



# *Laplace Transform Method*

– *Laplace Transform as Solution Tool*



# ***Laplace Transform Based Approach***



## ***Laplace Transform Definition***

**Laplace** transform is an important **tool** for solving **LTI** systems without explicit integration. **Laplace** transform is an integral **transform**, defined for a function **f(t)**, as below.

$$F(s) = \int_0^{\infty} e^{-st} f(t) dt = L(f)$$

The above **integral**, and the Laplace **transform**, exists if the integrand '**e<sup>-st</sup> f(t)**' goes to **zero** fast enough as  $t \rightarrow \infty$ .

Here, '**s**' is called **Laplace** variable and is a complex quantity defined as ' **$\sigma \pm j\omega$** ', where, ' **$\sigma$** ' represents the **real axis** while, ' **$j\omega$** ' represents the **imaginary axis** (s – plane).

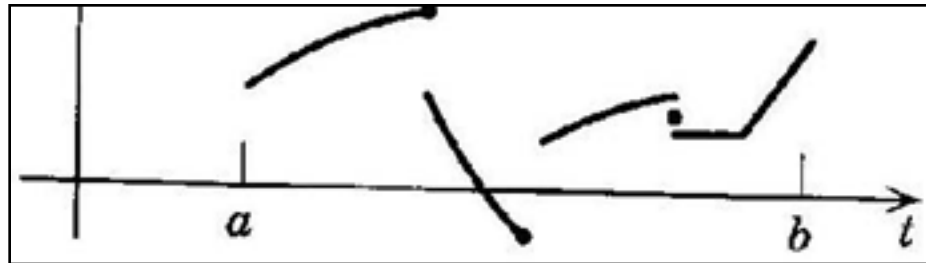


# *Properties of Laplace Transform*

**Laplace** transform is a **linear operation** and hence it is **applicable** only in the context of **LTI** systems.

**Laplace** transform, by virtue of **integral**, replaces every operation of **calculus** by an **algebraic** operation.

However,  $f(t)$  needs to be at least **piece-wise continuous**.





# Typical Laplace Transforms

	$f(t)$	$F(s)$
1	Unit impulse $\delta(t)$	1
2	Unit step $1(t)$	$\frac{1}{s}$
3	$t$	$\frac{1}{s^2}$
4	$\frac{t^{n-1}}{(n-1)!} \quad (n = 1, 2, 3, \dots)$	$\frac{1}{s^n}$
5	$t^n \quad (n = 1, 2, 3, \dots)$	$\frac{n!}{s^{n+1}}$
6	$e^{-at}$	$\frac{1}{s+a}$
7	$te^{-at}$	$\frac{1}{(s+a)^2}$
8	$\frac{1}{(n-1)!} t^{n-1} e^{-at} \quad (n = 1, 2, 3, \dots)$	$\frac{1}{(s+a)^n}$
9	$t^n e^{-at} \quad (n = 1, 2, 3, \dots)$	$\frac{n!}{(s+a)^{n+1}}$

**Transform table is bi-directional**  
i.e. we can get **F(s)** for a given **f(t)** or **vice versa**.



# *Laplace Transform as Solution Tool*

As **Laplace** transform involves **integration**, we can use it to **convert differential** equations into **algebraic** equations, using the following **properties**.

$$\mathcal{L}_{\pm} \left[ \frac{d^n}{dt^n} f(t) \right] = s^n F(s) - \sum_{k=1}^n s^{n-k} f^{(k-1)}(0_{\pm})$$

$$\text{where } f^{(k-1)}(t) = \frac{d^{k-1}}{dt^{k-1}} f(t)$$

$$\mathcal{L} \left[ \int_0^t f(t) dt \right] = \frac{F(s)}{s}$$



## *Laplace Transform Example*

Consider the following **2<sup>nd</sup> order** LTI system.

$$\ddot{c}(t) + 2\zeta\omega_n\dot{c}(t) + \omega_n^2c(t) = \omega_n^2r(t)$$

We can take **term-by-term** Laplace transform of the above **model** to arrive at the **algebraic** equation, as shown below.

$$\begin{aligned} \left[ s^2C(s) - \dot{c}(0) - sc(0) \right] + 2\zeta\omega_n \left[ sC(s) - c(0) \right] + \omega_n^2C(s) &= \omega_n^2R(s) \\ \left[ s^2 + 2\zeta\omega_ns + \omega_n^2 \right] C(s) &= \left\{ \dot{c}(0) + (s + 2\zeta\omega_n)c(0) \right\} + \omega_n^2R(s) \end{aligned}$$

The above **algebraic system** can be suitably **manipulated** and  $c(t)$  can be **obtained** through the table of **transforms**.



