004  
Qinhuangdao, China.  
Email: [cch@ysu.edu.cn](mailto:cch@ysu.edu.cn)

## Funding information

National Natural Science Foundation of  
China, Grant/Award Numbers:  
U20A20187, 618255304; Science Fund for  
Creative Research Groups of Hebei  
Province, Grant/Award Number:  
F2020203013; Science and Technology  
Development Grant of Hebei Province,  
Grant/Award Number: 20311803D; Hebei  
Innovation Capability Improvement Plan  
Project, Grant/Award Number:  
22567619H

## Abstract

In this article, the global adaptive prescribed-time stabilization problem of high-order ( $p$ -normal form) nonlinear systems with parametric uncertainties is studied. Compared with the existing results on time-optimal control, our control algorithm eliminates the residual term caused by unknown parameters for the first time. By introducing the new Lyapunov-like function and time-varying function, a prescribed-time controller with adaptive law is constructed, which ensures that the system state variables converge to the origin rather than a residual term, and the setting time can be arbitrarily specified. Moreover, with the help of the state transformations and the power integrator technique, the difficulties of system structure caused by the high power are overcome. Then, based on the differential equations, it is proved that the boundedness of all system signals. Finally, an example of underdrive system simulation is presented to verify the availabilities of the developed control algorithm.

## KEYWORDS

adaptive control, high-order nonlinear systems, parametric uncertainty, prescribed-time control

## 1 | INTRODUCTION

In many physical systems, such as tactical missile track and autonomous vehicle rendezvous, we usually need to complete tasks without errors within the time specified by the user to fulfill stringent time constraints. As a form of time-optimal control, finite/fixed-time control has been extensively studied during the past decades due to its appealing features, which makes the system stabilization within a time and has better robustness, faster convergence rate.<sup>1-3</sup> However, the estimation on settling time of finite/fixed-time control method rely on some designed parameters or initial conditions.<sup>4-8</sup> In order to achieve system stability in accurate time, a prescribed time control method is proposed in References 9-13. Prescribed-time control can ensure that state variables converge to origin by introducing the time-varying scaling functions. It is different from the traditional time optimal control that the convergence time can be arbitrarily specified as the practical permissible range. In Reference 14, the authors provided a prescribed-time smooth state feedback controller instead of the fractional-power controller to address stabilization problem. The inverse optimality prescribed-time stabilization problem of stochastic nonlinear systems was solved in Reference 15, and the complexity of calculation was reduced by proposing a nonscaling backstepping design method. According to the time-varying scaling high gain technique and a novel

temporal transformation with terminal time terms, the authors studied a class of state-dependent systems with generalized strict-feedback structure in Reference 16. However, all the above-mentioned works do not consider the influence of unknown parameters on the system.

As is well known uncertain factors such as unknown parameters exist widely in practical systems, which can cause instability or poor performance of systems. Recently, many excellent control methods for parameter uncertainty system have been reported, such as fuzzy/neural network control, dynamic gain method, adaptive control,<sup>17–21</sup> and so forth. Among those algorithms, adaptive control is a basic method to address system unknown parameters. An adaptive time-varying output feedback control algorithm was put forward in Reference 22 for  $p$ -normal uncertain nonlinear systems, where the nonlinear function has unknown growth rate. In Reference 23, by combining the sign function with adaptive control algorithm, the authors designed globally finite-time controller for  $p$ -normal nonlinear systems with external disturbance and unknown control coefficient. The fixed-time output tracking problem was investigated in Reference 24 for nonlinear systems with parameter uncertainty and output asymmetric time-varying constraints. However, these algorithms only ensured that state variables approach to neighborhood of origin within a finite time interval. The essential obstacle is that the uncertain parameters of systems, which may result in some residual terms in control design and analysis. In References 25 and 26, global finite-time stabilization problems were investigated, a two-phase analytical approaches was presented to deal with unknown parameters, which rendered that the system state converged to origin within a finite time, their global stability analysis was divided into two cases based on two sets related to the unknown parameters and initial conditions. Moreover, because of the existence of adaptive parameters, it is difficult to construct the whole Lyapunov function  $V(t)$  so that its derivative satisfies  $\dot{V} \leq -aV^\alpha - bV^\beta$  to make sure that the state converges to zero, where  $a, b > 0, \alpha > 1, 0 < \beta < 1$ . When the above two-phase analytical method of References 25 and 26 is extended to the fixed-time control, the settling-time still relies on the initial conditions. To our best knowledge, the adaptive fixed-time algorithms can only achieve bounded stability or local stabilization. Therefore, how to provide a unified control algorithm to eliminate the residual term and assign the settling-time in advance is a valuable research issue.

High-order ( $p$ -normal) nonlinear systems have received extensive attention because it has a more general construction and potential practical value. Due to the Jacobian linearization is uncontrollable and unobservable, studying high-order nonlinear uncertain system is more challenging than the widely studied strict feedback nonlinear systems. Motivated by the aforementioned survey, this article aims to propose a global adaptive prescribed-time control algorithm for nonlinear systems with  $p$ -normal form and unknown parameter. The main contributions of this article can be summarized as follows.

1. The global adaptive prescribed-time control problem of nonlinear systems with  $p$ -normal form is investigated for the first time. Different from the results of adaptive finite/fixed-time control, the settling time is irrelevant with the initial value and the system state can converge globally to zero.
2. A new coordinate transformation combining time-varying function is proposed, which is critical for the state converges to origin within the specified time. Moreover, different from the finite/fixed-time stability theory based on fractional Lyapunov differential inequality, which generates residual term when using adaptive technology. We design a novel Lyapunov-like function to handle this trouble, the global adaptive prescribed-time stabilization problem with zero error is thoroughly solved.

**Notations:** In this article,  $\mathbb{R}$ ,  $\mathbb{R}^+$ , and  $\mathbb{R}^n$  represent the set of real number, nonnegative real number, and the  $n$ -dimensional real spaces, respectively. We denote  $[\eta]^\alpha = |\eta|^\alpha \text{sign}(\eta)$ ; for any constant  $\alpha \geq 1$ ,  $[\eta]^\alpha$  is a continuously one-order differentiable function, and  $\frac{d[\eta]^\alpha}{dt} = \alpha|\eta|^{\alpha-1}$ , where  $|\eta|^0 = 1$  if  $\eta = 0$ .  $\mathbb{R}_{\text{odd}}^{\geq 1} := \{q | q = \frac{q_1}{q_2} \text{ with } q_1 \text{ and } q_2 \text{ are odd positive integers, and } q_1 \geq q_2\}$ ;  $\bar{\eta}_i = [\eta_1, \eta_2, \dots, \eta_i]^T \in \mathbb{R}^i$ .

## 2 | PROBLEM FORMULATION AND PREPARATION

In this article, we study the following high-order uncertain nonlinear system

$$\begin{cases} \dot{\eta}_i = g_i(t, \bar{\eta}_i) \eta_{i+1}^{p_i} + f_i(t, \theta_i, \bar{\eta}_i), & i = 1, 2, \dots, n-1, \\ \dot{\eta}_n = g_n(t, \bar{\eta}_n) u^{p_n} + f_n(t, \theta_n, \bar{\eta}_n), \end{cases} \quad (1)$$

where  $\eta_i \in \mathbb{R}$  and  $u \in \mathbb{R}$  are state variable and control input of the system, respectively;  $p_i \in \mathbb{R}_{\text{odd}}^{\geq 1}$  is the order of system;  $\theta_i$  is an unknown constant;  $g_i(t, \bar{\eta}_i): \mathbb{R}^+ \times \mathbb{R}^i \rightarrow \mathbb{R}$  is an unknown control gain and  $f_i(t, \theta_i, \bar{\eta}_i): \mathbb{R}^+ \times \mathbb{R} \times \mathbb{R}^i \rightarrow \mathbb{R}$  is the uncertain continuous nonlinear function.

**Objective:** The control objective of this article is to design a prescribed time controller with adaptive law, so that the state variables converge to the origin within any time  $T$ , and all signals of the system are bounded. In order to achieve the control objective, the following assumptions and lemmas are proposed, which are critical to controller design.

**Assumption 1.** For  $i = 1, \dots, n$ , the sign of  $g_i(t, \bar{\eta}_i)$  is unchanged, without loss of generality, supposing that it is strict positive, and there are known positive constants  $\bar{g}_i$  and  $\underline{g}_i$  such that  $\underline{g}_i \leq |g_i(\cdot)| \leq \bar{g}_i$ .

**Assumption 2.** For  $i = 1, \dots, n$ , there exists a known nonnegative  $C^1$  function  $\bar{\varphi}_i(t, \bar{\eta}_i)$  such that

$$|f_i(t, \theta_i, \bar{\eta}_i)| \leq \bar{\theta} \bar{\varphi}_i(t, \bar{\eta}_i) (|\eta_1| + |\eta_2| + \dots + |\eta_n|), \quad (2)$$

where  $\bar{\theta}$  is an unknown positive constant.

