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High-dimensional quantum key distribution based on multicore fiber using silicon photonic integrated circuits

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Quantum key distribution provides an efficient means to exchange information in an unconditionally secure way. Historically, quantum key distribution protocols have been based on binary signal formats, such as two polarization states, and the transmitted information efficiency of the quantum key is intrinsically limited to 1 bit/photon. Here we propose and experimentally demonstrate, for the first time, a high-dimensional quantum key distribution protocol based on space division multiplexing in multicore fiber using silicon photonic integrated lightwave circuits. We successfully realized three mutually unbiased bases in a four-dimensional Hilbert space, and achieved low and stable quantum bit error rate well below both the coherent attack and individual attack limits. Compared to previous demonstrations, the use of a multicore fiber in our protocol provides a much more efficient way to create high-dimensional quantum states, and enables breaking the information efficiency limit of traditional quantum key distribution protocols. In addition, the silicon photonic circuits used in our work integrate variable optical attenuators, highly efficient multicore fiber couplers, and Mach-Zehnder interferometers, enabling manipulating high-dimensional quantum states in a compact and stable manner. Our demonstration paves the way to utilize state-of-the-art multicore fibers for noise tolerance high-dimensional quantum key distribution, and boost silicon photonics for high information efficiency quantum communications.

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INTRODUCTION

Quantum key distribution (QKD) is an attractive quantum technology that provides a means to securely share secret keys between two clients (Alice and Bob).^{1–4} Traditional QKD is based on binary signal formats, such as the BB84 protocol where the quantum information is encoded in the polarization domain.⁵ Four polarization states create a set of two mutually unbiased basis (MUBs) in a two-dimensional Hilbert space which are used for establishing quantum keys between two parties. In these binary QKD systems the information efficiency is limited to 1 bit/photon. Recently, tremendous efforts have been put into developing novel protocols to increase the information efficiency.^{6–10} High-dimensional QKD (HD-QKD) based on qudit encoding (unit of information in a N dimension space) is an efficient technique to achieve high information efficiency for QKD systems.^{11–18} Furthermore, HD-QKD protocols exhibit higher resilience to noise, allowing for lower signal-to-noise ratio (SNR) of the received signal,¹⁰ which in turn may be translated into longer transmission distances.¹⁹ One interesting way to achieve HD-QKD is to use space division multiplexing technology, where the spatial dimension is used to carry the quantum states. In this context, optical angular momentum (OAM) modes has been proposed for HD-QKD protocol, and demonstrated over a free-space link using discrete components. QKD systems will eventually be merged with optical fiber transmission, so as to combine the classical internet with a quantum internet.^{20–23} However, the transmission of OAM modes over long distance fiber links is very challenging due to inter-modal crosstalk. An alternative approach is to use separate cores in multicore fibers (MCFs), which have been widely studied for in

ultra-high capacity fiber communication, owing to the large potential multiplicity of cores and the low crosstalk between cores.^{24, 25}

On the other hand, photonic integration has played a critical role in recent quantum information revolution by integrating functionalites of traditional discrete bulky components into ultra-compact chips.^{26, 27} Photonic integrated circuits (PICs) provide excellent optical phase stability, making them particularly suitable for manipulating quantum states in compact chips with low energy consumption. Binary QKD systems have been demonstrated using integrated circuits.^{28, 29} Silicon photonics technology has been a powerful means to combine the assets of integrated photonics with complementary metal-oxide semiconductor technologies. In this paper, we demonstrate the first HD-QKD protocol based on MCFs using silicon PICs. We prove that manipulation of the high dimensional quantum states in MCFs is feasible. We successfully prepare three mutually unbiased bases with four dimensional quantum states (ququart) and send these through a seven-core fiber. High fidelity is obtained in tomography measurements. Stable and low quantum bit error rate (QBER), below threshold limits, is achieved for more than 10 min.

RESULTS

Protocol definition

Most of the QKD protocols are based on the concept of MUBs. For instance the bases $\{B_0 = |\Phi_{0,i}\rangle, i = 0, 1, \dots, N-1\}$ and $\{B_1 = |\Phi_{1,j}\rangle, j = 0, 1, \dots, N-1\}$ are orthogonal bases of a N -dimensional Hilbert space H_N . These bases are defined as mutually unbiased if and

