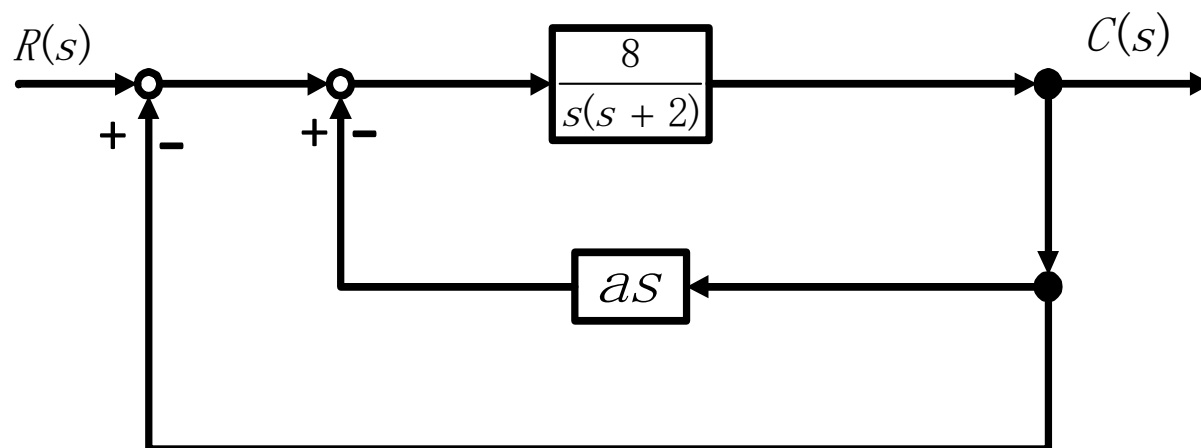


习题课

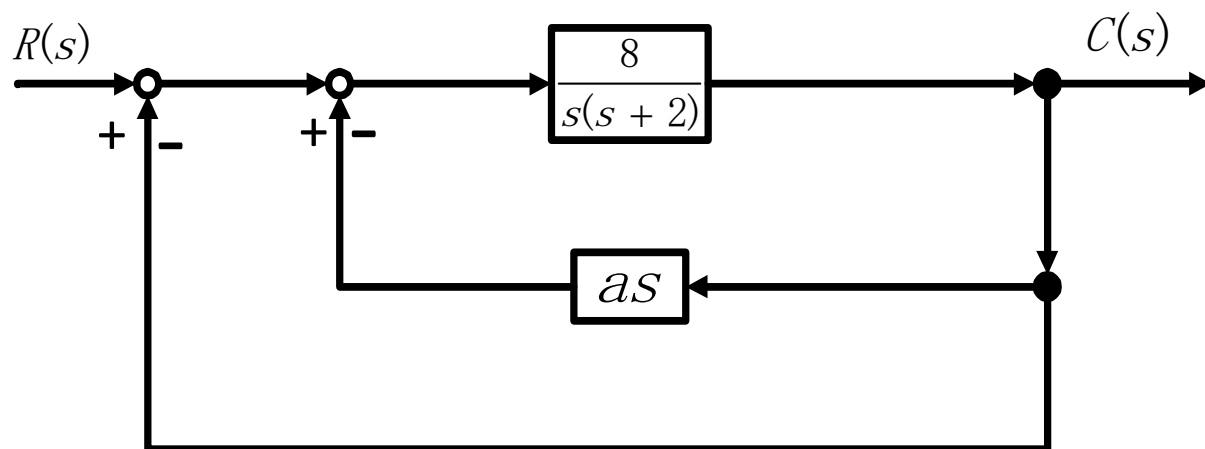
习题一



- 1、当 $a = 0$ 时，确定系统的阻尼比 ξ ，自然频率 ω_n 和单位斜坡输入时的稳态误差
- 2、当 $\xi = 0.7$ 时，确定参数 a 的值以及单位斜坡输入时的稳态误差
- 3、在保证 $\xi = 0.7$ 和稳态误差 $\text{ess}(\infty) = 0.25$ 时，确定参数 a 及前向通道增益 K

习题课

习题一



1、当 $a = 0$ 时

$$G(s) = \frac{8}{s(s+2)} = \frac{\omega_n^2}{s(s+2\zeta\omega_n)}$$

$$\omega_n = 2\sqrt{2} = 2.83, \zeta = \sqrt{2} / 4 = 0.35$$

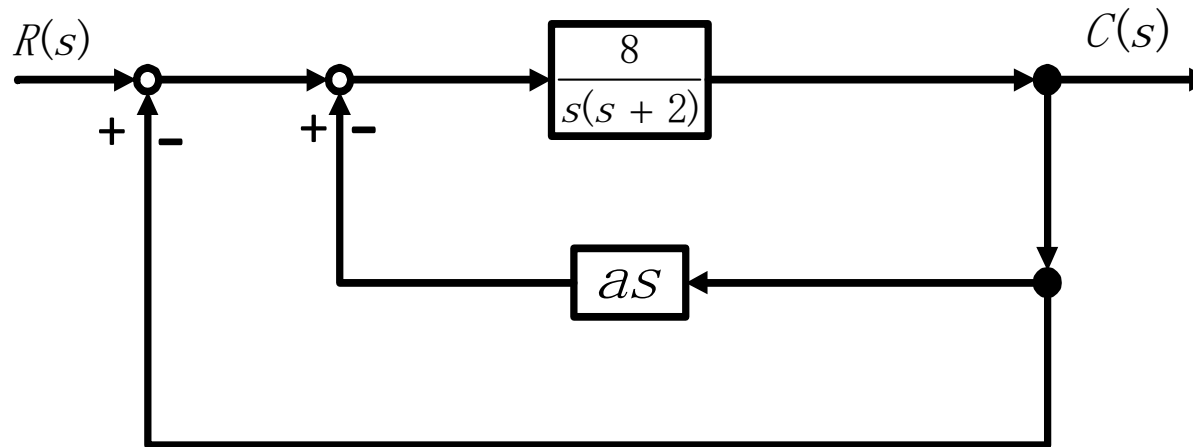
I型系统

$$K_v = 4$$

$$e_{ss}(\infty) = 1 / K_v = 0.25$$

习题课

习题一



2、当 $a \neq 0, \zeta = 0.7$

$$G(s) = \frac{8}{s(s+2+8a)} = \frac{\omega_n^2}{s(s+2\zeta\omega_n)}$$

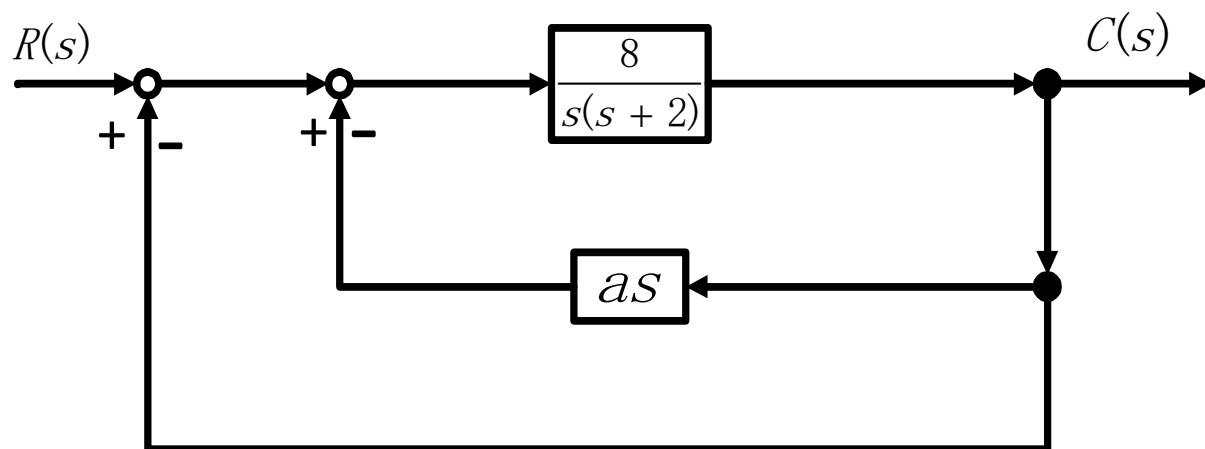
$$\omega_n = 2\sqrt{2} = 2.83, a = \frac{\zeta\omega_n - 1}{4} = 0.245$$

$$K_v = 8 / (2 + 8a) = 2.02$$

$$e_{ss}(\infty) = 1 / K_v = 0.495$$

习题课

习题一



3、当 $\zeta = 0.7, e_{ss}(\infty) = 0.25$

$$G(s) = \frac{K}{s(s+2+Ka)} = \frac{\omega_n^2}{s(s+2\zeta\omega_n)} \quad K_v = K / (2 + Ka)$$

