

Transmission Line and Voltage Standing Wave Ratio

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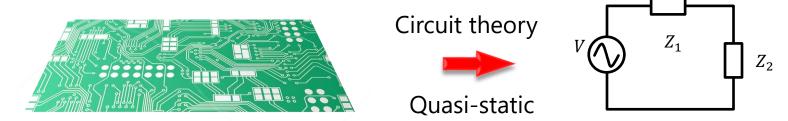
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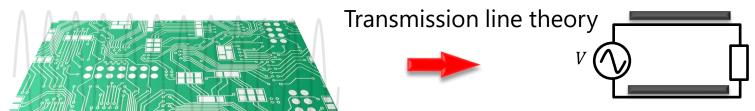


Circuit and transmission line theory

Low frequency: Wavelength is much larger than the circuit size



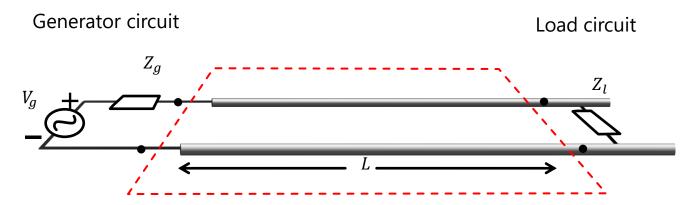
High frequency: Wavelength is much smaller than circuit size



Different phases at different locations in the circuit

White in circuit theory, the **length** of a transmission line is important to determine voltage and current

Transmission lines



A transmission line is a two-port network connecting a generator circuit at the sending end to a load at the receiving end.

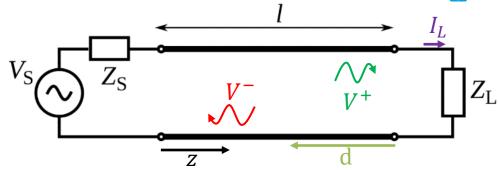
- It is a structure to carry alternating current and electromagnetic waves from one place to another
- E.g.: connecting radio transmitters and receivers with their antennas, distributing cable television signals, and computer network connections, etc.

Outline

- 1. Basics of travelling wave
 - 1. Current and Voltage waves in Transmission Lines
 - 2. Introduction on Characteristic impedance, load and mismatch terms
- Background information
 - 1. Intro to Transmission Line (TL)
 - 2. Intro to Standing wave
- 3. How can we measure mismatch and find load?
 - 1. Voltage Standing Wave Ratio
 - Load Calculation



1. Basics of travelling wave



- One of the mathematical ways to represent waves are the complex exponentials. In our notation $e^{-\gamma z}$ denotes forward propagating wave.
- $\gamma = \alpha + j\beta$ where α quantifies loss terms while β determines propagation properties
- Waves in TL can be decomposed as incident (forward going), and reflected (backward going)
- TL's have a characteristic impedance (Z_{ch}) determined by dimensions of the TL and the dielectric material (independent of length). The voltage and current waves within the TL scales with Z_{ch}

$$V(z) = V^{+}e^{-\gamma z} + V^{-}e^{+\gamma z}$$
 Voltage wave

We can write voltage at z=l or d=0

$$V_{L} = V^{+}e^{-\gamma l} + V^{-}e^{+\gamma l}$$

$$V_{L}^{+} \qquad V_{L}^{-}$$

$$V(d) = V_L^+ e^{+\gamma d} + V_L^- e^{-\gamma d}$$

$$I_L = I_L^+ + I_L^- = \frac{V_L^+}{Z_{ch}} - \frac{V_L^-}{Z_{ch}}$$

$$I(d) = rac{V_L^+ e^{+\gamma d} + V_L^- e^{-\gamma d}}{Z_{ch}}$$
 Current wave

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Introduction on Characteristic impedance, load and mismatch terms

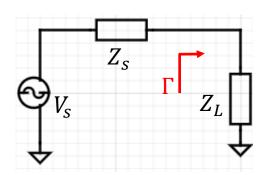
Resistance against applied voltage (R)

Storage due to change in applied voltage (X)



Impedance or Load Z_L =(R+jX)

- Frequency dependent due to X
- Complex number in general



Max. Power Transfer

$$Z_S = Z_L^*$$

$$P_{Load} = P_{gen}$$

Mismatch

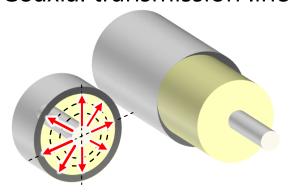
$$Z_S \neq Z_L$$

$$\Gamma = \frac{Z_L - Z_S}{Z_L + Z_S} \qquad P_{Load} = (1 - |\Gamma|^2) P_{gen}$$

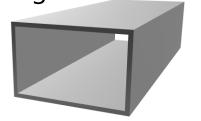
$$P_{Load} = (1 - |\Gamma|^2) P_{gen}$$

2. Background Information

Coaxial transmission line



Rectangular waveguide



Attenuation constant

$$\gamma = \alpha + j\beta \rightarrow$$
 Propagation constant

Transmission Line

 $\neq k_0$ due to dielectric

 $\beta \neq k_0$ due to cutoff relation

$$\beta = w\sqrt{\mu_0\epsilon_0\mu_r\epsilon_r}$$
$$\lambda_g = \lambda_0/\sqrt{\mu_r\epsilon_r}$$

