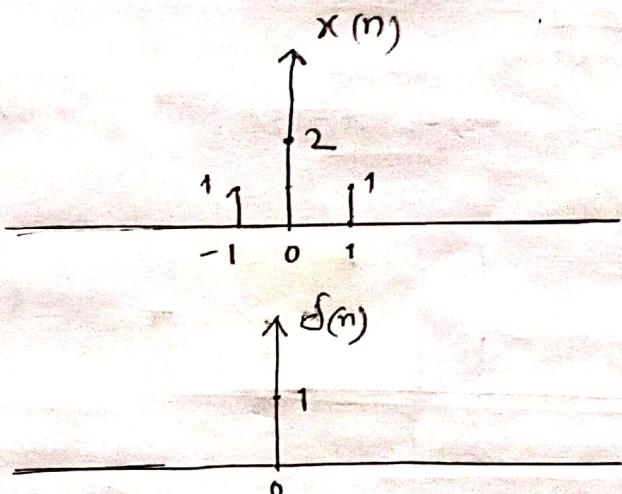


$$\delta(n) = \begin{cases} 1, & n=0 \\ 0, & \text{otherwise} \end{cases}$$

$$\delta(n-1) = \begin{cases} 1, & n=1 \\ 0, & \text{otherwise} \end{cases}$$

$$\delta(n+1) = \begin{cases} 1, & n=-1 \\ 0, & \text{otherwise} \end{cases}$$

$$x(n) = \{ 1, 2, 1 \}$$



$x[k]$ = particular value for a certain point

$$x[0]\delta[n] = 2 \times 1 = 2$$

$$x[1]\delta[n-1] = 1 \times 1 = 1$$

$$x[-1]\delta[n+1] = 1 \times 1 = 1$$

$$x[2]\delta[n+2] = 0 \times 1 = 0$$

$$x[2]\delta[n-2] = 0 \times 1 = 0$$

Adding all of these values will be equal to $x(n)$.

$$\sum_{k=-\infty}^{\infty} x[k]\delta[n-k] = x(n)$$

Any discrete signal can be decomposed with shifted

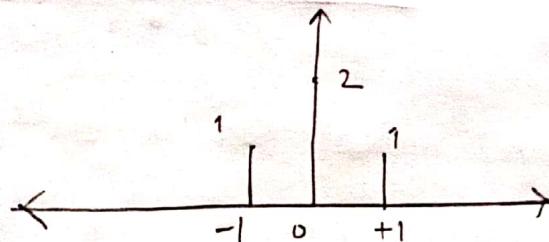
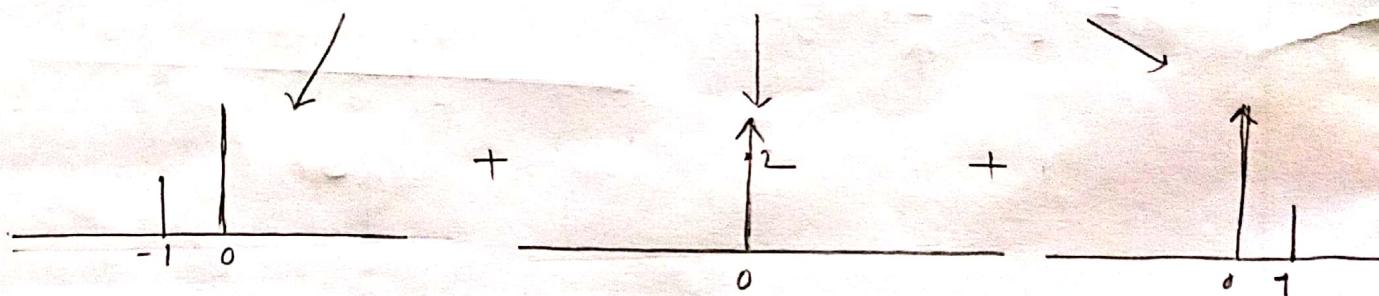
impulse signal.

Every discrete signal is a linear combination of shifted impulse signal.

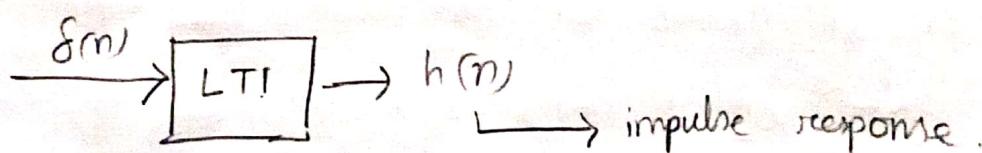
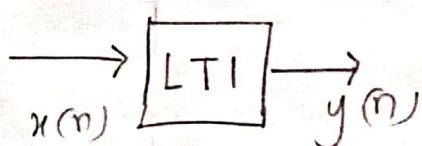
↳ unit sample signal.

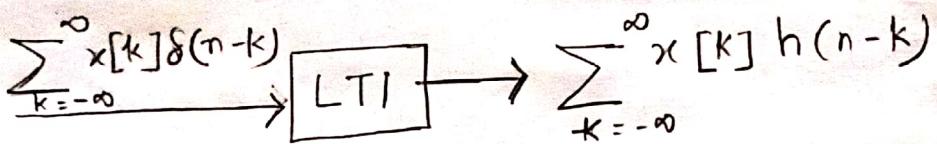
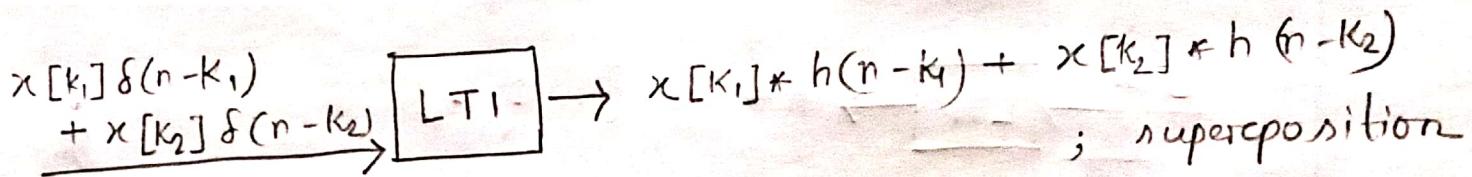
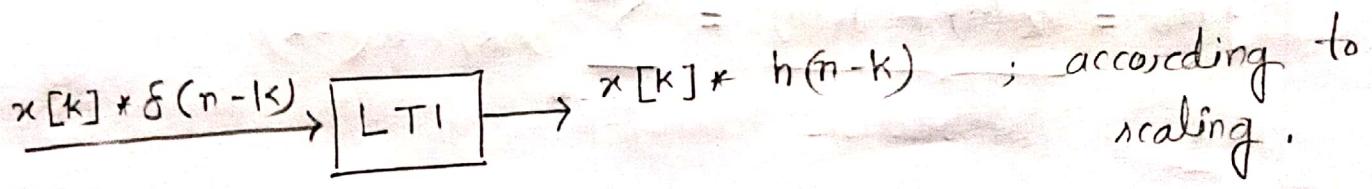
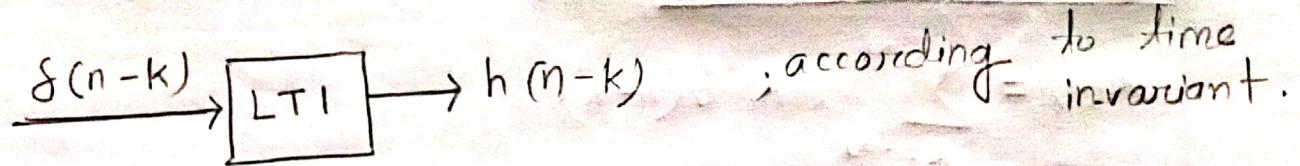
$$\sum_{k=-\infty}^{\infty} x(k) \delta(n-k)$$