**Remark 1.** Assumption 1 is universal in the stabilization problem of nonlinear systems, which represents global controllability and can be found in Reference 26. Assumption 2 shows that the system has an unknown nonlinear growth conditions, which can be found in References 15 and 23.

**Lemma 1** (27). Let  $\kappa > 0$ , for any  $\eta_i \in \mathbb{R}$  with  $i = 1, \dots, n$ , we have  $(|\eta_1| + |\eta_2| + \dots + |\eta_n|)^\kappa \leq a (|\eta_1|^\kappa + |\eta_2|^\kappa + \dots + |\eta_n|^\kappa)$ , where  $a = n^{\kappa-1}$  if  $\kappa \geq 1$  and  $a = 1$  if  $0 < \kappa < 1$ .

**Lemma 2** (28). If  $a_1 > 0$  and  $0 < a_2 \leq 1$ , then for  $\forall x, y \in \mathbb{R}$ , one has  $|[x]^{a_1 a_2} - [y]^{a_1 a_2}| \leq 2^{1-a_2} |[x]^{a_1} - [y]^{a_1}|^{a_2}$ .

**Lemma 3** (29). Let  $a_1$  and  $a_2$  be positive real numbers, there is a positive function  $\lambda(x, y)$  such that the inequality  $\delta |x|^{a_1} |y|^{a_2} \leq \lambda |x|^{a_1+a_2} + \frac{a_2}{a_1+a_2} \left( \frac{a_1}{(a_1+a_2)\lambda} \right)^{\frac{a_1}{a_2}} \delta^{\frac{a_1+a_2}{a_2}} |y|^{a_1+a_2}$  holds for  $\forall x, y \in \mathbb{R}$ ,  $\delta > 0$ .

### 3 | MAIN RESULTS

#### 3.1 | Controller design

The adaptive prescribed-time controller of nonlinear systems (1) for any given settling time will be designed. We first introduce the following state transformations

$$x_i = L(t)^{-m_i} \eta_i, \quad (3)$$

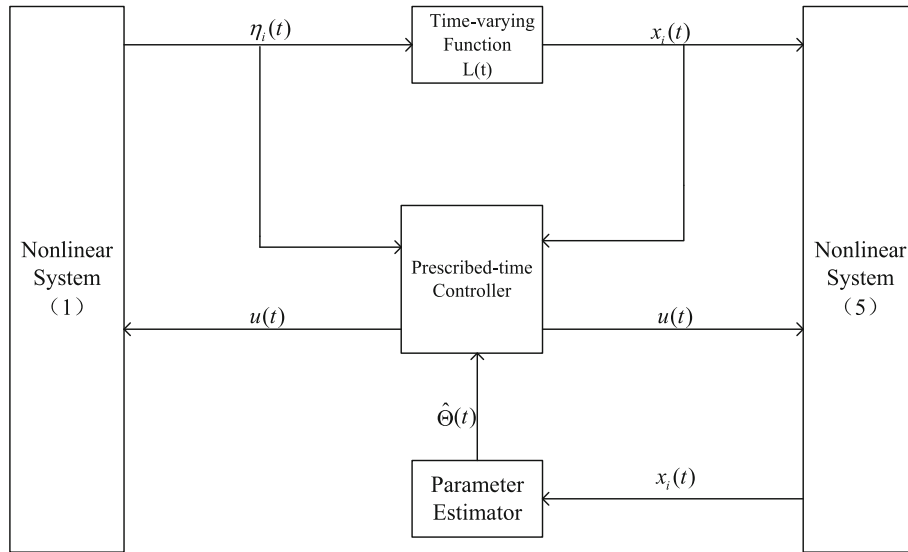
where  $m_n = 1$ ,  $m_i p_{i-1} = m_{i-1} - 1$ , and  $L(t)$  is a positive continuously differentiable bounded function, it has the following characteristics

$$\lim_{t \rightarrow T} L(t) = 0, \dot{L}(t) = c(t) \leq 0, \quad (4)$$

where  $c(t)$  is a bounded function. From (1) and (3), we have

$$\begin{cases} \dot{x}_i = -m_i \frac{\dot{L}(t)}{L(t)} x_i + \frac{g_i x_i^{p_i}}{L(t)} + L(t)^{-m_i} f_i(t, \theta_i, \bar{\eta}_i), \\ \dot{x}_n = -m_n \frac{\dot{L}(t)}{L(t)} x_n + \frac{g_n u^{p_n}}{L(t)} + L(t)^{-m_n} f_n(t, \theta_n, \bar{\eta}_n). \end{cases} \quad (5)$$

**Remark 2.** According to Figure 1, the original prescribed time stabilization problems of  $\eta_i$  are converted to the control problems of  $x_i$  by introducing time-varying function  $L(t)$ .  $L(t)$  as part of the control gain, plays an important role in ensuring that all of state variables converge to origin within prescribed time due to the characteristics of  $\lim_{t \rightarrow T} L(t) = 0$  and  $\dot{L}(t) \leq 0$ . From state transformation (5), it is easy to find that if  $x_i$  is bounded on  $t \in [0, T]$ , one has  $\lim_{t \rightarrow T} \eta_i(t) = 0$ . Moreover, parameter  $m_i$ , function  $L(t)$  and variable  $x_i$  determine the convergence rate of  $\eta_i$ , in particular, the system performances are determined by  $L(t)$ . Therefore, we need to choose appropriate time-varying functions to ensure the system



**FIGURE 1** Block diagram of control strategy

transient steady-state performance. Actually,  $L(t)$  can be chosen widely in this premise of its properties, such as designing function as  $L(t) = (T - t)^a, \log_b(T - t + 1)$ , where  $a \geq 1, b > 1$  are constants.

We define a positive and continuously differentiable function  $W_i(t, \bar{x}_i)$  as

$$W_i(t, \bar{x}_i) = \int_{x_i^*}^{x_i} \left[ |s|^{\frac{\sigma}{r_i}} - |x_i^*|^{\frac{\sigma}{r_i}} \right]^{\frac{2\sigma - r_i}{\sigma}} ds, \quad (6)$$

where  $i = 1, 2, \dots, n; \sigma \geq 1; r_1 = 1, r_i = p_i r_{i+1}$ . It is clear that  $W_i$  is a positive and continuously differentiable function, the virtual control laws  $x_1^*, x_2^*, \dots, x_{n+1}^*$  are defined as

$$\begin{aligned} x_1^* &= 0, & \xi_1 &= [x_1]^{\frac{\sigma}{r_1}} - [x_1^*]^{\frac{\sigma}{r_1}}, \\ x_2^* &= -\beta_1^{\frac{1}{p_1}} [\xi_1]^{\frac{r_2}{\sigma}}, & \xi_2 &= [x_2]^{\frac{\sigma}{r_2}} - [x_2^*]^{\frac{\sigma}{r_2}}, \\ &\vdots & &\vdots \\ x_{n+1}^* &= -\beta_n^{\frac{1}{p_n}} [\xi_n]^{\frac{r_{n+1}}{\sigma}}, & \xi_{n+1} &= [u]^{\frac{\sigma}{r_{n+1}}} - [x_{n+1}^*]^{\frac{\sigma}{r_{n+1}}}, \end{aligned} \quad (7)$$

where  $\beta_i(t, \bar{x}_i, \hat{\Theta})$  is a positive  $C^1$  function. Define an unknown positive constant in the form of

$$\Theta = \max \left\{ \bar{\theta}^{-2}, \bar{\theta}^{\frac{4\sigma}{2\sigma - r_i}}, \bar{\theta}^{\frac{4\sigma}{r_i}} \mid 2 \leq i \leq n \right\}. \quad (8)$$

Let  $\hat{\Theta}$  is the estimate of  $\Theta$ ,  $\tilde{\Theta} = \Theta - \hat{\Theta}$  is the estimate error and the adaptive updating law as

$$\dot{\hat{\Theta}} = \Gamma_n(\bar{x}_n, \hat{\Theta}) \quad (9)$$

with

$$\Gamma_i = \Gamma_{i-1} + \phi_i(t, \bar{x}_i, \hat{\Theta}) \xi_i^2, \quad i = 1, 2, \dots, n, \quad (10)$$

where  $\hat{\Theta}(0) \geq 0, \Gamma_0 = 0, \phi_i(t, \bar{x}_i, \hat{\Theta})$  is a positive  $C^1$  function. We introduce some useful propositions, the proof of which are given in the Appendix.

