### **IIT Bombay**

# Makerspace (MS101)

2023 (Autumn) EE-Lecture-10

Using Bipolar Junction Transistor (BJT) and Metal Oxide Field Effect Transistor (MOSFET) as Switch

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### Transistors: Introduction

A transistor is a semiconductor device with three terminals, used in analog & digital applications. It can be modelled as a 'two-port network', with 'input-dependent output variable'.

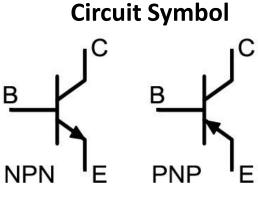
### **Commonly used transistors**

(a) Bipolar Junction Transistor (BJT)

**(b)** Metal Oxide Field Effect Transistor (MOSFET)

**Bipolar Junction Transistor** 

**MOSFET** (Enhancement mode)



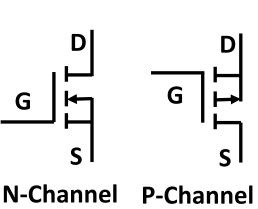
B: Base C: Collector E: Emitter β: Current gain

G: Gate

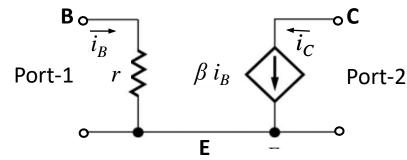
D: Drain

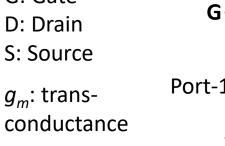
S: Source

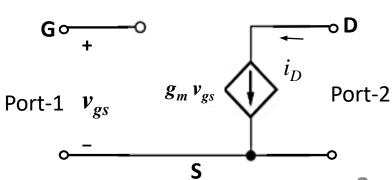
 $g_m$ : trans-



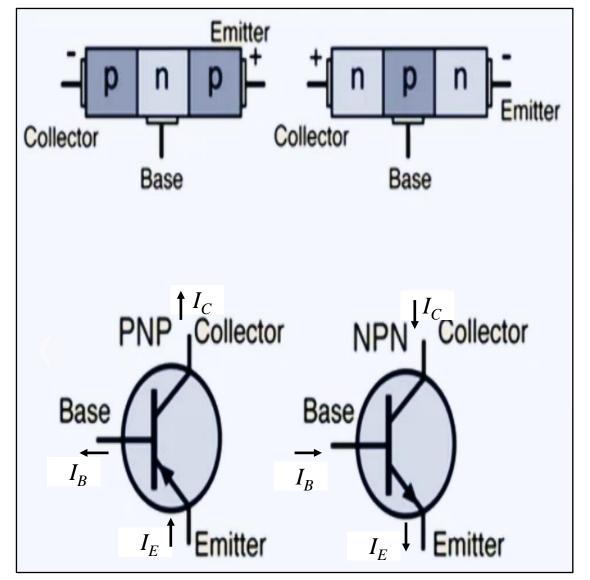
### **Simplified Small-Signal Model**







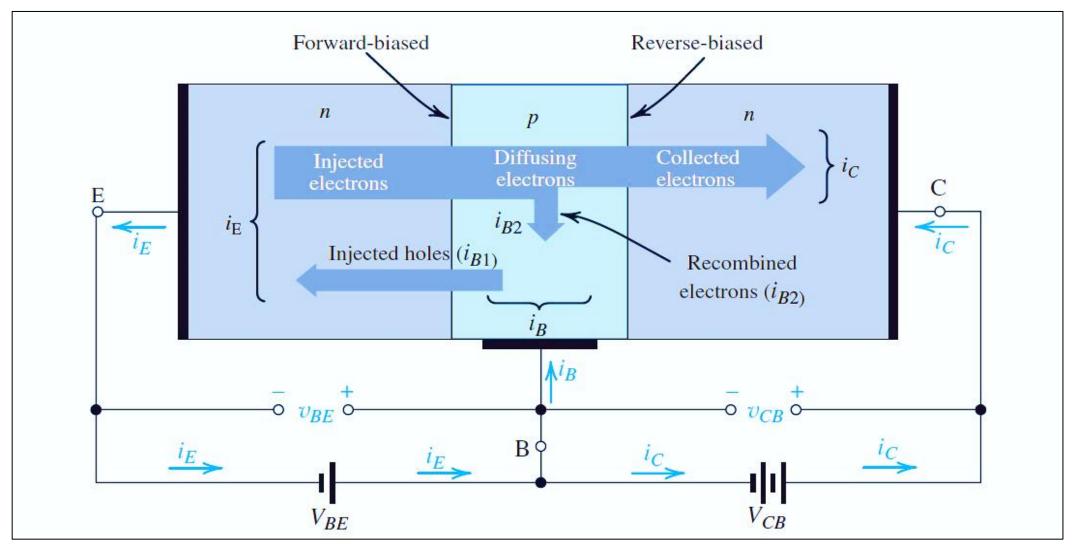
# Bipolar Junction Transistor (BJT)



