# AA203 Optimal and Learning-based Control

Course overview; control, stability, performance metrics





#### Course mechanics

#### Teaching team:

- Instructors: Ed Schmerling (OH: W 11am-12pm; Project OH: W 4:30-5:30pm)

  James Harrison (OH: M 10-11am; Project OH: Th 2-3pm)
- CAs: Matt Tsao and Spencer M. Richards (OH: Tu 4-6pm, Th 8:30-10:30am)

#### Logistics:

- Class info, lectures, and homework assignments on class web page: <a href="http://asl.stanford.edu/aa203/">http://asl.stanford.edu/aa203/</a>
- Forum: <a href="http://piazza.com/stanford/spring2021/aa203">http://piazza.com/stanford/spring2021/aa203</a>
- For urgent questions: <u>aa203-spr2021-staff@lists.stanford.edu</u>

### Course requirements

- Homework: there will be a total of four problem sets
- Homework submissions: <a href="https://www.gradescope.com/courses/257531">https://www.gradescope.com/courses/257531</a>
- Final project (details on the course website)
- Grading:
  - homework 60% (15% per HW)
  - final project 40%

#### Course material

 Course notes: a set of course notes will be provided covering all the content presented in the lectures

• Recitations: Friday lecture sessions (F 10:30-11:50AM, weeks 2—5) led by the CAs covering relevant tools (computational and mathematical)

 Textbooks that may be valuable for context or further reference are listed in the syllabus

### Prerequisites

- Strong familiarity with calculus (e.g., CME100)
- Strong familiarity with linear algebra (e.g., EE263 or CME200)
- Familiarity with optimization (e.g., EE364a, CME307, CS269o, AA222)
- To get the most out of this class, at least one of:
  - A course in machine learning (e.g., CS229, CS230, CS231n) or
  - A course in control (e.g., ENGR105, ENGR205, AA212)

Homework 0 (ungraded) is out now to gauge preparedness.

# Today's Outline

1. Context and course goals

2. State-space models

3. Problem formulation for optimal control

# Today's Outline

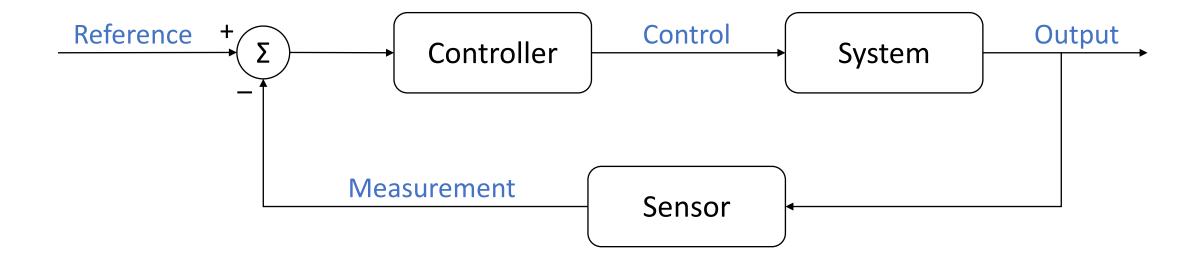
1. Context and course goals

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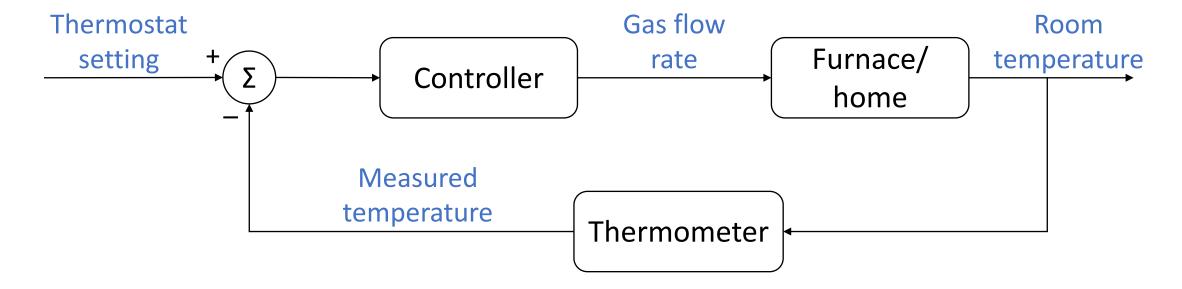
### Feedback control

