

# Chapter 2: Radio Wave Propagation Fundamentals

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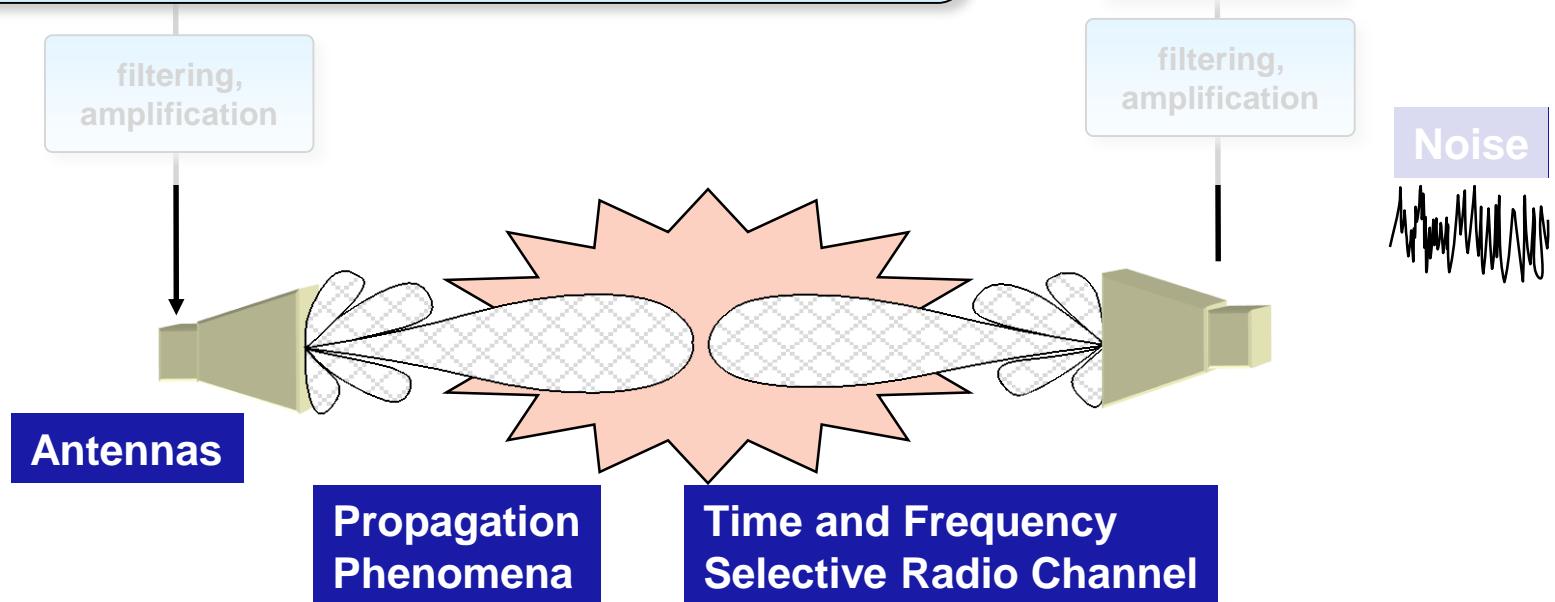
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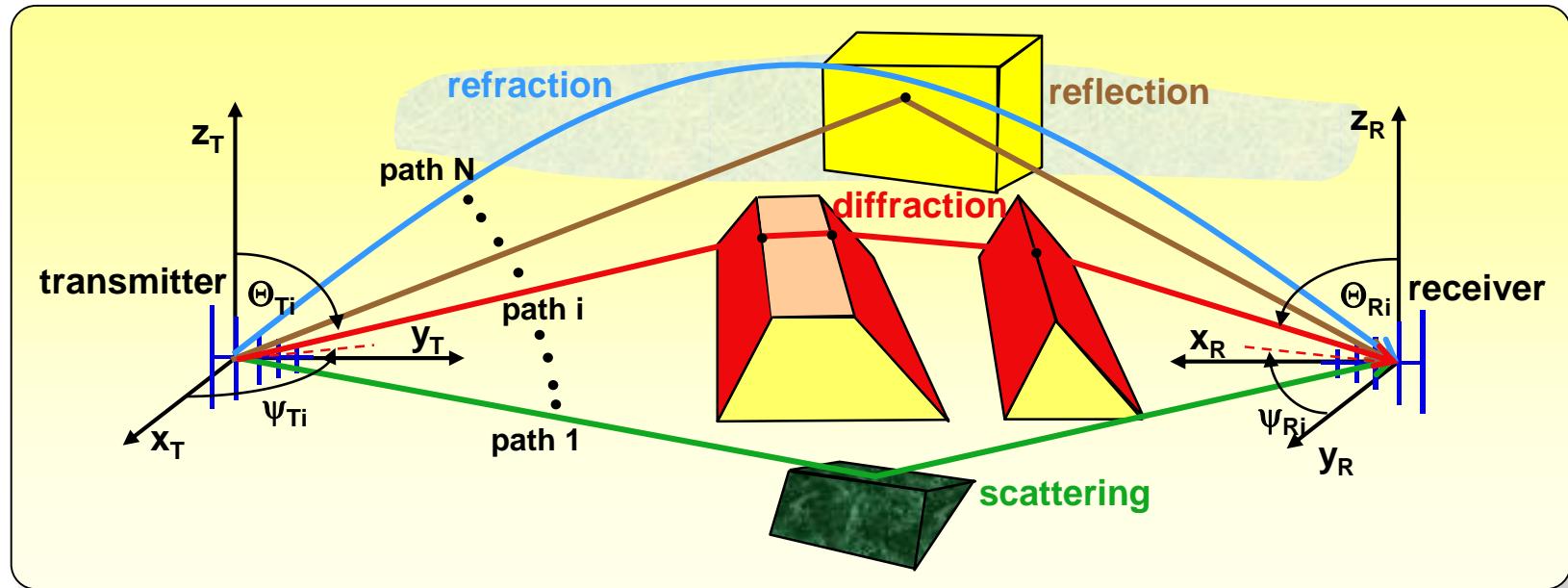
# Scope of the (Today's) Lecture

Effects during wireless transmission of signals:

- **physical phenomena that influence the propagation of electromagnetic waves**
- **no statistical description of those effects in terms of modulated signals**



# Propagation Phenomena



free space propagation:

- line of sight
- no multipath

reflection:

- plane wave reflection
- Fresnel coefficients

diffraction:

- knife edge diffraction

scattering:

- rough surface scattering
- volume scattering

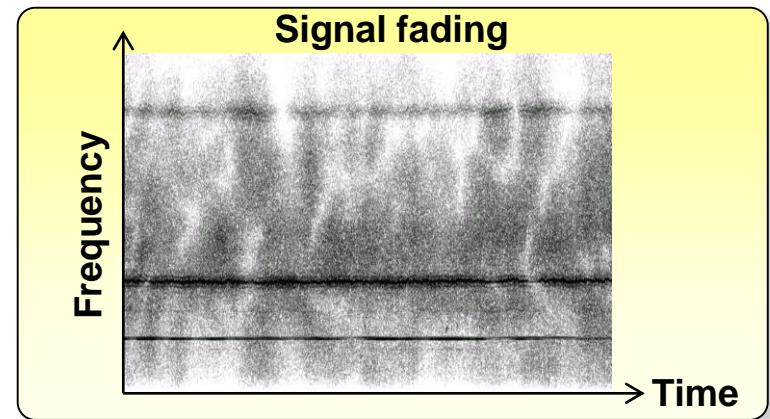
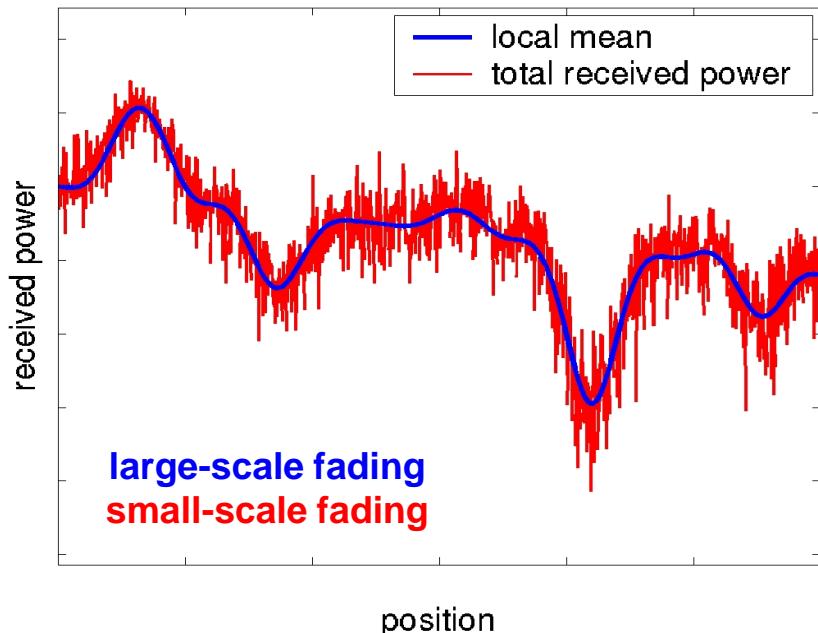
refraction in the troposphere:

- not considered

In general multipath propagation leads to fading at the receiver site

# The Received Signal

Fading is a **deviation** of the **attenuation** that a signal experiences over certain propagation media. It may vary with time, position and/or frequency



**Classification of fading:**

- **large-scale fading** (gradual change in local average of signal level)
- **small-scale fading** (rapid variations due to random multipath signals)

# Propagation Models

Propagation models (PM) are being used to predict:

- **average signal strength at a given distance from the transmitter**
- **variability of the signal strength in close spatial proximity to a particular location**

PM can be divided into:

- **large-scale models**

*(mean signal strength for large transmitter receiver separation)*

- **small-scale models**

*(rapid fluctuations of the received signal over very short travel distances)*

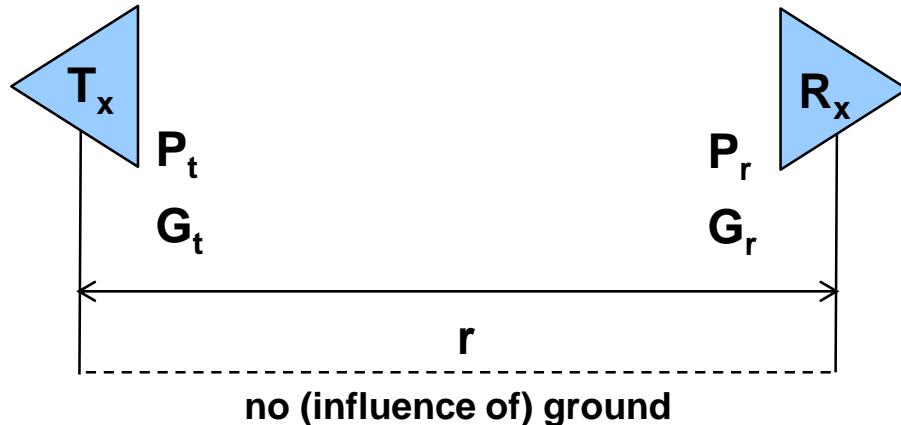
Severe multipath conditions in urban areas (*small-scale fading*)



# Large-Scale Propagation

Free Space Propagation

# Free Space Propagation



## Assumptions:

- **unobstructed line of sight (LOS)**
- **no multipath propagation**

**Received power:**

$$P_R = A_{eR} \cdot S_R$$

**Power density at Rx site:**

$$S_R = \frac{P_T G_T}{4\pi r^2}$$

**Antenna effective area:**

$$A_{eR} = \frac{\lambda^2}{4\pi} G_R$$

**Friis free space equation:**

$$P_R = \frac{\lambda^2}{4\pi} G_R \cdot \frac{P_T G_T}{4\pi r^2} = \left(\frac{\lambda}{4\pi r}\right)^2 G_R G_T P_T \propto \frac{1}{r^2}$$

# Received Power and Path Loss

**Using:**  $(P_R)^{dBm} = 10 \log \left( \frac{P_R}{1mW} \right)$

$$(P_R)^{dBm} = P_T^{dBm} + G_R^{dBi} + G_T^{dBi} - 20 \log \left( \frac{4\pi d}{\lambda} \right)$$

## Assumptions:

- polarization matched receiving antenna
- conjugate complex impedance matching of the receiver

## Path loss:

$$(P_L) = \frac{P_T}{P_R} \Rightarrow (P_L)^{dB} = 20 \log \left( \frac{4\pi d}{\lambda} \right) - G_R^{dBi} - G_T^{dBi}$$

## Isotropic path loss (no antenna gains):

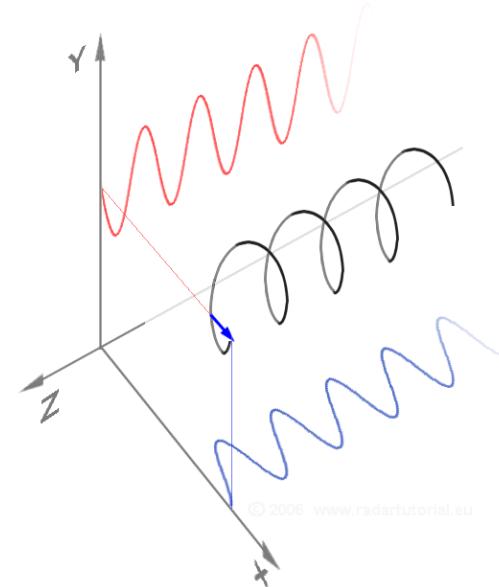
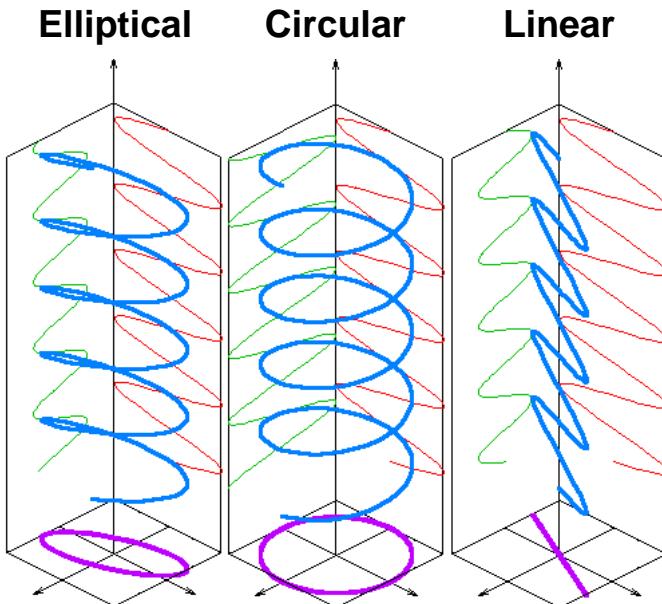
$$(P_L)^{dB} = 20 \log \left( \frac{4\pi d}{\lambda} \right)$$

