Principles of Robot Autonomy I

Trajectory tracking





Attendance Form

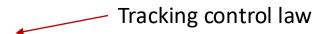


Agenda

- Trajectory tracking
 - Based on differential flatness techniques
 - Based on LQR techniques
- Readings
 - Chapter 3, sections 3.1 3.3 in D. Gammelli, J. Lorenzetti, K. Luo, G. Zardini,
 M. Pavone. *Principles of Robot Autonomy*. 2026.

Trajectory tracking

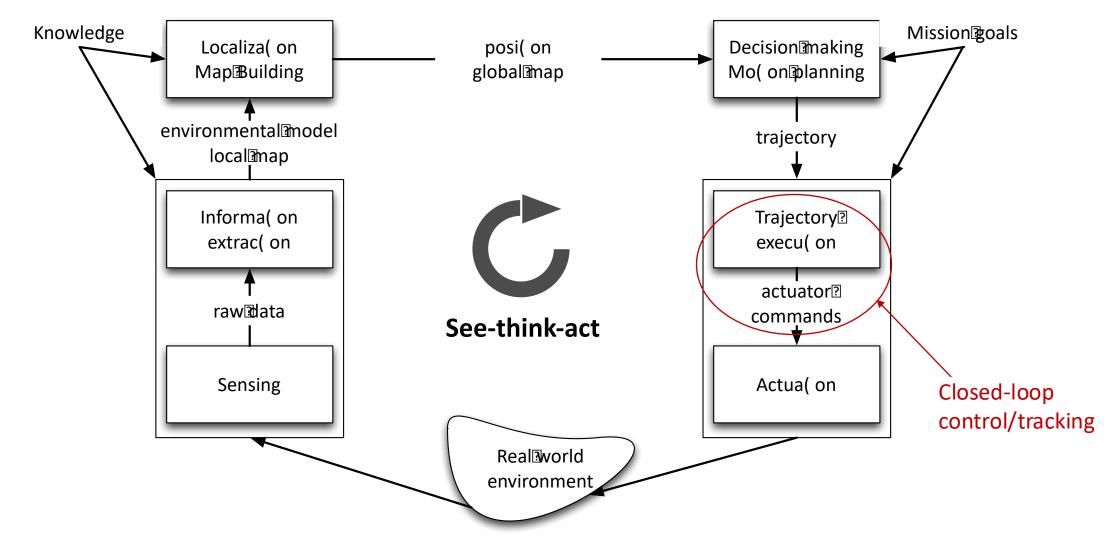
Back to two-step design strategy



$$\mathbf{u}^*(t) = \mathbf{u}_d(t) + \pi(\mathbf{x}(t), \mathbf{x}(t) - \mathbf{x}_d(t))$$

- Reference trajectory and control history (i.e., $\mathbf{x}_d(t)$ and $\mathbf{u}_d(t)$) are computed via open-loop techniques (e.g., differential flatness)
- For reference tracking (Problem 2(v-vi) in pset 1)
 - Geometric (e.g., pursuit) strategies
 - Linearization (either approximate or exact) + linear structure
 - Non-linear control
 - Optimization-based techniques (e.g., MPC)

The see-think-act cycle



Differential flatness (recap)

• A nonlinear system $\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u})$ is differentially flat if there exists a set of outputs $\mathbf{z} = \alpha(\mathbf{x}, \mathbf{u}, \dots, \mathbf{u}^{(p)})$ such that

$$\mathbf{x} = \beta(\mathbf{z}, \dot{\mathbf{z}}, \dots, \mathbf{z}^{(q)})$$

$$\mathbf{u} = \gamma(\mathbf{z}, \dot{\mathbf{z}}, \dots, \mathbf{z}^{(q)})$$

- One can then use any interpolation scheme (e.g., piecewise polynomial (spline)) to plan the trajectory of **z** in such a way as to satisfy the appropriate boundary conditions
- The evolution of the state variables **x**, together with the associated control inputs **u**, can then be computed algebraically from **z**

