Circuits and Transforms

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 $\begin{tabular}{ll} Abstract — This manual provides a simple introduction to Transforms \end{tabular}$

1 Definitions

1. The unit step function is

$$u(t) = \begin{cases} 1 & t > 0 \\ \frac{1}{2} & t = 0 \\ 0 & t < 0 \end{cases}$$
 (1.1)

2. The Laplace transform of g(t) is defined as

$$G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt$$
 (1.2)

2 Laplace Transform

1. In the circuit, the switch S is connected to position P for a long time so that the charge on the capacitor becomes $q_1 \mu C$. Then S is switched to position Q. After a long time, the charge on the capacitor is $q_2 \mu C$.

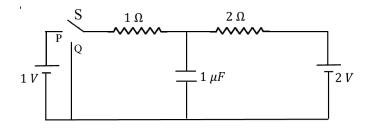
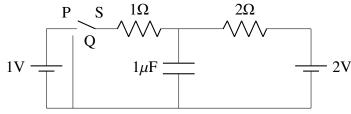


Fig. 2.1

2. Draw the circuit using latex-tikz.

Solution:

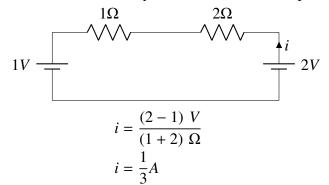


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3. Find q_1 .

Solution:

Current through the capacitor is 0 at steady state. So, current only flows in the outer loop.



Potential difference across the capacitor,

$$V_C = 2 - 2 \times \frac{1}{3} = \frac{4}{3}V$$

$$q_1 = CV_C$$

$$q_1 = 1\mu F \times \frac{4}{3}V$$

$$q_1 = \frac{4}{3}\mu C$$

4. Show that the Laplace transform of u(t) is $\frac{1}{s}$ and find the ROC.

Solution:

By the definition of Laplace transform

$$u(t) \xrightarrow{\mathcal{L}} \int_{-\infty}^{\infty} u(t)e^{-st}dt = \int_{0}^{\infty} e^{-st}dt$$
$$= \frac{e^{-st}}{-s} \Big|_{0}^{\infty} = \frac{1}{s}$$

 $\lim_{s\to\infty} e^{-st} = 0$ only when Re(s) > 0. The ROC is Re(s) > 0. 5. Show that

$$e^{-at}u(t) \stackrel{\mathcal{H}}{\longleftrightarrow} L\frac{1}{s+a}, \quad a > 0$$
 (2.1)

and find the ROC.

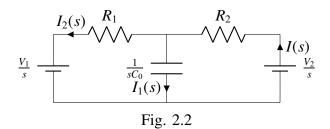
Solution:

By the definition of Laplace transform

$$e^{-at}u(t) \xrightarrow{\mathcal{L}} \int_{-\infty}^{\infty} e^{-at}u(t)e^{-st}dt$$
$$= \int_{0}^{\infty} e^{-(s+a)t}dt$$
$$= \frac{e^{-(s+a)t}}{-(s+a)}\Big|_{0}^{\infty} = \frac{1}{s+a}$$

 $\lim_{s\to\infty} e^{-(s+a)t} = 0$ only when Re(s+a) > 0. The ROC is Re(s+a) > 0 or Re(s) > -a.

6. Now consider the following resistive circuit transformed from Fig. 2.1 where



$$u(t) \stackrel{\mathcal{H}}{\longleftrightarrow} LV_1(s)$$
 (2.2)

$$2u(t) \stackrel{\mathcal{H}}{\longleftrightarrow} LV_2(s)$$
 (2.3)

Find the voltage across the capacitor $V_{C_0}(s)$. Solution:

Using the Transform from time domain to s-domain

$$V_1(s) = \mathcal{L}(u(t)) = \frac{V_1}{s}$$
$$V_2(s) = \mathcal{L}(2u(t)) = \frac{V_2}{s}$$

Potential across the capacitor is $V_{C_0}(s)$ and the assuming bottom is grounded.