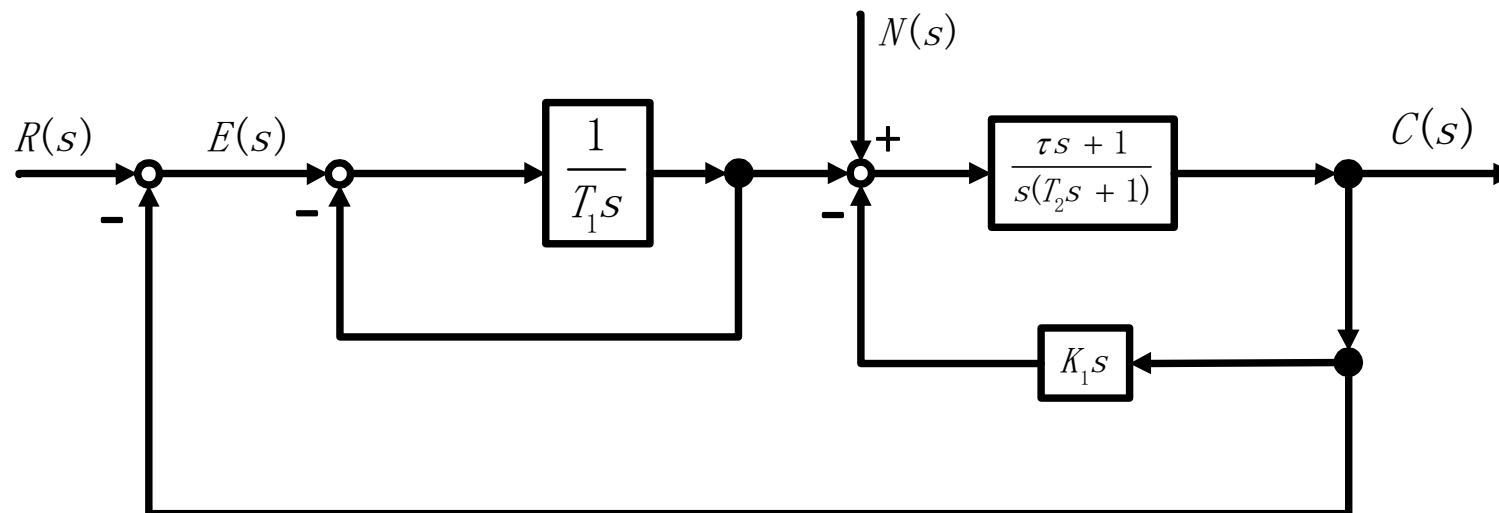
I型系统 $e_{ss}(\infty) = 1 / K_v = 2 / K + a$

代入

$$\zeta = 0.7, e_{ss}(\infty) = 0.25 \quad \Rightarrow \quad \begin{cases} \omega_n^2 = K \\ 1.4 \cdot \omega_n = 2 + Ka \\ 2 / K + a = 0.25 \end{cases} \Rightarrow \begin{cases} \omega_n = 5.6 \\ K = 31.36 \\ a = 0.186 \end{cases}$$

习题课

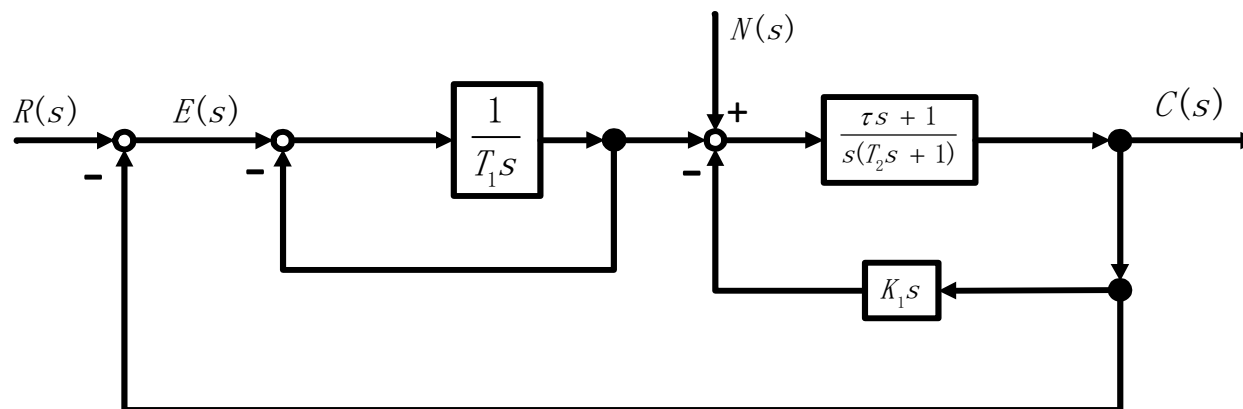
习题二



试鉴别系统对输入 $r(t)$ 和扰动 $n(t)$ 的类型

习题课

习题二



当 $n(t) = 0$ 时，开环传递函数

$$G_r(s) = \frac{\frac{1}{T_1 s}}{1 + \frac{1}{T_1 s}} \cdot \frac{\frac{\tau s + 1}{s(T_2 s + 1)}}{1 + \frac{K_1 s(\tau s + 1)}{s(T_2 s + 1)}} = \frac{\tau s + 1}{s(T_1 s + 1) [(T_2 s + 1) + K_1(\tau s + 1)]} \quad \text{I型系统}$$

当 $r(t) = 0$ 时，开环传递函数

$$\text{前向通道} \quad G_n(s) = \frac{\tau s + 1}{s(T_2 s + 1)} \quad \text{反馈通道} \quad H_n(s) = \frac{1}{T_1 s + 1} + K_1 s = \frac{K_1 T_1 s^2 + K_1 s + 1}{T_1 s + 1}$$

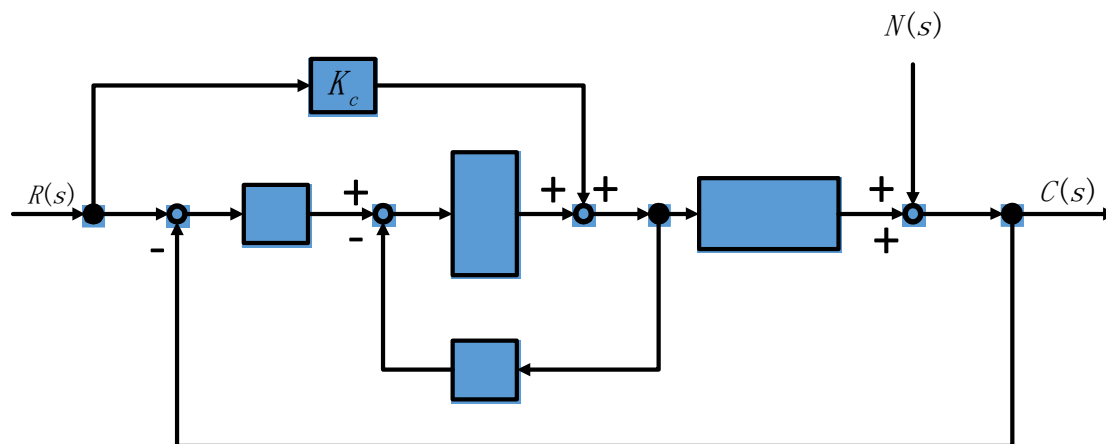
$$E_n(s) = R(s) - C_n(s) = -C_n(s) = -\frac{G_n(s)}{1 + G_n(s)H_n(s)} N(s)$$

$$= -\frac{(\tau s + 1)(T_1 s + 1)}{s(T_1 s + 1)(T_2 s + 1) + (\tau s + 1)(K_1 T_1 s^2 + K_1 s + 1)} N(s)$$

