## Linear System Theory Solution to Homework 9

- 1. (a) The eigenvalues of the system are -1, 1, 2.
  - (b) The controllability and observability matrices are full rank. Hence the system is controllable and observable. The system is unstable because two eigenvalues are on the right-half plane.
  - (c) Let  $\mathbf{b}_1 = [-1, -1, 2]^T$ ,  $\mathbf{b}_2 = [0, 1, 0]^T$ ,  $\lambda_1 = 1$ , and  $\lambda_2 = 2$ . Note that  $\lambda_1$  and  $\lambda_2$  are the unstable modes of the system. Then

$$\operatorname{rank} [\lambda_i \mathbf{I} - \mathbf{A}, \mathbf{b}_1] = 3, \quad i = 1, 2$$
  

$$\operatorname{rank} [\lambda_1 \mathbf{I} - \mathbf{A}, \mathbf{b}_2] = 2$$
  

$$\operatorname{rank} [\lambda_2 \mathbf{I} - \mathbf{A}, \mathbf{b}_2] = 3$$

By PBH rank test, the system can be stabilized by the first input only, but cannot be stabilized by the 2nd one, because the unstable mode  $\lambda_1$  cannot be changed by the 2nd input along.

(d) We can use the first input to do the state feedback; however, the mode  $\lambda = -1$  is not controllable with respect to the first input. Hence we transform the system into the modal form:

$$\Lambda = \mathbf{V}^{-1}\mathbf{A}\mathbf{V} = \operatorname{diag}(-1, 1, 2), \quad \tilde{\mathbf{b}}_1 = \mathbf{V}^{-1}\mathbf{b}_1$$

where  $\mathbf{V} = \begin{bmatrix} 1 & 0 & -1 \\ 1 & -1 & 0 \\ -1 & 1 & 1 \end{bmatrix}$  whose columns are eigenvectors of  $\mathbf{A}$  asso-

ciated with eigenvalues  $-1, \overline{1}, 2$ , respectively.

Notice that  $\tilde{\mathbf{b}}_1 = [0, 1, 1]^T$ . Let  $\mathbf{k}_m = [k_1, k_2, k_3]^T$ . Then

$$\mathbf{\Lambda} - \tilde{\mathbf{b}}_1 \mathbf{k}_m = \begin{bmatrix} -1 & 0 & 0 \\ -k_1 & 1 - k_2 & -k_3 \\ -k_1 & -k_2 & 2 - k_3 \end{bmatrix}$$

The characteristic polynomial of  $\mathbf{\Lambda} - \tilde{\mathbf{b}}_1 \mathbf{k}_m$  is

$$(s+1)((s-1+k_2)(s-2+k_3)-k_2k_3)=(s+1)^3$$

Then  $k_2 = -4$  and  $k_3 = 9$ .  $k_1$  can be assigned arbitrarily, and we choose  $k_1 = 0$ . Then  $\mathbf{K} = \begin{bmatrix} \mathbf{k}_m \mathbf{V}^{-1} \\ \mathbf{0} \end{bmatrix} = \begin{bmatrix} -4 & 9 & 5 \\ 0 & 0 & 0 \end{bmatrix}$ .

(e) From either output, the system is not detectable. Hence the observer-based controller with only one sensor feedback cannot stabilize the system.

2. For all  $\mathbf{x}(0) \in \mathbb{R}^n$ ,

$$\mathbf{y}(t) = \mathbf{C}e^{\mathbf{A}t}\mathbf{x}(0) \to \mathbf{0}, \text{ as } t \to \infty$$

Suppose that **A** is unstable and  $\lambda$  is one of the unstable eigenvalues of **A**. Let **v** be the eigenvector of **A** associated with  $\lambda$ . Then choose  $\mathbf{x}(0) = \mathbf{v}$  and we have  $\mathbf{y}(t) = e^{\lambda t} \mathbf{C} \mathbf{v} \to \mathbf{0}$ , as  $t \to \infty$ . Since  $\lambda$  is unstable, we have  $\mathbf{C} \mathbf{v} = \mathbf{0}$ . By PBH test, the system is unobservable, which is a contradiction.