**Proposition 1.** For  $i = 2, \dots, n$ , there is a positive constant  $\varphi_{i1}$  such that

$$\frac{[\xi_{i-1}]^{\frac{2\sigma-r_{i-1}}{\sigma}} g_{i-1} (x_i^{p_{i-1}} - x_i^{*p_{i-1}})}{L} \leq \frac{1}{3L} \xi_{i-1}^2 + \frac{\varphi_{i1}}{L} \xi_i^2. \quad (11)$$

**Proposition 2.** For  $i = 2, \dots, n$ , there are positive  $C^1$  functions  $\phi_{i1}(t, \bar{x}_i, \hat{\Theta})$  and  $\varphi_{i2}(t, \bar{x}_i, \hat{\Theta})$  such that

$$\begin{aligned} \frac{\partial W_i}{\partial x_i} L^{-m_i} f_i - m_i \frac{\dot{L}}{L} [\xi_i]^{\frac{2\sigma-r_i}{\sigma}} x_i &\leq \frac{1}{4} \sum_{k=1}^{i-1} \xi_k^2 + \frac{1}{3L} \xi_{i-1}^2 + \frac{1}{2L} \xi_i^2 \\ &+ \frac{\varphi_{i2}(t, \bar{x}_i, \hat{\Theta})}{L} \xi_i^2 + L\Theta \phi_{i1}(t, \bar{x}_i, \hat{\Theta}) \xi_i^2. \end{aligned} \quad (12)$$

**Proposition 3.** For  $i = 2, \dots, n$ , there are positive  $C^1$  functions  $k_{i1}(t, \bar{x}_i, \hat{\Theta})$ ,  $\varphi_{i3}(t, \bar{x}_i, \hat{\Theta})$  and  $\phi_{i2}(t, \bar{x}_i, \hat{\Theta})$  such that

$$\frac{\partial W_i}{\partial t} + \sum_{k=1}^{i-1} \frac{\partial W_i}{\partial x_k} \dot{x}_k \leq \sum_{k=1}^{i-1} \left( \frac{1}{2} \xi_k^2 + \frac{1}{3L} \xi_k^2 \right) + \frac{\xi_i^2}{2L} + \left( k_{i1} + \frac{\varphi_{i3}}{L} \right) \xi_i^2 + L\Theta \phi_{i2}(t, \bar{x}_i, \hat{\Theta}) \xi_i^2, \quad (13)$$

where  $\frac{\partial W_i}{\partial t}$  represents the time derivative of  $W_i$  with respect to the explicit variable  $t$ .

**Proposition 4.** For  $i = 2, \dots, n$ , there is a positive  $C^1$  function  $k_{i2}(t, \bar{x}_i, \hat{\Theta})$  such that

$$\sum_{k=1}^{i-1} \frac{\partial W_k}{\partial \hat{\Theta}} \phi_i \xi_i^2 + \frac{\partial W_i}{\partial \hat{\Theta}} \Gamma_i \leq \sum_{k=1}^{i-1} \frac{1}{4} \xi_k^2 + k_{i2}(t, \bar{x}_i, \hat{\Theta}) \xi_i^2. \quad (14)$$

Next, we will give specific controller design steps by inductive method.

**Step 1:** Choose the Lyapunov function as

$$V_1 = W_1 + V_\theta, \quad (15)$$

where  $V_\theta = \frac{L(t)\tilde{\Theta}^2}{2}$ . From (5), (6), and (15), the differential of  $V_1$  satisfies

$$\begin{aligned} \dot{V}_1 &\leq \frac{\dot{L}}{L} V_\theta - L\tilde{\Theta}\dot{\hat{\Theta}} + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} \left( -m_1 \frac{\dot{L}}{L} x_1 + \frac{g_1 x_2^{p_1}}{L} + L^{-m_1} f_1(t, \theta_1, \eta_1) \right) \\ &\leq \frac{-m_1 \dot{L}}{L} \xi_1^2 + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} g_1 \frac{x_2^{p_1} - x_2^{*p_1}}{L} + \bar{\theta} \bar{\varphi}_1(t, \eta_1) \xi_1^2 \\ &\quad + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} \frac{g_1 x_2^{*p_1}}{L} - L\tilde{\Theta}\dot{\hat{\Theta}} + \frac{\dot{L}}{L} V_\theta \\ &\leq \frac{-m_1 \dot{L}}{L} \xi_1^2 + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} g_1 \frac{x_2^{p_1} - x_2^{*p_1}}{L} + L\Theta \phi_1 \xi_1^2 \\ &\quad + \frac{\xi_1^2}{L} + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} \frac{g_1 x_2^{*p_1}}{L} - L\tilde{\Theta}\dot{\hat{\Theta}} + \frac{\dot{L}}{L} V_\theta, \end{aligned} \quad (16)$$

where  $\phi_1(t, x_1) \geq 0.25\bar{\varphi}_1^2$  is a positive function, then, we design the virtual controller as

$$\begin{aligned} x_2^* &= -\beta_1^{\frac{1}{p_1}}(t, \hat{\Theta}, x_1) [\xi_1]^{\frac{r_2}{\sigma}}, \\ \beta_1 &= \underline{g}_1^{-1}(-(m_1 + 2)\dot{L} + (L + 1)(n - 1) + 1 + L^2\hat{\Theta}\phi_1). \end{aligned} \quad (17)$$

Substituting (17) into (16) yields

$$\begin{aligned} \dot{V}_1 &\leq \frac{2\dot{L}}{L} \xi_1^2 - (1 + L^{-1})(n - 1) \xi_1^2 \\ &\quad + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} g_1 \frac{x_2^{p_1} - x_2^{*p_1}}{L} + L\tilde{\Theta}(\Gamma_1 - \dot{\hat{\Theta}}) + \frac{\dot{L}}{L} V_\theta. \end{aligned} \quad (18)$$

**Step 2:** Let  $V_2$  be defined as  $V_2 = V_1 + W_2$ , the derivative of  $V_2$  satisfying

$$\begin{aligned}\dot{V}_2 &= \dot{V}_1 + \frac{\partial W_2}{\partial x_1} \dot{x}_1 + \frac{\partial W_2}{\partial x_2} \dot{x}_2 + \frac{\partial W_2}{\partial t} + \frac{\partial W_2}{\partial \hat{\Theta}} \dot{\hat{\Theta}} \\ &\leq \frac{2\dot{L}}{L} \xi_1^2 - (1 + L^{-1})(n-1) \xi_1^2 \\ &\quad + [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} g_1 \frac{x_2^{p_1} - x_2^{*p_1}}{L} + L\tilde{\Theta}(\Gamma_1 - \dot{\hat{\Theta}}) + \frac{\dot{L}}{L} V_\theta \\ &\quad + [\xi_2]^{\frac{2\sigma-r_2}{\sigma}} \left( -m_2 \frac{\dot{L}}{L} x_2 + \frac{g_2 x_3^{p_2}}{L} + L^{-m_2} f_2(t, \theta_2, \bar{\eta}_2) \right) \\ &\quad + \frac{\partial W_2}{\partial x_1} \dot{x}_1 + \frac{\partial W_2}{\partial t} + \frac{\partial W_2}{\partial \hat{\Theta}} \dot{\hat{\Theta}}.\end{aligned}\quad (19)$$

Based on Propositions 1–4, we have

$$\begin{aligned}&\frac{\xi_1^2}{L} + \xi_1^2 + \frac{\varphi_2 \xi_2^2}{L} + k_2 \xi_2^2 + L\Theta \phi_2 \xi_2^2 \\ &\geq [\xi_1]^{\frac{2\sigma-r_1}{\sigma}} g_1 \frac{x_2^{p_1} - x_2^{*p_1}}{L} + \frac{\partial W_2}{\partial x_1} \dot{x}_1 + \frac{\partial W_2}{\partial t} + \frac{\partial W_2}{\partial \hat{\Theta}} \Gamma_2 \\ &\quad + [\xi_2]^{\frac{2\sigma-r_2}{\sigma}} \left( -m_2 \frac{\dot{L}}{L} x_2 + L^{-m_2} f_2(t, \theta_2, \bar{\eta}_2) \right),\end{aligned}\quad (20)$$

where  $\varphi_2(t, \bar{x}_2, \hat{\Theta}) = \varphi_{21} + \varphi_{22} + \varphi_{23} + 1$ ,  $k_2(t, \bar{x}_2, \hat{\Theta}) = k_{21} + k_{22}$ ,  $\phi_2(t, \bar{x}_2, \hat{\Theta}) = \phi_{21} + \phi_{22}$  are positive functions.

