Learning Control-Oriented Dynamical Structure from Data

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Abstract

Even for known nonlinear dynamical systems, feedback controller synthesis is a difficult problem that often requires leveraging the particular structure of the dynamics to induce a stable closed-loop system. For general nonlinear models, including those fit to data, there may not be enough known structure to reliably synthesize a stabilizing feedback controller. In this paper, we propose a novel state-dependent nonlinear tracking controller formulation based on a statedependent Riccati equation for general nonlinear control-affine systems. Our formulation depends on a nonlinear factorization of the system of vector fields defining the control-affine dynamics, which we show always exists under mild smoothness assumptions. We discuss how this factorization can be learned from a finite set of data. On a variety of simulated nonlinear dynamical systems, we empirically demonstrate the efficacy of learned versions of our controller in stable trajectory tracking. Alongside our method, we evaluate recent ideas in jointly learning a controller and stabilizability certificate for known dynamical systems; we show experimentally that such methods can be frail in comparison.

1. Introduction

Data-driven system identification and control algorithms are imperative to the operation of autonomous systems in complex environments. In particular, model-based algorithms equip an autonomous agent with the ability to learn how it and the system it is part of evolve over time. However, for general nonlinear systems including those learned from data, it is not always clear how to synthesize a *stabiliz*-

ing tracking controller. Effective control design often leverages specific system structure; some classic examples of this are the linear quadratic regulator (LQR), and the computed torque method and its variants for Lagrangian dynamical systems (Murray et al., 1994; Slotine & Li, 1987). A central goal of control-oriented learning (Richards et al., 2021; 2022) and this paper is to jointly learn a dynamics model and additional control-oriented structure that naturally encodes or reveals a stabilizing controller design.

Related Work An approach favoured by recent works has been to learn stabilizing controllers for nonlinear system models by simultaneously learning a parametric controller and a parametric control-theoretic certificate, such as a control Lyapunov function (CLF) or control contraction metric (CCM). This paradigm originates in works that learn stability certificates for nonlinear systems of the form $\dot{x} = f(x)$ or $x_{t+1} = f(x_t)$. Convergence of the state to x = 0 is guaranteed if a Lyapunov certificate function V can be found such that $\nabla V(x)^{\mathsf{T}} f(x) < 0$ or V(f(x)) - V(x) < 0, respectively, for each $x \neq 0$. Such functional inequalities serve as the cornerstone for methods that learn parametric certificates from data either via gradient descent on a loss function comprising sampled point violations (Richards et al., 2018; Boffi et al., 2020), or formal synthesis and verification (Abate et al., 2021). Similar functional inequalities appear in contraction theory (Lohmiller & Slotine, 1998) to describe the convergence of system trajectories to each other over time, and have been used in imitation learning to regularize fitted dynamics models towards stability (Sindhwani et al., 2018) or intrinsic stabilizability (Singh et al., 2021).

For *controlled* nonlinear systems like $\dot{x} = f(x) + B(x)u$, one can try jointly learning a parametric CLF V and parametric controller u = k(x) by penalizing violations of the inequality $\nabla V(x)^{\mathsf{T}}(f(x) + B(x)k(x)) < 0$ at sampled states. This concept underlies most prior work on learning certified stabilizing nonlinear controllers (Chang et al., 2019; Chang & Gao, 2021; Dawson et al., 2021; 2022). For tracking a trajectory $(\bar{x}(t), \bar{u}(t))$, Sun et al. (2020) jointly learn a CCM and a feedback controller $u = \pi(x, \bar{x}, \bar{u})$, again based on sampled inequality violations. Such approaches aspire to the closed-loop stability promised by satisfaction of this infinite dimensional constraint, yet it is

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unclear whether penalizing violations at a finite number of points is sufficient to achieve this in practice.

Rather than trying to fit a controller and certificate to data, one can leverage structure in the dynamics to inform stabilizing controller design. Lagrangian systems of the form $H(q)\ddot{q} + C(q,\dot{q})\dot{q} + q(q) = u$ with state $x := (q,\dot{q})$ are amenable to feedback linearization (Slotine & Li, 1991) by virtue of their double-integrator form, even when learned from data (Gupta et al., 2020; Richards et al., 2021; Djeumou et al., 2022). Hamiltonian structure as a physicsbased prior in learned models can be exploited to synthesize passivity-based controllers (Zhong et al., 2020; Li et al., 2022). Perhaps the most fundamental example of structure informing control is LQR, which for a linear system $\dot{x} = Ax + Bu$ describes an optimal stabilizing controller computable from a Riccati equation in terms of the system matrices (A, B) and chosen cost matrices (Q, R). Each of these designs is tailored to a subset of controlaffine dynamical systems, yet LQR can be extended to general control-affine systems of the form $\dot{x} = f(x) + B(x)u$ with the state-dependent coefficient (SDC) factorization f(x) = A(x)x, which always exists as long as f is differentiable and f(0) = 0 (Cimen, 2010). A feedback controller can then be implemented by solving the corresponding state-dependent Riccati equation (SDRE) in terms of (A(x), B(x)) in closed-loop. While such a controller is only locally stabilizing in theory, in practice it has a large region of attraction and has proven effective in automotive (Acarman, 2009), spacecraft (Cloutier & Zipfel, 1999). robotic (Watanabe et al., 2008), and process control (Banks et al., 2002).

Contributions In this work, we study how to jointly identify nonlinear dynamics models and control-oriented structures from data that can be naturally leveraged in stabilizing closed-loop tracking control design. To this end, we propose a novel tracking controller for general nonlinear control-affine systems based on SDRE feedback. While SDREs have seen use in fixed-point stabilization, our design is novel in its exact characterization and control of error dynamics for trajectory tracking. Our design relies on a generalized SDC factorization of the error dynamics that we show always exists for differentiable dynamics. We then propose a method to learn such structure from a finite data set, and thereby enable the use of our SDRE-based tracking controller. We compare our method of learning control-enabling structure to an adaptation of prior work that tries to jointly learn a dynamics model, controller, and stability certificate. In a variety of simulated nonlinear systems, we demonstrate that our learned controller performs well in closed-loop, and that controllers instead learned alongside dynamics models and parametric certificate functions can be brittle and data inefficient in practice.

