

AIC HW4

110011222 陳立珩

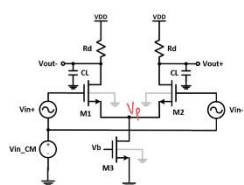
$$\text{NMOS: } I_{DS} = \frac{1}{2} \times \mu_n C_{OX} \times \frac{W}{L} (V_{GS} - |V_{th}|)^2 \quad \dots\dots \textcircled{1}$$

$$g_m = \frac{2 \cdot I_d}{V_{ov}} \dots\dots (2)$$

$$V_{ov} = \sqrt{2u_n C_{ox} \frac{W}{L} I_d} \quad \dots\dots (3)$$

PART1:

- I. 分析此電路的小訊號模型，得出 differential gain 為 $g_{mM1}(r_{oM1}(M2)||R_d)$ ，因此可知若要 $gain > 20$ ，則 g_{mM1} 要很大 $\Rightarrow I_{ss}$ 要大，i.e. W_{M3}/L_{M3} 要大 \Rightarrow choose $W_{M3}/L_{M3} = 25\mu m/1\mu m$ 。
- II. 若 V_b 太大，會使 $M1(M2)$ 的 source voltage 太大，造成 $M1(M2)$ 操作在不飽和區，因此使 V_b 能超過 $M3$ 的 V_{th} 就好，假設 $V_{th} = 0.4V \Rightarrow$ choose $V_b = 0.44V$ 。
- III. 若 V_{in_CM} 太大，根據電流①公式，為了維持電流相同，如果 V_{ov} 不變，代表 V_s 因此提高，會容易造成 $M3$ 操作在不飽和區，因此使 V_{in_CM} 能超過 $M3$ 的 V_{th} ，假設 $V_{th} = 0.4V \Rightarrow$ choose $V_{in_CM} = 0.5V$ 。
- IV. 因為 $M1(M2)$ 的 $r_{oM1(M2)}$ 與 gain 有關係，而且此題要求的 gain 很大，預估 R_d 會取很大，所以盡量讓 $r_{oM1(M2)}$ 更大來減少 $r_{oM1(M2)}||R_d$ 的值，根據 $hw1$ $r_o \propto L$ ，i.e. $L_{M1(M2)}$ 不能太小 choose $L_{M1(M2)} = 4\mu m$ 。
- V. 一開始假設讓 $M3$ 和 $M1(M2)$ 的 V_{ov} 相同，所以根據公式③， $V_{ovM3} = V_{ovM1(M2)}$ ，可以推得 $(\frac{W}{L})_{M3} = 2(\frac{W}{L})_{M1(M2)}$ ，取 $W_{M1(M2)} = 25\mu m$ ($m = 2$)。
- VI. 大概推算一下 $I_{ss} = \frac{1}{2}u_n C_{ox}(\frac{W}{L})_{M3} V_{ovM3}^2 = 9.68\text{ uA} \Rightarrow g_{mM1(M2)} = \frac{2(\frac{1}{2}I_{ss})}{V_{ovM1(M2)}} = 193\text{ uA}$ ，假設 $r_o \rightarrow \infty$ ， $Ad = g_{mM1(M2)}R_d > 20 \Rightarrow R_d = 100k\Omega$ 。
- VII. 驗證 mos region:



$$I_{SS} \approx 9.68 \text{ mA} \quad V_{TH} \approx 0.4 \text{ V} \quad \Delta n_{Cox} \approx 300 \times 10^{-6}$$

$$\frac{1}{2} I_{SS} = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L} \right)_{M1, M2} (V_{in-CM} - V_p - V_{TH})^2$$

$$V_{in_CM} - V_P - V_{TH} \approx 0.05 \text{ V}$$

$$\Rightarrow V_p \approx 0.05$$

$$\begin{cases} V_b > V_{TH} & \Rightarrow 0.44 > 0.4 \\ V_p > V_b - V_{TH} & \Rightarrow 0.05 > 0.44 - 0.4 \end{cases}$$

$$\therefore V_{in_CM} - V_p > V_{TH} \Rightarrow 0.5 - 0.05 > 0.4$$

$$V_{DD} - I_d R_d - V_P > V_{in,CM} - V_P - V_{TH} \quad \text{for } M_1 (M_2) \text{ on and sat.}$$

$$\Rightarrow 1,5 - 9,68_{44} \times 100k > 0,5 - 0,4$$

第 1 次測試:

Design parameter:

Vb	Vin_CM	W _{M3} /L _{M3}	W _{M1(M2)} /L _{M1(M2)}	Rd
0.44 V	0.5V	25um/1um	25 um/4um (m = 2)	100kΩ

Result:

subckt	0:mm1	0:mm2	0:mm3	v(voutp,voutn)/vip	= 16.9571
element	0:n 18.1	0:n 18.1	0:n 18.1	input resistance at vip	= 1.000e+20
region	Saturation	Saturation	Saturation	output resistance at v(voutp,voutn)	= 177.4316k
id	10.7036u	10.7036u	21.4073u	dcgain_in_db=	24.5870
ibs	-354.7274a	-354.7274a	-3.270e-21	at=	1.0000
ibd	-1.5775f	-1.5775f	-177.3596a	from=	1.0000
vgs	403.3866m	403.3866m	440.0000m	to=	100.0000g
vds	333.0237m	333.0237m	96.6134m	bw=	1.6703x
vbs	-96.6134m	-96.6134m	0.		
vth	349.4255m	349.4255m	387.9827m		
vdsat	84.2819m	84.2819m	88.7119m		
vod	53.9611m	53.9611m	52.0173m		
beta	3.7593m	3.7593m	7.7755m		
gam_eff	510.0843m	510.0843m	507.4460m		
gm	191.1409u	191.1409u	357.8438u		
gds	1.2720u	1.2720u	51.6959u		
gmb	37.3125u	37.3125u	74.5657u		
cdtot	71.7344f	71.7344f	45.7259f		
cgtot	1.2056p	1.2056p	161.5664f		
cstot	1.2176p	1.2176p	172.7527f		
cbtot	435.9652f	435.9652f	98.5710f		
cgs	1.0538p	1.0538p	133.7782f		
cgd	17.0470f	17.0470f	12.1195f		

觀察與微調:

M1、M2、M3 都進入了飽和區，且如預期三者的 V_{ov} 接近，很明顯可以發現 A_{DM} 還差一些，是因為 r_o 把 output resistance 分掉了，有兩個做法，第一個是將 R_d 繼續提高，但這樣做會讓 bw 降低，於是我選擇第二個方法，增加 $gm_{M1(M2)}$ ，根據公式②，如果將 $V_{ovM1(M2)}$ 降低就可以增加 $gm_{M1(M2)}$ 的值，再觀察公式①，可以發現如果提高 $(\frac{W}{L})_{M3}$ 的值，就可以降低 $V_{ovM1(M2)}$ ，試取 $(\frac{W}{L})_{M3} = \frac{50um}{4um}$ 。

第 2 次測試:

Design parameter:

Vb	Vin_CM	W _{M3} /L _{M3}	W _{M1(M2)} /L _{M1(M2)}	Rd
0.44 V	0.5V	25um/1um	50um/4um (m = 2)	100kΩ

Result:

subckt	0:mm1	0:mm2	0:mm3	v(voutp,voutn)/vip	= 19.8065
element	0:n 18.1	0:n 18.1	0:n 18.1	input resistance at vip	= 1.000e+20
region	Saturation	Saturation	Saturation	output resistance at v(voutp,voutn)	= 171.5873k
id	11.1959u	11.1959u	22.3918u	dcgain_in_db=	25.9362
ibs	-898.3297a	-898.3297a	-3.420e-21	at=	1.0000
ibd	-2.7380f	-2.7380f	-229.1217a	from=	1.0000
vgs	375.1906m	375.1906m	440.0000m	to=	100.0000g
vds	255.5985m	255.5985m	124.8094m	bw=	1.6170x
vbs	-124.8094m	-124.8094m	0.		
vth	355.0247m	355.0247m	387.7696m		
vdsat	67.9273m	67.9273m	88.8316m		
vod	20.1660m	20.1660m	52.2304m		
beta	7.5168m	7.5168m	7.7754m		
gam_eff	510.8295m	510.8295m	507.4460m		
gm	230.8634u	230.8634u	386.1202u		
gds	1.6559u	1.6559u	23.0819u		
gmb	44.8729u	44.8729u	80.2472u		
cdtot	144.6724f	144.6724f	41.5689f		
cgtot	2.0169p	2.0169p	160.0552f		
cstot	1.9074p	1.9074p	173.2041f		
cbtot	859.7789f	859.7789f	98.0280f		
cgs	1.6330p	1.6330p	133.4956f		
cgd	35.4268f	35.4268f	10.5633f		