# Polarization

**Orientation of Field Vectors and Reference Planes**

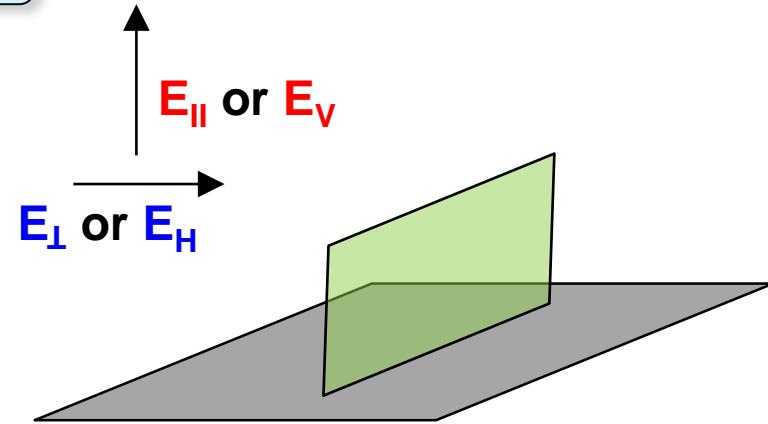
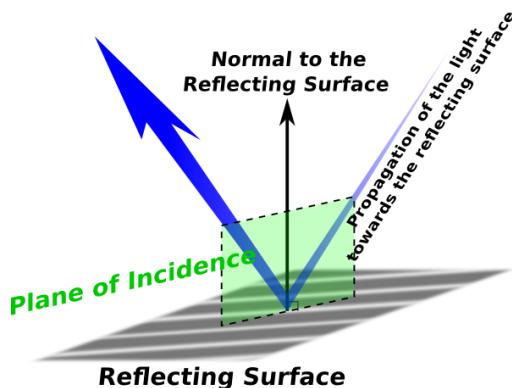
# Polarization of the EM Waves

Every **elliptically** polarized EM wave can be decomposed into a **horizontal** and a **vertical** component.



# Polarization: II, $\perp$ , V or H?

**Plane of incidence:** formed by the **normal vector to the reflecting surface** and **Poynting vector of the incidence wave**



**Polarization (E-field vector) with respect to the **plane of incidence**:**

- parallel (**II**)
- perpendicular ( **$\perp$** )

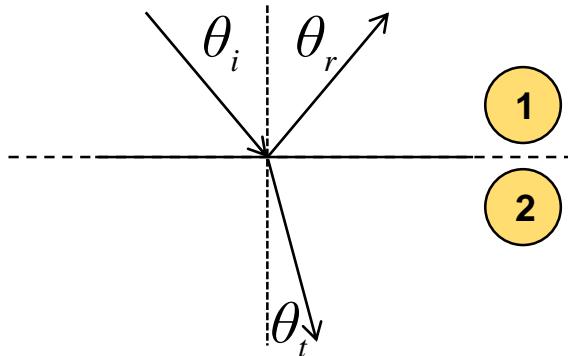
**Polarization (E-field vector) with respect to the **earth coordinates**:**

- vertical (**V**)
- horizontal (**H**)

# Reflection and Transmission

## Dielectric Boundary

# Snell's Law of Reflection



- surface large compared to the wave length
- smooth surface (otherwise scattering)
- three angles:
  - incidence
  - reflection
  - transmission / refraction

- Relation between angles through Fermat's principle (**principle of least time**):  
 - “*the rays of light (EM-waves) traverse the path of stationary optical length*”
- This results in\* Snell's laws:  
 - “*ratio of the sines of the angles of incidence and refraction is equivalent to the opposite ratio of the indices of refraction*”  
 - “*the incidence and reflection angles are equal and they are in the same plane*”

$$\frac{\sin(\theta_i)}{\sin(\theta_t)} = \frac{n_2}{n_1} \quad n_x = \sqrt{\epsilon_{r,x} \cdot \mu_{r,x}} \quad \theta_i = \theta_r$$

\*full derivation in Arthur Schuster: “An Introduction to the Theory of Optics”

# Which Part is Transmitted / Reflected?

Derivation procedure:

- ***Definition of the electric field strength of the incident wave***
- ***Reflected and transmitted field strengths***
- ***Faraday's law of induction***
- ***Boundary conditions at the border between two dielectric media***
- ***Decomposition of the incident waves on parallel and normal components***

# Fresnel Reflection & Transmission Coefficients

$$R_{\parallel} = \frac{\eta_1 \cos \Theta_i - \eta_2 \cos \Theta_t}{\eta_1 \cos \Theta_i + \eta_2 \cos \Theta_t} = \frac{E_r}{E_i}$$

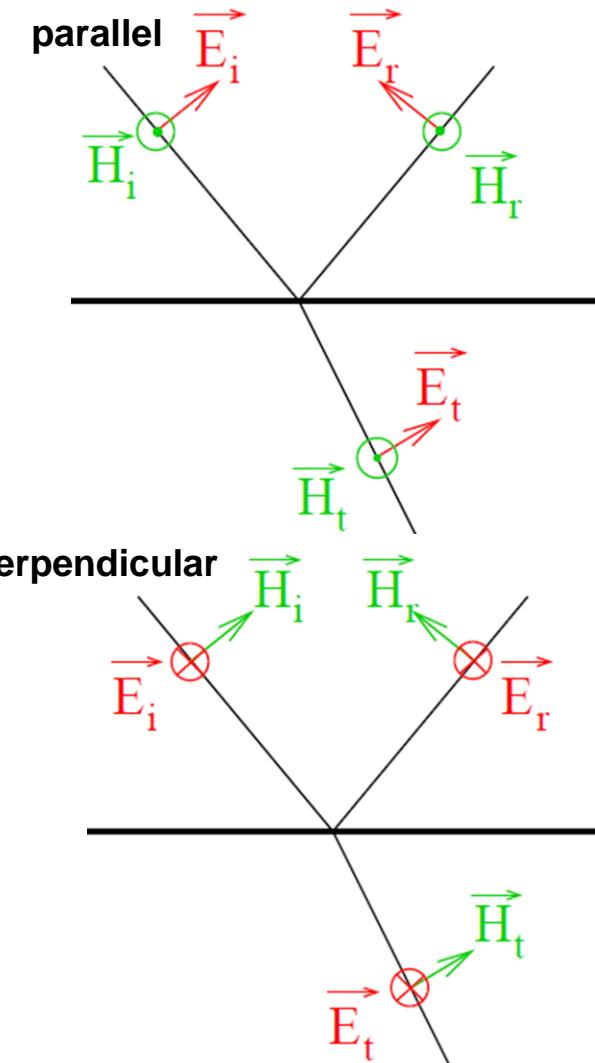
$$T_{\parallel} = \frac{2\eta_2 \cos \Theta_i}{\eta_1 \cos \Theta_i + \eta_2 \cos \Theta_t} = \frac{E_t}{E_i}$$

$$R_{\perp} = \frac{\eta_2 \cos \Theta_i - \eta_1 \cos \Theta_t}{\eta_2 \cos \Theta_i + \eta_1 \cos \Theta_t} = \frac{E_r}{E_i}$$

$$T_{\perp} = \frac{2\eta_2 \cos \Theta_i}{\eta_2 \cos \Theta_i + \eta_1 \cos \Theta_t} = \frac{E_t}{E_i}$$

where:  $\eta = \sqrt{\frac{j\omega\mu}{\sigma+j\omega\epsilon}}$

Fresnel coefficients are frequency dependent and in general complex



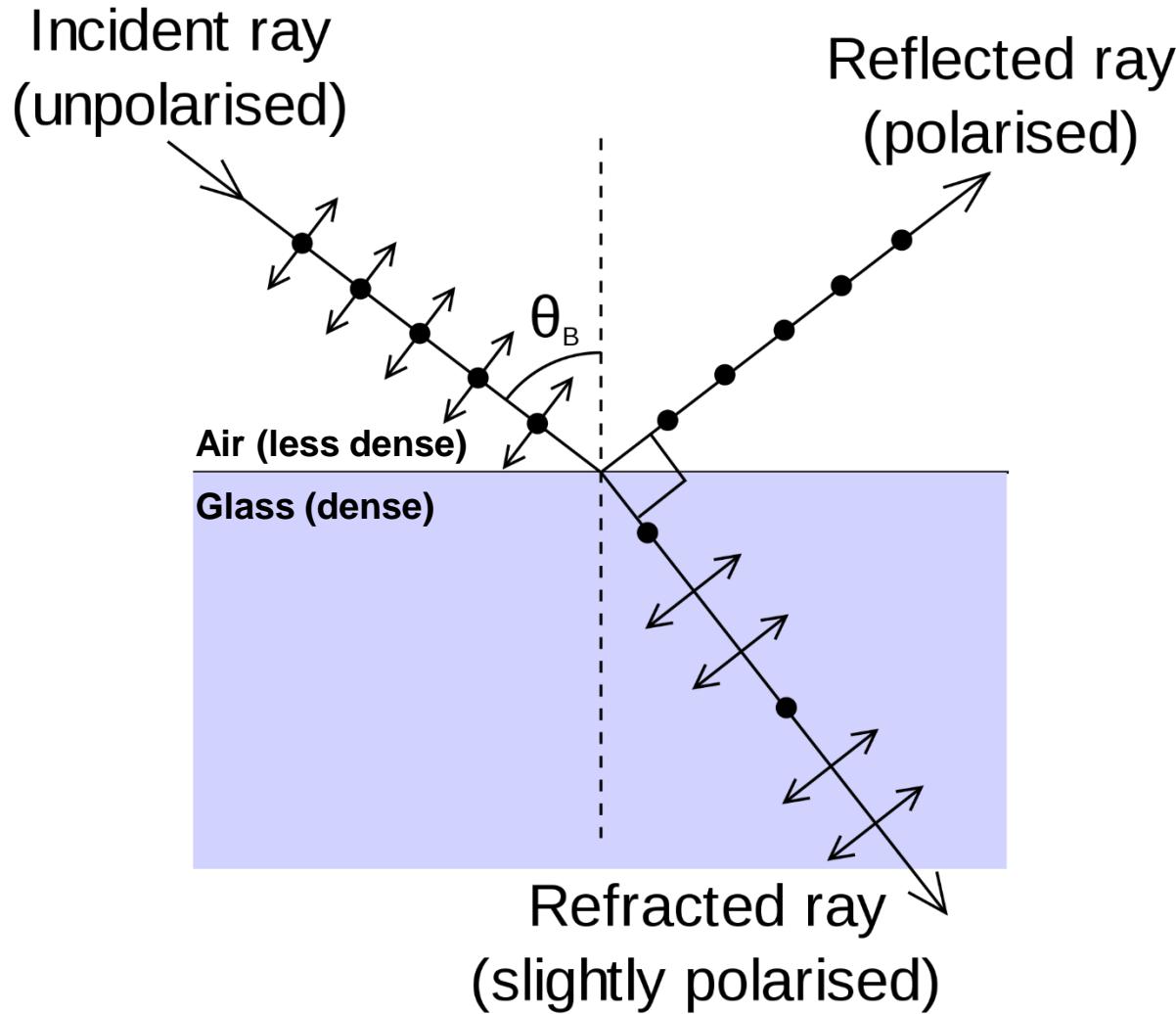
# Brewster's Angle (I)

Angle, where no reflection occurs is Brewster's Angle:

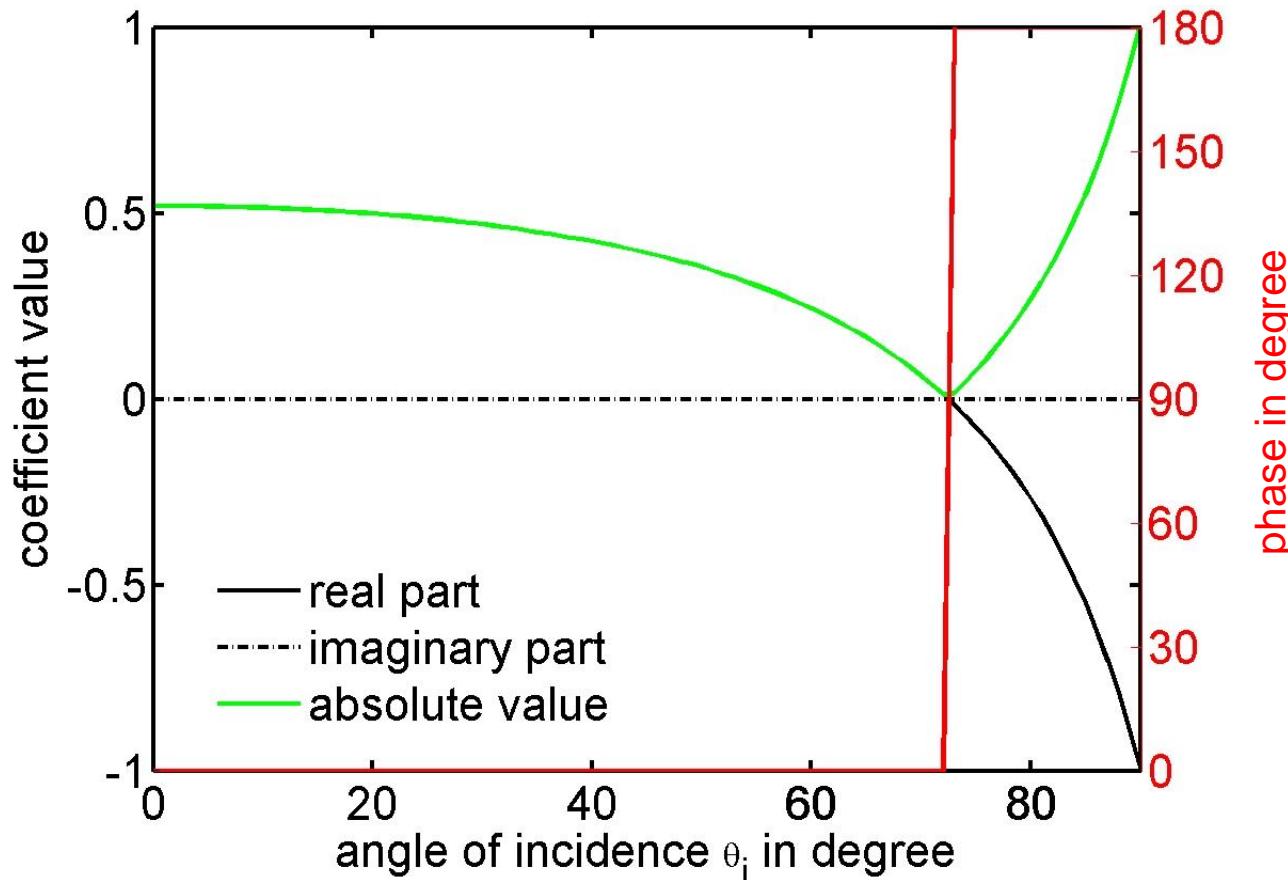
- exists only for parallel (II / V) polarization
- calculation by comparing the reflection coefficient to zero
- calculation by using “physical limitations”

$$R_{\parallel} = \frac{\eta_1 \cos \Theta_i - \eta_2 \cos \Theta_t}{\eta_1 \cos \Theta_i + \eta_2 \cos \Theta_t} \stackrel{!}{=} 0$$