2. Problem Statement

In this paper, we are interested in learning to control the nonlinear control-affine dynamical system

$$\dot{x} = f(x) + B(x)u = f(x) + \sum_{j=1}^{m} u_j b_j(x), \quad (1)$$

with state $x(t) \in \mathbb{R}^n$, control $u(t) \in \mathbb{R}^m$, drift $f: \mathbb{R}^n \to \mathbb{R}^n$, and actuator $B: \mathbb{R}^n \to \mathbb{R}^{n \times m}$ with columns $b_j: \mathbb{R}^n \to \mathbb{R}^n$, $j \in \{1, 2, \dots, m\}$. In particular, we want to determine a tracking controller of the form $u = \pi(x, \bar{x}(t), \bar{u}(t))$ such that (x(t), u(t)) converges to any dynamically feasible pair $(\bar{x}(t), \bar{u}(t))$, i.e., satisfying $\dot{\bar{x}} = f(\bar{x}) + B(\bar{x})\bar{u}$. While we know the dynamics take the form of Equation (1), the vector fields $(f, \{b_j\}_{j=1}^m)$ are otherwise unknown to us. Instead, we only have access to a finite pre-collected data set $\mathcal{D} := \{(x^{(i)}, u^{(i)}, \dot{x}^{(i)})\}_{i=1}^N$ of input-output measurements of Equation (1).

3. Nonlinear Tracking Control

In this section, we overview a number of methods for synthesizing a tracking controller $u=\pi(x,\bar{x}(t),\bar{u}(t))$ for any control-affine nonlinear system of the form in Equation (1). We begin with LQR-based methods, including our novel state-dependent-LQR tracking controller. We also discuss tracking controllers that are guaranteed to exponentially stabilize the resulting closed-loop dynamics provided an accompanying certificate function is found, namely either a control Lyapunov function (CLF) or a control contraction metric (CCM). For each controller, we identify any structure that is required in addition to the dynamics to compute the feedback signal; we term this *control-oriented* structure. We will then discuss how to jointly learn such structure along with a dynamics model from data in Section 4 to enable closed-loop tracking control.

3.1. Linearization-based LQR

Perhaps the simplest approach to tracking control is based on linearizing the dynamics in Equation (1) around the current target $(\bar{x}(t), \bar{u}(t))$. Specifically, in this method we first linearize the nonlinear dynamics of the tracking error $e(t) \coloneqq x(t) - \bar{x}(t)$ given by

$$\dot{e} = f(x) + B(x)u - f(\bar{x}) - B(\bar{x})\bar{u} \tag{2}$$

to get the approximation

$$\dot{e} \approx \underbrace{\left(\frac{\partial f}{\partial x}(\bar{x}) + \sum_{j=1}^{m} \bar{u}_{j} \frac{\partial b_{j}}{\partial x}(\bar{x})\right)}_{=:A(\bar{x},\bar{u})} e + B(\bar{x})(u - \bar{u}). \quad (3)$$

Then, with $(A(\bar{x}, \bar{u}), B(\bar{x}))$ and chosen positive-definite weight matrices (Q, R), we solve the Riccati equation

$$P(\bar{x}, \bar{u})A(\bar{x}, \bar{u}) + A(\bar{x}, \bar{u})^{\mathsf{T}}P(\bar{x}, \bar{u})$$
$$-P(\bar{x}, \bar{u})B(\bar{x})R^{-1}B(\bar{x})^{\mathsf{T}}P(\bar{x}, \bar{u}) = -Q$$
(4)

for the positive-definite solution $P(\bar{x}, \bar{u})$. We then compute the tracking controller

$$u = \pi_{\text{LQR}}(x, \bar{x}, \bar{u}) \coloneqq \bar{u} - R^{-1}B(\bar{x})^{\mathsf{T}}P(\bar{x}, \bar{u})e. \tag{5}$$

In practice, the linearization-based LQR tracking controller in Equation (5) can be effective as long as (x(t), u(t)) remains close to $(\bar{x}(t), \bar{u}(t))$, i.e., as long as the linearized error dynamics in Equation (3) remain a good approximation of original error dynamics in Equation (2). Overall, the linearization-based LQR tracking controller requires us to be able to evaluate and differentiate the vector fields $(f, \{b_i\}_{i=1}^m)$; no additional structures are required.

3.2. Nonlinear State-Dependent LQR

For general nonlinear systems, the linearization-based LQR tracking controller presented in the previous section is a good first choice. However, it can fail for nonlinear systems when (x(t), u(t)) strays too far from the target $(\bar{x}(t), \bar{u}(t))$, since then the linearized dynamics in Equation (3) are no longer a good approximation.

In this section, we introduce a novel *exact* nonlinear factorization of the error dynamics for general control-affine systems that resemble the linearized form in Equation (3). This factorization is based on the theory of SDC forms (Çimen, 2010; 2012), and thereby enables a feedback law based on solving an associated SDRE.

State-Dependent LQR for Regulation To begin, we first look at the simpler problem of regulating the state x(t) of the system $\dot{x} = f(x) + B(x)u$ to x = 0. For now, we assume that (x, u) = (0, 0) is an equilibrium pair, i.e., f(0) = 0. If $f : \mathbb{R}^n \to \mathbb{R}^n$ is differentiable, Çimen (2010, Proposition 1) shows we can write the dynamics as

$$\dot{x} = f(x) + B(x)u = A(x)x + B(x)u,$$
 (6)

where $f(x) \equiv A(x)x$ is an exact factorization known as a state-dependent coefficient (SDC) form of f. With chosen positive-definite matrices (Q,R), these factorized dynamics naturally enable the controller $u=K(x)x=-R^{-1}B(x)^{\rm T}P(x)x$, where P(x) is the positive-definite solution of the state-dependent Riccati equation (SDRE)

$$P(x)A(x) + A(x)^{\mathsf{T}}P(x) - P(x)B(x)R^{-1}B(x)^{\mathsf{T}}P(x) = -Q$$
 (7)

As its name implies, the SDRE is dependent on the *current* state x of the system. This contrasts with the Riccati equation for linearized LQR in Equation (4), which

does not depend on x and only depends on the target pair (\bar{x}, \bar{u}) due to linearization. Despite using an exact nonlinear factorization of the dynamics, the feedback law $u = -R^{-1}B(x)^{\mathsf{T}}P(x)x$ is only locally stabilizing in theory. In practice, state-dependent LQR control can induce a large region of attraction in the closed-loop system, especially relative to linearization-based control (Cimen, 2012).