Types	NPN, PNP
Terminals	emitter (E), base (B), collector (C)
Controlling quantity	Current-controlled device: Base Current ( $I_B$ ) controls Collector Current ( $I_C$ ).

BJTs have lower input resistance ( $\approx 10 \text{ k}\Omega$  to  $1 \text{ M}\Omega$  in common-emitter mode) and hence consume more power from the input signal source than MOSFETs.

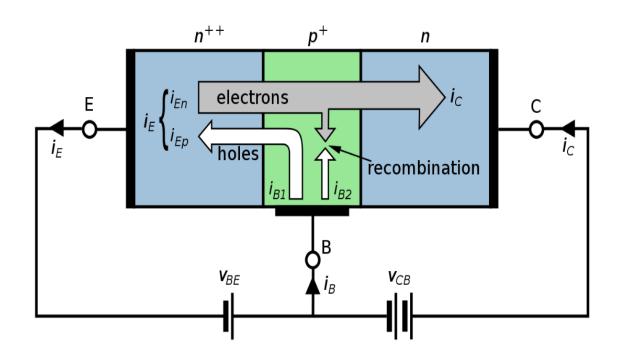
## Bipolar Junction Transistor Operation



NPN transistor biased in active mode (BE junction: forward biased, BC junction: reversed biased)

Source: Fig 6.3:-Sedra A S, Smith K C, "Microelectronic Circuits", Oxford University Press, 7Ed. ISBN: 9780199339136

# **BJT** Operation



### **Cricket analogy**

- The emitter is the bowler who shoots balls (electrons) toward the base.
- The base is the batsman, a tail-ender who swings away but connects with only 1-2% of the incoming balls (electrons).
- Most of the balls (electrons) are collected by the wicketkeeper.

### **Current relations**

$$KCL: I_E = I_B + I_C$$

Transistor action

$$\Delta I_C = \alpha \Delta I_E \ (\alpha \approx 0.90 - 0.99)$$

$$\Delta I_C = \beta \Delta I_B$$

 $I_C = \beta I_B$  is used as an approximate relation.

### Relation between $\alpha \& \beta$

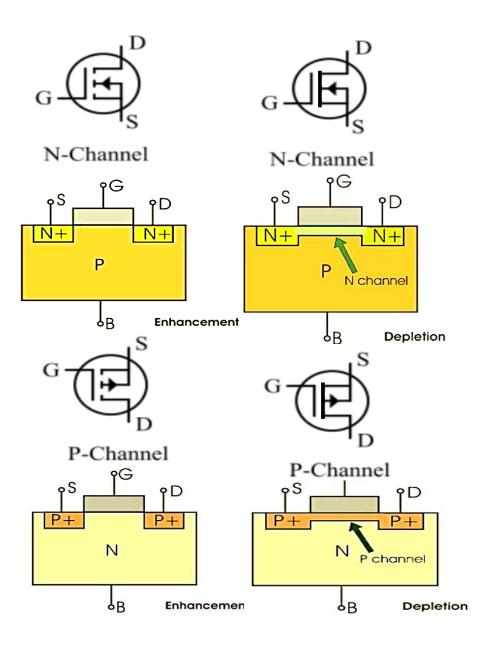
$$\Delta I_E = \Delta I_B + \Delta I_C \implies \Delta I_B = \Delta I_E - \Delta I_C = (1/\alpha - 1) \Delta I_C$$