## *Summary*

Laplace **transform** provides an **elegant way** of converting **differential** equations into **algebraic** form, which are significantly **easier to manipulate**.





# ***Transfer Function Based Solution***



## ***TF as LTI System Solution***

**TF** is an important **solution** building block and is defined as **ratio** of the Laplace transforms of **output and input** for a system (under **zero initial** conditions), as detailed below.

$$a_0 \frac{d^n y}{dt^n} + a_1 \frac{d^{n-1} y}{dt^{n-1}} + \cdots + a_{n-1} \frac{dy}{dt} + a_n y = b_0 \frac{d^m u}{dt^m} + b_1 \frac{d^{m-1} u}{dt^{m-1}} + \cdots + b_{m-1} \frac{du}{dt} + b_m u$$

$$\text{Transfer Function: } G(s) = \frac{L(\text{Output})}{L(\text{Input})} \Big|_{\text{Zero Initial Conditions}} = \frac{Y(s)}{U(s)}$$

$$(a_0 s^n + a_1 s^{n-1} + \cdots + a_{n-1} s + a_n) Y(s) = (b_0 s^m + b_1 s^{m-1} + \cdots + b_{m-1} s + b_m) U(s)$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_0 s^m + b_1 s^{m-1} + \cdots + b_{m-1} s + b_m}{a_0 s^n + a_1 s^{n-1} + \cdots + a_{n-1} s + a_n}$$



# ***Transfer Function Features***

Transfer function **G(s)** is the s-domain **unit impulse response**, as shown below.

$$Y(s) = G(s) \cdot U(s) \rightarrow \text{For } U(s) = \delta(s) = 1, \quad Y(s) = G(s)$$

In general, **G(s)** is represented in **polynomial & factored** forms, as shown below.

$$\begin{aligned} G(s) &= K \frac{(s^m + b_1 s^{m-1} + \dots + b_{m-1} s + b_m)}{(s^n + a_1 s^{n-1} + \dots + a_{n-1} s + a_n)} \\ &= K \frac{(s - z_1)(s - z_2) \cdots (s - z_m)}{s^k (s - p_{k+1})(s - p_{k+2}) \cdots (s - p_n)} \end{aligned}$$

$p_i$ 's: **poles**; roots of the **denominator** polynomial

$z_j$ 's: **zeros**; roots of the **numerator** polynomial.

**K**: **Gain** parameter

**k**: **System Type**



## ***Transfer Function Example***

Consider the **Laplace** transform of a **2<sup>nd</sup>** order system.

$$\left[ s^2 + 2\zeta\omega_n s + \omega_n^2 \right] C(s) = \left\{ \dot{c}(0) + (s + 2\zeta\omega_n) c(0) \right\} + \omega_n^2 R(s)$$

We can **write** the corresponding **transfer function**, by applying its **definition** as follows.

$$G(s) = \frac{C(s)}{R(s)} = \frac{\omega_n^2}{\left[ s^2 + 2\zeta\omega_n s + \omega_n^2 \right]} = \frac{\omega_n^2}{\left[ (s + \zeta\omega_n)^2 + \omega_d^2 \right]}$$
$$\omega_d^2 = \omega_n^2 (1 - \zeta^2); \quad G(s) = \frac{\omega_n^2}{(s + \zeta\omega_n + j\omega_d)(s + \zeta\omega_n - j\omega_d)}$$



## ***LTI System Responses Using TF***

**Transfer functions** are used to generate time **responses** of LTI systems based on the principle of **superposition**.

This involves (1) **decomposing**  $Y(s)$  into its characteristic components and (2) **mapping** these components to their time domain **counterparts**.

**Decomposition** is based on the **premise** that any complex **LTI** system can be **synthesized** as a linear **combination** of 1<sup>st</sup> and 2<sup>nd</sup> order **terms**.

**Partial fractions** is standard method for **decomposing**.



# ***Partial Fractions Concept***

**Partial fraction** decomposition uses method of **residues** to **decompose** an '**n<sup>th</sup>**' order fraction into a set of '**n**' **1<sup>st</sup>** order fractions.

