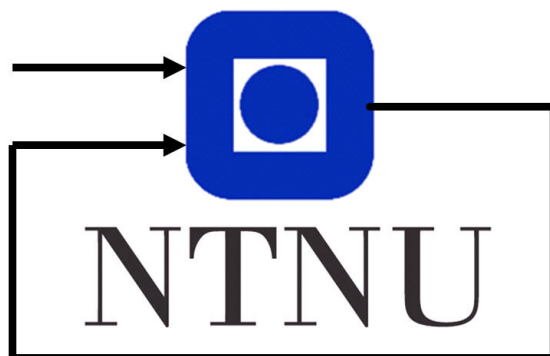
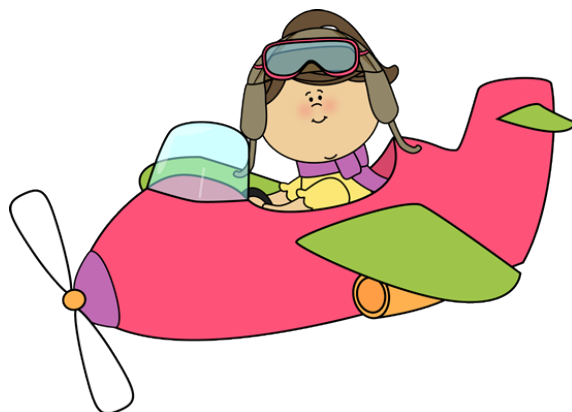


TTK4190 Guidance and Control of Vehicles

Assignment 2

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Problem 1 - Attitude Control of Satellite

Problem 1 a)

What is the ground speed of the aircraft (numerical value) in the absence of wind?

In the absence of wind, $V_w = 0$, we have $V_g = V_a = 580 \text{ km/h}$.

Problem 1 b)

Write down two expressions for the sideslip (crab) angle β in the absence of wind. One expression should depend on aircraft velocity and the other on aircraft heading.

$$\beta = \sin^{-1} \left(\frac{v_r}{\sqrt{u_r^2 + v_r^2 + w_r^2}} \right) \quad (1)$$

$$\beta = \chi_c - \psi \quad (2)$$

Problem 1 c)

Compute the Dutch-roll natural frequency and relative damping ratio for the aircraft. Can you, very briefly with your own words, describe how the Dutch roll mode affect the yaw and roll motion? How would the motion change with increased relative damping ratio?

$$\lambda_{dutch-roll} = \frac{Y_v + N_r}{2} \pm \sqrt{\left(\frac{Y_v + N_r}{2} \right)^2 - (Y_v N_r - N_v Y_r)} \quad (3)$$

With the constants found from the state space model

$$Y_V = -0.322 \quad (4)$$

$$N_R = -0.32 \quad (5)$$

$$N_V = 0.0426 \quad (6)$$

$$Y_R = -180.44 \quad (7)$$

We get

$$\lambda = -0.321 \pm 2.77i \quad (8)$$

$$\omega_n = 2.79 \quad (9)$$

$$\zeta = 0.1147 \quad (10)$$

Dutch roll will affect roll and yaw by inducing an oscillation in yaw which is "transferred" to roll. If you increase the damping ratio the oscillation will dampen faster.

Problem 1 d)

Compute the spiral-divergence mode. Is the mode unstable?

For the spiral-divergence mode we assume that $\dot{p} = \bar{p} = 0$ and that the rudder command is negligible. We have

$$\lambda_{spiral} = \frac{N_r L_v - N_v L_r}{L_v} \quad (11)$$

With the constants

$$L_v = -0.0657 \quad (12)$$

$$L_r = 0.46 \quad (13)$$

We get

$$\lambda_{spiral} = -0.022 \quad (14)$$

Since λ is less than 0 we do not have an unstable mode.

Problem 1 e)

Compute the roll mode. Is the roll mode faster or slower than the spiral-divergence mode?

We can find the roll mode by ignoring the heading dynamics and assuming a constant pitch angle. An approximation of the eigenvalue for the rolling mode is given by

$$\lambda_{roll} = L_p \quad (15)$$

$$= -2.87 \quad (16)$$

This mode is faster, as the pole is more negative.