$$\Rightarrow \Delta I_C = \alpha/(1-\alpha) \Delta I_B = \beta \Delta I_B$$

$$\Rightarrow \beta = \alpha/(1-\alpha)$$

Example: 
$$\alpha = 0.98 \Rightarrow \beta = 0.98/0.02 = 49$$

### MOSFETs: Introduction

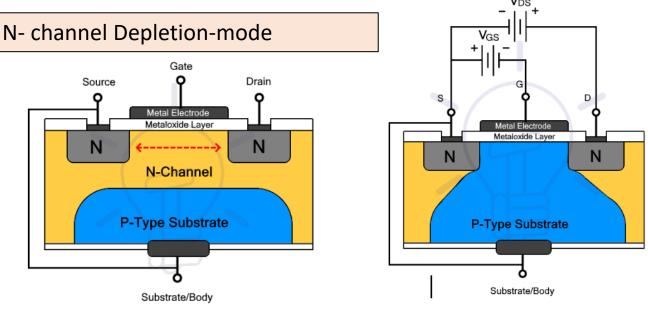


Types: N-channel & P-channel, fabricated in two modes					
(enhancement & d figures.	epletio	n) as sche	matically	shown in	
				•	

Types	N-channel enhancement-mode P-channel enhancement-mode N-channel depletion-mode P-channel depletion-mode
Terminals	source (S), drain (D), gate (G), substrate or body or bulk (B, often connected to S.
Controlling quantity	Voltage-controlled device: Gate-Source voltage ( $V_{GS}$ ) controls drain current ( $I_D$ )

MOSFETs have high input resistance (10 M $\Omega$  to 100 M $\Omega$ ) and hence consume less power from the input signal source than BJTs.

## MOSFET Operation



### **Device ON/OFF operation**

ON: Low resistance for drain-source current flow.

OFF: Very high resistance for drain-source current flow.

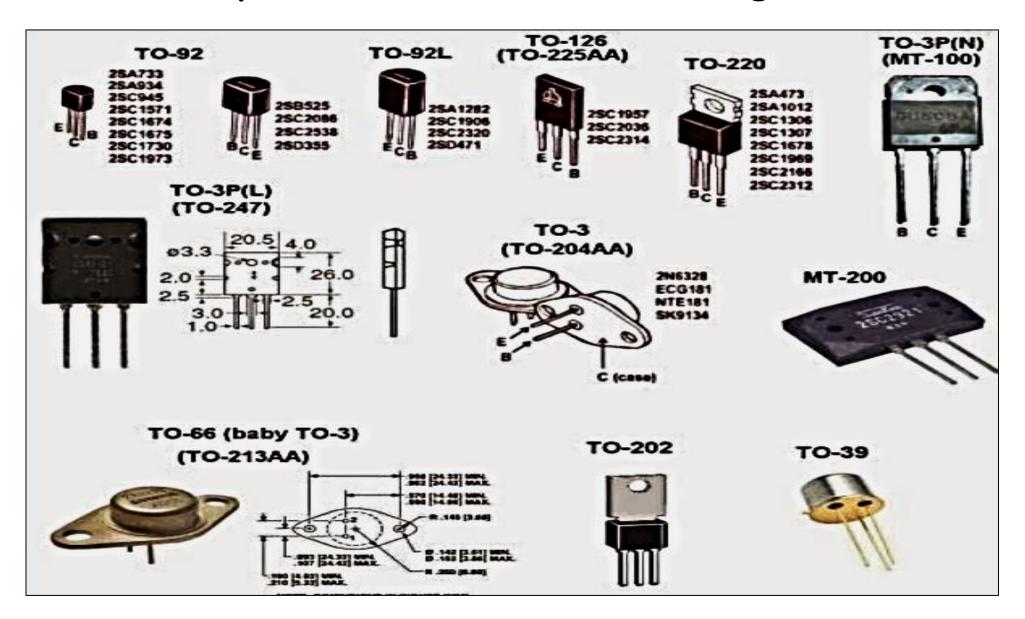
**Depletion mode**: Channel for drain-source current flow is available at  $V_{GS}$  = 0. It gets depleted for negative  $V_{GS}$ , and it is totally depleted for  $V_{GS}$  below the threshold voltage  $V_{\tau}$  (-ve for N-channel).