0型系统

习题课

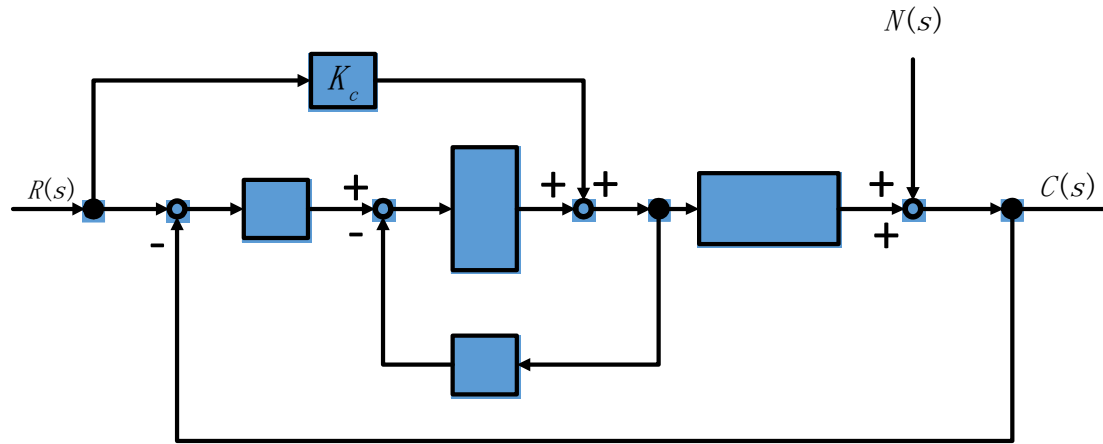
习题三



- 1、计算扰动 $n(t) = t$ 引起的稳态误差
- 2、设计 K_c ，是系统在 $r(t) = t$ 作用下无稳态误差

习题课

习题三



1、 $n(t) = t$

$$N(s) = \frac{1}{s^2} \quad \frac{C(s)}{N(s)} = \frac{s(s + K_2 K_3) (Ts + 1)}{s(s + K_2 K_3) (Ts + 1) + K_1 K_2 K_4}$$

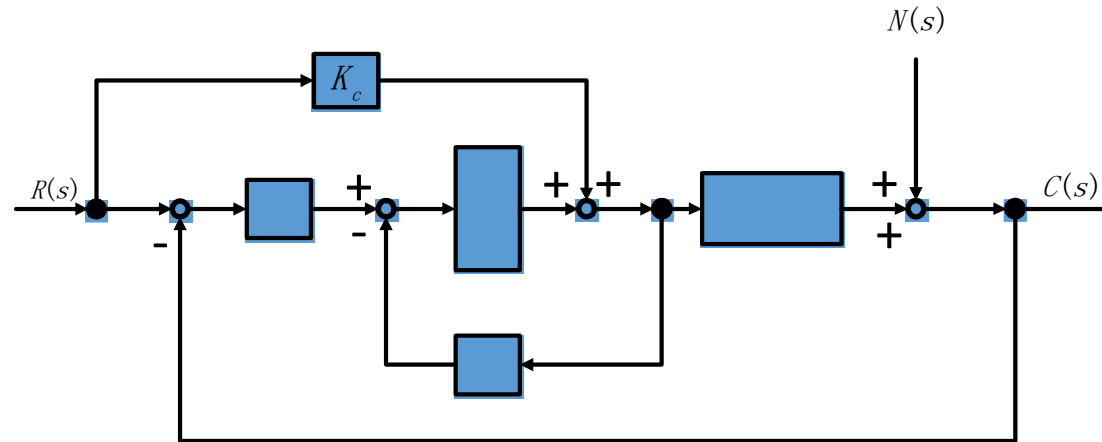
$$E_n(s) = -C(s) = -\frac{s(s + K_2 K_3) (Ts + 1)}{s(s + K_2 K_3) (Ts + 1) + K_1 K_2 K_4} N(s)$$

$$e_{ssn}(\infty) = \lim_{s \rightarrow 0} s E_n(s) = -\lim_{s \rightarrow 0} s \frac{s(s + K_2 K_3) (Ts + 1)}{s(s + K_2 K_3) (Ts + 1) + K_1 K_2 K_4} \cdot \frac{1}{s^2}$$

$$e_{ssn}(\infty) = -\frac{K_3}{K_1 K_4}$$

习题课

习题三



2、 $r(t) = t$

两个单独回路

$$L_1 = -\frac{K_2 K_3}{s}, L_2 = -\frac{K_1 K_2 K_4}{s^2(Ts + 1)}, \Delta = 1 - L_1 - L_2$$

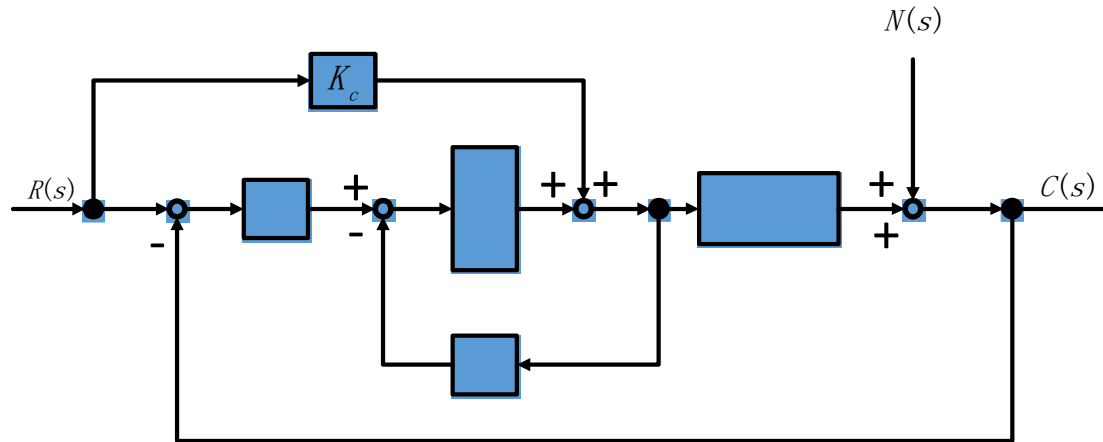
两条前向通道

$$p_1 = \frac{K_1 K_2 K_4}{s^2(Ts + 1)}, \Delta_1 = 1 \quad p_2 = \frac{K_c K_4}{s(Ts + 1)}, \Delta_2 = 1$$

$$\Phi(s) = \frac{C(s)}{R(s)} = \frac{\sum p_i \Delta_i}{\Delta} = \frac{K_c K_4 s + K_1 K_2 K_4}{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) + K_1 K_2 K_4}$$