Generalized SDC forms One of our contributions is extending SDRE-based feedback to tracking control for control-affine systems. To this end, we introduce a generalization of SDC forms in Proposition 1 below.

Proposition 1: Suppose $f: \mathbb{R}^n \to \mathbb{R}^d$ is differentiable. Then there exists $A: \mathbb{R}^n \times \mathbb{R}^n \to \mathbb{R}^{d \times n}$ such that

$$f(x) - f(\bar{x}) = A(\bar{x}, x - \bar{x})(x - \bar{x}) = A(\bar{x}, e)e,$$
 (8)

for all $x, \bar{x} \in \mathbb{R}^n$ with $e := x - \bar{x}$. Furthermore, A can be chosen such that $A(\bar{x}, 0) \equiv \frac{\partial f}{\partial x}(\bar{x})$.

Proof. Consider any curve $r(s) = \bar{x} + R(s)e$ where $R: [0,1] \to \mathbb{R}^{n \times n}$ is differentiable, R(0) = 0, and R(1) = I. Then by the fundamental theorem for line integrals,

$$f(x) - f(\bar{x}) \equiv \underbrace{\left(\int_0^1 \frac{\partial f}{\partial x} (\bar{x} + R(s)e)R'(s) \, ds\right)}_{=:A(\bar{x},e)} e. \quad (9)$$

Moreover,
$$A(\bar{x},0) = \int_0^1 \frac{\partial f}{\partial x}(\bar{x}) R'(s) ds = \frac{\partial f}{\partial x}(\bar{x}).$$

Proposition 1 describes a factorization of differentiable f that exactly quantifies $f(x)-f(\bar{x})$ between any (x,\bar{x}) . In the case where $x=\bar{x}$, the matrix factor $A(\bar{x},e)$ can reduce to the local Jacobian of f at \bar{x} . Much like the linear approximation $\frac{\partial f}{\partial x}(\bar{x})e$, the exact factorization $A(\bar{x},e)e$ is a function of the chosen "target" \bar{x} and the "error" $e:=x-\bar{x}$. It is precisely this perspective that now allows us to apply this generalized SDC form to tracking control.

State-Dependent LQR for Trajectory Tracking To construct our novel SDRE-based tracking controller, we analyze the form of the error dynamics for general controlaffine systems. Let $(\bar{x}(t), \bar{u}(t))$ be a dynamically feasible pair that we want to track. Then the dynamics of the tracking error $e \coloneqq x - \bar{x}$ are

$$\dot{e} = f(x) + B(x)u - f(\bar{x}) - B(\bar{x})\bar{u}
= f(x) - f(\bar{x}) + (B(x) - B(\bar{x}))\bar{u} + B(x)(u - \bar{u})
= \underbrace{\left(A_0(\bar{x}, e) + \sum_{j=1}^m \bar{u}_j A_j(\bar{x}, e)\right)}_{=:A_{\text{SDC}}(\bar{x}, \bar{u}, e)} e + B(x)v$$
, (10)

where $v := u - \bar{u}$, and $(A_0, \{A_j\}_{j=1}^m)$ are SDC factorizations of the vector fields $(f, \{b_j\}_{j=1}^m)$ such that

$$f(x) - f(\bar{x}) \equiv A_0(\bar{x}, e)e b_j(x) - b_j(\bar{x}) \equiv A_j(\bar{x}, e)e, \ \forall j \in \{1, 2, \dots, m\}$$
 (11)

A state-dependent Riccati equation similar to Equation (7) expressed in terms of $(A_{\text{SDC}}(\bar{x}, \bar{u}, e), B(x))$ and chosen positive-definite weight matrices (Q, R) can be solved for the positive-definite matrix $P_{\text{SDC}}(\bar{x}, \bar{u}, e)$. The associated nonlinear tracking controller is then

$$u = \pi_{\text{SDC}}(x, \bar{x}, \bar{u}) := \bar{u} - R^{-1}B(x)^{\mathsf{T}}P_{\text{SDC}}(\bar{x}, \bar{u}, e)e. \tag{12}$$

This controller reduces to the linearization-based LQR controller in Equation (5) if e=0 is always used, even when $x \neq \bar{x}$, since then the SDC factorizations in Equation (11) reduce to the corresponding Jacobians in Equation (3).

Our goal in using state-dependent LQR tracking control is to enable better tracking performance for highly nonlinear systems that may experience large deviations from the target trajectory, e.g., during fast or aggressive maneuvers. The key trade-off in the use of a more complex controller is the need for additional known control-oriented structure. In this case, that structure comprises the SDC factorizations $(A_0, \{A_j\}_{j=1}^m)$ that are not required in the simpler linearization-based LQR tracking controller. In Section 4, we will discuss how we can learn $(A_0, \{A_j\}_{j=1}^m)$ from data, and later in Section 5 we will show how this has a powerful regularization effect on learning models of dynamical systems for the purposes of closed-loop control. Before that, in the next section we overview alternative methods that couple a tracking controller with a certificate function guaranteeing closed-loop tracking convergence.

3.3. Exponential Stabilizability via Contraction Theory

Linearization-based and state-dependent LQR rely on approximate and exact factorized forms, respectively, of the system dynamics to construct tracking control laws. However, neither of these LQR controllers is guaranteed to stabilize the closed-loop error dynamics when the system is nonlinear. In this section, we review a family of tracking controllers that ensure exponential stability, i.e.,

$$||x(t) - \bar{x}(t)||_2 \le \alpha ||x(0) - \bar{x}(0)||_2 \exp(-\beta t),$$
 (13)

with overshoot $\alpha > 0$ and decay rate $\beta > 0$, for all $t \ge 0$.

