

MEEC/MIEEC

ELECTRONICS FOR MICRO-SYSTEMS

Lab#1 P1 A Temperature Meter System with 3 Sensors, Relay and GUI

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1 Introduction

explain the requirements and the main objectives of the project Se calhar dizer aqui quais sao as metricas por onde podemos avaliar a nossa solucao Linearidade de output para conseguir aproveitar melhor a resolucao do adc Consumo erro

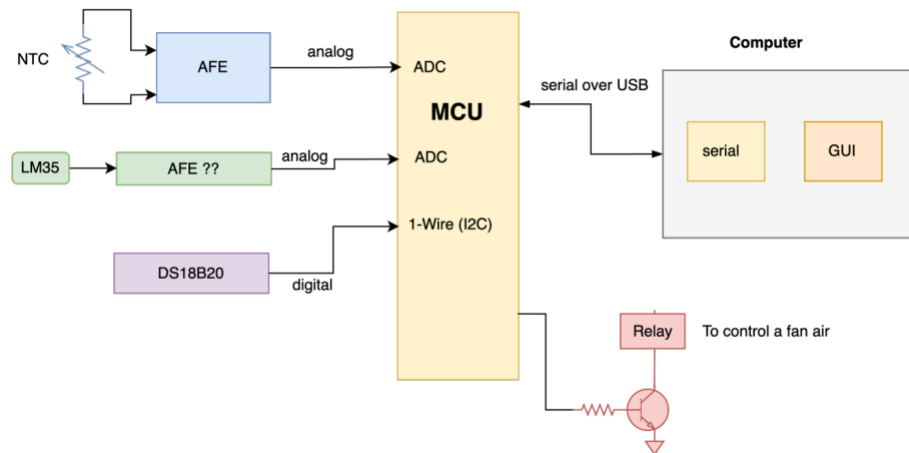


Figure 1: Temperature sensing system with 3 three types of sensors.

2 Temperature Sensors

2.1 NTC - Negative Temperature Coefficient

2.2 LM35 - Precision Centigrade Temperature Sensor

2.3 DS18B20 - Digital Thermometer

3 System Design

3.1 Analog FrontEnd (AFE) NTC

A parte onde definimos o intervalo de temperatura nao devia estar aqui pq é para todos os circuitos

In order to design the *NTF AFE* first it's necessary to define the temperature interval in which this circuit will work, thus it was define as $T \in [10^\circ; 40^\circ]$. Through the *NTC*'s datasheet the interval of its resistance values is $R_{NTC} \in [5,282k ; 19,98k]$

For an accurate reading of the temperature it was used the *Steinhart-Hart equation*.

$$\frac{1}{T} = A + B \cdot \ln(R_{NTC}) + C \cdot [\ln(R_{NTC})]^3 \quad (1)$$

In order to find the parameters A , B and C , its necessary to use 3 points from the datasheet. The points chosen were the two extremes and the middles point.

$$\begin{cases} R(283,15) = 1,998 \cdot 10^4 \Omega \\ R(298,15) = 10^4 \Omega \\ R(313,15) = 0,5282 \cdot 10^4 \Omega \end{cases} \Leftrightarrow \begin{cases} A = 1,2 \cdot 10^{-3} \\ B = 2,1 \cdot 10^{-4} \\ C = 1,3 \cdot 10^{-7} \end{cases} \quad (2)$$

Para poder dimensionar o AFE do NTC primeiro é necessario definir o intrevalo de temperaturas em que o circuito irá operar. Foi então decidido que seria adequado um temperatura $T \in [10^\circ; 40^\circ]$. E pelo datasheet do NTC foi obtido o intrevalo da sua resistencia $R_{NTC} \in [5,282k ; 19,98k]$

Para usar equacao Steinhart-Hart $\frac{1}{T} = A + B \cdot \ln(R) + C \cdot [\ln(R)]^3$, precisamos de usar 3 pontos para encontrar as constantes A , B e C . $R(T) = R_{NTC}$ onde T é a temperatura em kelvin e R_{NTC} é o valor da resistencia do thermistor *NTC*

The simplest way to convert the resistance to voltage, is to use a voltage divider circuit.