觀察與微調:

發現 $V_{ovM1(M2)}$ 降低使得 A_{DM} 提高了，繼續調高 $(\frac{W}{L})_{M3}$ 的值。試取 $(\frac{W}{L})_{M3} = \frac{50um}{4um}$ 。

第 3 次測試:

Design parameter:

Vb	Vin_CM	W _{M3} /L _{M3}	W _{M1(M2)} /L _{M1(M2)}	Rd
0.44 V	0.5V	25um/1um	55um/4um (m = 2)	100kΩ

Result:

subckt	0:mm1	0:mm2	0:mm3	v(voutp,voutn)/vip	=	20.1189
element	0:n_18.1	0:n_18.1	0:n_18.1	input resistance at	vip	= 1.000e+20
model	0:n_18.1	0:n_18.1	0:n_18.1	output resistance at v(voutp,voutn)		= 170.7975k
region	Saturation	Saturation	Saturation	dcgain_in_db=	26.0721	at= 1.0000
id	11.2366u	11.2366u	22.4733u	from=	1.0000	to= 100.0000g
ibs	-1.0155f	-1.0155f	-3.433e-21	bw=	1.6042x	
ibd	-2.9741f	-2.9741f	-235.8914a			
vgs	371.5030m	371.5030m	440.0000m			
vds	247.8396m	247.8396m	128.4970m			
vbs	-128.4970m	-128.4970m	0.			
vth	355.7577m	355.7577m	387.7417m			
vdsat	66.0495m	66.0495m	88.8472m			
vod	15.7453m	15.7453m	52.2583m			
beta	8.2683m	8.2683m	7.7754m			
gam_eff	510.9262m	510.9262m	507.4460m			
gm	235.5891u	235.5891u	388.3616u			
gds	1.7098u	1.7098u	21.1271u			
gmb	45.7584u	45.7584u	80.6950u			
cdtot	159.1119f	159.1119f	41.2243f			
cgtot	2.1392p	2.1392p	159.9281f			
cstot	1.9924p	1.9924p	173.2585f			
cbtot	943.5317f	943.5317f	97.9681f			
cgs	1.7008p	1.7008p	133.4636f			
cgd	39.1976f	39.1976f	10.4400f			

觀察與微調:

發現已達到題目要求， $A_{DM} > 20(V/V)$ ， $bw > 1.5M(Hz)$

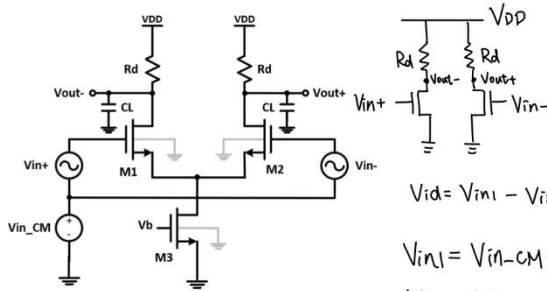
(b)

subckt	0:mm1	0:mm2	0:mm3
element	0:n_18.1	0:n_18.1	0:n_18.1
model	0:n_18.1	0:n_18.1	0:n_18.1
region	Saturation	Saturation	Saturation
id	11.2366u	11.2366u	22.4733u
ibs	-1.0155f	-1.0155f	-3.433e-21
ibd	-2.9741f	-2.9741f	-235.8914a
vgs	371.5030m	371.5030m	440.0000m
vds	247.8396m	247.8396m	128.4970m
vbs	-128.4970m	-128.4970m	0.
vth	355.7577m	355.7577m	387.7417m
vdsat	66.0495m	66.0495m	88.8472m
vod	15.7453m	15.7453m	52.2583m
beta	8.2683m	8.2683m	7.7754m
gam_eff	510.9262m	510.9262m	507.4460m
gm	235.5891u	235.5891u	388.3616u
gds	1.7098u	1.7098u	21.1271u
gmb	45.7584u	45.7584u	80.6950u
cdtot	159.1119f	159.1119f	41.2243f
cgtot	2.1392p	2.1392p	159.9281f
cstot	1.9924p	1.9924p	173.2585f
cbtot	943.5317f	943.5317f	97.9681f
cgs	1.7008p	1.7008p	133.4636f
cgd	39.1976f	39.1976f	10.4400f

(c)

****	small-signal transfer characteristics	
v(voutp,voutn)/vip		= 20.1189
input resistance at	vip	= 1.000e+20
output resistance at v(voutp,voutn)		= 170.7975k

(d)



$$V_{id} = V_{in1} - V_{in2} \quad V_{od} = (V_{out-}) - (V_{out+}) \quad A_d = \frac{V_{od}}{V_{id}}$$

$$V_{in1} = V_{in_CM} + V_{in+} \Rightarrow V_{out-} = A_{v1} \cdot V_{in1} = A_{v1} (V_{in_CM} + V_{in+})$$

$$V_{in2} = V_{in_CM} + V_{in-} \Rightarrow V_{out+} = A_{v2} \cdot V_{in2} = A_{v2} (V_{in_CM} + V_{in-})$$

$$\text{if } A_{v1} = A_{v2} = -g_m(R_D \parallel r_o) \\ \Rightarrow A_d = \frac{V_{od}}{V_{id}} = \frac{A_{v1} + A_{v2}}{2} = -g_m(R_D \parallel r_o)$$

$$= -235.5891 \times 10^{-6} \times (100k \parallel \frac{1}{1.7098 \times 10^{-6}}) = 20.118968 (\%)$$

$$\text{誤差} = \frac{20.1189 - 20.118968}{20.118968} \times 100\% = -3.416 \times 10^{-4}$$

(e)

```
***** ac analysis tnom= 25.000 temp= 25.000 *****
acm_in_db= -1.4903
```

$$\text{CMRR}_{\text{simulated}} = 20 \log(20.1189) - (-1.4903) = 27.562 \text{ dB}$$

(f)

g_{m1}	g_{m2}	R_D	r_{o1}	r_{o3}
$235.5891 \mu\Omega^{-1}$	$235.5891 \mu\Omega^{-1}$	$100k\Omega$	584863Ω	47332Ω
C_L	C_p	w		
1pF	4.026 pF	10kHz		

$R_D \parallel r_{o1} \parallel \frac{1}{C_L S}$	$r_{o3} \parallel \frac{1}{C_p S}$
85398.72Ω	47332.49Ω

$$|A_{c,CM}| = \frac{g_{m1} [R_D \parallel r_{o1} \parallel \frac{1}{C_L S}]}{(g_{m1} + g_{m2}) [r_{o3} \parallel \frac{1}{C_p S}] + 1} = \frac{235.5891 \cdot 10^{-6} \cdot 85398.72}{2 \cdot 235.5891 \cdot 10^{-6} \cdot 47332.49 + 1} = 0.86340126 \text{ (V/V)}$$

$$\text{CMRR}_{\text{hand}} = 20 \cdot \log \left(\left| \frac{A_{DM}}{A_{V,CM}} \right| \right) = 27.34786044 \text{ dB}$$

$$\text{CMRR}_{\text{error}} = \frac{\text{CMRR}_{\text{simulated}} - \text{CMRR}_{\text{hand}}}{\text{CMRR}_{\text{hand}}} \cdot 100\% = \left| \frac{27.562 \text{ dB} - 27.34786044 \text{ dB}}{27.34786044 \text{ dB}} \right| \cdot 100\% \\ = 0.783\%$$

