

2 Description of State Space

2.1 Definition

1. Input variables

We usually use $\mathbf{u}_t = \begin{bmatrix} u_1(t) \\ u_2(t) \\ \vdots \\ u_{n(t)} \end{bmatrix}$ to represent input variables.

2. State variables

We usually use $\mathbf{x}_t = \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_{n(t)} \end{bmatrix}$ to represent state variables. It is a least set to describe state of system.

3. Output variables

We usually use $\mathbf{y}_t = \begin{bmatrix} y_1(t) \\ y_2(t) \\ \vdots \\ y_{n(t)} \end{bmatrix}$ to represent output variables.

4. State equation

State equation is a first order differential equation that describe relationship between input variables and state variables. We can write it as:

$$\begin{cases} \dot{x}_1 = f_1(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_p, t) \\ \dot{x}_2 = f_2(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_p, t) \\ \vdots \\ \dot{x}_n = f_n(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_p, t) \end{cases} \quad (2.1.1)$$

Rewrite it as vector form:

$$\dot{\mathbf{x}}_t = \mathbf{f}(\mathbf{x}_t, \mathbf{u}_t, t) \quad (2.1.2)$$

5. Output equation

Output equation is a equation that describe relationship between state variables and output variables. We can write it as:

$$\begin{cases} y_1 = g_1(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_p, t) \\ y_2 = g_2(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_p, t) \\ \vdots \\ y_n = g_n(x_1, x_2, \dots, x_n; u_1, u_2, \dots, u_p, t) \end{cases} \quad (2.1.3)$$

Rewrite it as vector form:

$$\mathbf{y}_t = \mathbf{g}(\mathbf{x}_t, \mathbf{u}_t, t) \quad (2.1.4)$$

6. Description of State space of System

We can describe state space of system by equations as:

$$\begin{cases} \dot{\mathbf{x}}_t = \mathbf{f}(\mathbf{x}_t, \mathbf{u}_t, t) \\ \mathbf{y}_t = \mathbf{g}(\mathbf{x}_t, \mathbf{u}_t, t) \end{cases} \quad (2.1.5)$$

When the system is linear, we can write it as:

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{A}(t)\mathbf{x} + \mathbf{B}(t)\mathbf{u} \\ \mathbf{y} = \mathbf{C}(t)\mathbf{x} + \mathbf{D}(t)\mathbf{u} \end{cases} \quad (2.1.6)$$

2.2 Transfer function

Transfer function is a function that describe relationship between input and output of system. Given a system with different state, the transfer function is still the same which means it is not related to state of system in other words state variables.

Single input – Single output system

Given a linear single input-single output system, we have state space representation as:

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}u \\ y = \mathbf{C}\mathbf{x} + \mathbf{D}u \end{cases} \quad (2.2.1)$$

To get transfer function, we can use Laplace transform to get:

$$\begin{aligned} s\mathbf{X} - \mathbf{x}(0) &= \mathbf{A}\mathbf{X} + \mathbf{B}U(s) \\ Y(s) &= \mathbf{C}\mathbf{X}(s) + \mathbf{D}U(s) \end{aligned}$$

Laplace transfer:

$$\mathcal{L}[f(t)] = F(s) = \int_0^{\infty} f(t)e^{-st} dt$$

$$\mathcal{L}[kf(t)] = kF(s)$$

$$\mathcal{L}[f(t) + g(t)] = F(s) + G(s)$$

$$\mathcal{L}[e^{-at}f(t)] = F(s + a)$$

$$\mathcal{L}[e^{at}f(t)] = F(s - a)$$

$$\mathcal{L}[f(t - T)] = e^{-sT}F(s)$$

$$\mathcal{L}[f(at)] = \frac{1}{a}F\left(\frac{s}{a}\right)$$

$$\mathcal{L}\left[\frac{df}{dt}\right] = sF(s) - f(0)$$

$$\mathcal{L}\left[\frac{d^2f}{dt^2}\right] = s^2F(s) - sf(0) - f'(0)$$

$$\mathcal{L}\left[\frac{d^nf}{dt^n}\right] = s^nF(s) - s^{n-1}f(0) - s^{n-2}f'(0) - \dots - f^{(n-1)}(0)$$

$$\mathcal{L}\left[\int_0^t f(t) dt\right] = \frac{F(s)}{s}$$

$$f(\infty) = \lim_{s \rightarrow 0} sF(s)$$

$$f(0) = \lim_{s \rightarrow \infty} sF(s)$$

$f(t)$	$F(s)$
1	$\frac{1}{s}$
t	$\frac{1}{s^2}$
t^n	$\frac{n!}{s^{n+1}}$
e^{-at}	$s + a$
$\sin(\omega t)$	$\frac{\omega}{s^2 + \omega^2}$
$\cos(\omega t)$	$\frac{s}{s^2 + \omega^2}$
$u(t)$	$\frac{1}{s}$
$\delta(t)$	1

