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DELAY CHAIN AND METHOD OF OPERATING SAME

Abstract

A delay chain includes: delay stages coupled in series, each delay stage including a driver device coupled between an input node and an output node of the delay stage, and a capacitor digital-to-analog converter (CDAC) coupled to the output node of the delay stage; at least one of the CDACs being a separately adjustable (SA) type of CDAC (SA_CDAC) that is separately adjustable relative to at least another one of the CDACs; and a controller configured to adjust the at least one CDAC.

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Background/Summary

PRIORITY CLAIM [0001] This application claims the priority of U.S. Provisional Application No. 63/551,883, filed Feb. 9, 2024, which is incorporated herein by reference in its entirety.

BACKGROUND

[0002] The semiconductor integrated circuit (IC) industry produces a wide variety of analog and digital devices to address issues in a number of different areas. Developments in semiconductor process technology nodes have progressively reduced component sizes and tightened spacing resulting in progressively increased transistor density. ICs have become smaller.

Description

BRIEF DESCRIPTION OF THE DRAWINGS

[0003] One or more embodiments are illustrated by way of example, and not by limitation, in the figures of the accompanying drawings, wherein elements having the same reference numeral designations represent like elements throughout. The drawings are not to scale, unless otherwise disclosed.

[0004] FIGS. 1A-ID are corresponding schematic diagrams, in accordance with some embodiments.

[0005] FIGS. 2A-2E are corresponding schematic diagrams, in accordance with some embodiments.

[0006] FIG. 3 is a table, in accordance with some embodiments.

[0007] FIGS. 4A-4G are corresponding schematic diagrams, in accordance with some embodiments.

[0008] FIG. 5 is a table, in accordance with some embodiments.

[0009] FIGS. 6-7 are flowcharts of corresponding methods, in accordance with some embodiments.

[0010] FIG. 8 is a block diagram of an electronic design automation (EDA) system in accordance with some embodiments.

[0011] FIG. 9 is a block diagram of an integrated circuit (IC) manufacturing system, and an IC manufacturing flow associated therewith, in accordance with some embodiments.

DETAILED DESCRIPTION

[0012] The following disclosure discloses many different embodiments, or examples, for implementing different features of the subject matter. Examples of components, materials, values, steps, operations, arrangements, or the like, are described below to simplify the present disclosure. These are, of course, merely examples and are not intended to be limiting. Other components, values, operations, materials, arrangements, or the like, are contemplated. For example, the formation of a first feature over or on a second feature in the description that follows include embodiments in which the first and second features are formed in direct contact, and further include embodiments in which additional features are formed between the first and second features, such that the first and second features are in indirect contact. In addition, the present disclosure repeats reference numerals and/or letters in the various examples. This repetition is for the purpose of simplicity and clarity and does not in itself dictate a relationship between the various embodiments and/or configurations discussed.

[0013] Further, spatially relative terms, such as “beneath,” “below,” “lower,” “above,” “upper” and the like, are used herein for case of description to describe one element or feature's relationship to another element(s) or feature(s) as illustrated in the figures. The spatially relative terms are intended to encompass different orientations of the device in use or operation in addition to the orientation depicted in the figures. The apparatus is otherwise oriented (rotated 90 degrees or at other orientations) and the spatially relative descriptors used herein are likewise interpreted accordingly. In some embodiments, the term standard cell structure refers to a standardized

building block included in a library of various standard cell structures. In some embodiments, various standard cell structures are selected from a library thereof and are used as components in a layout diagram representing a circuit.

[0014] In some embodiments, a delay chain includes: delay stages coupled in series, each delay stage including a driver device coupled between an input node and an output node of the delay stage, and a capacitor digital-to-analog converter (CDAC) coupled to the output node of the delay stage; at least one of the CDACs being a separately adjustable (SA) type of CDAC (SA_CDAC) that is separately adjustable relative to at least another one of the CDACs; and a controller configured to adjust the at least one CDAC.

[0015] In some embodiments, the at least one SA_CDAC is a one-dimensional array of separately couplable capacitance (SCC) cells (SCC_cells). In some embodiments, 'couplable' is understood to mean capable of being coupled to something, e.g., a first SCC_cell of a given SA_CDAC is capable of being coupled to the output node of the given SA_CDAC. In some embodiments, for the first SCC_cell and a second SCC_cell of the given SA_CDAC, 'separately couplable' is understood to mean that coupling/decoupling the first SCC_cell to/from the output node of the given SA_CDAC is capable of being performed separately (or independently) from coupling/decoupling the second SCC_cell to/from the output node of the given SA_CDAC. Accordingly, in such embodiments, each of the first and second SCC_cells is described as being separately (or independently) couplable to the output node of the given SA_CDAC. In such embodiments, the one-dimensional arrays include rows and a single column, and each row includes a single SCC_cell; and the controller is further configured to selectively couple none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage. In some embodiments, the at least one SA_CDAC is a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells). In some embodiments, describing an SCC_cell as being controllable is understood to mean that the coupling between the SA_CDAC and the output node of the given SA_CDAC is selectively makeable or breakable. In some embodiments, for a first SCC_cell and a second SCC_cell of a given SA_CDAC, 'separately controllable' is understood to mean that coupling/decoupling the first SCC_cell to/from the output node of the given SA_CDAC is capable of being performed separately (or independently) from coupling/decoupling the second SCC_cell to/from the output node of the given SA_CDAC.

[0016] In combination with an even-row decoder (see FIGS. 2B and 2D, 4B and 4D, or the like) and an odd-row decoder (see FIGS. 2C and 4C, 2E and 4E, or the like) correspondingly included in each of the SCC_cells, the use of row-address signals RW(i) and column-addressing signals CL(i) to form a thermometer code reduces the total number of control lines needed for the SA_CDACs as compared to a CDAC-based delay chain according to another approach, where the CDACs of the other approach are not separately adjustable. The reduced total number of control lines for the SA_CDACs results in the corresponding delay chain occupying a smaller area (having a smaller footprint) as compared to the CDAC-based delay chain according to another approach.

[0017] FIG. 1A is a schematic diagram of a delay chain **102**, in accordance with some embodiments.

[0018] Delay chain **100** has an input node nd_IN and an output node nd_OUT. Delay chain **100** includes delay stages ds(0), . . . , ds(N-2) and ds(N-1) coupled in series between input node nd_IN and output node nd_OUT, where N is an integer and $2 \leq N$.

[0019] Taking delay stage ds(j) as being representative of each of delay stages ds(0)-ds(N-1), delay stage ds(j) includes: a driver device drv(j) coupled between an input node in(j) and an output node out(j) of the delay stage ds(j); and a capacitor digital-to-analog converter (CDAC), and more particularly a separately adjustable (SA) type of CDAC(j) (SA_CDAC(j)), coupled to output node out(j) of delay stage ds(j). In some embodiments, describing SA_CDAC(j) as being separately adjustable is understood as SA_CDAC(j) being separately adjustable relative to at least another one of the CDACs.

[0020] In FIG. 1A, each CDAC is an SA_CDAC. In some embodiments, at least one but fewer than all of SA_CDAC(0)-SA_CDAC(N-1) is a type of CDAC other than an SA_CDAC.

[0021] Each of SA_CDAC(0)-SA_CDAC(N-1) is a one-dimensional $M \times 1$ array of separately couplable capacitance (SCC) cells (SCC_cells) (see FIG. 2B-2C, 4B-4C, or the like). Each of SA_CDAC(0)-SA_CDAC(N-1) is configured and controlled to be a monotonic CDAC.

