Basic IIR Digital Filter Structures

- The causal IIR digital filters we are concerned with in this course are characterized by a real rational transfer function of z^{-1} or, equivalently by a constant real coefficient difference equation
- From the difference equation representation, it can be seen that the realization of the causal IIR digital filters requires some form of feedback

Basic IIR Digital Filter Structures

- An *N*-th order IIR digital transfer function is characterized by 2*N*+1 unique coefficients, and in general, requires 2*N*+1 multipliers and 2*N* two-input adders for implementation
- **Direct form IIR filters**: Filter structures in which the multiplier coefficients are precisely the coefficients of the transfer function

• Consider for simplicity a 3rd-order IIR filter with a transfer function

$$H(z) = \frac{P(z)}{D(z)} = \frac{p_0 + p_1 z^{-1} + p_2 z^{-2} + p_3 z^{-3}}{1 + d_1 z^{-1} + d_2 z^{-2} + d_3 z^{-3}}$$

• We can implement H(z) as a cascade of two filter sections as shown on the next slide

$$X(z) \longrightarrow H_1(z) \xrightarrow{W(z)} H_2(z) \longrightarrow Y(z)$$

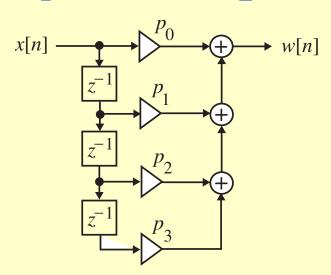
where

$$H_1(z) = \frac{W(z)}{X(z)} = P(z) = p_0 + p_1 z^{-1} + p_2 z^{-2} + p_3 z^{-3}$$

$$H_2(z) = \frac{Y(z)}{W(z)} = \frac{1}{D(z)} = \frac{1}{1 + d_1 z^{-1} + d_2 z^{-2} + d_3 z^{-3}}$$

• The filter section $H_1(z)$ can be seen to be an FIR filter and can be realized as shown below

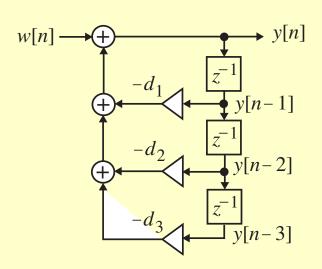
$$w[n] = p_0x[n] + p_1x[n-1] + p_2x[n-2] + p_3x[n-3]$$



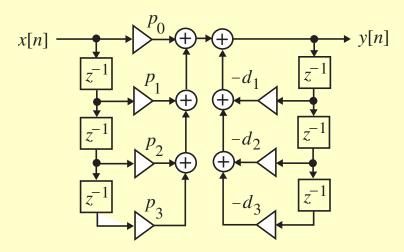
• The time-domain representation of $H_2(z)$ is given by

$$y[n] = w[n] - d_1y[n-1] - d_2y[n-2] - d_3y[n-3]$$

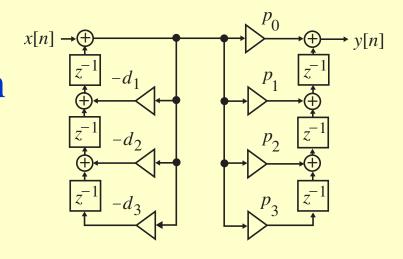
Realization of $H_2(z)$ follows from the above equation and is shown on the right



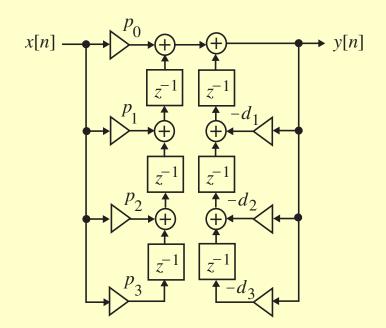
• A cascade of the two structures realizing $H_1(z)$ and $H_2(z)$ leads to the realization of H(z) shown below and is known as the **direct** form I structure