The equations are organized as follows:

$$\mathbf{X}(s) = (s\mathbf{I} - \mathbf{A})^{-1}[\mathbf{x}(0) + \mathbf{B}U(s)]$$

$$Y(s) = \mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}[\mathbf{x}(0) + \mathbf{B}U(s)] + DU(s)$$

Let initial condition be zero($\mathbf{x}(0) = 0$),we can get:

$$Y(s) = [\mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B} + D]U(s)$$

Thus,we can get transfer function as:

$$g(s) = \frac{Y(s)}{U(s)} = \mathbf{C}(s\mathbf{I} - \mathbf{A})^{-1}\mathbf{B} + D \quad (2.2.2)$$

Let $D = 0$,we can get:

$$g(s) = \frac{\mathbf{C} \operatorname{adj}(s\mathbf{I} - \mathbf{A})\mathbf{B}}{\det(s\mathbf{I} - \mathbf{A})} \quad (2.2.3)$$

Multi input – Multi output system

Given a multi input-multi output system,we define transfer function between i-th out y_i and j-th input u_j as:

$$g_{ij}(s) = \frac{Y_i(s)}{U_j(s)} \quad (2.2.4)$$

where $Y_i(s)$ is Laplace transform of $y_i(t)$ and $U_j(s)$ is Laplace transform of $u_j(t)$. Must mention that if we define transfer function in this way,we assume that all other inputs are zero.Because linear system satisfies the principle of superposition,so when we plus all inputs U_1, U_2, \dots, U_p ,we can get the i-th output Y_i as:

$$Y_i = \sum_{j=1}^p g_{ij}U_j \quad (2.2.5)$$

We can write it as matrix form:

$$Y(s) = G(s)U(s) \quad (2.2.6)$$

Thus given a linear multi input-multi output system, we have state space representation as:

$$\begin{cases} \dot{x} = Ax + Bu \\ y = Cx + Du \end{cases} \quad (2.2.7)$$

We can conduct as before to get transfer function as:

$$G(s) = C(sI - A)^{-1}B + D = \frac{C \operatorname{adj}(sI - A)B + D \det(sI - A)}{\det(sI - A)} \quad (2.2.8)$$

Regular

We say a transfer function is regular if and only if when

$$\lim_{s \rightarrow \infty} g(s) = c \quad (2.2.9)$$

where c is a constant. And a transfer function is strictly regular if and only if when

$$\lim_{s \rightarrow \infty} g(s) = 0 \quad (2.2.10)$$

2.3 Establishing State Space Model by Differential Equation

Given a single input and single output system, if we have differential equation as:

$$y^{(n)} + a_1 y^{(n-1)} + a_2 y^{(n-2)} + \dots + a_n y = b_0 u + b_1 u^{(1)} + b_2 u^{(2)} + \dots + b_m u^{(m)} \quad (2.3.1)$$

where $m \leq n$.

Condition 1: $m = 0$ let $b_0 = 1$, we have:

$$y^{(n)} + a_1 y^{(n-1)} + a_2 y^{(n-2)} + \dots + a_n y = b_0 u + b_1 u^{(1)} + b_2 u^{(2)} + \dots + b_m u^{(m)} \quad (2.3.2)$$

We can define state variables as:

$$\begin{cases} x_1 = y \\ x_2 = y^{(1)} \\ x_3 = y^{(2)} \\ \vdots \\ x_n = y^{(n-1)} \end{cases} \quad (2.3.3)$$

We can get state equation as:

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = x_3 \\ \vdots \\ \dot{x}_{n-1} = x_n \\ \dot{x}_n = -a_1 x_n - a_2 x_{n-1} - \dots - a_n x_1 + b_0 u \end{cases} \quad (2.3.4)$$

We can rewrite it as vector form:

$$\dot{\mathbf{x}} = \begin{bmatrix} 0 & 1 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & 1 \\ -a_n & -a_{n-1} & \dots & -a_1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 1 \end{bmatrix} u$$

$$y = [1 \ 0 \ \dots \ 0] \mathbf{x}$$
(2.3.5)

Condition 2: $m \neq n$

Controllable Canonical Form Method:

Let us note D as $\frac{d}{dt}$, we can rewrite Equation (2.3.1) as:

$$y = \frac{b_0 D^m + b_1 D^{m-1} + b_2 D^{m-2} + \dots + b_m}{D^n + a_1 D^{n-1} + a_2 D^{n-2} + \dots + a_n} u$$
(2.3.6)

