

1.
(1)

Let $g_a(t)$ be a continuous-time signal that is sampled uniformly at $t = nT$, generating the sequence $g[n]$ where

$$g[n] = g_a(nT), \quad -\infty < n < \infty, \quad (3.61)$$

with T being the *sampling period*. The reciprocal of T is called the *sampling frequency* F_T ; that is, $F_T = 1/T$. Now, the frequency-domain representation of $g_a(t)$ is given by its continuous-time Fourier transform (CTFT):

$$G_a(j\Omega) = \int_{-\infty}^{\infty} g_a(t) e^{-j\Omega t} dt, \quad (3.62)$$

whereas the frequency-domain representation of $g[n]$ is given by its discrete-time Fourier transform:

$$G(e^{j\omega}) = \sum_{n=-\infty}^{\infty} g[n] e^{-j\omega n}. \quad (3.63)$$

To establish the relations between these two different types of Fourier spectra, $G_a(j\Omega)$ and $G(e^{j\omega})$, we treat the sampling operation mathematically as a multiplication of the continuous-time signal $g_a(t)$ by a periodic impulse train $p(t)$:

$$p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT), \quad (3.64)$$

consisting of a train of ideal impulse functions⁷ with a period T , as indicated in Figure 3.14. The multiplication operation yields a continuous function of time for all values of t including $t = nT$:

$$g_p(t) = g_a(t)p(t) = \sum_{n=-\infty}^{\infty} g_a(nT)\delta(t - nT). \quad (3.65)$$

It should be noted that the signal $g_p(t)$ consists of a train of uniformly spaced impulses with the impulse at $t = nT$ weighted by the sampled value $g_a(nT)$ of $g_a(t)$ at that instant and is thus a distribution solely defined by its integration properties.

There are two different forms of the continuous-time Fourier transform $G_p(j\Omega)$ of $g_p(t)$. One form is obtained by taking the CTFT of Eq. (3.64), which results in a weighted sum of the continuous-time Fourier transforms of the shifted impulse functions $\delta(t - nT)$. The CTFT of $\delta(t - nT)$ is given by $e^{-j\Omega nT}$. Hence, from Eq. (3.65), $G_p(j\Omega)$ is given by

$$G_p(j\Omega) = \sum_{n=-\infty}^{\infty} g_a(nT) e^{-j\Omega nT}. \quad (3.66)$$

To derive the second form, we make use of the *Poisson's sum formula*, given by [Pap62]

$$\sum_{n=-\infty}^{\infty} \phi(t + nT) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \Phi(jk\Omega_T) e^{jk\Omega_T t}, \quad (3.67)$$

where $\Omega_T = 2\pi/T$ denotes the angular sampling frequency, and $\Phi(j\Omega)$ is the CTFT of the continuous-time function $\phi(t)$. A proof of the above identity is left as an exercise (Problem 3.56). For $t = 0$, Eq. (3.67) reduces to a simpler form

$$\sum_{n=-\infty}^{\infty} \phi(nT) = \frac{1}{T} \sum_{k=-\infty}^{\infty} \Phi(jk\Omega_T), \quad (3.68)$$

which is used to derive the second form of $G_p(j\Omega)$. Now from the frequency-shifting property of the CTFT, the CTFT of $g_a(t) e^{-j\Psi t}$ is given by $G_a(j(\Omega + \Psi))$. Substituting $\phi(t) = g_a(t) e^{-j\Psi t}$ in Eq. (3.68), we arrive at

$$\sum_{n=-\infty}^{\infty} g_a(nT) e^{-j\Psi nT} = \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a(j(k\Omega_T + \Psi)). \quad (3.69)$$

By replacing Ψ with Ω in the above equation, we arrive at the alternative form of the continuous-time Fourier transform of $g_p(t)$ given by

$$G_p(j\Omega) = \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a(j(\Omega + k\Omega_T)). \quad (3.70)$$

As can be seen from the above equation, $G_p(j\Omega)$ is a periodic function of frequency Ω consisting of a sum of shifted and scaled replicas of $G_a(j\Omega)$, shifted by integer multiples of Ω_T and scaled by $1/T$. The

We now establish the relation between the discrete-time Fourier transform $G(e^{j\omega})$ of the sequence $g[n]$ and the continuous-time Fourier transform $G_a(j\Omega)$ of the analog signal $g_a(t)$. If we compare Eq. (3.63) with Eq. (3.66) and make use of Eq. (3.61), we observe that