Applying KCL at the upper junction, we have,

$$\begin{split} &\frac{V_{C_0}(s)-V_1(s)}{R_1}+\frac{V_{C_0}(s)-V_2(s)}{R_2}+\frac{V_{C_0}(s)}{\frac{1}{sC_0}}=0\\ &V_{C_0}(s)\left(\frac{1}{R_1}+\frac{1}{R_2}+sC_0\right)=\frac{V_1(s)}{R_1}+\frac{V_2(s)}{R_2}\\ &V_{C_0}(s)=\frac{V_1(s)R_2+V_2(s)R_1}{R_1+R_2+sC_0R_1R_2}\\ &V_{C_0}(s)=\frac{R_2+2R_1}{s(R_1+R_2+sC_0R_1R_2)} \end{split}$$

after substitution.

7. Find $v_{C_0}(t)$. Plot using python.

Solution:

Factoring $V_{C_0}(s)$, we have

$$V_{C_0}(s) = \left(\frac{1}{R_1 + R_2}\right) \left(\frac{(R_2 + 2R_1)}{s} - \frac{R_1 R_2 C_0 (R_2 + 2R_1)}{R_1 + R_2 + s C_0 R_1 R_2}\right)$$

$$= \left(\frac{R_2 + 2R_1}{R_1 + R_2}\right) \left(\frac{1}{s} - \frac{R_1 R_2 C_0}{R_1 + R_2 + s C_0 R_1 R_2}\right)$$

$$= \left(\frac{R_2 + 2R_1}{R_1 + R_2}\right) \left(\frac{1}{s} - \frac{1}{\frac{1}{R_2 C_0} + \frac{1}{R_1 C_0} + s}\right)$$

Applying inverse Laplace transform,

$$v_{C_0}(t) = \frac{R_2 + 2R_1}{R_1 + R_2} \left(u(t) - e^{-\left(\frac{1}{R_1 C_0} + \frac{1}{R_2 C_0}\right)t} u(t) \right)$$
(2.4)

Using the values $R_1 = 1 \Omega$, $R_2 = 2 \Omega$, $C_0 = 1 \mu F$, we get

$$v_{C_0}(t) = \frac{4}{3}u(t)\left(1 - e^{-1.5 \times 10^6 t}\right)$$

The following code yields $v_{C_0}(t)$.

wget https://github.com/Sigma1084/EE3900/blob/master/cktsig/code/Ex2_plotVt.py

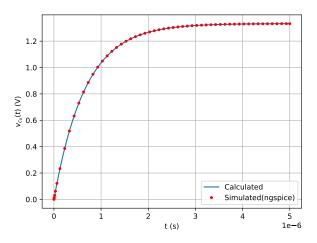


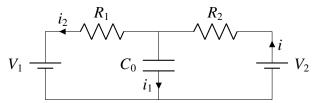
Fig. 2.3

8. Verify your result using ngspice. The following code yields $v_{C_0}(t)$. using ngspice

wget https://github.com/Sigma1084/EE3900/ blob/master/cktsig/code/Ex2.spice

9. Obtain Fig. 2.2 using the equivalent differential equation.

Solution:



Using KVL on the left and right loops, we get

$$i_2(t) R_1 + V_1 - \int_0^t \frac{i_1(t)}{C_0} dt = 0$$
 (2.5)

$$-i(t) R_2 + V_2 - \int_0^t \frac{i_1(t)}{C_0} dt = 0$$
 (2.6)

Taking the Laplace Transform after multiplying u(t) in both equations,

$$\mathcal{L}\left(R_1 i_2(t) u(t) + V_1 u(t) - u(t) \int_0^t \frac{i_1}{C_0} dt\right) = 0$$

$$\mathcal{L}\left(-R_2 i(t) u(t) + V_2 u(t) - u(t) \int_0^t \frac{i_1}{C_0} dt\right) = 0$$

Suppose $i(t)u(t) \xrightarrow{\mathcal{L}} I(s)$ and $Vu(t) \xrightarrow{\mathcal{L}} \frac{V}{s}$

$$\implies \mathcal{L}\left(u(t)\int_0^t \frac{i_1}{C_0}dt\right) = \frac{1}{sC_0}\mathcal{L}\left(u(t)i_1\right) = \frac{I_1(s)}{sC_0}$$