Let us discuss the case when $m < n$

let $\tilde{y}^{(n)} + a_1 \tilde{y}^{(n-1)} + a_2 \tilde{y}^{(n-2)} + \dots + a_{n-1} \tilde{y}^{(1)} + a_n \tilde{y} = u$ also as $\tilde{y} = \frac{1}{D^n + a_1 D^{n-1} + a_2 D^{n-2} + \dots + a_n} u$

we can get:

$$y = b_0 \tilde{y}^{(m)} + b_1 \tilde{y}^{(m-1)} + b_2 \tilde{y}^{(m-2)} + \dots + b_m \tilde{y}$$
(2.3.7)

We choose state variables as $x_1 = \tilde{y}, x_2 = \tilde{y}^{(1)}, \dots, x_n = \tilde{y}^{(n-1)}$. We can get state equation as:

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = x_3 \\ \vdots \\ \dot{x}_{n-1} = x_n \\ \dot{x}_n = -a_n x_1 - a_{n-1} x_2 - \dots - a_1 x_n + u \end{cases}$$
(2.3.8)

and output equation as:

$$y = b_m x_1 + b_{m-1} x_2 + \dots + b_0 x_{m+1}$$
(2.3.9)

We can rewrite it as vector form:

$$\dot{\mathbf{x}} = \begin{bmatrix} 0 & 1 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & 1 \\ -a_n & -a_{n-1} & \dots & -a_1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 1 \end{bmatrix} u$$

$$y = [b_m, \dots, b_0, 0, \dots, 0] \mathbf{x}$$
(2.3.10)

Let us discuss the case when $m = n$, we can rewrite Equation (2.3.6) as:

$$y = \left[b_0 + \frac{(b_1 - b_0 a_1) D^{n-1} + \dots + (b_n - b_0 a_n)}{D^n + a_1 D^{n-1} + \dots + a_n} \right] u$$
(2.3.11)

Also let $\tilde{y}^{(n)} + a_1 \tilde{y}^{(n-1)} + a_2 \tilde{y}^{(n-2)} + \dots + a_{n-1} \tilde{y}^{(1)} + a_n \tilde{y} = u$ We can get:

$$y = (b_n - b_0 a_n) \tilde{y}^{(n-1)} + (b_{n-1} - b_0 a_{n-2}) \tilde{y}^{(n-2)} + \dots + (b_1 - b_0 a_1) \tilde{y} + b_0 u$$
(2.3.12)

Thus we can write state equation in vector form in familiar way as:

$$\dot{x} = \begin{bmatrix} 0 & 1 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & 1 \\ -a_n & -a_{n-1} & \dots & -a_1 \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 1 \end{bmatrix} u \quad (2.3.13)$$

$$y = [b_n - b_0 a_n, b_{n-1} - b_0 a_{n-1}, \dots, b_1 - b_0 a_1] x + b_0 u$$

Undetermined Canonical Form Method: W.l.o.g, we assume that the equation is in the form of:

$$y^{(n)} + a_1 y^{(n-1)} + a_2 y^{(n-2)} + \dots + a_n y = b_0 u^{(n)} + b_1 u^{(n-1)} + b_2 u^{(n-2)} + \dots + b_n u \quad (2.3.14)$$

We can define state variables as:

$$\begin{cases} x_1 = y - \beta_0 u \\ x_2 = \dot{x}_1 - \beta_1 u = \dot{y} - \beta_0 \dot{u} - \beta_1 u \\ x_3 = \dot{x}_2 - \beta_2 u = \ddot{y} - \beta_0 \ddot{u} - \beta_1 \dot{u} - \beta_2 u \\ \vdots \\ x_n = \dot{x}_{n-1} - \beta_{n-1} u = y^{(n-1)} - \beta_0 u^{(n-1)} - \beta_1 u^{(n-2)} - \dots - \beta_{n-1} u \end{cases} \quad (2.3.15)$$

Thus we have:

$$\begin{cases} y = x_1 + \beta_0 u \\ \dot{y} = x_2 + \beta_0 \dot{u} + \beta_1 u \\ \ddot{y} = x_3 + \beta_0 \ddot{u} + \beta_1 \dot{u} + \beta_2 u \\ \vdots \\ y^{(n-1)} = x_n + \beta_0 u^{(n-1)} + \beta_1 u^{(n-2)} + \dots + \beta_{n-1} u \end{cases} \quad (2.3.16)$$

Let us introduce a new variables $x_{n+1} = \dot{x}_n - \beta_n u = \dot{x}_{n-1} - \beta_{n-1} u = y^{(n)} - \beta_0 u^{(n)} - \beta_1 u^{(n-1)} - \dots - \beta_n u$. Thus we have:

$$y^{(n)} = x_{n+1} + \beta_0 u^{(n)} + \beta_1 u^{(n-1)} + \dots + \beta_n u \quad (2.3.17)$$