习题课

习题三



$$\Phi(s) = \frac{C(s)}{R(s)} = \frac{\sum p_i \Delta_i}{\Delta} = \frac{K_c K_4 s + K_1 K_2 K_4}{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) + K_1 K_2 K_4}$$

$$E_r(s) = R(s) - C(s) = [1 - \Phi(s)] \bullet R(s)$$

$$= \frac{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) - K_c K_4 s}{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) + K_1 K_2 K_4} \bullet R(s)$$

$$e_{ssr}(\infty) = \lim_{s \rightarrow 0} s E_r(s) = \lim_{s \rightarrow 0} s \cdot \frac{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) - K_c K_4 s}{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) + K_1 K_2 K_4} \cdot \frac{1}{s^2}$$

$$= \lim_{s \rightarrow 0} \frac{s(Ts + 1) + K_2 K_3 Ts + (K_2 K_3 - K_c K_4)}{s^2(Ts + 1) + K_2 K_3 s(Ts + 1) + K_1 K_2 K_4} = 0$$



$$K_2 K_3 - K_c K_4 = 0$$

$$K_c = \frac{K_2 K_3}{K_4}$$

习题课

习题四

设单位反馈系统的开环传递函数如下：

$$G(s) = \frac{K^*(s + 2)}{(s + 6)(s^2 + 2s + s)}$$

试绘制K在正负反馈时的根轨迹图，并确定系统无超调时K的范围

习题课

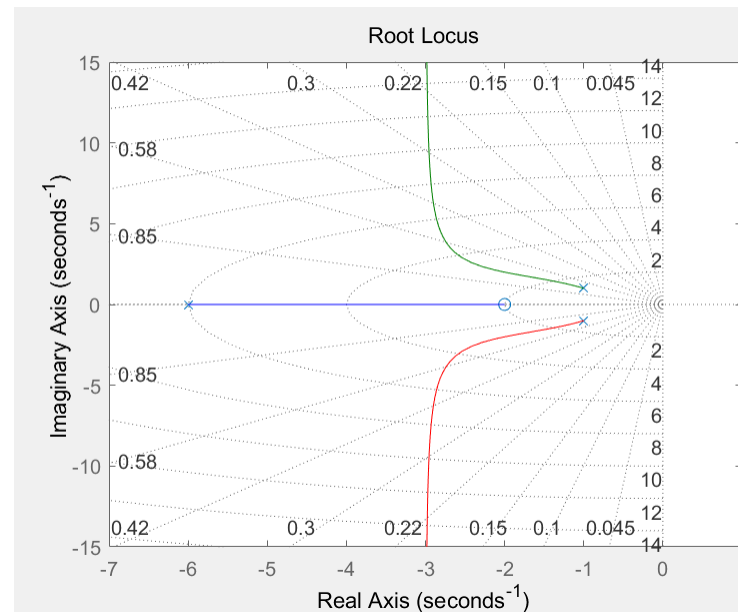
习题四

$$G(s) = \frac{K^*(s+2)}{(s+6)(s^2+2s+2)} = \frac{K^*(s+2)}{(s+6)(s+1-j)(s+1+j)}$$

负反馈: $K : 0 \rightarrow +\infty$ 实轴上的分布: $[-6, -2]$

根轨迹的渐近线: $\sigma_\alpha = \frac{-1+j-1-j-6+2}{3-1} = -3, \varphi_\alpha = \pm \frac{\pi}{2}$

根轨迹的起始角: $\theta_{p1} = 180^\circ + \varphi_{z1p1} - \theta_{p2p1} - \theta_{p3p1} = 180^\circ + 45^\circ - 90^\circ - \arctan \frac{1}{5} = 123.69^\circ$
 $\theta_{p2} = -123.69^\circ$



习题课

习题四

$$G(s) = \frac{K^*(s+2)}{(s+6)(s^2+2s+2)} = \frac{K^*(s+2)}{(s+6)(s+1-j)(s+1+j)}$$

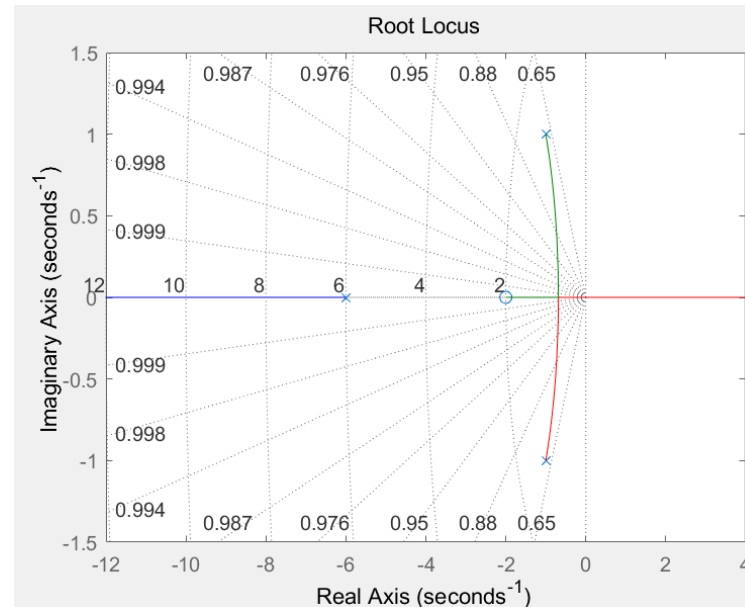
正反馈: $K : -\infty \rightarrow 0$

实轴上的分布: $[-6, -\infty), [-2, +\infty)$

根轨迹的分离点: $\frac{1}{d+6} + \frac{1}{d+1-j} + \frac{1}{d+1+j} = \frac{1}{d+2} \quad d = -0.685$

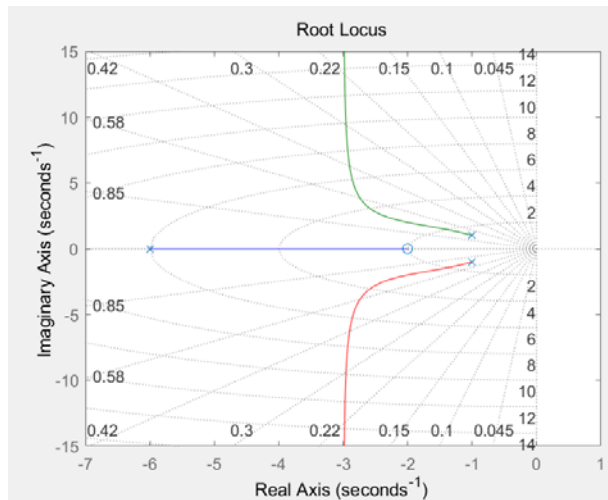
根轨迹的起始角: $\theta_{p1} = \varphi_{z1p1} - \theta_{p2p1} - \theta_{p3p1} = 45^\circ - 90^\circ - \arctan \frac{1}{5} = -56.31^\circ$

$$\theta_{p2} = 56.31^\circ$$

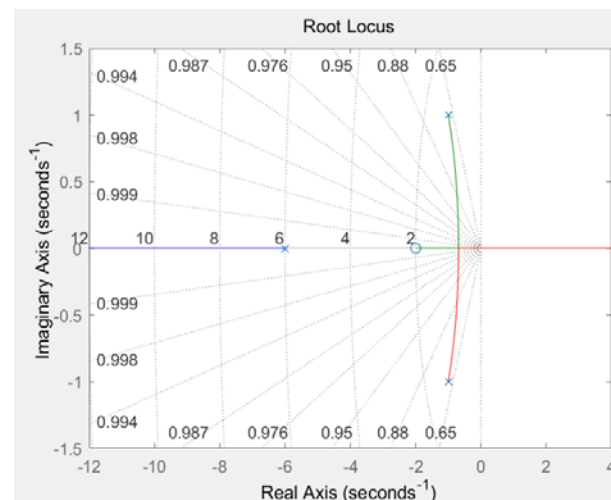


习题课

习题四
$$G(s) = \frac{K^*(s + 2)}{(s + 6)(s^2 + 2s + 2)} = \frac{K^*(s + 2)}{(s + 6)(s + 1 - j)(s + 1 + j)}$$



×



√

使系统无超调，闭环极点须位于实轴之上

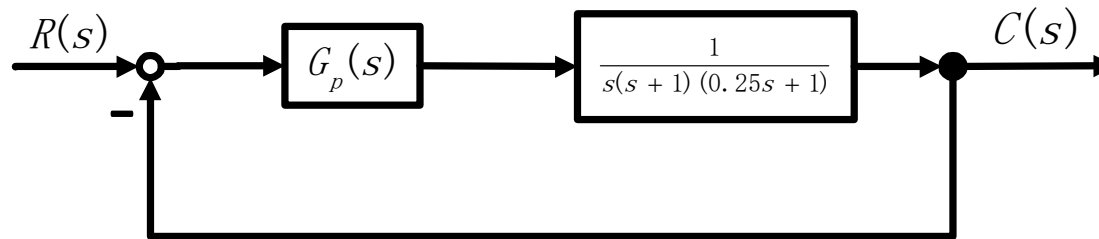
系统闭环特征方程 $D(s) = (s + 6)(s^2 + 2s + 2) + K^*(s + 2) = s^3 + 8s^2 + 14s + 12 + K^*(s + 2) = 0$

