

# MATH 525

## Section 3.5: The Extended Golay Code

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## The Extended Golay Code



Marcel Golay

- In 1949 Marcel Golay noticed that

$$\binom{23}{0} + \binom{23}{1} + \binom{23}{2} + \binom{23}{3} = 2^{11},$$

so he started to look for a perfect  $(23, 12, 7)$ -linear code, and succeeded. We will study this code, now known as the Golay code, in the next section.

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- The *extended* Golay code,  $C_{24}$ , is the linear code of length 24, dimension 12, and distance 8, whose generator matrix is:

$$G = [I_{12}|B]$$

where  $B$  is the  $12 \times 12$  matrix

$$B = \left[ \begin{array}{c|c} B_1 & j^T \\ \hline j & 0 \end{array} \right] \text{ where } j = [1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1]$$

and  $B_1$  is displayed on the next slide.

- The following concept will be used on the next slide: The *left-cyclic shift* of the vector  $(v_1, v_2, v_3, \dots, v_n) \in K^n$  is the vector

$$(v_2, v_3, \dots, v_n, v_1).$$

For example, the left-cyclic shift of 1011 is 0111.

$$B = \left[ \begin{array}{cccccccccccc|c} 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 \\ 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1 \\ 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 \\ 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 \\ 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 1 & 1 \\ 1 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 1 \\ 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 1 & 1 \\ \hline 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 0 \end{array} \right]$$

Each row of the  $11 \times 11$  submatrix  $B_1$  is a left-cyclic shift of the previous row.

[illegible]

$$B = \begin{bmatrix} \mathbf{b}_1 & 1 & 1 & 0 & 1 & | & 1 & 1 & 0 & 0 & | & 0 & 1 & 0 & 1 \\ \mathbf{b}_2 & 1 & 0 & 1 & 1 & | & 1 & 0 & 0 & 0 & | & 1 & 0 & 1 & 1 \\ \mathbf{b}_3 & 0 & 1 & 1 & 1 & | & 0 & 0 & 0 & 1 & | & 0 & 1 & 1 & 1 \\ \mathbf{b}_4 & 1 & 1 & 1 & 0 & | & 0 & 0 & 1 & 0 & | & 1 & 1 & 0 & 1 \\ \mathbf{b}_5 & 1 & 1 & 0 & 0 & | & 0 & 1 & 0 & 1 & | & 1 & 0 & 1 & 1 \\ \mathbf{b}_6 & 1 & 0 & 0 & 0 & | & 1 & 0 & 1 & 1 & | & 0 & 1 & 1 & 1 \\ \mathbf{b}_7 & 0 & 0 & 0 & 1 & | & 0 & 1 & 1 & 0 & | & 1 & 1 & 1 & 1 \\ \mathbf{b}_8 & 0 & 0 & 1 & 0 & | & 1 & 1 & 0 & 1 & | & 1 & 1 & 0 & 1 \\ \mathbf{b}_9 & 0 & 1 & 0 & 1 & | & 1 & 0 & 1 & 1 & | & 1 & 0 & 0 & 1 \\ \mathbf{b}_{10} & 1 & 0 & 1 & 1 & | & 0 & 1 & 1 & 1 & | & 0 & 0 & 0 & 1 \\ \mathbf{b}_{11} & 0 & 1 & 1 & 0 & | & 1 & 1 & 1 & 0 & | & 0 & 0 & 1 & 1 \\ \mathbf{b}_{12} & 1 & 1 & 1 & 1 & | & 1 & 1 & 1 & 1 & | & 1 & 1 & 1 & 0 \end{bmatrix}$$

## Properties of $C_{24}$

- ①  $|C| = 2^{12} = 4096$ .
- ② A parity-check matrix for  $C_{24}$  is  $\begin{bmatrix} B \\ I_{12} \end{bmatrix}$ .
- ③ Another parity-check matrix is  $H = \begin{bmatrix} I_{12} \\ B \end{bmatrix}$ .

Proof.