**Enhancement mode:** No channel is available at  $V_{GS} = 0$ . The channel is formed for  $V_{GS}$  above the threshold voltage  $V_T$  (+ve for N-channel).

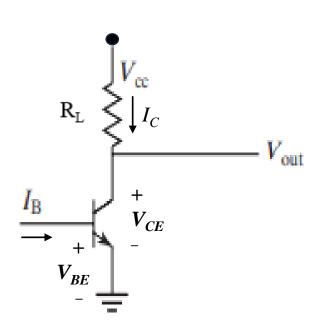
Channel-enhancement-mode	<u>-</u>     <del> </del>
P-Type Substrate  Substrate/Body Substrate/Body Pe: https://www.electricaltechnology.org/2021/06/mosfet.h	Metal Electrode Metaloxide Layer N + + + + + + N N-Channel  P-Type Substrate  Substrate/Body

MOSFET Type	Condition for Switching
N-channel Enhancement	OFF for $V_{GS} < V_T$ ( 0.6 to 1 V)
N-channel Depletion	OFF for $V_{GS} < V_T$ (—1.3 to 0.0 V)
P-channel Enhancement	OFF for $V_{GS} > V_T$ (—1.0 to 0.6 V)
P- channel Depletion	OFF for $V_{GS} > V_T$ ( 0.0 to 1.3 V)

## Commercially Available Transistor Packages



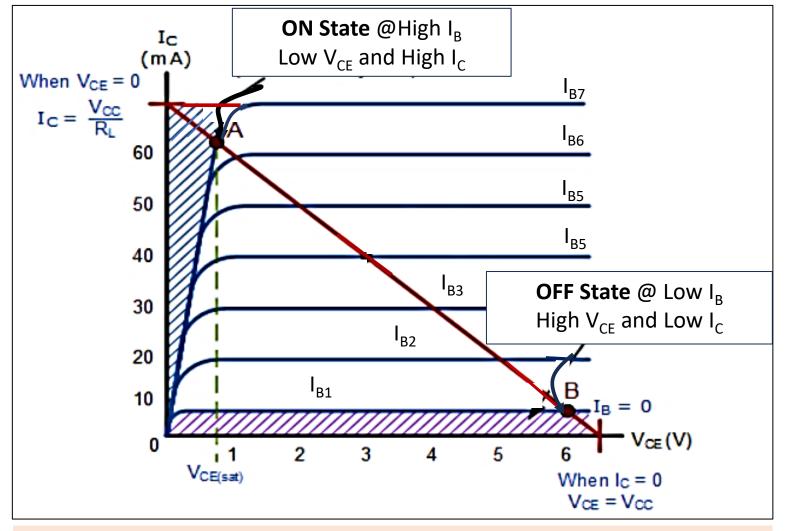
### Switching Characteristics of BJT (on $V_{CE}$ - $I_C$ Plane)



BJT is switched ON / OFF by changing the base current  $I_B$ .

OFF: 
$$V_{BE} < V_{\gamma}$$
.  $(V_{\gamma} \approx 0.5 \text{ V})$   
 $I_{B} \approx 0$ .  $V_{CE} \approx V_{CC}$ .  $I_{C} \approx 0$ .

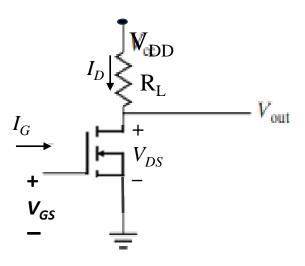
ON: 
$$I_B > I_C / \beta$$
.  
 $V_{BE} = V_{BES}$ .  $V_{CE} = V_{CES}$ .  
 $I_C = (V_{CC} - V_{CES}) / R_L$   
 $(V_{BES} \approx 0.8 \text{ V}. V_{CES} \approx 0.2 \text{ V})$ 



Typical ' $V_{CE}$  -  $I_C$ ' characteristics for BJT (base currents:  $I_{B1} < I_{B2} < I_{B3} < I_{B4} < I_{B5} < I_{B6} < I_{B7}$ ). Operating points for switching action: A (on) & B (off).