$$= \dots + x(-1) \delta(n+1) + x(0) \delta(n) + x(1) \delta(n-1) + \dots$$



⊗ LTI —





$$x(n) \rightarrow \boxed{\text{LTI}} \rightarrow y(n) = \sum_{k=-\infty}^{\infty} x[k] h(n-k)$$

$$= x(n) * \delta(n)$$

→ convolve (?)

This is called convolution of $x(n)$, a discrete time signal.

Given, $x(n) = \{1, 2, 3\}$

$$h(n) = \{2, -1, 0, 2\}$$

$$y(n) = ?$$

Homework.

LTI systems make signal analysis easier.

$$x(n) = \sum_{k=-\infty}^{\infty} x(k) \delta(n-k)$$

$$y(n) = \underbrace{x(n) * h(n)}_{\text{convolution}} = \sum_{k=-\infty}^{\infty} x(k) \cdot h(n-k)$$

$$\delta(n) \xrightarrow{\text{LTI}} h(n) = \{1, 1, 1\}$$

$$\delta(n) = \{1\}$$

$$y(n) = ?$$

Method - 1:

$$x(n) = \left\{ \begin{array}{l} 2 \\ \uparrow \\ 0 \\ 2 \end{array} \right\}$$

$$h(n) = \left\{ \begin{array}{l} 1 \\ \uparrow \\ 1 \\ 1 \end{array} \right\}$$

$$y(n) = x(n) * h(n) = ?$$

$$x(n) = \sum_{k=-\infty}^{\infty} x(k) \delta(n-k) = x[0] \delta(n) + x[1] \delta(n-1) + x[2] \delta(n-2)$$

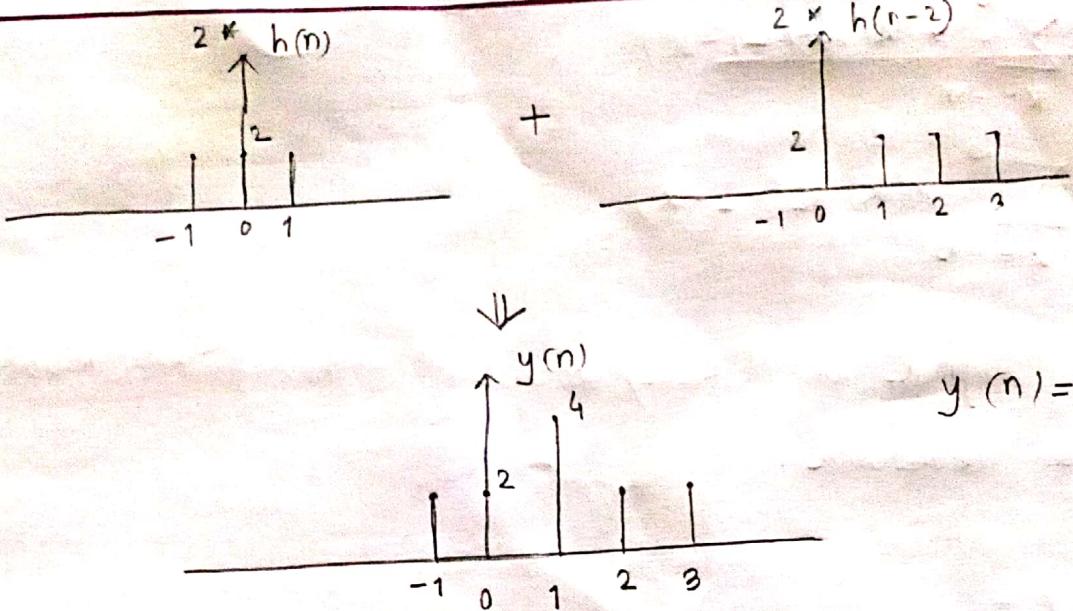
$$= 2 \delta(n) + 0 + 2 \delta(n-2)$$

$$x(n) = 2 \delta(n) + 2 \delta(n-2)$$

$$2 \delta(n) \xrightarrow{\text{LTI}} 2 h(n)$$

$$2 \delta(n-2) \xrightarrow{\text{LTI}} 2 h(n-2)$$

$$y(n) = x(n) * h(n) = 2 h(n) + 2 h(n-2)$$



* If sequence is long, just show 2-4 values around the origin.

