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If I represent this as state space model Taking $x = \begin{pmatrix} q_1 \\ q_2 \end{pmatrix}$

And here $x(0) = \begin{pmatrix} q_1(0) \\ q_2(0) \end{pmatrix} = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$

:. State space model can be represented as

 $\dot{x} = \begin{pmatrix} -q_1 \\ -2q_2 \end{pmatrix} = \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix} x$ (5.0.1.5)

Abstract—This manual is an introduction to control 5.0.2. Consider the linear system : systems based on GATE problems.Links to sample Python codes are available in the text.

1 STABILITY

2 ROUTH HURWITZ CRITERION

- 3 Compensators
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- 5 STATE SPACE MODEL
- 5.0.1. Consider the circuit given below, here R_1 = $1K\Omega$, $R_2 = 2K\Omega$, $C_1 = C_2 = 1mF$. Both the capacitors were charged to 1 C. Find Differential equations for q_1 and q_2 .

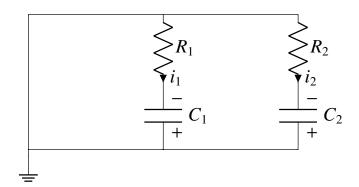


Fig. 5.0.1: Circuit diagram

Solution: We know that, $i_1 = -\dot{q}_1$, $i_2 = -\dot{q}_2$ ('-' as charge is depleting)

Applying KVL on 1st branch we get:

$$-\dot{q}_1 R_1 - \frac{q_1}{c_1} = 0 (5.0.1.1)$$

$$\dot{q}_1 = -q_1 \tag{5.0.1.2}$$

Now, Applying KVL 0n 2nd branch we get:

$$-\dot{q}_2 R_2 - \frac{q_2}{c_2} = 0 (5.0.1.3)$$

$$\dot{q}_2 = -2q_2 \tag{5.0.1.4}$$

$$\dot{x} = \begin{pmatrix} -1 & 0\\ 0 & -2 \end{pmatrix} x \tag{5.0.2.1}$$

with initial condition: $x(0) = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$. Find x(t)

(A)
$$\mathbf{x}(t) = \begin{pmatrix} e^{-t} & te^{-2t} \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

(B)
$$x(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

(C) $x(t) = \begin{pmatrix} e^{-t} & t^2 e^{-2t} \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$
(D) $x(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$

(C)
$$\mathbf{x}(t) = \begin{pmatrix} e^{-t} & t^2 e^{-2t} \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

(D)
$$\mathbf{x}(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

Solution: Given expression is the state equation, as it can be written in the following form

$$\dot{x} = Ax + BU \tag{5.0.2.2}$$

Here,
$$A = \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix}$$
 and $U = 0$

Solution to this can be given by

$$x(t) = \phi(t)x(0)$$
 (5.0.2.3)

Where $\phi(t)$ is called the state-transition matrix and is given by

$$\phi(t) = \mathcal{L}^{-1}\left(\left(sI - A\right)^{-1}\right)$$
 (5.0.2.4)

$$\begin{pmatrix} sI - A \end{pmatrix} = \begin{pmatrix} s+1 & 0 \\ 0 & s+2 \end{pmatrix}$$
(5.0.2.5)

$$(sI - A)^{-1} = \begin{pmatrix} \frac{1}{s+1} & 0\\ 0 & \frac{1}{s+2} \end{pmatrix}$$
 (5.0.2.6)

$$\mathcal{L}^{-1}\left(\left(sI-A\right)^{-1}\right) = \begin{pmatrix} e^{-t} & 0\\ 0 & e^{-2t} \end{pmatrix}$$
 (5.0.2.7)

Given
$$x(0) = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$

$$\therefore x(t) = \begin{pmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix}$$
 (5.0.2.8)

$$\therefore option(D) \qquad (5.0.2.9)$$

*Derivation for state transition matrix:

$$\dot{x} = Ax \tag{5.0.2.10}$$

Applying laplace transform, we get

$$sX(s) - x(0) = AX(s)$$
 (5.0.2.11)

$$(sI - A)X(s) = x(0)$$
 (5.0.2.12)

$$X(s) = (sI - A)^{-1} x(0)$$
 (5.0.2.13)

(5.0.2.14)

Taking inverse laplace transform

$$x(t) = \mathcal{L}^{-1}\left(\left(sI - A\right)^{-1}\right)x(0) \qquad (5.0.2.15)$$

 \rightarrow Where $\mathcal{L}^{-1}((sI - A)^{-1})$ is called State transition matrix.