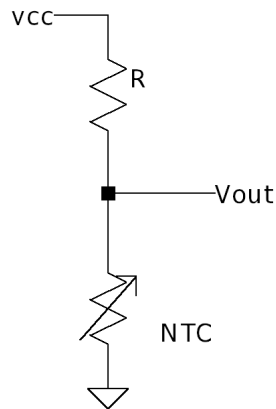
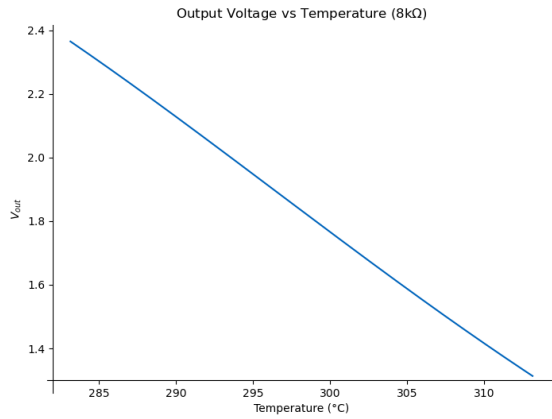


Figure 2: NTC voltage divider.

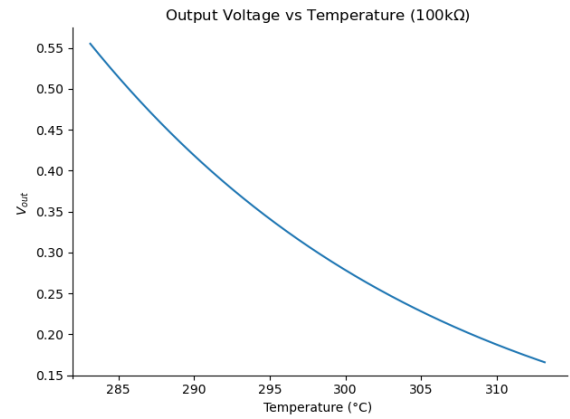
This approach has a few problems:

- The output resistance is really high $R \parallel NTC$,
- The output voltage is highly non linear, which is a problem because this way some *ADC* resolution is lost.

The first problem is solved through a buffer at the entrance of the *AFE*, and the second is somewhat mitigated by using a resistor value around $8k\Omega$. To achieve this value it was used the beta model for the *NTC* and through a python script, the resistor value was iterated until an almost linear output was achieved.



(a) $8K\Omega$



(b) $100K\Omega$

Figure 3: Output Voltage vs Temperature

With $R = 8k$ the output signal is $V_{out} \in [1.31V; 2.36V]$. But the *ADC* as resolution of 12 bits and a voltage range of $3.3V$, meaning that to have the best resolution possible the signal needs an offset and a gain of 2.87 and -1.75 respectively. It's important to note that a $0.1V$ margin was added to V_{out} in order to ensure that the *AFE* works properly in the specified range. This gain and offset was achieved through the following topology.

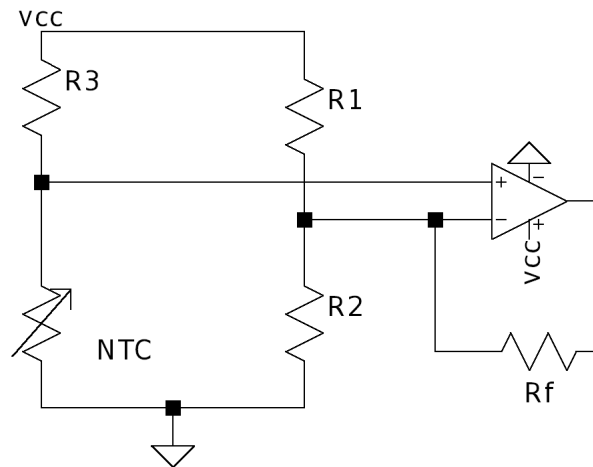


Figure 4: NTC's AFE topology

Nao esquecer de fazer referencia ao codigo funcao AfeNtc()

As already specified the R_3 value is $8K \Omega$ and for the purpose circuit dimensioning the circuit the positive node of the opamp will be treated as a variable V_p .

Using python the circuit function $V_{out}(V_p)$:

$$V_{out} = V_p \left[1 + R_f \cdot \left(\frac{1}{R_1} + \frac{1}{R_2} \right) \right] - V_{cc} \cdot \frac{R_f}{R_1} \quad (3)$$

Now it's clear to see what part of the equation is responsible for the circuit gain and offset.

Setting $R_f = 10K$ the values that satisfy the equation system are $R_1 = 9348\Omega$ and $R_2 = 11760\Omega$. In the simulation this values are a bit too close to the limits so in order to have better margins and to use the available resistors the final values are $R_1 = 8.2K\Omega$ and $R_2 = 12K\Omega$.