带入分离点 $d = -0.685$

$$K^* = -4.443 \quad K^* \leq -4.443$$

习题课

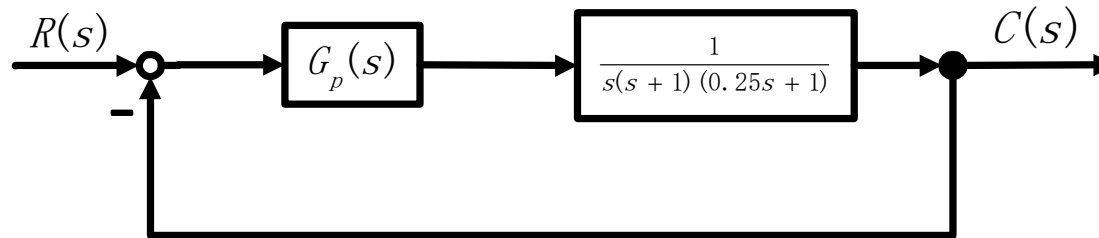
习题五



- 1、当 $G_p(s) = K_p$ 时，绘制以 K_p 为参数变量的根轨迹
- 2、为使系统的阶跃响应呈现衰减振荡形式，试确定 K_p 范围
- 3、当 $G_p(s) = K_p(1+0.5s)$ 时，试绘制以 K_p 为参数变量的根轨迹

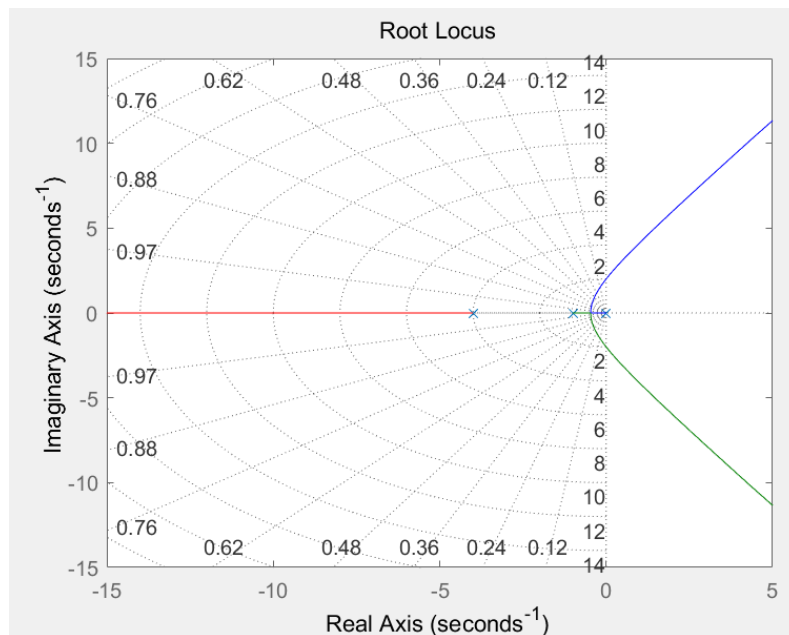
习题课

习题五



1、当 $G_p(s) = K_p$ 时，绘制以 K_p 为参数变量的根轨迹

$$G(s) = \frac{K_p}{s(s+1)(0.25s+1)} = \frac{4K_p}{s(s+1)(s+4)}$$



渐近线 $\sigma_\alpha = -\frac{5}{3}, \varphi_\alpha = \pm 60^\circ, 180^\circ$

分离点 $\frac{1}{d} + \frac{1}{d+1} + \frac{1}{d+4} = 0 \quad d = -0.465$

$$4K_p = 0.465 \times 0.535 \times 3.535 = 0.879$$

$K_p = 0.22$ 分离点处根轨迹增益

$$s^3 + 5s^2 + 4s + 4K_p = 0$$

$K_p = 5$ 全零行

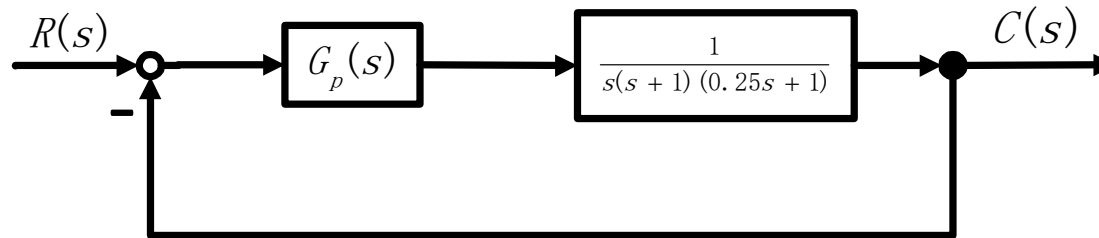
$$5s^2 + 4K_p = 5s^2 + 20 = 0$$

$$\omega = \pm 2$$

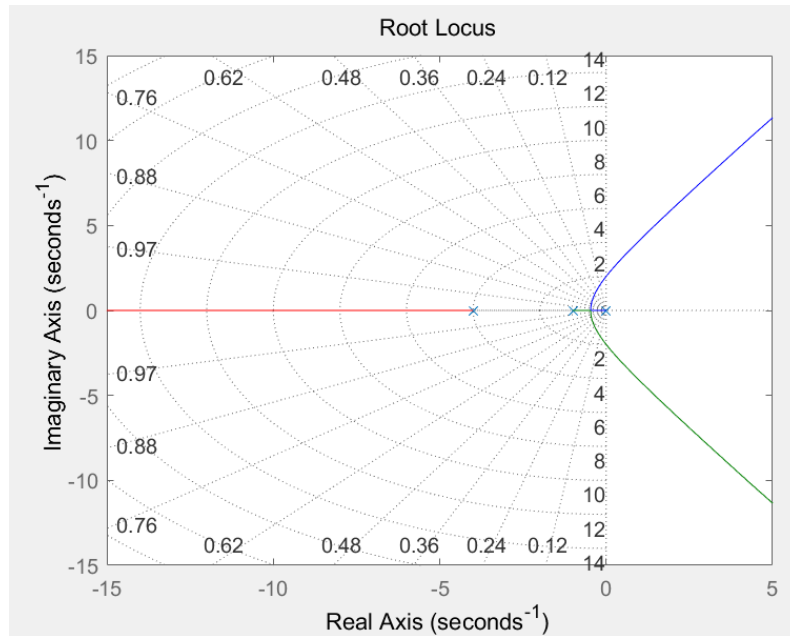
s^3	1	4
s^2	5	$4K_p$
s^1	$20 - 4K_p$	0
s^0	5	$4K_p$

习题课

习题五



2、为使系统的阶跃响应呈现衰减振荡形式，试确定 K_p 范围



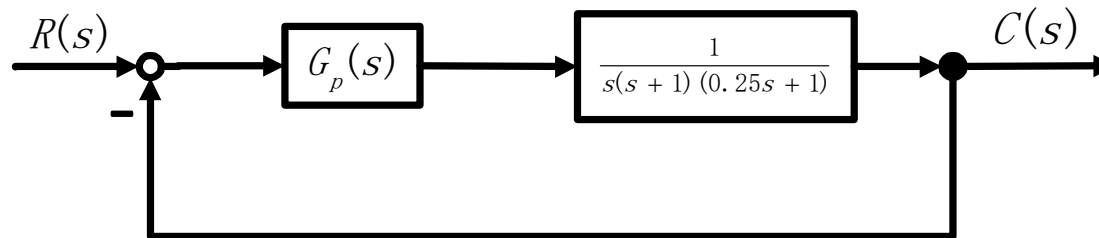
$K_p = 0.22$ 分离点处根轨迹增益

$K_p = 5$ 全零行

$$0.22 < K_p < 5$$

习题课

习题五



3、当 $G_p(s) = K_p(1+0.5s)$ 时，试绘制以 K_p 为参数变量的根轨迹

$$G_p(s) = K_p(1 + 0.5s) \quad G(s) = \frac{K_p(1 + 0.5s)}{s(s+1)(0.25s+1)} = \frac{2K_p(s+2)}{s(s+1)(s+4)}$$