Method-2:

$$x(n) = \{2, 0, 2\} \quad ; \quad y(n) = ?$$

$$h(n) = \{1, 1, 1\}$$

$$\begin{aligned} y(n) &= x(n) * h(n) \\ &= \sum_{k=-\infty}^{\infty} x(k) * h(n-k) \end{aligned}$$

$$y(0) = \sum_{k=-\infty}^{\infty} x(k) * h(-k) =$$

$$y(1) = \sum_{k=-\infty}^{\infty} x(k) * h(1-k) =$$

$$y(2) = \sum_{k=-\infty}^{\infty} x(k) * h(2-k) =$$

⋮

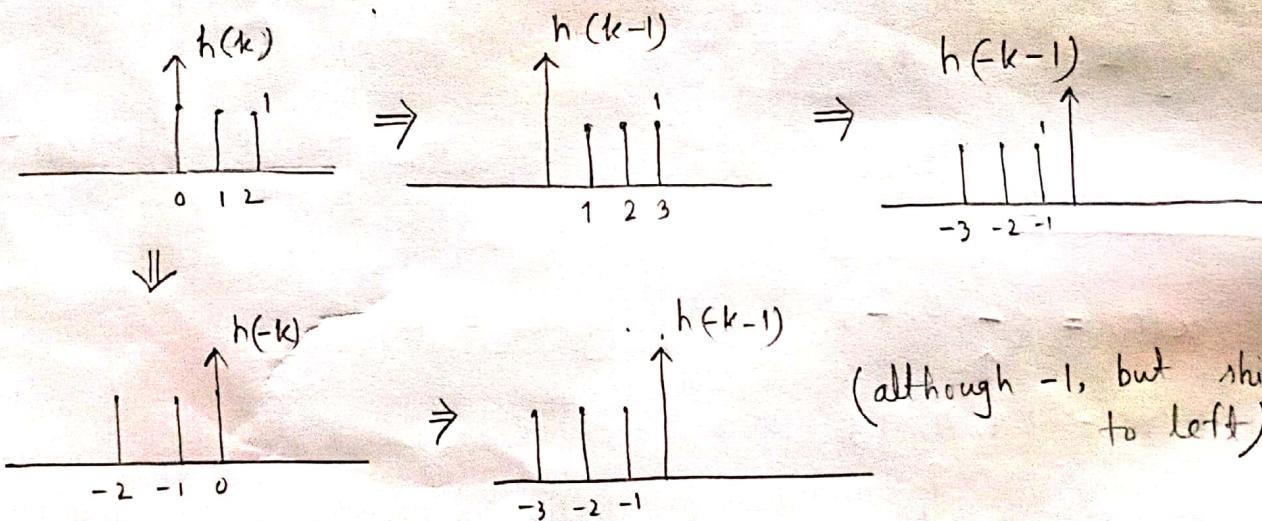
$$y(-1) = \sum_{k=-\infty}^{\infty} x(k) h(-1-k)$$

$$y(-2) = \sum_{k=-\infty}^{\infty} x(k) h(-2-k)$$

If a signal is folded/reversed, the direction of shifting is also reversed.

Example -

$$h(-k-1) = ?$$



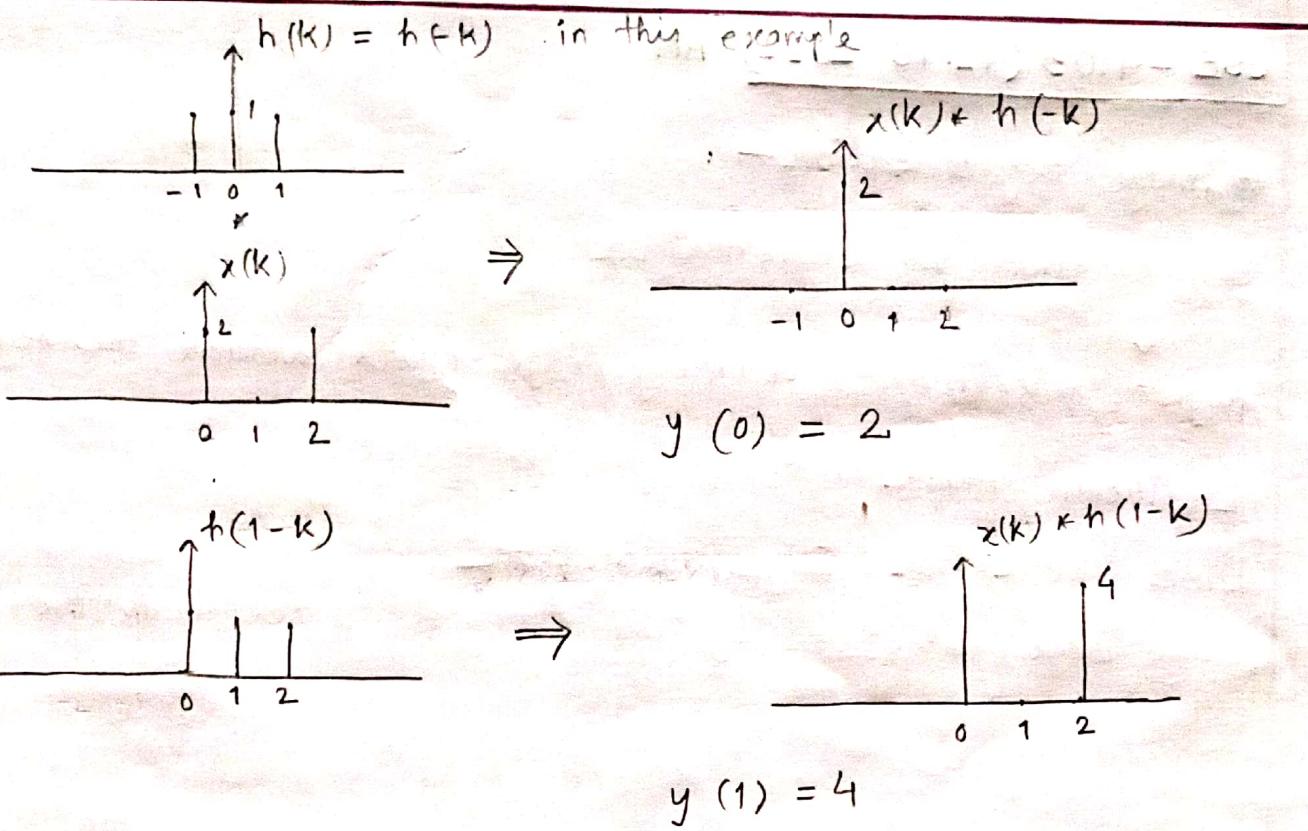
→ i) folding $\rightarrow h(-k)$

ii) shifting $\rightarrow h(p-k)$

; for each n .

iii) multiplication $\rightarrow x(k) * h(p-k)$

iv) summation $\rightarrow \sum_k x(k) * h(p-k)$



Similarly calculate for $y(2)$, $y(-1)$, $y(-2)$.

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$

$$y(n) = x_1(n) * x_2(n) = \sum_{k=-\infty}^{\infty} x_1(k) x_2(n-k)$$

General formula
of convolution.

