

# ME424 Project 2 Report

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## 1. Full state space model:

$$\begin{cases} \ddot{z} = \frac{1}{m_c + m_p \sin^2 \theta} [u + m_p \sin \theta (L \dot{\theta}^2 + g \cos \theta)] \\ \ddot{\theta} = \frac{1}{L(m_c + m_p \sin^2 \theta)} [-u \cos \theta - m_p L \dot{\theta}^2 \cos \theta \sin \theta - (m_c + m_p) g \sin \theta] \end{cases}$$

$$\text{where } x = \begin{bmatrix} z \\ \theta \\ \dot{z} \\ \dot{\theta} \end{bmatrix} = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix}$$

$$\begin{aligned} \dot{x} &= \begin{bmatrix} \dot{z} \\ \dot{\theta} \\ \ddot{z} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} \dot{z} \\ \dot{\theta} \\ \frac{1}{m_c + m_p \sin^2 \theta} [u + m_p \sin \theta (L \dot{\theta}^2 + g \cos \theta)] \\ \frac{1}{L(m_c + m_p \sin^2 \theta)} [-u \cos \theta - m_p L \dot{\theta}^2 \cos \theta \sin \theta - (m_c + m_p) g \sin \theta] \end{bmatrix} \\ &= \begin{bmatrix} x_3 \\ x_4 \\ \frac{1}{m_c + m_p \sin^2 x_2} [u + m_p \sin x_2 (L x_4^2 + g \cos x_2)] \\ \frac{1}{L(m_c + m_p \sin^2 x_2)} [-u \cos x_2 - m_p L x_4^2 \cos x_2 \sin x_2 - (m_c + m_p) g \sin x_2] \end{bmatrix} \\ &= \begin{bmatrix} x_3 \\ x_4 \\ \frac{1}{10 + \sin^2 x_2} [u + \sin x_2 (0.5 x_4^2 + 9.81 \cos x_2)] \\ \frac{1}{5 + 0.5 \sin^2 x_2} [-u \cos x_2 - 0.5 x_4^2 \cos x_2 \sin x_2 - 107.91 \sin x_2] \end{bmatrix} = f(x, u) \end{aligned}$$

## 2. Angle dynamic model:

$$\begin{aligned}
 x_\theta &= \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix} = \begin{bmatrix} x_{\theta 1} \\ x_{\theta 2} \end{bmatrix} \\
 \dot{x}_\theta &= \begin{bmatrix} \dot{\theta} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} \dot{x}_{\theta 1} \\ \dot{x}_{\theta 2} \end{bmatrix} = \begin{bmatrix} x_{\theta 2} \\ \frac{1}{L(m_c + m_p \sin^2 x_{\theta 1})} \left[ -u \cos x_{\theta 1} - m_p L x_{\theta 2}^2 \cos x_{\theta 1} \sin x_{\theta 1} - (m_c + m_p) g \sin x_{\theta 1} \right] \end{bmatrix} \\
 &= \begin{bmatrix} x_{\theta 2} \\ \frac{1}{5 + 0.5 \sin^2 x_{\theta 1}} \left[ -u \cos x_{\theta 1} - 0.5 x_{\theta 2}^2 \cos x_{\theta 1} \sin x_{\theta 1} - 107.91 \sin x_{\theta 1} \right] \end{bmatrix} = f_\theta(x_\theta, u)
 \end{aligned}$$

## 3. Linearized model for angle dynamics:

$$\langle \hat{x}_\theta, \hat{u} \rangle = ([\pi, 0]^T, 0),$$

where  $\cos \hat{\theta} = 1, \sin \hat{\theta} = 0$

$$\begin{aligned}
 \dot{x}_\theta &= \begin{bmatrix} \dot{x}_{\theta 1} \\ \dot{x}_{\theta 2} \end{bmatrix} = \begin{bmatrix} f_{\theta 1} \\ f_{\theta 2} \end{bmatrix} = \begin{bmatrix} x_{\theta 2} \\ \frac{1}{L(m_c + m_p \sin^2 x_{\theta 1})} \left[ -u \cos x_{\theta 1} - m_p L x_{\theta 2}^2 \cos x_{\theta 1} \sin x_{\theta 1} - (m_c + m_p) g \sin x_{\theta 1} \right] \end{bmatrix} \\
 \hat{A} &= \left. \frac{\partial f_\theta}{\partial x_\theta} \right|_{\langle \hat{x}_\theta, \hat{u} \rangle = ([\pi, 0]^T, 0)} = \begin{bmatrix} \frac{\partial f_{\theta 1}}{\partial x_{\theta 1}} & \frac{\partial f_{\theta 1}}{\partial x_{\theta 2}} \\ \frac{\partial f_{\theta 2}}{\partial x_{\theta 1}} & \frac{\partial f_{\theta 2}}{\partial x_{\theta 2}} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{10791}{500} & 0 \end{bmatrix} \\
 \hat{B} &= \left. \frac{\partial f_\theta}{\partial u} \right|_{\langle \hat{x}_\theta, \hat{u} \rangle = ([\pi, 0]^T, 0)} = \begin{bmatrix} \frac{\partial f_{\theta 1}}{\partial u} \\ \frac{\partial f_{\theta 2}}{\partial u} \end{bmatrix} = \begin{bmatrix} 0 \\ -\frac{1}{5} \end{bmatrix} \\
 \dot{x}_\theta &= \begin{bmatrix} 0 & 1 \\ -\frac{10791}{500} & 0 \end{bmatrix} \begin{bmatrix} x_{\theta 1} \\ x_{\theta 2} \end{bmatrix} + \begin{bmatrix} 0 \\ -\frac{1}{5} \end{bmatrix} u + \begin{bmatrix} 0 \\ -\frac{10791\pi}{500} \end{bmatrix}
 \end{aligned}$$