(g)

```

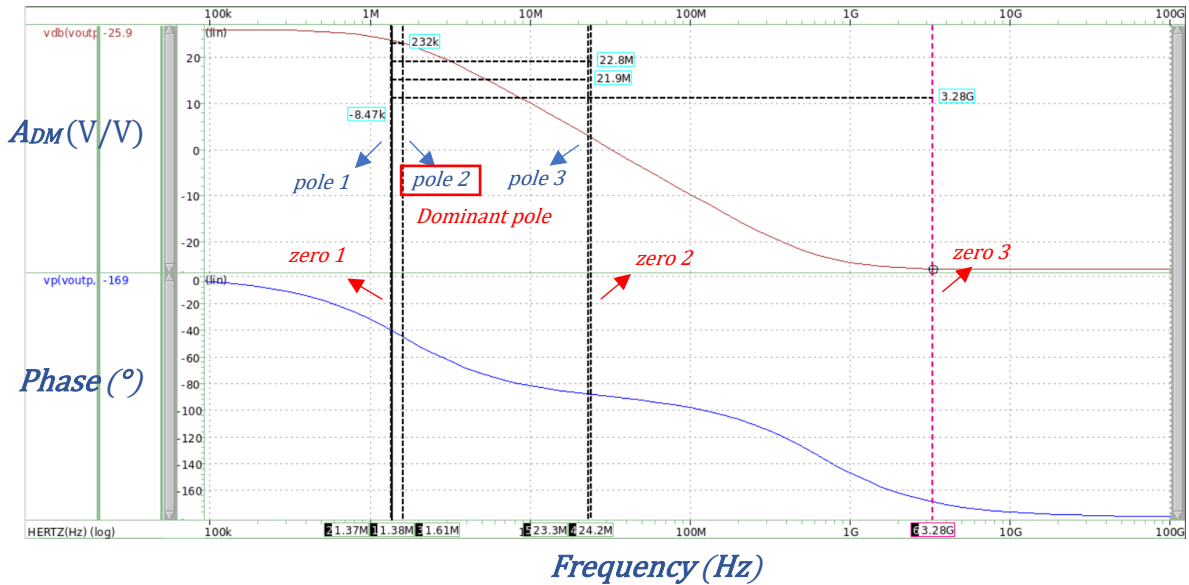
input = 0:vip          output = v(voutp,voutn)

      poles (rad/sec)          poles ( hertz)
real      imag      real      imag
-8.64647x    0.      -1.37613x    0.
-10.1024x    0.      -1.60784x    0.
-146.429x    0.      -23.3049x    0.

      zeros (rad/sec)          zeros ( hertz)
real      imag      real      imag
-8.59328x    0.      -1.36766x    0.
-151.986x    0.      -24.1892x    0.
20.6049g     0.      3.27937g     0.

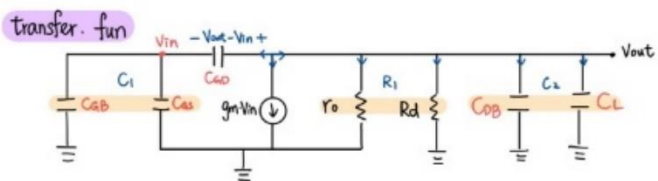
**** constant factor = 4.58096m

```



從圖中可以發現 pole 1 和 zero 1 的值非常相近，所以其實會互相抵銷，使得 dominant pole 出現在 pole 2 的位置也幾乎就是 -3dB bandwidth 的位置。

(h)



$$C_1 = C_{GB} \parallel C_{GS} = C_{GB} + C_{GS}$$

$$C_2 = C_{DB} \parallel C_L = C_{DB} + C_L$$

$$R_1 = r_o \parallel R_d$$

$$(V_{out} - V_{in}) S C_{GD} + g_m V_{in} + \frac{V_{out}}{R_1} + V_{out} \cdot S C_2 = 0$$

$$\Rightarrow V_{out} \cdot (sC_{GD} + \frac{1}{R_1} + sC_2) = V_{in} (sC_{GD} - g_m)$$

$$\Rightarrow \frac{V_{out}}{V_{in}} = \frac{sC_{GD} - g_m}{sC_{GD} + \frac{1}{R_1} + sC_2} = \frac{\overbrace{sC_{GD} - g_m}^{a-b}}{\underbrace{s(C_{GD} + C_2) + \frac{1}{R_1}}_c} = \frac{a-b}{c+s+d}$$

$$\text{Zero} = as - b = 0 \Rightarrow s = \frac{b}{a}$$

pole: $cs+d=0 \Rightarrow s = \frac{-d}{c}$

-3dB pole

$$\frac{\text{gain, -3dB}}{\text{gain, max}} = \frac{1}{\sqrt{2}} = \left| \frac{as+b}{cs+d} \right| \cdot \frac{1}{|\text{gain, max}|}$$

$$\Rightarrow \frac{1}{\sqrt{2}} |\text{gain, max}| = \sqrt{\frac{b^2 + w^2 a^2}{d^2 + w^2 c^2}} = A$$

$$\Rightarrow A^2 d^2 + w^2 \cdot A^2 c^2 = b^2 + w^2 a^2$$

$$\Rightarrow W^2(A^2c^2 - a^2) = b^2 - A^2d^2$$

$$\Rightarrow W = \left(\frac{b^2 - A^2 d^2}{A^2 c^2 - a^2} \right)^{\frac{1}{2}}$$

CL	CGS	CDB	CGD	C2
1E-12	1.7008E-12	1.199143 E-13	39.1976E-15	1.1199143E-12
R1	gm			
85398.55506	388.3616E-6			

$$a = C_{GD} \quad , \quad b = gm \quad , \quad c = C_{GD} + C_2 \quad , \quad d = \frac{1}{ro||Rd} \quad , \quad A = \frac{1}{\sqrt{2}} \cdot |gain_{max}|$$

a	b	c	d	A
3.91976E-14	2.355891E-4	1.1591119 E-12	1.17098E-5	14.22621062

from transfer function : $as + b = 0$

$$\Rightarrow s = \text{zero} = \frac{b}{a} = \frac{2.355891 \cdot 10^{-4}}{3.91976 \cdot 10^{-14}} = 6010293998 \left(\frac{\text{rad}}{\text{sec}} \right) = 956567999.2 \text{ (Hz)}$$

from transfer function : $cs + d = 0$

$$\Rightarrow s = \text{pole} = \frac{-d}{c} = \frac{-1.17098 \cdot 10^{-5}}{1.1591119 \cdot 10^{-12}} = 10099801.41 \left(\frac{\text{rad}}{\text{sec}} \right) = 1607433.318 \text{ (Hz)}$$

$$-3\text{dB BW} = \omega = \left(\frac{b^2 - A^2 d^2}{A^2 c^2 - a^2} \right)^{\frac{1}{2}} = \left(\frac{(2.355891 \cdot 10^{-4})^2 - (14.22621062)^2 (1.17098 \cdot 10^{-5})^2}{(1.1591119 \cdot 10^{-12})^2 - (3.91976 \cdot 10^{-14})^2} \right)^{\frac{1}{2}}$$

$$= 1.60786077 \text{ (MHz)}$$

	hand calculation	hspice	error
pole	1.607433318 MHz	1.60780 MHz	0.0228%
-3dB Bandwidth	1.607860770 MHz	1.60420 MHz	0.2275%

推測造成誤差的原因為電路中的寄生電容, 且因此式較為複雜, 故計算時, 各值的四捨五入也會造成最終計算結果有些許誤差。

(i)