[0022] In FIG. 1A, taking delay stage ds(j) as being representative of each of delay stages ds(0)-ds(N-1), driver device drv(j) includes an inverter INV coupled between input node in(j) and output node out(j) of delay stage ds(j). In some embodiments, driver device drv(j) includes a buffer (not shown) coupled between input node in(j) and output node out(j) of delay stage ds(j).

[0023] In FIG. 1A, a signal provided to input node nd_IN of delay chain 102 experiences an overall propagation delay as the signal propagates to output node nd_OUT of delay chain 102. The overall propagation delay is the sum of incremental propagation delays introduced as the signal correspondingly propagates through delay stages ds(0)-ds(N-1) of delay chain 102. Where SA_CDAC(j) is representative of each of SA_CDAC(0)-SA_CDAC(N-1), SA_CDAC(j) represents a controllable capacitive load on output node out(j) of delay stage ds(j). Adjusting the capacitive load represented by SA_CDAC(j) has an effect of adjusting the incremental propagation delay introduced by corresponding delay stage ds(j). In some embodiments, separately adjusting the capacitive loads correspondingly represented by SA_CDAC(j)-SA_CDAC(N-1) is referred to as tuning the overall propagation delay of delay chain 102. As each of SA_CDAC(0)-SA_CDAC(N-1) is monotonic, delay chain 102 facilitates monotonic delay tuning.

[0024] In FIG. 1A, SA_CDAC(0)-SA_CDAC(N-1) are controlled via addressing by signals CL(0)-CL(N-1) and RW(0)-RW(M-1). A controller (112 FIG. 1B) is configured to generate signals CL(0)-CL(N-1) and RW(0)-RW(M-1) based on a digital thermometer encoding scheme (see FIG. 2A-2E, 3, 4A-4E, 5, or the like). Accordingly, delay chain 102 is described as a digitally controlled type of delay chain.

[0025] In general, CDACs, e.g., SA_CDAC(0)-SA_CDAC(N-1), facilitate clock synchronization, communication in general such as circuit-to-circuit communication (e.g., in a compute-in-memory (CIM) system), digital-to-analog conversion (DAC), analog-to-digital conversion (ADC), waveform generation, audio applications, or the like. As delay chain 102 facilitates monotonic delay tuning, delay chain 102 facilitates, e.g., low jitter clock generation for a high speed serializer/deserializer (SERDES) device, a die-to-die input/output (I/O) system, or the like.

[0026] FIG. 1B is a schematic diagram of an array 104, in accordance with some embodiments.

[0027] Array 104 includes: SCC_cells 106(0,0)-106(0,N-1); SCC_cells 108(1,0)-108(1,N-1); and SCC_cells 106(M-1,0)-106(M-1,N-1). In FIG. 1B, the SCC_cells are treated as black boxes; see FIGS. 2B-2C, 4B-4C, or the like, for a discussion of internal configurations of the SCC_cells.

[0028] Together, SCC_cells 106(0,0), 108(1,0) and 106(M-1,0) represent SA_CDAC(0) of delay stage ds(0). Together, SCC_cells 106(0,N-2), 108(1,N-2) and 106(M-1,N-2) represent SA_CDAC(1) of delay stage ds(N-2). Together, SCC_cells 106(0,N-1), 108(1,N-1) and 106(M-1,N-1) represent SA_CDAC(N-1) of delay stage ds(N-1).

[0029] To facilitate coordinated control of delay line 102, the N instances of one-dimensional $M \times 1$ arrays of SCC_cells in FIG. 1A, i.e., SA_CDAC(0)-SA_CDAC(N-1), are treated as a whole according to some embodiments. Treated as a whole, SA_CDAC(0)-SA_CDAC(N-1) together represent an $M \times N$ array of SCC_cells. SA_CDAC(0)-SA_CDAC(N-1) correspondingly represent columns col_0-col_N-1 of array 104.

[0030] Array 104 receives: column-addressing signals CL(0)-CL(N-1) corresponding to columns col(0)-col(N-1) in array 104; row-addressing signals RW(0)-RW(M-1) corresponding to rows rw_0-rw_M-1 and row-addressing signal RW(M) corresponding to an imaginary row rw_M. Signals CL(0)-CL(N-1) and RW(0)-RW(M-1) are generated by a controller 112. Controller 112 is further configured to generate signals CL(0)-CL(N-1) and RW(0)-RW(M-1) based on a digital thermometer encoding scheme (see FIG. 2A-2E, 3, 4A-4E, 5, or the like) and thereby selective

couple some, none or all of the SCC_cells correspondingly in SA_CDAC(0)-SA_CDAC(N-1) to corresponding nodes out(0)-out(N-1) in delay stages ds(0)-ds(N-1).

[0031] In array **104**, even-numbered rows include SCC_cells **106**(*i,j*), where each of *i* and *j* are corresponding non-negative integers, $0 \leq i \leq (M-1)$ and $0 \leq j \leq (N-1)$. Odd-numbered rows include SCC_cells **108**(*i,j*), where *i* is an odd number. Even row *rw_0* of array **104** includes SCC_cells **106**(0,0)-**106**(0,N-1). Odd row *rw_1* of array **104** includes SCC_cells **108**(1,0)-**108**(1,N-1). Row *rw_M-1* of array **104** includes SCC_cells **106**(M-1,0)-**106**(M-1,N-1).

[0032] Each of SCC_cells **106**(*i,j*) and **108**(*i,j*) has three inputs and one output. Each of SCC_cells **106**(*i*,0)-**106**(*i*,N-2) and **108**(*i*,0)-**108**(*i*,N-2) is configured to receive three different signals correspondingly on the three inputs (see FIG. 2D, 4D, or the like). By contrast, the SCC_cells in the last column *col_N-1* receive fewer signals. each of SCC_cells **106**(*i*,N-1) and **108**(*i*,N-1) is configured to receive two different signals on three inputs, i.e., two of the inputs are configured to receive the same signal (see FIG. 2E, 4E, or the like).

[0033] In FIG. 1B, the SCC_cells in the even rows have the default SCC-cell configuration and the SCC_cells in the odd rows have a different configuration as compared to the SCC_cells in the even rows. As such, the SCC_cells in rows *rw_0*, *rw_2*, etc., are different than the SCC_cells in rows *rw_1*, *rw_3*, etc. In some embodiments, the SCC_cells in array **104** are described as having row-number-parity-specific configurations. The use of row-number-parity-specific configurations for the SCC_cells in array **104** facilitates gray-code-compliant adjustment of SA_CDAC(0)-SA_CDAC(N-1), i.e., of the SCC_cells in array **104**, and gray-code-compliant adjustment facilitates monotonic adjustment of the same.

[0034] In array **104**, each of the SCC_cells in the even rows include the same components. Each of the SCC_cells in the odd rows include the same components of the SCC_cells in the even rows however one of the inputs in each of the SCC_cells in each of the even rows is inverted as indicated by prefixed inversion bubble **110**. In some embodiments, each inverted input is described as being an active low input. Regarding FIG. 1B, in some embodiments, the SCC_cells in the odd rows have the default SCC-cell configuration and the SCC_cells in the even rows have the different configuration as compared to the SCC_cells in the odd rows which represents the converse of the even versus odd rows SCC_cell configuration relationship shown in FIG. 1B.

[0035] In general, the output of each of SCC_cells **106**(*i,j*) and **108**(*i,j*) in column *col(j)* is coupled to output node out(*j*) of the delay stage ds(*j*). The output of each of SCC_cells **106**(0,0), **108**(1,0) and **106**(M-1,0) in column *col*(0) is coupled to output node out(0) of the delay stage ds(0). The output of each of SCC_cells **106**(0,N-2), **108**(1,N-2) and **106**(M-1,N-2) in column *col*(N-2) is coupled to output node out(N-2) of the delay stage ds(N-2). The output of each of SCC_cells **106**(0,N-1), **108**(1,N-1) and **106**(M-1,N-1) in column *col*(N-1) is coupled to output node out(N-1) of the delay stage ds(N-1).