## 4. Discrete Time Angel Dynamics Model:

$$\begin{aligned}
 x_\theta[k+1] &= x_\theta(t+T) = x_\theta(t) + \dot{x}_\theta(t)T \\
 &= \begin{bmatrix} x_{\theta 1} \\ x_{\theta 2} \end{bmatrix} + 0.005 \times \left[ \hat{A}x_\theta + \hat{B}u + f(\hat{x}, \hat{u}) \right] \\
 &= I \begin{bmatrix} x_{\theta 1} \\ x_{\theta 2} \end{bmatrix} + 0.005 \times \begin{bmatrix} 0 & 1 \\ \frac{10791}{500} & 0 \end{bmatrix} \begin{bmatrix} x_{\theta 1} \\ x_{\theta 2} \end{bmatrix} + 0.005 \times \begin{bmatrix} 0 \\ \frac{1}{5} \end{bmatrix} u + \begin{bmatrix} 0 \\ -0.10791\pi \end{bmatrix} \\
 &= \begin{bmatrix} 1 & 0.005 \\ 0.10791 & 1 \end{bmatrix} \begin{bmatrix} x_{\theta 1} \\ x_{\theta 2} \end{bmatrix} + \begin{bmatrix} 0 \\ 0.001 \end{bmatrix} u + \begin{bmatrix} 0 \\ -0.339 \end{bmatrix} \\
 y(k) &= \theta_k = \begin{bmatrix} 1 & 0 \end{bmatrix} x_\theta(k) \\
 A &= \begin{bmatrix} 1 & 0.005 \\ 0.10791 & 1 \end{bmatrix} \\
 B &= \begin{bmatrix} 0 \\ 0.001 \end{bmatrix} \\
 C &= \begin{bmatrix} 1 & 0 \end{bmatrix}
 \end{aligned}$$

## 5. Drake Simulation Setup and Testing

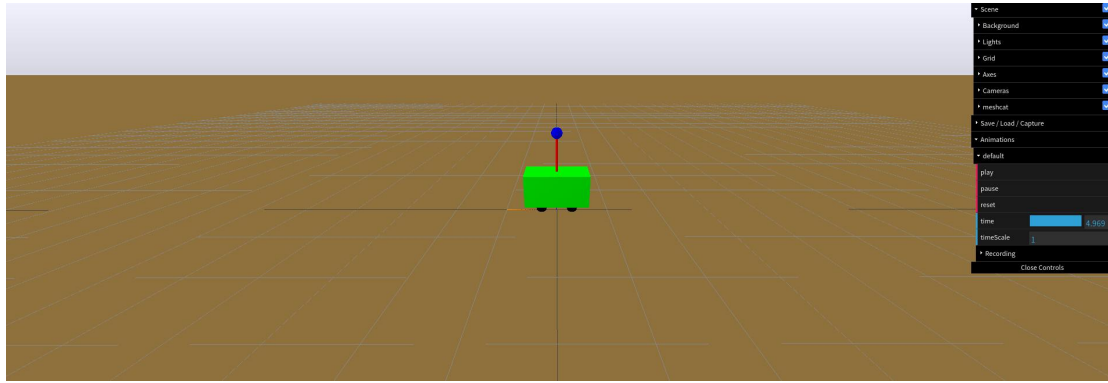
Initial state vectors are shown as below:

```
# start simulation
UprightState = np.array([0, np.pi, 0, 0]) # the state of the cart-pole is organized as [z, theta, zdot, theta_dot]
UprightState = np.array([0, np.pi/2, 0, 0]) # the state of the cart-pole is organized as [z, theta, zdot, theta_dot]
UprightState = np.array([0, np.pi/3, 0, 0]) # the state of the cart-pole is organized as [z, theta, zdot, theta_dot]
```

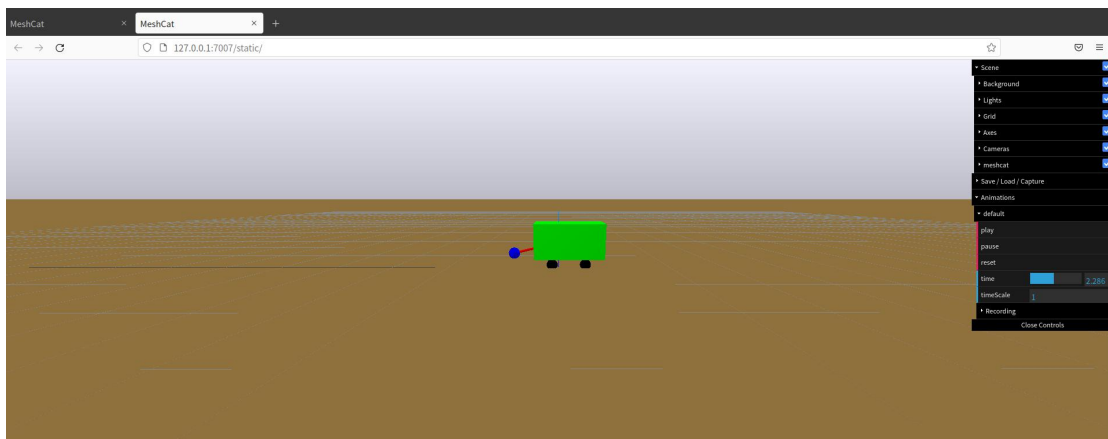
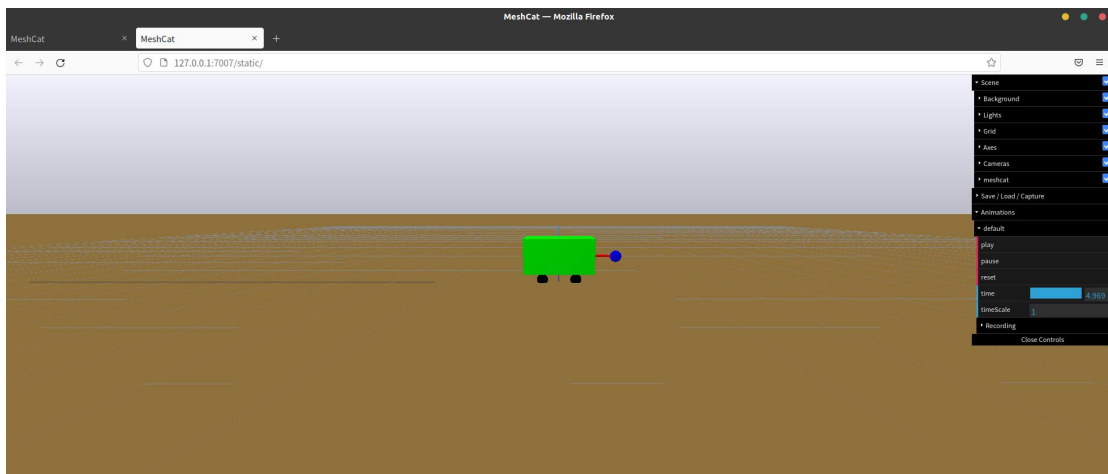
All snapshots shown as below are the simulations with no disturbance.

```
context = simulator.get_mutable_context()
# context.SetContinuousState(UprightState + np.array([0.1, 0.3, 0.3, 0.1])) # 有扰动偏移 disturbance
context.SetContinuousState(UprightState) # 无扰动偏移 No disturbance
```