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only if

$$|\langle \Phi_{0,i} | \Phi_{1,j} \rangle|^2 = \frac{1}{N} \quad (1)$$

In other words in a set of MUBs $\{B_0, B_1, B_2, \dots, B_k, B_n\}$ if we measure a state in B_n basis, and this state was prepared in B_k basis (with $k \neq j$), all the outputs are equally probable. It has been proven that the maximum number of the MUBs in a Hilbert space of dimension N is $N+1$, where N is a integer power of a prime number.³⁰ In our analysis we consider the case of three MUBs reported in Eq. (2) for an Hilbert space of four dimension ($N=4$). The main difference is related to the tolerable threshold of the QBER and the maximum achievable value of secret key rate. A more detailed discussion follows in next paragraph. The set of the three MUBs, used in the proposed HD-QKD system, can be defined as a linear combination of:

$$|A\rangle = \frac{1}{\sqrt{2}} (1\ 0\ 0\ 0)$$

$$|B\rangle = \frac{1}{\sqrt{2}} (0\ 1\ 0\ 0)$$

$$|C\rangle = \frac{1}{\sqrt{2}} (0\ 0\ 1\ 0)$$

$$|D\rangle = \frac{1}{\sqrt{2}} (0\ 0\ 0\ 1)$$

$$M_0 = \begin{pmatrix} |A\rangle + |B\rangle \\ |A\rangle - |B\rangle \\ |C\rangle + |D\rangle \\ |C\rangle - |D\rangle \end{pmatrix} \quad M_1 = \begin{pmatrix} |A\rangle + |C\rangle \\ |A\rangle - |C\rangle \\ |B\rangle + |D\rangle \\ |B\rangle - |D\rangle \end{pmatrix} \quad M_2 = \begin{pmatrix} |A\rangle + |D\rangle \\ |A\rangle - |D\rangle \\ |B\rangle + |C\rangle \\ |B\rangle - |C\rangle \end{pmatrix} \quad (2)$$

We can easily prove that the set M_0, M_1, M_2 satisfies the mutually unbiased assumption, $|\langle M_0 | M_1 \rangle|^2 = |\langle M_0 | M_2 \rangle|^2 = |\langle M_1 | M_2 \rangle|^2 = 1/4$. Physically, $|A\rangle, |B\rangle, |C\rangle$, and $|D\rangle$ represent the quantum states related to the four individual cores of a multi-core fiber, as shown in the inset of Fig. 1a. By tuning the Mach-Zehnder interferometers (MZIs), situated in the transmitter chip, Alice

creates a quantum superposition between cores. In this way she prepares one of the states in the three MUBs. A random number generator is used for basis and states choice. Before the quantum measurement, Bob randomly choose one of the three MUBs tuning the corresponding MZI. In such a way, Bob creates interference between different cores and he correctly measures the quantum states sent by Alice. As in the BB84 protocol, after the measurements, a distillation process is required. In this procedure Alice and Bob discard all the data related to a different basis choice. At this point Alice and Bob share an identical quantum key, useful for encryption and decryption of the plain text.

Secret key rate

One of the most important parameters in a communication system is the achievable rate. In particular, in a QKD system, the major criterion is represented by the secret key rate: number of bit/s or bit/pulse that Alice and Bob can establish as useful key. A general formula for the standard protocols can be written as:

$$R \geq I_{AB} - \min(I_{AE}, I_{BE}), \quad (3)$$

where I_{AB} represents the classical mutual information between Alice and Bob ($I_{XY} = H(X) - H(X|Y)$), and the marginal entropy is defined as $H(X) = \sum_{x \in X} p(x) \log p(x)$. The right term of Eq. (3), $\min(I_{AE}$ and I_{BE}), is related to the quantum mutual information between Alice and Eve or Bob and Eve. In the following analysis we consider only the mutual information between Alice and Eve, but a more complete analysis can be done: in this way the secret key rate value is estimated with a lower bound. Furthermore, we work under the assumption of trusted-device scenario, in which Eve cannot modify the efficiency of Bob's detectors. In case of ququart QKD system the final secret key rate formula must be adapted. The mutual information between Alice and Bob is:

$$I_{AB} = \log_2(N) + F \log_2(F) + (1-F) \log_2\left(\frac{1-F}{N-1}\right), \quad (4)$$

where N is the dimension of the Hilbert space and $F=1-D$ represents the fidelity of Bob. D is defined as the disturbance in the communication link. In order to extract a bound on the final secret key rate, we should define the best strategy for Eve. In the following analysis we consider two different kinds of Eve's strategy. Individual attacks (IAs), where Eve monitors the ququarts independently from each other, and coherent attacks (CAs). CAs instead are less conservative on Eve's strategy, in fact she can