# Brewster's Angle (II)



# Brewster's Angle (III)

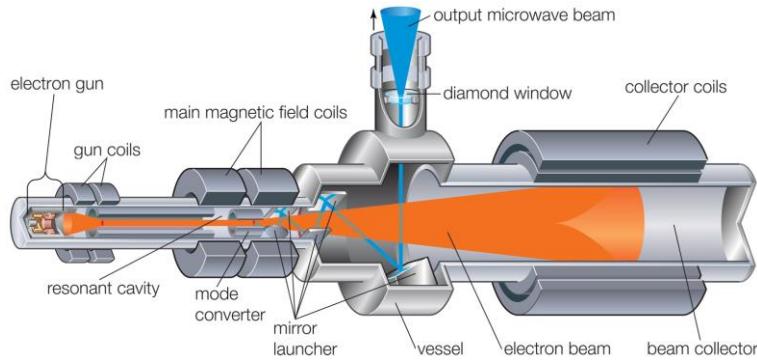


# Brewster's Angle (IV)

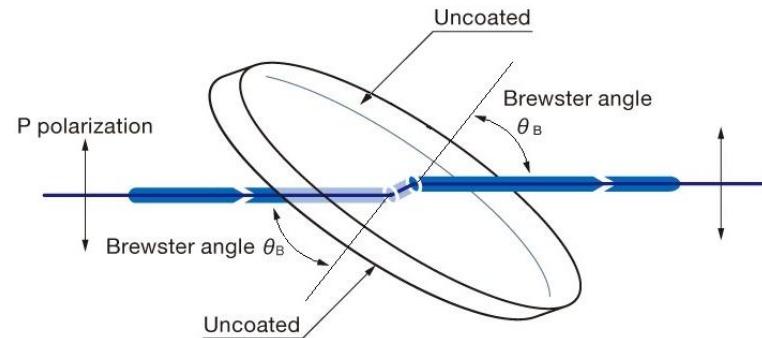
## Operation principle of Brewster window:

- used for windows in optical or quasi optical systems
- window with normal incidence → reflection loses at window
- window tilted at Brewster's angle → no reflection loses at window

## Microwave gyrotron



## Brewster window



# Total Internal Reflection (I)

When does the total internal reflection appears?

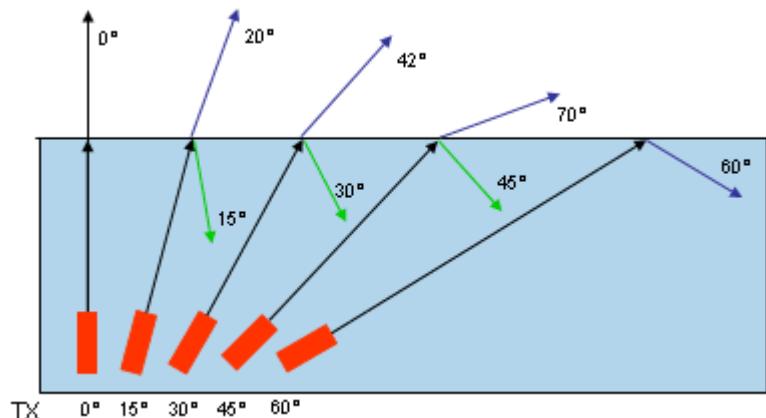
- a ray must strike the medium's boundary at an angle larger than the critical angle
- calculation by comparing the transmission angle to 90 degree

$$n_i \sin \theta_i = n_t \sin \theta_t \Big|_{\theta_t=90^\circ}$$

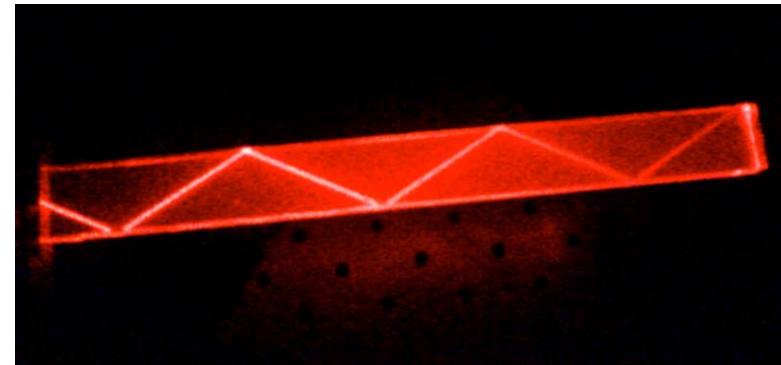
$$\theta_c = \arcsin \left( \frac{n_t}{n_i} \right)$$

critical angle exists only for  $n_t < n_i$

Increasing the incidence angle



Total reflection of red laser light in PMMA

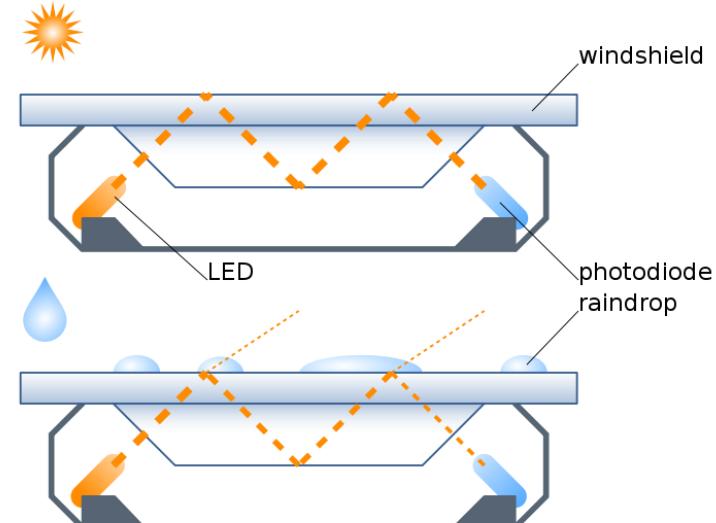


# Total Internal Reflection (II)

## Operation principle of rain sensors:

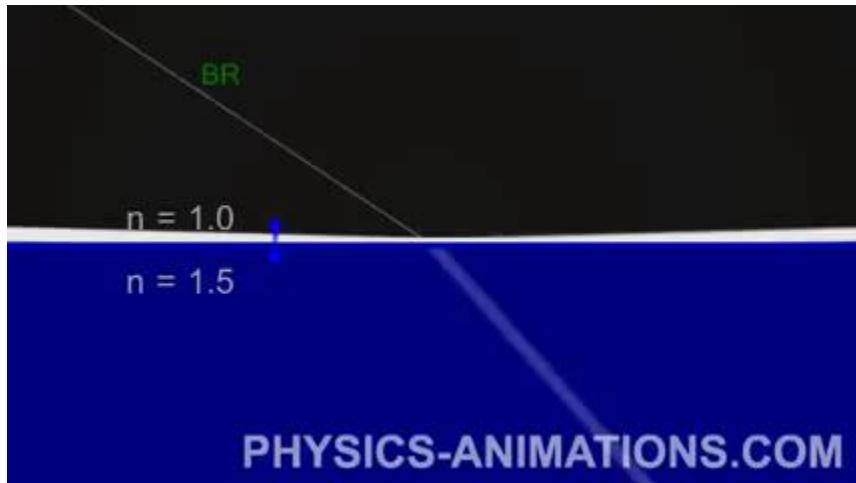
- IR-beam projected on the glass-air interface at a specific angle
- total inner reflection in dry conditions
- partial transmission to the second medium if windshield is wet
- reduced receive power triggers the sensor