Problem 2 - Autopilot for course hold using aileron and successive loop closure

Problem 2 a)

Find numerical values for a_{ϕ_1} and a_{ϕ_2} based on the state-space model.

$$\delta_a \left(\frac{a_{\phi_2}}{s + a_{\phi_1}} \right) = p \Rightarrow \frac{p}{\delta_a} = \frac{a_{\phi_2}}{s + a_{\phi_1}} \quad (17)$$

$$\dot{p} = -10.6\beta - 2.87p + 0.46r - 0.65\delta_a \quad (18)$$

$$\dot{p} = -2.87p - 0.65\delta_a \quad (19)$$

$$\Rightarrow p(s + 2.87) = -0.65\delta_a \quad (20)$$

$$\Rightarrow \frac{p}{\delta_a} = -\frac{0.65}{s + 2.87} \quad (21)$$

Therefore

$$a_{\phi_1} = 2.87 \quad (22)$$

$$a_{\phi_2} = -0.65 \quad (23)$$

Problem 2 b)

Find numerical values for the five gains $k_{p\phi}$, $k_{d\phi}$, $k_{i\phi}$, $k_{i\chi}$, and $k_{p\chi}$.

First we find the control gains of the inner loop which control roll angle and roll rate.

We find the control gains $k_{p\phi}$ and $k_{d\phi}$ based on the desired response of the closed-loop dynamics.

The transfer function from ϕ^c to ϕ is given by

$$H_{\phi/\phi^c}(s) = \frac{k_{p\phi} a_{\phi_2}}{s^2 + (a_{\phi_1} + a_{\phi_2} k_{d\phi})s + k_{p\phi} a_{\phi_2}} \quad (24)$$

This can be recognized as the canonical second-order transfer function

$$\frac{\phi(s)}{\phi^c(s)} = \frac{\omega_{n_\phi}^2}{s^2 + 2\zeta_\phi \omega_{n_\phi} s + \omega_{n_\phi}^2} \quad (25)$$

If we then equate the denominator polynomial coefficients we get

$$\omega_{n_\phi}^2 = k_{p\phi} a_{\phi_2} \quad (26)$$

$$2\zeta_\phi \omega_{n_\phi} = a_{\phi_2} + a_{\phi_2} k_{d\phi} \quad (27)$$

$$(28)$$

The proportional gain is selected so that the ailerons saturate when the roll error is e_ϕ^{max} where e_ϕ^{max} is a design parameter. Therefore we get

$$k_{p\phi} = \frac{\delta_a^{max}}{e_\phi^{max}} \text{sgn}(a_{a\phi 2}) \quad (29)$$

$$= -\frac{30}{15} \quad (30)$$

$$= -2 \quad (31)$$

The δ_a^{max} is the maximum deflection of the ailerons. The natural frequency of the roll loop is therefore given by

$$\omega_{n\phi} = \sqrt{|a_{\phi 2}| \frac{\delta_a^{max}}{e_\phi^{max}}} = \sqrt{1.3} \quad (32)$$

$$(33)$$

Which gives

$$k_{d\phi} = \frac{2\zeta_\phi \omega_{n\phi} - a_{\phi 1}}{a_{\phi 2}} \quad (34)$$

$$= \frac{2 \cdot 0.707 \cdot \sqrt{1.3} - 2.87}{-0.65} \quad (35)$$

$$= 1.935 \quad (36)$$

Where the damping ratio ζ_ϕ is a design parameter given as $\zeta_\phi = 0.707$.

An integrator is then added to the loop which gives a new transfer function

$$H_{\phi/\phi^c}(s) = \frac{k_{p\phi} a_{\phi 2} \left(s + \frac{k_{i\phi}}{k_{p\phi}} \right)}{s^3 + (a_{\phi 1} + a_{\phi 2} k_{d\phi}) s^2 + k_{p\phi} a_{\phi 2} s + k_{i\phi} a_{\phi 2}} \quad (37)$$

Since $a_{\phi 2}$ and $a_{\phi 1}$ is known, $k_{i\phi}$ can be effectively selected using root locus techniques. The closed-loop poles of the system are given by

$$1 + k_{i\phi} \left(\frac{a_{\phi 2}}{s(s^2 + (a_{\phi 1} + a_{\phi 2} k_{d\phi})s + k_{p\phi} a_{\phi 2})} \right) = 0 \quad (38)$$

From the root locus analysis in fig. 1 we see that no $k_{i\phi} > 0$ results in all poles being in the left half-plane. For all three poles to be stable, we should choose:

$$-3 < k_{i\phi} \leq 0 \quad (39)$$

We find that choosing $k_{i\phi} = 0$, thereby entirely removing the integral gain, gives the best results. This is discussed further in the next subtask.