# Switching Characteristics of N-Channel Enhancement-Mode MOSFET

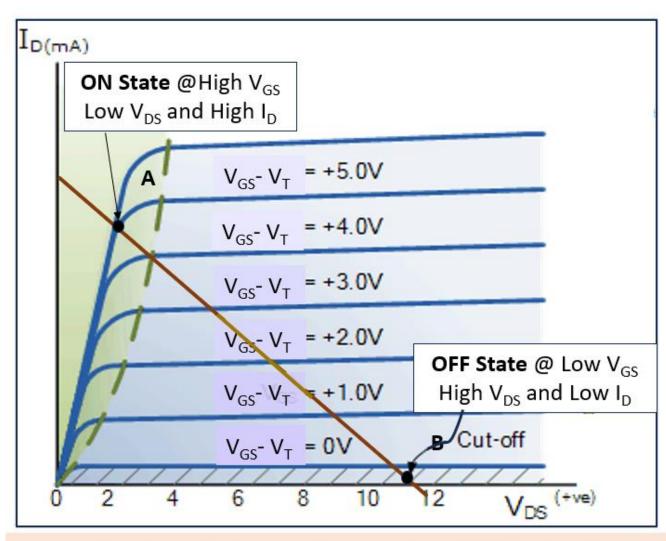
(on V<sub>DS</sub>-I<sub>D</sub> Plane)



- Enhancement MOSFET starts conducting when the drain-source channel is formed.
- N-Channel MOSFET: N-channel is formed.
- $V_T$  is the minimum value of  $V_{GS}$  for the channel formation.  $V_{GS} > V_T$  for MOSFET to be ON. Channel resistance decreases as  $V_{GS}$  increases further.

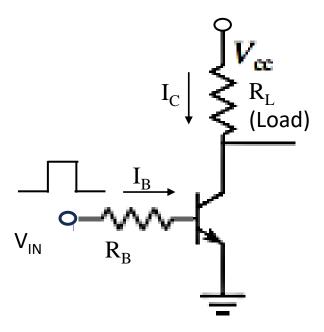
OFF: 
$$V_{GS} < V_T$$
.  $I_G \approx 0$ .  $I_D \approx 0$ .  $V_{DS} \approx V_{DD}$ 

ON: 
$$V_{GS} > V_T$$
 + a few V.  
 $I_G \approx 0$ .  $V_{DS} \approx 0$ .  $I_D \approx V_{DD} / R_L$ .



Typical 'V<sub>DS</sub>-I<sub>D</sub>' characteristics for N channel Enhancement mode MOSFET A & B: operating points for switching action

### BJT & MOSFET Switching with Vin as Control Input

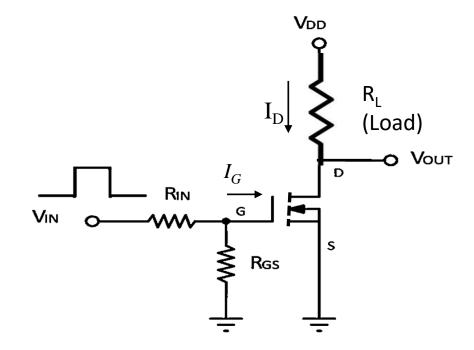


Switch circuit using NPN BJT & control input V<sub>in</sub>.

 $R_B$  is for limiting  $I_B$ .

OFF:  $I_B \approx 0$ .

ON:  $I_B > I_C/\beta$ .



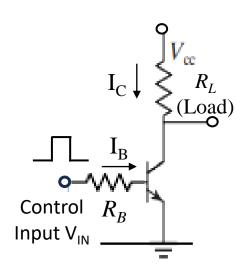
Switch circuit using N-Channel MOSFET & control input V<sub>in</sub>.

 $R_{in}$  and  $R_{GS}$  are for voltage attenuation, if  $V_{in}$ -peak is large.

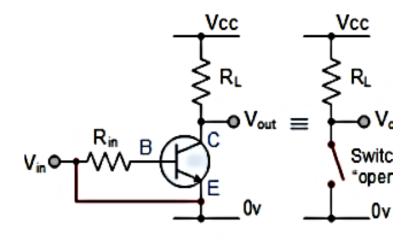
OFF:  $V_{GS} < V_T$ .