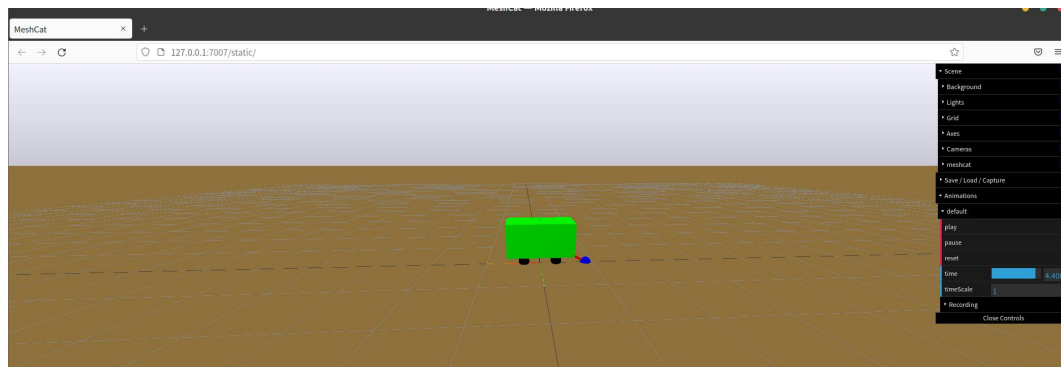
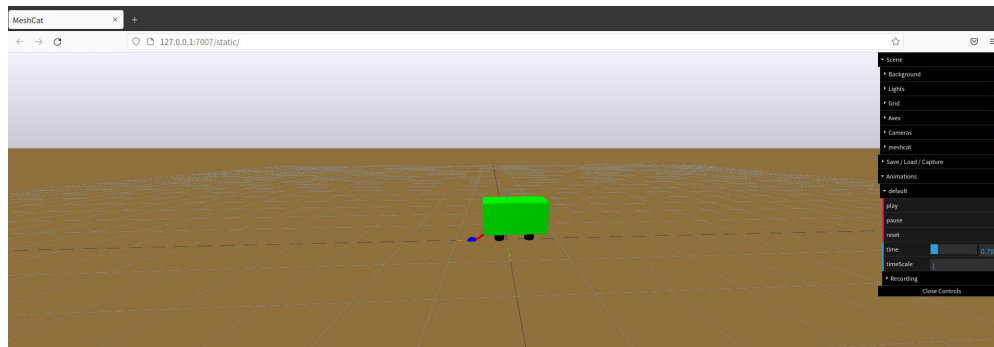
$$\theta_{init} = \pi$$



$\theta_{init} = \pi/2$  ([GIF file\(q5\\_0.5pi.gif\)](#) is attached in the document)



$\theta_{init} = \pi/3$  ([GIF file\(q5\\_0.33pi.gif\)](#) is attached in the document)



## 6. Simulation studies for state-feedback control

### (a) Eigenvalue assignment

To find  $K$  such that  $A-BK$  has eigenvalue  $z_{1,2} = e^{s_{1,2} * T}$ , where  $s_{1,2} = -2 \pm j$ , we should:

The first step:

converting  $A$ ,  $B$  matrices to a python code form:

The matrix shown as below is  $f(x,u)$

```
vector_field
✓ 0.0 Python
```

$$\begin{bmatrix} \theta_d \\ -1.9[m_y \sin(\theta) \cos(\theta) - l_f \cos(\theta) - g(m_y + m_b) \sin(\theta)] \\ 1.9[m_y \sin^2(\theta)] \end{bmatrix}$$

The second step:

Calculating  $A$  and  $B$  matrix based on Part 3.

```

pfx = vector_field.jacobian([theta, theta_d])
pfpu = vector_field.jacobian([fx])

# g = 9.81
# mc = 10
# mp = 1
# l = 0.5

pfx_f = sym.lambdify((q, fx, mc, mp, l, theta, theta_d), pfx)

pfpu_f = sym.lambdify((q, fx, mc, mp, l, theta, theta_d), pfpu)
Ahat = pfx_f([9.81, 0, 10, 1, 0.5, np.pi, 0])
Bhat = pfpu_f([9.81, 0, 10, 1, 0.5, np.pi, 0])

✓ 0.3s Python

Ahat
✓ 0.3s Python
array([[ 0. ,  1. ],
       [21.982, -0. ]])

Bhat
✓ 0.0s Python
array([[0. 1.],
       [0. 2.]])