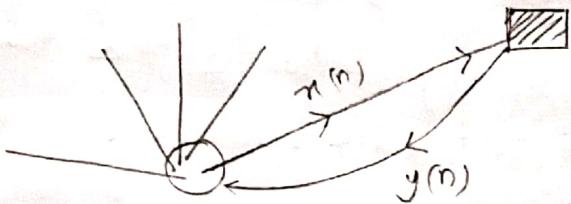
$$y(n) = x_2(n) * x_1(n) = \sum_{k=-\infty}^{\infty} x_2(k) x_1(n-k)$$

④ Prove that,

$$y(n) = x_1(n) * x_2(n) = x_2(n) * x_1(n) \Rightarrow \text{homework.}$$

④ Correlation -

Measuring relation between two signals.



$x(n)$ = transmitted signal
 $y(n)$ = received signal.

$y(n)$ is the delayed version of $x(n)$.

$$y(n) = \alpha x(n-D) + w(n) \rightarrow \text{random noise signal.}$$

↳ attenuation factor. depends on channel/environment.

If $x(n)$ and $y(n)$ are less correlated, then something unwanted exists in the environment.

$$\pi_{xy}(n) = \pi(n) + y(n)$$

If something unwanted exists → take action → find position
else

How radar system works.

use the delay to
find position

$$x(n) * y(n) = \sum_{k=-\infty}^{\infty} x(k) * y(n-k) \Rightarrow \text{convolution}$$

$$r_{xy}(n) = x(n) * y(n) = \sum_{k=-\infty}^{\infty} x(k) * y(k-n) \Rightarrow \text{correlation}$$

$$= \sum_{k=-\infty}^{\infty} x(k+n) \cdot y(n) \quad (\text{check equivalency as homework})$$

Example —

$$x(n) = \{1, 2, 3\} \quad , \quad y(n) = \{1, 0, 1, 2\}$$

↑ ↑

$$\text{Find } r_{xy}(n) = x(n) * y(n) = ?$$

Find similarity / dissimilarity between the two signals

Find how the signals are correlated.

$$r_{xy}(n) = \sum_{k=-\infty}^{\infty} x(k) y(k-n)$$

$$r_{xy}(0) = \sum_{k=-\infty}^{\infty} x(k) y(k).$$

$$r_{xy}(1) = \sum_{k=-\infty}^{\infty} x(k) y(k-1)$$

$$r_{xy}(2) = \sum_{k=-\infty}^{\infty} x(k) y(k-2)$$

$$r_{xy}(-1) = \sum_{k=-\infty}^{\infty} x(k) y(k+1)$$

$$r_{xy}(-2) = \sum_{k=-\infty}^{\infty} x(k) y(k+2)$$

Process —

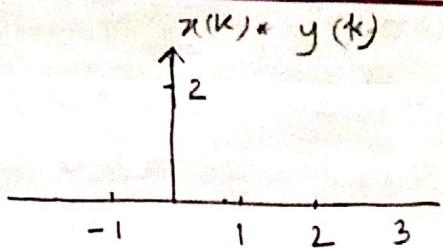
i) shifting $\rightarrow y(k-p)$

ii) multiplication $\rightarrow x(k) * y(k-p)$

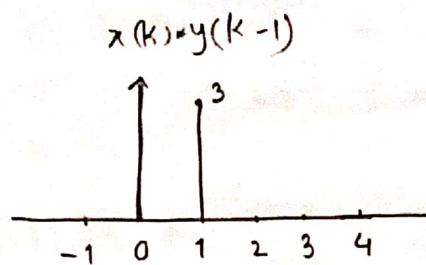
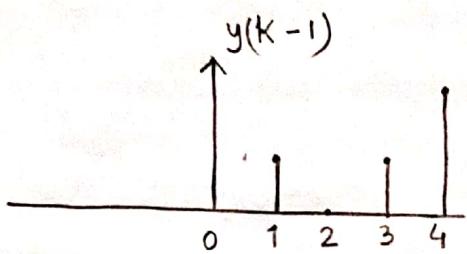
iii) summation $\rightarrow \sum_k x(k) * y(k-p)$

In convolution,
folding is necessary.
In correlation,
no folding





$$\Rightarrow r_{xy}(0) = 0 + 2 + 0 + 0 = 2$$



$$\Rightarrow r_{xy}(1) = 3$$

Similarly find out $r_{xy}(2)$, $r_{xy}(-1)$, $r_{xy}(-2)$ etc.

$$r_{xy}(n) = \{ \dots, \underset{\uparrow}{2}, 3, \dots \}$$

Correlation

Auto correlation

(Same i/p signal)

$$r_{xx}(n)$$

Cross correlation

(Different i/p signal)

$$r_{xy}(n)$$

* Proof of commutative law of convolution \rightarrow

$$y(n) = x_1(n) * x_2(n) = x_2(n) * x_1(n)$$

$$x_1(n) * x_2(n) = \sum_{k=-\infty}^{\infty} x_1(k) x_2(n-k)$$

$$x_2(n) * x_1(n) = \sum_{k=-\infty}^{\infty} x_2(k) x_1(n-k)$$

let,

$$n-k=p \quad , \quad k = -\infty, \text{ then } p = n+\infty = \infty$$
$$\Rightarrow k = n-p \quad , \quad k = \infty, \text{ then } p = n-\infty = -\infty$$

$$x_1(n) * x_2(n) = \sum_{p=-\infty}^{\infty} x_1(n-p) \cdot x_2(p)$$

$$= \sum_{p=-\infty}^{\infty} x_1(n-p) \cdot x_2(p) \quad [\text{reversing the order of summation does not change the result}]$$

$$= \sum_{p=-\infty}^{\infty} x_2(p) x_1(n-p)$$

$$= x_2(n) * x_1(n)$$



Homework -

$$\pi_{xy}(n) = \pi_{yx}(n), \text{ verify if it is true.}$$

If not equal, how will they be equal?

Hint $\pi_{xy}(n) = -\pi_{yx}(n).$

Linear Constant Coefficient Difference Equation -

$$y(n) = \alpha y(n-1) + x(n)$$

↓
o/p ↓
i/p

$$y(0) = \alpha y(-1) + x(0)$$

$$y(1) = \alpha y(0) + x(1) = \alpha^2 y(-1) + \alpha x(0) + x(1)$$

$$y(2) = \alpha y(1) + x(2) = \alpha^3 y(-1) + \alpha^2 x(0) + \alpha x(1) + x(2)$$

⋮ ↴ initial condition

$$y(n) = \alpha^{n+1} y(-1) + \alpha^n x(0) + \alpha^{n-1} x(1) + \dots + \alpha x(n-1) + x(n)$$

$$= \alpha^{n+1} y(-1) + \sum_{k=0}^n \alpha^k x(n-k)$$

case - 1 :

If the system is relaxed, i.e. the system only depends on present condition/ and all previous conditions are zero,

$$y(-1) = 0$$

$$y(n) = \sum_{k=0}^n \alpha^k x(n-k)$$

case - 2 :

If the input signal is zero, $x(n) = 0$

$$y_{z_1}(n) = \alpha^{n+1} y(-1)$$

$$y(n) = y_{z_1}(n) + y_{z_2}(n)$$

$$y(n) = a_0 y(n-1) - y(n-5) = x(n) + x(n-2)$$

$$a_0 y(n) + a_1 y(n-1) + a_2 y(n-2) + \dots + a_n y(n-N)$$

$$= b_0 x(n) + b_1 x(n-1) + b_2 x(n-2) + \dots + b_m x(n-M)$$

$$\Rightarrow \sum_{k=0}^N a_k y(n-k) = \sum_{k=0}^M b_k x(n-k)$$

↳ n^{th} order equation in general.