This integrated-circuit temperature sensor, generates an output voltage linearly proportional to the Centigrade temperature.

Hence, in the specified conditions $V_{out} \in [0.1; 0.4]$. To increase resolution as done in [NTC's AFE subsection](#), it's needed to add gain and an offset. For this purpose a differential amplifier can be used.



Although this circuit achieves the expected output, the amount of components, increases noise and power consumption, thus the used topology was a non-inverting amplifier.

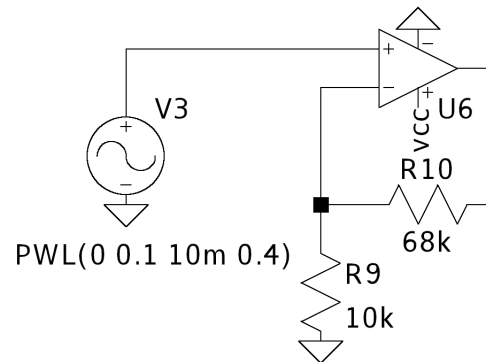


Figure 6: AFE for the LM35 sensor.

This way the input impedance is high, only one OpAmp and two resistors are used. But this comes with the cost of lost resolution. With a gain of 8 $V_{out} \in [0.8V, 3.2V]$ 0.8V of range in the *ADC* are lost.

3.3 Output Relay

Lastly the *MCU* will turn on and off a fan. For this purpose a relay will open and close the fan circuit. Because the *MCU's* *I/O* can drive the relay a *NPN* transistor is used to get enough current. Thus the following circuit also needs to be dimensioned. It's important to note that the diode is needed to protect the circuit.

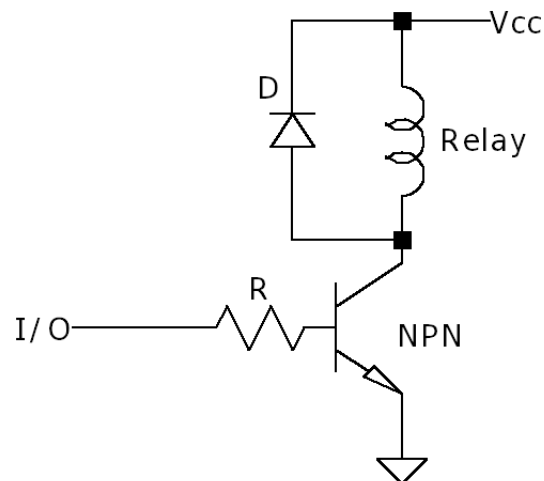


Figure 7: Relay circuit.

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$$R = \frac{I/O_{HIGH} - V_{BE}}{I_r} \cdot \beta \Leftrightarrow R = \quad (5)$$

4 Simulations

4.1 NTC

For the purpose of testing the following circuit was designed. At first with linear voltage supply since as seen in [section](#), this voltage is almost linear.

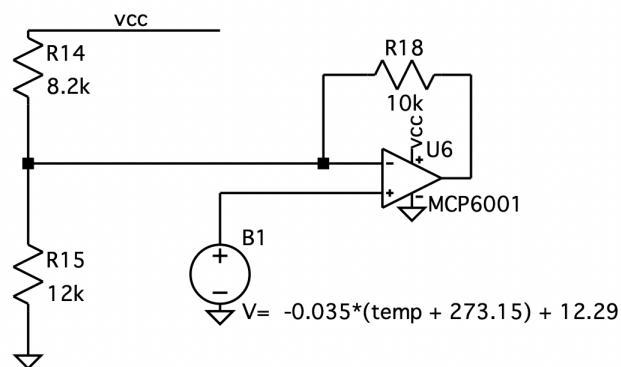


Figure 8: Linear supply NTC test circuit.

Producing the following results.