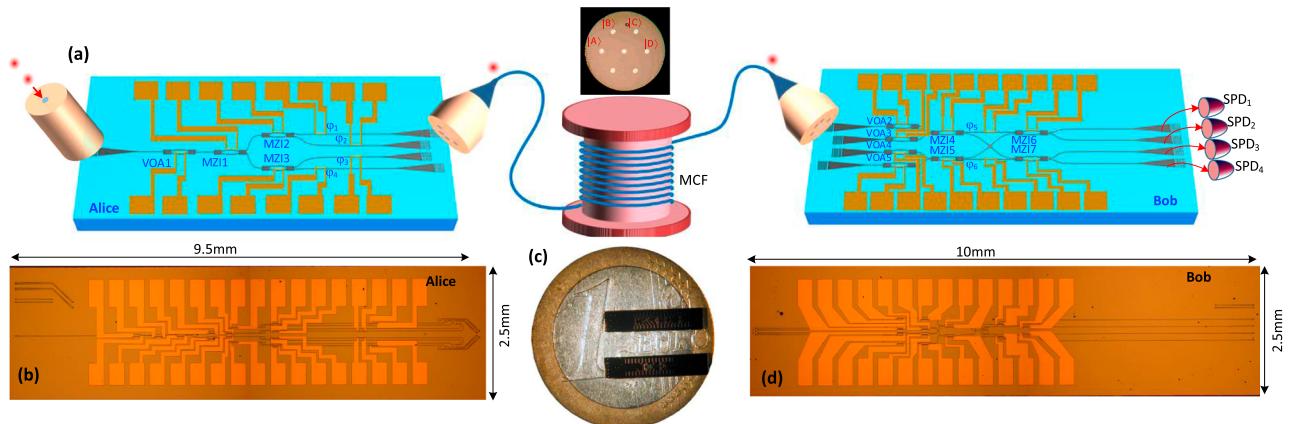


Fig. 1 **a** Schematic of the HD-QKD based on MCF using silicon PIC for Alice and Bob. The inset shows the cross-section of the multi-core fiber, where four cores are used. **b, d** Shows the fabricated silicon PIC for Alice and Bob, respectively. **c** Presents the picture of the fabricated chips with a 1 euro coin, indicating the compact size of the silicon PICs

monitor more quantum states jointly. Intuitively CAs are less stringent than IAs.

Let's start our analysis under the assumption of IAs where Eve uses a universal quantum cloning machine for qudits. By focusing on the two MUBs we retrieve the mutual information between Alice and Eve depending on the number of states N . In the case of ququarts encoding we define four different quantum states as $|0\rangle, |1\rangle, |2\rangle, |3\rangle$. The most general symmetric eavesdropping strategy for ququarts can be written as:

$$|0\rangle|E\rangle \xrightarrow{U} \sqrt{1-D}|0\rangle|E_{00}\rangle + \sqrt{\frac{D}{N}}|1\rangle|E_{01}\rangle + \sqrt{\frac{D}{N}}|2\rangle|E_{02}\rangle + \sqrt{\frac{D}{N}}|3\rangle|E_{03}\rangle$$

$$|1\rangle|E\rangle \xrightarrow{U} \sqrt{\frac{D}{N}}|0\rangle|E_{10}\rangle + \sqrt{1-D}|1\rangle|E_{11}\rangle + \sqrt{\frac{D}{N}}|2\rangle|E_{12}\rangle + \sqrt{\frac{D}{N}}|3\rangle|E_{13}\rangle$$

$$|2\rangle|E\rangle \xrightarrow{U} \sqrt{\frac{D}{N}}|0\rangle|E_{20}\rangle + \sqrt{\frac{D}{N}}|1\rangle|E_{21}\rangle + \sqrt{1-D}|2\rangle|E_{22}\rangle + \sqrt{\frac{D}{N}}|3\rangle|E_{23}\rangle$$

$$|3\rangle|E\rangle \xrightarrow{U} \sqrt{\frac{D}{N}}|0\rangle|E_{30}\rangle + \sqrt{\frac{D}{N}}|1\rangle|E_{31}\rangle + \sqrt{1-D}|2\rangle|E_{32}\rangle + \sqrt{\frac{D}{N}}|3\rangle|E_{33}\rangle$$

Under the assumption of symmetric attacks (Eve treats all the input states with the same strategy) and following the approach defined in¹⁰ the fidelity is: $F = \text{Tr}(|\psi_i\rangle\langle\psi_i|\rho_B^{\text{out}})$, where ρ_B^{out} is the reduced density operator of the state send to Bob. The maximum value of Eve's mutual information is:

$$I_{AE} = \log_2(N) + F_E \log_2(F_E) + (1 - F_E) \log_2\left(\frac{1 - F_E}{N - 1}\right), \quad (5)$$

where the corresponding optimal fidelity for Eve is:

$$F_E = \frac{F}{N} + \frac{(N-1)(1-F)}{N} + \frac{2}{N} \sqrt{(N-1)F(1-F)} \quad (6)$$

In other words, Eq. (5) represents the ability of Eve to distinguish between N non-orthogonal states. In the case of IAs, the term I_{AE} can be easily formulated using F_E instead of F in Eq. (4). By using Eq. (3) we can define an upper bound on the disturbance D introduced by Eve. We highlight that this limit increases with the Hilbert space dimension (see Supplementary Information Table S1). As evidence we report in Fig. 2 the theoretical mutual information in case of different Hilbert space dimension (from $N=2$ to $N=8$) and under the assumption of IAs and CAs. The