Substitute $y, \dot{y}, \dots, y^{(n)}$ into Equation (2.3.14), we can get:

$$\begin{aligned} & (x_{n+1} + a_1 x_n + \dots + a_n x_1) + \beta_0 u^{(n)} + (\beta_1 + a_1 \beta_0) u^{(n-1)} + \\ & (\beta_2 + a_1 \beta_1 + a_2 \beta_0) u^{(n-2)} + \dots + (\beta_n + a_1 \beta_{n-1} + a_2 \beta_{n-2} + \dots + a_n \beta_0) u \\ & = b_0 u^{(n)} + b_1 u^{(n-1)} + b_2 u^{(n-2)} + \dots + b_n u \end{aligned} \quad (2.3.18)$$

Compare the coefficients of $u^{(n)}, u^{(n-1)}, \dots, u$, we can get:

$$\begin{cases} x_{n+1} + a_1 x_n + \dots + a_n x_1 & = 0 \\ \beta_0 & = b_0 \\ \beta_1 + a_1 \beta_0 & = b_1 \\ \beta_2 + a_1 \beta_1 + a_2 \beta_0 & = b_2 \\ \vdots & \\ \beta_n + a_1 \beta_{n-1} + a_2 \beta_{n-2} + \dots + a_n \beta_0 & = b_n \end{cases} \quad (2.3.19)$$

we can rewrite it as matrix form:

$$\begin{bmatrix} b_0 \\ b_1 \\ b_2 \\ \vdots \\ b_n \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & \dots & 0 \\ a_1 & 1 & 0 & \dots & 0 \\ a_2 & a_1 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ a_n & a_{n-1} & a_{n-2} & \dots & 1 \end{bmatrix} \begin{bmatrix} \beta_0 \\ \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} \quad (2.3.20)$$

In summary, we can get state equation as:

$$\begin{cases} \dot{x}_1 = \dot{y} - \beta_0 \dot{u} = x_2 + \beta_1 u \\ \dot{x}_2 = \ddot{y} - \beta_0 \ddot{u} - \beta_1 \dot{u} = x_3 + \beta_2 u \\ \vdots \\ \dot{x}_{n-1} = y^{(n-1)} - \beta_0 u^{(n-1)} - \beta_1 u^{(n-2)} - \dots - \beta_{n-2} \dot{u} = x_n + \beta_{n-1} u \\ \dot{x}_n = y^{(n)} - \beta_0 u^{(n)} - \beta_1 u^{(n-1)} - \dots - \beta_{n-1} \dot{u} = -a_n x_1 - a_{n-1} x_2 - \dots - a_1 x_n + \beta_n u \end{cases} \quad (2.3.21)$$

and output equation as:

$$y = x_1 + \beta_0 u \quad (2.3.22)$$

We can rewrite it as vector form:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & 1 \\ -a_n & -a_{n-1} & -a_{n-2} & \dots & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} \beta_1 \\ \beta_2 \\ \vdots \\ \beta_n \end{bmatrix} u \quad (2.3.23)$$

$$y = [1, 0, 0, \dots, 0] \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \beta_0 u$$

2.4 Establishing State Space Model by Transfer Function

For a actual physical system, the transfer function of the system is always regular.

First, let us discuss the situation where the system is restrict regular. If we have a differential equation of system as:

$$y^{(n)} + a_{n-1}y^{(n-1)} + a_{n-2}y^{(n-2)} + \dots + a_1\dot{y} + a_0y = b_{n-1}u^{(n-1)} + \dots + b_1\dot{u} + b_0u \quad (2.4.1)$$

Then we have transfer function as:

$$g(s) = \frac{Y(s)}{U(s)} = \frac{b_1s^{n-1} + b_2s^{n-2} + \dots + b_n}{s^n + a_1s^{n-1} + \dots + a_{n-1}s + a_n} \quad (2.4.2)$$

We have:

$$Y(s) = U(s) \frac{b_1s^{-1} + b_2s^{-2} + \dots + b_ns^{-n}}{1 + a_1s^{-1} + \dots + a_{n-1}s^{-(n-1)} + a_ns^{-n}} \quad (2.4.3)$$

Introduce a intermediate variables $Z(s) = \frac{U(s)}{1 + a_1s^{-1} + \dots + a_{n-1}s^{-(n-1)} + a_ns^{-n}}$ We can get:

$$Y(s) = b_1s^{-1}Z(s) + b_2s^{-2}Z(s) + \dots + b_ns^{-n}Z(s) \quad (2.4.4)$$

Let us inverse Laplace transform of $Z(s)$, we can get:

$$z^{(n)} + a_{n-1}z^{(n-1)} + a_{n-2}z^{(n-2)} + \dots + a_1\dot{z} + a_0z = u \quad (2.4.5)$$