[0036] One input of each of SCC_cells **106**(0,0), **108**(1,0) and **106**(M-1,0) in column *col*(0) is configured to receive column-addressing signal CL(0). One input of each of SCC_cells **106**(0,N-2), **108**(1,N-2) and **106**(M-1,N-2) in column *col*(N-2) is configured to receive column-addressing signal CL(N-2). One input of each of SCC_cells **106**(0,N-1), **108**(1,N-1) and **106**(M-1,N-1) in column *col*(N-1) is configured to receive a column-addressing signal CL(N-1).

[0037] Two inputs of each of SCC_cells **106**(0,0)-**106**(0,N-2) in row *rw_0* are configured to receive row-addressing signal RW(0). One input of each of SCC_cells **106**(0,0)-**106**(0,N-2) in row *rw_0* is configured to receive row-addressing signal RW(1). Two inputs of SCC_cell **106**(0,N-1) in row *rw_0* are configured to receive row-addressing signal RW(1).

[0038] Two inputs of each of SCC_cells **108**(1,0)-**108**(1,N-2) in row *rw_1* are configured to receive row-addressing signal RW(1). One input of each of SCC_cells **108**(1,0)-**108**(1,N-2) in row *rw_1* is configured to receive row-addressing signal RW(2). Two inputs of SCC_cell **108**(1,N-1) in row *rw_1* are configured to receive row-addressing signal RW(2).

[0039] Two inputs of each of SCC_cells **106**(M-1,0)-**106**(M-1,N-2) in row *rw_M-1* are

configured to receive row-addressing signal $RW(M-1)$. One input of each of SCC_cells **106**(M-1,0)-**106**(M-1,N-2) in row rw_M-1 is configured to receive row-addressing signal $RW(M)$. Two inputs of each of SCC_cells **106**(M-1,0)-**106**(M-1,N-2) in row rw_M-1 are configured to receive row-addressing signal $RW(M)$.

[0040] In FIG. 1B, row-addressing signal $RW(0)$ is coupled to a first reference voltage VDD. Row-addressing signal $RW(M)$ corresponding to an imaginary row rw_M is coupled to a second reference voltage VSS. Column-addressing signal $CL(N-1)$ is coupled to either the first reference voltage VDD or the second reference voltage VSS. In some embodiments, the first and second reference voltages have corresponding values other than VDD and VSS.

[0041] FIG. 1C is a schematic diagram of an oscillator **114**, in accordance with some embodiments.

[0042] Oscillator **114** represents an application of delay chain **102** in that oscillator **114** includes digitally-controlled delay chain. Accordingly, oscillator **114** is described as a digitally-controlled type of oscillator. In some embodiments, oscillator **114** is described as a phase-locked loop (PLL).

[0043] In FIG. 1C, output node nd_OUT of delay chain **102** is coupled to input node nd_IN of delay chain **102** via a feedback coupling **116**.

[0044] FIG. 1D is a schematic diagram of a delay line **118**, in accordance with some embodiments.

[0045] Delay line **118** represents an application of delay chain **102** in that delay line **118** includes digitally-controlled delay chain. Accordingly, delay line **118** is described as a digitally-controlled type of delay line. In some embodiments, delay line **118** is described as a delay-locked loop (DLL).

[0046] In FIG. 1D, input node nd_IN of delay chain **102** receives a clock signal CLK which experiences an overall propagation delay as signal CLK propagates to output node nd_OUT of delay chain **102**. A signal CLK_dlyd is generated by delay chain **102** on output node nd_OUT which represents a delay version of signal CLK, and where the text string dlyd is an abbreviation of delayed.

[0047] FIG. 2A is a schematic diagram of an array **204**, in accordance with some embodiments.

[0048] In some embodiments, FIG. 2A is described as a simplified version of FIG. 1B and accordingly array **204** is described as a simplified version of array **104**. FIG. 2A shows more of the columns in array **204** than FIG. 1B shows of the columns in array **104**, the latter showing three columns, namely $col(0)$, $col(N-2)$ and $col(N-1)$. By contrast, FIG. 2A shows four columns of array **204**, namely columns $col(0)$, $col(1)$, $col(j)$ and $col(N-1)$.

[0049] In FIG. 2A, for simplicity of illustration, it is assumed that the number of rows M is $M=9$ and the number of columns N is $N=8$. As such, array **204** includes rows rw_0 - rw_8 and columns col_0 - col_7 . In some embodiments, M and N have corresponding values other than $M=9$ and $N=8$.

[0050] Even-numbered rows include SCC_cells **206B**(i,j) and **206D**(i,j). Odd-numbered rows include SCC_cells **208C**(i,j) and **208E**(i,j), where i is an odd number. For example, even row rw_0 of array **204** includes SCC_cells **206B**(0,0)-**206B**(0,6) and **208D**(0,7). For example, odd row rw_1 of array **204** includes SCC_cells **208C**(1,0)-**208C**(1,6) and **208E**(1,7).

[0051] Internal configurations of even-row SCC_cells **206B**(i,j) and **206D**(i,j) are shown correspondingly in FIGS. 2B and 2D. Internal configurations of odd-row SCC_cells **208C**(i,j) and **208D**(i,j) are shown correspondingly in FIGS. 2C and 2E.

[0052] In FIG. 2A, SCC_cells **206B**(i,j), **206D**(i,j), **208C**(i,j) and **208E**(i,j) are controlled via addressing by signals $CL(0)$ - $CL(N-1)$ and $RW(0)$ - $RW(9)$. A controller (see **112** FIG. 1B) is configured to generate signals $CL(0)$ - $CL(N-1)$ and $RW(0)$ - $RW(9)$ based on a digital thermometer encoding scheme (see FIG. 3, or the like).

[0053] It is to be recalled that each column $col(j)$ represents $SA_CDAC(j)$ in corresponding delay stage $ds(j)$ (see FIGS. 1A-1B). In the example of FIG. 2A, each of SCC_cells **206B**(i,j), **206D**(i,j), **208C**(i,j) and **208E**(i,j) includes a capacitor (see FIG. 2B-2E, or the like) that is separately couplable to corresponding output nodes $out(0)$ - $out(7)$ of delay stage $ds(0)$ - $ds(7)$ in accordance with the digital thermometer encoding scheme.

[0054] In the example of FIG. 2A, $M \times N = 9 \times 8 = 72$ capacitors are separately couplable to

corresponding output nodes out(0)-out(7). Signal RW(0) is coupled to voltage VDD. Signal RW(M), which corresponds to an imaginary row rw_9 is coupled to voltage VSS.

[0055] In FIG. 2A, the digital thermometer code is formed from signals RW(0)-RW(8) and CL(0)-CL(7).