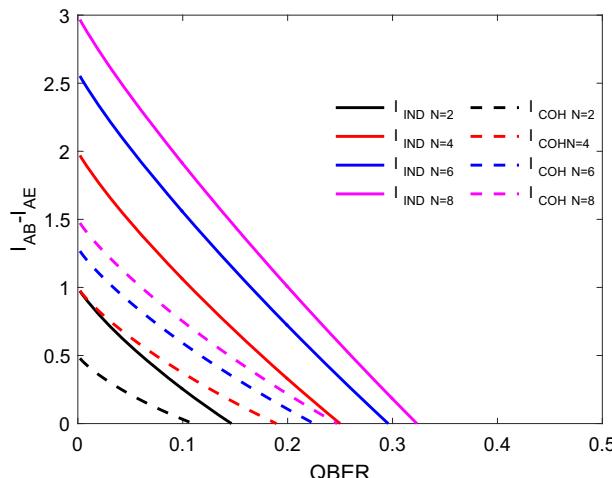


Fig. 2 Mutual information compared with different Hilbert space dimension. The mutual information thresholds between Alice and Bob/Eve in case of different Hilbert space dimensions and different Eve's strategy. These curves are obtained in the case of two MUBs with individual and coherent attacks

intersections between the x -axis and the color filled/dashed lines represent the QBER thresholds. The higher N used in the system, the higher QBER threshold is allowed in the key generation. This means that higher dimension used in the QKD protocol leads to higher resilience to noise, allowing lower SNR for the received signal, which can be translated into longer transmission distances. Following the same approach it is possible to define a bound on Eve's information related to the case of $N+1$ MUBs.¹⁰

Let's now consider a different strategy for Eve. As introduced above in case of CAs, Eve collects a finite number of ququart pulses, $n < \infty$. In this case under the assumption made in¹⁰ and³¹, it is possible to define an upper bound on Eve's information. The condition $I_{AB} \geq \log_2(N)/2$ is a sufficient case for guaranteeing the security under CAs, if the key dimension I is much greater than n .³¹ To be stressed that all the previous bounds are true under the assumption of infinity key length, but a different approach can be used for defining security under the case of finite-key scenario.

High-dimensional decoy state analysis

Most of the QKD systems encode the bit information using different degrees of freedom like polarization, phase and time, carried by a train of weak coherent pulses (WCPs). This idea, although unconditionally secure from a theoretical point of view, is subjected to some limitation on a practical realization. In fact, WCPs are obtained by attenuated laser with a mean photon number per pulse, μ , lower than one. However, the Poissonian distribution of the laser does not guarantee a non zero probability of multi-photon states. In that case, Eve can easily stop all the single photon pulses and split the multi-photon pulses (PNS—photon number splitting attack). In such a way Eve can measure the same information as Bob and after information reconciliation Alice, Eve, and Bob share the same key. In order to overcome this problem a simple strategy can be adopted-decoy state method. During the transmission process, Alice randomly changes the mean photon number per pulse.³² A intensity tunable laser or a real time variable optical attenuator (VOA1) can be used to create different levels of mean photon numbers. From a theoretical point of view this technique, used with infinite different number of decoy states, permits to avoid the PNS attack and to increment the key rate for further distances. However, as reported and proved in,³² two different decoy values (plus a vacuum measurement), are sufficient to establish a good compromise between rate and security. Our analysis follows the same procedure as introduced in,³² by using a practical implementation of decoy-state method with two weak decoy-states (with average photon flux $\nu < \mu < 1$) and a vacuum state. The final secret key rate can be written as:

$$R \geq \frac{1}{M} [Q_0 \log_2(N) + Q_1 (\log_2(N) - h_{HD}(e_1)) - Q_u h_{HD}(D)f(D)], \quad (7)$$

where M is the number of bases used in the protocol (in our QKD experiment we used $M=2$), D is the overall QBER as already defined is the disturbance in the communication link, e_1 is the error-rate of the single-photon state and $f(D)$ is the efficiency of the error correction code ($f(D) = 1.05$). The Shannon entropy in case of high-dimensionality can be adapted as $h_{HD}(x) = -x \log_2(x/(D-1)) - (1-x) \log_2(1-x)$. The gain of the signal decoy state (i.e., the probability of obtaining a detection when the signal state μ is sent) is defined like $Q_u = Y_0 + 1 - \exp(-\eta\mu)$, where Y_0 is the yield of the vacuum state (probability of the dark counts) and η is the overall efficiency ($\eta = \eta_{Bob} \eta_{AB} \eta_q$). The link transmittance parameter $\eta_{AB} = 10^{-\alpha_f L/10}$, where α_f [dB/km] is the fiber losses (average losses of a MCF about 0.2 dB/km) and L is the link distance in kilometer. e_0 is the QBER associated to the vacuum state ($e_0 = (N-1)/N$) corresponding to the probability of a random dark count in one of the N detectors. The values of Q_u and D can be directly measured when Alice send the signal state to Bob. Q_0 can be