Consider **n<sup>th</sup>** **order** system, along with its **decomposed form**, as given below.

$$G(s) = K \frac{(s - z_1)(s - z_2) \cdots (s - z_m)}{(s - p_1)(s - p_2) \cdots (s - p_n)} = \frac{A_1}{s - p_1} + \frac{A_2}{s - p_2} + \cdots + \frac{A_n}{s - p_n}$$

Here, **A<sub>1</sub>** to **A<sub>n</sub>** are called the **residues**.



## ***Partial Fractions Concept***

The **residues** represent the **contributions** of each of the **factors** to the total **response** and are obtained as follows.

$$\text{Distinct Poles: } A_i = \left[ (s - p_i) Y(s) \right] \big|_{s=p_i}; \quad i = 1, n$$

$$\text{Multiple Poles: } G(s) = K \frac{(s - z_1)(s - z_2) \cdots (s - z_m)}{(s - p_1)^k (s - p_{k+1}) \cdots (s - p_n)}$$

$$Y(s) = \frac{A_1}{s - p_1} + \frac{A_2}{(s - p_1)^2} + \cdots + \frac{A_k}{(s - p_1)^k} + \cdots + \frac{A_i}{s - p_i}$$

$$A_j = \frac{1}{(k - j)!} \frac{d^{k-j}}{ds^{k-j}} \left[ (s - p_1)^k Y(s) \right] \big|_{s=p_1}; \quad j = 1, k; \quad i = k + 1, n$$

We can also **compare** the coefficients of applicable **numerator** polynomial in **certain cases**.



## *Partial Fractions Example – Distinct*

Obtain unit impulse response of the given TF.

$$G(s) = \frac{(s+3)}{(s+1)(s+2)}$$

$$\begin{aligned} Y(s) &= \frac{A_1}{s+1} + \frac{A_2}{s+2} \\ &= \frac{(A_1 + A_2)s + (2A_1 + A_2)}{(s+1)(s+2)} \end{aligned}$$

$$A_1 + A_2 = 1; \quad 2A_1 + A_2 = 3$$

$$A_1 = 2; \quad A_2 = -1$$

$$p_1 = -1; \quad p_2 = -2; \quad z_1 = -3; \quad K = 1$$

$$Y(s) = \frac{(s+3)}{(s+1)(s+2)} = \frac{A_1}{s+1} + \frac{A_2}{s+2}$$

$$A_1 = [(s+1)Y(s)] = \left[ \frac{(s+3)}{(s+2)} \right] \Big|_{s=-1} = 2$$

$$A_2 = [(s+2)Y(s)] = \left[ \frac{(s+3)}{(s+1)} \right] \Big|_{s=-2} = -1$$

$$Y(s) = \frac{2}{s+1} - \frac{1}{s+2} = \frac{(s+3)}{(s+1)(s+2)}$$

$$y(t) = L^{-1}Y(s) = 2e^{-t} - e^{-2t}$$





## *Partial Fractions Example – Multiple*

Obtain **unit impulse response** of the given system.

$$G(s) = \frac{s^2 + 2s + 3}{(s + 1)^3}$$

$$\begin{aligned} b_1 s^2 + (2b_1 + b_2)s \\ + (b_1 + b_2 + b_3) \\ = s^2 + 2s + 3 \\ b_1 = 1; \quad b_2 = 0; \quad b_3 = 2 \end{aligned}$$

$$p_1 = -1, -1, -1; \quad z_1 = -1 \pm j\sqrt{2}; \quad K = 1$$

$$Y(s) = \frac{b_1}{s+1} + \frac{b_2}{(s+1)^2} + \frac{b_3}{(s+1)^3}$$

$$b_3 = \left[ (s+1)^3 \frac{B(s)}{A(s)} \right]_{s=-1} = 2$$

$$b_2 = \frac{d}{ds} \left[ (s+1)^3 \frac{B(s)}{A(s)} \right]_{s=-1} = 0$$

$$b_1 = \frac{1}{2!} \frac{d^2}{ds^2} \left[ (s+1)^3 \frac{B(s)}{A(s)} \right]_{s=-1} = 1$$

$$Y(s) = \frac{1}{s+1} + \frac{2}{(s+1)^3} \rightarrow y(t) = (1+t^2)e^{-t}$$



## *Summary*

Transfer function is the **building block** for obtaining the response of **LTI systems** and its decomposition using **partial fractions** provides a **convenient** methodology.