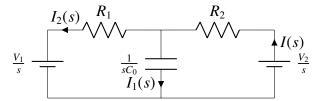
(Using the properties of Laplace Transform)

The above equations after transform

$$R_1 I_2(s) + \frac{V_1}{s} - \frac{I_1(s)}{sC_0} = 0 {(2.7)}$$

$$-R_2I(s) + \frac{V_2}{s} - \frac{I_1(s)}{sC_0} = 0$$
 (2.8)

Now after applying the transformation, from time domain to s-domain, we replace the capacitor with an equivalent resistor of resistence $\frac{1}{sC_0}$ and replace I by I(s), V by $\frac{V}{s}$ The equivalent s-domain circuit is now,



3 Initial Conditions

1. Find q_2 in Fig. 2.1

Solution:

Current through the capacitor is 0 in steady state. Since the current flows only in the outer loop, let the current in the outer loop be I.

$$I = \frac{2V}{(1+2)\Omega} = \frac{2}{3}A$$

So, the potential difference across the capacitor

$$V_{C_0} = 2 - \left(\frac{2}{3} \times 2\right) = \frac{2}{3}V$$

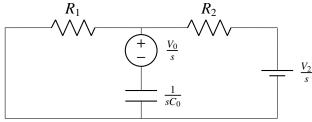
$$q_2 = CV_{C_0}$$

$$q_2 = 1\mu C \times \frac{2}{3}V$$

$$q_2 = \frac{2}{3}\mu C$$

2. Draw the equivalent s-domain resistive circuit when S is switched to position Q. Use variables R_1, R_2, C_0 for the passive elements. Use latex-tikz.

Solution:



3. $V_{C_0}(s) = ?$

Solution:

Applying KCL at the upper junction and assuming the bottom is grounded, we have,

$$\begin{split} &\frac{V_{C_0}(s)}{R_1} + \frac{V_{C_0}(s) - \frac{V_2}{s}}{R_2} + \frac{V_{C_0}(s) - \frac{V_0}{s}}{\frac{1}{sC_0}} = 0\\ &V_{C_0}(s) \left(\frac{1}{R_1} + \frac{1}{R_2} + sC_0\right) = \frac{V_2}{sR_2} + V_0C_0\\ &V_{C_0}(s) = \frac{\frac{V_2}{sR_2} + V_0C_0}{\frac{1}{R_1} + \frac{1}{R_2} + sC_0}\\ &V_{C_0}(s) = V_0 \left(\frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2}\right) + s}\right)\\ &+ \frac{V_2}{R_2 \left(\frac{1}{R_1} + \frac{1}{R_2}\right)} \left(\frac{1}{s} - \frac{1}{\frac{1}{C_0} \left(\frac{1}{R_1} + \frac{1}{R_2}\right) + s}\right) \end{split}$$

4. $v_{C_0}(t) = ?$ Plot using python.

Solution:

Applying inverse Laplace transform to $V_{C_0}(s)$, we have,

$$\begin{split} v_{C_0}(t) &= V_0 e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}} u(t) \\ &+ \frac{V_2}{R_2\left(\frac{1}{R_1} + \frac{1}{R_2}\right)} \left(1 - e^{-\left(\frac{1}{R_1} + \frac{1}{R_2}\right)\frac{t}{C_0}}\right) u(t) \end{split}$$

Substituting the values $R_1 = 1, R_2 = 2, C_0 = 1\mu F, V_0 = \frac{4}{3}V, V_2 = 2V$

$$v_{C_0}(t) = \frac{2}{3} \left(1 + e^{-\left(1.5 \times 10^6\right)t} \right) u(t)$$
 (3.1)

The following code yields $v_{C_0}(t)$.

wget https://github.com/Sigma1084/EE3900/blob/master/cktsig/code/Ex3_plotVt.py

5. Verify your result using ngspice. The following code yields $v_{C_0}(t)$. using ngspice

wget https://github.com/Sigma1084/EE3900/blob/master/cktsig/code/Ex3.spice

6. Find $v_{C_0}(0-)$, $v_{C_0}(0+)$ and $v_{C_0}(\infty)$.

Solution:

Using the initial conditions, we have,

$$v_{C_0}(0-) = \frac{q_1}{C_0} = \frac{4}{3}V$$

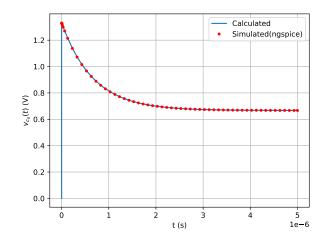


Fig. 3.1

Using (3.1), we have,

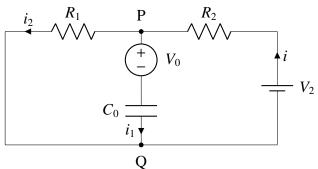
$$v_{C_0}(0+) = \frac{2}{3} \left(1 + e^{-\left(1.5 \times 10^6\right)0} \right) u(0^+) = \frac{4}{3} V$$

$$v_{C_0}(\infty) = \frac{2}{3} \left(1 + e^{-\left(1.5 \times 10^6\right)\infty} \right) u(\infty) = \frac{2}{3} V$$

7. Obtain the Fig. in problem 3.2 using the equivalent differential equation.

Solution:

Constructing the circuit,



Using KVL on both the left and outer loops, we have

$$-i_2(t) R_1 + V_0 + \int_0^t \frac{i_1(t)}{C_0} dt = 0$$
$$-i(t) R_2 + V_2 - i_2(t) R_1 = 0$$

Taking the Laplace Transform after multiplying u(t) in both equations,

$$\mathcal{L}\left(-R_1 i_2(t) u(t) + V_0 u(t) + u(t) \int_0^t \frac{i_1}{C_0} dt\right) = 0$$

$$\mathcal{L}\left(-R_2 i(t) u(t) + V_2 u(t) - R_1 i_2(t) u(t)\right) = 0$$

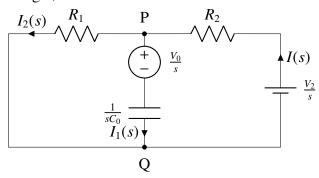
Suppose $i(t)u(t) \xrightarrow{\mathcal{L}} I(s)$ and $Vu(t) \xrightarrow{\mathcal{L}} \frac{V}{s}$

$$\implies \mathcal{L}\left(u(t)\int_0^t \frac{i_1}{C_0}dt\right) = \frac{1}{sC_0}\mathcal{L}\left(u(t)i_1\right) = \frac{I_1(s)}{sC_0}$$

(Using the properties of Laplace Transform) The above equations after transform

$$R_1 I_2(s) + \frac{V_0}{s} + \frac{I_1(s)}{sC_0} = 0$$
$$-R_2 I(s) + \frac{V_2}{s} - R_1 I_2(s) = 0$$

Now after applying the transformation, from time domain to s-domain, we replace the capacitor with an equivalent resistor of resistence $\frac{1}{sC_0}$ and replace I by I(s), V by $\frac{V}{s}$, and hence we get,



4 BILINEAR TRANSFORM

1. In Fig. 2.1, Consider the case when S is switched to Q right in the beginning. Formulate the differential equation.

Solution:

Special Case of (2.5), putting $V_1 = 0$, we get

$$i_2(t) R_1 - \int_0^t \frac{i_1(t)}{C_0} dt = 0$$
 (4.1)

$$-i(t) R_2 + V_2 - \int_0^t \frac{i_1(t)}{C_0} dt = 0$$
 (4.2)

The above equations after transform

$$R_1 I_2(s) - \frac{I_1(s)}{sC_0} = 0 (4.3)$$

$$-R_2I(s) + \frac{V_2}{s} - \frac{I_1(s)}{sC_0} = 0 {(4.4)}$$

- 2. Find H(s) considering the outure voltage at the capacitor.
- 3. Plot H(s). What kind of filter is it?

4. Using trapezoidal rule for integration, formulate the difference equation by considering

$$y(n) = y(t)|_{t=n}$$
 (4.5)

- 5. Find H(z).
- 6. How can you obtain H(z) from H(s)?