- 3. The eigenvalues of the system are  $\mu$  (with multiplicity 2) and  $\pm j\omega$ . The corresponding eigenvectors are  $\mathbf{v}_1 = [1, 0, 0, 0]^T$  (for  $\mu$ ) and  $\mathbf{v}_2 = [0, 0, \pm j, 1]^T$  (for  $\pm j\omega$ ). For the system to be observable,  $\mathbf{C}\mathbf{v}_1 = \begin{bmatrix} c_{11} \\ c_{21} \end{bmatrix} \neq \begin{bmatrix} 0 \\ 0 \end{bmatrix}$ . This implies  $c_{11}$  and  $c_{21}$  cannot be zero simultaneously. On the other hand,  $\mathbf{C}\mathbf{v}_2 = \omega \begin{bmatrix} 1 \\ \pm j \end{bmatrix} \neq \mathbf{0}$ , because  $\omega > 0$ . Hence the system is observable if  $c_{11}$  and  $c_{21}$  are not zero simultaneously.
- 4. (a) Since **Q** is symmetric positive semidefinite with rank r, it has r positive eigenvalues  $\lambda_1, \dots, \lambda_r$  and n-r zero eigenvalues. Moreover, **Q** can be diagonalized by orthogonal matrix as follows:

$$\mathbf{Q} = \left[ egin{array}{cc} \mathbf{Q}_1 & \mathbf{Q}_2 \end{array} 
ight] \left[ egin{array}{cc} oldsymbol{\Lambda} & oldsymbol{0} \ oldsymbol{0} & oldsymbol{0} \end{array} 
ight] \left[ egin{array}{cc} \mathbf{Q}_1^T \ \mathbf{Q}_2^T \end{array} 
ight] = \mathbf{Q}_1 oldsymbol{\Lambda} \mathbf{Q}_1^T \end{array}$$

where  $\mathbf{\Lambda} = \operatorname{diag}(\lambda_1, \dots, \lambda_r), \ \mathbf{Q}_1 \in \mathbb{R}^{n \times r}, \ \operatorname{and} \ \mathbf{Q}_2 \in \mathbb{R}^{n \times (n-r)}.$  Define  $\mathbf{C} = \operatorname{diag}(\sqrt{\lambda_1}, \dots, \sqrt{\lambda_r})\mathbf{Q}_1^T$ . Then  $\mathbf{Q} = \mathbf{C}^T\mathbf{C}$ .

(b) 
$$(\mathbf{OC})^T \mathbf{OC} = \mathbf{C}^T \mathbf{O}^T \mathbf{OC} = \mathbf{C}^T \mathbf{C} = \mathbf{Q}$$

Hence **OC** is also a square root of **Q**.

(c) We have

$$\mathbf{C}_1^T \mathbf{C}_1 = \mathbf{C}_2^T \mathbf{C}_2 \tag{1}$$

Because  $\mathbf{C}_1^T \in \mathbb{R}^{n \times r}$  has r linearly independent columns, it has a left inverse  $(\mathbf{C}_1\mathbf{C}_1^T)^{-1}\mathbf{C}_1$ . Multiplying both sides of (1) by  $(\mathbf{C}_1\mathbf{C}_1^T)^{-1}\mathbf{C}_1$  leads to

$$\mathbf{C}_1 = \left(\mathbf{C}_1 \mathbf{C}_1^T\right)^{-1} \mathbf{C}_1 \mathbf{C}_2^T \mathbf{C}_2 = \mathbf{O} \mathbf{C}_2$$

where  $\mathbf{O} = (\mathbf{C}_1 \mathbf{C}_1^T)^{-1} \mathbf{C}_1 \mathbf{C}_2^T$ . Then we need to show that  $\mathbf{O} \in \mathbb{R}^{r \times r}$  is orthogonal, i.e  $\mathbf{O}^T \mathbf{O} = \mathbf{I}$ . Notice that (1) becomes

$$\mathbf{C}_1^T \mathbf{C}_1 = \left(\mathbf{O} \mathbf{C}_2\right)^T \mathbf{O} \mathbf{C}_2 = \mathbf{C}_2^T \mathbf{O}^T \mathbf{O} \mathbf{C}_2 = \mathbf{C}_2^T \mathbf{C}_2 \tag{2}$$