```

The third step:

Computing the matrix A and the matrix B in Part 4, the form of discrete time angel dynamics model based on the continuous-time linearized model in Part 3.

```

import scipy.signal as sig
I = np.eye(2)
A = I + 0.005*Ahat
B = 0.005*Bhat
print(A)
print(B)

✓ 0.3s Python
[[1.  0.005 ]
 [0.10791 1.  ]]
[[0.  ]
 [0.001]]

```

$$A = \begin{bmatrix} 1 & 0.005 \\ 0.10791 & 1 \end{bmatrix}$$

$$B = \begin{bmatrix} 0 \\ 0.001 \end{bmatrix}$$

The fourth step:

Applying the function `sig.place_poles(matrix A, matrix B, z_desired).gain_matrix` to find gain K, which leads to the desired eigenvalues of A-BK.

```

z_desired = np.array([2+1j, -2-1j])
z_desired = np.exp(z_desired*0.005)
cart_pole_gain = sig.place_poles(A, B, z_desired).gain_matrix
print(cart_pole_gain)

✓ 0.0s Python
[[132.06140054 19.9250837 ]]

```

## (b) Closed-loop simulation

After calculating K, we should test the control  $u = -kx_\theta$  is feasible or not, or the efficiency of adjustment to stable.

```

class myController(LeafSystem):
    def __init__(self, K):
        LeafSystem.__init__(self)
        self.DeclareVectorInputPort("u", BasicVector(4))
        self.DeclareVectorOutputPort("y", BasicVector(1), self.CalcOutputY)
        self.K = K
    def CalcOutputY(self, context, output):
        statex = self.get_input_port(0).Eval(context)
        y = -np.dot(self.K, (statex-np.array([0, np.pi, 0, 0])))
        # print(statex, y, statex-np.array([0, np.pi, 0, 0]))
        output.SetFromVector([y])

```

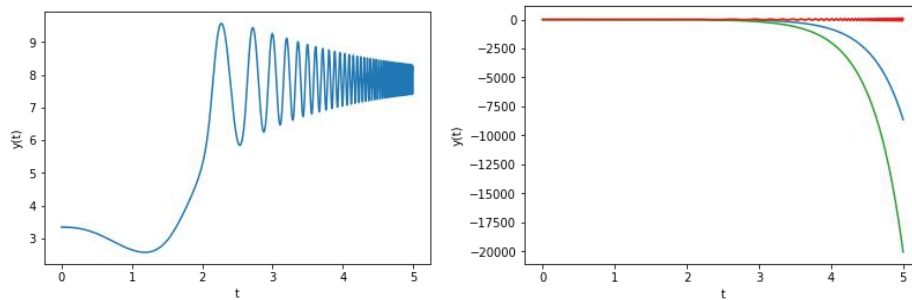
Since  $K_1$  and  $K_3$  is meant for  $z$ , and factor of  $z$  could make a difference to  $\theta$ , so we prepared 2 sets of K with different  $K_1$  and  $K_3$ .

(i) *eigenvalues* =  $-2 \pm j$ ,

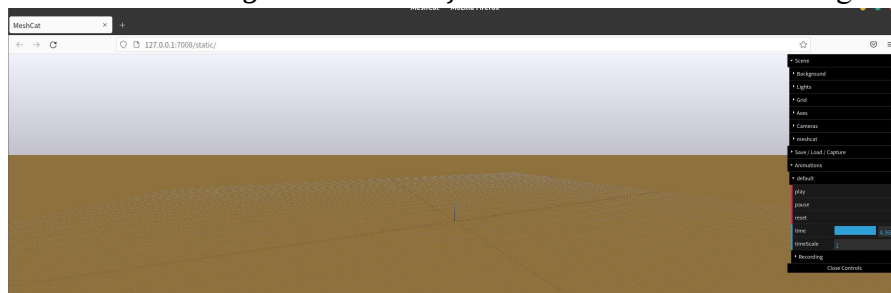
$K = [-15.29051988 \text{ } 132.66140054 \text{ } -18.85830785 \text{ } 19.9250837]$

```
builder = DiagramBuilder()
# K = np.array([-15.29051988, 220.55525994, -18.85830785, 44.42915392])
K = np.array([-15.29051988, 132.66140054, -18.85830785, 19.9250837]) # Question7 -2-j -2+j
# K = np.array([-15.29051988, 444.48486861, -18.85830785, 21.48420104]) # Question7 -2-8j -2+8j
# K = np.array([0, 132.66140054, -6, 19.9250837]) # Question7 -2-j -2+j]
#K = np.array([0, 444.48486861, -6, 21.48420104]) # Question7 -2-8j -2+8j
```

The diagram is shown as below:



We can see that  $\theta$  converges to a value far more than  $\pi$  and  $z$  diverges.