To find the course-hold gains we look at the outer loop assuming that the inner loop has been adequately tuned such that $H_{\phi/\phi^c} \approx 1$. Using the transfer function H_χ given by

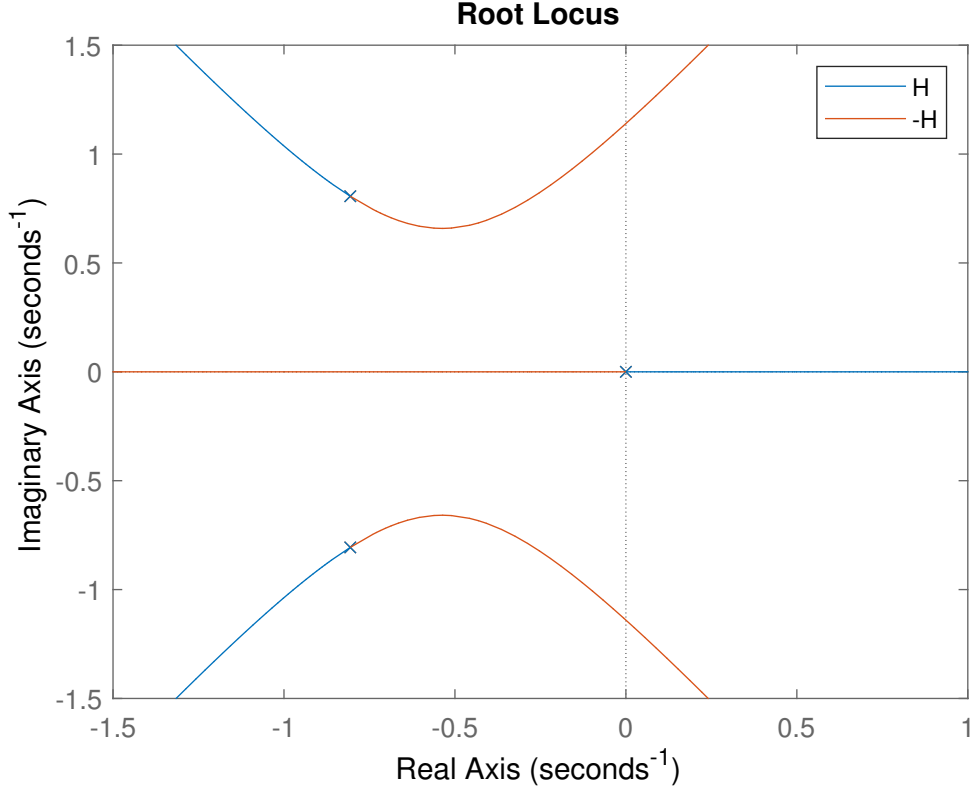


Figure 1: Root locus analysis

$$H_\chi = \frac{2\zeta_\chi \omega_{n_\chi} s + \omega_{n_\chi}^2}{s^2 + 2\zeta_\chi \omega_{n_\chi} s + \omega_{n_\chi}^2} \quad (40)$$

We use the natural frequency and damping to find the final two gains.

$$\omega_{n_\chi} = \frac{1}{W_\chi} \omega_{n_\phi} \quad (41)$$

$$k_{p_\chi} = 2\zeta_\chi \omega_{n_\chi} V_g / g \quad (42)$$

$$k_{i_\chi} = \omega_{n_\chi}^2 V_g / g \quad (43)$$

The inner loop needs to be at least 5-10 times faster than the outer loop. This is achieved by setting the bandwidth separation W_χ . Through a number of simulations, we found that $\zeta_\chi = 1.8$ and $W_\chi = 18$ gives the best response. This gives gains

$$\omega_{n_\chi} = 0.063 \quad (44)$$

$$k_{p_\chi} = 3.725 \quad (45)$$

$$k_{i_\chi} = 0.065 \quad (46)$$

$$(47)$$

Problem 2 c)