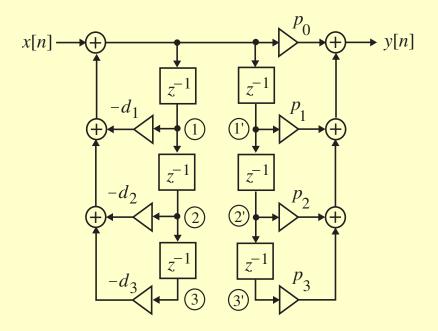


- Note: The direct form I structure is noncanonic as it employs 6 delays to realize a 3rd-order transfer function
- A transpose of the direct form I structure is shown on the right and is called the direct form I_t
 structure



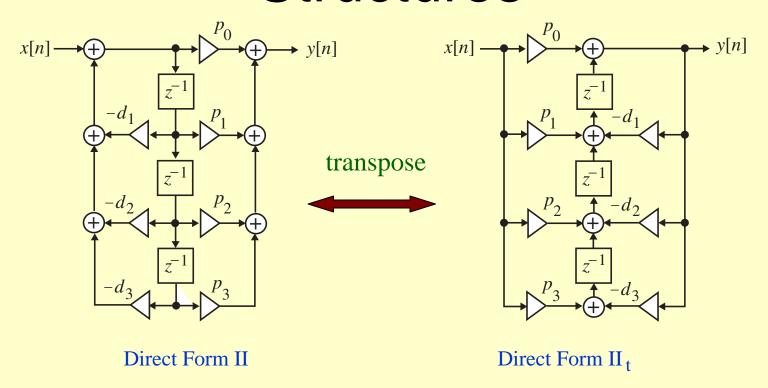
 Various other noncanonic direct form structures can be derived by simple block diagram manipulations as shown below





• Observe in the direct form structure shown below, the signal variable at nodes ① and ① are the same, and hence the two top delays can be shared

- Likewise, the signal variables at nodes (2) and (2) are the same, permitting the sharing of the middle two delays
- Following the same argument, the bottom two delays can be shared
- Sharing of all delays reduces the total number of delays to 3 resulting in a canonic realization shown on the next slide along with its transpose structure

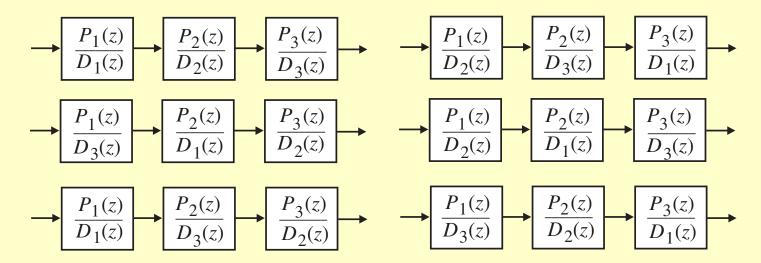


• Direct form realizations of an *N*-th order IIR transfer function should be evident

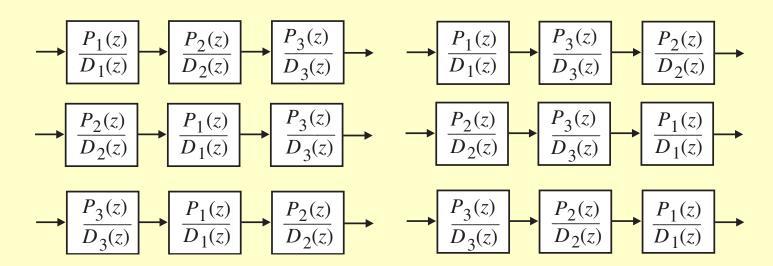
- By expressing the numerator and the denominator polynomials of the transfer function as a product of polynomials of lower degree, a digital filter can be realized as a cascade of low-order filter sections
- Consider, for example, H(z) = P(z)/D(z) expressed as

$$H(z) = \frac{P(z)}{D(z)} = \frac{P_1(z)P_2(z)P_3(z)}{D_1(z)D_2(z)D_3(z)}$$

 Examples of cascade realizations obtained by different pole-zero pairings are shown below



 Examples of cascade realizations obtained by different ordering of sections are shown below