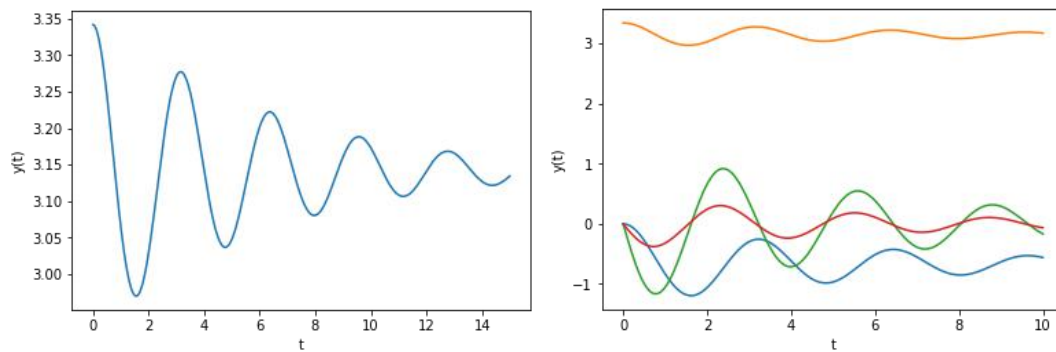
The cart is lost in the simulation, which indicates the failure (seen in [q6b\\_1.gif](#)). We can see that under such conditions, the cart pole is unstable. To eliminate the effect of other factor we used another set of  $K$ .

(ii) *eigenvalues* =  $-2 \pm j$ ,

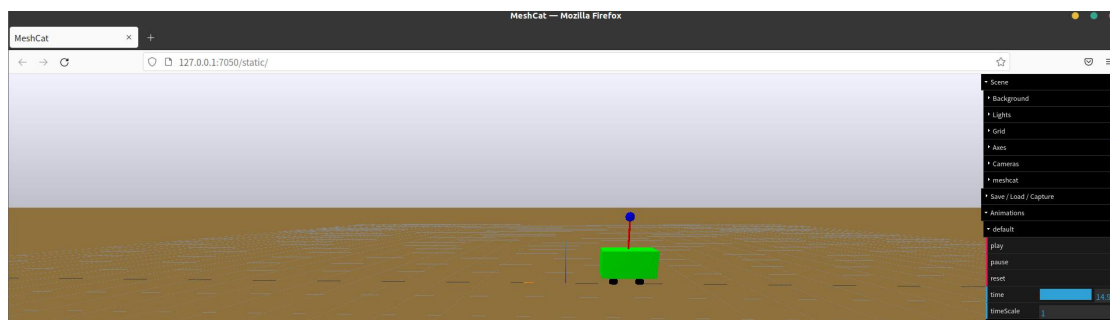
$K = [0 \text{ } 132.66140054 \text{ } -6 \text{ } 19.9250837]$

```
builder = DiagramBuilder()
# K = np.array([-15.29051988, 220.55525994, -18.85830785, 44.42915392])
# K = np.array([-15.29051988, 132.66140054, -18.85830785, 19.9250837]) # Question7 -2-j -2+j
# K = np.array([-15.29051988, 444.48486861, -18.85830785, 21.48420104]) # Question7 -2-8j -2+8j
K = np.array([0, 132.66140054, -6, 19.9250837]) # Question7 -2-j -2+j
#K = np.array([0, 444.48486861, -6, 21.48420104]) # Question7 -2-8j -2+8j
#K = np.array([-0. , 124.96 , -6.13149847, 17.06574924])
```

The diagram is shown as below:



We can see that  $\theta$  and  $z$  both converge, and  $\theta$  converges to  $\pi$



[q6b\\_2.gif](#) is attached

## 7. Repeat with different eigenvalues

First calculate  $K$ :

```

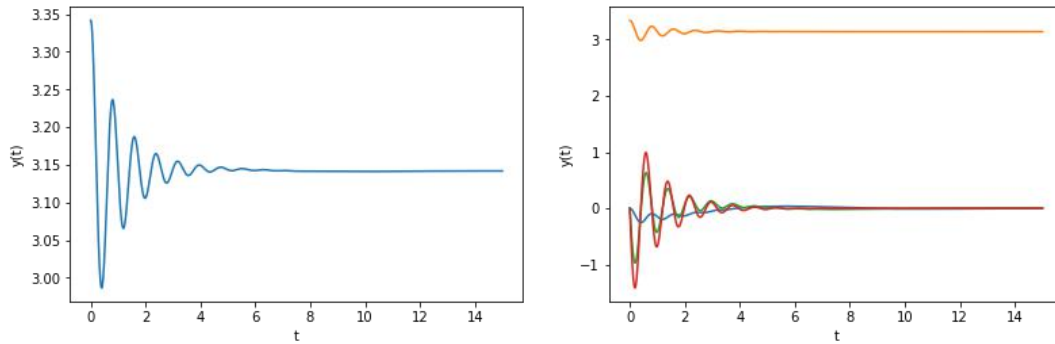
> s_desired = np.array([-2+8j, -2-8j])
# s_desired = np.array([-2+1j, -2-1j])
z_desired = np.exp(s_desired*0.005)
cart_pole_gain = sig.place_poles(A, B, z_desired).gain_matrix
print(cart_pole_gain)
[38] ✓ 0.3s
... [[444.48486861 21.48420104]]