[0056] The digital thermometer encoding scheme of FIG. 2A separately couples none, some or all the capacitors in SCC_cells **206B(i,j)**, **206D(i,j)**, **208C(i,j)** and **208E(i,j)** to corresponding output nodes out(0)-out(7) row-wise sequentially as follows: row rw(0) from **206B(0,0)** to **206B(0,1)** to . . . **206B(0,j)** to . . . **206D(0,7)**; then row rw(1) from **208C(1,0)** to . . . **208C(1,1)** to . . . **208C(1,j)** to . . . **208E(1,7)**; then row rw(2) from **206B(2,0)** to . . . **206D(2,7)**; then row rw(3) from **208C(3,0)** to . . . **208E(3,7)**; . . . then row rw(7) from **208C(7,0)** to . . . **208C(7,1)** to . . . **208C(7,j)** to . . . **208D(7,7)**; and then row rw(8) from **206B(8,0)** to . . . **206B(8,1)** to . . . **206B(8,j)** to . . . **206D(8,7)**. Such row-wise sequential coupling of capacitors changes the coupling-state of only one of the corresponding capacitors changes for each change in the value of the thermometer code, and thus is gray-code compliant and monotonic. Gray code encoding produces a string of binary digits for which any two successive values of the string differ only in the value of one of the binary digits of the string. In some embodiments, a gray code is referred to as a reflected binary (RB) code. In some embodiments, gray-code-compliant adjustment of SA_CDAC(0)-SA_CDAC(N-1) connotes such row-wise sequential coupling of capacitors in that the coupling-state of only one of the corresponding capacitors changes for each change in the value of the thermometer code.

[0057] In some embodiments, when the capacitor of SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is coupled to corresponding output node out(j), then SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is described as being turned on. And when the capacitor of SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is not coupled to corresponding output node out(j), then SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is described as being turned off.

[0058] In some embodiments, when the capacitor of SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is coupled to corresponding output node out(j), then SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is described as being in a thermometer-code-ON state. And when the capacitor of SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is not coupled to corresponding output node out(j), then SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is described as being in a thermometer-code-OFF state. In the example of FIG. 2A, the following is assumed: SCC_cells **206B(i,j)**, **206D(i,j)**, **208C(i,j)** and **208E(i,j)** included in a first range from SCC_cell **206B(0,0)** to SCC_cell **206B(4,0)** are in the thermometer-code-ON state; and SCC_cells **206B(i,j)**, **206D(i,j)**, **208C(i,j)** and **208E(i,j)** included in a second range from SCC_cells **206B(4,1)** through **206D(8,7)** are in the thermometer-code-OFF state.

[0059] In combination with decoders (see FIG. 2B-2D, 4B-4D, or the like) correspondingly included in each of SCC_cell **206B(i,j)**, **206D(i,j)**, **208C(i,j)** and **208E(i,j)**, the use of signals RW(0)-RW(8) and CL(0)-CL(7) to form the thermometer code reduces the total number of control lines needed for the SA_CDACs corresponding to columns col(0)-col(7) as compared to a CDAC-based delay chain according to another approach. The reduced total number of control lines for the SA_CDACs corresponding to columns col(0)-col(7) results in the corresponding delay chain occupying a smaller area (having a smaller footprint) as compared to the CDAC-based delay chain according to another approach.

[0060] FIG. 2B is a schematic diagram of SCC_cell **206B(i,j)** for i=even and j={0, . . . , N-2}, in accordance with some embodiments.

[0061] FIG. 2B extends the example of FIG. 2A. SCC_cell **206B(i,j)** is used in even rows of array **204** of FIG. 2A, or the like. Recalling that N=8 in the example of FIG. 2A, SCC_cell **206B(i,j)** represents SCC_cells **206B(0,0)**-**206B(8,6)**.

[0062] SCC_cell **206B(i,j)** includes a decoder **220**, a switch P(i,j) and a capacitor. Switch P(i,j) is coupled between output node out(j) of corresponding delay stage ds(j) and the capacitor. The capacitor is coupled between switch P(i,j) and VSS. A control input of switch P(i,j) is coupled to

the output of decoder **220**. Decoder **220** has three inputs correspondingly coupled to signals CL(j), RW(i) and RW(i+1).

[0063] In FIG. **2B**, switch P(i,j) is a field-effect transistors (FET) having a P-type dopant used for positive-channel metal-oxide semiconductor (PMOS) transistor technologies, i.e., a PFET. When the output of decoder **220** is a logical zero, PFET P(i,j) turns on, i.e., switch P(i,j) is in the closed state and the capacitor is coupled to output node out(j). When the output of decoder **220** is a logical one, PFET P(i,j) turns off, i.e., switch P(i,j) is in the open state and, in effect, the capacitor is not coupled to output node out(j). In some embodiments, the switch is an FET having an N-type dopant used for negative-channel metal-oxide semiconductor (NMOS) transistor technologies, i.e., an NFET.

[0064] Regarding FIG. **2B**, in some embodiments, the capacitor correspondingly of each instance of SCC_cell **206B**(i,j) is described as being separately couplable to output node out(j). In some embodiments, the description of each instance of SCC_cell **206B**(i,j) as being separately couplable to output node out(j) is understood to mean that controlling a first instance of switch P(i,j) of a first instance of SCC_cell **206B**(i,j) to turn on/off is performed separately (or independently) from controlling a second instance of switch P(α , β) of a second instance of SCC_cell **206B**(i,j) to turn on/off. Accordingly, in such embodiments, the on/off state of the first instance of switch P(i,j) is described as being separately (or independently) controllable relative to the on/off state of the second instance of switch P(i,j). Decoder **220** includes a logical AND gate and a logical NOR gate. The AND gate has two inputs correspondingly configured to receive signals CL(j) and RW(i). The NOR gate has two inputs correspondingly configured to receive the output of the AND gate and RW(i+1). In some embodiments, decoder **220** is described as a logical AND-OR-INVERT (AOI) gate.

[0065] In some embodiments, the thermometer-code-ON state is described as driving SCC_cell **206B**(i,j), **206D**(i,j), **208C**(i,j) or **208E**(i,j) according to a duty cycle of 100%, or as SCC_cell **206B**(i,j), **206D**(i,j), **208C**(i,j) or **208E**(i,j) having a duty cycle of 100%. In such embodiments, conversely, the thermometer-code-OFF state is described as driving SCC_cell **206B**(i,j), **206D**(i,j), **208C**(i,j) or **208E**(i,j) according a duty cycle of 0% or as SCC_cell **206B**(i,j), **206D**(i,j), **208C**(i,j) or **208E**(i,j) having a duty cycle of 0%. In some embodiments, is described as being controlled according to a binary coupling scheme.

[0066] FIG. **2C** is a schematic diagram of SCC_cell **208C**(i,j) for i=odd and j={0, . . . , N-2}, in accordance with some embodiments.

[0067] FIG. **2C** extends the example of FIG. **2A**. SCC_cell **208C**(i,j) is used in odd rows of array **204** of FIG. **2A**, or the like. Recalling that N=8 in the example of FIG. **2A**, SCC_cell **208C**(i,j) represents SCC_cells **208C**(1,0)-**208C**(7,6).

[0068] SCC_cell **208C**(i,j) of FIG. **2C** is similar to SCC_cell **206B**(i,j) of FIG. **2B**. In some embodiments, the capacitor correspondingly of each instance of SCC_cell **206C**(i,j) is described as being separately couplable to output node out(j). For brevity, the discussion will focus on differences of SCC_cell **208C**(i,j) as compared to SCC_cell **206B**(i,j) rather than on similarities.

[0069] SCC_cell **208C**(i,j) differs from SCC_cell **206B**(i,j) in that one of the inputs of SCC_cell **208C**(i,j) is inverted. More particularly, one of the inputs of the AND gate of decoder **222**, namely the input of the AND gate configured to receive signal CL(j), is inverted as indicated by prefixed inversion bubble **210**. In some embodiments, the inverted input is described as being an active low input. In some embodiments, SCC_cell **208C**(i,j) does not have the inverted input but instead SCC_cell **206B**(i,j) has the inverted input, i.e., the AND gate of decoder **220** which is configured to receive signal CL(j) is inverted.

[0070] Similar to decoder **220** of FIG. **2B**, in some embodiments, decoder **222** of FIG. **2C** is described as a logical AND-OR-INVERT (AOI) gate.

