

CONTENTS

| | | |
|---|-------------------------|---|
| 1 | Stability | 1 |
| 2 | Routh Hurwitz Criterion | 1 |
| 3 | Compensators | 1 |
| 4 | Nyquist Plot | 1 |
| 5 | State space model | 1 |

Abstract—This manual is an introduction to control systems based on GATE problems. Links to sample Python codes are available in the text.

1 STABILITY

2 ROUTH HURWITZ CRITERION

3 COMPENSATORS

4 NYQUIST PLOT

5 STATE SPACE MODEL

5.0.1. Consider the circuit given below, here $R_1 = 1\text{K}\Omega$, $R_2 = 2\text{K}\Omega$, $C_1 = C_2 = 1\text{mF}$. Both the capacitors were charged to 1 C. Find Differential equations for q_1 and q_2 .

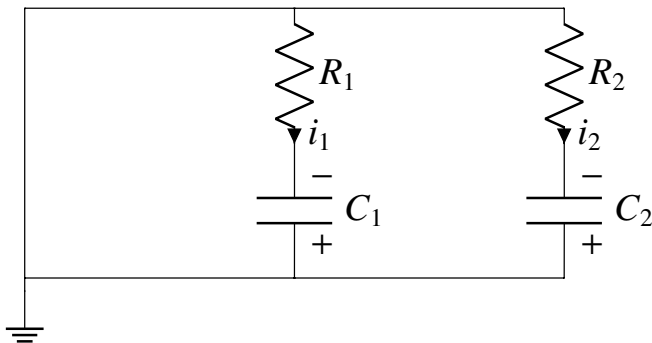


Fig. 5.0.1: Circuit diagram

Solution: We know that, $i_1 = -\dot{q}_1$, $i_2 = -\dot{q}_2$ ('-' as charge is depleting)

Applying KVL on 1st branch we get:

$$-\dot{q}_1 R_1 - \frac{q_1}{C_1} = 0 \quad (5.0.1.1)$$

$$\dot{q}_1 = -q_1 \quad (5.0.1.2)$$

Now, Applying KVL On 2nd branch we get:

$$-\dot{q}_2 R_2 - \frac{q_2}{C_2} = 0 \quad (5.0.1.3)$$

$$\dot{q}_2 = -2q_2 \quad (5.0.1.4)$$

If I represent this as state space model

$$\text{Taking } x = \begin{pmatrix} q_1 \\ q_2 \end{pmatrix}$$

$$\text{And here } x(0) = \begin{pmatrix} q_1(0) \\ q_2(0) \end{pmatrix} = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

\therefore State space model can be represented as

$$\dot{x} = \begin{pmatrix} -q_1 \\ -2q_2 \end{pmatrix} = \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix} x \quad (5.0.1.5)$$

5.0.2. Consider the linear system :

$$\dot{x} = \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix} x \quad (5.0.2.1)$$

with initial condition: $x(0) = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$. Find $x(t)$

$$(A) \ x(t) = \begin{pmatrix} e^{-t} & te^{-2t} \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

$$(B) \ x(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

$$(C) \ x(t) = \begin{pmatrix} e^{-t} & t^2 e^{-2t} \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

$$(D) \ x(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

Solution: Given expression is the state equation, as it can be written in the following form

$$\dot{x} = Ax + BU \quad (5.0.2.2)$$

Here, $A = \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix}$ and $U = 0$

Solution to this can be given by

$$x(t) = \phi(t)x(0) \quad (5.0.2.3)$$

Where $\phi(t)$ is called the state-transition matrix and is given by

$$\phi(t) = \mathcal{L}^{-1} \left((sI - A)^{-1} \right) \quad (5.0.2.4)$$

$$(sI - A) = \begin{pmatrix} s+1 & 0 \\ 0 & s+2 \end{pmatrix} \quad (5.0.2.5)$$

$$(sI - A)^{-1} = \begin{pmatrix} \frac{1}{s+1} & 0 \\ 0 & \frac{1}{s+2} \end{pmatrix} \quad (5.0.2.6)$$

$$\mathcal{L}^{-1} \left((sI - A)^{-1} \right) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{pmatrix} \quad (5.0.2.7)$$

Given $x(0) = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$

$$\therefore x(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix} \quad (5.0.2.8)$$

$$\therefore \text{option}(D) \quad (5.0.2.9)$$

**Derivation for state transition matrix:*

$$\dot{x} = Ax \quad (5.0.2.10)$$

Applying laplace transform, we get

$$sX(s) - x(0) = AX(s) \quad (5.0.2.11)$$

$$(sI - A)X(s) = x(0) \quad (5.0.2.12)$$

$$X(s) = (sI - A)^{-1} x(0) \quad (5.0.2.13)$$

$$(5.0.2.14)$$

Taking inverse laplace transform

$$x(t) = \mathcal{L}^{-1} \left((sI - A)^{-1} \right) x(0) \quad (5.0.2.15)$$

→ Where $\mathcal{L}^{-1} \left((sI - A)^{-1} \right)$ is called State transition matrix.

5.0.3. Draw block diagram for the the above mentioned state model.

Solution: Here, clearly Transfer function is given by

$$X(s) = (sI - A)^{-1} x(0) \quad (5.0.3.1)$$

So, over here open loop gain, $G(s) = (sI - A)^{-1}$
Input being constant = $x(0)$, and output is $X(s)$

∴ From the circuit signal flows as current i

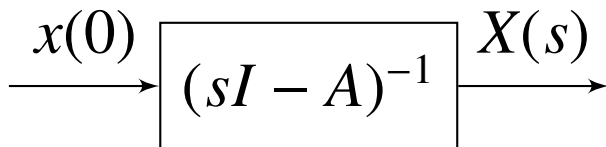


Fig. 5.0.3: Block diagram