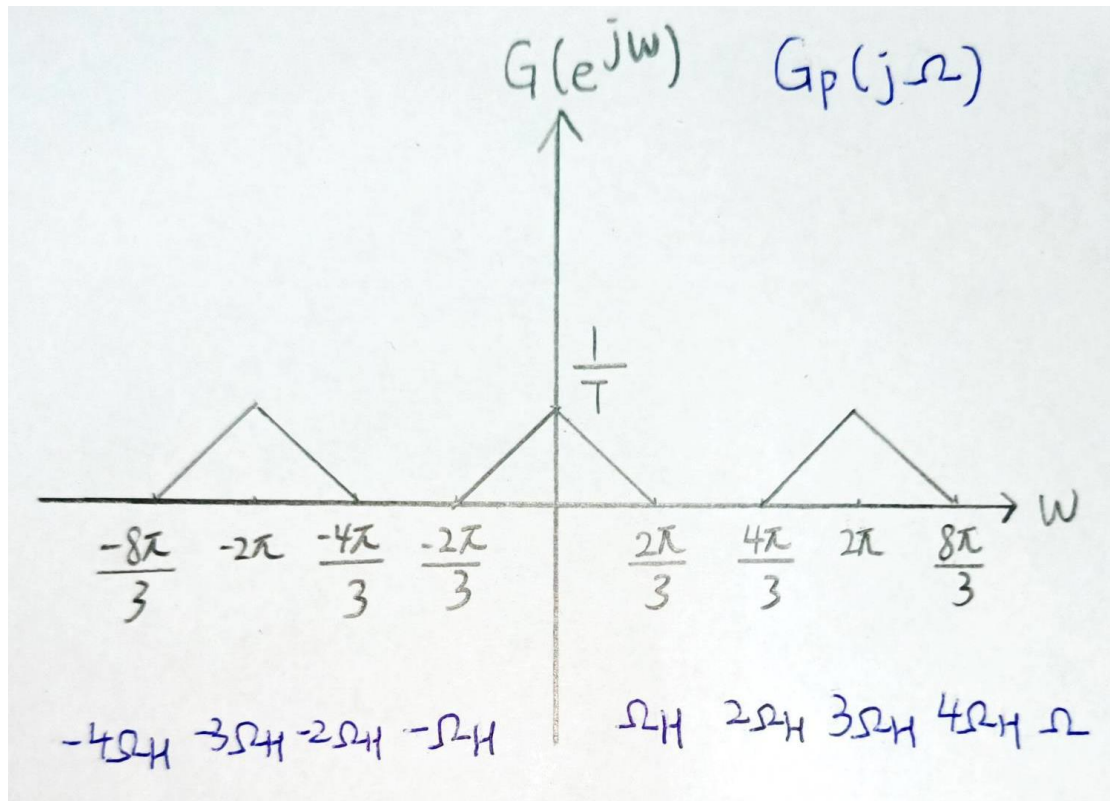
$$G(e^{j\omega}) = G_p(j\Omega) \Big|_{\Omega=\omega/T}, \quad (3.74a)$$

or equivalently,

$$G_p(j\Omega) = G(e^{j\omega}) \Big|_{\omega=\Omega T}. \quad (3.74b)$$

Therefore, from the above and Eq. (3.69), we arrive at the desired result given by

$$\begin{aligned} G(e^{j\omega}) &= \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a(j\Omega + jk\Omega_T) \Big|_{\Omega=\omega/T} \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a\left(j\frac{\omega}{T} + jk\Omega_T\right) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a\left(j\frac{\omega}{T} + j\frac{2\pi k}{T}\right), \end{aligned} \quad (3.75a)$$



2.

(1)

$$g[n] \xleftrightarrow{DTFT} G(e^{j\omega})$$

$$g[n - n_d] \xleftrightarrow{DTFT} G(e^{j\omega})e^{-jn_d\omega}$$

$$|G(e^{j\omega})e^{-jn_d\omega}| = |G(e^{j\omega})||e^{-jn_d\omega}| = |G(e^{j\omega})|$$

(2)

NO

3.

$$G(e^{j(\omega+2\pi)})$$

$$= \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a\left(j \frac{(\omega + 2\pi)}{T} + j \frac{2\pi k}{T}\right)$$

$$= \frac{1}{T} \sum_{k=-\infty}^{\infty} G_a\left(j \frac{\omega}{T} + j \frac{2\pi(k+1)}{T}\right)$$

$$= \frac{1}{T} \sum_{n=-\infty}^{\infty} G_a\left(j \frac{\omega}{T} + j \frac{2\pi n}{T}\right)$$

$$= G(e^{j\omega})$$

4.