To this end, *contraction theory* (Lohmiller & Slotine, 1998) seeks to construct certifiably stabilizing controllers for any control-affine system of the form Equation (1) by analyzing the stabilizability of the variational dynamics

$$\dot{\delta}_{x} = \underbrace{\left(\frac{\partial f}{\partial x}(x) + \sum_{j=1}^{m} u_{j} \frac{\partial b_{j}}{\partial x}(x)\right)}_{=:A(x,u)} \delta_{x} + B(x)\delta_{u}, \quad (14)$$

where δ_x and δ_u are virtual displacements in the tangent spaces at x and u, respectively. The high-level idea of contraction theory is to stabilize this infinite family of linear variational systems pointwise everywhere with a variational feedback law for δ_u , then path-integrate to get a stabilizing feedback law for u in the original system (Lohmiller & Slotine, 1998; Manchester & Slotine, 2017). Let $M: \mathbb{R}^n \to \mathbb{S}^n_{\succ 0}$ be a uniformly positive-definite matrix-valued function, i.e., such that $\underline{\lambda}I \preceq M(x) \preceq \overline{\lambda}I$ for some constants $\underline{\lambda}, \overline{\lambda} > 0$ and all $x \in \mathbb{R}^n$. Denote the time-derivative of M(x) as $\dot{M}(x,u)$, with ij-th element

$$\dot{M}_{ij}(x,u) := \nabla M_{ij}(x)^{\mathsf{T}} (f(x) + B(x)u). \tag{15}$$

Then M(x) is a control contraction metric (CCM) for the system in Equation (1) if there exist a constant $\beta > 0$ and a variational controller $\delta_u = \delta_\pi(\delta_x, x, u)$ such that

$$\delta_x^{\mathsf{T}} \Big(\dot{M}(x, u) + M(x) A(x, u) + A(x, u)^{\mathsf{T}} M(x) \Big) \delta_x + 2\delta_x^{\mathsf{T}} M(x) B(x) \delta_\pi (\delta_x, x, u) \le -2\beta \delta_x^{\mathsf{T}} M(x) \delta_x$$
(16)

for all δ_x , x, and u. Given a CCM, an exponentially stabilizing tracking controller of the form

$$u = \pi_{\text{CCM}}(x, \bar{x}, \bar{u}) = \bar{u} + k(x, \bar{x}) \tag{17}$$

can be constructed by geodesic integration between x and \bar{x} (Manchester & Slotine, 2017; Singh et al., 2019; 2021), with overshoot $\alpha = \sqrt{\overline{\lambda}/\underline{\lambda}}$, decay rate β , and $k(\bar{x},\bar{x}) \equiv 0$. Alternatively, a differentiable controller of the form in Equation (17) achieves this same result if

$$\dot{M}(x,u) + \left(A(x,u) + B(x)\frac{\partial k}{\partial x}(x,\bar{x})\right)^{\mathsf{T}}M(x) + M(x)\left(A(x,u) + B(x)\frac{\partial k}{\partial x}(x,\bar{x})\right) \leq -2\beta M(x)$$
(18)

for all x, \bar{x} , and \bar{u} (Manchester & Slotine, 2017).

The exponential stability of the error dynamics in closed-loop with the tracking controller in Equation (17) is *certified* by the CCM M. Once again we see that attaining better closed-loop performance requires additional controloriented structure; in this case, this structure comprises the certificate M and the closed-loop contraction condition in Equation (18) that must be satisfied for *all* x, \bar{x} , and \bar{u} .

4. Jointly Learning Dynamics, Controllers, and Control-Oriented Structure

In the previous section, we introduced a number of tracking controllers for nonlinear control-affine systems. We also highlighted how increasing the complexity of the tracking controller often promises improved closed-loop performance with the requirement of knowing additional structure of the problem. For linearization-based LQR, only the

vector fields $(f, \{b_j\}_{j=1}^m)$ and their derivatives are needed. For state-dependent LQR, we also need to know the SDC factorizations $(A_0, \{A_j\}_{j=1}^m)$ of $(f, \{b_j\}_{j=1}^m)$. For CCM-based tracking control, we need to know (f, B) and a CCM M that together satisfy the constraint in Equation (18) for all x, \bar{x} , and \bar{u} . Even when (f, B) are known, synthesizing SDC factorizations (e.g., via the line integral in Equation (9)) or a CCM is a difficult problem that requires leveraging further structure in the dynamics (e.g., sparsity). This is generally not possible when (f, B) are learned from data for an unknown system using complex parametric function approximators (e.g., neural networks).

In this section, we describe our main contribution to learning how to control control-affine dynamical systems when we only have access to a finite labelled data set $\mathcal{D} \coloneqq \{(x^{(i)}, u^{(i)}, \dot{x}^{(i)})\}_{i=1}^N$ of input-output measurements of Equation (1). Specifically, we describe a few methods jointly learning a dynamics model and a tracking controller with unconstrained optimization, and focus on how this involves additionally modeling and learning control-oriented structure to enable a particular feedback law.

Learning dynamics from data Each method in this section learns a model of the dynamics in Equation (1). To this end, we define the regression loss

$$L_{\text{reg}}^{\text{dyn}}(f, B, \mathcal{D}) = \sum_{(x, u, \dot{x}) \in \mathcal{D}} \|\dot{x} - f(x) - B(x)u\|_{2}^{2}.$$
 (19)

If we instantiate (f,B) with parametric functions, such as neural networks, we can do gradient descent on this loss to fit (f,B) to the data. Thus, a naïve approach and our first baseline for learning how to control Equation (1) is to fit a differentiable model of (f,B) to the data $\mathcal D$ and then apply linearization-based tracking LQR from Section 3.1.