Then, the virtual control law is designed as

$$\begin{aligned}x_3^* &= -\beta_2^{\frac{1}{p_2}}(t, \hat{\Theta}, \bar{x}_2) [\xi_2]^{\frac{r_3}{\sigma}}, \\ \beta_2 &= \frac{1}{g_2} (-2\dot{L} + (L+1)(n-2) + \varphi_2 + Lk_2 + L^2 \hat{\Theta} \phi_2).\end{aligned}\quad (21)$$

Combined with (19)–(21), we have

$$\begin{aligned}\dot{V}_2 &\leq \left( \frac{2\dot{L}}{L} - (1 + L^{-1})(n-2) \right) \sum_{k=1}^2 \xi_k^2 + \frac{\dot{L}}{L} V_\theta \\ &\quad + [\xi_2]^{\frac{2\sigma-r_2}{\sigma}} g_2 \frac{x_3^{p_2} - x_3^{*p_2}}{L} + \left( L\tilde{\Theta} - \frac{\partial W_2}{\partial \hat{\Theta}} \right) (\Gamma_2 - \dot{\hat{\Theta}}).\end{aligned}\quad (22)$$

**Inductive step  $i$  ( $3 \leq i \leq n$ ):** Based on the above results, we have

$$\begin{aligned}\dot{V}_{i-1} &\leq \left( \frac{2\dot{L}}{L} - (1 + L^{-1})(n-i+1) \right) \sum_{k=1}^{i-1} \xi_k^2 + [\xi_{i-1}]^{\frac{2\sigma-r_{i-1}}{\sigma}} g_{i-1} \frac{x_i^{p_{i-1}} - x_i^{*p_{i-1}}}{L} + \frac{\dot{L}}{L} V_\theta \\ &\quad + \left( L\tilde{\Theta} - \sum_{k=1}^{i-1} \frac{\partial W_k}{\partial \hat{\Theta}} \right) (\Gamma_{i-1} - \dot{\hat{\Theta}}).\end{aligned}\quad (23)$$

We define the virtual controller as

$$\begin{aligned}x_{i+1}^* &= -\beta_i(t, \hat{\Theta}, \bar{x}_i)^{\frac{1}{p_i}} [\xi_i]^{\frac{r_{i+1}}{\sigma}}, \\ \beta_i &= \frac{1}{g_i} (-2\dot{L} + (L+1)(n-i) + \varphi_i + Lk_i + L^2 \hat{\Theta} \phi_i),\end{aligned}\quad (24)$$

where  $\varphi_i(t, \bar{x}_i, \hat{\Theta}) = \varphi_{i1} + \varphi_{i2} + \varphi_{i3} + 1$ ,  $k_i(t, \bar{x}_i, \hat{\Theta}) = k_{i1} + k_{i2}$ , and  $\phi_i(t, \bar{x}_i, \hat{\Theta}) = \phi_{i1} + \phi_{i2}$  are positive functions. Define the Lyapunov function as  $V_i = V_{i-1} + W_i$ , based on Propositions 1–4, we can obtain

$$\begin{aligned} \dot{V}_i &\leq \dot{V}_{i-1} + \sum_{k=1}^{i-1} \frac{\partial W_i}{\partial x_k} \dot{x}_k + \frac{\partial W_i}{\partial x_i} \dot{x}_i + \frac{\partial W_i}{\partial t} + \frac{\partial W_i}{\partial \hat{\Theta}} \dot{\hat{\Theta}} \\ &\leq \left( \frac{2\dot{L}}{L} - (1 + L^{-1})(n - i) \right) \sum_{k=1}^i \xi_k^2 + \xi_i^{\frac{2\sigma-r_i}{\sigma}} g_i \frac{x_{i+1}^{p_i} - x_{i+1}^{*p_i}}{L} + \frac{\dot{L}}{L} V_\theta \\ &\quad + \left( L\tilde{\Theta} - \sum_{k=1}^i \frac{\partial W_k}{\partial \hat{\Theta}} \right) (\Gamma_i - \dot{\hat{\Theta}}). \end{aligned} \quad (25)$$

**Step  $n$ :** We choose the Lyapunov function  $V_n = V_{n-1} + W_n$ , using the inductive argument above, we know the differential of  $V_n$  satisfies

$$\begin{aligned} \dot{V}_n &\leq \frac{2\dot{L}}{L} \sum_{k=1}^n \xi_k^2 + \left( L\tilde{\Theta} - \sum_{k=1}^n \frac{\partial W_k}{\partial \hat{\Theta}} \right) (\Gamma_n - \dot{\hat{\Theta}}) \\ &\quad + [\xi_n]^{\frac{2\sigma-r_n}{\sigma}} g_n \frac{u^{p_n} - x_{n+1}^{*p_n}}{L} + \frac{\dot{L}}{L} V_\theta. \end{aligned} \quad (26)$$

The controller with adaptive law is designed as

$$\begin{aligned} u &= -\beta_n^{\frac{1}{p_n}}(t, \hat{\Theta}, \bar{x}_n) [\xi_n]^{\frac{r_{n+1}}{\sigma}}, \\ \dot{\hat{\Theta}} &= \Gamma_n = \sum_{k=1}^n \phi_k \xi_k^2, \end{aligned} \quad (27)$$

where  $\beta_n = \frac{1}{g_n}(-2\dot{L} + \varphi_n + Lk_n + L^2\hat{\Theta}\phi_n)$ ,  $\varphi_n(t, \bar{x}_n, \hat{\Theta}) = \varphi_{n1} + \varphi_{n2} + \varphi_{n3} + 1$ ,  $k_n(t, \bar{x}_n, \hat{\Theta}) = k_{n1} + k_{n2}$  and  $\phi_n(t, \bar{x}_n, \hat{\Theta}) = \phi_{n1} + \phi_{n2}$  are positive functions. Let  $x_{n+1}^* = u$ , then, since  $W_i = \int_{x_i^*}^{x_i} [s]^{\frac{\sigma}{r_i}} - [x_i^*]^{\frac{\sigma}{r_i}}]^{\frac{2\sigma-r_i}{\sigma}} ds \leq 2\xi_i^2$ ,  $\dot{L} \leq 0$ , we have

$$\dot{V}_n \leq \frac{2\dot{L}}{L} \sum_{k=1}^n \xi_k^2 + \frac{\dot{L}}{L} V_\theta \leq \frac{\dot{L}}{L} V_n. \quad (28)$$

### 3.2 | Prescribed-time stability analysis

Now, we give the following theorem to summarize the main results of this article.

**Theorem 1.** For the nonlinear systems (1) with Assumptions 1 and 2, by designing the controller and adaptive law (27), the states of systems can converge to zero within a arbitrarily specified time  $T$  and other signals are bounded.

*Proof.* Based on the definition of  $L(t)$  and  $\dot{V}_n(t) \leq \frac{\dot{L}(t)}{L(t)} V_n(t)$ , we have

$$V_n(t) \leq \frac{L(t)}{L(0)} V_n(0). \quad (29)$$

Because  $V_n(t) = \sum_{i=1}^n \int_{x_i^*}^{x_i} [s]^{\frac{\sigma}{r_i}} - [x_i^*]^{\frac{\sigma}{r_i}}]^{\frac{2\sigma-r_i}{\sigma}} ds + \frac{L(t)\tilde{\Theta}^2}{2}$  and  $\lim_{t \rightarrow T} L(t) = 0$ , we can conclude  $\lim_{t \rightarrow T} \|\bar{\xi}_n\| = 0$  and  $|\tilde{\Theta}| \leq \sqrt{\frac{2V_n(0)}{L(0)}}$ , hence we know  $\beta_i(\cdot)$  is a bounded function and  $\lim_{t \rightarrow T} |x_i^*| = \lim_{t \rightarrow T} |\beta_{i-1}^{\frac{1}{p_{i-1}}} [\xi_{i-1}]^{\frac{r_i}{\sigma}}| = 0$ . From (3) and (7), we can get that  $\lim_{t \rightarrow T} \|\bar{x}_n\| = 0$  and  $\lim_{t \rightarrow T} \|\bar{\eta}_n\| = 0$ . Since  $\lim_{t \rightarrow T} u(t) = \lim_{t \rightarrow T} \dot{\hat{\Theta}}(t) = 0$  and  $\lim_{t \rightarrow T} \|\bar{\eta}_n(t)\| = 0$ , based on the solution continuation properties,  $f_i(\cdot)$  vanish at the origin, when setting  $u(t) = 0$  for all  $t \geq T$ , we can easy to verify that  $\|\bar{\eta}_n(t)\| = 0$ . In other words, the states of system reach to zero within the prescribed time  $T$  and will always



keep in equilibrium point for  $t \in [T, \infty]$ . Then, based on the definition of  $V_n(t)$  and (29), we can directly conclude that the state  $\tilde{\xi}_n$ ,  $\bar{x}_n$  and the estimate error  $\tilde{\Theta}$  are all bounded. Since  $\Theta$  is a constant, we know the estimate  $\hat{\Theta}$  is also bounded. Then, it can be proved that virtual controllers  $x_i^*(t)$ , the control law  $u(t)$  and the adaptive law  $\dot{\hat{\Theta}}(t)$  are bounded. ■

**Remark 3.** In this article, by introducing the time-varying scaling transformations  $x_i = L(t)^{-m_i} \eta_i$ , the original prescribed time stabilization problems are converted to the control problems of transformed system. Then, by constructing a new coordinate transformation combining time-varying function with estimation error, we propose a new Lyapunov-like function  $V_\theta = \frac{L(t)\tilde{\Theta}^2}{2}$ . Essentially,  $\lim_{t \rightarrow T} L(t) = 0$  can ensure  $\lim_{t \rightarrow T} V_\theta = 0$ , then taking  $L(t)\tilde{\Theta}$  as a whole to design adaptive law and controller. In other words, our goal is that the whole Lyapunov function converges at the rate of  $L(t)$ , therefore, we construct function  $V_\theta$ , which also converges in  $L(t)$ . The bound of the controller must be taken into account in prescribed-time control, we can ensure functions  $\beta_i$  only contain the positive power of time-varying function  $L(t)$ . Hence the controller  $u$  and error function  $\tilde{\Theta}$  are bounded. Based on this, a new framework for adaptive prescribed-time control is established, that eliminate the residual term, and ensure that the states converge to the zero within a time that independent of the initial conditions.