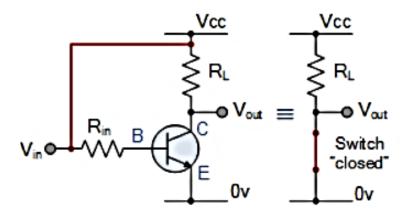
ON:  $V_{GS} > V_T + a$  few V.

# BJT Switching Operation



BJT (NPN) circuit with control input Vin (binary levels: 0,  $V_{cc}$ ).





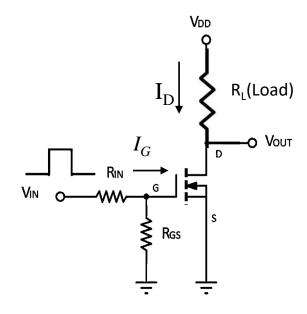
(a) 
$$V_{in} = 0$$

- $V_{BE} = 0 < 0.5 \text{ V}$ .
- BJT operates as open switch.  $I_B \approx 0$ .  $V_{CE} \approx V_{CC}$ .  $I_C \approx 0$ .
- BE junction is not forward biased. BC junction is reverse biased.

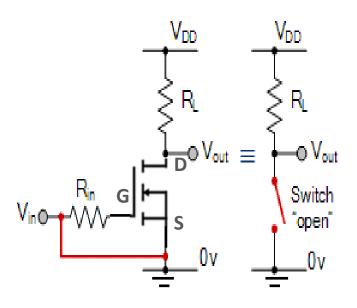
(b) 
$$V_{in} = V_{CC}$$

- BE junction is forward biased, with  $I_B$  limited by  $R_B$ , with  $V_{BE}$  reaching its saturation value  $V_{BES}$  ( $\approx 0.8$  V). Collector voltage drops until BC junction gets forward biased with  $V_{BE}$  reaching its saturation value  $V_{CES}$  ( $\approx 0.2$  V).
- BJT operates as closed switch.
- $V_{BE} = 0.8 \text{ V. } V_{out} = V_{CES} = 0.2 \text{ V.}$   $I_B = (V_{CC} - 0.8) / R_B.$  $I_C = (V_{CC} - V_{CES}) / R_I.$

### MOSFET Switching Operation



MOSFET (N-channel) circuit with control input V<sub>in</sub> (binary levels: 0,  $V_{DD}$ ). Usually R<sub>in</sub> is short and R<sub>GS</sub> is open

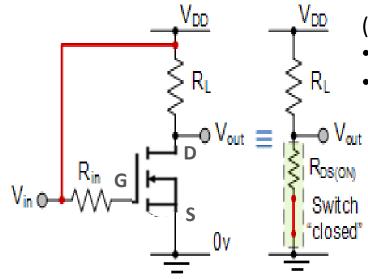


Let  $V_{DD} = 12 \text{ V}$ ,  $V_T = 2 \text{ V}$ ,  $R_{IN} = 0$ ,  $R_{GS} = \infty$ .

(a) 
$$V_{in} = 0$$

- (a)  $V_{in} = 0$   $V_{GS} = 0 < V_T$ . MOSFET operates as open switch.  $I_G \approx 0$ .  $V_{DS} \approx V_{DD}$ .  $I_D \approx 0$ .

$$I_G \approx 0$$
.  $V_{DS} \approx V_{DD}$ .  $I_D \approx 0$ 

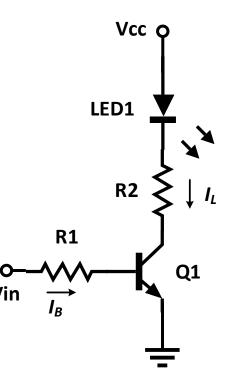


(b) 
$$V_{in} = V_{DD}$$

- (b)  $V_{in} = V_{DD}$   $V_{GS} = 12 \text{ V} > V_T$ . MOSFET operates as closed switch.

$$I_G \approx 0$$
.  $V_{DS} \approx 0$ .  $I_D = V_{DD}/R_L$ .