Rain sensor in the rear view mirror

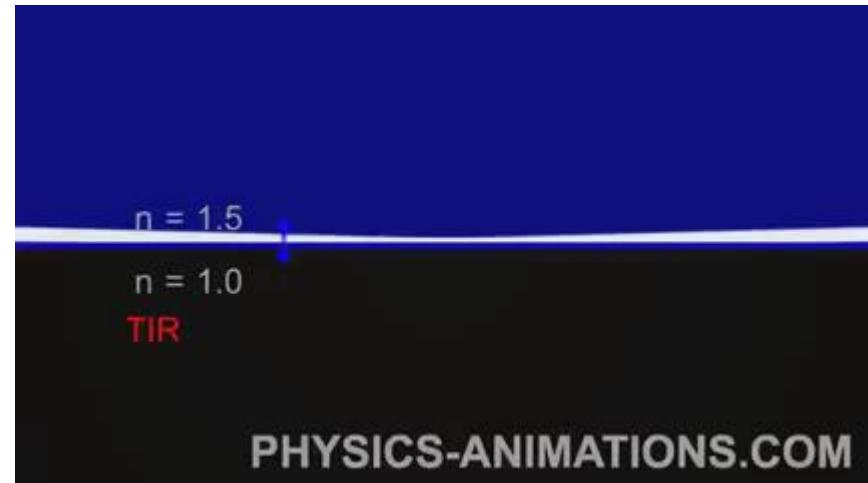


# Visualization Parallel Pol – E-Field

Parallel Pol – Air to Glass

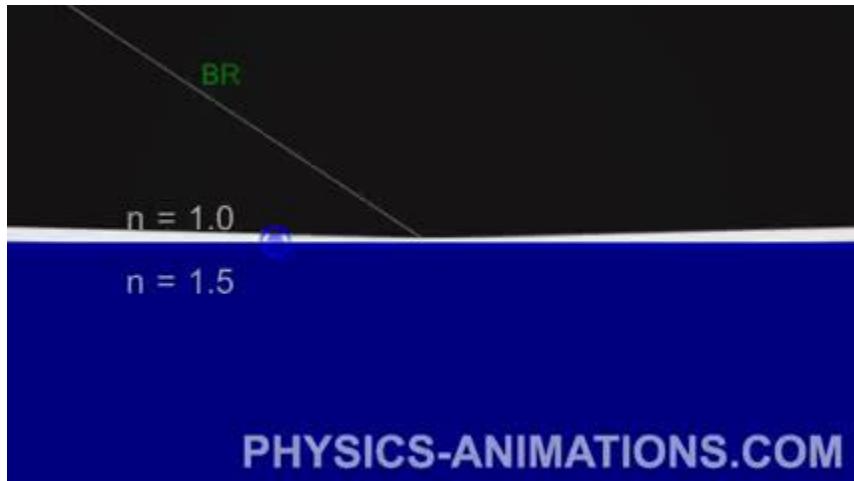


Parallel Pol – Glass to Air

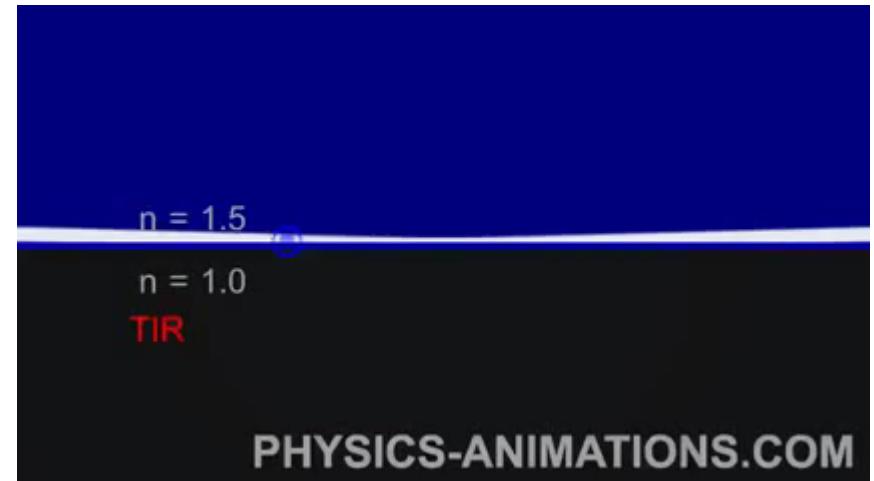


# Visualization Perpendicular Pol – E-Field

Perpendicular Pol – Air to Glass



Perpendicular Pol – Glass to Air



# Reflection and (no) Transmission

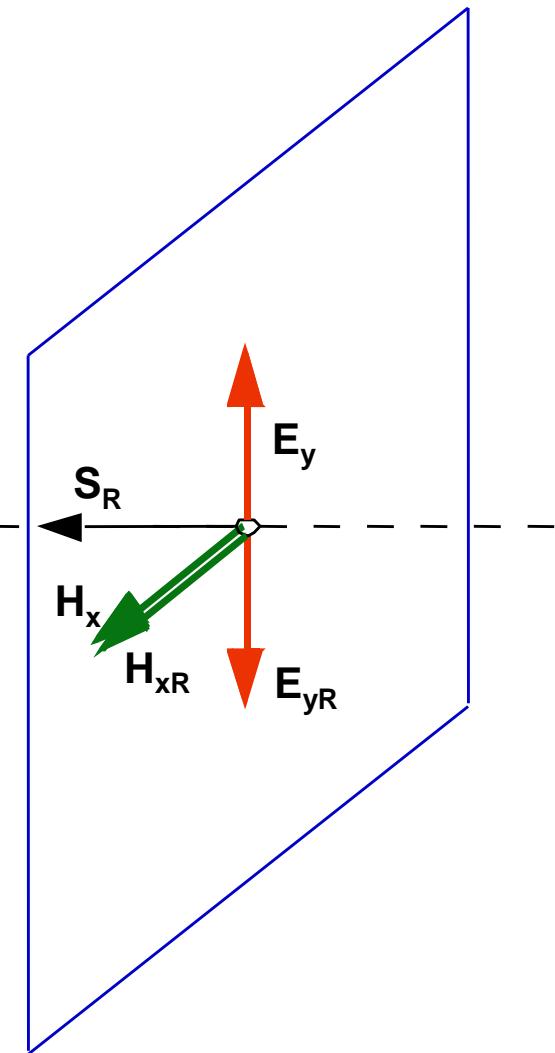
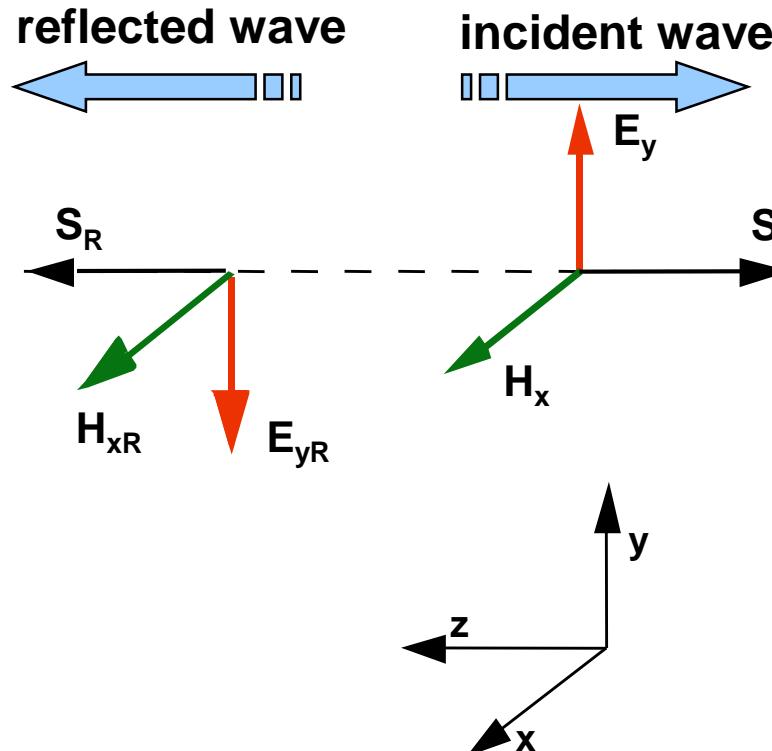
Perfect Electric Conductor (PEC)

# Orthogonal PEC Reflection

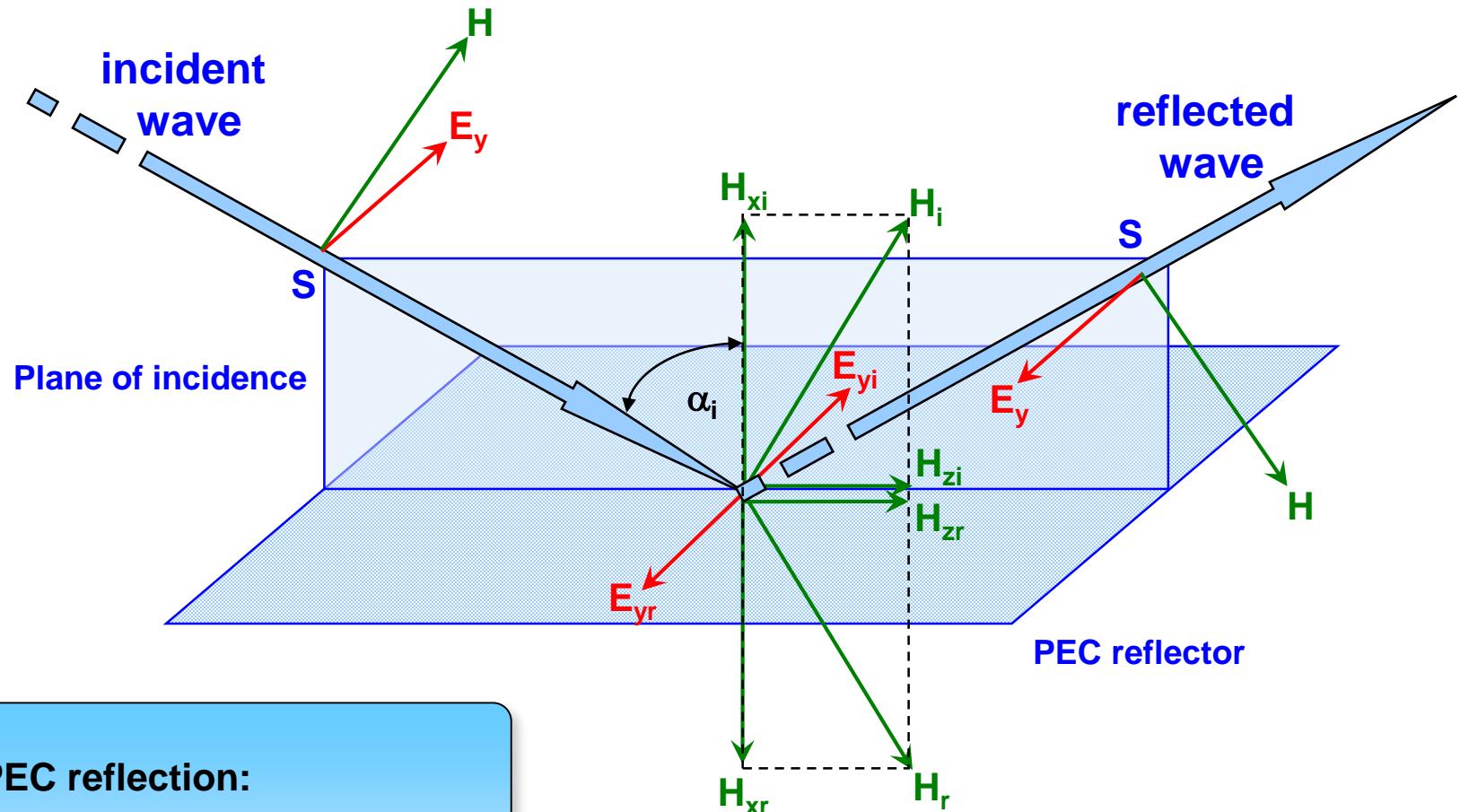
Boundary conditions:

$$\sum E_{\tan} = E_{\tan,i} + E_{\tan,r} = 0$$

$$\sum H_{norm} = H_{norm,i} + H_{norm,r} = 0$$



# PEC Reflection, Orthogonal Polarization



## PEC reflection:

- $R_{||} = +1$
- $R_{\perp} = -1$  (to ensure  $E_{tan} = 0$ )

$$\alpha_i = \alpha_r$$

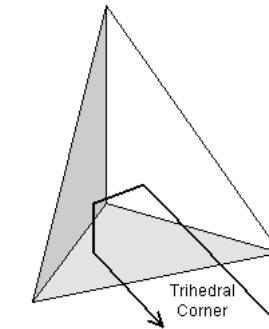
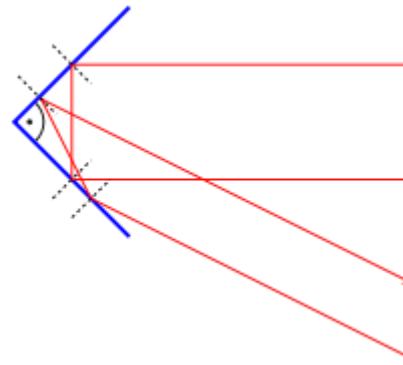
# PEC Reflection: Applications

- Radar calibration with metallic:**
- dihedral
  - trihedral (corner reflector)



Satellite radar calibration

Reflection in the direction of incidence:

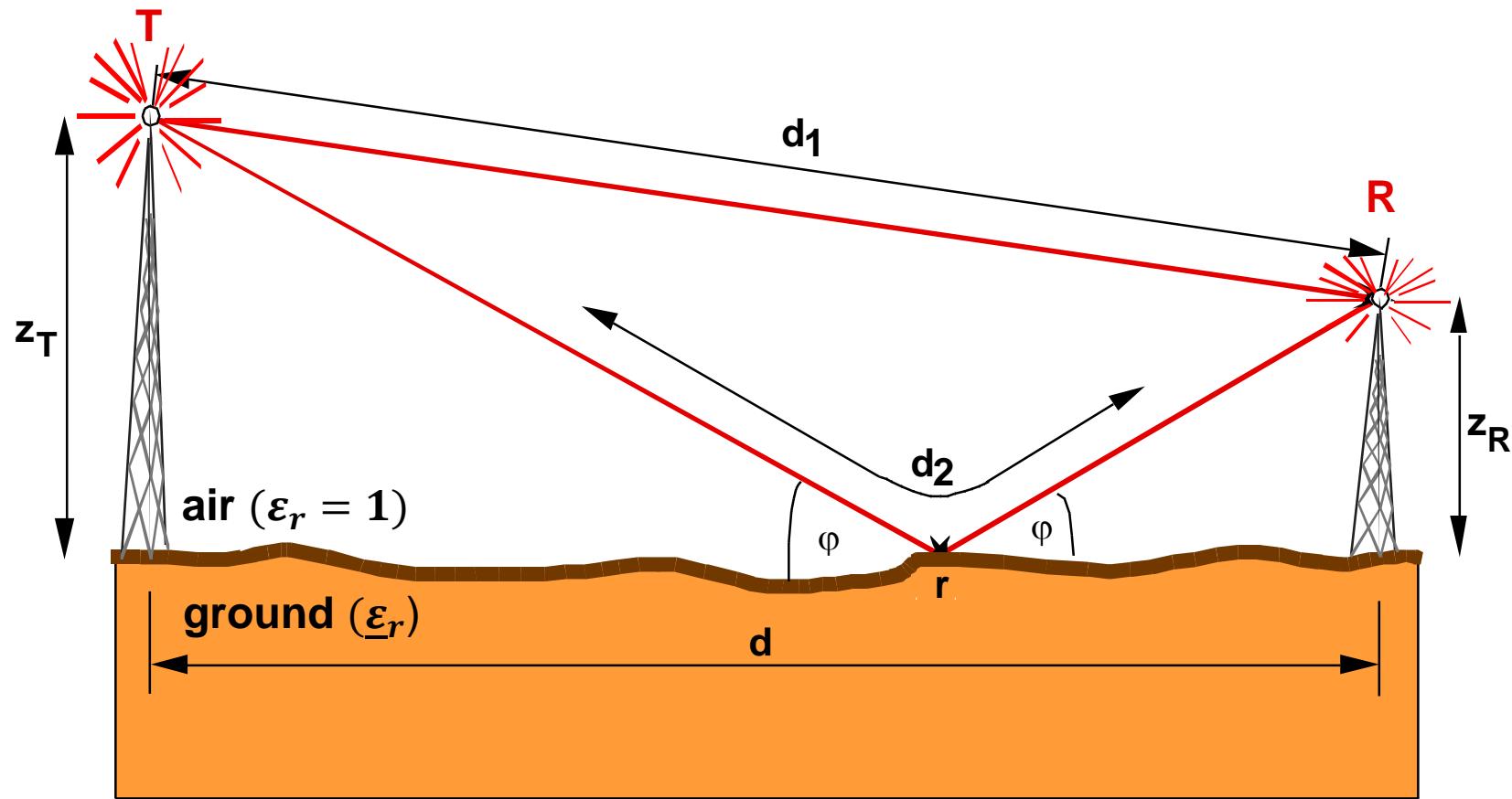


Buoy with dihedral



# Two-Ray Propagation Model

# Geometry

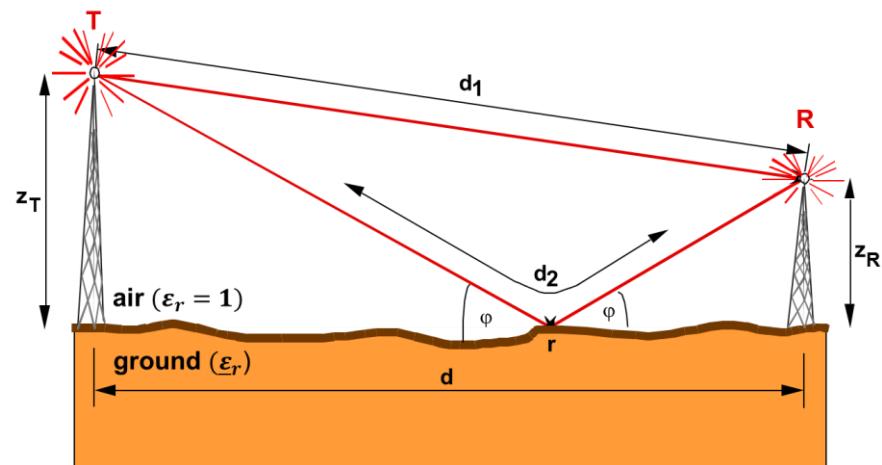


**Two-Ray model is based on geometrical optics and predicts large-scale fading**

# Assumptions

## Assumptions in two-ray model:

- ground is PEC
- $d \gg z_T, z_R$



$$P_{R\perp,\parallel} = \left(\frac{\lambda_0}{4\pi d}\right)^2 G_R G_T P_T \cdot \begin{cases} 4 \cos^2\left(\frac{k_0 z_T z_R}{d}\right) & \text{for } \parallel \\ 4 \sin^2\left(\frac{k_0 z_T z_R}{d}\right) & \text{for } \perp \end{cases}$$

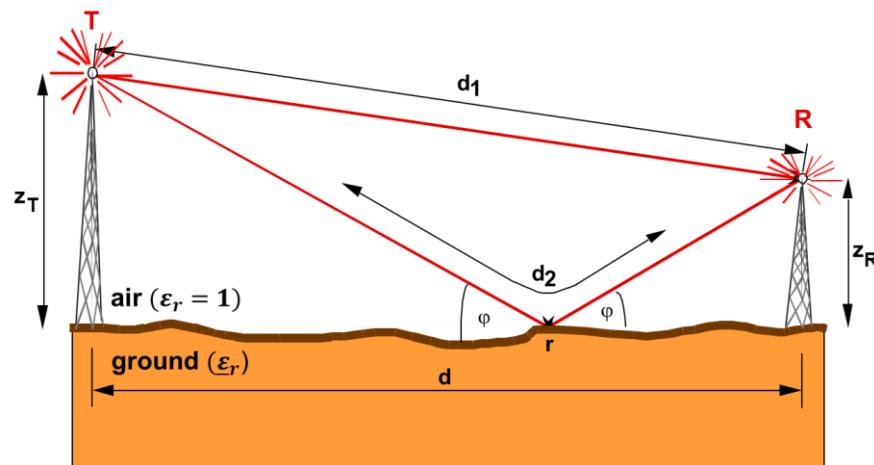
## Observations:

- the received power  $P_R$  oscillates like a  $\sin^2$  or  $\cos^2$  with distance
- the minimum value of  $P_R$  is 0
- the maximum value of  $P_R$  is  $4 \cdot P_{R,\text{freespace}}$  (+ 6 dB)

# Large Distances

## Conditions:

- $d \gg k_0 z_T z_R$
- $\cos^2 x \rightarrow 1$
- $\sin^2 x \rightarrow x^2$



$$P_{R\perp,\parallel} = \begin{cases} 4 \left( \frac{\lambda_0}{4\pi d} \right)^2 G_R G_T P_T & \text{for } \parallel \\ 4 \left( \frac{\lambda_0}{4\pi d} \right)^2 G_R G_T P_T \cdot \left( \frac{k_0 z_T z_R}{d} \right)^2 = P_T G_T G_R \frac{(z_R z_T)^2}{d^4} & \text{for } \perp \end{cases}$$

## Observations:

- parallel pol: 20 dB / decade, perpendicular pol: 40 dB / decade
- perpendicular pol: independent on frequency
- perpendicular pol: **antenna height gain** (double  $z_T$  or  $z_R \rightarrow$  quadruple  $P_R$ )

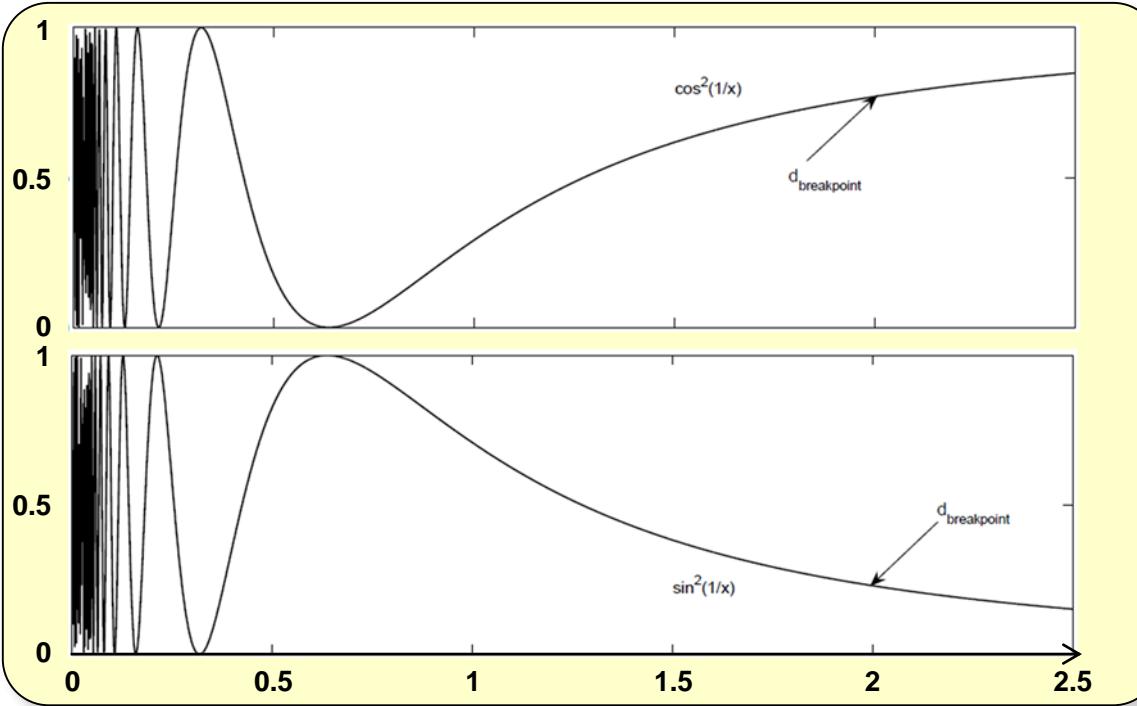
# Breakpoint

**Definition:**

The **breakpoint** is the distance  
where the argument of the  
 **$\sin^2$  and  $\cos^2$**  terms equals 0.5

$$\Rightarrow \frac{k z_R z_T}{d} \leq \frac{1}{2}$$

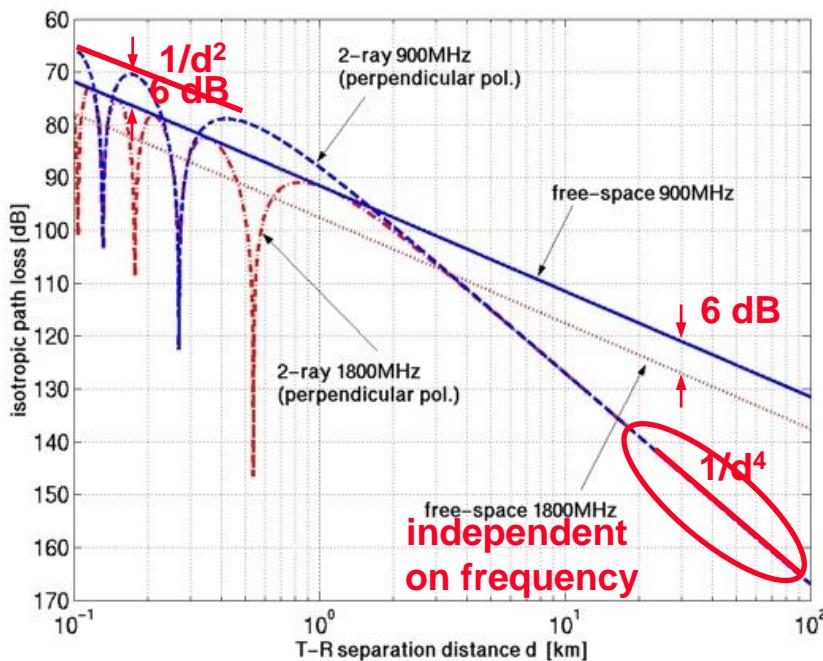
$$d_{\text{breakpoint}} = 2k_0 z_T z_R = \frac{4\pi z_T z_R}{\lambda}$$



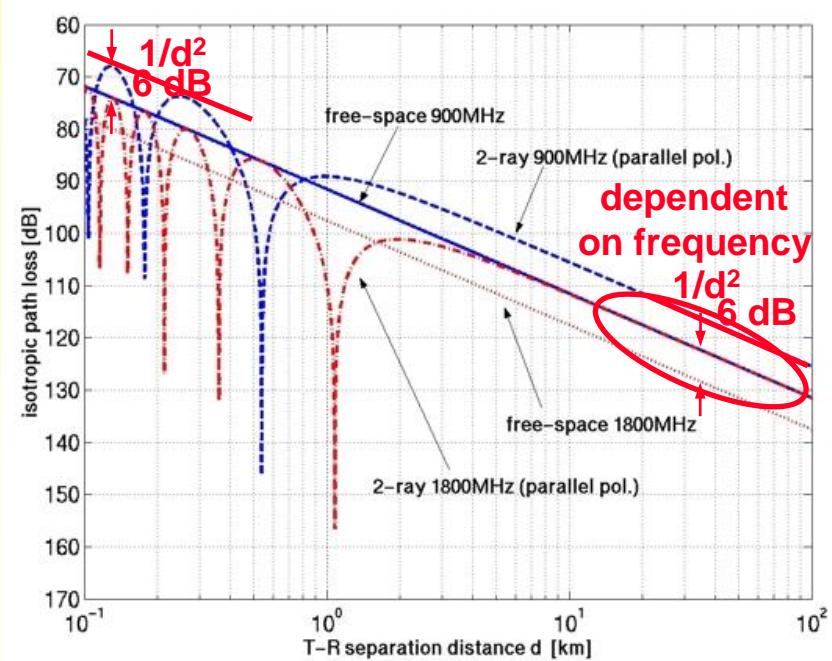
**Beyond the  
breakpoint there  
are no oscillations!**

# Polarization Dependence

perpendicular polarization



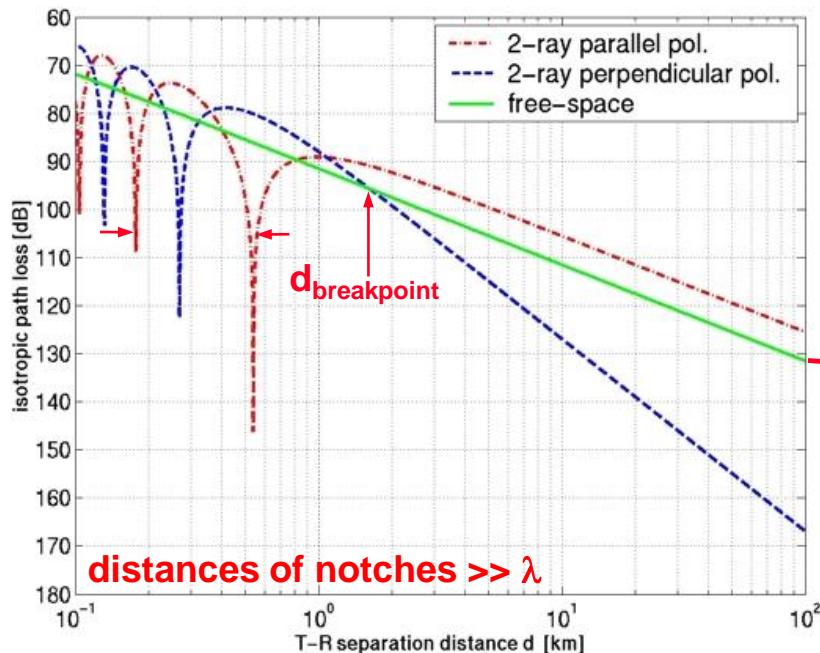
parallel polarization



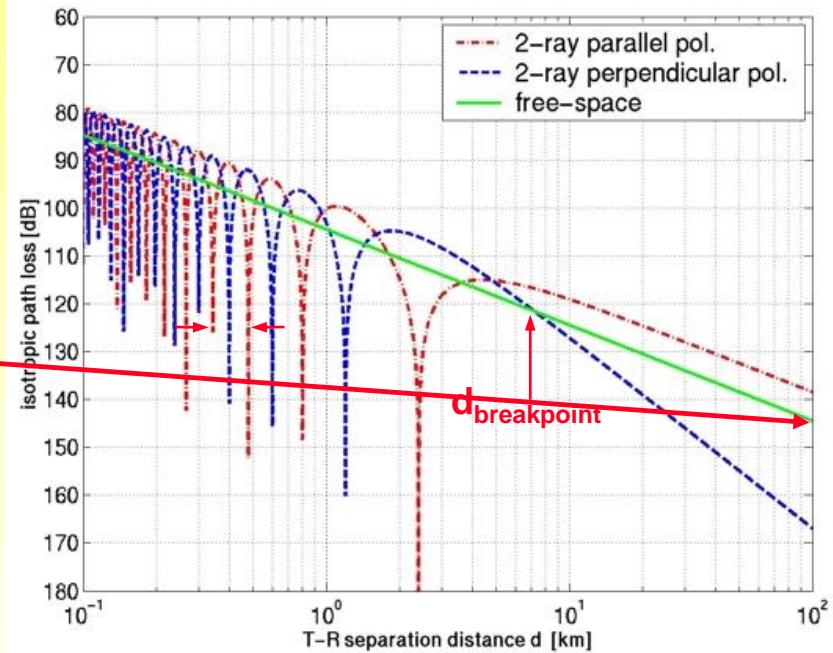
$$P_{R\perp,\parallel} = \begin{cases} 4 \left( \frac{\lambda_0}{4\pi d} \right)^2 G_R G_T P_T (\parallel \text{ polarization}) \propto \frac{1}{d^2} \\ 4 \left( \frac{\lambda_0}{4\pi d} \right)^2 G_R G_T P_T \cdot \left( \frac{k_0 z_T z_R}{d} \right)^2 = P_T G_T G_R \frac{(z_R z_T)^2}{d^4} \text{ for } \perp \end{cases}$$