**Remark 4.** It should be noted that due to the existence of unknown parameter  $\tilde{\Theta}$ , we cannot introduce  $-\frac{1}{2}\tilde{\Theta}^\alpha$  to the right hand side of the derivative of the whole Lyapunov function. So, the traditional finite/fixed-time methods based on fractional Lyapunov differential inequality, cannot be directly used to stability analysis. References 23-26 made an attempt to propose a two-phase control strategy that the global finite-time stability analysis was divided into two parts because of the residual term. For example, by defining a set  $\Delta$  associated with unknown parameters and initial value, Li et al.<sup>26</sup> get  $\dot{V}_n \leq -\frac{1}{8}\bar{\mu}V_n^{\frac{2\sigma}{2\sigma-1}} - \gamma\bar{\mu}(\frac{1}{4} - \bar{V}_n)\sum_{i=1}^n \xi_i^2$ , where  $\bar{V}_n < \frac{1}{4}$  when state  $x(t) \in \Delta$ . If the initial value  $x(0) \in \Delta$ , the state will converge to origin with the time  $T_1(x(0))$ ; if  $x(0) \notin \Delta$ , the state will reached the boundary of  $\Delta$  within a finite-time  $T_2(x(0))$ , then from the previous discussion, the states globally converge to origin within finite time  $T_1(x(0)) + T_2(x(0))$ . Obviously, the setting time relies on the initial values. For the fixed-time two-phase control strategy, we only can get  $T_1$  is independent of the initial values, the setting-time  $T_1 + T_2(x(0))$  still depends on the initial value. In this article, we proposed a new control strategy that does not generate the residual sets, and the convergence time can be prespecified and irrelevant to the initial value.

## 4 | SIMULATION EXAMPLE

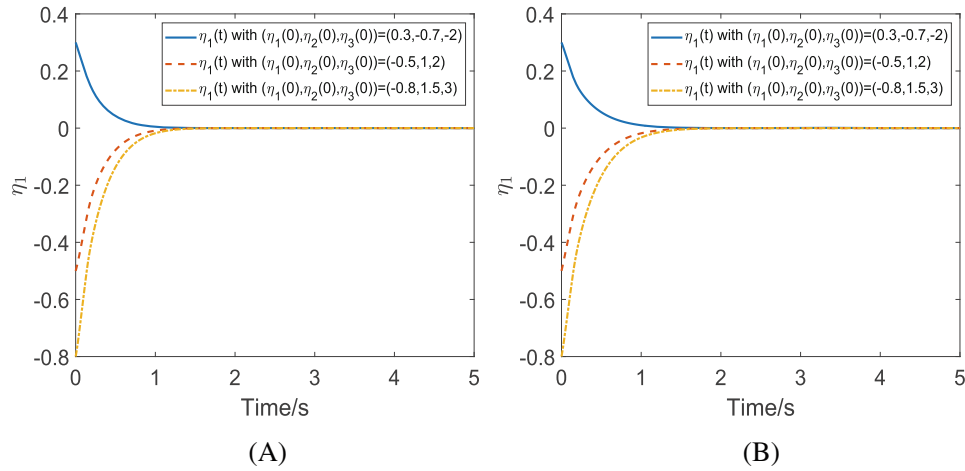
In this article, we consider the following practical model for a class of underactuated systems with weak coupling<sup>26,30,31</sup> to demonstrate the availabilities of our developed control algorithm.

$$\begin{aligned}\ddot{\omega} &= \frac{g}{l} \sin \omega + \frac{k_s}{b_2 l} (x - l \sin \omega)^3 \cos \omega, \\ \ddot{x} &= -\frac{k}{b_1} x - \frac{k_s}{b_1} (x - l \sin \omega)^3 + \frac{\hat{u}}{b_1},\end{aligned}\quad (30)$$

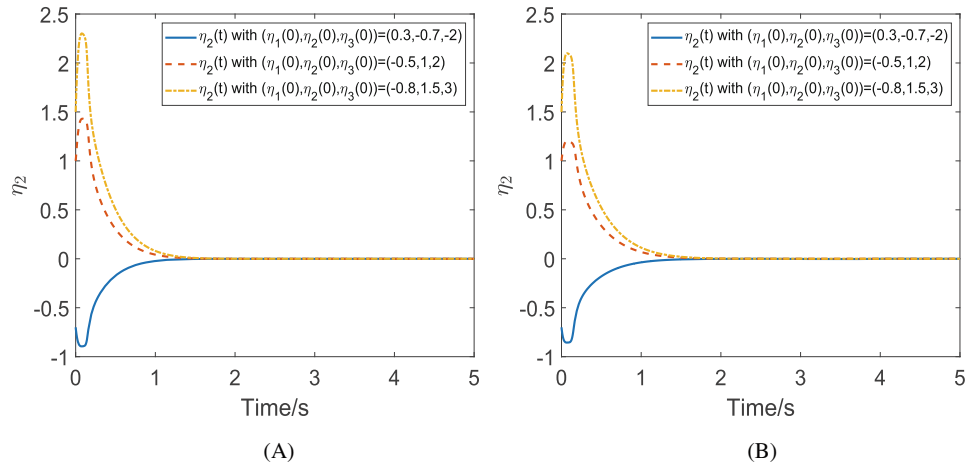
where  $b_1$  and  $b_2$  are the mass of smooth pedestal platform and the mass of inverted pendulum, respectively.  $\omega$  and  $x$  are the angle of the inverted pendulum and the displacement of mass  $b_1$ ;  $k_s, l, g$  are the spring coefficients, length of the bar and gravity coefficient, respectively. Our objective is to stabilize the mechanical system within a user-prescribed time, by defining the smooth change of coordinates  $\eta_1 = \omega, \eta_2 = \dot{\eta}_1, \eta_3 = (x - l \sin \omega)\sqrt[3]{\cos \omega}, \eta_4 = \dot{\eta}_3 = u$  on  $(\omega, \dot{\omega}, x, \dot{x}) \in \left(-\frac{\pi}{2}, \frac{\pi}{2}\right) \times \mathbb{R}^3$ . According to,<sup>26</sup> the stabilization problem of the fourth-order system is also solvable if the following third-order system form<sup>26</sup> is solvable.

$$\begin{aligned}\dot{\eta}_1 &= \eta_2, \\ \dot{\eta}_2 &= \frac{k_s}{b_2 l} \eta_3^3 + \frac{g}{l} \sin \eta_1, \\ \dot{\eta}_3 &= u.\end{aligned}\quad (31)$$

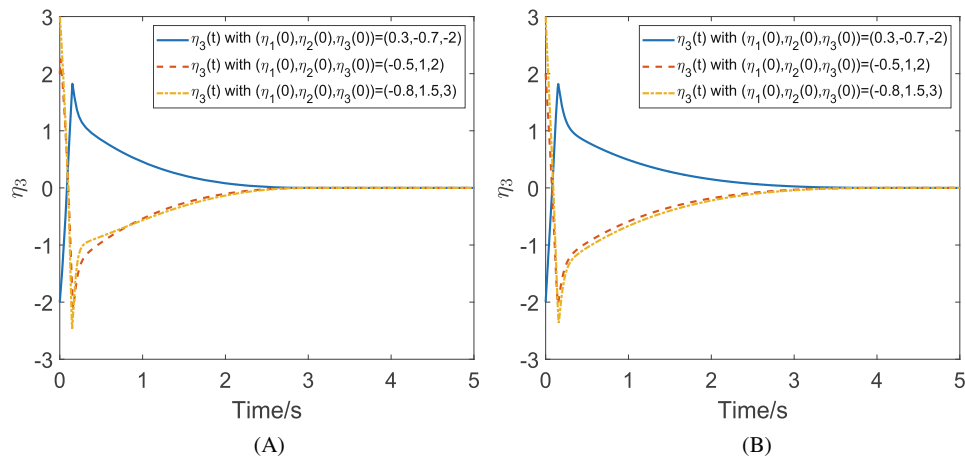




**FIGURE 2** Response of  $\eta_1$  under different initial values. (A) Prescribed-time  $T = 3$  s and (B) prescribed-time  $T = 4$  s