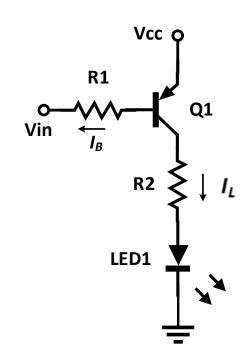
NPN
Switch for
Load
Connected
to +ve
Supply End



$$V_{cc} = 5 \text{ V.}$$
  $V_{in}$ : 0 (LED off), 5 V (LED on).  $\beta > 50$ .  $I_L$  for full brightness = 10 mA. LED drop = 2 V.

$$I_L = [V_{CC} - V_{LED} - V_{CES}]/R_2 = [5 - 2 - 0.2]/R_2 > 10 \text{ mA}$$
  $\Rightarrow R_2 < 2.8/10 \text{ k}\Omega = 280 \Omega$ .  
Let  $R_2 = 270 \Omega$ .  $\Rightarrow I_L = 10.3 \text{ mA}$ .  
 $I_B = [(V_{in})_{high} - V_{BES} - 0]/R_1 > I_L/\beta_{min}$   $\Rightarrow (5 - 0.8 - 0)/R_1 > 10.3/50 \Rightarrow R_1 < 20.38 \text{ k}\Omega$ .  
Let  $R_1 = 18 \text{ k}\Omega$ .

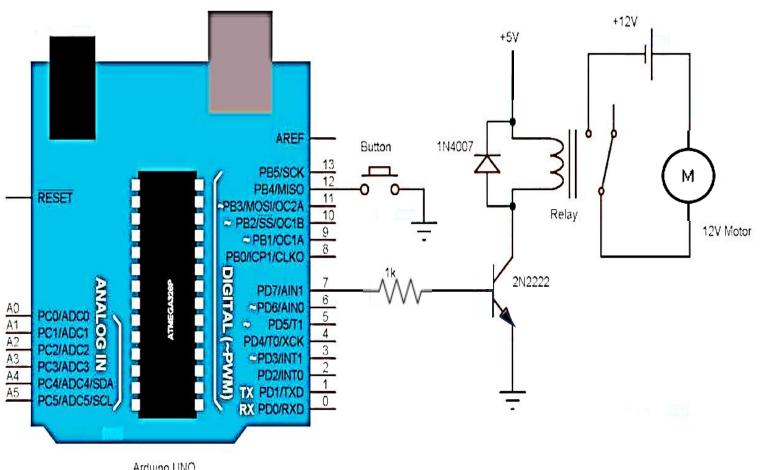




$$I_L$$
 for full brightness = 10 mA. LED drop = 2 V.   
 $I_L = [V_{CC} - V_{ECS} - V_{LED}]/R_2 = [5 - 0.2 - 2]/R_2 > 10$  mA   
 $\Rightarrow R_2 < 2.8/10 \text{ k}\Omega = 280 \Omega$ .   
Let  $R_2 = 270 \Omega$ ,  $\Rightarrow I_L = 10.3$  mA.   
 $I_B = [V_{CC} - V_{EBS} - (V_{in})_{low}]/R_1 > I_L/\beta_{min}$    
 $\Rightarrow (5 - 0.8 - 0)/R_1 > 10.3/50 \Rightarrow R_1 < 20.38 \text{ k}\Omega$ .   
Let  $R_1 = 18 \text{ k}\Omega$ .

 $V_{cc} = 5 \text{ V.}$   $V_{in} : 0 \text{ (LED on)}, 5 \text{ V (LED off)}. \beta > 50.$ 

# Relay Switching Using Micro-Controller (Arduino)



Arduino UNC

Arduino Relay Control Circuit Diagram

Source: https://www.electronicshub.org/arduino-relavcontrol/

### Example

Control of a relay with 'Arduino' digital output pin (PD7) and NPN transistor (2N2222).

Relay coil current = 60 mA. Transistor  $\beta_{min} = 30$ .

Set the pin PD7 to 'Hi'. It will put the transistor in ON state, allowing current to flow through the relay coil making the relay ON.

 $I_B > \text{Relay current} / \beta_{\text{min}} = 2 \text{ mA}.$ 

Diode connected in parallel with the relay avoids sudden change of current in the relay coil.

# Questions and Discussions