• There are altogether a total of 36 different cascade realizations of

$$H(z) = \frac{P_1(z)P_2(z)P_2(z)}{D_1(z)D_2(z)D_3(z)}$$

based on pole-zero-pairings and ordering

 Due to finite wordlength effects, each such cascade realization behaves differently from others

• Usually, the polynomials are factored into a product of 1st-order and 2nd-order polynomials:

$$H(z) = p_0 \prod_{k} \left(\frac{1 + \beta_{1k} z^{-1} + \beta_{2k} z^{-2}}{1 + \alpha_{1k} z^{-1} + \alpha_{2k} z^{-2}} \right)$$

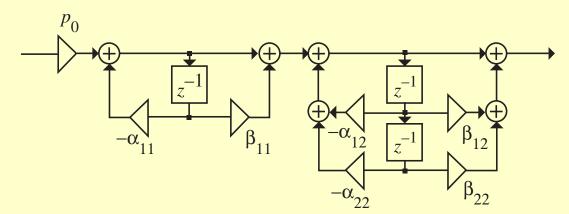
• In the above, for a first-order factor

$$\alpha_{2k} = \beta_{2k} = 0$$

Consider the 3rd-order transfer function

$$H(z) = p_0 \left(\frac{1 + \beta_{11} z^{-1}}{1 + \alpha_{11} z^{-1}} \right) \left(\frac{1 + \beta_{12} z^{-1} + \beta_{22} z^{-2}}{1 + \alpha_{12} z^{-1} + \alpha_{22} z^{-2}} \right)$$

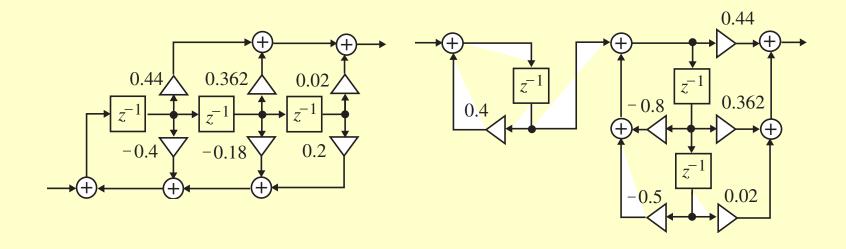
One possible realization is shown below



• Example - Direct form II and cascade form realizations of

$$H(z) = \frac{0.44z^{-1} + 0.362z^{-2} + 0.02z^{-3}}{1 + 0.4z^{-1} + 0.18z^{-2} - 0.2z^{-3}}$$
$$= \left(\frac{0.44 + 0.362z^{-1} + 0.02z^{-2}}{1 + 0.8z^{-1} + 0.5z^{-2}}\right) \left(\frac{z^{-1}}{1 - 0.4z^{-1}}\right)$$

are shown on the next slide



Direct form II

Cascade form

- A partial-fraction expansion of the transfer function in z^{-1} leads to the **parallel form I** structure
- Assuming simple poles, the transfer function H(z) can be expressed as

$$H(z) = \gamma_0 + \sum_{k} \left(\frac{\gamma_{0k} + \gamma_{1k} z^{-1}}{1 + \alpha_{1k} z^{-1} + \alpha_{2k} z^{-2}} \right)$$

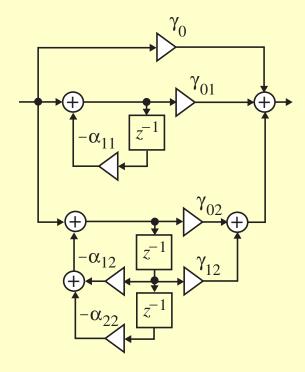
• In the above for a real pole $\alpha_{2k} = \gamma_{1k} = 0$

- A direct partial-fraction expansion of the transfer function in z leads to the parallel form II structure
- Assuming simple poles, the transfer function H(z) can be expressed as