**FIGURE 3** Response of  $\eta_2$  under different initial values. (A) Prescribed-time  $T = 3$  s and (B) prescribed-time  $T = 4$  s



**FIGURE 4** Response of  $\eta_3$  under different initial values. (A) Prescribed-time  $T = 3$  s and (B) prescribed-time  $T = 4$  s

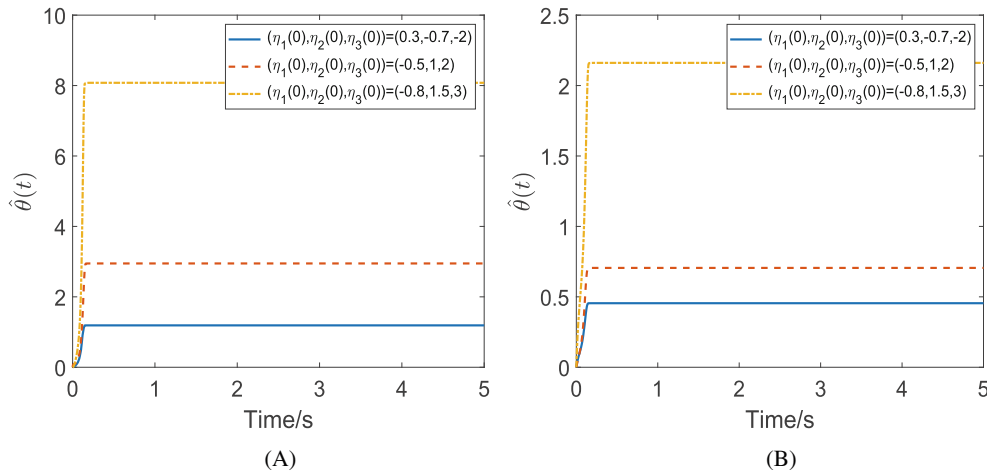


FIGURE 5 Response of  $\hat{\Theta}(t)$  under different initial values. (A) Prescribed-time  $T = 3$  s and (B) prescribed-time  $T = 4$  s

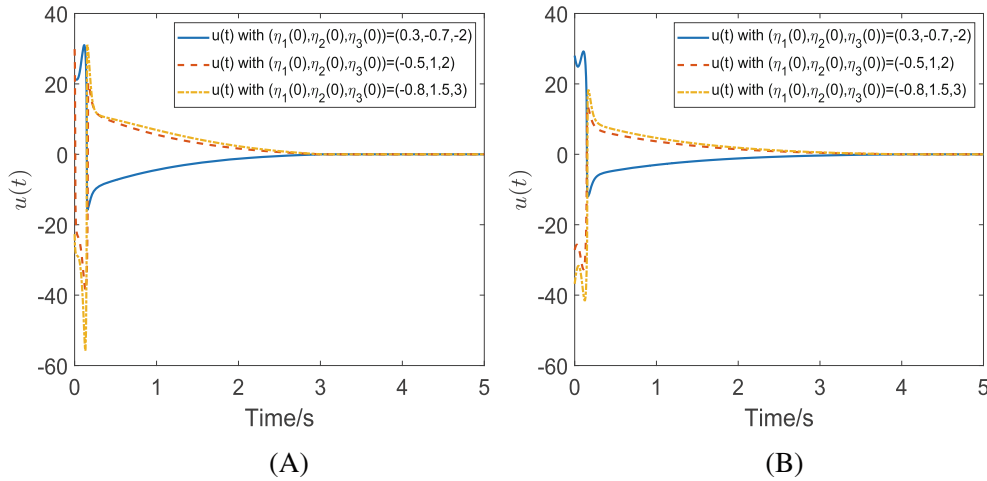


FIGURE 6 Response of  $u(t)$  under different initial values. (A) Prescribed-time  $T = 3$  s and (B) prescribed-time  $T = 4$  s

Based on our proposed algorithm, we can take the design parameters as  $m_1 = 5, m_2 = 4, m_3 = 1, r_1 = 1, r_2 = 1, r_3 = \frac{1}{3}, \sigma = 1, \bar{\theta} = \max \left\{ 1, \frac{g}{l} \right\}$ . Designing time-varying function as  $L(t) = T - t$ . The initial values are chosen as  $[\eta_1(0), \eta_2(0), \eta_3(0), \hat{\Theta}(0)]^T = [0.3, -0.7, -2, 0]^T, [-0.5, 1, 2, 0]^T$ , and  $[-0.8, 1.5, 3, 0]^T$ , respectively. Then, by designing prescribed-time as  $T = 3$  s and  $T = 4$  s, respectively, we can obtain the following simulation results.

The simulation results are showed in Figures 2–6. According to Figures 2–4, it is obvious to get that the state variable converges to zero within  $T = 3$  s and  $T = 4$  s, where the setting time  $T$  of the system can be arbitrarily specified by user, and independent of any design parameters and initial values. From Figures 5 and 6, we can know the estimated values  $\hat{\Theta}$  and controller  $u(t)$  are bounded and  $\lim_{t \rightarrow T} u(t) = 0$  due to  $\lim_{t \rightarrow T} \|\bar{\eta}_n\| = 0$ . According to the simulation results, it can be concluded that our proposed adaptive prescribed-time control algorithm is effective. Moreover, from the part figures A and B, we can find that the control gain actually depends on given convergence time, larger control gain will lead to faster convergence. Theoretically, the convergence time is any prescribed time, since it does not rely on the initial values. However, if the prescribed time  $T$  decreases, the control gain will become large. In practical application, we need to balance the relationship between the convergence time and the control gain to find an optimal situation.

## 5 | CONCLUSION

This article has proposed a novel adaptive prescribed-time algorithm to handle the stabilization problem for a class of uncertainty systems with  $p$ -normal form. By utilizing the time-varying scaling transformation, an adaptive controller has been structured to ensure that system state variables converge to the origin within the specified time and all the signals are bounded. Different from the existing results, our control algorithm does not generate residual sets, hence avoiding classified discussion caused by residual sets. Finally, the availabilities of the developed algorithm are demonstrated by a underactuated simulation example. Further research will focus on the global adaptive prescribed-time tracking problem without residual sets.

## FUNDING INFORMATION

This work was partially supported by National Natural Science Foundation of China (U20A20187, 618255304), Science Fund for Creative Research Groups of Hebei Province (F2020203013), Science and Technology Development Grant of Hebei Province (20311803D), and Hebei Innovation Capability Improvement Plan Project (22567619H).

## CONFLICT OF INTEREST

The authors declare that they have no conflict of interest to this work.

## DATA AVAILABILITY STATEMENT

The data that support the findings of this study are available from the corresponding author upon reasonable request.

## ORCID

Changchun Hua  <https://orcid.org/0000-0001-6311-2112>

Hao Li  <https://orcid.org/0000-0001-8164-3830>

Kuo Li  <https://orcid.org/0000-0002-9049-5558>

## REFERENCES

- Huang X, Lin W, Yang B. Global finite-time stabilization of a class of uncertain nonlinear systems. *Automatica*. 2005;41(5):881-888.
- Huang J, Wen C, Wang W, Song Y-D. Adaptive finite-time consensus control of a group of uncertain nonlinear mechanical systems. *Automatica*. 2015;51:292-301.
- Hong Y, Jiang Z-P. Finite-time stabilization of nonlinear systems with parametric and dynamic uncertainties. *IEEE Trans Automat Contr*. 2006;51(12):1950-1956.
- Li Y, Li K, Tong S. Adaptive neural network finite-time control for multi-input and multi-output nonlinear systems with positive powers of odd rational numbers. *IEEE Trans Neural Netw Learn Syst*. 2020;31(7):2532-2543.
- Sui S, Xu H, Tong S, Chen CLP. Prescribed performance fuzzy adaptive output feedback control for nonlinear mimo systems in a finite-time. *IEEE Trans Fuzzy Syst*. 2022;30(9):3633-3644.
- Hua C, Ning P, Li K, Guan X. Fixed-time prescribed tracking control for stochastic nonlinear systems with unknown measurement sensitivity. *IEEE Trans Cybern*. 2022;52(5):3722-3732.
- Shi S, Xu S, Liu W, Zhang B. Global fixed-time consensus tracking of nonlinear uncertain multiagent systems with high-order dynamics. *IEEE Trans Cybern*. 2020;50(4):1530-1540.
- Liu Y, Liu X, Jing Y. Adaptive practical preassigned finite-time stability for a class of pure-feedback systems with full state constraints. *Int J Robust Nonlinear Control*. 2019;29(10):2978-2994.
- Hua C-C, Ning P, Li K. Adaptive prescribed-time control for a class of uncertain nonlinear systems. *IEEE Trans Automat Contr*. 2021. doi:10.1109/TAC.2021.3130883
- Jimenez-Rodriguez E, Sanchez-Torres JD, Loukianov AG. On optimal predefined-time stabilization. *Int J Robust Nonlinear Control*. 2017;27(17):3620-3642.
- Zhou B. Finite-time stabilization of linear systems by bounded linear time-varying feedback. *Automatica*. 2020;113:108760.
- Ni J, Tang Y, Shi P. A new fixed-time consensus tracking approach for second-order multiagent systems under directed communication topology. *IEEE Trans Syst Man Cybern Syst*. 2021;51(4):2488-2500.
- Holloway J, Krstic M. Prescribed-time output feedback for linear systems in controllable canonical form. *Automatica*. 2019;107:77-85.
- Song Y, Wang Y, Holloway J, Krstic M. Time-varying feedback for finite-time robust regulation of normal-form nonlinear systems. Proceedings of the 2016 IEEE 55th Conference on Decision and Control (CDC); 2016:3837-3842.
- Li W, Krstic M. Stochastic nonlinear prescribed-time stabilization and inverse optimality. *IEEE Trans Automat Contr*. 2022;67(3):1179-1193.