(1)

$$\begin{aligned} H(e^{j\omega}) &= \sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k} \\ &= \sum_{k=0}^{N-1} h[k] e^{-j\omega k} \\ &= \frac{1}{N} (1 + e^{-j\omega} + e^{-2j\omega} + \dots + e^{-j(N-1)\omega}) \end{aligned}$$

sum of a geometric sequence of N terms

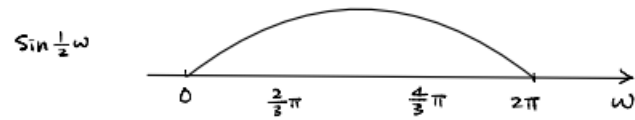
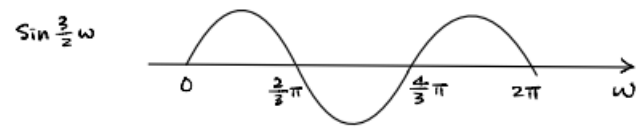
and common ratio $e^{-j\omega}$

$$H(e^{j\omega}) = \frac{1}{N} \frac{1 - e^{-jN\omega}}{1 - e^{-j\omega}}$$

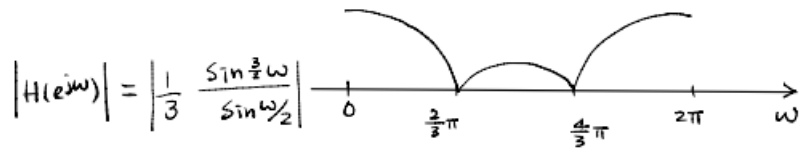
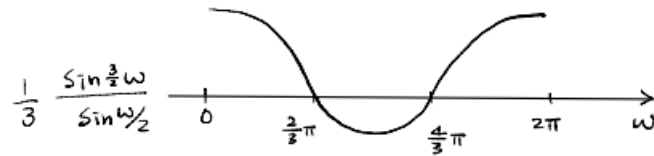
$$\begin{aligned} H(e^{j\omega}) &= \frac{1}{N} \frac{1 - e^{-jN\omega}}{1 - e^{-j\omega}} \\ &= \frac{1}{N} \frac{e^{j\frac{N}{2}\omega}}{e^{j\frac{\omega}{2}}} \frac{e^{j\frac{N}{2}\omega} - e^{-j\frac{N}{2}\omega}}{e^{j\frac{\omega}{2}} - e^{-j\frac{\omega}{2}}} \\ &= e^{-j\frac{1}{2}(N-1)\omega} \frac{1}{N} \frac{\sin \frac{N}{2}\omega}{\sin \frac{\omega}{2}} \end{aligned}$$

(2)

$$N = 3 \quad |H(e^{j\omega})| = \left| \frac{1}{N} \frac{\sin \frac{3}{2}\omega}{\sin \frac{1}{2}\omega} \right|$$

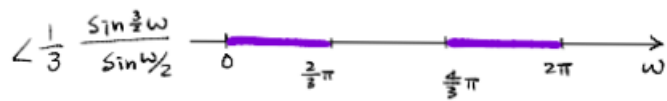


$$\lim_{\omega \rightarrow 0} \frac{1}{3} \frac{\sin \frac{3}{2}\omega}{\sin \frac{1}{2}\omega} = \frac{1}{3} \lim_{\omega \rightarrow 0} \frac{\frac{3}{2} \cos \frac{3}{2}\omega}{\frac{1}{2} \cos \frac{1}{2}\omega} = 1$$

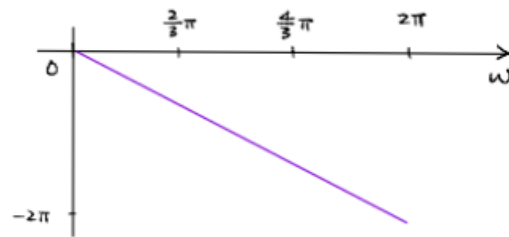


$$\angle H(e^{j\omega}) = \angle e^{-j\frac{1}{2}(N-1)\omega} + \angle \frac{1}{3} \frac{\sin \frac{3}{2}\omega}{\sin \frac{\omega}{2}}$$

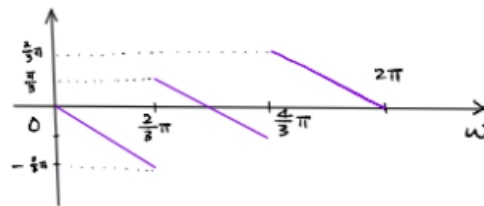
— π



$$\angle e^{-j\frac{1}{2}(3-1)\omega} = \angle e^{-j\omega} = -\omega$$



$$\angle H(e^{j\omega}) = \angle e^{-j\frac{1}{2}(N-1)\omega} + \angle \frac{1}{3} \frac{\sin \frac{3}{2}\omega}{\sin \frac{\omega}{2}}$$



(3)

$$\frac{d}{d\omega} \arg(H(e^{j\omega})) = -\frac{1}{2}(N-1)$$

(4)

YES

$$\begin{aligned} H(e^{j\omega}) &= \alpha + \beta e^{-j\omega} + \alpha e^{-j2\omega} \\ &= e^{-j\omega} (\alpha e^{j\omega} + \beta + \alpha e^{-j\omega}) \\ &= e^{-j\omega} (2\alpha \cos(\omega) + \beta) \end{aligned}$$

$$\frac{d}{d\omega} \arg(H(e^{j\omega})) = -1$$