Learning SDC factorizations For state-dependent LQR, we need to learn the SDC factorizations denoted by $A := (A_0, \{A_j\}_{j=1}^m)$. For this, we define the regression loss

$$L_{\text{reg}}^{\text{SDC}}(\mathcal{A}, \mathcal{D}) = \sum_{\substack{(x, u, \dot{x}), \\ (\bar{x}, \bar{u}, \dot{\bar{x}}) \in \mathcal{D}}} \|\dot{e} - A_{\text{SDC}}(\bar{x}, \bar{u}, e)e - B(x)v\|_2^2, \quad (20)$$

which sums over *pairs* of labelled samples in the data set \mathcal{D} . We also need \mathcal{A} to be a set of valid SDC factorizations, for which we define the *unlabelled* data set $\mathcal{D}^{\text{SDC}}_{\text{aux}} = \{(x^{(i)}, \bar{x}^{(i)})\}_{i=1}^{N^{\text{SDC}}_{\text{aux}}}$ and the auxiliary loss

$$L_{\text{aux}}^{\text{SDC}}(f, B, \mathcal{A}, \mathcal{D}_{\text{aux}}^{\text{SDC}}) = \sum_{(x,\bar{x}) \in \mathcal{D}_{\text{aux}}^{\text{SDC}}} \left(\|f(x) - f(\bar{x}) - A_0(\bar{x}, e)e\|_2^2 + \sum_{j=1}^m \|b_j(x) - b_j(\bar{x}) - A_j(\bar{x}, e)e\|_2^2 \right). \tag{21}$$

Overall, we can learn (f, B, A) instantiated as parametric functions via gradient descent on the composite loss

$$\begin{split} L^{\text{SDC}}(f, B, \mathcal{A}, \mathcal{D}, \mathcal{D}_{\text{aux}}^{\text{SDC}}) &= L_{\text{reg}}^{\text{dyn}}(f, B, \mathcal{D}) + L_{\text{reg}}^{\text{SDC}}(\mathcal{A}, \mathcal{D}) \\ &+ L_{\text{aux}}^{\text{SDC}}(f, B, \mathcal{A}, \mathcal{D}_{\text{aux}}^{\text{SDC}}) \end{split} \tag{22}$$

This total loss is *semi-supervised* in that it is a function of both labelled and unlabelled data \mathcal{D} and $\mathcal{D}_{\text{aux}}^{\text{SDC}}$, respectively. Ideally, we would want to constrain \mathcal{A} to be a set of SDC factorizations of (f,B) consistent with Equation (8). Since we cannot straightforwardly enforce Equation (8) by construction, we use the auxiliary loss term in Equation (22) as a penalty-based relaxation, with as many unlabelled samples in $\mathcal{D}_{\text{aux}}^{\text{SDC}}$ as possible. This idea of relaxing pointwise functional constraints with sampling-based penalty terms is a common approach to learning global control-oriented structure (Richards et al., 2018; Singh et al., 2021; Sun et al., 2020; Dawson et al., 2022) and more generally in semi-infinite optimization (Zhang et al., 2010).

Learning CCMs This method is founded on the literature concerning joint learning of dynamics, controllers, and stability certificates (Singh et al., 2021; Sun et al., 2020; Dawson et al., 2022). For CCM-based tracking control, we need to learn a dynamics model (f,B), a uniformly positive-definite CCM M, and a feedback controller $u=\bar{u}+k(x,\bar{x})$ such that $k(\bar{x},\bar{x})\equiv 0$, that altogether satisfy the inequality in Equation (18) for all x,\bar{x} , and \bar{u} . We take some cues from Sun et al. (2020) to setup a loss function that will allow us to train all three components together with gradient descent, albeit with some adjustments to accommodate our lack of any knowledge of the dynamics (f,B) (which Sun et al. (2020) assume are known).

We first specify the desired overshoot $\alpha>0$, decay rate $\beta>0$, and eigenvalue lower bound $\underline{\lambda}>0$ as hyperparameters, and construct a candidate CCM M as

$$M(x) = \underline{\lambda}I + L(x)L(x)^{\mathsf{T}},\tag{23}$$

where $L:\mathbb{R}^n \to \mathbb{R}^{n \times n}$ is any parametric matrix function. This construction ensures $M(x) \succeq \underline{\lambda} I$ for all x. To ensure $k(\bar{x},\bar{x}) \equiv 0$, we follow Proposition 1 and let $k(x,\bar{x}) = K(x,\bar{x})(x-\bar{x})$ for any parametric function $K:\mathbb{R}^n \to \mathbb{R}^n \to \mathbb{R}^m \times n$. With the closed-loop variational matrix

$$\bar{A}(x,\bar{x},\bar{u}) = \frac{\partial f}{\partial x}(x) + \sum_{i=1}^{m} u_{j} \frac{\partial b_{j}}{\partial x}(x) + B(x) \frac{\partial k}{\partial x}(x,\bar{x}),$$

we collect terms of the inequality from Equation (18) in

$$C(x, \bar{x}, \bar{u}) = \dot{M}(x, u) + \bar{A}(x, \bar{x}, \bar{u})^{\mathsf{T}} M(x) + M(x) \bar{A}(x, \bar{x}, \bar{u}) + 2\beta M(x),$$
(24)

with $u=\bar{u}+k(x,\bar{x})=\bar{u}+K(x,\bar{x})(x-\bar{x}).$ Finally, with the unlabelled data set $\mathcal{D}^{\text{CCM}}_{\text{aux}}=\{(x^{(i)},\bar{x}^{(i)}),\bar{u}^{(i)})\}_{i=1}^{N^{\text{CCM}}},$

we define the auxiliary loss

$$L_{\text{aux}}^{\text{CCM}}(f, B, M, K, \mathcal{D}_{\text{aux}}^{\text{CCM}})$$

$$= \sum_{(x, \bar{x}, \bar{u}) \in \mathcal{D}_{\text{aux}}^{\text{CCM}}} \left(\max(0, \lambda_{\max}(C(x, \bar{x}, \bar{u}))) + \max(0, \lambda_{\max}(M(x)) - \alpha^2 \underline{\lambda}) \right)$$
(25)

where $\lambda_{\max}(\cdot)$ denotes the maximum eigenvalue operator. Overall, we can learn (f,B,M,K) instantiated as parametric functions via gradient descent on the total loss

$$L^{\text{CCM}}(f, B, M, K, \mathcal{D}, \mathcal{D}_{\text{aux}}^{\text{CCM}})$$

$$= L_{\text{reg}}^{\text{dyn}}(f, B, \mathcal{D}) + L_{\text{aux}}^{\text{CCM}}(f, B, M, K, \mathcal{D}_{\text{aux}}^{\text{CCM}}). \tag{26}$$