We already have an integral action in the outer course loop. This is typically sufficient, and so we should be able to achieve stability without any integral action in the inner roll loop. Also, we need the inner loop to be significantly faster than the outer loop. The integral will add both delay and instability, and thus we figure we are better off without it. This was confirmed by testing various values for k_{i_ϕ} , seeing no improvement caused by the integrator.

Problem 2d)

We have the maneuver

$$\chi_{ref} = [0 \quad 10 \quad -5 \quad -10 \quad 5 \quad 10 \quad 20 \quad 20] \quad (48)$$

With time vector

$$t = [0 \quad 60 \quad 140 \quad 230 \quad 300 \quad 380 \quad 500 \quad 600] \quad (49)$$

Which gives the simulation results seen in figure 2. Our autopilot is kind of slow, taking about 80-90 seconds to reach the desired course after a step of up to 15 degrees. However, a too fast response would be uncomfortable for the passengers of the plane. It is also assumed that the plane would not have to change course very frequently on a typical flight, meaning the extra seconds for stabilization are perfectly acceptable.

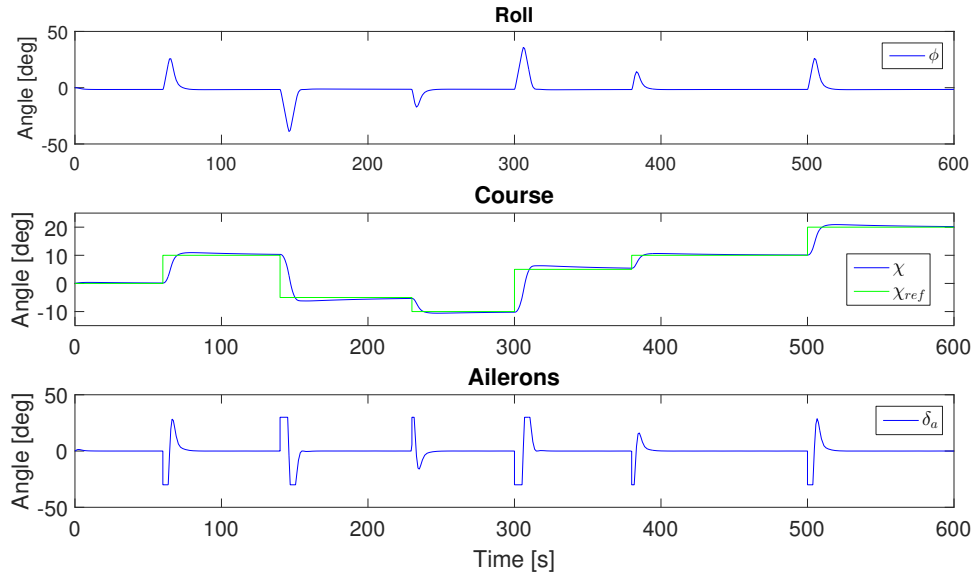


Figure 2: Roll, course inputs and response and aileron inputs for simplified model

We see that the ailerons reach saturation for a few seconds at each change in course reference, but overall the response is smooth.

Problem 2e)

Comparing figure 2 and figure 3, we see very little difference. Figure 3, based on the state-space model, shows slightly more oscillations and overshoot than the simplified model, but the