Since  $\mathbf{C}_2^T$  also has full column rank, it has a left inverse  $\mathbf{C}_2^{\dagger} = \left(\mathbf{C}_2\mathbf{C}_2^T\right)^{-1}\mathbf{C}_2$ . Multiplying  $\mathbf{C}_2^{\dagger}$  from the left of the third equality of (2) and  $\mathbf{C}_2^{\dagger T}$  from the right yields  $\mathbf{O}^T\mathbf{O} = \mathbf{I}$ . Therefore,  $\mathbf{O}$  is an orthogonal matrix.

(d) There is an orthogonal matrix O so that  $C = OC_0$ . Then

$$\left[\begin{array}{c} \mathbf{C} \\ \lambda \mathbf{I} - \mathbf{A} \end{array}\right] = \left[\begin{array}{c} \mathbf{O} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{array}\right] \left[\begin{array}{c} \mathbf{C}_0 \\ \lambda \mathbf{I} - \mathbf{A} \end{array}\right]$$

Since O is invertible,

$$\operatorname{rank}\left[egin{array}{c} \mathbf{C} \ \lambda \mathbf{I} - \mathbf{A} \end{array}
ight] = \operatorname{rank}\left[egin{array}{c} \mathbf{C}_0 \ \lambda \mathbf{I} - \mathbf{A} \end{array}
ight]$$

Thus (C, A) is observable.

5. Let  $C = \sqrt{Q}$  and y = Cx. The cost function can then be expressed as

$$J = \int_0^\infty (\|\mathbf{y}(t)\|^2 + \mathbf{u}^T(t)\mathbf{R}\mathbf{u}(t))dt$$

Let  $\mathbf{u}^*$  denote the optimal control that minimizes J and let  $\mathbf{x}^*$  and  $\mathbf{y}^*$  be the corresponding state and output under the optimal control. Then  $\mathbf{u}^*(t) = -\mathbf{K}\mathbf{x}^*(t)$ , and the optimal cost  $J^* < \infty$  since  $(\mathbf{A}, \mathbf{B})$  is assumed controllable. It then follows that  $\mathbf{y}^*(t) \to \mathbf{0}$  and  $\mathbf{u}^*(t) \to \mathbf{0}$  as  $t \to \infty$ , since  $\mathbf{R} > 0$ . Since both  $\mathbf{u}^*$  and  $\mathbf{y}^*$  are solutions of a linear system, each component of which is a linear combination of exponential functions (or products of polynomial and exponential functions). The derivatives, of all orders, of  $\mathbf{y}^*$  and  $\mathbf{u}^*$  must also converge to zero. It then follows that

$$\mathbf{C}\mathbf{x}^* = \mathbf{y}^* \to \mathbf{0} \text{ as } t \to \infty$$

$$\mathbf{C}\mathbf{A}\mathbf{x}^* = \dot{\mathbf{y}}^*(t) - \mathbf{C}\mathbf{B}\mathbf{u}^*(t) \to \mathbf{0} \text{ as } t \to \infty$$
:

Then we conclude that

$$\begin{bmatrix} \mathbf{C} \\ \mathbf{CA} \\ \vdots \\ \mathbf{CA}^{n-1} \end{bmatrix} \mathbf{x}^*(t) \to \mathbf{0} \text{ as } t \to \infty$$

By assumption, the observability matrix has full column rank, it follows that  $\mathbf{x}^*(t) \to \mathbf{0}$  as  $t \to \infty$  and the closed-loop system is stable.