directly estimated as $Q_0 = e^{-\mu Y_0}$. Note that $Q_0 \log_2(N)$ indicates the gain relative to the dark counts, where Eve has no information. However, the gain and the error of the single photon state (Q_1 and e_1) must be approximated by the following formulas:

$$Q_1 \geq \left(\frac{e^{-\mu} u^2}{\mu v - \nu^2} \right) \left(Q_v e^\nu - Q_u e^\mu \frac{\nu^2}{\mu^2} - Y_0 \frac{\mu^2 - \nu^2}{\mu^2} \right) \quad (8)$$

$$e_1 \leq \frac{\mu(E_v Q_v e^\nu - e_0 Y_0)}{\nu Q_1 e^\mu} \quad (9)$$

By using the decoy-state method is possible to avoid the main problem of the multi-photon state generated by the WCPs, used in the current experiment. This technique allows to establish an higher secret key rate for longer distances comparing the performance of the same protocol without decoy-state.

Experimental results

The experimental scheme of the proposed HD-QKD based on MCF using silicon PICs is shown in Fig. 1a (the detailed experimental setup is presented in Supplementary Information, Fig. S2). WCPs (10 ns wide, see also [Supplementary Information](#)) are injected into the transmitter chip (Alice, Fig. 1b). The silicon PIC for Alice is used to prepare the high-dimensional quantum states. The detailed schematic of the chip design is found in Fig. S1 in [Supplementary Information](#). A VOA1 is used to thermally tune the pulse power to prepare the decoy states. Cascaded MZIs associated with the four thermally tunable phase shifters are used to prepare the quantum

states in the three MUB sets with mean number of photons per pulse, μ , lower than 1. As an example, MZI 1 can be set so that light goes to the upper arm. At the same time, MZI 2 can be set as a 3 dB coupler. In this case, M_0 basis is prepared, and states $|A\rangle + |B\rangle$ and $|A\rangle - |B\rangle$ can be prepared by controlling the phase difference between the two output arms of MZI 2 to 0 or π , which is obtained through thermally controlling φ_1 and φ_2 . Similarly, the other quantum states can be prepared by configuring the corresponding MZIs and phase shifters. The total insertion loss of Alice chip is 15 dB, which includes the coupling losses of the grating coupler with the input single mode fiber and the grating coupler based MCF fan-in/fan-out (FI/FO), and all the losses by other components on the chip. The quantum states are coupled to four cores of a multi-core fiber through an apodized grating coupler based MCF FI/FO.^{33, 34} After the MCF, the quantum states are coupled into Bob's chip (Fig. 1d) through the MCF FI/FO coupler, and randomly measured in one of the MUBs by configuring the corresponding phase shifters and MZIs. For instance, by configuring MZI 4 as a 3 dB coupler and setting MZI 6 and MZI 7 to directly go through, Bob is configured to receive the quantum states in base M_0 . In order to measure correctly the quantum states, it is critical to obtain balanced losses for the four spatial channels, which is enabled by four VOAs introduced in Bob's chip. The total insertion loss of Bob chip is 8 dB. We firstly measured the state tomography of the 3 MUBs. Figure. 3a, b show the theoretically and the experimental data acquired for the three MUBs. In this measurements, a weak coherent pulse with $\mu < 0.1$ at 10 kHz was injected to Alice, and Bob acquires data for 30 s. The resulting fidelity of the