480.0000m	1.000e+20	32.5407k	4.6728
481.0000m	1.000e+20	33.7857k	4.8002
482.0000m	1.000e+20	35.2950k	4.9565
483.0000m	1.000e+20	37.1470k	5.1503
484.0000m	1.000e+20	39.4535k	5.3939
485.0000m	1.000e+20	42.3765k	5.7053
486.0000m	1.000e+20	46.1569k	6.1108
487.0000m	1.000e+20	51.1587k	6.6508
488.0000m	1.000e+20	57.9240k	7.3853
489.0000m	1.000e+20	67.2311k	8.4012
490.0000m	1.000e+20	80.0127k	9.8036
491.0000m	1.000e+20	96.8021k	11.6562
492.0000m	1.000e+20	116.4050k	13.8342
493.0000m	1.000e+20	135.2069k	15.9430
494.0000m	1.000e+20	149.6557k	17.5865
495.0000m	1.000e+20	159.0365k	18.6756
496.0000m	1.000e+20	164.6168k	19.3418
497.0000m	1.000e+20	167.8247k	19.7376
498.0000m	1.000e+20	169.6156k	19.9661
499.0000m	1.000e+20	170.5210k	20.0843
500.0000m	1.000e+20	170.7975k	20.1189
501.0000m	1.000e+20	170.5210k	20.0761
502.0000m	1.000e+20	169.6156k	19.9441
503.0000m	1.000e+20	167.8247k	19.6884
504.0000m	1.000e+20	164.6168k	19.2398
505.0000m	1.000e+20	159.0365k	18.4750
506.0000m	1.000e+20	149.6557k	17.2115
507.0000m	1.000e+20	135.2069k	15.2908
508.0000m	1.000e+20	116.4050k	12.8138
509.0000m	1.000e+20	96.8021k	10.2457
510.0000m	1.000e+20	80.0127k	8.0532
511.0000m	1.000e+20	67.2311k	6.3859
512.0000m	1.000e+20	57.9240k	5.1710
513.0000m	1.000e+20	51.1587k	4.2857
514.0000m	1.000e+20	46.1576k	3.6283

```

subckt element 0:mm1 0:mm2 0:mm3
model 0:n 18.1 0:n 18.1 0:n 18.1
region
id Saturation Saturation Saturation
ids -977.5306a -977.5313a -3.416e-21
ibid -4.6628f -1.3704f -227.0741a
vgs 359.3060m 393.3060m 440.0000m
vds 466.3191m 49.7187m 123.6940m
vbs -123.6940m -123.6940m 0.
vth 353.9444m 355.5670m 387.7780m
vdsat 61.7662m 76.2651m 88.8260m
vov 5.3616m 37.7384m 52.2220m
beta 8.2628m 8.2752m 7.7754m
gam_eff 510.8002m 510.8002m 507.4460m
gm 197.8424u 218.3358u 385.3946u
gds 1.1075u 143.7415u 23.7291u
gmb 38.4543u 42.8474u 80.1021u
19.9661
cgtot 148.7160f 604.6153f 41.6795f
cstot 1.9376p 2.6641p 160.0960f
cbtot 1.7304p 2.4264p 173.1873f
cgs 935.7874f 978.3265f 98.8466f
cgs 1.4557p 2.1474p 133.5055f
cgd 38.3875f 225.3306f 10.6032f

```

```

**** small-signal transfer characteristics
v(voutp,voutn)/vip = 5.1503
input resistance at vip = 1.000e+20
output resistance at v(voutp,voutn) = 37.1472k

```

```

subckt element 0:mm1 0:mm2 0:mm3
model 0:n 18.1 0:n 18.1 0:n 18.1
region
id Saturation Saturation Saturation
ids -15.0011u -9.5710u -22.4322u
ibid -1.0005f -1.0005f -3.426e-21
ibid -1.5322f -4.4404f -232.4000a
vgs 385.4043m 361.4043m 440.0000m
vds 67.2907m 436.2994m 126.5957m
vbs -126.5957m -126.5957m 0.
vth 356.0825m 354.6499m 387.7501m
vdsat 72.1715m 62.3373m 88.8391m
vov 29.3218m 6.7635m 52.2430m
beta 8.2730m 8.2642m 7.7754m
gam_eff 510.8764m 510.8764m 507.4460m
gm 240.4363u 202.7995u 387.2352u
gds 63.9814u 1.2280u 22.1000u
gmb 46.9741u 39.3675u 80.4701u
19.9661
cgtot 351.4104f 149.6165f 41.3980f
cstot 2.4590p 1.9648p 159.9921f
cbtot 964.7501f 935.8226f 97.9908f
cgs 2.0086p 1.4593p 133.4799f
cgd 117.6992f 38.3247f 10.5020f

```

```

**** small-signal transfer characteristics
v(voutp,voutn)/vip = 5.1710
input resistance at vip = 1.000e+20
output resistance at v(voutp,voutn) = 57.9240k

```


(j)

$$FoM = \frac{total\ current(uA)}{input\ range(mV) \cdot -3dB\ bandwidth\ (MHz)} = \frac{22.4733\ (uA)}{29\ (mV) \cdot 1.6042\ (MHz)} = 0.4830703025$$

優化:

題目說明 $FoM = \frac{total\ current(uA)}{input\ range(mV) \cdot -3dB\ bandwidth\ (MHz)}$ 越小越好，根據測試發現

如果調小 V_{in_CM} ，可以大幅提升 input range 的範圍，但是一旦調太小有可能會使得 V_{GS} 變得比 V_{th} 小，mos 進入 cutoff，所以非常需要注意 $V_{ovM1(M2)}$ 的

值，不能因為為了讓 A_{DM} 增加，而不斷去調高 $(\frac{W}{L})_{M3}$ 的值，使得 $V_{ovM1(M2)}$ 太

小，而需要適時的調高 R_d 來達到 A_{DM} 增加的目的，當然這麼做就會降低 bw 的頻寬，所以需要作出取捨，同時我也注意到如果降低 I_{ss} ，雖然會降低 $g_{mM1(M2)}$ 造成需要更大的 R_d ，但 bw 下降的幅度比 I_{ss} 還多，也有助於降低 Fom。至於

A_{CM} 根據公式 $A_{C,CM} = \frac{g_{m1} [R_D \parallel r_{o1} \parallel \frac{1}{C_{LS}}]}{(g_{m1} + g_{m2}) [r_{o3} \parallel \frac{1}{C_{PS}}] + 1}$ 如果提高 r_{o3} 有助於降低 A_{CM} ，根

據 hw1 $r_o \propto L$ ， L_{M3} 不能取太小，且電阻不能太大，代表說為了要達到 $A_{DM} > 20(V/V)$ 的要求， I_D 不能太小。

大致流程:

1. 先大概取一個 L_{M3} 的值，取 $L_{M3}=3\mu m$ ，再跑看看 hspice 發現 V_{th} 大約等於 0.34V，所以取一個較接近的 V_b 值，取 $V_b=0.35\ V$ 。 V_{in_CM} 取 0.45V。
2. 經過幾次測試後發現電流不太能夠小於 20 uA，我取電流大約為 21uA，經過公式①計算後和實際跑 hspice 後大概推得 W_{M3}/L_{M3} 大約要等於 60，所以我設 $5.2\mu m/3\mu m(m=35)$ 。
3. 跑看看 hspice 發現電流不夠大，於是加大 V_b 值，取 0.352 V。
4. 經過幾次測試後發現如果 range 要 39mV 的話， $V_{odM1(M2)}$ 不能太小，要大於 15.6mV 左右。
5. 先隨意代一個 $W_{M1(M2)}/L_{M1(M2)}$ 值和 R_d 值，再來從 gain 和 4. 來決定最終值。
6. 跑看看 A_{CM} 發現值蠻大的，根據經驗如果升高 V_{in_CM} 和減少 $V_{odM1(M2)}$ 可以降低 A_{CM} ，但因為不希望動到電流，所以加大 V_{in_CM} ，大約要取到 $V_{in_CM}=0.4565V$ 。
7. 將 $V_{odM1(M2)}$ 控制到約 15.6mV 左右，發現 gain 超過 20V/V。
根據經驗發現降低電流減少 A_{DM} 比增加電阻可以得到更好的 Fom 值，所以我打算降低電流使得 A_{DM} 接近 20V/V。經過測試可以得到 $W_{M3}/L_{M3}=5.15\mu m/3\mu m(m=35)$ ，要同時注意 $V_{odM1(M2)}$ 有沒有小於約 15.6mV，不然到時候測 input range 會進到 cut off，所以在這個過程中需要