Tracking a reference signal

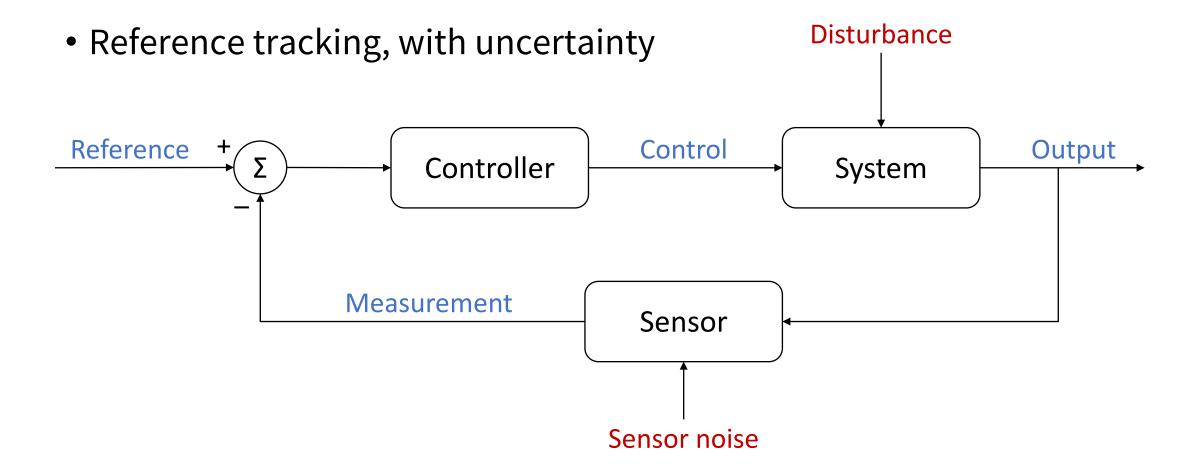


#### Feedback control

Tracking a reference signal

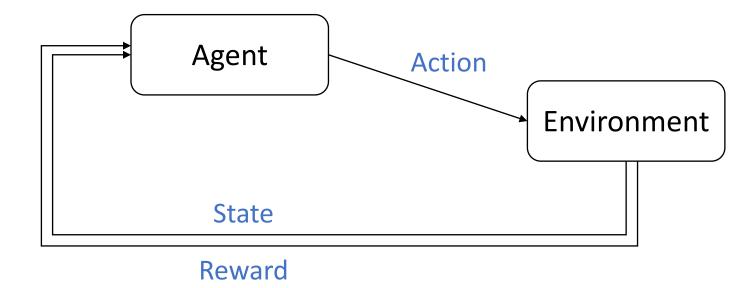


#### Feedback control



# Reinforcement learning

• A brief aside...



#### Feedback control desiderata

- Stability: multiple notions; loosely system output is "under control"
- Tracking: the output should track the reference "as closely as possible"
- Disturbance rejection: the output should be "as insensitive as possible" to disturbances/noise
- Robustness: controller should still perform well up to "some degree of" model misspecification