Much like in the state-dependent LQR case, this total loss is semi-supervised, although the auxiliary data set \mathcal{D}_{aux}^{CCM} also requires samples of the input \bar{u} . This loss function can be viewed as an unconstrained relaxation of the approach from Singh et al. (2021), who instead use pointwise inequalities derived from Equation (16) as exact constraints in an optimization over (f, B, M). However, Singh et al. (2021) only use linear-in-parameter approximators for (f, B, M)to construct a bi-convex program between (f, B) and M, investigate the regularizing effect of fitting (f, B, M) on the predictive capabilities of (f, B) in closed-loop, and do not learn a controller. In contrast, the modified setup described above jointly learns a dynamics model, certificate function, and controller that can each be expressed with complex parametric functions, so that in the next section we can compare with the learning setups for linearizationbased LOR and our novel state-dependent LOR.

5. Experiments

In this section, we experimentally investigate the three methods described previously for jointly learning a dynamics model, stabilizing tracking controller, and/or some control-oriented structure enabling the use thereof, namely:

- Naïve LQR learning: Fit a control-affine form (f,B) to labelled data $\mathcal{D} := \{(x^{(i)},u^{(i)},\dot{x}^{(i)})\}_{i=1}^N$ via gradient descent on the regression loss in Equation (19). Then perform linearized LQR.
- *CCM learning:* Jointly fit (f,B), a CCM M, and a gain matrix function K to labelled data \mathcal{D} and unlabelled data $\mathcal{D}^{\text{CCM}}_{\text{aux}} = \{(x^{(i)}, \bar{x}^{(i)}), \bar{u}^{(i)})\}_{i=1}^{N^{\text{CCM}}}$ via gradient descent on the composite loss in Equation (26). Then apply the controller $u = \bar{u} + K(x, \bar{x})(x \bar{x})$.
- Our state-dependent LQR ("SDC learning"): Jointly fit (f,B) and SDC factorizations $(A_0,\{A_j\}_{j=1}^m)$ to labelled data \mathcal{D} and unlabelled data $\mathcal{D}^{\text{SDC}}_{\text{aux}} = \{(x^{(i)},\bar{x}^{(i)})\}_{i=1}^{N_{\text{aux}}^{\text{SDC}}}$ via gradient descent on the composite loss in Equation (22). Then perform state-dependent LQR.

We evaluate these methods on two nonlinear benchmark systems: a planar spacecraft with an offset center-of-mass, and the classic planar vertical-take-off-and-landing (PV-TOL) vehicle (Hauser et al., 1992).

Example: Spacecraft Our planar spacecraft has mass m and rotational moment of inertia J. Its center of mass is offset at the position $(d_x,d_y)\in\mathbb{R}^2$ with respect to the body-fixed frame. Its state is $x=(p_x,p_y,\theta,\dot{p}_x,\dot{p}_y,\dot{\theta})\in\mathbb{R}^6$, where (p_x,p_y) is its position and θ is its heading angle. The control is $u=(F_x,F_y,M)\in\mathbb{R}^3$, where (F_x,F_y) are the applied thrusts along the body-fixed x-axis and y-axis, respectively, and M is the applied moment. The controlaffine dynamics of the spacecraft are given by

$$f(x) = \frac{1}{m} \begin{bmatrix} \dot{p}_x \\ \dot{p}_y \\ \dot{\theta} \\ \dot{\theta}^2 d_x \\ \dot{\theta}^2 d_y \\ 0 \end{bmatrix}, B(x) = \frac{1}{mJ} \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ J + d_y^2 & -d_x d_y & d_y \\ -d_x d_y & J + d_x^2 & -d_x \\ m d_y & -m d_x & m \end{bmatrix}.$$

Due to the non-zero offset (d_x, d_y) , the nonlinear term $\dot{\theta}^2$ is introduced into the drift dynamics.

Example: PVTOL We consider a "planar quadrotor" PVTOL with mass m, rotational moment of inertia J, moment arm length ℓ between the center of mass and each of two rotors, and gravitational acceleration g. Its state is $x=(p_x,p_y,\phi,v_x,v_y,\dot{\phi})\in\mathbb{R}^6$, where (p_x,p_y) is its position, ϕ is its roll angle, and (v_x,v_y) is its velocity in the body-fixed frame. The control is $u=(F_R,F_L)\in\mathbb{R}^3$, where F_R and F_L are the applied thrusts by the right and left rotors, respectively, along the body-fixed y-axis. The control-affine dynamics of this PVTOL are given by

$$f(x) = \begin{bmatrix} v_x \cos \phi - v_y \sin \phi \\ v_x \sin \phi + v_y \cos \phi \\ \dot{\phi} \\ v_y \dot{\phi} - g \sin \phi \\ -v_x \dot{\phi} - g \cos \phi \\ 0 \end{bmatrix}, \ B(x) = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ \frac{1}{m} & \frac{1}{m} \\ \ell/J & -\ell/J \end{bmatrix}.$$

The PVTOL is a highly nonlinear, underactuated, non-minimum-phase dynamical system, and thus serves as a challenging benchmark for learning-based control.