不斷嘗試 $W_{M1(M2)}/L_{M1(M2)}$ 和 R_d ，最後得到

$W_{M1(M2)}/L_{M1(M2)} = 14.6\mu\text{m}/4.4\mu\text{m}$ ($m = 8$)， $R_d = 101.5\text{k}\Omega$

8. 再跑看看 A_{CM} 發現值還大了一點點，所以稍稍提升 V_{in_CM} 值，
 $V_{in_CM} = 0.4566\text{V}$ 。
9. 重複回到步驟 7.，因為 V_{in_CM} 也會影響到 $V_{odM1(M2)}$ 。
10. 跑看看 input range，發現在邊界的地方 $M2$ 進到 linear，推測應該是電阻太大造成 V_{DS} 值過小，但觀察發現離飽和區不遠了，所以稍微減少 R_d 值，
取 $R_d = 100.6\text{k}\Omega$ 。
11. 重複步驟 7.、10.。
12. 經過多次嘗試後我取得 Fom 最小的設計。

第 4 次測試：

Design parameter

Vb	Vin_CM	W_{M3}/L_{M3}	$W_{M1(M2)}/L_{M1(M2)}$	R_d
0.352 V	0.4566V	5.166 $\mu\text{m}/3\mu\text{m}$ ($m=35$)	14.685 $\mu\text{m}/4.4\mu\text{m}$ ($m = 8$)	100.690k Ω

Result

subckt	element	$\theta:mm1$	$\theta:mm2$	$\theta:mm3$
model		$\theta:n_18.1$	$\theta:n_18.1$	$\theta:n_18.1$
region	Saturation	Saturation	Saturation	Saturation
id	10.7673u	10.7673u	21.5346u	
ibs	-828.6809a	-828.6809a	-3.790e-21	
ibd	-3.6873f	-3.6873f	-1.4297f	
vgs	363.1444m	363.1444m	352.0000m	
vds	322.3845m	322.3845m	93.4556m	
vbs	-93.4556m	-93.4556m	0.	
vth	347.4730m	347.4730m	339.9587m	
vdsat	65.1735m	65.1735m	63.3044m	
vod	15.6714m	15.6714m	12.0413m	
beta	8.0034m	8.0034m	18.0797m	
gam_eff	510.0001m	510.0001m	507.4459m	
gm	225.7648u	225.7648u	446.9981u	
gds	1.3566u	1.3566u	28.4525u	
gmb	44.5799u	44.5799u	93.6443u	
cdtot	167.0234f	167.0234f	341.5147f	
cgtot	2.5128p	2.5128p	2.6471p	
cstot	2.3345p	2.3345p	2.4190p	
cbtot	1.1000p	1.1000p	1.3501p	
cgs	1.9944p	1.9944p	2.0341p	
cgd	40.8139f	40.8139f	90.0559f	

$v(v_{outp}, v_{outn})/v_{ip}$	=	20.0001
input resistance at vip	=	1.000e+20
output resistance at $v(v_{outp}, v_{outn})$	=	177.1775k
dcgain_in_db=	26.0207	at= 1.0000
	from= 1.0000	to= 100.0000g
bw=	1.5355x	

觀察與微調：

發現達到題目要求， $A_{DM} > 20(\text{V/V})$ ， $bw > 1.5\text{M(Hz)}$ 。

Input range:

432.0000m	1.000e+20	37.6000k	5.0433
433.0000m	1.000e+20	38.5897k	5.1331
434.0000m	1.000e+20	39.8134k	5.2467
435.0000m	1.000e+20	41.3348k	5.3905
436.0000m	1.000e+20	43.2425k	5.5736
437.0000m	1.000e+20	45.6629k	5.8090
438.0000m	1.000e+20	48.7794k	6.1155
439.0000m	1.000e+20	52.0640k	6.5212
440.0000m	1.000e+20	58.3188k	7.0674
441.0000m	1.000e+20	65.7333k	7.8155
442.0000m	1.000e+20	75.8994k	8.8485
443.0000m	1.000e+20	89.6267k	10.2534
444.0000m	1.000e+20	107.0080k	12.0474
445.0000m	1.000e+20	126.1945k	14.0503
446.0000m	1.000e+20	143.5414k	15.8921
447.0000m	1.000e+20	156.3746k	17.2916
448.0000m	1.000e+20	164.6472k	18.2318
449.0000m	1.000e+20	169.6606k	18.8367
450.0000m	1.000e+20	172.6676k	19.2297
451.0000m	1.000e+20	174.4955k	19.4931
452.0000m	1.000e+20	175.6272k	19.6754
453.0000m	1.000e+20	176.3359k	19.8039
454.0000m	1.000e+20	176.7759k	19.8940
455.0000m	1.000e+20	177.0341k	19.9546
456.0000m	1.000e+20	177.1579k	19.9900
457.0000m	1.000e+20	177.1688k	20.0022
458.0000m	1.000e+20	177.0686k	19.9909
459.0000m	1.000e+20	176.8401k	19.9534
460.0000m	1.000e+20	176.4419k	19.8845
461.0000m	1.000e+20	175.7966k	19.7742
462.0000m	1.000e+20	174.7665k	19.6054
463.0000m	1.000e+20	173.1081k	19.3475
464.0000m	1.000e+20	170.3899k	18.9472
465.0000m	1.000e+20	165.0649k	18.3121
466.0000m	1.000e+20	158.3596k	17.2083
467.0000m	1.000e+20	146.5059k	15.7406
468.0000m	1.000e+20	129.9318k	13.6013
469.0000m	1.000e+20	110.8033k	11.1589
470.0000m	1.000e+20	92.8346k	8.8773
471.0000m	1.000e+20	78.3394k	7.0398
472.0000m	1.000e+20	67.5182k	5.6656
473.0000m	1.000e+20	59.6223k	4.6575

subckt	0:mm1	0:mm2	0:mm3
element	0:n 18.1	0:n 18.1	0:n 18.1
model	0:n 18.1	0:n 18.1	0:n 18.1
region	Saturation	Saturation	Saturation
id	7.7784u	13.5770u	21.3554u
lbs	-778.4346a	-778.4355a	-3.758e-21
ibd	-6.3559f	-1.1877f	-1.3431f
vgs	345.2110m	392.4110m	352.0000m
vds	629.0072m	45.1401m	87.7890m
vbs	-87.7890m	-87.7890m	0.
vth	345.2109m	347.3161m	339.9812m
vdscat	58.9548m	79.0165m	63.2952m
vod	96.1623n	45.0949m	12.0180m
beta	7.9956m	8.0132m	18.0798m
gam eff	509.8488m	509.8488m	507.4459m
gm	172.0187u	209.5318u	441.3017u
gds	938.6565m	181.5911u	35.0938u
gmb	34.0047u	41.8985u	92.4931u
cdtot	155.6146f	914.4145f	352.2538f
cgtpot	2.1570p	3.2971p	2.6516p
cstot	1.8597p	2.9894p	2.4169p
cbtot	1.0887p	1.1542p	1.3513p
cgs	1.5586p	2.6326p	2.0353p
cgd	41.2163f	351.5168f	94.2886f

subckt	0:mm1	0:mm2	0:mm3
element	0:n 18.1	0:n 18.1	0:n 18.1
model	0:n 18.1	0:n 18.1	0:n 18.1
region	Saturation	Saturation	Saturation
id	13.2747u	8.2260u	21.5007u
lbs	-818.3326a	-818.3318a	-3.784e-21
ibd	-1.4486f	-5.9563f	-1.4119f
vgs	379.7115m	348.9115m	352.0000m
vds	71.0811m	579.4372m	92.2885m
vbs	-92.2885m	-92.2885m	0.
vth	348.1452m	346.3042m	339.9633m
vdscat	72.3168m	59.9533m	63.3025m
vod	31.5663m	2.6073m	12.0367m
beta	8.0094m	7.9979m	18.0797m
gam eff	509.9690m	509.9690m	507.4459m
gm	244.2460u	180.4157u	445.9237u
gds	55.4178u	992.3078m	29.6805u
gmb	48.5001u	35.5882u	93.4274u
cdtot	385.9337f	156.9172f	343.5657f
cgtpot	2.9250p	2.2164p	2.6480p
cstot	2.7150p	1.9403p	2.4186p
cbtot	1.1274p	1.0886p	1.3503p
cgs	2.4072p	1.6326p	2.0343p
cgd	128.3238f	41.0091f	90.8615f

```
**** small-signal transfer characteristics
v(voutp,voutn)/vip = 5.1331
input resistance at vip = 1.000e+20
output resistance at v(voutp,voutn) = 38.5898k
```

```
**** small-signal transfer characteristics
v(voutp,voutn)/vip = 5.6656
input resistance at vip = 1.000e+20
output resistance at v(voutp,voutn) = 67.5182k
```

input range = 472(mV) - 433(mV) = 39(mV)