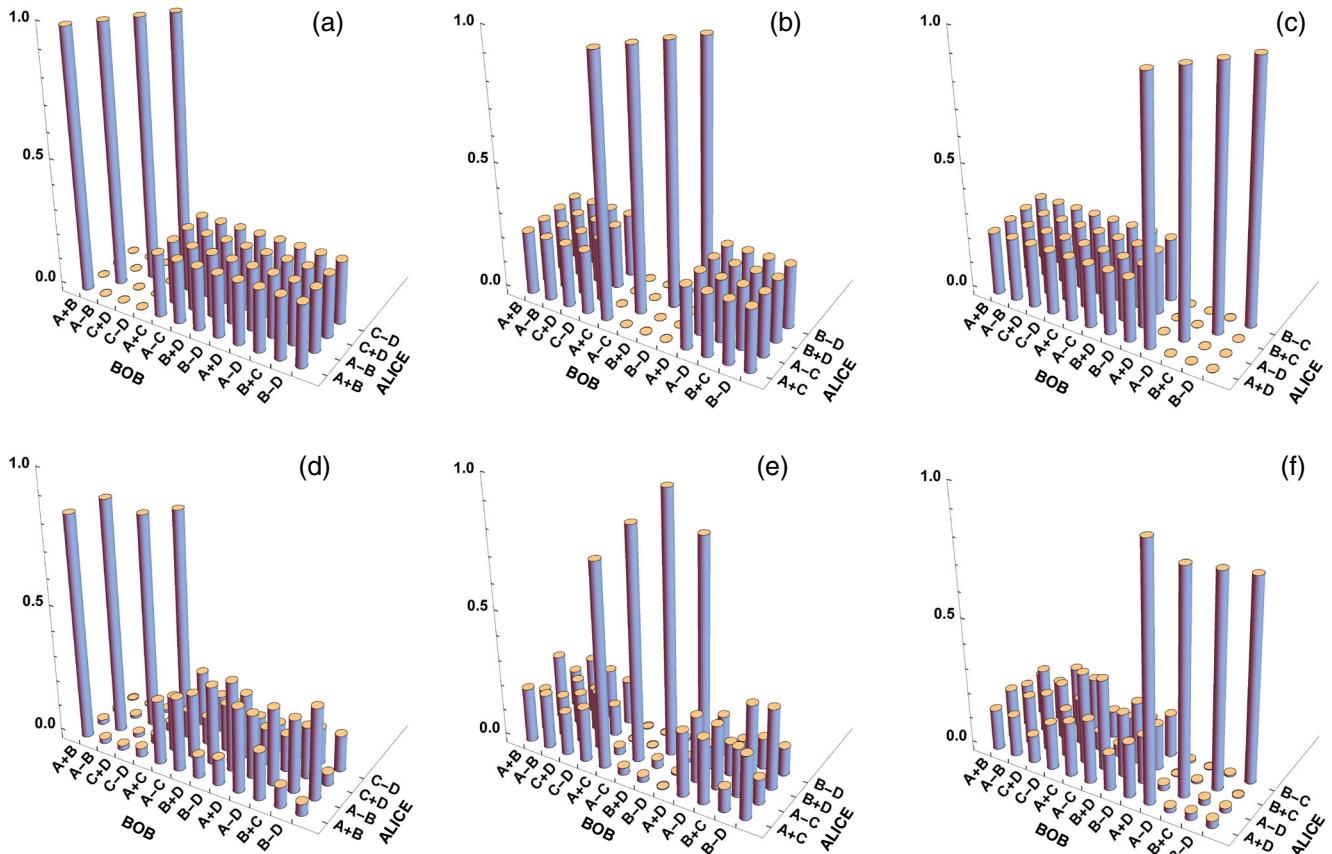


Fig. 3 State tomography of the three MUBs: **a, b, c** represent the theoretical matrices of the different MUBs; **d, e, f** show the experimental measurements of the same basis. The state tomography were obtained with 30 s of integration time using a CW 1550 nm laser modulated with a 10 ns pulse at 10 kHz of repetition rate and a mean photon per pulse of 0.2 ± 0.006 . By using the (classical) definition of fidelity ($F(x, y) = \sum_i (p_i q_i)^{1/2}$, where x and y are random variables and q_i and p_i are the vector of probability distribution), we obtain an average fidelity of 0.977 ± 0.01

Table 1. Measured parameter of the decoy-state method used in the current experiment

Parameter	Experimental value
Q_u	$43.8 \times 10^{-2} \pm 0.9 \times 10^{-2}$
E_u	0.136 ± 0.011
Q_v	$24.3 \times 10^{-2} \pm 0.7 \times 10^{-2}$
E_v	0.118 ± 0.013
Q_0	$1.2 \times 10^{-5} \pm 7 \times 10^{-8}$

In particular we used a two-value ($v < \mu < 1$) and vacuum decoy-state.

corresponding matrices are 0.973 ± 0.006 ; 0.981 ± 0.0064 ; 0.975 ± 0.006 with an average value of 0.976 ± 0.0109 . Secondly, as a proof of concept, 3 m of MCF is used to prove a real-time QKD experiment with two MUBs with a 5 kHz repetition rate (currently limited by the on-chip thermal heaters, see [supplementary information](#)). Alice, by using a PRBS sequence (8 bit) generated by an FPGA, randomly chooses one of the four states in one of the two bases and set the corresponding heaters in order to create the required quantum state. Moreover, by randomly tuning the VOA1 we implement the decoy-state technique. In particular the decoy values are reported in Table 1, and represents the signal and decoy state parameters. By using the same procedure, Bob's heaters are configured in one of the two MUBs. The choice of the basis is randomly decided by a PRBS sequence (7 bit), generated by the same FPGA. In the succeeding distillation process, counts measured in the wrong basis are discarded. In Fig. 4a we presents the QBER as a function of time and with an average value of $\mu = 0.276 \pm 0.006$ and $v = 0.153 \pm 0.004$ photon per pulse. Good and stable QBER performance with an average value of 13% is achieved for more than 10 min, which is well below both coherent attach limit and individual attach limit. In Fig. 4b we report the measured value of the decoy-state gain Q_u . The average value of $43.8 \times 10^{-2} \pm 0.9 \times 10^{-2}$ indicates the low losses present on Alice side and the stability of the MCFs grating coupler.