Solution \Rightarrow Finding an independent formula.

* Solve, $y(n) = a y(n-1) + x(n)$ where $y(-1) = 2$,
 $x(n) = u(n)$

$$y(n) = a^{n+1} \cdot y(-1) + \sum_{k=0}^n a^k x(n-k)$$

$$= 2 \cdot a^{n+1} + \sum_{k=0}^n a^k u(n-k)$$

Now find, $y(0), y(1), y(2), \dots$

④ Solve,

$$\sum_{k=0}^N a_k y(n-k) = \sum_{k=0}^M b_k x(n-k)$$

let,

$$y(n) = y_h(n) + y_p(n)$$

↓ ↓

homogeneous particular
solution solution

(i/p = 0)

↓

general solution



④ Solution of constant coefficient difference equation -

$$y(n) + a_1 y(n-1) + \dots + a_N y(n-N) = b_0 x(n) + b_1 x(n-1) + \dots + b_M x(n-M)$$

$$\Rightarrow \sum_{k=0}^N a_k y(n-k) = \sum_{k=0}^M b_k x(n-k)$$

No input = homogeneous solution.

$$y(n) = y_h(n) + y_p(n)$$

$$\sum_{k=0}^N a_k y(n-k) = 0$$

$$\Rightarrow a_0 y(n) + a_1 y(n-1) + a_2 y(n-2) + \dots + a_N y(n-N) = 0 \quad \text{--- (1)}$$

let,

$$y_h(n) = \lambda^n$$

$$y(n-k) = \lambda^{n-k} \text{ must be satisfied.}$$

From eq - (1),

$$\lambda^n + a_1 \lambda^{n-1} + a_2 \lambda^{n-2} + \dots + a_N \lambda^{n-N} = 0 \quad [a_0 = 1]$$

$$\Rightarrow \lambda^{n-N} (\lambda^n + a_1 \lambda^{n-1} + a_2 \lambda^{n-2} + \dots + a_n) = 0 \quad \left| \begin{array}{l} N \rightarrow \text{order} \\ n \rightarrow \text{index in time series} \end{array} \right.$$

$$\lambda^{n-N} = 0$$

$$\lambda^n + a_1 \lambda^{n-1} + a_2 \lambda^{n-2} + \dots + a_n = 0$$

After solving, we find N roots.

If all λ are distinct, then $\lambda_1, \lambda_2, \lambda_3, \dots, \lambda_N$

So, $y_n(n) = c_1 \lambda_1^n + c_2 \lambda_2^n + c_3 \lambda_3^n + \dots + c_N \lambda_N^n \Rightarrow$ homogeneous solution.

Example -

$$y(n) + a_1 y(n-1) = x(n), \text{ where } x(n) = u(n)$$

Order 1 cause only one difference equation.

$$y(n) = y_h(n) + y_p(n)$$

let, $y_h(n) = \lambda^n \quad & \quad x(n) = 0.$

$$y(n) + a_1 y(n-1) = x(n) \quad \text{--- (1)}$$

$$\Rightarrow \lambda^n + a_1 \lambda^{n-1} = 0$$

$$\Rightarrow \lambda^{(n-1)} (\lambda + a_1) = 0$$

$$\Rightarrow \lambda + a_1 = 0 \quad (\lambda^{n-1} \neq 0)$$

$$\Rightarrow \lambda = -a_1$$

$$y_h(n) = c (-a_1)^n \quad \text{--- (II)}$$

$$\frac{x(n)}{\text{A (constant)}}$$

$$A^M$$

$$A^n$$

$$n^m$$

$$\frac{y_p(n)}{k(\text{constant}) \cdot x(n)}$$

$$k^M \cdot x(n)$$

$$k^n \cdot x(n)$$

$$(k_1 n + k_2 n^2 + k_3 n^3 + \dots + k_m n^m) \cdot x(n)$$

let,

$$y_p(n) = k \cdot x(n) = k \cdot u(n)$$

From eq ①,

$$k \cdot u(n) + a_1 k u(n-1) = u(n) \quad \text{--- (III)}$$

n = such a minimum value that no term vanishes from this equation.

putting $n = 1$ in eq - (III),

$$k + a_1 k = 1$$

$$\Rightarrow k(1+a_1) = 1$$

$$\Rightarrow k = \frac{1}{1+a_1}$$

$$\therefore y_p(n) = \frac{1}{1+a_1} \cdot u(n)$$

$$y_p(n) = \frac{1}{1+a_1} ; \quad n \geq 0$$

$$y_t(n) = y_h(n) + y_p(n)$$

$$= c(a_1)^n + \frac{1}{1+a_1} \cdot u(n)$$

$$= c(a_1)^n + \frac{1}{1+a_1} ; \quad n \geq 0 \quad \text{--- (IV)}$$

$$y(0) = c + \frac{1}{1+a_1} \quad \text{--- (V)}$$

from eq - ①,

$$y(0) + a_1 y(-1) = u(0)$$

$$\Rightarrow y(0) = 1 - a_1 y(-1) \quad \text{--- (VI)}$$

$$c + \frac{1}{1+a_1} = 1 - a_1 y(-1)$$

$$\Rightarrow c = -a_1 y(-1) + \frac{a_1}{1+a_1}$$

$$y_t(n) = c + a_1^n + \frac{1}{1+a_1}$$

$$= (-a_1 y(-1) + \frac{a_1}{1+a_1}) \cdot (-a_1)^n + \frac{1}{1+a_1}$$

$$= (-a_1)^{n+1} y(-1) - \frac{(-a_1)^{n+1}}{1+a_1} + \frac{1}{1+a_1}$$

$$= (-a_1)^{n+1} y(-1) + \frac{-(-a_1)^{n+1} + 1}{1+a_1}$$

$$= (-a_1)^{n+1} y(-1) + \frac{1 - (-a_1)^{n+1}}{1+a_1}$$

This is the complete solution.

$$y(0) = (-a_1)^1 y(-1) + \frac{1+a_1}{1+a_1} = (-a_1) y(-1) + 1, \text{ verified.}$$

⊕ Z- Transformation of Discrete Time Signal -

Why Z-transformation is needed?

