

Satellite Dish Azimuth and Elevation Analysis

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I. INTRODUCTION

II. NWZ FRAME ANALYSIS

Let $\beta \in (0, 2\pi]$ and $\eta \in [-\pi/2, \pi/2]$ denote the azimuth and elevation of the satellite, respectively, regarding the antenna dish NWZ reference frame.

$$x = r \cos \eta \cos \beta \quad (1)$$

$$y = r \cos \eta \sin \beta \quad (2)$$

$$z = r \sin \eta \quad (3)$$

where

$$r = \sqrt{x^2 + y^2 + z^2} \quad (4)$$

$$\eta = \arcsin(z/r) \quad (5)$$

$$\beta = \arctan 2(y, x) \quad (6)$$

It is important to notice that when the satellite is precisely at the zenith ($x = y = 0$ or $\eta = \pi/2$), the azimuth is not well defined. Such pathological case will be later considered.

A. Speed

$$\begin{aligned} \dot{r} &= \frac{\partial r}{\partial x} \dot{x} + \frac{\partial r}{\partial y} \dot{y} + \frac{\partial r}{\partial z} \dot{z} \\ &= \nabla r \cdot \mathbf{v} \\ &= \hat{\mathbf{r}} \cdot \mathbf{v} \end{aligned} \quad (7)$$

B. Speed of XY-Projection

Let $s = \sqrt{x^2 + y^2}$

$$\begin{aligned} \dot{s} &= \frac{\partial s}{\partial x} \dot{x} + \frac{\partial s}{\partial y} \dot{y} \\ &= \frac{x\dot{x} + y\dot{y}}{s} \end{aligned}$$

C. Elevation Rate ($\eta \neq \pi/2$)

Considering

$$\sin \eta = \frac{z}{r} \quad \text{and} \quad \cos \eta = \frac{s}{r}, \quad (9)$$

then

$$\begin{aligned} \cos \eta \cdot \dot{\eta} &= \frac{\dot{z}}{r} - z \frac{\dot{r}}{r^2} \\ \sin \eta \cdot \dot{\eta} &= -\frac{\dot{s}}{r} + s \frac{\dot{r}}{r^2} \end{aligned} \quad (10)$$

which leads to

$$\begin{aligned} \cos^2 \eta \cdot \dot{\eta} &= \frac{s\dot{z}}{r^2} - sz \frac{\dot{r}}{r^3} \\ \sin^2 \eta \cdot \dot{\eta} &= -\frac{\dot{s}z}{r^2} + sz \frac{\dot{r}}{r^3} \end{aligned}$$

Thus

$$\dot{\eta} = \frac{s\dot{z} - \dot{s}z}{r^2} \quad (11)$$

D. Azimuth Rate ($\eta \neq \pi/2$)

Considering

$$\sin \beta = \frac{y}{\sqrt{x^2 + y^2}} \quad \text{and} \quad \cos \beta = \frac{x}{\sqrt{x^2 + y^2}} \quad (12)$$

Then

$$\begin{aligned} \cos \beta \cdot \dot{\beta} &= \frac{d}{dt} \left(\frac{y}{\sqrt{x^2 + y^2}} \right) = \frac{\dot{y}\sqrt{x^2 + y^2} - y \frac{d}{dt} \sqrt{x^2 + y^2}}{x^2 + y^2} \\ \sin \beta \cdot \dot{\beta} &= -\frac{d}{dt} \left(\frac{x}{\sqrt{x^2 + y^2}} \right) = \frac{-\dot{x}\sqrt{x^2 + y^2} + x \frac{d}{dt} \sqrt{x^2 + y^2}}{x^2 + y^2} \end{aligned} \quad (13)$$

$$\text{Since } \frac{d}{dt} \sqrt{x^2 + y^2} = \frac{x\dot{x} + y\dot{y}}{\sqrt{x^2 + y^2}}$$

$$\cos \beta \cdot \dot{\beta} = \frac{\dot{y}}{\sqrt{x^2 + y^2}} - \frac{y}{x^2 + y^2} \frac{x\dot{x} + y\dot{y}}{\sqrt{x^2 + y^2}}$$

$$\sin \beta \cdot \dot{\beta} = \frac{-\dot{x}}{\sqrt{x^2 + y^2}} + \frac{x}{x^2 + y^2} \frac{x\dot{x} + y\dot{y}}{\sqrt{x^2 + y^2}}$$