[0071] FIG. **2D** is a schematic diagram of SCC_cell **206D**(i,7) for i=even and j=7, in accordance with some embodiments.

[0072] FIG. 2D extends the example of FIG. 2A. SCC_cell 206D(i,7) is used in even rows of the last column col_N-1 of the array, i.e., col_7 in the context of array 204 of FIG. 2A, or the like.

[0073] SCC_cell 206D(i,7) of FIG. 2D is similar to SCC_cell 206B(i,j) of FIG. 2B. In some embodiments, the capacitor correspondingly of each instance of SCC_cell 206D(i,j) is described as being separately couplable to output node out(j). For brevity, the discussion will focus on differences of SCC_cell 206D(i,7) of FIG. 2D as compared to SCC_cell 206B(i,j) of FIG. 2B rather than on similarities.

[0074] Though SCC_cell 206D(i,7) has three inputs, nevertheless SCC_cell 206D(i,7) is configured to receive fewer signals than SCC_cell 206B(i,j) of FIG. 2B. SCC_cell 206D(i,7) is configured to receive two different signals on three inputs, i.e., two of the inputs are configured to receive the same signal. SCC_cell 206D(i,7) is not configured to receive signal RW(i). In particular, not only is one of the inputs of the NOR gate of decoder 220 configured to receive signal RW(i+1), but also one of the inputs of the AND gate of decoder 220 is configured to receive signal RW(i+1). By contrast, the corresponding input of the AND gate of decoder 220 of FIG. 2B is configured to receive signal RW(i).

[0075] FIG. 2E is a schematic diagram of SCC_cell 208E(i,7) for i=odd and j=7, in accordance with some embodiments.

[0076] FIG. 2E extends the example of FIG. 2A. SCC_cell 208E(i,7) is used in odd rows of the last column col_N-1 of the array, i.e., col_7 in the context of array 204 of FIG. 2A, or the like.

[0077] SCC_cell 208E(i,7) of FIG. 2E is similar to SCC_cell 208C(i,j) of FIG. 2C. In some embodiments, the capacitor correspondingly of each instance of SCC_cell 206D(i,j) is described as being separately couplable to output node out(j). For brevity, the discussion will focus on differences of SCC_cell 208E(i,7) of FIG. 2E as compared to SCC_cell 208C(i,j) of FIG. 2C rather than on similarities.

[0078] Though SCC_cell 208E(i,7) has three inputs, nevertheless SCC_cell 208E(i,7) is configured to receive fewer signals than SCC_cell 208C(i,j) of FIG. 2C. SCC_cell 208E(i,7) is configured to receive two different signals on three inputs, i.e., two of the inputs are configured to receive the same signal. SCC_cell 208E(i,7) is not configured to receive signal RW(i). In particular, not only is one of the inputs of the NOR gate of decoder 222 configured to receive signal RW(i+1), but also one of the inputs of the AND gate of decoder 222 is configured to receive signal RW(i+1). By contrast, the corresponding input of the AND gate of decoder 222 of FIG. 2C is configured to receive signal RW(i).

[0079] In each of SCC_cell 206D(i,7) of FIG. 2D and SCC_cell 208E(i,7) of FIG. 2E, configuring two of the inputs to receive signal RW(i+1) has the corresponding effects of causing decoders 220 and 222 to function as logical inverter gates with respect to signal RW(i+1). In other words, in each of SCC_cell 206D(i,7) of FIG. 2D and SCC_cell 208E(i,7) of FIG. 2E, configuring two of the inputs to receive signal RW(i+1) has the corresponding effects of causing the state of signal CL(7) to have no effect on the output of decoders 220 and 222. In some embodiments, in light of such effects, signal CL(7) is described as a ‘don't care’ signal with respect to the operation of decoders 220 and 222. Accordingly, in such embodiments, signal CL(7) is configured according to another purpose, e.g., to reduce energy consumption, or noise, or the like.

[0080] A truth table for SCC_cell 206D(i,7) is shown below as Table 1.

TABLE-US-00001 TABLE 1 (SCC_cell 206D(i, 7)) CL(7) RW(i + 1) AND OR NOR 0 0 0 0 1 0 1 0 1 0 1 0 0 0 1 1 1 1 1 0

[0081] A truth table for SCC_cell 208E(i,7) is shown below as Table 2. The inversion of signal CL(7) is referred to as CL(7)_bar.

TABLE-US-00002 TABLE 2 (SCC_cell 208E(i, 7)) CL(7) CL(7)_bar RW(i + 1) AND OR NOR 0 1 0 0 0 1 0 1 1 1 1 0 1 0 0 0 0 1 1 0 1 0 1 0

[0082] FIG. 3 is a table 324 showing partial values of a thermometer code, in accordance with some embodiments.

[0083] The table of FIG. 3 extends the example of FIG. 2A. As columns col_0-col_7 of array **204** of FIG. 2A represent corresponding SC_CDACs, Table **324** provides examples of how a thermometer code is used to adjust each of the SA_CDACs separately relative to the other SA_CDACs.

[0084] Table **324** includes: a first column (column 1) that shows the row-numbers assigned to the rows in Table **324**; a second column (column 2) that shows the count of SCC_cells **206B(i,j)**, **206D(i,j)**, **208C(i,j)** and **208E(i,j)** that are in a thermometer-code-ON state; a third column (column 3) that shows signals RW(0)-RW(7) as corresponding binary digits concatenated to form the upper eight bits of the example thermometer code; a fourth column (column 4) that is specific to the 'last' ON-row (discussed below) of array **204**, and where column 4 shows signals CL(0)-CL(7) as corresponding binary digits concatenated to form the lower eight bits of the example thermometer code; and a fifth column (column 5) that is specific to the 'last' ON-row (discussed below) of array **204**, and where column 5 shows which, if any, of columns col_0-col_7 have SCC_cells **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** is in the thermometer-code-ON state.

[0085] The full example code represented by Table **324** is formed by concatenating signals RW(0)-RW(7) and CL(0)-CL(7) as corresponding binary digits representing the upper eight bits and lower eight bits of the example thermometer code in the context of the example of FIG. 2A. Table **324** is a partial listing of the permutations and combinations of values for the example thermometer code in the context of the example of FIG. 2A.

[0086] Column 3 of Table **324** considers each of rows rw_0-rw_7 as a whole in terms of the thermometer-code-ON state. Accordingly, each row rw_i as a whole is represented by a corresponding single binary digit in column 3 of Table **324**. If all SCC_cells in the row rw_i are in the thermometer-code-ON state, then the binary digit representing row_i in column 3 of Table **324** is set to one (logical high). If fewer than all SCC_cells in the row rw_i are in the thermometer-code-ON state, then the binary digit representing row_i in column 3 of Table **324** is set to zero (logical low).

[0087] In some embodiments, the 'last' ON-row of Table **324** is the row (row rw_i) of Table **324** that satisfies the following conditions: (1) row_i has one or more SCC_cells in the thermometer-code-ON state; and (2A) row_i is row_0 or (2B) all SCC_cells in the preceding row rw_i-1 are in the thermometer-code-ON state.

[0088] Column 5 of Table **324** is specific to the 'last' ON-row and thus considers each SCC_cell in the 'last' ON-row in terms of its thermometer-code-ON state. That is, if SCC_cell in column col_i is in the thermometer-code-ON state, then the binary digit representing col_i in column 5 of Table **324** is set to one (logical high). If the SCC_cell in the row col_i in the thermometer-code-OFF state, then the binary digit representing col_i in column 5 of Table **324** is set to zero (logical low).

[0089] Regarding row 1 of Table **324**, no SCC_cell (none **206B(i,j)**, **206D(i,j)**, **208C(i,j)** nor **208E(i,j)**) is in the thermometer-code-ON state. For the other rows of Table **324**, one or more of SCC_cells **206B(i,j)**, **206D(i,j)**, **208C(i,j)** or **208E(i,j)** are in the thermometer-code-ON state.