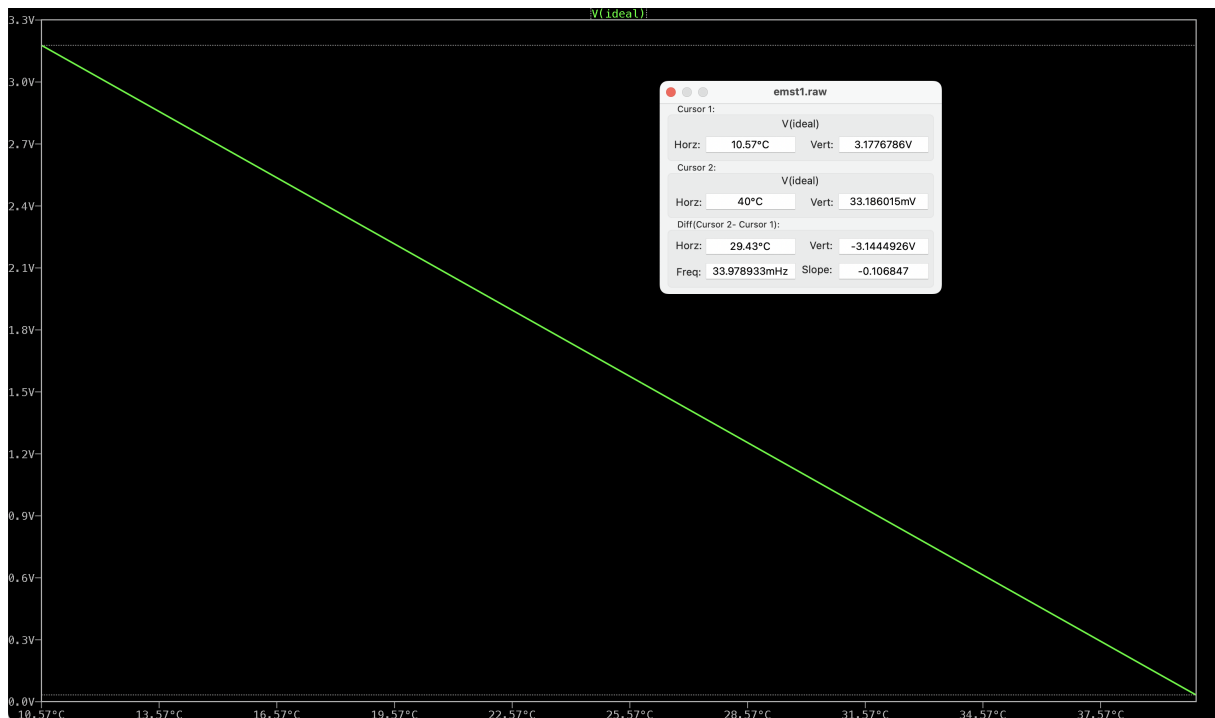


Figure 9: Linear supply NTC test results.

Confirming the effectiveness of the *AFE* circuit a more realistic circuit was tested.

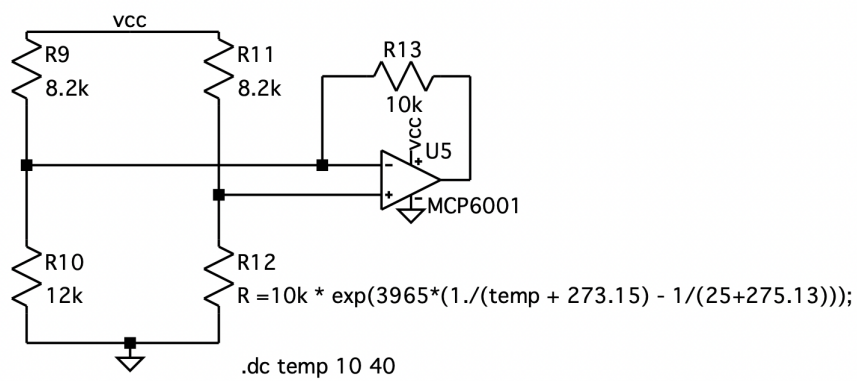


Figure 10: Realistic NTC test circuit.

With following results.