```

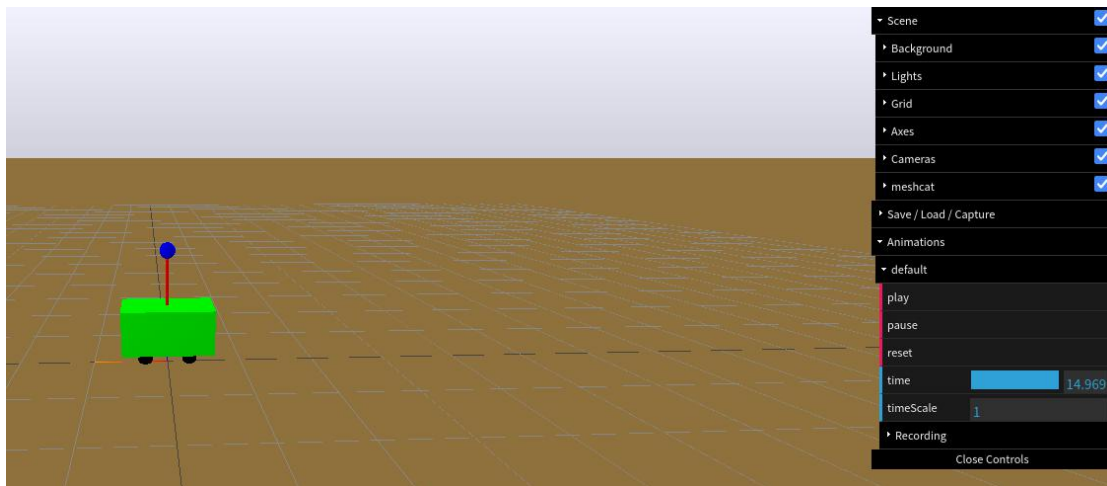
(i) **eigenvalues** =  $-2 \pm 8j$ ,

$K = [ -15.29051988 \quad 444.48486861 \quad -18.85830785 \quad 21.48420104 ]$





Both converges



[q7b\\_1.gif](#) is attached

### Analysis:

Compared with

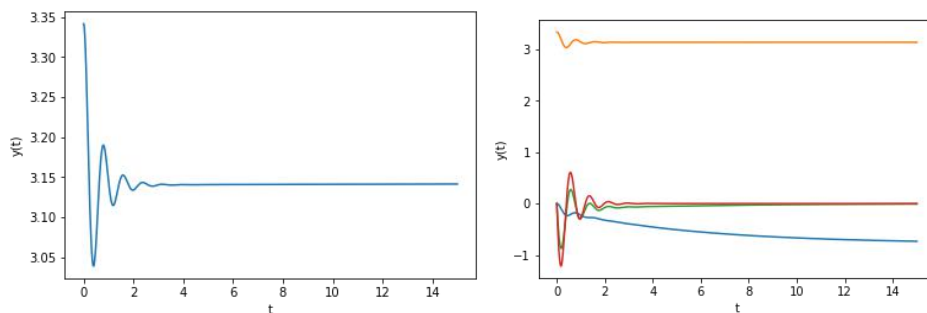
$K = \begin{bmatrix} -15.29051988 & 132.66140054 & -18.85830785 & 19.9250837 \end{bmatrix}$  of  
eigenvalues  $= -2 \pm j$

$K = \begin{bmatrix} -15.29051988 & 444.48486861 & -18.85830785 & 21.48420104 \end{bmatrix}$  of  
eigenvalues  $= -2 \pm 8j$  takes shorter time to converge to  $\pi$  and be in a stable state,  
which is around 5 second, whose respond is faster.

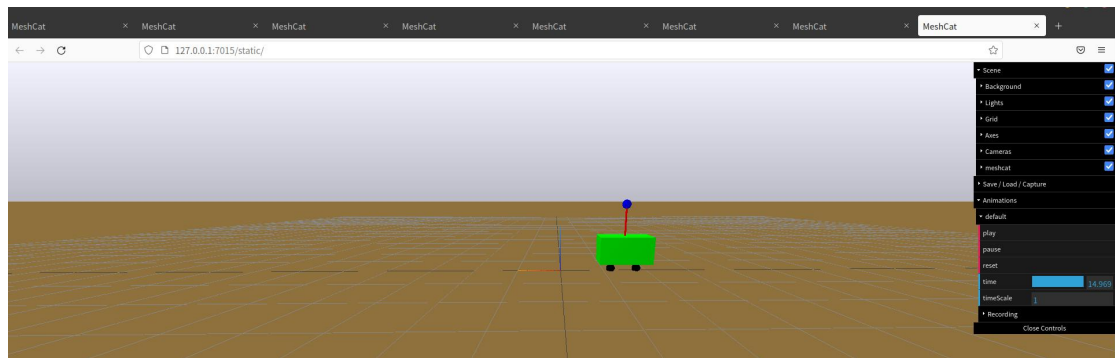
But for the previous one, it doesn't show a typically control plotting, and it also  
converges to a value much larger than  $\pi$ . And its 's converge time is much more than  
5 seconds.

(ii) **eigenvalues**  $= -2 \pm 8j$ ,

$K = \begin{bmatrix} 0 & 444.48486861 & -6 & 21.48420104 \end{bmatrix}$



Both converges



[q7b\\_2.gif](#) is attached

### Analysis:

Compared with

$K = [0 \ 132.66140054 \ -6 \ 19.9250837]$  of eigenvalues  $= -2 \pm j$

$K = [0 \ 444.48486861 \ -6 \ 21.48420104]$  of eigenvalues  $= -2 \pm 8j$  takes shorter time to converge to  $\pi$  and be in a stable state, where the previous one is more than 15 seconds, where the latter is around 2 second, whose response is faster.

## 8. Output feedback control design:

The code of observer is shown as below:

The first one is to compute L of the specific eigenvalues  $-9 \pm 2j$

```
# Calculate L
import scipy.signal as sig
A = np.array([[1, 0.005], [0.10791, 1]])
B = np.array([[0], [0.001]])
C = np.array([[1, 0]])
s_desired = np.array([-9+2j, -9-2j])
z_desired = np.exp(s_desired*0.005)
L = sig.place_poles(A.T, C.T, z_desired).gain_matrix
L = L.T
print(L)

✓ 0.3s

[[0.0881064]
 [0.51427411]]
```