⇒ changing domain to find similarities / dissimilarities among signals.

Why use LTI systems in real life?

⇒ Linear so multiple signals can be handled.

Time invariant so delays do not change how the signal behaves.

Z transformation is used to analyze LTI systems.

Definition -

$$x(n) \xrightarrow{Z} X(z) \equiv Z[x(n)]$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}$$

→ expanding this term will convert it into an equation of z .

$$x(n) \xleftarrow{Z} X(z) \quad (\text{transform \& inverse transform})$$

Examples -

Ex $x(n) = \{1, 2, 3, 4\}$, $X(z) = ?$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}$$

$$\begin{aligned} &= \sum_{n=0}^{3} x(n) \cdot z^{-n} = x(0)z^0 + x(1)z^{-1} + x(2)z^{-2} + x(3)z^{-3} \\ &= 1 + \frac{2}{z} + \frac{3}{z^2} + \frac{4}{z^3} \end{aligned}$$

(Ans).

Hence, the power of z indicates position and the coefficient indicates value of x at that position.

$$x(n) = \{ \underset{\uparrow}{5}, 2, 3, 4 \}$$

$$X(z) = \{ 5z^0 + \underset{\uparrow}{2z^{-1}} + \underset{\uparrow}{3z^{-2}} + \underset{\uparrow}{4z^{-3}} \}$$

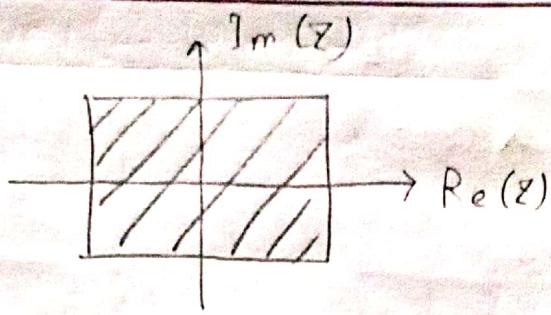
$$x(n) = \{ \underset{\uparrow}{5}, 2, 3, 4 \}$$

④ Region of convergence -

How far a series spreads.

If at some point $X(z) = \infty$, then at that point $X(z)$ diverges.

Here, z is a complex number.

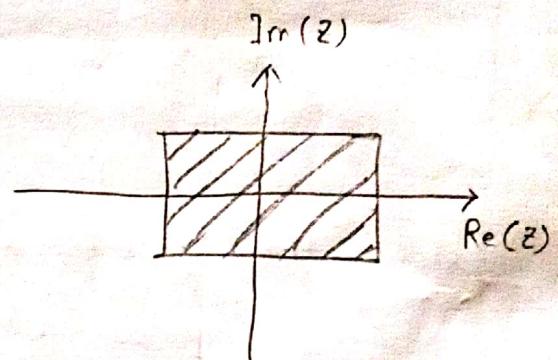


$X(z)$ exists at all values of z except at $z = 0$. \Rightarrow the region of convergence for $X(z)$.

$$\textcircled{*} \quad x(n) = \{1, 2, 3, 4\}$$

$$X(z) = \sum_{n=0}^{-3} x(n) \cdot z^{-n}$$

$$= -4 + 3z + 2z^2 + z^3$$



$X(z)$ exists at all values of z except at $z = \infty$.

$$\textcircled{*} \quad x(n) = \{1, \underset{\uparrow}{2}, 3, 4\}$$

$$X(z) = \sum_{n=-1}^2 x(n) \cdot z^{-n}$$

$$= 1 \cdot z^1 + 2 \cdot z^0 + 3z^{-1} + 4z^{-2}$$

$$= z + 2 + 3/z + 4/z^2$$

ROC = all values except 0 or ∞ .

$$\oplus \quad x(n) = \delta(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n} = x(0) z^0 = 1.$$

$$\therefore \delta(n) \xrightarrow{z} 1$$

$$\oplus \quad x(n) = \delta(n-2)$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n} = x(2) z^{-2}$$
$$= \frac{1}{z^2}$$

$$\delta(n-2) \xrightarrow{z} z^{-2}$$

$$\delta(n-k) \xrightarrow{z} z^{-k}$$

$$\delta(n+2) \xrightarrow{z} z^2$$

$$\delta(n+k) \xrightarrow{z} z^k$$

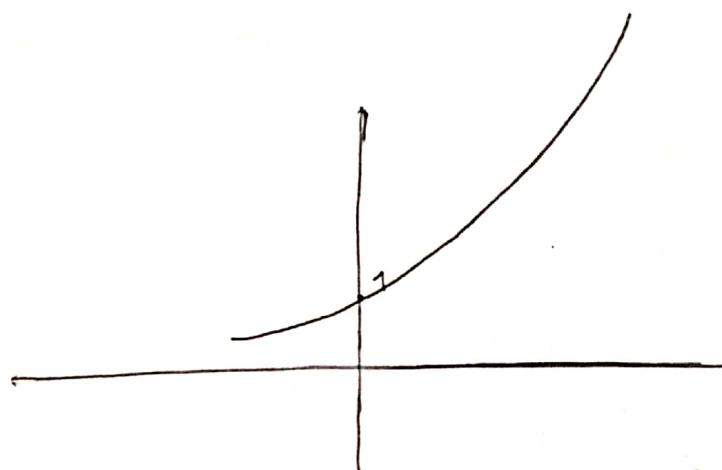
$$x_1(n) = \{-1, 2, 3\}$$

$$\begin{aligned} X(z) &= \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n} \quad z \in \mathbb{C} \\ &= \sum_{n=-1}^1 x(n) \cdot z^{-n} \end{aligned}$$