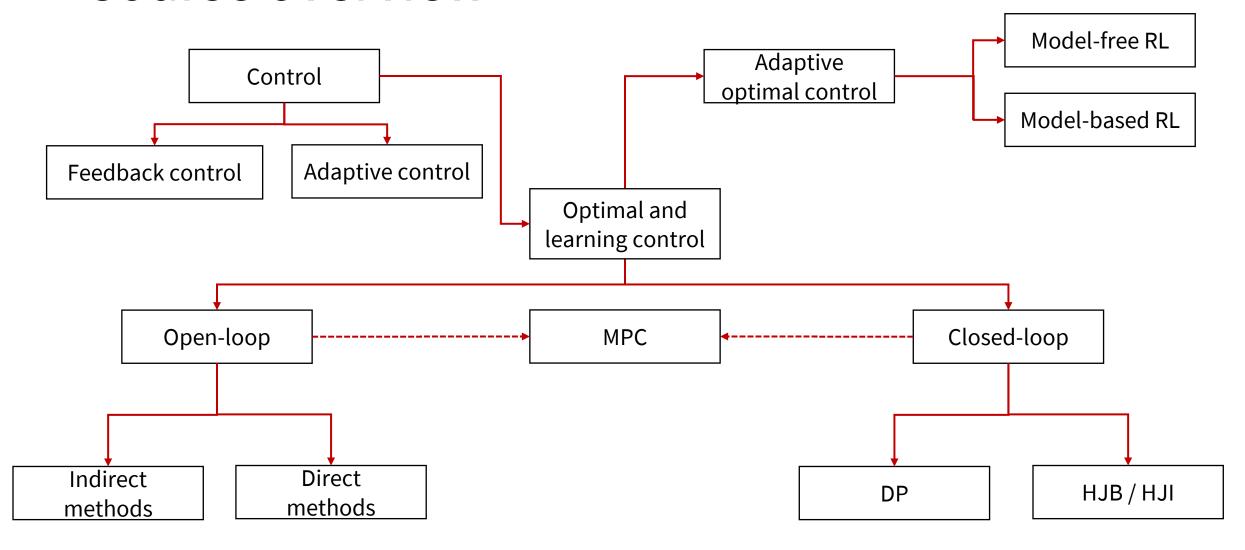
# What's missing?

• Performance: mathematical quantification of the above desiderata, and providing a control that best realizes the tradeoffs between them

 Planning: providing an appropriate reference trajectory for the controller to track (particularly nontrivial, e.g., when controlling mobile robots)

• Learning: a controller that adapts to an initially unknown, or possibly time-varying system

#### Course overview



## Course goals

To learn the *theoretical* and *implementation* aspects of main techniques in optimal and learning-based control

To provide a *unified framework and context* for understanding and relating these techniques to each other

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#### Mathematical model

$$\dot{x}_1(t) = f_1(x_1(t), x_2(t), \dots, x_n(t), u_1(t), u_2(t), \dots, u_m(t), t) 
\dot{x}_2(t) = f_2(x_1(t), x_2(t), \dots, x_n(t), u_1(t), u_2(t), \dots, u_m(t), t) 
\vdots 
\dot{x}_n(t) = f_n(x_1(t), x_2(t), \dots, x_n(t), u_1(t), u_2(t), \dots, u_m(t), t)$$

#### Where

- $x_1(t), x_2(t), \ldots, x_n(t)$  are the state variables
- $u_1(t), u_2(t), \ldots, u_m(t)$  are the control inputs

#### Mathematical model

In compact form

$$\dot{\mathbf{x}}(t) = \mathbf{f}(\mathbf{x}(t), \mathbf{u}(t), t)$$

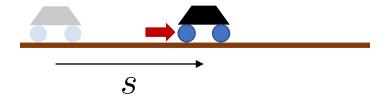
- a history of control input values during the interval  $[t_0,t_f]$  is called a control history and is denoted by  ${\bf u}$
- a history of state values during the interval  $[t_0, t_f]$  is called a *state trajectory* and is denoted by  ${\bf x}$

## Illustrative example

Double integrator: point mass under controlled acceleration

$$\ddot{s}(t) = a(t)$$

$$\begin{bmatrix} \dot{s} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} v \\ a \end{bmatrix}$$



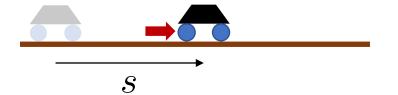
### Example system

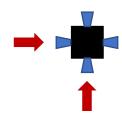
Double integrator: point mass under controlled acceleration

$$\begin{bmatrix} \dot{s} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} s \\ v \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} a \end{bmatrix}$$

$$\dot{\mathbf{x}}(t) = A \quad \mathbf{x}(t) + B \quad \mathbf{u}(t)$$

$$\begin{bmatrix} \dot{\mathbf{s}} \\ \dot{\mathbf{v}} \end{bmatrix} = \begin{bmatrix} 0 & I \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \mathbf{s} \\ \mathbf{v} \end{bmatrix} + \begin{bmatrix} 0 \\ I \end{bmatrix} \begin{bmatrix} \mathbf{a} \end{bmatrix}$$

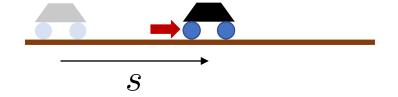




### Example controller

Let's drive from  $[5, 0]^T$  to  $[0, 0]^T$ .