16. Krishnamurthy P, Khorrami F, Krstic M. A dynamic high-gain design for prescribed-time regulation of nonlinear systems. *Automatica*. 2020;115:108860.
17. Wang F, Chen B, Lin C, Zhang J, Meng X. Adaptive neural network finite-time output feedback control of quantized nonlinear systems. *IEEE Trans Cybern*. 2018;48(6):1839-1848.
18. Li Y, Tong S, Li T. Composite adaptive fuzzy output feedback control design for uncertain nonlinear strict-feedback systems with input saturation. *IEEE Trans Cybern*. 2015;45(10):2299-2308.
19. Zhang X, Lin Y. Adaptive control of nonlinear time-delay systems with application to a two-stage chemical reactor. *IEEE Trans Automat Contr*. 2015;60(4):1074-1079.
20. Wu J, Li J, Zong G, Chen W. Global finite-time adaptive stabilization of nonlinearly parametrized systems with multiple unknown control directions. *IEEE Trans Syst Man Cybern Syst*. 2017;47(7):1405-1414.
21. Hua C, Meng R, Li K. Hybrid threshold strategy-based adaptive tracking control for nonlinear stochastic systems with full-state constraints. *Int J Robust Nonlinear Control*. 2020;30(15):6189-6206.
22. Man Y, Liu Y. Global output-feedback stabilization for high-order nonlinear systems with unknown growth rate. *Int J Robust Nonlinear Control*. 2017;27(5):804-829.
23. Min H, Xu S, Gu J, Cui G. Adaptive finite-time control for high-order nonlinear systems with multiple uncertainties and its application. *IEEE Trans Circuits Syst I-Regul Pap*. 2020;67(5):1752-1761.
24. Jin X. Adaptive fixed-time control for mimo nonlinear systems with asymmetric output constraints using universal barrier functions. *IEEE Trans Automat Contr*. 2019;64(7):3046-3053.
25. Sun Z-Y, Shao Y, Chen C-C. Fast finite-time stability and its application in adaptive control of high-order nonlinear system. *Automatica*. 2019;106:339-348.
26. Li T, Yang J, Wen C, Zhang C. Global adaptive finite-time stabilization of uncertain time-varying p-normal nonlinear systems without homogeneous growth nonlinearity restriction. *IEEE Trans Automat Contr*. 2019;64(11):4637-4644.
27. Liu L, Xie X-J. Output-feedback stabilization for stochastic high-order nonlinear systems with time-varying delay. *Automatica*. 2011;47(12):2772-2779.
28. Ding S, Li S, Zheng WX. Nonsmooth stabilization of a class of nonlinear cascaded systems. *Automatica*. 2012;48(10):2597-2606.
29. Li S, Ahn CK, Guo J, Xiang Z. Global output feedback sampled-data stabilization of a class of switched nonlinear systems in the P-normal form. *IEEE Trans Syst Man Cybern Syst*. 2021;51(2):1075-1084.
30. Rui C, Reyhanoglu M, Kolmanovsky I, Cho S, McClamroch N. Nonsmooth stabilization of an underactuated unstable two degrees of freedom mechanical system. *Proceedings of the 36th IEEE Conference on Decision and Control*; vol. 4, 1997:3998-4003.
31. Qian C, Lin W. A continuous feedback approach to global strong stabilization of nonlinear systems. *IEEE Trans Automat Contr*. 2001;46(7):1061-1079.

**How to cite this article:** Hua C, Li H, Li K. Global adaptive prescribed-time stabilization for high-order nonlinear systems. *Int J Robust Nonlinear Control*. 2023;33(3):1669-1682. doi: 10.1002/rnc.6433

## APPENDIX

*Proof of Proposition 1.* Based on Lemmas 2 and 3 and  $p_i = \frac{r_i}{r_{i+1}}$ , we have

$$\begin{aligned}
 & L^{-1} |\xi_{i-1}|^{\frac{2\sigma-r_{i-1}}{\sigma}} g_{i-1} \left| x_i^{p_{i-1}} - x_i^{*p_{i-1}} \right| \\
 & \leq \bar{g}_{i-1} L^{-1} |\xi_{i-1}|^{\frac{2\sigma-r_{i-1}}{\sigma}} \left| [x_i]^{\frac{\sigma}{r_i} \frac{r_{i-1}}{\sigma}} - [x_i^*]^{\frac{\sigma}{r_i} \frac{r_{i-1}}{\sigma}} \right| \\
 & \leq \bar{g}_{i-1} 2^{1-\frac{r_{i-1}}{\sigma}} L^{-1} |\xi_{i-1}|^{\frac{2\sigma-r_{i-1}}{\sigma}} |\xi_i|^{\frac{r_{i-1}}{\sigma}} \\
 & \leq \frac{1}{3L} \xi_{i-1}^2 + \frac{\varphi_{i1}}{L} \xi_i^2,
 \end{aligned} \tag{A1}$$

where  $\varphi_{i1}$  is a positive designed real number. ■

*Proof of Proposition 2.* We know  $m_{i-1} - m_i \geq 1$ , since  $L < L(0)$ , there is a known function  $\bar{\varphi}_{i1}(t, \bar{x}_i)$  such that  $\bar{\varphi}_{i1}(t, \bar{x}_i) \geq \bar{\varphi}_i = \bar{\varphi}_i(t, L^{m_1} x_1, \dots, L^{m_i} x_i)$ , then

$$\frac{\partial W_i}{\partial x_i} L^{-m_i} f_i(t, \theta_i, \bar{x}_i) \leq \bar{\theta} \bar{\varphi}_i(t, \bar{\eta}_i) |\xi_i|^{2-\frac{r_i}{\sigma}} (L^{m_1-m_i} |x_1| + \dots + |x_i|)$$

$$\begin{aligned}
&\leq \max \{L^{m_1-m_i}, L^{m_2-m_i}, \dots, L^0\} \bar{\theta} \tilde{\varphi}_i(t, \bar{x}_i) \\
&\quad \times |\xi_i|^{2-\frac{r_i}{\sigma}} (|\xi_1|^{\frac{r_1}{\sigma}} + \dots + \beta_{i-1}^{\frac{1}{p_{i-1}}} |\xi_{i-1}|^{\frac{r_i}{\sigma}} + |\xi_i|^{\frac{r_i}{\sigma}}) \\
&\leq \frac{1}{4} \sum_{k=1}^{i-1} \xi_k^2 + \frac{\xi_i^2}{2L} + L\Theta \phi_{i1}(t, \bar{x}_i, \hat{\Theta}) \xi_i^2,
\end{aligned} \tag{A2}$$

we know  $\phi_{i1}(t, \bar{x}_i, \hat{\Theta})$  is a known positive  $C^1$  function.