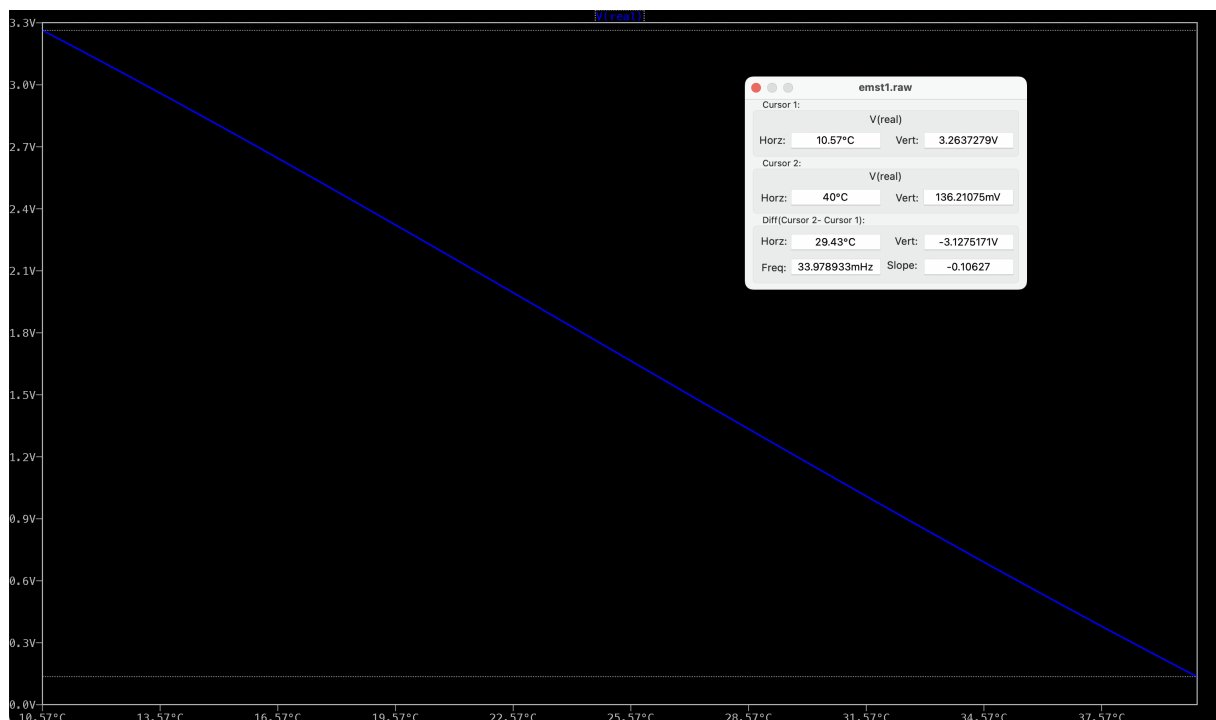


Figure 11: Realistic NTC test results.

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Comparing both results the difference is minimal. Confirming that the output is nearly linear in relation to the temperature variation.



Figure 12: NTC Linear and Realistic results comparison.

4.2 LM35

For this sensor the simulation is straight forward, since the output is linear and proportional to the temperature, it can be simulated with variable voltage supply.

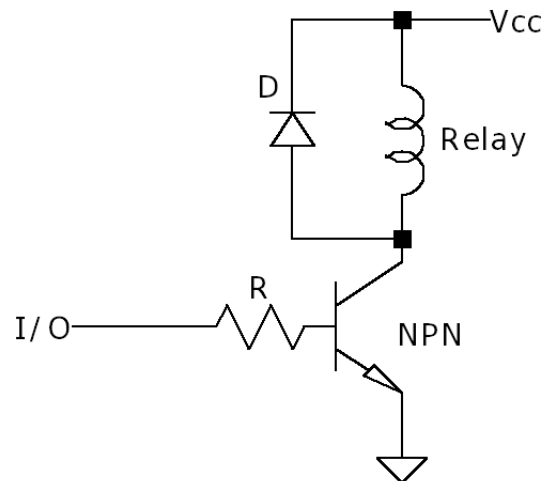


Figure 13: Relay circuit.

Giving the following results.

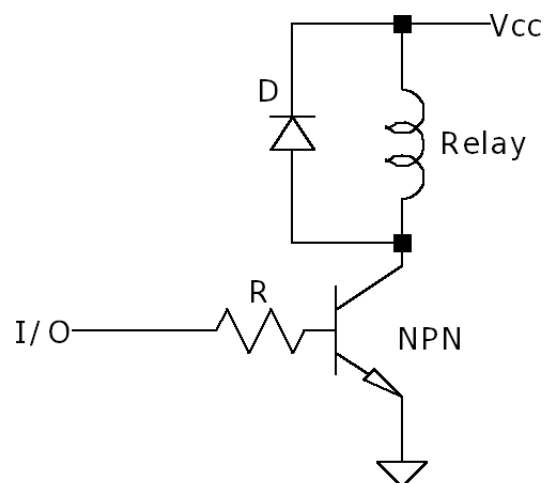


Figure 14: Relay circuit.

5 Implementation and Experimental Tests

6 Results Analysis

7 Conclusion