DISCUSSION

The results reported in Fig. 4a, b show a stable and good demonstration of the HD-protocol. The average value of 13% of the QBER, obtained over 10 min of measurement, represents a good point for the key extraction. Moreover, thanks to the on-chip VOA1, we implemented a real-time decoy state technique. The experiment was performed using two of three MUBs prepared and showed in Fig. 3. We would like to highlight that the maximum number of MUBs in an Hilbert space of four-dimension is five. However, by using five MUBs we will compromise the final secret key rate. In fact, although the security increases using more bases, the secret key rate (Eq. (7)) scales dependently with the number of the bases. As reported in Table S1 of the Supplementary information, for a space of four-dimension and in the case of IAs, less than 6.5% of disturbance can be gained in the QBER threshold respect to a leak of a factor 6/10 on the key rate. The user must choose a trade off between security and rate, depending on the actual conditions of the channel. As a comparison, the case of multiple independent quantum keys (one per each core) in a MCF must be considered. The final composed rate would possibly be higher as compared to the HD one. However more strict requirements must be satisfied. As an example, the QBER of the classical BB84 protocol must be lower than 11% (Fig. 2), and after a long transmission distance the fluctuation of phase and polarization due to the perturbation in the fiber becomes relevant. Furthermore, if we introduce the concept of photon information efficiency ¹⁴ our HD-QKD allows a

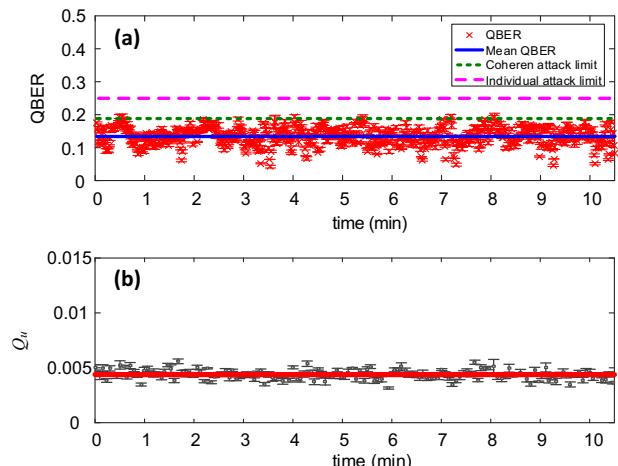


Fig. 4 Experimental results. **a** QBER for 10 min of acquired data. The blue line represents the average QBER about 13%, which is well below the limits of both coherent attach and individual attacks. **b** Measured value of the signal-state gain Q_u . The average value of $43.8 \times 10^{-2} \pm 0.9 \times 10^{-2}$ indicates the low losses present on Alice side and the stability of the MCFs grating coupler

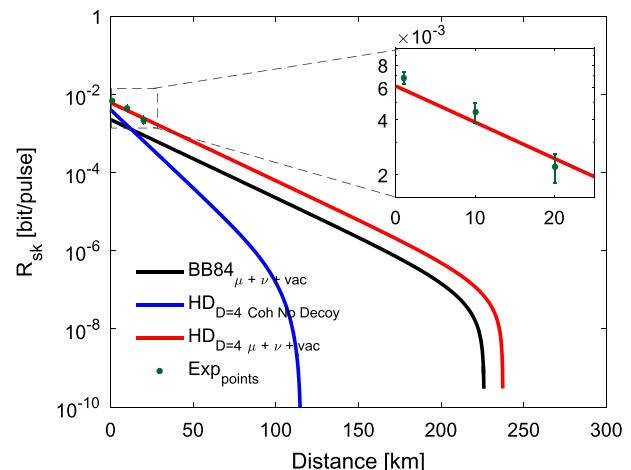


Fig. 5 Comparison between experimental data and theoretical simulation. We overlapped the experimental data with the expected theoretical rate as a function of the link distance. The parameters used are: $\rho_{\text{dark}} = 2 \times 10^{-8}$, $\eta_d = 0.1$, $a_{\text{loss}} = 0.2 \text{ dB/km}$, $u = 0.3$, $v = 0.15$ for the decoy rates. Blue solid line is the rate without decoy state in case of WCP HD-QKD. Red and black solid lines are the simulated decoy state key rate for qubit or ququart encoding

gain of 0.65 bit/detected photon (from 0.806 to 1.461). In Fig. 5 we show the simulation of the secret key rate as a function of the link distance in the case of HD-QKD with decoy state, BB84 QKD with decoy state and HD-QKD without decoy state. It follows that the best performance in terms of rate and achievable distance is reached with the case of HD-QKD and decoy-state. We also traced, as a comparison, the experimental data acquired by using the four VOAs integrated in Bob's chip in order to simulate the link losses. It is well known in optical communication, that phase and polarization of coherent light is changed in a long optical fiber transmission. This effect is mainly due to the environment changes during transmission. In a MCF, each core acts independently from the other with very low crosstalk. However the phase and polarization of photons drift independently in each core. In our experiment this effect was mitigated by the short link distance, but in a real QKD system it must be considered. A close