Trajectory tracking for differentially flat systems

Example: dynamically extended unicycle model

$$\dot{x}(t) = V \cos(\theta(t))$$

$$\dot{y}(t) = V \sin(\theta(t))$$

$$\dot{V}(t) = a(t)$$

$$\dot{\theta}(t) = \omega(t)$$

• The system is differentially flat with flat outputs (x, y), in particular

$$\begin{bmatrix} \ddot{x}(t) \\ \ddot{y}(t) \end{bmatrix} = \underbrace{\begin{bmatrix} \cos(\theta) & -V\sin(\theta) \\ \sin(\theta) & V\cos(\theta) \end{bmatrix}}_{:=J} \begin{bmatrix} a \\ \omega \end{bmatrix} := \begin{bmatrix} w_1 \\ w_2 \end{bmatrix}$$

Trajectory tracking for differentially flat systems

 Then one can use the following virtual control law for trajectory tracking:

$$w_1 = \ddot{x}_d + k_{px}(x_d - x) + k_{dx}(\dot{x}_d - \dot{x})$$

$$w_2 = \ddot{y}_d + k_{py}(y_d - y) + k_{dy}(\dot{y}_d - \dot{y})$$

where k_{px} , k_{dx} , k_{py} , $k_{dy} > 0$ are control gains

 Such a law guarantees exponential convergence to zero of the Cartesian tracking error

Trajectory tracking for differentially flat systems

- More broadly, suppose system is differentially flat: the full state and control trajectories can be computed from flat outputs $(\mathbf{z}, \dot{\mathbf{z}}, \dots, \mathbf{z}^{(q)})$
- Define: $\mathbf{z}^{(q+1)} = \mathbf{w}$
- One can then design a tracking controller by using linear control techniques; in particular, for a given reference flat output \mathbf{z}_d , define the *component-wise* error

$$e_i := z_i - z_{i,d}$$
, which implies $e_i^{(q+1)} = w_i - w_{i,d}$

• For guaranteed convergence to zero of tracking error, one can set

$$w_i = w_{i,d} - \sum_{j=0}^{q} k_{i,j} e_i^{(j)},$$

with the gains $\{k_{i,j}\}$ chosen so as to enforce stability

LQR-based methods

The previous approach only works for differentially flat systems

- How can we control more general classes of systems?
 - Nonlinear control techniques
 - Linear-quadratic regulation (LQR) methods

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Linear-quadratic regulator (LQR)

- How can we regulate (i.e., drive to the origin) the linear system $\dot{\mathbf{x}}(t) = A\mathbf{x}(t) + B\mathbf{u}(t)$ with minimum control effort?
- We define the optimal control problem

$$\begin{aligned} & \min_{\mathbf{u}} \quad \mathbf{x}(T)' F \mathbf{x}(T) + \int_{0}^{T} (\mathbf{x}(t)' Q \mathbf{x}(t) + \mathbf{u}(t)' R \mathbf{u}(t)) dt \\ & \text{s.t.} \quad \dot{\mathbf{x}}(t) = A \mathbf{x}(t) + B \mathbf{u}(t) \end{aligned}$$

• The optimal solution is of the form $\mathbf{u}_t = K_t \mathbf{x}_t$ where $K_t = -R^{-1}B'P_t$ and the matrix P_t solves the continuous time Riccati diff. equation:

$$\dot{P}_t = -A'P_t - P_tA + P_tBR^{-1}B'P_t - Q \text{ with } P_T = F$$

• Note: this results holds even in the more general case $\dot{\mathbf{x}}(t) = A(t)\mathbf{x}(t) + B(t)\mathbf{u}(t)$ (just plug A(t) and B(t) in the Riccati equation)