⊗ Infinite length -

$$x(n) = a^n u(n) \quad ; \quad u(n) \Rightarrow \text{unit step signal.}$$

$$\text{Given, } a = 2, \quad x(n) = 2^n u(n)$$

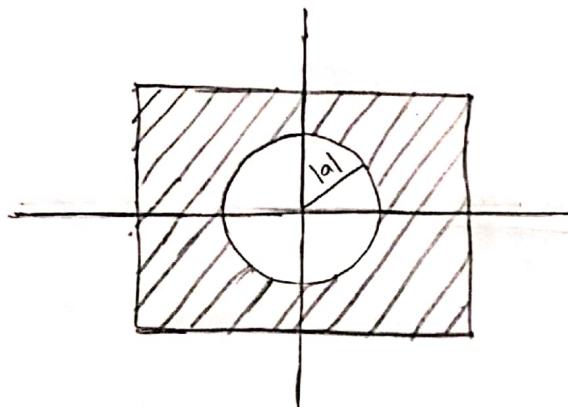


$$\begin{aligned} X(z) &= \sum_{n=-\infty}^{\infty} a^n u(n) \cdot z^{-n} = \sum_{n=0}^{\infty} a^n \cdot z^{-n} \\ &= \sum_{n=0}^{\infty} (az^{-1})^n \\ &= (az^{-1})^0 + (az^{-1})^1 + (az^{-1})^2 + \dots \\ &= 1 + (az^{-1})^1 + (az^{-1})^2 + \dots \end{aligned}$$

$$\text{Summation} = \frac{1}{1 - az^{-1}}, \quad |az^{-1}| < 1$$

$$X(z) = \frac{1}{1 - az^{-1}} \quad (\text{Ans})$$

| | |
|-------------------------|---------------------------------------|
| $ az^{-1} < 1$ | $ z $ represents a family of circles. |
| $\Rightarrow a/z < 1$ | radius $> a $ |
| $\Rightarrow a < z $ | |



$$\textcircled{*} \quad x(n) = b^n u(-n-1)$$

$$X(z) = \sum_{n=-\infty}^{\infty} b^n u(-n-1) z^{-n}$$

let
 $n = -l$

$$= \sum_{n=-1}^{-\infty} b^n z^{-n}$$

$$= \sum_{n=-1}^{-\infty} (bz^{-1})^n$$

$$= \sum_{l=1}^{\infty} (b^{-1}z)^l$$

$$= (b^{-1}z)^1 + (b^{-1}z)^2 + \dots$$

$$= (b^{-1}z) \{ 1 + (b^{-1}z) + (b^{-1}z)^2 + \dots \}$$

$$X(z) = (b^{-1}z) \cdot \frac{1}{1 - b^{-1}z}$$

$$= \cancel{(b^{-1}z)} \cdot \frac{1}{\cancel{(b^{-1}z)}(bz^{-1} - 1)}$$

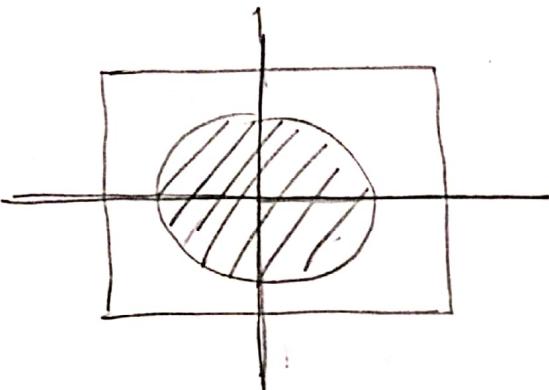
$$= \frac{1}{bz^{-1} - 1}$$

$$X(z) = \cancel{-\frac{1}{1 - bz^{-1}}} \quad \underline{(Ans)}$$

$$|b^{-1}z| < 1$$

ROC \rightarrow

$$\Rightarrow |z| < |b|$$



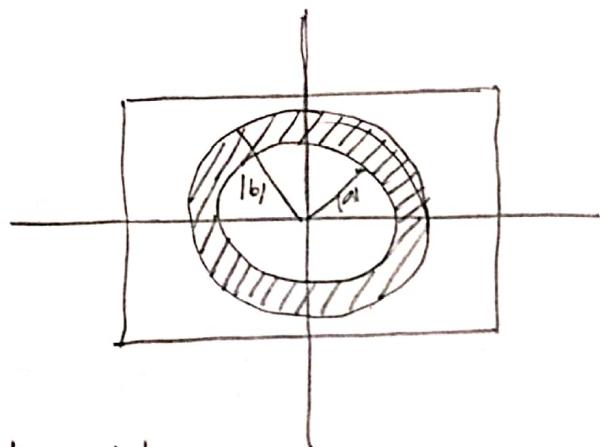
$$\oplus \quad x(n) = a^n u(n) + b^n u(-n-1)$$

$$\begin{aligned}
 X(z) &= \sum_{n=-\infty}^{\infty} x(n) z^{-n} \\
 &= \sum_{n=-\infty}^{\infty} [a^n u(n) + b^n u(-n-1)] z^{-n} \\
 &= \sum_{n=-\infty}^{\infty} a^n u(n) z^{-n} + \sum_{n=-\infty}^{\infty} b^n u(-n-1) z^{-n} \\
 &= \frac{1}{1 - az^{-1}} - \frac{1}{1 - bz^{-1}} \quad (\text{Ans})
 \end{aligned}$$

$|z| > |a|$

ROC \rightarrow

i) $|a| < |z| < |b|$
when $a < b$



ii) If $a > b$, $X(z)$ does not exist.
Conflicting statement.

$$\otimes \quad x(n) = u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} u(n)z^{-n} = 1 + z^{-1} + z^{-2} + \dots$$
$$= \frac{1}{1-z^{-1}}$$

$$\frac{1}{1-az^{-1}} \Rightarrow x(n) = a^n u(n)$$

$$\frac{-1}{1-bz^{-1}} \Rightarrow x(n) = b^n u(-n-1)$$

Coefficient of $z^{-1} \Rightarrow a \Rightarrow a^n$

$x(z)$ if negative $\Rightarrow u(-n-1)$.

$$\otimes \quad x(n) = 2^n u(n) + 5^n u(n)$$

$$X(z) = \frac{1}{1-2z^{-1}} + \frac{1}{1-5z^{-1}}$$

$$\text{ROC} \rightarrow |z| > 2 \quad \text{and} \quad |z| > 5$$

$$\text{so } |z| > 5 \quad \text{i.e. } \max(a, b)$$

If z is less than both a and b then

$$\text{ROC} = \min(a, b).$$

$$\textcircled{2} \quad x(n) = 5^n u(n) + 3^n u(-n-1)$$

Practice more

$$X(z) = \frac{1}{1-5z^{-1}} - \frac{1}{1-3z^{-1}}$$

$$|z| > 5 \quad \text{and} \quad |z| < 3$$

In this case, $X(z)$ does not exist.