then

$$x \cdot \dot{\beta} = \dot{y} - \frac{y}{x^2 + y^2} (x\dot{x} + y\dot{y})$$

$$y \cdot \dot{\beta} = -\dot{x} + \frac{x}{x^2 + y^2} (x\dot{x} + y\dot{y})$$

(8) Finally

$$\dot{\beta} = \begin{cases} \frac{\dot{y}}{x} - \frac{y}{x} \frac{x\dot{x} + y\dot{y}}{x^2 + y^2}, & x \neq 0 \\ \frac{-\dot{x}}{y}, & x = 0 \end{cases} \quad (14)$$

or

$$\dot{\beta} = \frac{x\dot{y} - \dot{x}y}{x^2 + y^2} \quad (15)$$

III. OPTIMAL CONTROL

During a pass, assume that the satellite rises over the horizon at the instant $t_{\text{sr}} \in \mathbb{R}$ and that it sets at $t_{\text{ss}} > t_{\text{sr}}$. Thus, the ground station antenna should be able to follow the satellite during the interval $I = [t_{\text{sr}}, t_{\text{ss}}]$. Moreover, let $\hat{\mathbf{r}}_{\text{sat}} \in \mathcal{C}^1(I; \mathbb{R}^{3 \times 1})$, that maps $I \ni t \mapsto \hat{\mathbf{r}}_{\text{sat}}(t) \in \mathbb{R}^{3 \times 1}$, denote the unit vector from ground station in the satellite direction. Mathematically, it can be said that the ground station antenna should track $\hat{\mathbf{r}}_{\text{sat}}$ as close as possible.

Consider that the azimuth $\beta \in [0, 2\pi)$ and elevation $\eta \in [0, \pi/2]$ of the antenna can be controlled throughout an input signal $\mathbf{u} \in \mathcal{C}^0(I; \mathbb{R}^{n \times 1})$ by the state law

$$\dot{\mathbf{x}} = \Phi(t, \mathbf{x}, \mathbf{u}) \quad (16)$$

while constrained by a set (i -indexed) of state-control restrictions on the form

$$G_i(t, \mathbf{x}, \mathbf{u}) \leq 0 \quad (17)$$

where the state is define by $\mathbf{x} = [\eta \ \beta \ \dot{\eta} \ \dot{\beta}]^T$.

Consider $D = [0, \pi/2] \times [0, 2\pi)$ and let $\hat{\mathbf{r}}_{\text{ant}} : D \rightarrow \mathbb{R}^{3 \times 1}$, that maps $D \ni (\eta, \beta) \mapsto \hat{\mathbf{r}}_{\text{ant}}(\eta, \beta) \in \mathbb{R}^{3 \times 1}$, be the antenna pointing direction, which is defined by

$$\hat{\mathbf{r}}_{\text{ant}}(\eta, \beta) = \begin{bmatrix} \cos \eta \cdot \cos \beta \\ \cos \eta \cdot \sin \beta \\ \sin \eta \end{bmatrix}. \quad (18)$$

In this scenario, the antenna control unit (ACU) should be able to keep $\hat{\mathbf{r}}_{\text{ant}}$ as close as possible to $\hat{\mathbf{r}}_{\text{sat}}$ by minimising the cost functional

$$J(\mathbf{x}) = \int_I \|\hat{\mathbf{r}}_{\text{ant}} - \hat{\mathbf{r}}_{\text{sat}}\|^2 d\tau. \quad (19)$$

As established in [?, Troutman, Chapter 10], the solution is to consider the lagrangian multipliers λ and μ_i and the hamiltonian

$$h(t, \mathbf{x}, \mathbf{u}, \lambda, \mu) = g(t, \mathbf{x}) + \lambda^T \Phi(t, \mathbf{x}, \mathbf{u}) + \sum_i \mu_i G_i(t, \mathbf{x}, \mathbf{u}), \quad (20)$$

where the lagragian is define by

$$g(t, \mathbf{x}) = \|\hat{\mathbf{r}}_{\text{ant}}(\eta(t), \beta(t)) - \hat{\mathbf{r}}_{\text{sat}}(t)\|^2. \quad (21)$$

(See conditions for unique solution in Troutman, section 10.1)