# Frequency Dependence

**f = 900 MHz**

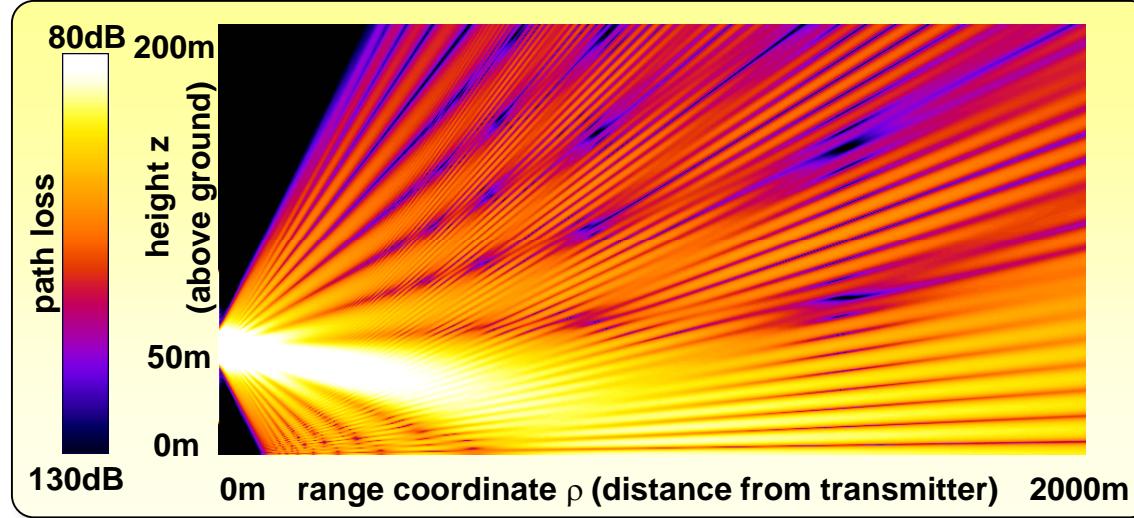


**f = 4 GHz**

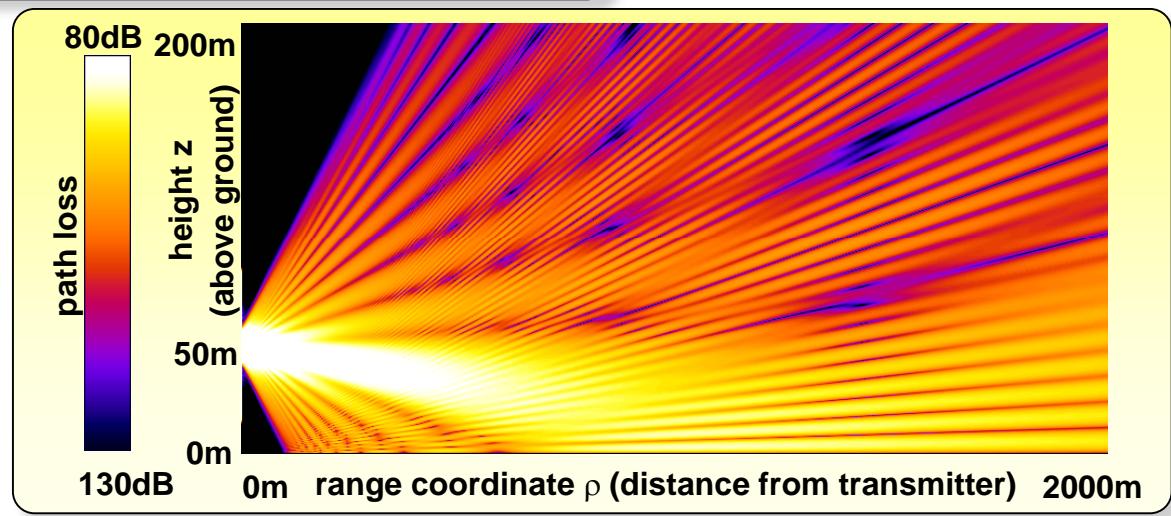


$$d_{\text{breakpoint}} = 2k_0 z_T z_R = \frac{4\pi z_T z_R}{\lambda}$$

# Path Loss Prediction



vertical (parallel)  
polarization

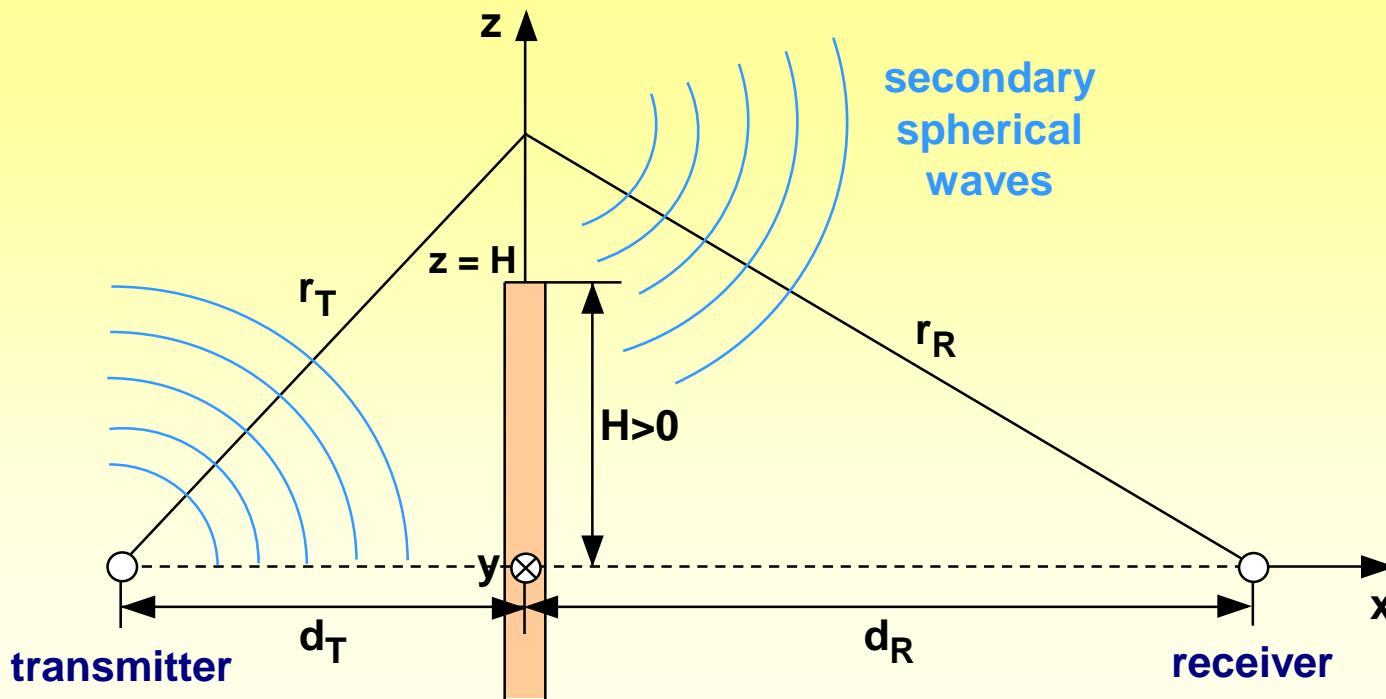


horizontal (perpendicular)  
polarization

# Diffraction

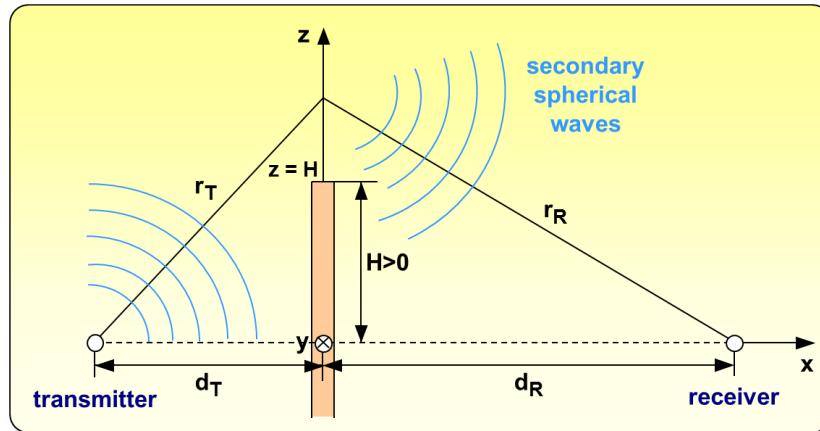
## Diffraction on Absorbing Half-Plates

# Knife Edge Diffraction: Geometry



- obstacle: semi-infinite, infinitely thin, absorbing plate
- calculate behavior behind the plate: *Huygens' principle*
- wave propagation behind the plate: sum of secondary waves

# Knife Edge Diffraction: Model



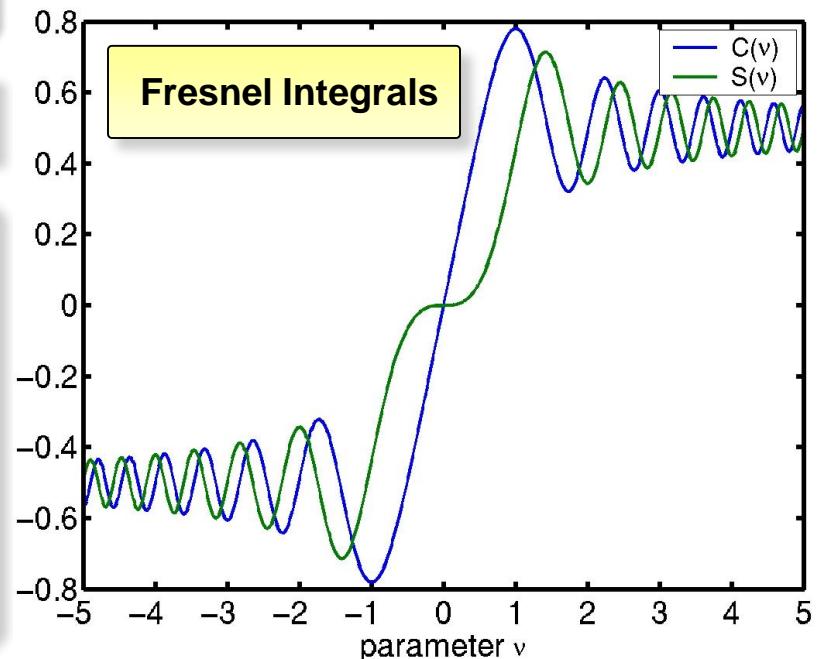
**Assumptions in knife edge model:**

- cylindrical waves (2D problem)
- $T_x$  and  $R_x$  at same height
- $|H| \ll d_T, d_R$

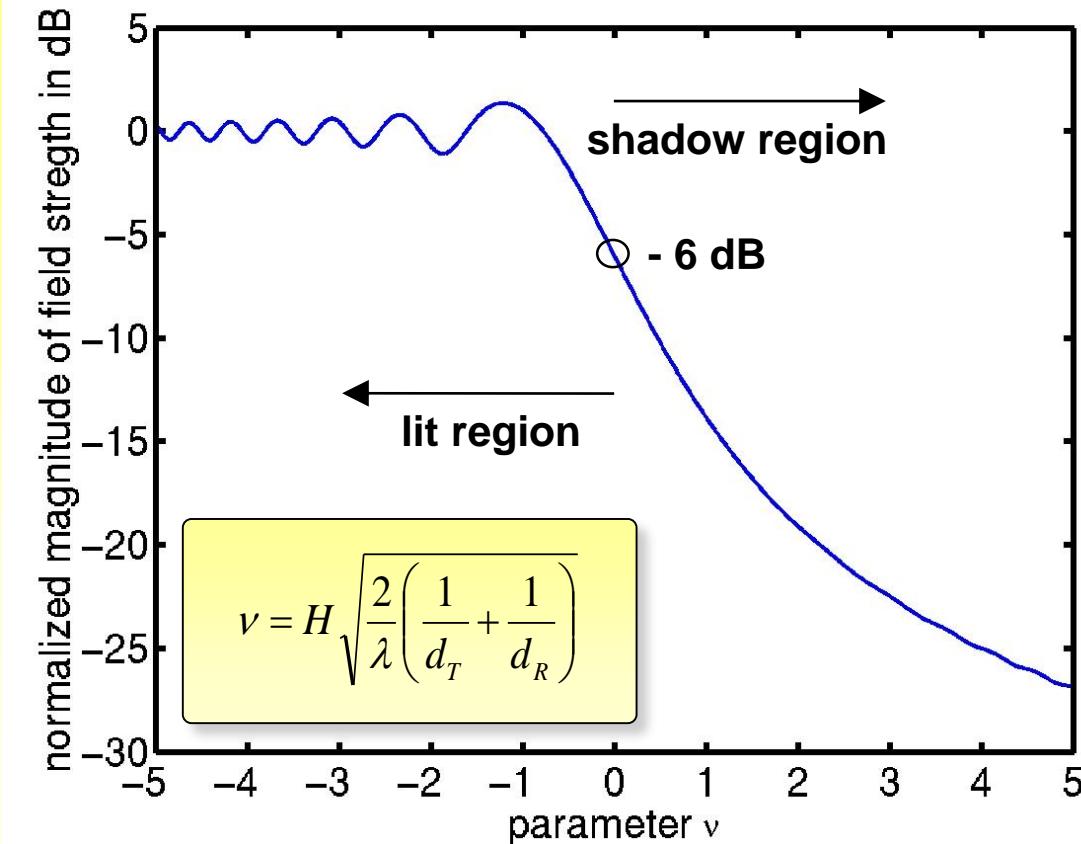
Field-strength relative to free space (no obstacle):

$$\left| \frac{E}{E_{H \rightarrow \infty}} \right| = \frac{1}{\sqrt{2}} \sqrt{\left( \frac{1}{2} - C(v) \right)^2 + \left( \frac{1}{2} - S(v) \right)^2}$$

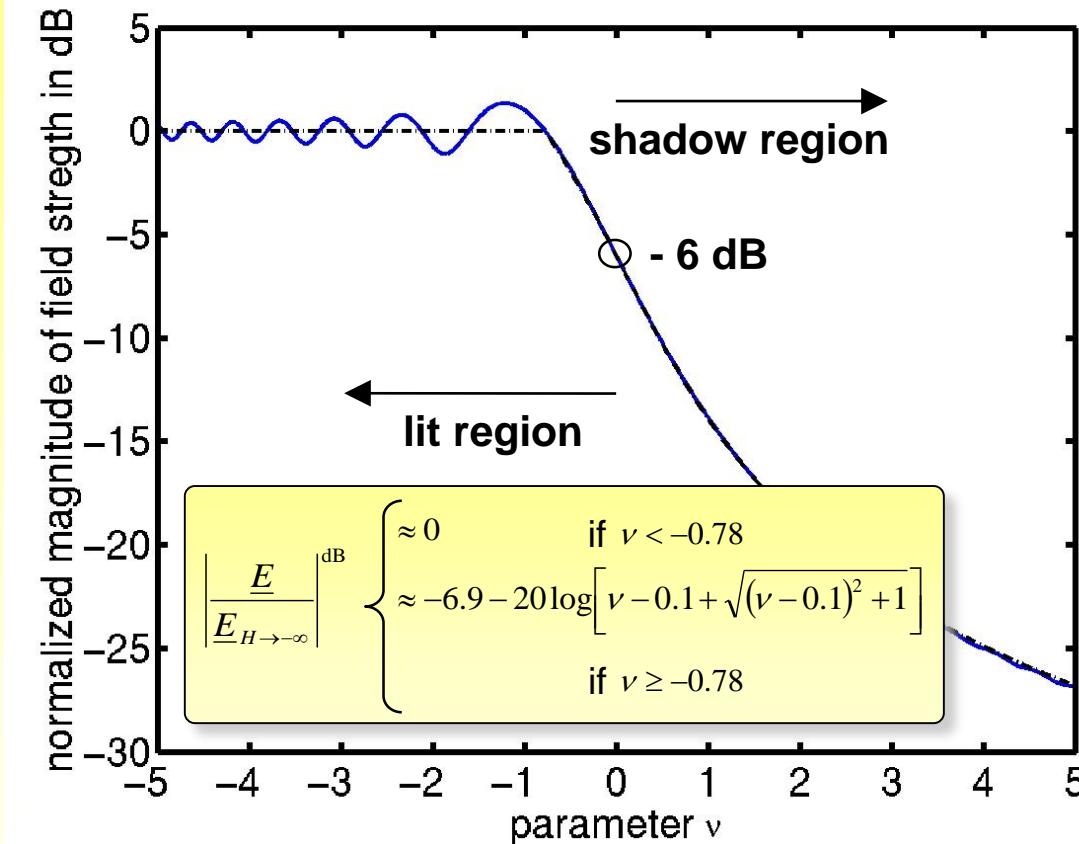
$$v = H \sqrt{\frac{2}{\lambda} \left( \frac{1}{d_T} + \frac{1}{d_R} \right)}$$