[0090] For rows 2-9 of Table **324**, only row rw_0 of array **204** has any SCC_cell that is in the thermometer-code-ON state. for rows 2—As such, rows 2-9 of Table **324**, column 5 of Table **324** is understood as referring to row rw_0 of array **204** as the 'last' ON-row. For rows 2-9 of Table **324**, the values of the upper eight bits (column 3) of the example thermometer code are all zero.

[0091] Regarding rows 10-17 of Table **324**, column 5 of Table **324** is understood as referring to row rw_1 of array **204** as the 'last' ON-row. For each of rows 10-17 of Table **324**, the values of the upper eight bits (column 3) of the example thermometer code are 10000000, where the left most binary digit is set to one (logical high).

[0092] In column 4 of Table **324**, the binary digit representing signal CL(7) is shown as the letter x rather than being set to one (logical high) or zero (logical low). The use of letter x in column 4 of Table **324** is intended to indicate that signal CL(7) is a 'don't care' signal with respect to the operation of decoders **220** and **222**, as discussed above.

[0093] FIG. 4A is a schematic diagram of an array **404**, in accordance with some embodiments.

[0094] Array **404** of FIG. 4A is similar to array **204** of FIG. 2A. As such, in FIG. 4A, for simplicity of illustration, it is assumed likewise that the number of rows M is $M=9$ and the number of columns N is $N=8$. As such, array **404** likewise includes rows rw_0 - rw_8 and columns col_0 - col_7 . In some embodiments, M and N have corresponding values other than $M=9$ and $N=8$. For brevity, the discussion will focus on differences of array **404** as compared to array **204** rather than on similarities.

[0095] In array **404**, even-numbered rows include SCC_cells **406B**(i,j) and **406D**(i,j). Odd-numbered rows include SCC_cells **408C**(i,j) and **408E**(i,j), where i is an odd number. For example, even row rw_0 of array **404** includes SCC_cells **406B**(0,0)-**406B**(0,6) and **406D**(0,7). For example, odd row rw_1 of array **404** includes SCC_cells **408C**(1,0)-**408C**(1,6) and **408E**(1,7).

[0096] In array **404**, all but one of SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) are controlled according to the binary coupling scheme (see FIG. 2A), whereas all SCC_cells **206B**(i,j), **206D**(i,j), **208C**(i,j) and **208E**(i,j) in array **204** of FIG. 2A are controlled according to the binary coupling scheme. In array **404**, one of SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) is controlled according to a fractional coupling scheme (discussed below). In some embodiments, array **404** is described as being controlled according to a hybrid binary and fractional coupling scheme.

[0097] Internal configurations of even-row SCC_cells **406B**(i,j) and **406D**(i,j) correspondingly are shown in FIGS. 4B and 4D. Internal configurations of odd-row SCC_cells **408C**(i,j) and **408E**(i,j) correspondingly are shown in FIGS. 4C and 4E.

[0098] In FIG. 4A, SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) are controlled via addressing by signals $CL(-1)$, $CL(0)$ - $CL(N)$ and $RW(0)$ - $RW(9)$. A controller (see **112** FIG. 1B) is configured to generate signals $CL(-1)$, $CL(0)$ - $CL(N)$ and $RW(0)$ - $RW(9)$ based on a digital thermometer encoding scheme (see FIG. 5, or the like). Signal $CL(-1)$ represents a signal for an imaginary column col_{-1} that precedes col_0 . Signal $CL(8)$, i.e., $CL(N)$, represents a signal for an imaginary column col_8 .

[0099] In FIG. 4A, the digital thermometer code is formed from signals $RW(0)$ - $RW(8)$, $CL(-1)$ and $CL(0)$ - $CL(8)$.

[0100] In the example of FIG. 4A, the following is assumed: SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) in a first range from SCC_cell **406B**(0,0) to SCC_cell **406B**(4,0) are in the thermometer-code-ON state; SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) in a second range from SCC_cell **406B**(4,2) (not shown) through SCC_cell **406D**(8,7) are assumed to be in the thermometer-code-OFF state; and SCC cell **406B**(4,1) is in a fractional state. Accordingly, SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) in the first range are described as being driven according to a duty cycle of 100%. SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) in the second range are described as being driven according to a duty cycle of 0%. In some embodiments, by being in the fractional state, SCC cell **406B**(4,1) is understood as being driven according to a partial duty cycle, DCY, in a range $0\% < DCY < 100\%$.

[0101] In general, the SCC_cell **406B**(i,j), **406D**(i,j), **408C**(i,j) or **408E**(i,j) which is in the fractional state is the first SCC_cell **406B**(i,j), **406D**(i,j), **408C**(i,j) or **408E**(i,j) that otherwise would be in the thermometer-code-OFF state but for the use of the fractional coupling scheme. In some embodiments, the SCC_cell **406B**(i,j), **406D**(i,j), **408C**(i,j) or **408E**(i,j) which is in the fractional state is referred to as the fractionally coupled SCC_cell. In some embodiments, Here, the first SCC_cell **406B**(i,j), **406D**(i,j), **408C**(i,j) or **408E**(i,j) that otherwise would be in the thermometer-code-OFF state but for the use of the fractional coupling scheme is referred to as the otherwise-first SCC_cell in the thermometer-code-OFF state.

[0102] When each of SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) is in the thermometer-code-ON state, then each of SCC_cells **406B**(i,j), **406D**(i,j), **408C**(i,j) and **408E**(i,j) adds one unit of capacitance, CPC, to output node $out(j)$ of corresponding delay stage $ds(j)$. By contrast, when

SCC_cell **406B(i,j)**, **406D(i,j)**, **408C(i,j)** or **408E(i,j)** is the fractionally coupled SCC_cell, then the fractionally coupled SCC_cell adds a fraction of one CPC(see FIGS. **4F-4G**) to output node out(j) of corresponding delay stage ds(j), where the magnitude of the fraction is proportional to the value of DCY multiplied by CPC such that $0\% < \text{DCY} * \text{CPC} < 100\%$. The use of the fractional coupling scheme for the otherwise-first SCC_cell in the thermometer-code-OFF state, i.e., the use of the hybrid binary and fractional coupling scheme, achieves better/finer granularity as compared to the purely a binary coupling scheme while nevertheless maintaining monotonicity.

[0103] In combination with decoders (see FIG. **4B-4D**, or the like) correspondingly included in each of SCC_cell **406B(i,j)**, **406D(i,j)**, **408C(i,j)** and **408E(i,j)**, the use of signals RW(0)-RW(8) and CL(0)-CL(7) to form the thermometer code reduces the total number of control lines needed for the SA_CDACs corresponding to columns col(0)-col(7) as compared to a CDAC-based delay chain according to another approach. The reduced total number of control lines for the SA_CDACs corresponding to columns col(0)-col(7) results in the corresponding delay chain occupying a smaller area (having a smaller footprint) as compared to the CDAC-based delay chain according to another approach.

[0104] FIG. **4B** is a schematic diagram of SCC_cell **406B(i,j)** for $i=\text{even}$ and $j=\{0, \dots, N-2\}$, in accordance with some embodiments.

[0105] FIG. **4B** extends the example of FIG. **4A**. SCC_cell **406B(i,j)** is used in even rows of array **404** of FIG. **4A**, or the like. Recalling that $N=8$ in the example of FIG. **4A**, SCC_cell **406B(i,j)** represents SCC_cells **406B(0,0)-406B(8,6)**. SCC_cell **406B(i,j)** of FIG. **4B** is similar to SCC_cell **206B(i,j)** of FIG. **2B**. For brevity, the discussion will focus on differences of SCC_cell **406B(i,j)** as compared to SCC_cell **206B(i,j)** rather than on similarities.