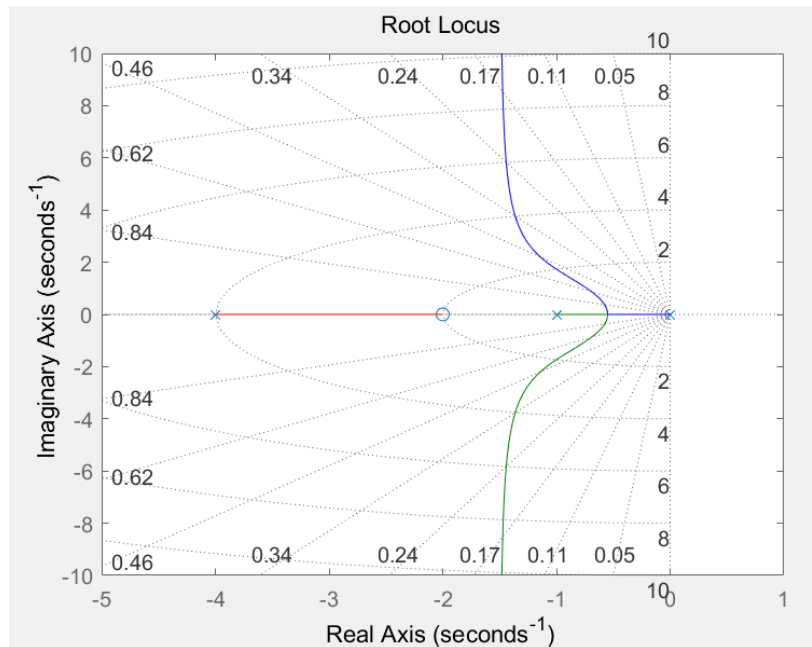
渐近线

$$\sigma_a = \frac{-5+2}{3-1} = -1.5, \varphi_a = \pm 90^\circ$$

分离点

$$\frac{1}{d} + \frac{1}{d+1} + \frac{1}{d+4} = \frac{1}{d+2}$$

$$d = -0.55$$



习题课

习题六

设单位反馈系统开环传递函数如下，试用奈奎斯特判据判断系统稳定性

$$(1) \quad G(s) = \frac{250(s+1)}{s^2(s+5)(s+15)}$$

$$(2) \quad G(s) = \frac{(s+1)^2}{s^2(3s+1)(0.1s+1)^2}$$

习题课

习题六

$$(1) \quad G(s) = \frac{250(s+1)}{s^2(s+5)(s+15)}$$

$$\begin{aligned} G(j\omega) &= \frac{250(j\omega+1)}{-\omega^2(j\omega+5)(j\omega+15)} \\ &= -\frac{250(75+19\omega^2)}{\omega^2(25+\omega^2)(225+\omega^2)} - j\frac{250(55-\omega^2)}{\omega(25+\omega^2)(225+\omega^2)} \end{aligned}$$

$$G(j0+) = -\infty - j\infty \quad G(j\infty) = 0$$

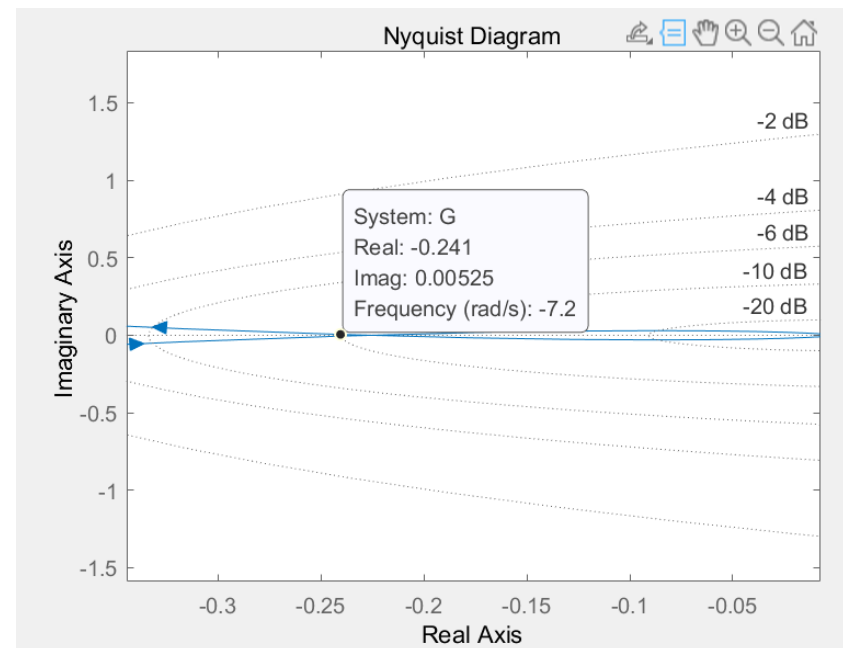
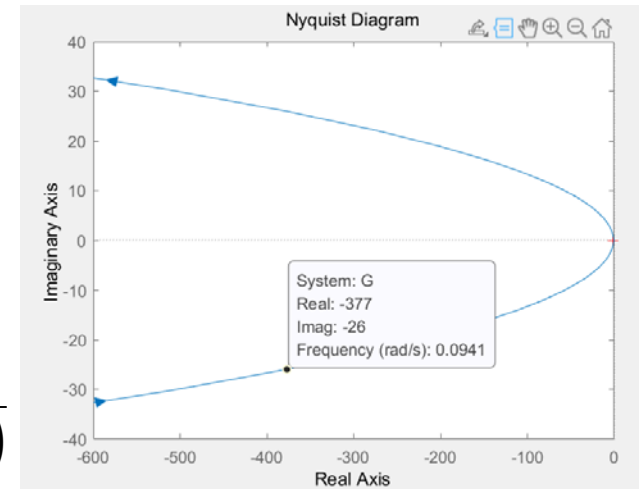
与实轴交点 $\text{Im}[G(j\omega)] = 0$

$$\omega_x = 7.42, G(j\omega_x) = \text{Re}[G(j\omega_x)] = -0.23$$

$$\nu = 2 \quad N_- = 0, N_+ = 0$$

$$N = N_+ - N_- = 0 \quad Z = P - 2N = 0$$

系统稳定



习题课

习题六

$$(2) \quad G(s) = \frac{(s+1)^2}{s^2(3s+1)(0.1s+1)^2}$$

$$\begin{aligned} G(j\omega) &= \frac{(j\omega+1)^2}{-\omega^2(j3\omega+1)(j0.1\omega+1)^2} \\ &= -\frac{1+4.79\omega^2+0.55\omega^4}{\omega^2(1+9\omega^2)(1+0.01\omega^2)^2} + j\frac{0.03\omega^4-2.01\omega^2+1.2}{\omega(1+9\omega^2)(1+0.01\omega^2)^2} \end{aligned}$$