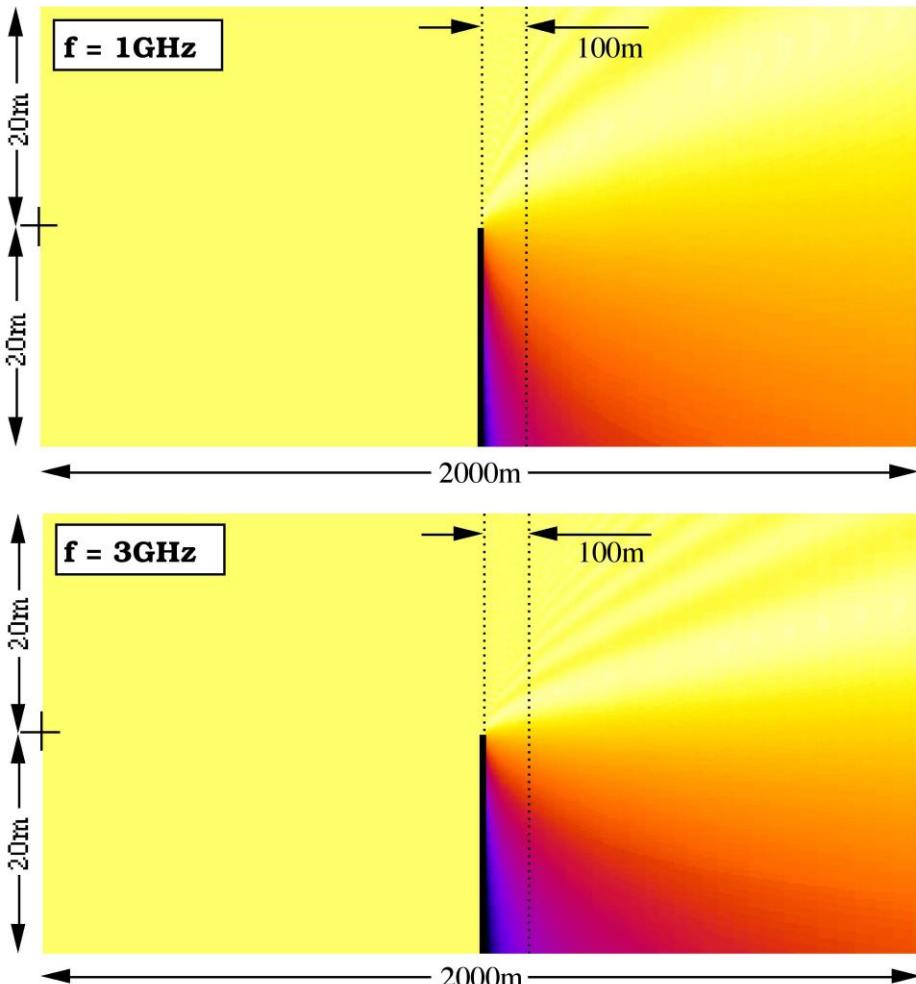
# Knife Edge Diffraction: Electric Field (I)



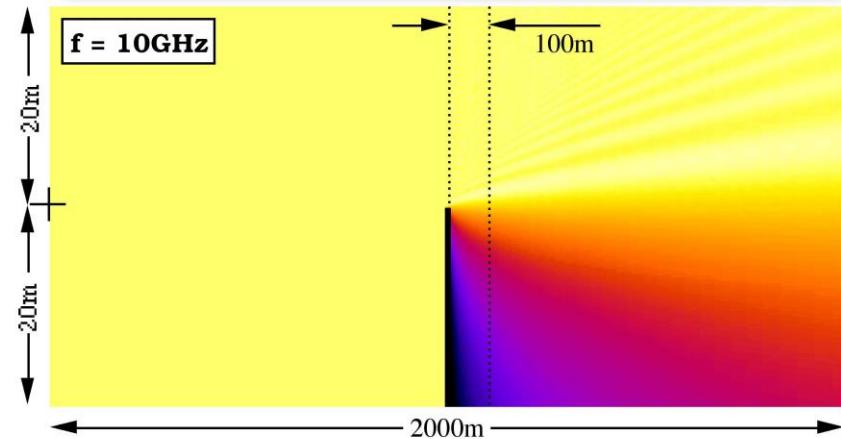
# Knife Edge Diffraction: Electric Field (II)



# Knife Edge Diffraction: Frequency Dependence (I)



- field strength normalized to free space level
- isotropic Tx antenna
- semi-infinite, absorbing plate
- $f = 1 \text{ GHz}, 3 \text{ GHz}, 10 \text{ GHz}$



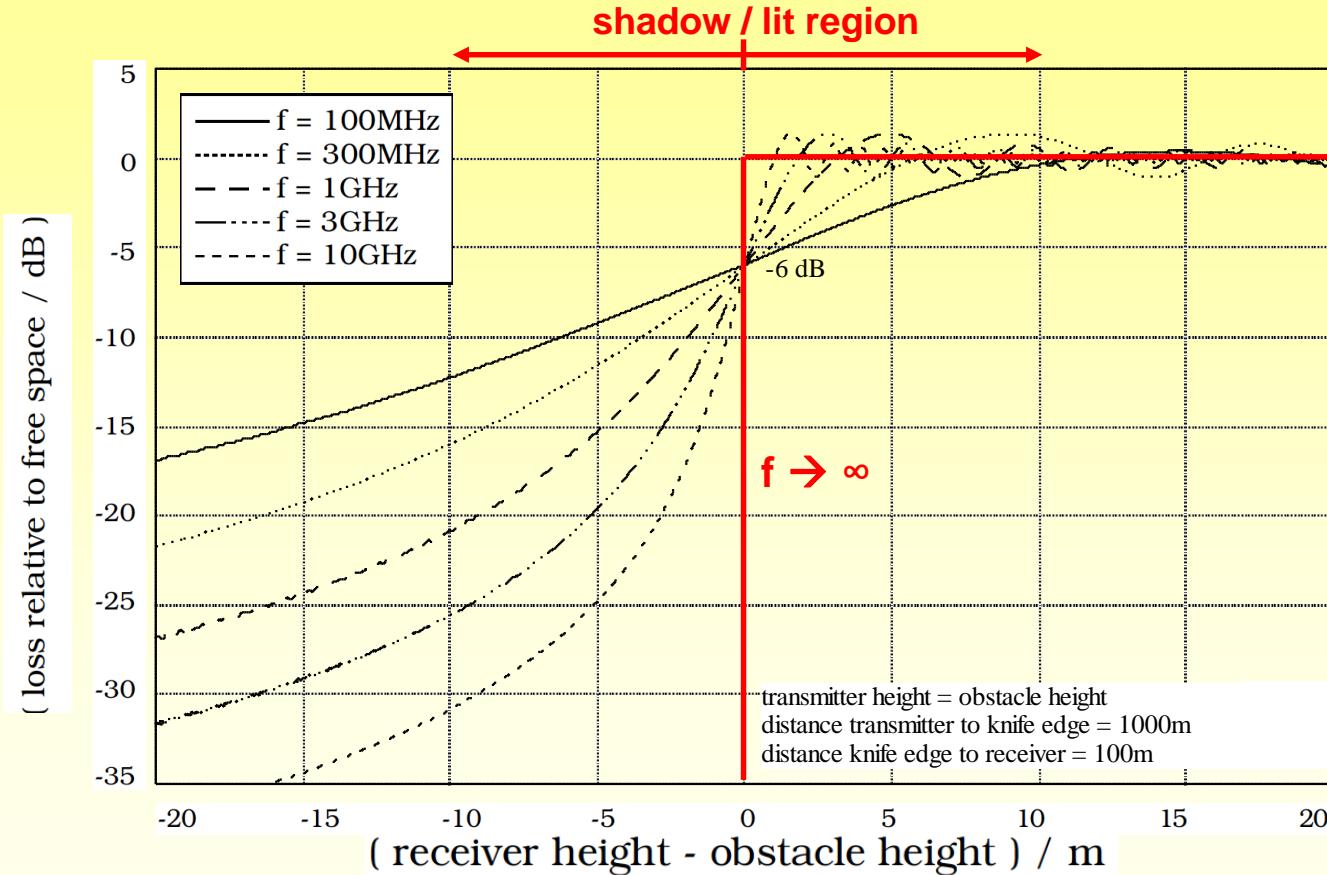
diffraction loss increases with frequency

$$\propto \sqrt{f} \quad \text{for } \nu \gg 1$$

normalized fieldstrength  $|E/E_0|$

- 45dB      5dB

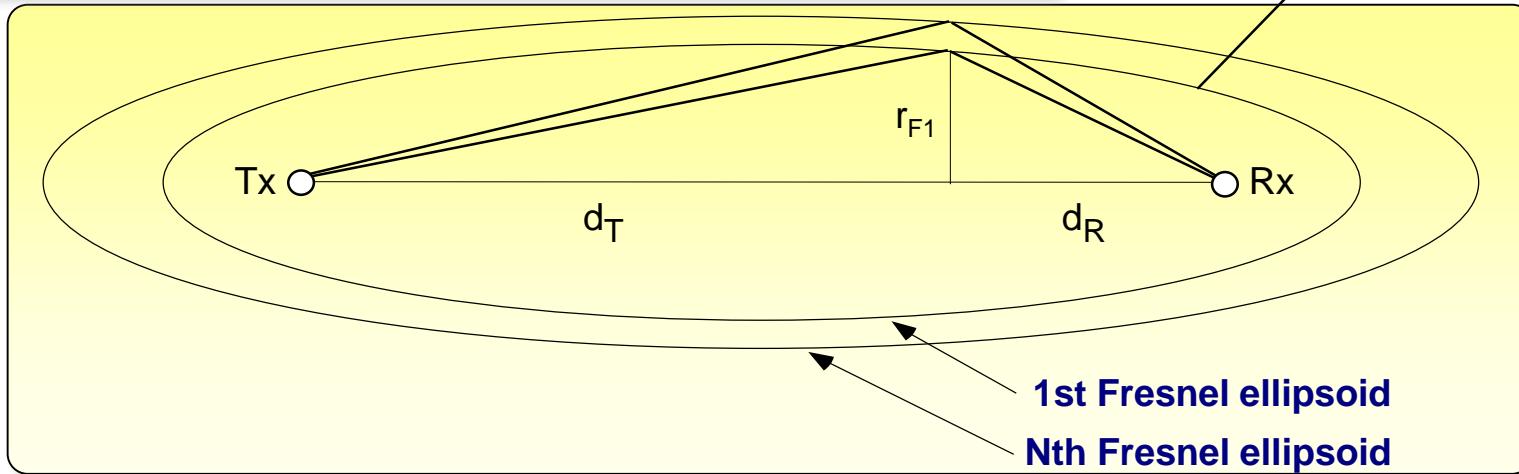
# Knife Edge Diffraction: Frequency Dependence (II)



# Fresnel Ellipsoids

**N<sup>th</sup> Fresnel zone is bounded by an ellipsoid, where the Tx-Rx-path is N half wavelengths longer than the direct Tx-Rx-path  $d_T + d_R$  between Tx and Rx**

$$d_{FN} = d_T + d_R + N \cdot \lambda/2$$



$$\sqrt{d_T^2 + R_{FN}^2} + \sqrt{d_R^2 + R_{FN}^2} - d_T - d_R \stackrel{\sqrt{1+x^2} \approx 1 + \frac{1}{2}x^2}{\approx} \frac{1}{2} \left( \frac{1}{d_T} + \frac{1}{d_R} \right) R_{FN}^2 \stackrel{!}{=} \frac{N\lambda}{2}$$

$$\Rightarrow R_{FN} = \sqrt{N\lambda \frac{d_T d_R}{d_T + d_R}}$$

**Radius of N<sup>th</sup> Fresnel ellipsoid**

# When to Neglect the Knife Edge Diffraction?

Relate Fresnel radius  $R_{FN}$  with diffraction parameter  $\nu$ :

$$R_{FN} = \sqrt{N\lambda \frac{d_T d_R}{d_T + d_R}}$$

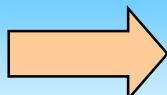
$$\frac{d_T d_R}{d_T + d_R} = \left( \frac{R_{FN}}{\sqrt{N\lambda}} \right)^2$$

$$\nu = H \sqrt{\frac{2}{\lambda} \left( \frac{1}{d_T} + \frac{1}{d_R} \right)} = H \sqrt{\frac{2}{\lambda} \left( \frac{d_T d_R}{d_T + d_R} \right)^{-1}}$$

$$\nu = \frac{H}{R_{FN}} \sqrt{2N}$$

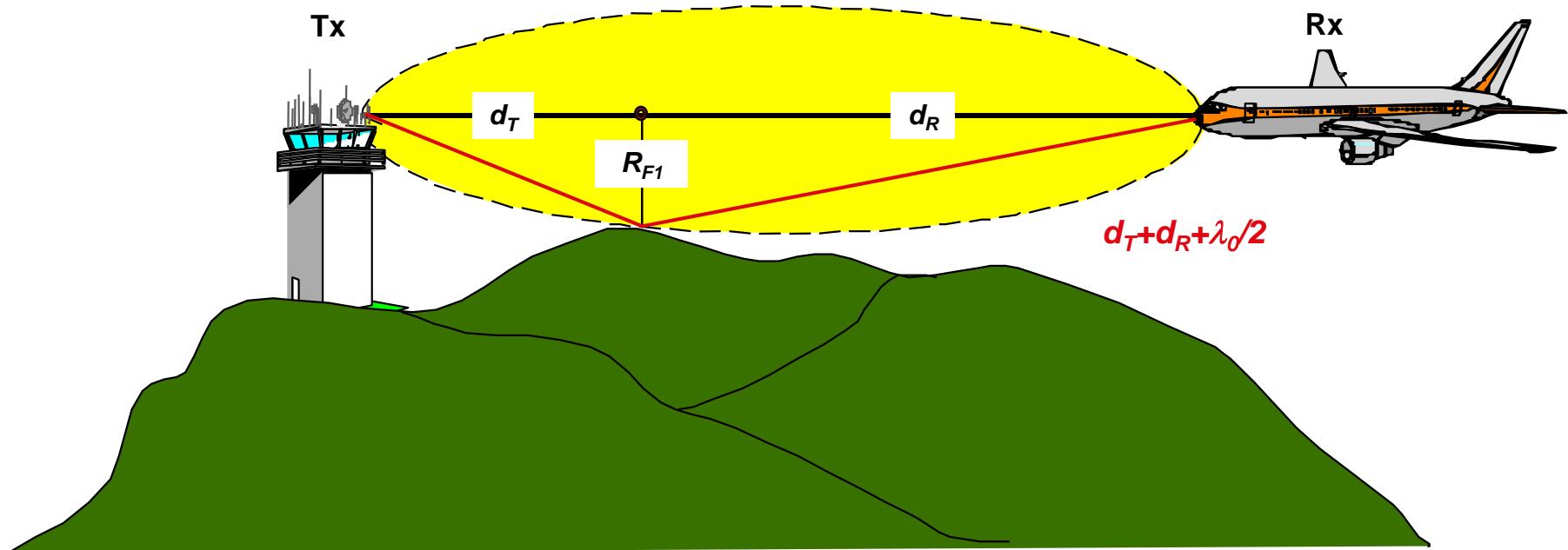
If the knife edge does not extend into 1<sup>st</sup> Fresnel zone, then the error compared to free space propagation is less than 1.1 dB:

$$\begin{aligned} -H &> R_{F1} \\ \nu &< -\sqrt{2} \end{aligned}$$



If the knife edge does not extend into the 1<sup>st</sup> Fresnel zone, then knife edge diffraction can be neglected