Training Details For each system, we begin by uniformly sampling points $\{(x^{(i)},u^{(i)})\}_{i=1}^N$ from a bounded state-control set $\mathcal{X} \times \mathcal{U} \subset \mathbb{R}^n \times \mathbb{R}^m$, and evaluating the true dynamics to form the labelled data \mathcal{D} . Both \mathcal{X} and \mathcal{U} are described in Appendix A, along with other implementation details and hyperparameters. We additionally uniformly sample unlabelled data sets $\mathcal{D}_{\text{aux}}^{\text{CCM}}$ and $\mathcal{D}_{\text{aux}}^{\text{SDC}}$ for use with

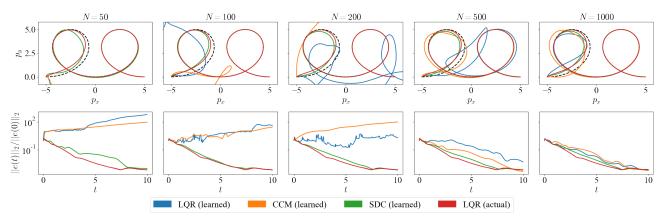


Figure 1. Trajectory tracking results for the PVTOL system on a double loop-the-loop trajectory. The top row qualitative depicts the closed-loop trajectories for each method overlayed on the reference trajectory (black dashed). The bottom row shows the normalized tracking error over time. Plots proceed from left to right with an increasing amount N of labelled training data. Our SDC method is the only learning-based approach that successfully tracks the trajectory for all N.

the CCM and SDC learning methods, respectively, from \mathcal{X} and \mathcal{U} . We vary the labelled training set size N to investigate the data efficiency of each method, with a constant number of auxiliary points $N_{\rm aux}^{\rm CCM} = N_{\rm aux}^{\rm SDC} = 10000$. Each function in $(f, B, M, K, A_0, \{A_j\}_{j=1}^m)$ is approximated as a feedforward neural network with the same number of fully connected hidden layers, and appropriately shaped input and output dimensions using Python and JAX (Bradbury et al., 2018). For each method, the appropriate subset of these functions is trained via the Adam optimizer (Kingma & Ba, 2015) on the corresponding loss function. Training is performed for a large number of epochs while the loss on a held-out validation set is monitored; for each method, the model parameters corresponding to the lowest validation loss are chosen for testing. This training procedure is repeated for each method across 5 random seeds.

Testing and Results In order to test the controllers learned with each method, we must first generate dynamically feasible trajectories for tracking. We first evaluate the PVTOL system qualitatively; we leverage its differential flatness (Ailon, 2010) to generate a feasible pair $(\bar{x}(t), \bar{u}(t))$ yielding the double loop-the-loop shape in Figure 1. For a single random seed, we plot the closedloop trajectory from using each learned controller to track the loop-the-loop. We also plot the results of using linearization-based LQR with the true dynamics as a form of ground truth. We repeat this test for various sizes Nof the labelled training data set \mathcal{D} , and plot the trajectories in (p_x, p_y) -space and the normalized tracking error $\frac{\|e(t)\|_2}{\|e(0)\|_2}$ over time. Our SDC method is the only learning-based method that succeeds for every size N, while the learned LQR and CCM controllers outright fail for smaller data set sizes. This is initial evidence of the data efficiency in learning SDC factorizations for state-dependent LQR.

For more thorough testing, we want to generate many tra-

jectories in a manner applicable to both the spacecraft and PVTOL. To this end, for each system we generate $N_{\text{test}} = 100$ feasible trajectories $\mathcal{T}_{\text{test}} \coloneqq \{(\bar{x}^{(k)}(t), \bar{u}^{(k)}(t))\}_{k=1}^{N_{\text{test}}}$ by solving the optimal control problem

$$\begin{aligned} & \underset{\bar{x}(\cdot),\bar{u}(\cdot)}{\text{minimize}} & \int_{0}^{T} \left(\bar{x}(t)^{\mathsf{T}} Q \bar{x}(t) + \bar{u}(t)^{\mathsf{T}} R \bar{u}(t) \right) dt \\ & \text{subject to } & \dot{\bar{x}}(t) = f(\bar{x}(t)) + B(\bar{x}(t)) \bar{u}(t), \\ & \bar{x}(0) = \bar{x}_{0}^{(k)}, \; \bar{x}(T) = 0, \; u_{\text{lb}} \leq \bar{u}(t) \leq u_{\text{ub}} \end{aligned}$$

for different initial conditions $\bar{x}_0^{(k)}$ sampled uniformly from \mathcal{X} , where (Q,R) are positive-definite weight matrices and $(u_{\mathrm{lb}},u_{\mathrm{ub}})$ are control input bounds. Specifically, we use CasADi (Andersson et al., 2019) to transcribe this problem into a nonlinear multiple shooting optimization that is passed to and solved by the Ipopt solver. Then, for each system, test trajectory, and tracking controller, we uniformly sample an initial state $x_0^{(k)} \neq \bar{x}_0^{(k)}$ from \mathcal{X} for the system, and simulate the closed-loop system.

Figure 2 displays the normalized tracking error $\frac{\|e(t)\|_2}{\|e(0)\|_2}$ over time for both the spacecraft and the PVTOL, for various training set sizes N. For each method, system, and N, the median normalized tracking error across the test trajectory set $\mathcal{T}_{\text{test}}$ is plotted along with shaded regions denoting the interquartile range over time. Once again we observe the data efficiency of our SDC learning method for both systems. Our method even *outperforms the ground truth* LQR controller at higher values of N for the PVTOL, despite having to learn the dynamics. The naïve learned LQR method unsurprisingly converges to performance similar to the ground truth LQR controller as N increases. Notably, the CCM-based controller has the most trouble overall, thereby highlighting its data inefficiency and brittleness.

To complete our assessment, we repeat the training procedure and tests underlying Figure 2 with 5 random seeds,

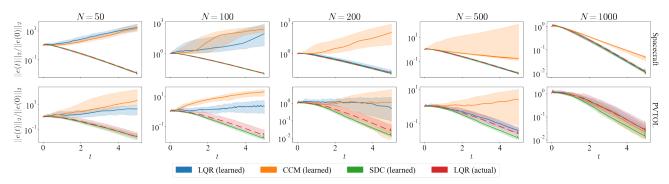


Figure 2. Trajectory tracking results for both the spacecraft and PVTOL systems for $N_{\text{test}} = 100$ trajectories each. The top and bottom rows show the normalized tracking error over time for the spacecraft and PVTOL, respectively. Plots proceed from left to right with an increasing amount N of labelled training data. Colored lines represent the median across all trajectories at each time t, while shaded regions depict interquartile ranges. Our SDC method consistently outperforms the considered baseline learning methods.