$$G(j0+) = -\infty + j\infty \quad G(j\infty) = 0$$

与实轴交点 $\text{Im}[G(j\omega)] = 0$

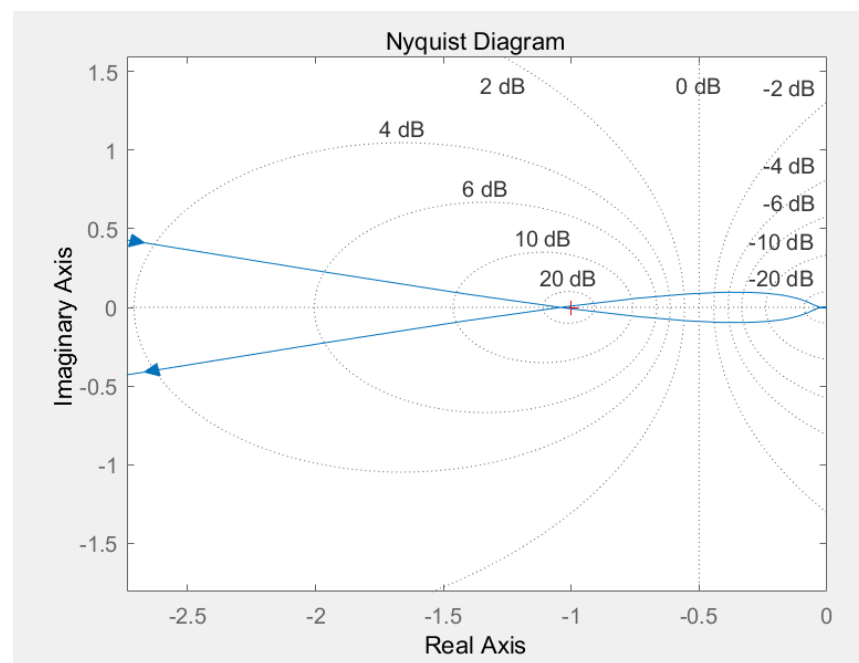
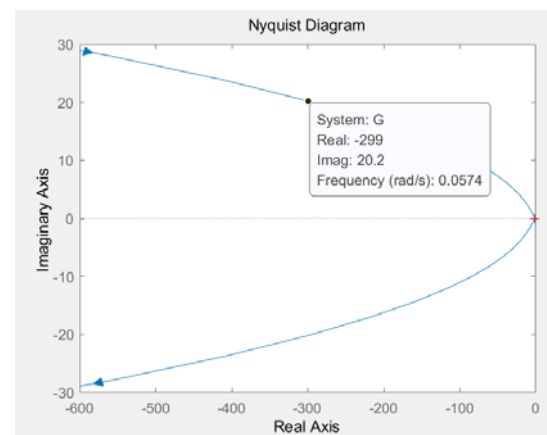
$$\omega_{x2} = 8.149, G(j\omega_{x2}) = \text{Re}[G(j\omega_{x2})] = -0.025$$

$$\omega_{x1} = 0.776, G(j\omega_{x1}) = \text{Re}[G(j\omega_{x1})] = -1.04$$

曲线在II、III象限变化

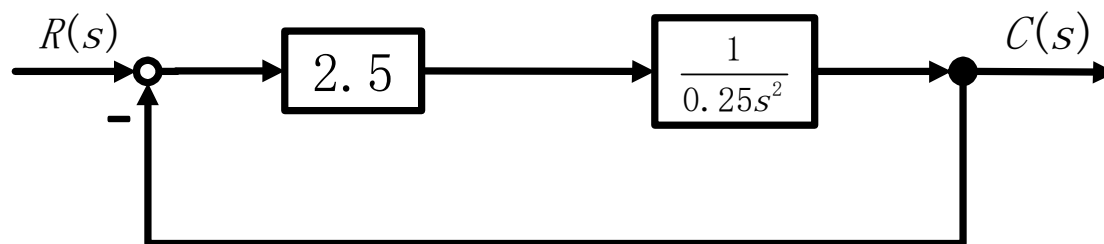
$$N_- = 1, N_+ = 1 \quad N = N_+ - N_- = 0$$

$$Z = P - 2N = 0 \quad \text{系统稳定}$$



习题课

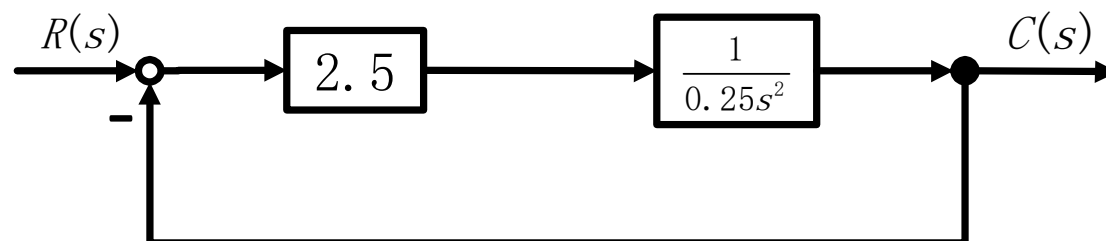
习题七



- 1、试设计串联校正网络，使系统相角裕度为50度、截止频率为6rad/s
- 2、若不采用串联校正，采用速度反馈校正，确定使系统单位阶跃响应的超调量不超过15%的反馈系数

习题课

习题七



1、试设计串联校正网络，使系统相角裕度为50度、截止频率为6rad/s

$$G_0(s) = \frac{10}{s^2} \quad \omega_c' = \sqrt{10} = 3.16, \gamma' = 0^\circ$$

试用超前校正网络

$$G_c(s) = \frac{1 + T_2 s}{1 + T_1 s}, T_2 > T_1 \quad \text{要求} \quad \omega_c = \omega_c'' = 6$$

$$\frac{0 - 11.126}{\lg \frac{1}{T_2} - \lg \omega_m} = 20$$

$$\lg T_2 = -0.22, T_2 = 0.6$$

$$G(s) = G_c(s)G_0(s) = \frac{10(1 + T_2 s)}{s^2(1 + T_1 s)}$$

$$L_0(\omega_m) = 20 \lg 10 - 20 \lg \omega_m^2 = -11.126 \text{ dB}$$

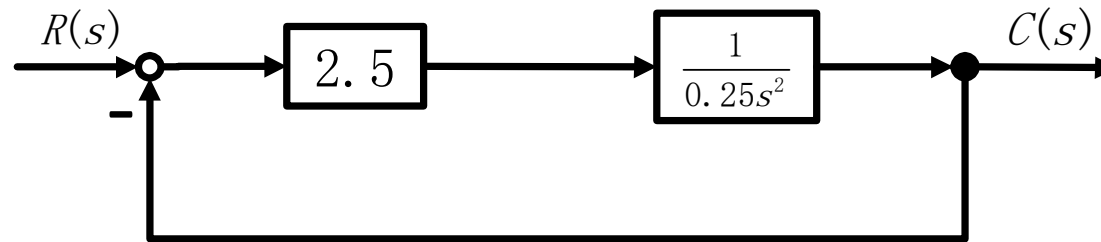
$$L_c(\omega_m) = -L_0(\omega_m) = 11.126 \text{ dB}$$