[0106] SCC_cell **406B(i,j)** has four inputs whereas SCC_cell **206B(i,j)** has three inputs. As compared to SCC_cell **206B(i,j)**, the additional fourth input of SCC_cell **406B(i,j)** is configured to receive a fractional coupling signal FRCT_ev (see FIG. **4F**).

[0107] SCC_cell **406B(i,j)** includes a decoder **420** whereas SCC_cell **206B(i,j)** includes decoder **220**. Decoder **420** of SCC_cell **406B(i,j)** has four inputs whereas decoder **220** of SCC_cell **206B(i,j)** has three inputs. As compared to decoder **220**, the additional fourth input of decoder **420** is configured to receive fractional coupling signal FRCT_ev.

[0108] FIG. **4C** is a schematic diagram of SCC_cell **408C(i,j)** for $i=\text{odd}$ and $j=\{0, \dots, N-2\}$, in accordance with some embodiments.

[0109] FIG. **4C** extends the example of FIG. **4A**. SCC_cell **408C(i,j)** is used in odd rows of array **404** of FIG. **4A**, or the like. Recalling that $N=8$ in the example of FIG. **4A**, SCC_cell **408C(i,j)** represents SCC_cells **408C(1,0)-408C(7,6)**.

[0110] SCC_cell **408C(i,j)** of FIG. **4C** is similar to SCC_cell **208C(i,j)** of FIG. **2C**. For brevity, the discussion will focus on differences of SCC_cell **408C(i,j)** as compared to SCC_cell **208C(i,j)** rather than on similarities.

[0111] SCC_cell **408C(i,j)** of FIG. **4C** has four inputs whereas SCC_cell **208C(i,j)** of FIG. **2C** has three inputs. As compared to SCC_cell **208C(i,j)**, the additional fourth input of SCC_cell **408C(i,j)** is configured to receive fractional coupling signal FRCT_od (see FIG. **4G**).

[0112] SCC_cell **408C(i,j)** includes a decoder **422** whereas SCC_cell **208C(i,j)** includes decoder **222**. Decoder **422** of SCC_cell **408C(i,j)** has four inputs whereas decoder **222** of SCC_cell **208C(i,j)** has three inputs. As compared to decoder **222**, the additional fourth input of decoder **422** is configured to receive fractional coupling signal FRCT_od.

[0113] FIG. **4D** is a schematic diagram of SCC_cell **406D(i,7)** for $i=\text{even}$ and $j=7$, in accordance with some embodiments.

[0114] FIG. **4D** extends the example of FIG. **4A**. SCC_cell **406D(i,7)** is used in even rows of the last column col_{N-1} of the array, i.e., col₇ in the context of array **404** of FIG. **4A**, or the like. SCC_cell **406D(i,7)** of FIG. **4D** is similar to SCC_cell **206D(i,j)** of FIG. **2D**. For brevity, the discussion will focus on differences of SCC_cell **406D(i,7)** of FIG. **4D** as compared to SCC_cell

206D(i,j) of FIG. 2D rather than on similarities.

[0115] SCC_cell **406D(i,j)** of FIG. 4D has four inputs whereas SCC_cell **206D(i,j)** of FIG. 2D has three inputs. As compared to SCC_cell **206D(i,j)**, the additional fourth input of SCC_cell **406D(i,j)** is configured to receive a fractional coupling signal FRCT_ev (see FIG. 4F).

[0116] SCC_cell **406D(i,j)** includes a decoder **420** whereas SCC_cell **206D(i,j)** includes decoder **220**. Decoder **420** of SCC_cell **406D(i,j)** has four inputs whereas decoder **220** of SCC_cell **206D(i,j)** has three inputs. As compared to decoder **220**, the additional fourth input of decoder **420** is configured to receive fractional coupling signal FRCT_ev.

[0117] FIG. 4E is a schematic diagram of SCC_cell **408E(i,j)** for $i=\text{odd}$ and $j=7$, in accordance with some embodiments.

[0118] FIG. 4E extends the example of FIG. 4A. SCC_cell **408E(i,j)** is used in odd rows of the last column col_N-1 of the array, i.e., col_7 in the context of array **404** of FIG. 4A, or the like.

SCC_cell **408E(i,j)** of FIG. 4E is similar to SCC_cell **208E(i,j)** of FIG. 2E. For brevity, the discussion will focus on differences of SCC_cell **408E(i,j)** of FIG. 4E as compared to SCC_cell **208E(i,j)** of FIG. 2E. rather than on similarities.

[0119] SCC_cell **406E(i,j)** of FIG. 4E has four inputs whereas SCC_cell **206E(i,j)** of FIG. 2E has three inputs. As compared to SCC_cell **206E(i,j)**, the additional fourth input of SCC_cell **406E(i,j)** is configured to receive a fractional coupling signal FRCT_od (see FIG. 4G).

[0120] SCC_cell **406E(i,j)** includes a decoder **422** whereas SCC_cell **206E(i,j)** includes decoder **222**. Decoder **422** of SCC_cell **406E(i,j)** has four inputs whereas decoder **222** of SCC_cell **206E(i,j)** has three inputs. As compared to decoder **222**, the additional fourth input of decoder **422** is configured to receive fractional coupling signal FRCT_od.

[0121] FIG. 4F is a schematic diagram of a selector **440**, in accordance with some embodiments.

[0122] FIG. 4F extends the example of FIG. 4A. Selector **440** is used for even rows of array **404** of FIG. 4A, or the like. Selector **440** is configured to generate fractional coupling signal FRCT_ev (see FIGS. 4B and 4D) and thereby drive the otherwise-first SCC_cell in the thermometer-code-OFF state.

[0123] Selector **440** includes a four-input AND gate **410** and a two-input AND gate **444**. AND gate **410** has two non-inverted inputs and two inverted inputs. The two inverted inputs are indicated by prefixed inversion bubbles **410**, only one of which is called out with reference number **410** for simplicity of illustration. In some embodiments, the inverted inputs are described as being active low inputs.

[0124] A first one of the two non-inverted inputs of AND gate **442** is configured to receive row-addressing signal RW(i-1). A second one of the two non-inverted inputs of AND gate **442** is configured to receive column-addressing signal CL(j). A first one of the two inverted inputs of AND gate **442** is configured to receive row-addressing signal RW(i). A second one of the two inverted inputs of AND gate **442** is configured to receive column-addressing signal CL(j-1). AND gate **442** is configured to generate an even-row fractional selection signal frc_sel_rw_ev (discussed below).

[0125] Regarding AND gate **444**, a first one of the inputs of AND gate **444** is configured to receive a duty signal DTY. In some embodiments, duty signal DTY is a pulse width modulation (PWM) signal with a constant pulse width or a randomized pulse width. In some embodiments, PWM is described as pulse-duration modulation (PDM) or pulse-length modulation (PLM). Duty signal DTY is configured with a waveform that represents the partial duty cycle, DCY, in the range $0\% < \text{DCY} < 100\%$.

[0126] A second one of the inputs of AND gate **444** is configured to receive even-row fractional selection signal frc_sel_rw_ev from AND gate **442**. AND gate **444** is configured to generate fractional coupling signal FRCT_ev.

[0127] Even-row fractional selection signal frc_sel_rw_ev is used to select the otherwise-first SCC_cell in the thermometer-code-OFF state, where the latter is to be driven by duty signal DTY.

A truth table for fractional control signal FRCT_ev is shown below in part as Table 3.