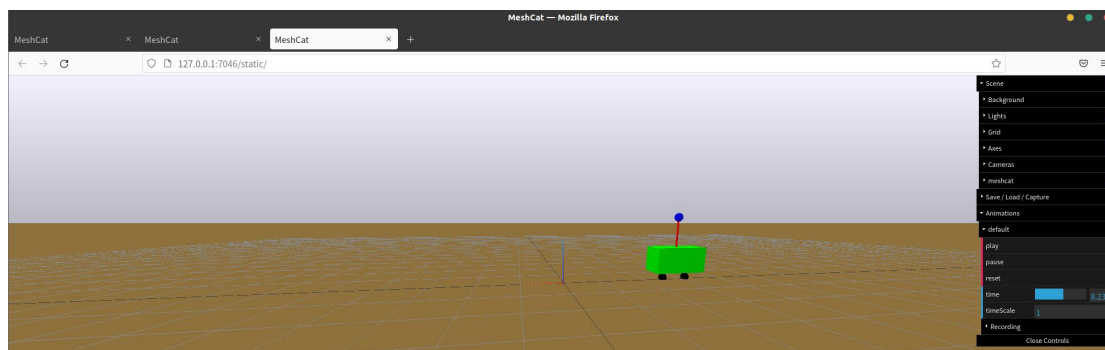
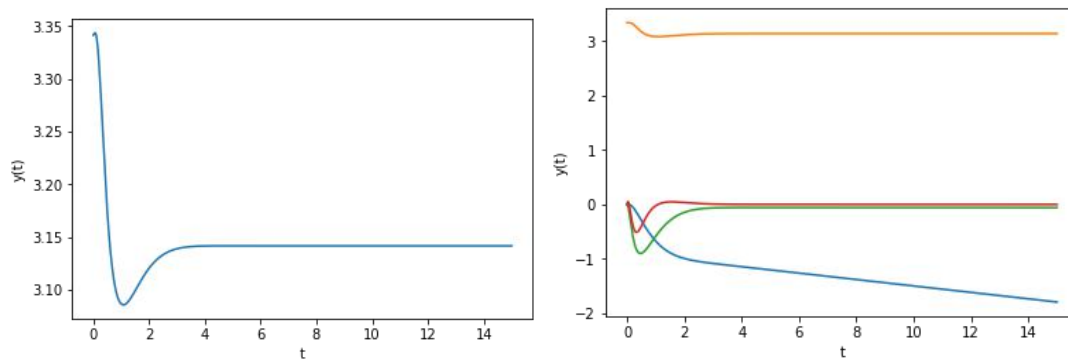
The second one is to set up the model of DT Luenberger observer

```
# Define the system.
class DTObserver(LeafSystem):
    def __init__(self, A, B, C, L, T):
        LeafSystem.__init__(self)
        self.DeclareDiscreteState(2)
        # Define the input
        self.DeclareVectorInputPort("uk", BasicVector(1))
        self.DeclareVectorInputPort("yk", BasicVector(4))
        # Define the output
        self.DeclareVectorOutputPort("x_estimated", BasicVector(2), self.CalcOutputY, set([self.all_state_ticket()]))
        self.DeclarePeriodicDiscreteUpdate(T) # One second timestep.
        self.A = A
        self.B = B
        self.C = C
        self.L = L

    def DoCalcDiscreteVariableUpdates(self, context, events, discrete_state):
        xk = context.get_discrete_state_vector().CopyToVector()
        # xk1 = np.array([[xk[1]], [xk[3]]])
        # print(xk1)
        uk = self.get_input_port(0).Eval(context)
        yk = self.get_input_port(1).Eval(context)
        yk1 = yk[1] - np.pi
        A = self.A
        B = self.B
        C = self.C
        L = self.L
        # print(A.dot(xk1))
        # xk1 = np.mat(xk1.reshape((2,1)))
        xnext = A.dot(xk) + B.dot(uk) + L.dot(yk1 - C.dot(xk))
        # xnext1 = np.array([[xk[0]], [xnext[0]], [xk[2]], [xnext[1]]])
        discrete_state.get_mutable_vector().SetFromVector(xnext)

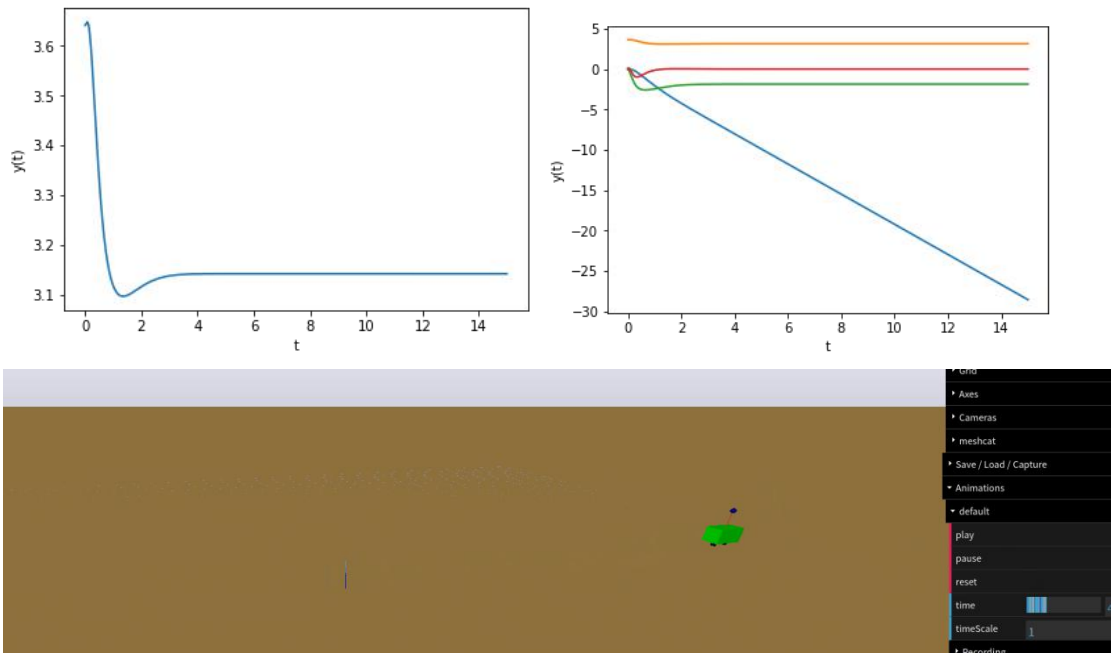
    def CalcOutputY(self, context, output):
        x = context.get_discrete_state_vector().CopyToVector()
        y = x
        output.SetFromVector(y)
```