This follows from the observation that

$$G \cdot H = [I_{12}|B] \cdot \begin{bmatrix} I_{12} \\ B \end{bmatrix} = I_{12} + B^2 = I_{12} + BB^T$$

(because  $B = B^T$ ). By direct inspection,  $\mathbf{b}_1 \cdot \mathbf{b}_1 = 1$  and  $\mathbf{b}_1 \cdot \mathbf{b}_i = 0$  for all  $i > 1$ . From the cyclic structure of  $B_1$ , it follows that  $\mathbf{b}_i \cdot \mathbf{b}_j = 0$  whenever  $i \neq j$  (note that if  $j > i$ , then  $\mathbf{b}_i \cdot \mathbf{b}_j = \mathbf{b}_1 \cdot \mathbf{b}_{j-i}$ ). Therefore,  $I_{12} + BB^T = I_{12} + I_{12} = \mathbf{0}$ , i.e.,  $G \cdot H = \mathbf{0}$ . □

- ④  $C_{24}$  is self-dual, i.e.,  $C_{24}^\perp = C_{24}$ .
- ⑤  $d(C_{24}) = 8$ .

Proof.

The proof of the last statement is divided into two parts:

- Part 1) Show that the weight of any codeword in  $C_{24}$  is congruent to zero modulo 4.
  - Part 2) Show that there is no word of weight 4.
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Set  $H = \left[ \frac{I_{12}}{B} \right]$  as the parity-check matrix for  $C_{24}$ .

**Algorithm for Decoding the Extended Golay Code  $C_{24}$ :**

- Input: The received vector  $r = (r_1, r_2, \dots, r_{24}) \in K^{24}$ .
  - The output will be  $u$ , the estimated error vector.
- 1) Compute  $s = r \cdot H$ .
  - 2) If  $\text{wt}(s) \leq 3$  then  $u = [s, 0]$ . EXIT.
  - 3) If  $\text{wt}(s + b_i) \leq 2$  for row  $i$  of  $B$  then  $u = [s + b_i, e_i]$ . EXIT.
  - 4) Compute  $sB$ .
  - 5) If  $\text{wt}(sB) \leq 3$  then  $u = [0, sB]$ . EXIT.
  - 6) If  $\text{wt}(sB + b_i) \leq 2$  for row  $i$  of  $B$  then  $u = [e_i, sB + b_i]$ . EXIT.
  - 7) Request retransmission or declare failure.

**Example**

Decode the following received words, assuming the code being used is  $C_{24}$ :

- (a)  $r = (0000 \ 0100 \ 0101 \ 1000 \ 1111 \ 0001)$ .
- (b)  $r = (1000 \ 0100 \ 1010 \ 1100 \ 1100 \ 1000)$ .
- (c)  $r = (1000 \ 0110 \ 1010 \ 1000 \ 1100 \ 1000)$ .

**Helpful calculations:**

In (a),  $s_1 = rH = (0101 \ 0010 \ 0000)$ .

In (b),  $s_1 = rH = (1111 \ 0101 \ 0001)$ .

In (c),  $s_1 = rH = (0100 \ 1111 \ 1010)$  and  $s_2 = s_1 B = (1111 \ 1000 \ 1111)$ .

To see an example of how syndromes are calculated, turn to the next slide.

### Example

Calculate the syndrome of  $r = (\underbrace{0000\ 0100\ 0101}_{\text{left half}}\ \underbrace{1000\ 1111\ 0001}_{\text{right half}})$  in part (a) of the previous example. Recall:

$$s = r \cdot H = r \cdot \left[ \frac{I_{12}}{B} \right],$$

and observe that the right half of  $r$  has 1s in positions 1, 5, 6, 7, 8, and 12. It follows that  $s$  is equal to:

	left half of $r \rightarrow$	0000 0100 0101
	$\mathbf{b}_1 \rightarrow$	1101 1100 0101
	$\mathbf{b}_5 \rightarrow$	1100 0101 1011
	$\mathbf{b}_6 \rightarrow$	1000 1011 0111
+	$\mathbf{b}_7 \rightarrow$	0001 0110 1111
	$\mathbf{b}_8 \rightarrow$	0010 1101 1101
	$\mathbf{b}_{12} \rightarrow$	1111 1111 1110
	$s \rightarrow$	0101 0010 0000