Proposal: use a linear feedback control law.



$$a = -k_p s - k_d v$$

$$\begin{bmatrix} \dot{s} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} s \\ v \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} k_p & k_d \end{bmatrix} \begin{bmatrix} s \\ v \end{bmatrix}$$

$$\begin{bmatrix} \dot{s} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -k_p & -k_d \end{bmatrix} \begin{bmatrix} s \\ v \end{bmatrix} \qquad \left( \dot{\mathbf{x}}(t) = (A - BK)\mathbf{x}(t) \right)$$

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## Analyzing stability

$$\begin{bmatrix} \dot{s} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -k_p & -k_d \end{bmatrix} \begin{bmatrix} s \\ v \end{bmatrix}$$

$$\begin{bmatrix} s(t) \\ v(t) \end{bmatrix} = \exp\left( \begin{bmatrix} 0 & 1 \\ -k_p & -k_d \end{bmatrix} t \right) \begin{bmatrix} s(0) \\ v(0) \end{bmatrix}$$

$$\begin{bmatrix} s(t) \\ v(t) \end{bmatrix} = V^{-1} \begin{bmatrix} e^{\lambda_+ t} & 0 \\ 0 & e^{\lambda_- t} \end{bmatrix} V \begin{bmatrix} s(0) \\ v(0) \end{bmatrix} \qquad \begin{bmatrix} s(t) \\ v(t) \end{bmatrix} = e^{\lambda t} V^{-1} \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} V \begin{bmatrix} s(0) \\ v(0) \end{bmatrix}$$

where 
$$\lambda_{\pm} = \left(-k_d \pm \sqrt{k_d^2 - 4k_p}\right)/2$$

$$\begin{bmatrix} s(t) \\ v(t) \end{bmatrix} = e^{\lambda t} V^{-1} \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} V \begin{bmatrix} s(0) \\ v(0) \end{bmatrix}$$

where 
$$\lambda = -k_d/2$$
, if  $k_d^2 - 4k_p = 0$ 

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## Analyzing stability

$$\begin{bmatrix} s(t) \\ v(t) \end{bmatrix} = V^{-1} \begin{bmatrix} e^{\lambda_{+}t} & 0 \\ 0 & e^{\lambda_{-}t} \end{bmatrix} V \begin{bmatrix} s(0) \\ v(0) \end{bmatrix}$$
 or where  $\lambda_{\pm} = \left( -k_d \pm \sqrt{k_d^2 - 4k_p} \right) / 2$ 

 $\operatorname{Re}(\lambda)$   $\rightarrow$  exponential growth (> 0), exponential decay (< 0), or constant (=0)

 $\operatorname{Im}(\lambda) \rightarrow \operatorname{sinusoidal} \operatorname{oscillation}$ 

$$\begin{bmatrix} s(t) \\ v(t) \end{bmatrix} = e^{\lambda t} V^{-1} \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} V \begin{bmatrix} s(0) \\ v(0) \end{bmatrix}$$

where  $\lambda = -k_d/2$ , if  $k_d^2 - 4k_p = 0$ 

system comes to a system exponentially converges to 0

system drifts off system oscillates

at least one eigenvalue has positive real part; system blows up

### Mathematical definitions of stability

#### Many notions:

- Asymptotic stability
  - Global: all trajectories converge to the equilibrium
  - Local: all trajectories starting near the equilibrium converge to the equilibrium
- Exponential stability
  - Same as asymptotic stability, but with exponential rate
- Marginal stability
- Bounded-input, bounded-output stability
- Lyapunov stability

$$\min \int_0^{t_f} \|\mathbf{x}(t)\|_2 + \|\mathbf{u}(t)\|_2 dt$$
s.t. 
$$\dot{\mathbf{x}}(t) = A\mathbf{x}(t) + B\mathbf{u}(t)$$

$$\mathbf{x}(0) = \mathbf{x_0}$$

$$\min \int_0^{t_f} ||\mathbf{x}(t)||_2 + ||\mathbf{u}(t)||_2 dt$$
s.t.  $\dot{\mathbf{x}}(t) = A\mathbf{x}(t) + B\mathbf{u}(t)$ 