loop phase stabilization system can be implemented for a longer transmission distance, as reported in.³⁵ In order to avoid influence by polarization drift, two different strategies can be used. Firstly, a two dimensional grating coupler associated with an MZI³⁶ can be used to couple with the cores of the MCF, so that the polarization for each core can be tuned independently. Moreover, the polarization dependence can be solved by an on-chip polarization diversity circuit^{37, 38} for Bob's chip. In addition, Bob is performing interference between cores for receiving quantum states in different basis, thus the time delay between each cores should be matched after long transmission. This problem can be solved by introducing a cascaded delay²⁸ on Alice chip.

The pulse rate used in this experiment of 10 kHz (for characterization) and 5 kHz (for the real-time QKD experiment) represents the main limitation for a long link deployment of this technology. The main limitation is the slow thermal dynamics of the heaters used in our experiment. In fact, despite a very high extinction ratio and good stability, the thermal tunable MZIs on Alice's and Bob's chips have low switching time of 66 and 27 μ s for rise and fall time, as detailed in *Supplementary Information*. In order to overcome this problem, two main ideas can be implemented. Different kinds of material, such as graphene, that has ultra-high thermal conductivity enabling sub-microsecond tuning speed can be investigated.^{39, 40} Ultra-high speed p-i-n or p-n junction,^{41, 42} or silicon-lithium niobate hybrid integration⁴³ can also be utilized for the phase shifters to further increase the repetition rate. The insertion loss of receiver chip is critical for final key rate. Our current device has an insertion loss of 8 dB, which attributes from the ~4 dB coupling loss of the grating couplers. More efficient grating couplers⁴⁴ can be applied to improve the coupling loss. In addition, significant improvements will be achieved by introducing an aluminum (Al) mirror below the grating couplers to improve the coupling efficiency.^{45, 46}

Once these technical problems is solved, the idea of space multiplexing HD-QKD can be considered feasible for an end-to-end high-rate transmission over long distance. Moreover, thanks to the advantages of state of the art silicon photonics, laser sources can be integrated on Alice chip in order to create a miniaturized quantum transmitter. Transistors, switches, digital-analog converters (DACs) and other electronic devices can also be integrated on a silicon substrate creating a stable and powerful chip. On the receiver side, fully demonstration of single photon detector on chip was already proved during the last years.⁴⁷ The best solution for HD-QKD is based on single photon arrays. In this way the scalability of the process can be improved and higher capacity protocols can be implemented.

MATERIALS AND METHODS

Device fabrication

The silicon PICs was fabricated on the commercial SOI wafer with top silicon thickness of 250 nm and buried oxide layer (BOX) of 1 μ m. A single step of standard SOI processing, including e-beam lithography and inductively coupled plasma etching was first used to fabricate the whole silicon PIC simultaneously. A 1500 nm thick layer of SiO₂ was then deposited on top of the chip by plasma-enhanced chemical vapor deposition. The chip surface was then polished, and the top SiO₂ was thinned down to 1 μ m accordingly. The 1 μ m SiO₂ was used as an isolation layer between the silicon waveguide and the Ti heaters fabricated later to avoid potential optical losses. Afterwards, the 100 nm thick titanium heaters are formed by e-beam lithography followed by metal deposition and liftoff process. After that, the thick Au/Ti contact layer was fabricated by UV lithography followed by metal deposition and liftoff process. The chip was then cleaved and wire-bonded to a PCB board, and controlled by field programmable gate array (FPGA) for system experiment.

Electronic design

The chip-to-chip HD-QKD based on space-division multiplexing is feasible thank to a real time control of the different MZIs presented on the silicon

chip. These MZIs, as reported above, are controlled by heaters: conductor material which change his property when a voltage is applied. In order to tune in real-time these MZIs, different electrical signals are required in the transmitter and receiver side. An Altera FPGA board emits eight digital parallel outputs every 0.2 ms, which are converted into analog voltages by eight DACs. Then, these analog signals are send to the transmitter and the receiver PCB board by flat cables.

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AUTHOR CONTRIBUTIONS

Y. D., D. B., X. C., and X. Z. proposed the idea. Y. D. designed and fabricated the silicon PICs. K. D. designed the electrical controlling circuits. Y. D., D. B., and K. D. performed the system experiment. D. B. carried out the theoretical analysis on the proposed protocol. Y. D., D. B., K. R., and L. K. Oxenlwe discussed the results. All authors contributed to the writing of the manuscript.

ADDITIONAL INFORMATION

Supplementary Information accompanies the paper on the *npj Quantum Information* website ([doi:10.1038/s41534-017-0026-2](https://doi.org/10.1038/s41534-017-0026-2)).

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