Tracking LQR — linear case

- Consider the linear system $\dot{\mathbf{x}}(t) = A\mathbf{x}(t) + B\mathbf{u}(t)$ and assume we would like to track a reference trajectory $\left(\mathbf{x}_d(t), \mathbf{u}_d(t)\right)_t$
- Define error variables $\delta \mathbf{x}(t) \coloneqq (\mathbf{x}(t) \mathbf{x}_d(t))$ and $\delta \mathbf{u}(t) \coloneqq (\mathbf{u}(t) \mathbf{u}_d(t))$, which leads to the dynamical system $\delta \dot{\mathbf{x}}(t) = A\delta \mathbf{x}(t) + B\delta \mathbf{u}(t)$
- Define optimal control problem

$$\min_{\mathbf{u}} \quad \delta \mathbf{x}(T)' F \, \delta \mathbf{x}(T) + \int_{0}^{T} (\delta \mathbf{x}(t)' Q \, \delta \mathbf{x}(t) + \delta \mathbf{u}(t)' R \, \delta \mathbf{u}(t)) dt$$
s.t.
$$\delta \dot{\mathbf{x}}(t) = A \delta \mathbf{x}(t) + B \delta \mathbf{u}(t)$$

• Optimal solution is $\delta \mathbf{u}_t = K_t \, \delta \mathbf{x}_t$ (same K_t as before), which leads to control $\mathbf{u}(t) = \mathbf{u}_d(t) + \delta \mathbf{u}(t)$

Tracking LQR — nonlinear case

- Consider the non-linear system $\dot{\mathbf{x}}(t) = \mathbf{f}(\mathbf{x}(t),\mathbf{u}(t))$ and assume we would like to track a reference trajectory $\left(\mathbf{x}_d(t),\mathbf{u}_d(t)\right)_t$
- Key idea: make the system "linear" by linearizing around $(\mathbf{x}_d(t), \mathbf{u}_d(t))_t$:

$$\dot{\mathbf{x}}(t) \approx \mathbf{f} \big(\mathbf{x}_d(t), \mathbf{u}_d(t) \big) + \frac{\partial \mathbf{f}}{\partial \mathbf{x}} \big(\mathbf{x}_d(t), \mathbf{u}_d(t) \big) \big(\mathbf{x}(t) - \mathbf{x}_d(t) \big) + \frac{\partial \mathbf{f}}{\partial \mathbf{u}} \big(\mathbf{x}_d(t), \mathbf{u}_d(t) \big) \big(\mathbf{u}(t) - \mathbf{u}_d(t) \big)$$

$$\dot{\mathbf{x}}(t) + A(t) \big(\mathbf{x}(t) - \mathbf{x}_d(t) \big) + B(t) \big(\mathbf{u}(t) - \mathbf{u}_d(t) \big)$$

• As before, we get a linear system in the error variables:

$$\delta \dot{\mathbf{x}}(t) = A(t)\delta \mathbf{x}(t) + B(t)\delta \mathbf{u}(t)$$

• Optimal solution is $\delta \mathbf{u}_t = K_t \, \delta \mathbf{x}_t$ which leads to control $\mathbf{u}(t) = \mathbf{u}_d(t) + \delta \mathbf{u}(t)$ (where K_t is again obtained from the Riccati eq.)

Transferring results to discrete case

- Same ideas apply for discrete-time systems, i.e., $\mathbf{x}_{k+1} = \mathbf{f}(\mathbf{x}_k, \mathbf{u}_k)$
- Key difference: control is $\mathbf{u}_k = \mathbf{u}_k^d + \delta \mathbf{u}_k$, with $\delta \mathbf{u}_k = K_k \delta \mathbf{x}_k$, where $K_k = -(R + B_k' P_{k+1} B_k)^{-1} B_k' P_{k+1} A_k$ and P_k is iteratively obtained from the discrete-time Riccati equation:

$$P_k = A'_k P_{k+1} A_k - A'_k P_{k+1} B_k (R + B'_k P_{k+1} B_k)^{-1} B'_k P_{k+1} A_k + Q$$
 with $P_N = F$

- Similarly as before:
 - $A_k = \frac{\partial \mathbf{f}}{\partial \mathbf{x}} (\mathbf{x}_k^d, \mathbf{u}_k^d)$ and $B_k = \frac{\partial \mathbf{f}}{\partial \mathbf{u}} (\mathbf{x}_k^d, \mathbf{u}_k^d)$
 - *R*, *Q*, *F* have the same interpretation as before, namely tracking penalty, control effort penalty, and final tracking error penalty

Next time