Acm:

```
***** ac analysis tnom= 25.000 temp= 25.000 *****
acm_in_db= 1.0104
```

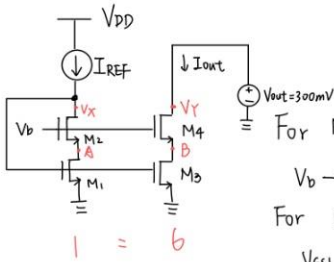
CMRR_{simulated} = 20 log(20.0001) - 1.0104 = 25.0103 dB > 25 dB

FoM:

$$\text{FoM} = \frac{\text{total current}(\mu\text{A})}{\text{input range}(\text{mV}) * -3\text{dB bandwidth}(\text{MHz})} = \frac{21.5346(\mu\text{A})}{39(\text{mV}) * 1.5355(\text{MHz})} = 0.35960223$$

PART2

(a)



For M_2 to be sat.

$$V_b - V_{TH2} \leq V_x (= V_{GS1})$$

For M_1 to be sat

$$V_{GS1} - V_{TH1} \leq V_A (= V_b - V_{GS2})$$

Thus

$$V_{GS2} + (V_{GS1} - V_{TH1}) \leq V_b \leq V_{GS1} + V_{TH2}$$

it exists if

$$V_{GS2} + (V_{GS1} - V_{TH1}) \leq V_{GS1} + V_{TH2}$$

$$\Rightarrow V_{GS2} - V_{TH2} \leq V_{TH1}$$

Let

$$V_{GS2} = V_{GS4} \Rightarrow V_b = V_{GS2} + (V_{GS1} - V_{TH1}) \\ = V_{GS4} + (V_{GS3} - V_{TH3})$$

$$\text{if } V_{GS2} = V_{GS4} \quad V_A = V_B$$

$$\Rightarrow V_{out, min} = V_B + V_{ov}$$

$$I_{REF} = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right)_1 (V_x - V_{TH1})^2 = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right)_2 (V_b - V_A - V_{TH2})^2$$

$$\text{Let } \left(\frac{W}{L}\right)_1 = \left(\frac{W}{L}\right)_2 \quad L_1 = L_2 \Rightarrow V_{TH1} \approx V_{TH2}$$

$$\Rightarrow V_x - V_{TH} = V_b - V_A - V_{TH}$$

$$\Rightarrow V_b = V_x + V_A = V_{GS2} + (V_{GS1} - V_{TH1}) = V_b - V_A + V_x - V_{TH}$$

$$\text{if we assume } V_x = 410 \text{ mV}, V_{TH} = 360 \text{ mV}$$

$$\Rightarrow V_A = 50 \text{ mV} \Rightarrow V_b = 460 \text{ mV}$$

For M_2 to be on, sat.

$$V_b - V_A \geq V_{TH2} \Rightarrow 460 \text{ mV} - 50 \text{ mV} \geq 360 \text{ mV}$$

$$V_b - V_{TH2} \leq V_x (= V_{GS1}) \Rightarrow 460 \text{ mV} - 360 \text{ mV} \leq 410 \text{ mV}$$

For M_1 to be on, sat

$$V_x \geq V_{TH1} \Rightarrow 410 \text{ mV} \geq 360 \text{ mV}$$

$$V_{GS1} - V_{TH1} \leq V_A (= V_b - V_{GS2}) \Rightarrow 410 \text{ mV} - 360 \text{ mV} \leq 50 \text{ mV}$$

I. 在 cascode stage 出現的電阻公式 R_{on} 可以推得 $R_{out} \cong (g_{m4} + g_{mb4})r_{o3}r_{o4}$ 但因為 r_o 不好去計算，所以換個方式觀察。

II. 因為此題的 spec 電阻要大於 700k，觀察 $R_{out} = \frac{V_{out}}{I_{out}}$ (at $V_{out} 300 \text{ mV}$)，

可以推斷如果 current mirror 越快達到 120 μA ，也就是如果能讓 M_4 在越

小的 V_{out} 下達到飽和區， $\frac{I_{out}}{V_{out}}$ 更小 $\Rightarrow R_{out}$ 更大。所以讓 M_4 的 V_{ov} 越

小越好 $\Rightarrow \frac{W_{M4}}{L_{M4}}$ 取大， $\frac{W_{M4}}{L_{M4}} = \frac{60 \mu\text{m}}{1 \mu\text{m}}$ ($m=5$)。

III. 因為 V_b 也會隨著 V_{out} 增加 (sweep from 0V to 1.8V) 升高到穩定值，如果能

讓 M_3 的 V_{ov} 越小 $\Rightarrow M_3$ 越快進入飽和 $\Rightarrow \frac{W_{M3}}{L_{M3}}$ 取大

$$, \frac{W_{M3}}{L_{M3}} = \frac{60 \mu\text{m}}{1 \mu\text{m}} \quad (m=5)。$$

IV. 根據 current mirror 和電流公式①，M1 與 M3 的 size 需成比例 1:6，且如果電流要準確，也需讓 V_A 、 V_B 同電壓，因此 M2 和 M4 的 size 也需成比

例 1:6，所以 $\frac{W_{M1}}{L_{M1}} = \frac{50\mu m}{1\mu m}$ 、 $\frac{W_{M2}}{L_{M2}} = \frac{50\mu m}{1\mu m}$ 。

第 1 次測試:

Design parameter: $\frac{W_{M4}}{L_{M4}} = \frac{W_{M3}}{L_{M3}} = \frac{60\mu m}{1\mu m}$ (m=5), $\frac{W_{M1}}{L_{M1}} = \frac{W_{M2}}{L_{M2}} = \frac{50\mu m}{1\mu m}$ (m=1)。

$V_b=460mV$

Result:

**** mosfets				***** dc transfer curves tnom= 25.000 temp= 25.000 *****			
subckt				deriv= 8.8870u			
element				rout= 112.5239k			
model				***** operating point information tnom= 25.000 temp= 25.000 *****			
region				***** operating point status is all simulation time is 0.			
id				node =voltage node =voltage node =voltage			
ibs				+0:a = 55.2105m 0:b = 54.3728m 0:vb = 460.0000m			
ibd				+0:vdd = 1.8000 0:vout = 300.0000m 0:vss = 0.			
vgs				+0:vx = 406.8730m			
vds							
vbs							
vth							
vdsat							
vod							
beta							
gam eff							
gm							
gds							
gmb							
cdtot							
cgtot							
cstot							
cbtot							
cgs							
cgd							

觀察與微調:

發現 Rout 還不夠大，猜測或許是因為 V_B 小使得 M3 隨著 V_B 增加，進入飽和區時的 V_{out} 值較接近 300mV，所以電流還未完全穩定，所以如果能稍微提高 V_b 的值，電流變化不大的情況下， V_B 的值就會提高，或許能讓 M3 早一些進入飽和區，電流更快在 $V_{out} = 300mV$ 時達到 120 uA。