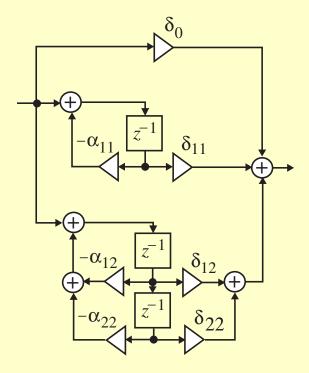
$$H(z) = \delta_0 + \sum_{k} \left(\frac{\delta_{1k} z^{-1} + \delta_{2k} z^{-2}}{1 + \alpha_{1k} z^{-1} + \alpha_{2k} z^{-2}} \right)$$

• In the above for a real pole $\alpha_{2k} = \delta_{2k} = 0$

• The two basic parallel realizations of a 3rdorder IIR transfer function are shown below



Parallel form I



Parallel form II

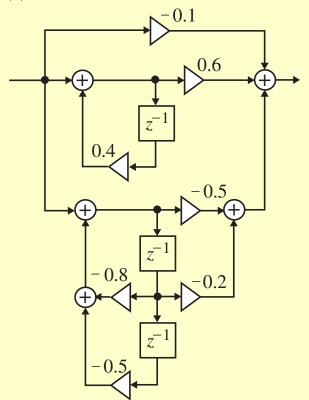
• Example - A partial-fraction expansion of

$$H(z) = \frac{0.44z^{-1} + 0.362z^{-2} + 0.02z^{-3}}{1 + 0.4z^{-1} + 0.18z^{-2} - 0.2z^{-3}}$$

in z^{-1} yields

$$H(z) = -0.1 + \frac{0.6}{1 - 0.4z^{-1}} + \frac{-0.5 - 0.2z^{-1}}{1 + 0.8z^{-1} + 0.5z^{-2}}$$

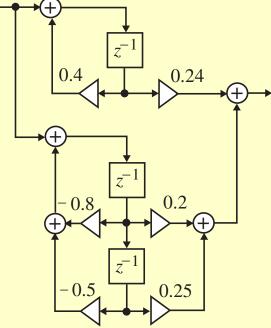
• The corresponding parallel form I realization is shown below



• Likewise, a partial-fraction expansion of H(z) in z yields

$$H(z) = \frac{0.24z^{-1}}{1 - 0.4z^{-1}} + \frac{0.2z^{-1} + 0.25z^{-2}}{1 + 0.8z^{-1} + 0.5z^{-2}}$$

 The corresponding parallel form II realization is shown on the right



- The cascade form requires the factorization of the transfer function which can be developed using the M-file zp2sos
- The statement sos = zp2sos(z,p,k) generates a matrix sos containing the coefficients of each 2nd-order section of the equivalent transfer function H(z) determined from its pole-zero form

• sos is an $L \times 6$ matrix of the form

$$\mathtt{sos} = \begin{bmatrix} \mathtt{p}_{01} & \mathtt{p}_{11} & \mathtt{p}_{21} & \mathtt{d}_{01} & \mathtt{d}_{11} & \mathtt{d}_{21} \\ \mathtt{p}_{02} & \mathtt{p}_{12} & \mathtt{p}_{22} & \mathtt{d}_{02} & \mathtt{d}_{12} & \mathtt{d}_{22} \\ \mathtt{p}_{0L} & \mathtt{p}_{1L} & \mathtt{p}_{2L} & \mathtt{d}_{0L} & \mathtt{d}_{1L} & \mathtt{d}_{2L} \end{bmatrix}$$

whose *i*-th row contains the coefficients $\{p_{i\ell}\}$ and $\{d_{i\ell}\}$, of the the numerator and denominator polynomials of the *i*-th 2nd-order section

- L denotes the number of sections
- The form of the overall transfer function is given by

$$H(z) = \prod_{i=1}^{L} H_i(z) = \prod_{i=1}^{L} \frac{p_{0i} + p_{1i}z^{-1} + p_{2i}z^{-2}}{d_{0i} + d_{1i}z^{-1} + d_{2i}z^{-2}}$$

• Program 6_1 can be used to factorize an FIR and an IIR transfer function

• Note: An FIR transfer function can be treated as an IIR transfer function with a constant numerator of unity and a denominator which is the polynomial describing the FIR transfer function

- Parallel forms I and II can be developed using the functions residuez and residue, respectively
- Program 6_2 uses these two functions