According to (7) and Lemma 3, we can obtain

$$\begin{aligned}
-m_i \frac{\dot{L}}{L} [\xi_i]^{\frac{2\sigma-r_i}{\sigma}} x_i &\leq -m_i \frac{\dot{L}}{L} |\xi_i|^{\frac{2\sigma-r_i}{\sigma}} (|\xi_i|^{\frac{r_i}{\sigma}} + \beta_{i-1}^{\frac{1}{p_{i-1}}} |\xi_{i-1}|^{\frac{r_i}{\sigma}}) \\
&\leq \frac{1}{3L} |\xi_{i-1}|^2 + \frac{\varphi_{i2}(t, \bar{x}_{i-1}, \hat{\Theta})}{L} |\xi_i|^2,
\end{aligned} \tag{A3}$$

where  $\varphi_{i2}(t, \bar{x}_{i-1}, \hat{\Theta})$  is a  $C^1$  function. Combining (A2) and (A3), we can get Proposition 2. ■

*Proof of Proposition 3.* According to (7), there exists a nonnegative function  $\bar{\varphi}(t, \hat{\Theta}, \bar{x}_{i-1})$  such that

$$\begin{aligned}
\frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial t} &= \frac{\partial(-\beta_{i-1}^{\frac{\sigma}{r_{i-1}}} [\xi_{i-1}])}{\partial t} \\
&\leq |\xi_{i-1}| \left| \frac{\partial(\beta_{i-1}^{\frac{\sigma}{r_{i-1}}})}{\partial t} \right| + \beta_{i-1}^{\frac{\sigma}{r_{i-1}}} |\xi_{i-2}| \left| \frac{\partial(\beta_{i-2}^{\frac{\sigma}{r_{i-2}}})}{\partial t} \right| + \dots + \beta_{i-1}^{\frac{\sigma}{r_{i-1}}} \beta_{i-2}^{\frac{\sigma}{r_{i-2}}} \dots \beta_2^{\frac{\sigma}{r_2}} |\xi_1| \left| \frac{\partial(\beta_1^{\frac{\sigma}{r_1}})}{\partial t} \right| \\
&\leq \bar{\varphi}(t, \hat{\Theta}, \bar{x}_{i-1}) \sum_{k=1}^{i-1} |\xi_k|
\end{aligned} \tag{A4}$$

by a simple calculation, we have

$$\begin{aligned}
\frac{\partial W_i}{\partial t} &\leq \left( \frac{2\sigma - r_i}{\sigma} \right) \left| \frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial t} \right| \int_{x_i^*}^{x_i} \left( s^{\frac{\sigma}{r_i}} - x_i^{\frac{\sigma}{r_i}} \right)^{1-\frac{r_i}{\sigma}} ds \\
&\leq \left( \frac{2\sigma - r_i}{\sigma} \right) \bar{\varphi}(t, \hat{\Theta}, \bar{x}_{i-1}) \sum_{k=1}^{i-1} |\xi_k| 2^{1-\frac{r_i}{\sigma}} |\xi_i| \\
&\leq \frac{1}{4} \sum_{k=1}^{i-1} \xi_k^2 + k_{i1} \xi_i^2,
\end{aligned} \tag{A5}$$

where  $k_{i1}(t, \bar{x}_i, \hat{\Theta})$  is a known positive  $C^1$  function. Then, based on mathematical induction, there exists a  $C^1$  function  $\lambda(t, \bar{x}_{i-1}, \hat{\Theta})$  such that

$$\left| \frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial x_j} \right| \leq \left( \sum_{l=1}^{i-1} |\xi_l|^{1-\frac{r_l}{\sigma}} \right) \lambda(t, \bar{x}_{i-1}, \hat{\Theta}). \tag{A6}$$

Then according to (5) and (A2), we have

$$\begin{aligned}
\frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial x_j} \dot{x}_j &\leq \left( \sum_{l=1}^{i-1} |\xi_l|^{1-\frac{r_l}{\sigma}} \right) \tilde{\lambda}(t, \bar{x}_{i-1}, \hat{\Theta}) \\
&\quad \times \left( \frac{\beta_{j-1}^{\frac{1}{p_{j-1}}} |\xi_{j-1}|^{\frac{r_j}{\sigma}} + \left( \beta_j^{\frac{1}{p_j}} + 1 \right) |\xi_j|^{\frac{r_j}{\sigma}}}{L} + \frac{|\xi_{j+1}|^{\frac{r_j}{\sigma}}}{L} + L^{-m_j} f_j(t, \theta_j, \bar{\eta}_j) \right)
\end{aligned}$$

$$\leq \left( L^{-1} \sum_{l=1}^{i-1} \bar{\lambda}_l(t, \bar{x}_{i-1}, \hat{\Theta}) |\xi_l| \right) + \bar{\theta}^{\frac{\sigma}{r_i}} \left( \sum_{l=1}^{i-1} \bar{\varphi}_l(t, \bar{x}_{i-1}, \hat{\Theta}) |\xi_l| \right). \quad (\text{A7})$$

Further, in view of (A6), there are positive  $C^1$  functions  $\varphi_{i3}(t, \bar{x}_i, \hat{\Theta})$ ,  $\phi_{i2}(t, \bar{x}_i, \hat{\Theta})$  such that

$$\begin{aligned} \sum_{k=1}^{i-1} \frac{\partial W_i}{\partial x_k} \dot{x}_k &\leq \left( 2 - \frac{r_i}{\sigma} \right) \sum_{k=1}^{i-1} \left| \frac{\partial(x_i^*)^{\frac{\sigma}{r_i}}}{\partial x_k} \dot{x}_k \right| \int_{x_i^*}^{x_i} \left( s^{\frac{\sigma}{r_i}} - x_i^{\frac{\sigma}{r_i}} \right)^{1-\frac{r_i}{\sigma}} ds \\ &\leq \left( \frac{2\sigma - r_i}{\sigma} \right) 2^{1-\frac{r_i}{\sigma}} |\xi_i| \sum_{k=1}^{i-1} \left| \frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial x_k} \dot{x}_k \right| \\ &\leq \sum_{k=1}^{i-1} \left( \frac{1}{4} \xi_k^2 + \frac{1}{3L} \xi_k^2 \right) + \frac{\varphi_{i3}(t, \bar{x}_i, \hat{\Theta})}{L} \xi_i^2 + \frac{1}{2L} \xi_i^2 + L\Theta \phi_{i2}(t, \bar{x}_i, \hat{\Theta}) \xi_i^2. \end{aligned} \quad (\text{A8})$$

Combining (A5) and (A8), we can get (13). ■

*Proof of Proposition 4.* Based on the proof of (12), we can find a function  $\gamma_i(t, \hat{\Theta}, \bar{x}_{i-1})$  yields

$$\begin{aligned} \frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial \hat{\Theta}} &= \frac{\partial(-\beta_{i-1}^{\frac{r_{i-1}}{r_i}} \xi_{i-1})}{\partial \hat{\Theta}} \\ &\leq |\xi_{i-1}| \left| \frac{\partial(\beta_{i-1}^{\frac{r_{i-1}}{r_i}})}{\partial \hat{\Theta}} \right| + \beta_{i-1}^{\frac{\sigma}{r_{i-1}}} |\xi_{i-2}| \left| \frac{\partial(\beta_{i-2}^{\frac{r_{i-2}}{r_{i-1}}})}{\partial \hat{\Theta}} \right| \\ &\quad + \cdots + \beta_{i-1}^{\frac{\sigma}{r_{i-1}}} \beta_{i-2}^{\frac{\sigma}{r_{i-2}}} \cdots \beta_2^{\frac{\sigma}{r_2}} |\xi_1| \left| \frac{\partial(\beta_1^{\frac{r_1}{r_2}})}{\partial \hat{\Theta}} \right| \\ &= \gamma_i(t, \hat{\Theta}, \bar{x}_{i-1}). \end{aligned} \quad (\text{A9})$$

According to the expression of  $\beta_i$ , we have  $\gamma_i(t, \hat{\Theta}, \bar{x}_{i-1})$  is a  $C^1$  function, then we have

$$\begin{aligned} \sum_{k=1}^{i-1} \frac{\partial W_k}{\partial \hat{\Theta}} \phi_i \xi_i^2 + \frac{\partial W_i}{\partial \hat{\Theta}} \Gamma_i &\leq \sum_{k=1}^{i-1} \frac{\partial W_k}{\partial \hat{\Theta}} \phi_i \xi_i^2 \\ &\quad + \left( \frac{2\sigma - r_i}{\sigma} \right) \left| \frac{\partial(-x_i^*)^{\frac{\sigma}{r_i}}}{\partial \hat{\Theta}} \right| \int_{x_i^*}^{x_i} \left( s^{\frac{\sigma}{r_i}} - x_i^{\frac{\sigma}{r_i}} \right)^{1-\frac{r_i}{\sigma}} ds \Gamma_i \\ &\leq \tilde{\gamma} 2^{1-\frac{r_i}{\sigma}} |\xi_i| \sum_{k=1}^i \phi_k \xi_k^2 + \sum_{k=1}^{i-1} \frac{\partial W_k}{\partial \hat{\Theta}} \phi_i \xi_i^2 \\ &\leq \sum_{k=1}^{i-1} \frac{1}{4} \xi_k^2 + k_{i2}(t, \bar{x}_i, \hat{\Theta}) \xi_i^2, \end{aligned} \quad (\text{A10})$$

where  $k_{i2}(t, \bar{x}_i, \hat{\Theta})$  is a known positive  $C^1$  function. ■