$$\gamma = 180^\circ + \angle G(j6) = \arctan 6T_2 - \arctan 6T_1$$

$$T_1 = 0.076$$

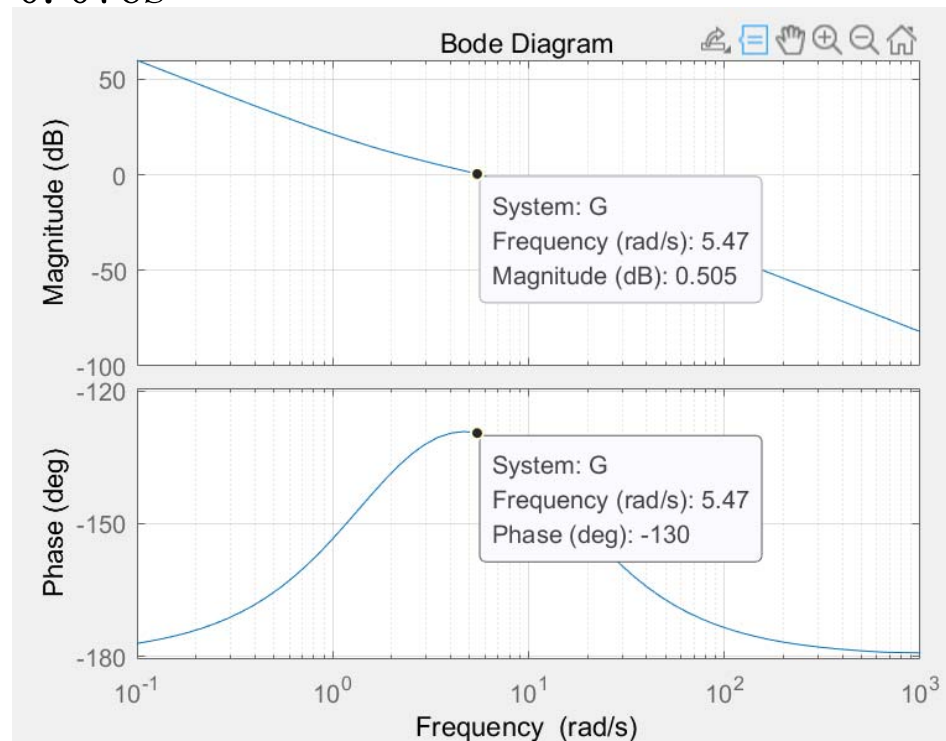
习题课

习题七



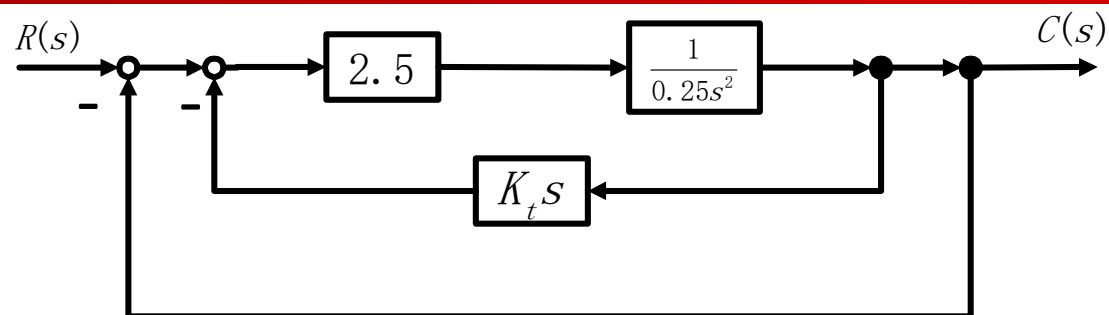
校正前后相角

$$G_c(s) = \frac{1 + 0.6s}{1 + 0.076s} \quad \varphi_0(\omega) = -180^\circ; \quad \varphi(\omega) = -180^\circ + \arctan 0.6\omega - \arctan 0.076\omega$$



习题课

习题七



2、若不采用串联校正，采用速度反馈校正，确定使系统单位阶跃响应的超调量不超过15%的反馈系数

开环
传递函数

$$G(s) = \frac{10 / s^2}{1 + 10K_t s / s^2} = \frac{10}{s(s + 10)K_t}$$

闭环
传递函数

$$\Phi(s) = \frac{10}{s^2 + 10K_t s + 10} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_n^2 = 10, 2\zeta\omega_n = 10K_t \quad \sigma\% = 100e^{-\pi\zeta/\sqrt{1-\zeta^2}}\% \leq 15\%$$

$$\zeta \geq 0.517$$

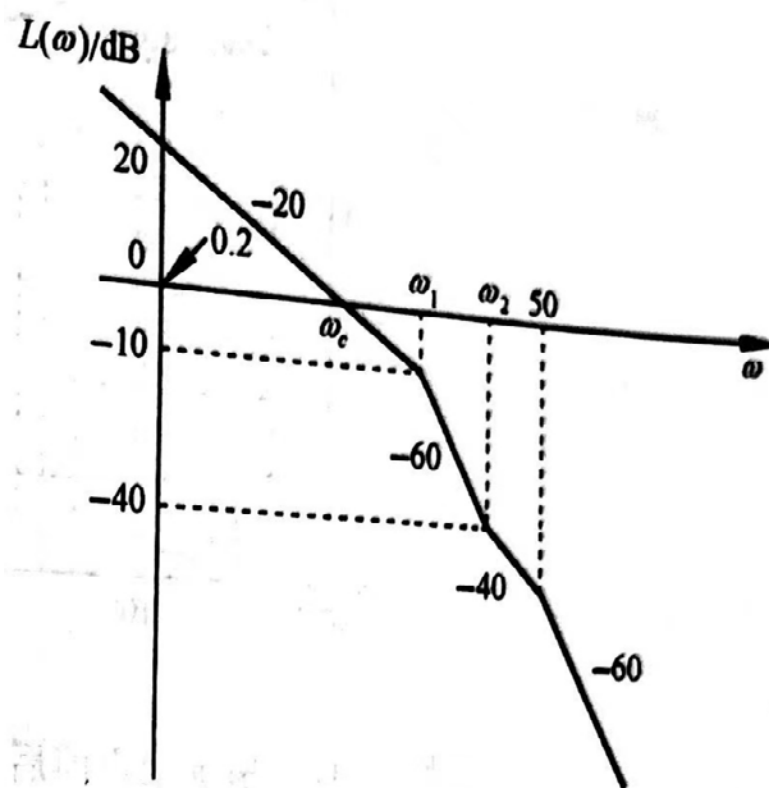
$$\text{取 } \zeta = 0.55 \quad K_t = \frac{2\zeta\omega_n}{10} = 0.348$$

$$\sigma\% = 100e^{-\pi\zeta/\sqrt{1-\zeta^2}}\% = 12.6\% < 15\%$$

习题课

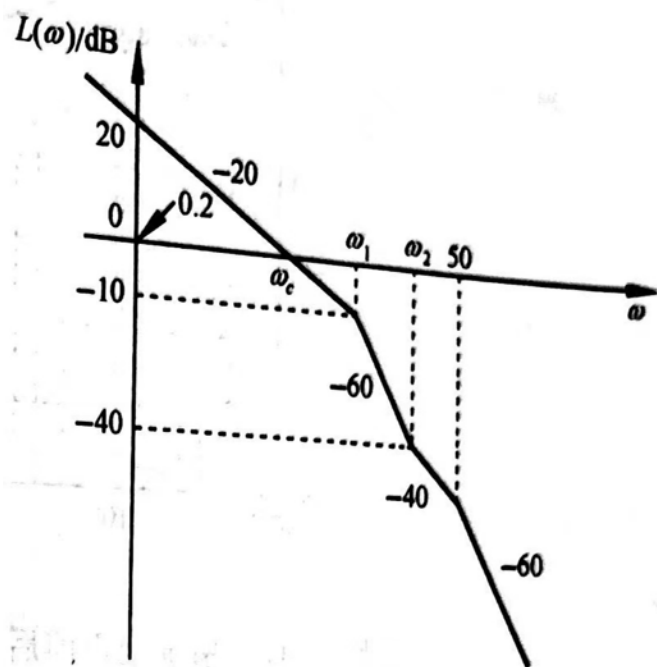
习题八

已知某最小相位系统开环对数特性曲线渐近线如图，试写出开环传递函数 $G_0(s)$ 的一种表达式



习题课

习题八



$$G_0(s) = \frac{K(\frac{1}{\omega_2} s + 1)}{s(\frac{1}{\omega_1} s + 1)^2(\frac{1}{50} s + 1)}$$

$$20 \lg \frac{K}{0.2} = 20 \Rightarrow K = 2$$

$$20 \lg \frac{\omega_c}{0.2} = 20 \Rightarrow \omega_c = 2$$

$$20 \lg \frac{\omega_1}{\omega_c} = 10 \Rightarrow \omega_1 = 6.32$$

$$60 \lg \frac{\omega_2}{\omega_1} = 30 \Rightarrow \omega_2 = 20$$



$$G_0(s) = \frac{2(\frac{1}{20} s + 1)}{s(\frac{1}{6.32} s + 1)^2(\frac{1}{50} s + 1)}$$

$$\gamma = 180^\circ + \angle G(\omega_c)$$

$$= (90^\circ + \arctan \frac{\omega_c}{20} - \arctan \frac{\omega_c}{6.32} - \arctan \frac{\omega_c}{50}) \Big|_{\omega_c=2} = 58.3^\circ$$