取 $V_b=500mV$ 。

第 2 次測試:

Design parameter: $\frac{W_{M4}}{L_{M4}} = \frac{W_{M3}}{L_{M3}} = \frac{60\mu m}{1\mu m}$ (m=5), $\frac{W_{M1}}{L_{M1}} = \frac{W_{M2}}{L_{M2}} = \frac{50\mu m}{1\mu m}$ (m=1)。

$V_b=500mV$

Result:

subckt				***** dc transfer curves tnom= 25.000 temp= 25.000 *****			
element				deriv= 3.0858u			
model				rout= 324.0639k			
region				***** operating point information tnom= 25.000 temp= 25.000 *****			
id				***** operating point status is all simulation time is 0.			
ibs				node =voltage node =voltage node =voltage			
ibd				+0:a = 88.1251m 0:b = 87.0873m 0:vb = 500.0000m			
vgs				+0:vdd = 1.8000 0:vout = 300.0000m 0:vss = 0.			
vds				+0:vx = 399.2933m			
vbs							
vth							
vdsat							
vod							
beta							
gam eff							
gm							
gds							
gmb							
cdtot							
cgtot							
cstot							
cbtot							
cgs							
cgd							

觀察與微調:

與預期一樣，如果提高 V_B 的值，就可讓 M3 更早進入飽和區，電流在 $V_{out}=300mV$ 時提高了，代表電流更早穩定， R_{out} 更大，但和標準還差了一點，於是繼續加大 V_b 。

取 $V_b=550mV$ 。

第 3 次測試:

Design parameter: $\frac{W_{M4}}{L_{M4}} = \frac{W_{M3}}{L_{M3}} = \frac{60um}{1um}$ (m=5), $\frac{W_{M1}}{L_{M1}} = \frac{W_{M2}}{L_{M2}} = \frac{50um}{1um}$ (m=1)。

$V_b=550mV$

Result:

subckt	0:mm1	0:mm2	0:mm3	0:mm4
element	0:n_18.1	0:n_18.1	0:n_18.1	0:n_18.1
model	Saturation	Saturation	Saturation	Saturation
region	id	id	id	id
id	20.0000u	20.0000u	129.1429u	120.1429u
ibs	-2.994e-21	-465.8938a	-1.793e-20	-2.7595f
ibd	-465.8870a	-1.4286f	-2.7595f	-6.4560f
vgs	396.9599m	420.5430m	396.9599m	421.7696m
vds	129.4570m	267.5028m	128.2304m	171.7696m
vbs	0.	-129.4570m	0.	-128.2304m
vth	387.3335m	412.9554m	387.2757m	413.4209m
vdsat	67.9430m	68.0482m	68.0100m	68.4034m
vod	9.6264m	7.5875m	9.6841m	8.3488m
beta	15.5335m	15.5868m	93.2040m	93.5274m
gam_eff	507.4459m	510.9513m	507.4459m	510.9192m
gm	416.0795u	423.0616u	2.4983m	2.5304m
gds	13.3167u	5.7736u	81.8960u	49.4711u
gmb	87.3477u	81.8969u	524.5171u	491.2481u
cdtot	77.1896f	70.3553f	463.5270f	435.8225f
cgtot	252.2189f	243.8267f	1.5141p	1.4765p
cstot	259.2146f	249.0509f	1.5500p	1.5051p
cbtot	194.6846f	181.5247f	1.1681p	1.0993p
cgs	185.6486f	179.1737f	1.1147p	1.0899p
cgd	19.3128f	18.4694f	116.0239f	112.5260f

```
***** dc transfer curves tnom= 25.000 temp= 25.000 *****
deriv= 1.3197u
rout= 757.7500k
```

```
***** operating point information tnom= 25.000 temp= 25.000 *****
***** operating point status is all simulation time is 0.
node =voltage node =voltage node =voltage
+0:a = 129.4570m 0:b = 128.2304m 0:vb = 550.0000m
+0:vdd = 1.8000 0:vout = 300.0000m 0:vss = 0.
+0:vx = 396.9599m
```

觀察與微調:

達到題目要求， $R_{out} > 700k\Omega$ 。

計算驗證:

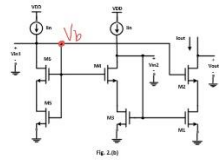
$R_{out} \cong (g_{m4} + g_{mb4})r_{o3}r_{o4}$

$$= (2.5304 \cdot 10^{-3} + 491.2481 \cdot 10^{-6}) \cdot \left(\frac{1}{81.8960 \cdot 10^{-6}} \right) \cdot \left(\frac{1}{49.4711 \cdot 10^{-6}} \right)$$

$$= 745813.9558 \Omega$$

$$R_{out \text{ error}} = \left| \frac{757750 - 745813.9558}{745813.9558} \right| \cdot 100\% = 1.6\%$$

(b)



$$\text{Assume } \left(\frac{W}{L}\right)_{M1} = \left(\frac{W}{L}\right)_{M2} = 6 \left(\frac{W}{L}\right)_{M3} = 6 \left(\frac{W}{L}\right)_{M4}$$

$$\left(\frac{W}{L}\right)_{M1, M2} = \frac{1}{4} \left(\frac{W}{L}\right)_{M5, 6}$$

$$(V_{GS} - V_{TH})_{M5, 6} = 2(V_{GS} - V_{TH})_{M3}$$

$$\Rightarrow V_{in1} = V_b = V_{GS, M5, 6} = 2V_{ov, M3} + V_{TH, M5, 6} = 2V_{ov} + V_{TH}$$

$$V_{in2} = V_{GS3} = V_{TH, M3} + V_{ov, M3} = V_{TH} + V_{ov}$$

$$V_{out} = V_{DS1} + V_{DS2} = V_{ov, M1} + V_{ov, M2}$$

$$\Rightarrow V_{out} = 2V_{ov, M2} = 2V_{ov}$$

V_{DD}

by (a) we can get $V_{ov} \approx 9.6mV$ and $V_{GS, M2} \approx 420mV$

$$V_b = 2V_{ov} + V_{TH} = V_{ov} + V_{GS, M2} \approx 430mV \dots \dots \textcircled{1}$$



$$I_{IN} = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right)_{M5} (V_b - V_{TH, M5})^2 = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right)_{M6} (V_b - V_{TH, M6})^2$$

$$\frac{1}{4n} (V_b - V_{TH, M5})^2 = \frac{1}{n} (V_b - V_{TH, M6})^2 \dots \dots \textcircled{2}$$

$$\begin{cases} V_c \geq V_b - V_{TH, M5} & (\text{for } M5 \text{ sat}) \\ V_b \geq V_{TH, M5} & (\text{for } M5 \text{ on}) \end{cases}$$

$$\begin{cases} V_b - V_c \geq V_b - V_c & (\text{for } M6 \text{ sat}) \\ V_b - V_c \geq V_{TH, M6} & (\text{for } M6 \text{ on}) \end{cases}$$

$\dots \dots \textcircled{3}$

by $\textcircled{1} \textcircled{2} \textcircled{3}$ and (a) 取 $V_b = 470mV$ $V_c = 30mV$

假設 $L_{M5} < L_{M6} \Rightarrow V_{TH, M5} = 440mV$

$V_{TH, M6} = 340mV$

$$\Rightarrow \frac{1}{4-n} (470mV - 440mV)^2 = \frac{1}{n} (470mV - 30mV - 340mV)^2$$

$\Rightarrow n \approx 3.5$ (驗證符合一開始假設 $L_{M5} < L_{M6}$)

by $\left(\frac{W}{L}\right)_{M1, M2} = \frac{1}{4} \left(\frac{W}{L}\right)_{M5, 6}$, $\Rightarrow W = 50\mu m$

take $W_{M5} = 50\mu m$ $L_{M5} = 3.5\mu m$

$W_{M6} = 50\mu m$ $L_{M6} = 0.5\mu m$

第 1 次測試:

Design parameter: $\frac{W_{M4}}{L_{M4}} = \frac{W_{M3}}{L_{M3}} = \frac{60\mu m}{1\mu m}$ ($m=5$), $\frac{W_{M1}}{L_{M1}} = \frac{W_{M2}}{L_{M2}} = \frac{50\mu m}{1\mu m}$ ($m=1$)。

$\frac{W_{M5}}{L_{M5}} = \frac{50\mu m}{3.5\mu m}$ ($m=1$), $\frac{W_{M6}}{L_{M6}} = \frac{50\mu m}{0.5\mu m}$ ($m=1$)。

Result:

subckt	0:mm1	0:mm2	0:mm3	0:mm4	0:mm5	0:mm6	***** operating point information tnom= 25.000 temp= 25.000 *****
element	0:n_18.1	0:n_18.1	0:n_18.1	0:n_18.1	0:n_18.1	0:n_18.1	***** operating point status is all simulation time is 0. *****
region	Saturation	Saturation	Saturation	Saturation	Subth	Saturation	node =voltage node =voltage node =voltage
id	118.9567u	118.9567u	20.0000u	20.0000u	20.0000u	20.0000u	+0:a = 44.4601m 0:b = 43.7067m 0:c = 26.4895m
ibs	-1.775e-20	-940.5868a	-2.994e-21	-160.0061a	-2.994e-21	-95.3333a	+0:vb = 446.8642m 0:vdd = 1.8000 0:vout = 300.0000m
ibd	-940.5513a	-6.4560f	-160.0001a	-1.4878f	-95.3273a	-1.6082f	+0:vss = 0. 0:vx = 413.4145m
vgs	413.4145m	403.1574m	413.4145m	402.4041m	446.8642m	420.3747m	
vds	43.7067m	256.2933m	44.4601m	368.9544m	26.4895m	420.3747m	
vbs	0.	-43.7067m	0.	-44.4601m	0.	-26.4895m	
vth	387.9146m	395.5579m	387.9760m	394.9095m	447.5388m	337.1020m	
vdsat	75.1679m	67.4777m	75.0914m	67.3980m	70.2272m	101.1466m	
vod	25.4999m	7.5995m	25.4385m	7.4946m	-674.5779u	83.2728m	
beta	93.2515m	93.3099m	15.5415m	15.5503m	32.3998m	4.3011m	
gam_eff	507.4459m	508.6564m	507.4459m	508.6771m	507.4459m	508.1831m	
gm	1.9916m	2.5107m	335.9836u	422.1281u	348.1071u	309.5045u	
gds	1.6363m	34.0540u	266.6457u	5.0378u	575.1132u	2.0863u	
gmb	421.7711u	511.6096u	71.1344u	85.7675u	70.1558u	62.1721u	
cdtot	710.9319f	431.7491f	117.2022f	69.9311f	96.7734f	70.7775f	
cgtot	1.7814p	1.4737p	296.4363f	245.1614f	135.7262f	1.1223p	
cstot	1.7891p	1.5144p	298.0257f	252.2242f	149.5446f	1.1642p	
cbtot	1.1945p	1.1293p	199.0480f	186.3564f	155.1980f	404.7903f	
cgs	1.3472p	1.0761p	224.4894f	178.8085f	85.0151f	998.9430f	
cgd	210.4275f	111.2369f	34.5368f	18.4537f	26.6011f	17.0002f	

觀察與微調:

發現 V_b 比預估的值還要小，所以造成 M_5 進入 cut off，要讓 V_b 的值大一些，可以發現 V_b 其實就是 M_5 的 V_G ，觀察公式①如果要增加 V_b 的值，可以減少 $\frac{W_{M5}}{L_{M5}}$ ，因為變動到 L 就會同時影響到 V_{th} ，所以我打算調小 W 來降低 $\frac{W_{M5}}{L_{M5}}$ 。

試取 $\frac{W_{M5}}{L_{M5}} = \frac{40\mu m}{3.5\mu m}$ ，因為 M_5 和 M_6 要 W 要相同，所以 $\frac{W_{M6}}{L_{M6}} = \frac{40\mu m}{0.5\mu m}$ 。

第 2 次測試:

Design parameter: $\frac{W_{M4}}{L_{M4}} = \frac{W_{M3}}{L_{M3}} = \frac{60\mu m}{1\mu m}$ ($m=5$)， $\frac{W_{M1}}{L_{M1}} = \frac{W_{M2}}{L_{M2}} = \frac{50\mu m}{1\mu m}$ ($m=1$)。

$\frac{W_{M5}}{L_{M5}} = \frac{40\mu m}{3.5\mu m}$ ($m=1$)， $\frac{W_{M6}}{L_{M6}} = \frac{40\mu m}{0.5\mu m}$ ($m=1$)。

Result:

subckt	0:mm1	0:mm2	0:n:18.1	0:mm3	0:mm4	0:mm5	0:mm6	***** operating point information tnom= 25.000 temp= 25.000 *****
element	0:n:18.1	0:n:18.1	0:n:18.1	0:n:18.1	0:n:18.1	0:n:18.1	0:n:18.1	***** operating point status is all simulation time is 0.
model	Saturation	Saturation	Saturation	Saturation	Saturation	Saturation	Saturation	node =voltage node =voltage node =voltage node =voltage
region	id	id	id	id	id	id	id	+0:a = 56.2369m 0:b = 55.3915m 0:c = 25.9627m
id	119.3818u	119.3818u	20.0000u	20.0000u	20.0000u	20.0000u	20.0000u	+0:vb = 461.2516m 0:vdd = 1.8000 0:vout = 300.0000m
ibs	-1.781e-20	-1.1920f	-2.994e-21	-202.3884a	-3.009e-21	-75.1287a	-1.3347f	+0:vss = 0. 0:vx = 406.4158m
ibd	-1.1920f	-6.4560f	-202.3825a	-1.4626f	-75.1227a	-1.3347f	-1.3347f	
vgs	406.4158m	405.8601m	406.4158m	405.0147m	461.2516m	435.2889m	435.2889m	
vds	55.3915m	244.6085m	56.2369m	350.1788m	25.9627m	435.2889m	435.2889m	
vbs	0.	-55.3915m	0.	-56.2369m	0.	-25.9627m	-25.9627m	
vth	387.8263m	398.0781m	387.8870m	397.4973m	447.6499m	337.0325m	337.0325m	
vdsat	71.9307m	67.6426m	71.8588m	67.4966m	76.3698m	110.4754m	110.4754m	
vod	18.5895m	7.7820m	18.5288m	7.5174m	13.6017m	98.2564m	98.2564m	
beta	93.2321m	93.3419m	15.5382m	15.5557m	25.9232m	3.4432m	3.4432m	
gam_eff	507.4459m	508.9752m	507.4459m	508.9982m	507.4459m	508.1685m	508.1685m	
gm	2.1712m	2.5187m	364.9194u	422.2780u	321.3872u	287.2352u	287.2352u	
gds	980.3291u	35.1740u	158.4394u	5.1258u	606.9979u	1.9690u	1.9690u	
gmb	458.7350u	509.7144u	77.0801u	85.2167u	64.7027u	57.4187u	57.4187u	
cdtot	590.2004f	432.0342f	97.6706f	70.0563f	86.0848f	56.6172f	56.6172f	
cgtot	1.6684p	1.4748p	277.7324f	244.9788f	122.1807f	910.0527f	910.0527f	
estot	1.6878p	1.5144p	281.1760f	251.7912f	134.7305f	947.7303f	947.7303f	
cbtot	1.1864p	1.1249p	197.7124f	185.7178f	124.7011f	323.3170f	323.3170f	
cgs	1.2570p	1.0791p	209.3749f	178.8849f	81.1418f	814.4899f	814.4899f	
cgd	162.6341f	111.2918f	26.8170f	18.4535f	24.6231f	13.6124f	13.6124f	

觀察與微調:

V_b 的值與預估的一樣提高了，所有 mos 都進入飽和區。

$I_{out}=119.33818 \text{ uA}$ 約等於 $6 \cdot I_{in} = 120\text{uA}$ 。