$$\mathbf{x}(0) = \mathbf{x_0}, \ \mathbf{x}(t_f) = \mathbf{x_f}$$

$$\min \int_{0}^{t_f} \mathbf{x}(t)^T Q \mathbf{x}(t) + \mathbf{u}(t)^T R \mathbf{u}(t) dt$$
s.t.  $\dot{\mathbf{x}}(t) = A \mathbf{x}(t) + B \mathbf{u}(t)$ 

$$\mathbf{x}(0) = \mathbf{x_0}, \ \mathbf{x}(t_f) = \mathbf{x_f}$$

min 
$$\int_0^{t_f} \mathbf{x}(t)^T Q \mathbf{x}(t) + \|\mathbf{u}(t)\|_1 dt$$
s.t. 
$$\dot{\mathbf{x}}(t) = A \mathbf{x}(t) + B \mathbf{u}(t)$$

$$\mathbf{x}(0) = \mathbf{x_0}, \ \mathbf{x}(t_f) = \mathbf{x_f}$$

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#### Problem formulation

- Mathematical description of the system to be controlled
- Statement of the constraints
- Specification of a performance criterion

### Performance measure

$$J = h(\mathbf{x}(t_f), t_f) + \int_{t_0}^{t_f} g(\mathbf{x}(t), \mathbf{u}(t), t) dt$$

- h and g are scalar functions
- $t_f$  may be specified or free

#### Constraints

initial and final conditions (boundary conditions)

$$\mathbf{x}(t_0) = \mathbf{x}_0, \qquad \mathbf{x}(t_f) = \mathbf{x}_f$$

constraints on state trajectories

$$\underline{X} \le \mathbf{x}(t) \le \overline{X}$$

control authority

$$\underline{U} \le \mathbf{u}(t) \le \overline{U}$$

and many more...

#### Constraints

- A control history which satisfies the control constraints during the entire time interval  $[t_0, t_f]$  is called an admissible control
- A state trajectory which satisfies the state variable constraints during the entire time interval  $\left[t_0,t_f\right]$  is called an admissible trajectory

### Optimal control problem

Find an admissible control **u**\* which causes the system

$$\dot{\mathbf{x}}(t) = \mathbf{f}(\mathbf{x}(t), \mathbf{u}(t), t)$$

to follow an *admissible trajectory* **x**\* that minimizes the performance measure

$$J = h(\mathbf{x}(t_f), t_f) + \int_{t_0}^{t_f} g(\mathbf{x}(t), \mathbf{u}(t), t) dt$$

### Optimal control problem

#### Comments:

- minimizer  $(\mathbf{x}^*, \mathbf{u}^*)$  called optimal trajectory-control pair
- existence: in general, not guaranteed
- uniqueness: optimal control may not be unique
- minimality: we are seeking a global minimum
- for maximization, we rewrite the problem as  $\min_{\mathbf{u}} -J$

### Form of optimal control

- 1. if  $\mathbf{u}^* = \pi(\mathbf{x}(t), t)$ , then  $\pi$  is called optimal control law or optimal policy (closed-loop)
  - important example:  $\pi(\mathbf{x}(t), t) = F \mathbf{x}(t)$
- 2. if  $\mathbf{u}^* = e(\mathbf{x}(t_0), t)$ , then the optimal control is *open-loop* 
  - optimal only for a particular initial state value

#### Discrete-time formulation

- System:  $\mathbf{x}_{k+1} = \mathbf{f}(\mathbf{x}_k, \mathbf{u}_k, k), k = 0, ..., N-1$
- Control constraints:  $\mathbf{u}_k \in U$
- Cost:

$$J(\mathbf{x}_0; \mathbf{u}_0, ..., \mathbf{u}_{N-1}) = h_N(\mathbf{x}_N) + \sum_{k=0}^{N-1} g_k(\mathbf{x}_k, \mathbf{u}_k, k)$$

Decision-making problem:

$$J^*(\mathbf{x}_0) = \min_{\mathbf{u}_k \in U, k=0,...,N-1} J(\mathbf{x}_0; \mathbf{u}_0, ..., \mathbf{u}_{N-1})$$

Extension to stochastic setting will be covered later in the course

### Next class

Introduction to learning;
System identification and adaptive control