$$\lambda_g = \lambda_0 / \sqrt{\mu_r \epsilon_r}$$

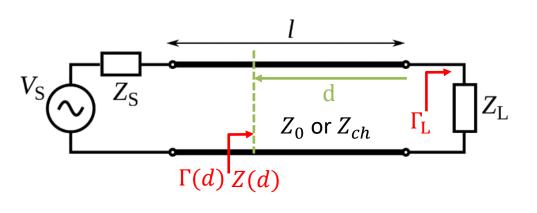
Phase velocity

$$u_p = \frac{w}{\beta}$$

$$\beta = k_0 \sqrt{1 - \left(\frac{f_c}{f}\right)^2}$$

 $k_0 = \frac{2\pi}{\lambda_0}$ wave number, f_c cut-off

$$\lambda_w = \frac{\lambda_0}{\sqrt{1 - \left(\frac{\lambda_0}{2a}\right)^2}}$$



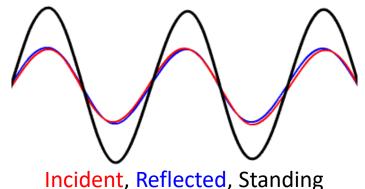
$$\Gamma(d) = \frac{Z(d) - Z_0}{Z(d) + Z_0} = \Gamma_L e^{-2\gamma d}$$

$$\gamma = \alpha + i\beta$$

$$Z(d) = Z_0 \frac{Z_L + Z_0 \tanh(\gamma d)}{Z_0 + Z_L \tanh(\gamma d)} \xrightarrow{\text{lossless}} Z_{lossless}(d) = Z_0 \frac{Z_L + jZ_0 \tan(\beta d)}{Z_0 + jZ_L \tan(\beta d)}$$



Standing Wave

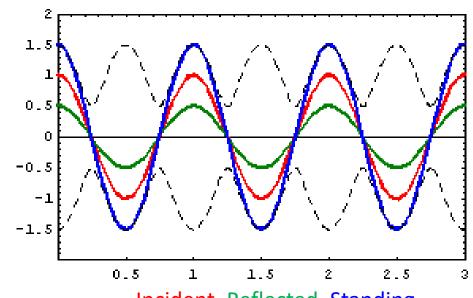


A measure of how much standing waves are present is the Voltage standing wave ratio (VSWR)

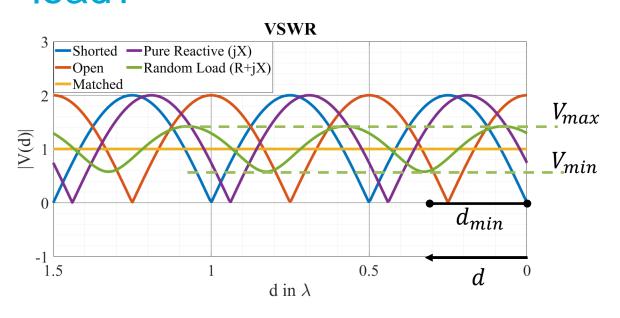
$$VSWR = \frac{V_{max}}{V_{min}} = \frac{1 + |\Gamma|}{1 - |\Gamma|}$$



If they don't have the same amplitude Oscillate between a max and a min



3. How can we measure mismatch and find load?



$$VSWR(S) = \frac{V_{max}}{V_{min}} = \frac{1+|\Gamma|}{1-|\Gamma|}$$

$$1 \leq S < \infty$$

$$\downarrow \qquad \qquad \downarrow$$
 $|\Gamma_L = 0| \qquad \qquad |\Gamma_L = 1|$
Matched
Load
• $Z_L = 0$

•
$$Z_L = \infty$$
 Open Load

•
$$Z_L = jX$$

Purely Reactive Load

Shorted Load



How to calculate load by VSWR graph?

Assuming lossless TL we can express reflection coeff. As;

$$Z_L = Z_0 \frac{1 - \Gamma_L}{1 + \Gamma_L}$$
 and $\Gamma_L = |\Gamma_L| e^{j\theta_L}$

1. Determine $|\Gamma_L|$ $S = \frac{1 + |\Gamma_L|}{1 - |\Gamma_L|} \quad \text{yields} \quad |\Gamma_L| = \frac{S - 1}{S + 1}$

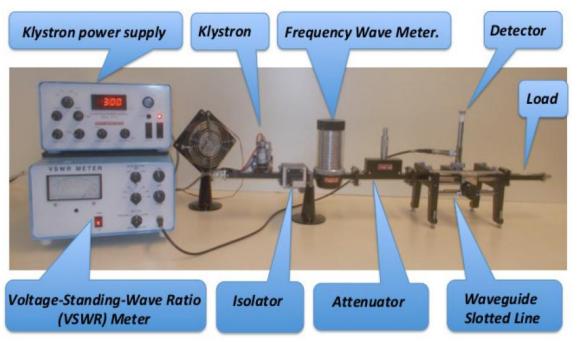
2. Determine
$$\theta_L$$

At $d = d_{min}$ voltage reflected and voltage incident has π phase difference

$$\frac{V_{reflected}}{V_{incident}} = \frac{V_L^- e^{-j\beta d_{min}}}{V_L^+ e^{j\beta d_{min}}} = \Gamma_L e^{-j2\beta d_{min}} = |\Gamma_L| e^{j\theta_L} e^{-j2\beta d_{min}}$$

Phase term should be $-\pi$. So $\theta_L = 2\beta d_{min} - \pi$

Experiment Setup



Summary of experiment procedure;

- Determine operating frequency of the source using Frequency wave meter
- 2. Calibrate VSWR meter by sliding probe on slotted line
- Measure VSWR values of different loads (open, matched load, antenna)
- Measure VSWR values of short circuit under different piston locations