~~Properties~~ Properties -

i) Linearity :

$$x_1(n) \xrightarrow{Z} X_1(z)$$

$$x_2(n) \xrightarrow{Z} X_2(z)$$

$$a_1 x_1(n) + a_2 x_2(n) \xrightarrow{Z} a_1 X_1(z) + a_2 X_2(z)$$

$$x(n) = 6 \underbrace{(5^n u(n))}_{x_1} + 2 \underbrace{(3^n u(-n-1))}_{x_2}$$

$$X(z) = 6 \cdot X_1(z) + 2 \cdot X_2(z)$$

ii) Shifting :

$$x(n) \xrightarrow{Z} X(z)$$

$$x(n+k) \xrightarrow{Z} z^k X(z)$$

$$x(n) = \{1, \underset{\uparrow}{2}, 3\}$$

$$X(z) = z + 2 + 3z^{-1}$$

$$x(n-1) \xrightarrow{z} z^{-1} \cdot X(z)$$

$$= z^{-1} (z + 2 + 3z^{-1})$$

$$= 1 + 2z^{-1} + 3z^{-2}$$

$$x(n-1) = \{1, 2, 3\}$$

↑ same, shifting property proved.

$$X(z) = 1 + 2z^{-1} + 3z^{-2}$$

Properties of z -transform \equiv LTI systems properties.

iii) Time reversal :

$$x(n) \xrightarrow{z} X(z)$$

$$x(-n) \xrightarrow{z} X(z^{-1})$$

$$\left. \begin{array}{l} x(n) = u(n) \\ X(z) = \frac{1}{1-z} \end{array} \right| \begin{array}{l} x_1(n) = u(n) \xrightarrow{z} \frac{1}{1-z^{-1}} \\ X_1(z) = \frac{1}{1-z^{-1}} \end{array}$$

$$x(n) = u(n)$$

$$X(z) = \sum_{n=-\infty}^{\infty} u(n) \cdot z^{-n} = \sum_{n=0}^{-\infty} z^{-n} = 1 + z + z^2 + \dots$$

$$= \frac{1}{1-z}$$

iv) Scaling :

$$x(n) \xrightarrow{z} X(z)$$

$$\alpha^n x(n) \xrightarrow{z} X(\alpha^{-1}z)$$

$$x(n) = \alpha^n u(n) = \alpha^n x_1(n)$$

$$x_1(n) \xrightarrow{z} \frac{1}{1-z^{-1}} \Rightarrow X_1(z) = \frac{1}{1-z^{-1}}$$

$$X(z) = X_1(\alpha^{-1}z) = \frac{1}{1 - (\alpha^{-1}z)^{-1}} = \frac{1}{1 - \alpha z^{-1}}$$

$$\text{If } x(n) = a^n u(n) + b^n$$

$$\text{find } X_1(z) \rightarrow X_2(z) = X_1(\alpha^{-1}z) \rightarrow X(z) = X_2(b^{-1}z)$$

④ Properties cont. -

v) Differentiation:

$$\text{if } x(n) \xleftrightarrow{z} X(z)$$

$$\text{then } n \cdot x(n) \xleftrightarrow{z} -z \frac{d}{dz} \{X(z)\}$$

Example -

$$x_1(n) = n \underbrace{a^n u(n)}_{x_1(z) = ?}$$

$$x_1(n) = n x(n)$$

$$X(z) = \frac{1}{1 - az^{-1}}$$

$$X_1(z) = -z \frac{d}{dz} \{X(z)\}$$

$$= -z \frac{d}{dz} \left[\frac{1}{1 - az^{-1}} \right]$$

$$\begin{aligned} &= z \cdot \frac{1}{(1 - az^{-1})^2} \cdot \frac{d}{dz} (1 - az^{-1}) \\ &= \frac{z}{(1 - az^{-1})^2} \cdot \frac{a}{z^2} \\ &= \frac{az^{-1}}{(1 - az^{-1})^2} \end{aligned}$$

vi) Convolution:

$$x_1(n) \xleftrightarrow{z} X_1(z)$$

$$x_2(n) \xleftrightarrow{z} X_2(z)$$

$$x(n) = x_1(n) * x_2(n) \xleftrightarrow{z} X(z) = X_1(z) \cdot X_2(z)$$

Time domain \rightarrow convolution

z -domain \rightarrow multiplication.

Example -

$$x_1(n) = \{1, \underset{\uparrow}{2}, 3\}$$

$$x_2(n) = \{2, \underset{\uparrow}{1}, 0, 2\}$$

\Rightarrow Home work.

$$\text{Find } x(n) = x_1(n) * x_2(n) = ?$$

$$X_1(z) = z + 2 + 3z^{-1}$$

$$X_2(z) =$$

Example -

$$x_1(n) = \{2, \underset{\uparrow}{0}, 2\}$$

$$x_2(n) = \{1, \underset{\uparrow}{1}, 1\}$$

$$X_1(z) = 2 + 2z^{-2}$$

$$X_2(z) = z + 1 + z^{-1}$$

$$X_1(z), X_2(z) = (2 + 2z^{-2}) \cdot (z + 1 + z^{-1})$$

$$= 2z + 2z^{-1} + 2 + 2z^{-2} + 2z^{-1} + 2z^{-3}$$

$$= 2z + 2 + 4z^{-1} + 2z^{-2} + 2z^{-3}$$

$$x(n) = \{2, \underset{\uparrow}{2}, 4, 2, 2\}$$

④ Prove that,

$$x(n) = x_1(n) * x_2(n) \xleftarrow{Z} x_1(z) \cdot x_2(z) = X(z)$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} \{x_1(n) * x_2(n)\} z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} \left\{ \sum_{k=-\infty}^{\infty} x_1(k) * x_2(n-k) \right\} z^{-n}$$

↔

$$= \sum_{k=-\infty}^{\infty} x_1(k) \underbrace{\sum_{n=-\infty}^{\infty} x_2(n-k)}_{z^{-n}} \quad X_2(z)$$

$$= \sum_{k=-\infty}^{\infty} x_1(k) \cdot z^{-k} \cdot X_2(z)$$

$$= X_2(z) \sum_{k=-\infty}^{\infty} x_1(k) \cdot z^{-k}$$

$$= X_2(z) \cdot X_1(z)$$

$$= X_1(z) \cdot X_2(z) \quad [\text{Proved}]$$

vii) Correlation :

$$x_1(n) \leftrightarrow z \rightarrow X_1(z)$$

$$x_2(n) \leftrightarrow z \rightarrow X_2(z)$$

$$x(n) = x_1(n) * x_2(n) \leftrightarrow z \rightarrow X(z) = X_1(z) \cdot X_2(z^{-1})$$

↓
Find $X_2(z)$

Replace z with z^{-1}