# Fresnel Ellipsoids: Example



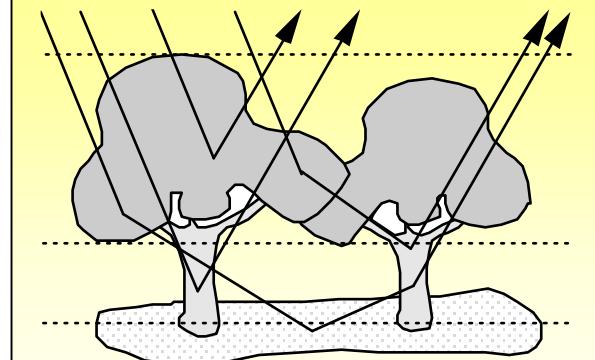
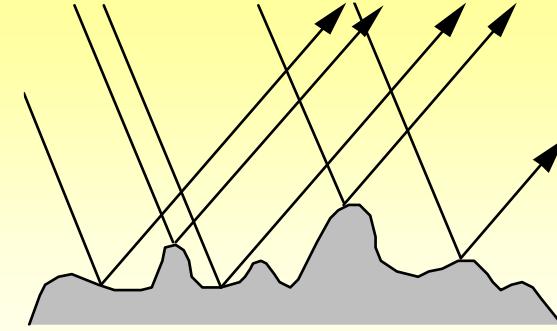
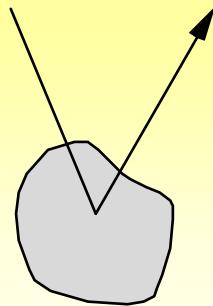
# Scattering

**Scattering of Incident Energy on Rough Surfaces**

# Different Types of Scattering

point scattering

distributed scattering

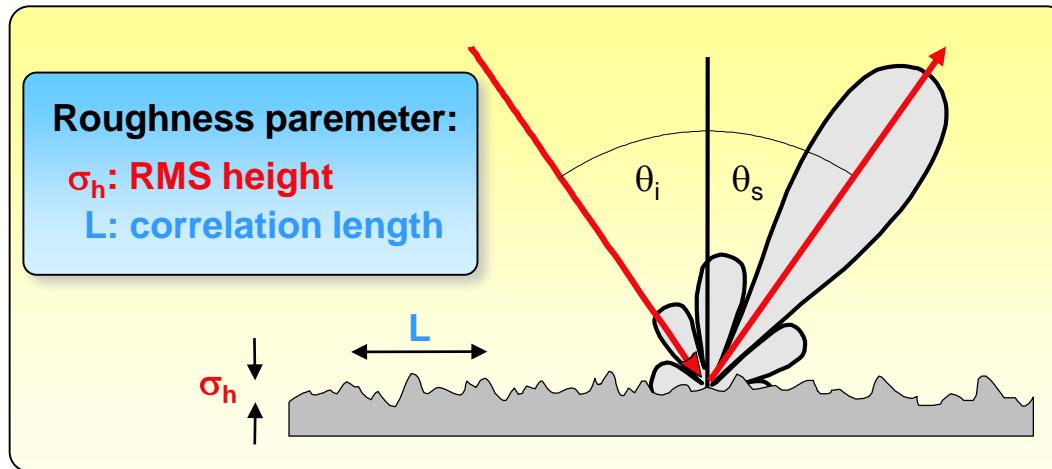
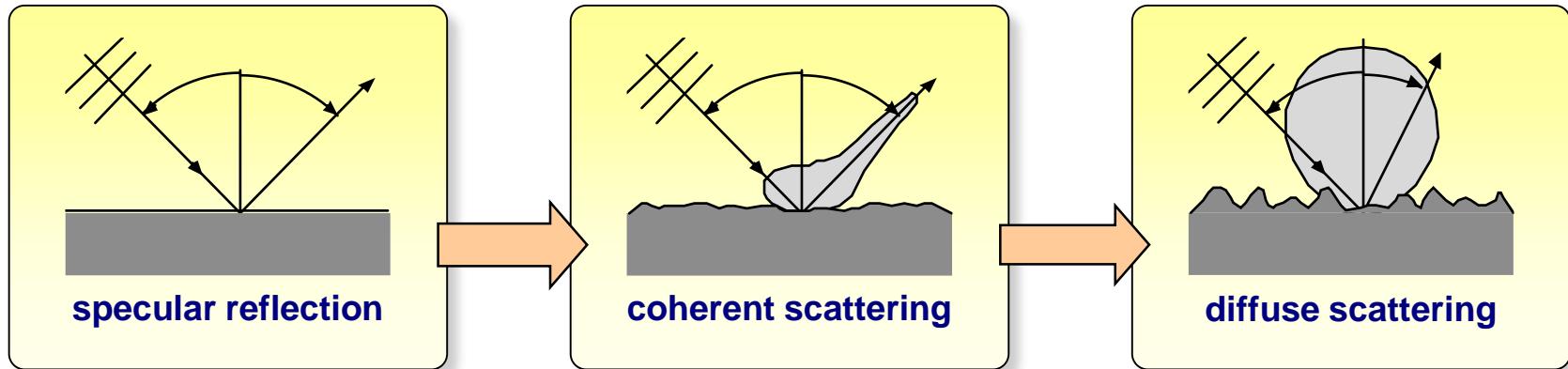


Simple targets  
(plate, sphere,  
cylinder, etc.)

rough surface scattering

volume scattering

# From Specular Reflection to Incoherent Scattering



**Roughness criteria:**

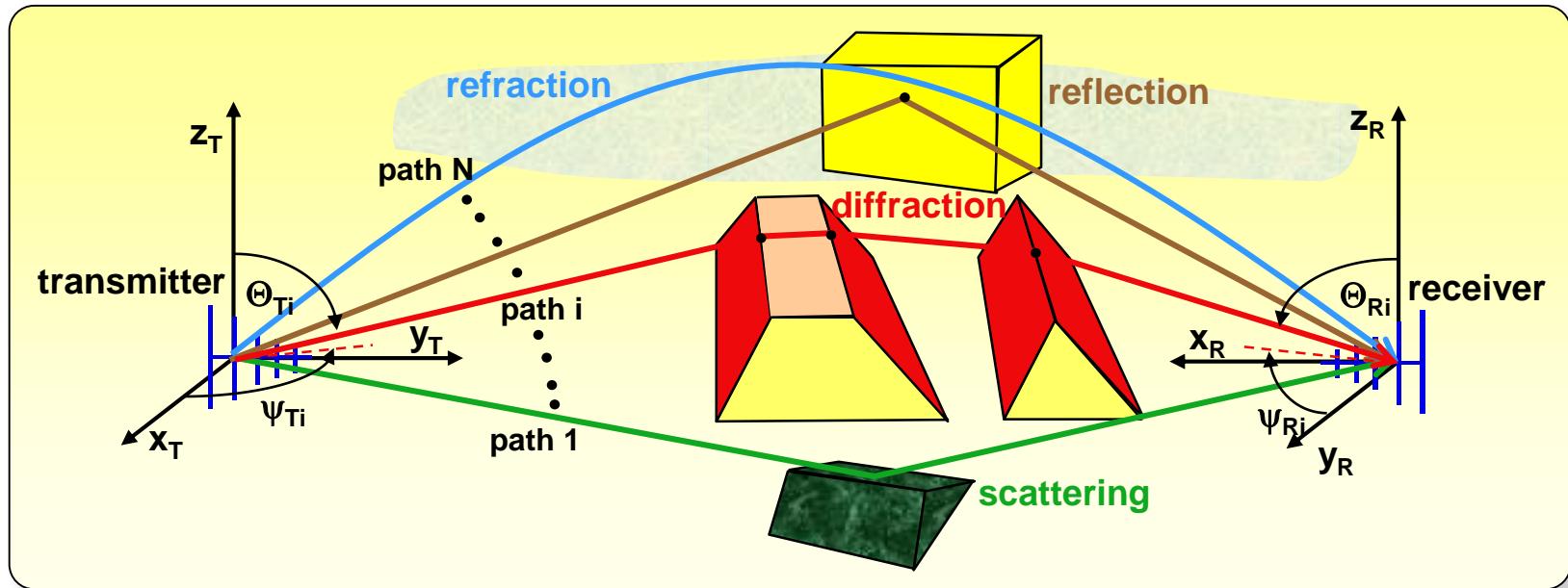
**Rayleigh:** 
$$\sigma_h \leq \frac{\lambda_0}{8 \cos \theta_i}$$

**Fraunhofer:** 
$$\sigma_h \leq \frac{\lambda_0}{32 \cos \theta_i}$$

# Multipath Propagation

Combination of all Wave Propagation Effects

# Propagation Phenomena



free space propagation:

- line of sight
- no multipath

reflection:

- plane wave reflection
- Fresnel coefficients

diffraction:

- knife edge diffraction

scattering:

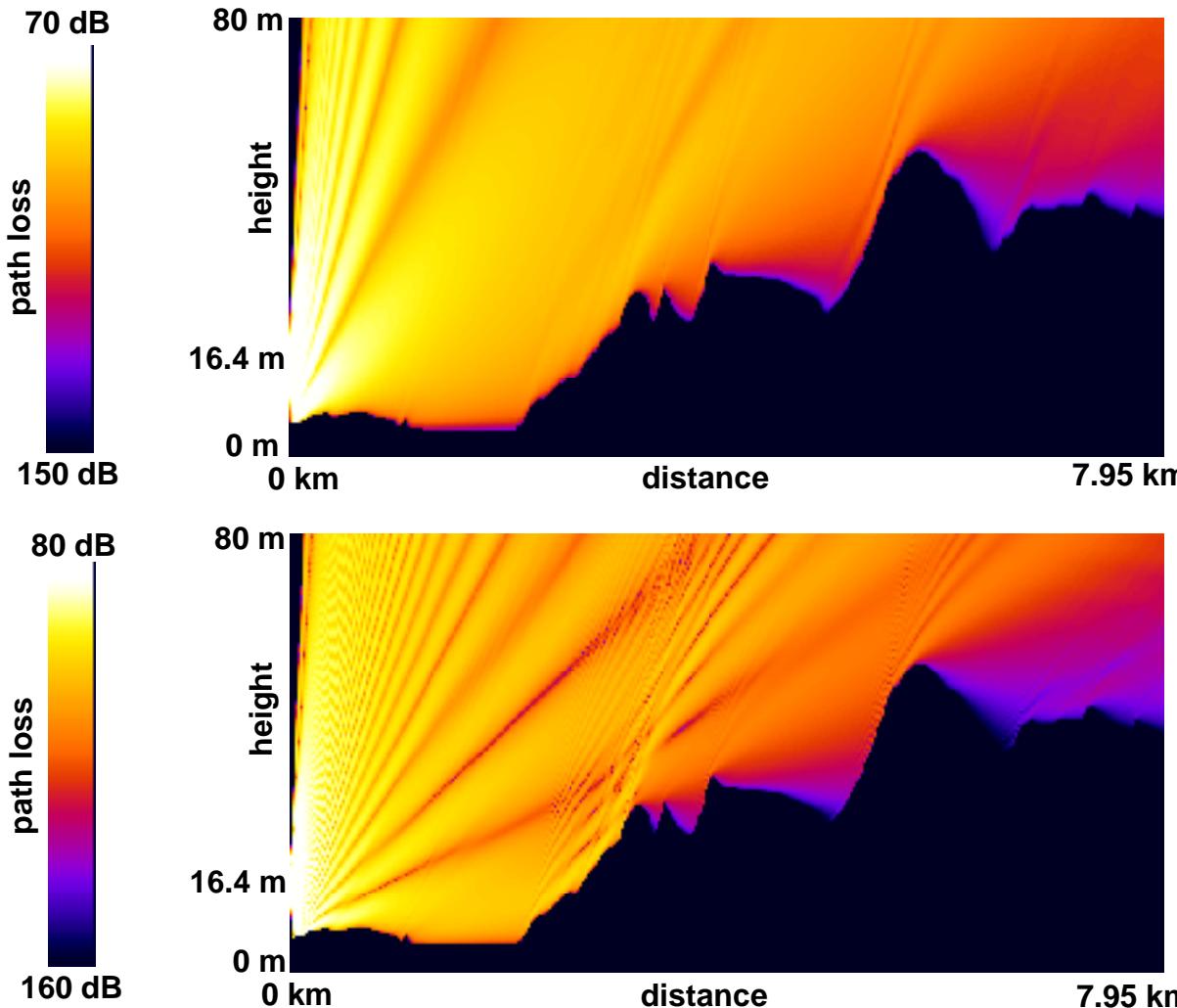
- rough surface scattering
- volume scattering

refraction in the troposphere:

- not considered

In general multipath propagation leads to fading at the receiver site

# Path Loss Prediction over Natural Terrain



- $f = 435 \text{ MHz}$

- Tx height = 16.4 m
- vertical polarization

- $f = 1900 \text{ MHz}$