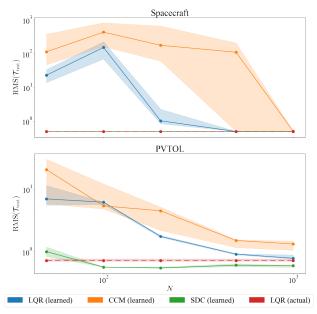


Figure 3. RMS tracking error as a function of the labelled training data set size N, averaged across $N_{\rm test}=100$ test trajectories (see Equation (27)). Colored lines denote medians across 5 random seeds, while shaded regions depict interquartile ranges. Our SDC method outperforms all other methods, even the ground-truth LQR on the PVTOL system.

and aggregate the results in Figure 3. To do this, we consider the average root mean squared (RMS) error

$$RMS(\mathcal{T}_{test}) := \frac{1}{N_{test}} \sum_{k=1}^{N_{test}} \sqrt{\frac{1}{T} \int_{0}^{T} \frac{\|e^{(k)}(t)\|_{2}^{2}}{\|e^{(k)}(0)\|_{2}^{2}}} dt \quad (27)$$

across all test trajectories for each random seed and training set size N. In Figure 3, we plot the median and interquartile range of $\mathrm{RMS}(\mathcal{T}_{\mathrm{test}})$ across random seeds as a function of N. From this plot, we can see an even starker contrast between the performance of our SDC learning method and the others. For the spacecraft, our method matches the performance of the ground truth LQR, which is not surprising

given that the spacecraft dynamics are only mildly nonlinear. For the highly nonlinear PVTOL, our method begins outperforming the ground truth LQR at only N=100. Meanwhile, both the learned LQR and CCM controllers struggle until more training data is used, thereby highlighting their data inefficiency compared to our method.

6. Conclusions and Future Work

In this paper, we studied how to jointly learn a dynamics model and a stabilizing tracking controller from only a finite data set of input-output measurements of an unknown dynamical system. We highlighted the importance of not only learning the dynamics, but also control-oriented structure that enables performant controller design. For this purpose, we proposed a novel state-dependent LQR tracking controller that relies on learning SDC factorizations of the dynamics. Inspired by the literature, we compared our method to navely learning a model for linearization-based LQR, and to methods that couple learned controllers with learned certificate functions. Overall, we found that our method was performant and data-efficient.

Future Work We view this paper in part as a critique of methods that try to enforce closed-loop stabilizability guarantees by penalizing sampled violations of certificate conditions like Equation (18). As we have demonstrated, such methods are not data efficient in learning good controllers, although the performance guarantees they are meant to certify (e.g., exponential stability) are attractive. Unlike these methods, our method learns intrinsic structure in the dynamics to enable control, rather than simultaneously learning a parametric controller. Thus, an interesting avenue for future research lies in building system models that are intrinsically stabilizable. This could build off of existing work in parameterizing dynamics models in part by stability certificates such that they are stable by construction (Manek & Kolter, 2019; Revay et al., 2021), albeit for the controlled case.

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A. Hyperparameters and Implementation Details

Physical Parameters For the spacecraft, we set its mass to m=0.5, rotational moment of inertia to J=0.005, and its center-of-mass offset to $(d_x,d_y)=(0.1,0.1)$. For the PVTOL, we set the its mass to m=0.5, arm length to $\ell=0.25$, rotational moment of inertia to J=0.005, and gravitational acceleration to g=9.81.

Hyperparameters Each function in $(f, B, M, K, A_0, \{A_j\}_{j=1}^m)$ is approximated as a feedforward neural network with two hidden layers and 128 hidden tanh activation units per layer. We use the Adam optimizer (Kingma & Ba, 2015) with a learning rate of 10^{-3} and otherwise default hyperparameters. Training is performed for 50000 epochs while the loss on a held-out validation set of size 0.10N is monitored, where N is the size of the labelled training data set. For each method, the model parameters corresponding to the lowest validation loss are chosen for testing.

For the CCM-based learning method, since Equation (18) is homogeneous in M(x), we choose $\underline{\lambda}=0.1$ without loss of generality. Additionally, we fix the overshoot $\alpha=10$ and the decay rate $\beta=0.5$ in the auxiliary loss Equation (26). For both the CCM and SDC learning methods, we use $N_{\rm aux}^{\rm CCM}=N_{\rm aux}^{\rm SDC}=10000$ unlabelled samples.

Sampling For sampling states and inputs, we draw uniformly from bounded sets $\mathcal{X} \subset \mathbb{R}^n$ and $\mathcal{U} \subset \mathbb{R}^m$, respectively. For the spacecraft, we use

$$\mathcal{X} = \{ x \in \mathbb{R}^6 \mid -c \leq x \leq c, \ c := (1, 1, \pi, 0.2, 0.2, 0.25) \}$$

$$\mathcal{U} = \{ u \in \mathbb{R}^3 \mid -c \leq u \leq c, \ c := (1, 1, 0.1) \}$$
(28)

For the PVTOL, we use

$$\mathcal{X} = \{ x \in \mathbb{R}^6 \mid -c \le x \le c, \ c := (10, 10, \pi/3, 2, 1, \pi/3) \},$$

$$\mathcal{U} = \{ u \in \mathbb{R}^2 \mid (0.1mq, 0.1mq) \le u \le (2mq, 2mq) \},$$
(29)

where m and g are the vehicle mass and gravitational acceleration, respectively. We also use \mathcal{U} to define the control bounds in the optimal control problem for generating test trajectories.

Testing When generating test trajectories with the optimal control problem in Section 5, we use Q = I and R = I in the cost function for both systems. For simulating the linearization-based and state-dependent LQR controllers, we also use Q = I and R = I in their corresponding Riccati equations.