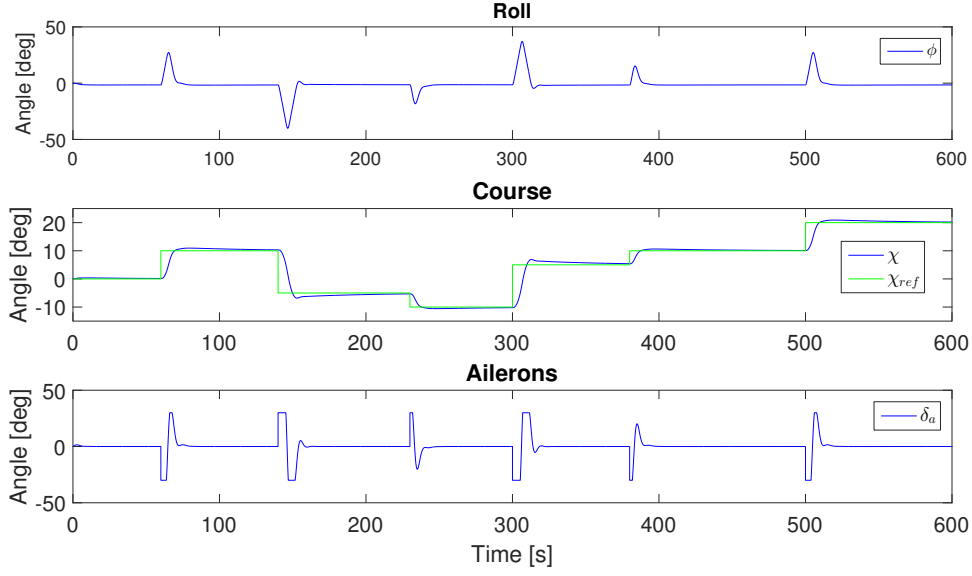


Figure 3: Roll, course inputs and response and aileron inputs for real dynamics

performance is still satisfactory. This indicates that the simplified model is a good approximation for the true dynamics. There are only some minor errors in the transient response.

Problem 2f)

Wind-up occurs whenever the aileron input is saturated. By looking at our results we find that when this happens there is a slight overshoot. Though it is not very significant, we implemented an anti-windup scheme to see if we improve performance. This solution handles the problem by using the inputs exceeding saturation limits as an added term in the integrator gain, scaled by its own gain k_{aw} . The input to the integrator then becomes:

$$k_{i_\chi} + k_{aw} * (\delta_{a,unsat} - \delta_{a,sat}) \quad (50)$$

$$(51)$$

We found from testing that the value of $k_{aw} = 0.02$ yields performance slightly better than without the anti-windup effect. This can be seen in figure 4. Here, a comparison is made between the anti-windup being active and inactive. There is noticeably less overshoot in the course angle with anti-windup, meaning its inclusion is validated.

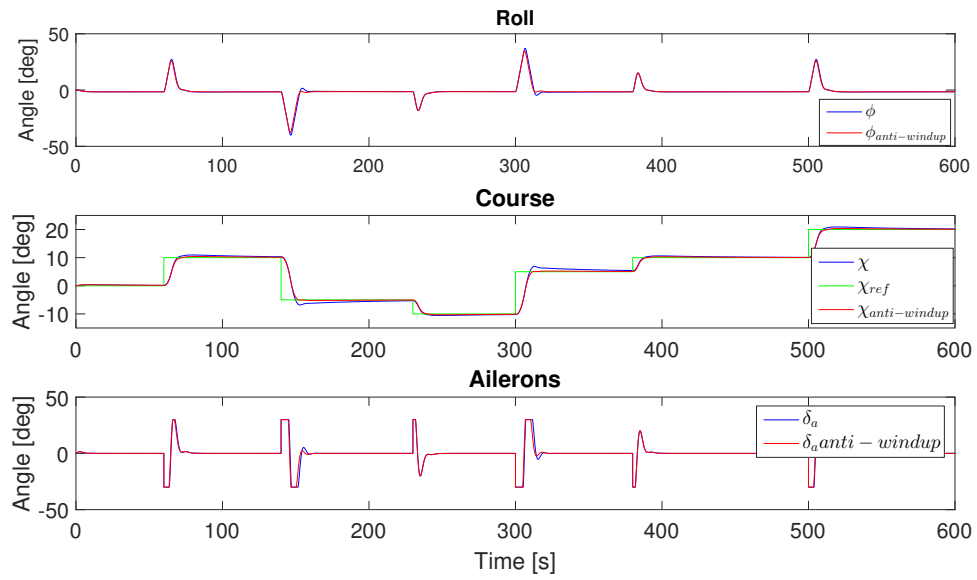


Figure 4: Roll, course inputs and response and aileron inputs for real dynamics with and without anti-windup