5.0.3. Draw block diagram for the the above mentioned state model.

Solution: Here, clearly Transfer function is given by

$$X(s) = (sI - A)^{-1} x(0)$$
 (5.0.3.1)

Input being constant = x(0), and output is X(s)

$$X_1(s) = \frac{1}{s+1} x_1(0) \tag{5.0.3.2}$$

$$X_2(s) = \frac{1}{s+2} x_2(0)$$
 (5.0.3.3)

Lets consider a closed loop negative unity feed back system, We know its open loop gain

$$=\frac{G(s)}{1+G(s)}\tag{5.0.3.4}$$

We can re-write the above to transfer equations

$$X_1(s) = \frac{\frac{1}{s}}{\frac{1}{s} + 1} x_1(0)$$
 (5.0.3.5)

$$X_2(s) = \frac{\frac{1}{s+1}}{\frac{1}{s+1} + 1} x_2(0)$$
 (5.0.3.6)

Below are the block the block diagrams for $X_1(s)$ and $X_2(s)$

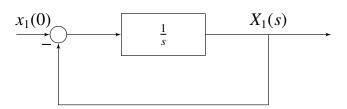


Fig. 5.0.3: block diagram for X_1

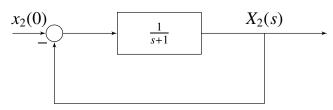


Fig. 5.0.3: block diagram for X_2

5.0.4. Describe how an oscillator works

Solution: Oscillators convert a DC input (the supply voltage) into an AC output (the waveform)

Given Below is basic block diagram

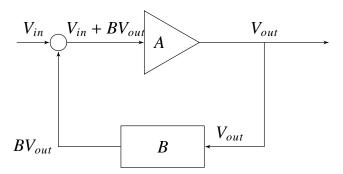


Fig. 5.0.4: block diagram for oscillator

Resonant frequency, is the frequency at which oscillator oscillates, it depends on R/L/C components of the circuit it's been fed back through.

Oscillators work because they overcome the losses of their positive feedback circuit either in the form of a capacitor, inductor or both. In other words, an oscillator is a an amplifier which uses positive feedback that generates an output frequency without the use of an input signal.

Oscillators gain can be given as follows:

From the above circuit we have:

A = Amplifiers gain, B = positive feedback gain, G = total gain of the circuit

$$A(V_{in} + BV_{out}) = V_{out}$$
 (5.0.4.1)

$$G = \frac{V_{out}}{V_{in}} = \frac{A}{1 - AB} \tag{5.0.4.2}$$

We get,

sustained oscillations when AB = 1

Exponentially decaying oscillation for AB < 1Exponentially increasing oscillation AB > 1 (unstable)

5.0.5. For Hartley oscillator frequency generated can be given as

$$f = \frac{1}{2\pi\sqrt{(L_1 + L_2)C}}$$
 (5.0.5.1)

Taking,

$$L_1 = 1\mu H \tag{5.0.5.2}$$

$$L_1 = 1\mu H$$
 (5.0.5.2)
 $L_2 = 1\mu H$ (5.0.5.3)
 $C = 1.2 nF$ (5.0.5.4)

$$C = 1.2pF$$
 (5.0.5.4)

We get f = 103 MHz

Feedback factor for Hartley given by:

$$Feedback factor = \frac{L_2}{L_1} = 1 (5.0.5.5)$$

W.K.T, AB = 1

∴ Minimum amplification Gain = 1