**Disturbance = 0.2**



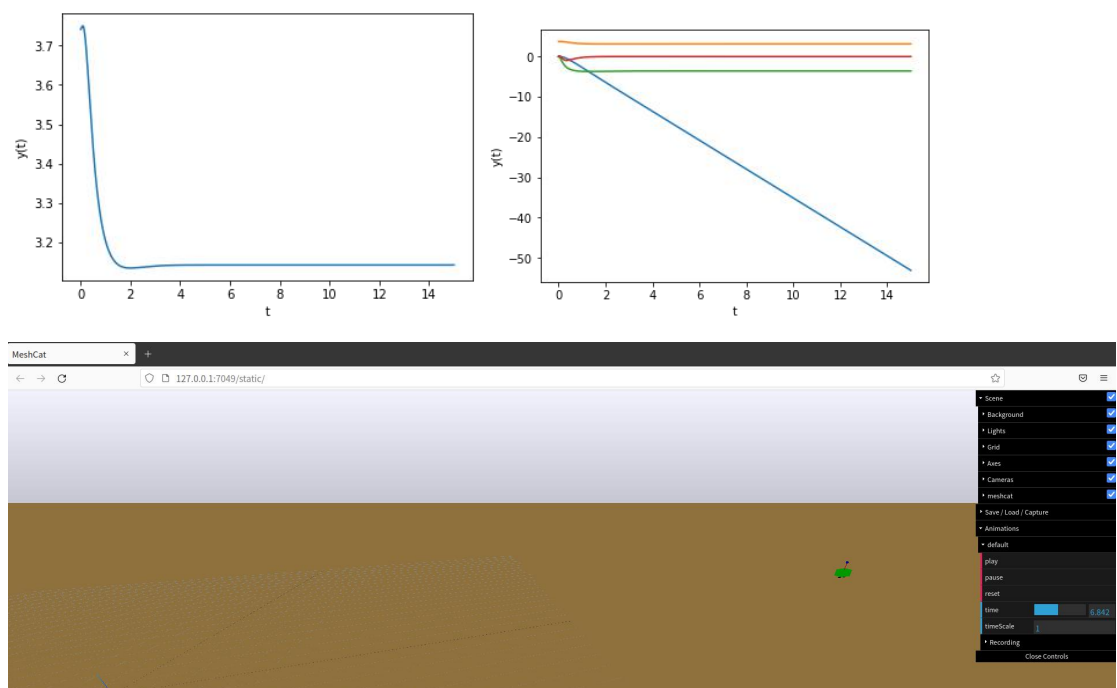
[q81.gif](#)

**Disturbance = 0.5**



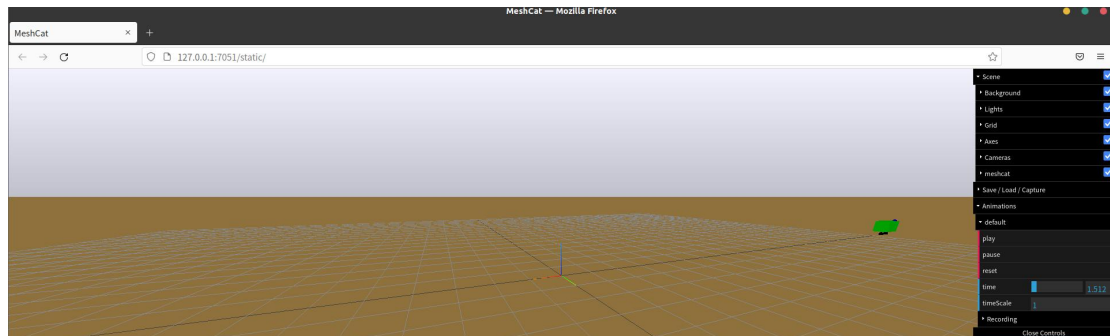
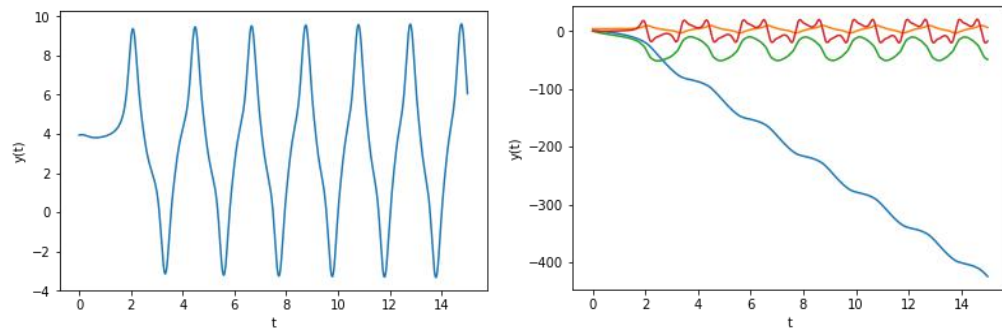
[q82.gif](#)

**Disturbance = 0.6**



[q83.gif](#)

**Disturbance = 0.8**



[q84.gif](http://q84.gif)

We find out that when the initial conditions is in a specific range, the ange of the pole will remain in a stable and desired angle. If the disturbance exceed the range, like the diagrams above when disturbance = 0.8, it may oscillate in a period, and also for z, it will diverge.