[0128] In some embodiments, the combinatorial logic of selector **440** is described as follows. For $i=\text{even}$, set $\text{frc_sel_rw_ev}=1$ when $\text{RW}(i-1)=1$, $\text{RW}(i)=0$, $\text{CL}(j-1)=0$ and $\text{CL}(j)=1$. More particularly, for $i=\text{even}$, set $\text{frc_sel_rw_ev}=\text{RW}(i-1)*(\sim\text{RW}(i))*(\sim\text{CL}(j-1))*\text{CL}(j)$ and set $\text{FRCT_ev}=\text{DTY}*\text{frc_sel_rw_ev}$, where $(\sim\text{RW}(i))$ and $(\sim\text{CL}(j-1))$ correspondingly are the inverses of $\text{RW}(i)$ and $\text{CL}(j-1)$.

[0129] FIG. **4G** is a schematic diagram of a selector **446**, in accordance with some embodiments.

[0130] FIG. **4G** extends the example of FIG. **4A**. Selector **446** is used for odd rows of array **404** of FIG. **4A**, or the like. Selector **446** is configured to generate fractional coupling signal FRCT_od (see FIGS. **4C** and **4E**) and thereby drive the otherwise-first SCC_cell in the thermometer-code-OFF state.

[0131] Selector **446** of FIG. **4G** is similar to selector **440** of FIG. **4F**. For brevity, the discussion will focus on differences of selector **446** of FIG. **4G** as compared to selector **440** of FIG. **4F** rather than on similarities.

[0132] Regarding selector **446**, the second non-inverted input and the second inverted input of AND gate **442** of selector **446** are configured differently as compared to the second non-inverted input and the second inverted input of AND gate **442** of selector **440**.

[0133] The second non-inverted input of AND gate **442** of selector **446** is configured to receive column-addressing signal $\text{CL}(j-1)$ rather than signal $\text{CL}(j)$ as in selector **440**. A second inverted input of AND gate **442** of selector **446** is configured to receive column-addressing signal $\text{CL}(j)$ rather than signal $\text{CL}(j-1)$ as in selector **440**. A truth table for fractional control signal FRCT_od is shown in part below as Table 3.

[0134] In some embodiments, the combinatorial logic of selector **446** is described as follows. For $i=\text{odd}$, set $\text{frc_sel_rw_od}=1$ when $\text{RW}(i-1)=1$, $\text{RW}(i)=0$, $\text{CL}(j-1)=1$ and $\text{CL}(j)=0$. More particularly, for $i=\text{odd}$, set $\text{frc_sel_rw_od}=\text{RW}(i-1)*(\sim\text{RW}(i))*\text{CL}(j-1)*(\sim\text{CL}(j))$ and set $\text{FRCT_od}=\text{DTY}*\text{frc_sel_rw_od}$, where $(\sim\text{CL}(j))$ is the inverse of $\text{CL}(j)$.

[0135] The truth table for fractional control signals FRCT_ev and FRCT_od is shown below as Table 3.

TABLE-US-00003 TABLE 3 (signals FRCT_ev and FRCT_od) Row # parity RW(i - 1) RW(i) CL(j - 1) CL(j) FRCT_ev FRCT_od EVEN 1 0 0 1 1 n/a ODD 1 0 1 0 n/a 1

[0136] For all permutations and combinations of signals $\text{CL}(j-1)$, $\text{CL}(j)$, $\text{RW}(i-1)$ and $\text{RW}(i)$ other than those shown above in Table 3, $\text{FRCT_ev}=0$ and $\text{FRCT_od}=0$.

[0137] FIG. **5** is a table **548** showing partial values of a thermometer code, in accordance with some embodiments.

[0138] The table of FIG. **5** extends the example of FIG. **4A**. As columns col_0-col_7 of array **404** of FIG. **4A** represent corresponding SC_CDACs, Table **548** provides examples of how a thermometer code is used to adjust each of the SA_CDACs separately relative to the other SA_CDACs.

[0139] Table **548** includes: a first column (column 1) that shows the row-numbers assigned to the rows in Table **548**; a second column (column 2) that shows signals $\text{RW}(0)\text{-RW}(7)$ as corresponding binary digits concatenated to form the upper eight bits of the example thermometer code; a third column (column 3) that shows signal $\text{CL}(0)$ as a binary digit which forms the middle binary digit of the example thermometer code; a fourth column (column 4) that shows signals $\text{CL}(1)\text{-CL}(7)$ as corresponding binary digits concatenated to form the lower seven bits of the example thermometer code; a fifth column (column 5) which shows signal $\text{CL}(8)$ as a binary digit, where signal $\text{CL}(8)$ corresponds to an imaginary column col_8; and a sixth column (column 6) that shows which of columns col_0-col_7 corresponding has an SCC_cell **406B**(i,j), **406D**(i,j), **408C**(i,j) or **408E**(i,j) in the fractional state. In column 6 of Table **548**, the binary digit have the larger font represents the column which has SCC_cell **406B**(i,j), **406D**(i,j), **408C**(i,j) or **408E**(i,j) in the fractional state.

[0140] The full example code represented by Table **548** is formed by concatenating signals $\text{RW}(0)\text{-}$

RW(7) and CL(0)-CL(7) as corresponding binary digits representing the upper eight bits and lower eight bits of the example thermometer code in the context of the example of FIG. 4A. Table 548 is a partial listing of the permutations and combinations of values for the example thermometer code in the context of the example of FIG. 4A.

[0141] Column 2 of Table 548 is similar to column 3 of Table 324. Together, columns 3 and 4 of Table 548 are similar to column 4 of Table 324.

[0142] In row 12 of Table 548, for example, the value of the thermometer code is 1000 0000 0 000 001, which represents a base 10 value of 11.4375. The base 2 equivalent of 11.4375 (base 10) includes a base 2 integer portion of 0000 1011 and base 2 fractional portion of 0111.

[0143] FIG. 6 is a flowchart of method 600, in accordance with some embodiments.

[0144] Method 600 is an example of a method by which SA_CDACs, e.g., SA_CDAC(0)-SA_CDAC(N-1) of FIG. 1A, the SA_DCACs corresponding to array 204 of FIG. 2A, the SA_DCACs corresponding to array 404 of FIG. 4A, or the like, are operated. Method 600 includes blocks 604-616. Flow according to method 600 begins at block 604.

[0145] At block 604, delay stages are coupled in series as a delay chain. An example of the delay chain is delay chain 102 of FIG. 1A, or the like. Examples of the delay stages are delay stages ds(0)-ds(N-1) of FIG. 1A, or the like. Within block 604, flow proceeds to block 606.

[0146] At block 606, for each delay stage, an i_th driver device is coupled between input and output nodes of the i_th delay stage. Examples of the i_th driver device includes driver devices drv(0)-drv(N-1) of FIG. 1A, or the like. Examples of the input nodes include input nodes in (0)-in (N-1) correspondingly of driver devices drv(0)-drv(N-1) of FIG. 1A, or the like. Examples of the output nodes include output nodes out(0)-out(N-1) correspondingly of driver devices drv(0)-drv(N-1) of FIG. 1A, or the like. From block 606, flow proceeds to block 608.

[0147] At block 608, for each delay stage, the i_th SC_CDAC is coupled to the i_th output node of the i_th delay stage. Examples of the i_th SC_CDAC coupled to the i_th output node of the i_th delay stage include SC_DAC(0)-SC_CDAC(N-1) correspondingly coupled to output nodes out(0)-out(N-1) of delay stages ds(0)-ds(N-1) of FIG. 1A, or the like. Within block 608, flow proceeds to block 610.

[0148] At block 610, for each SCC_cell in the i_th SA-CDAC, the corresponding switch and the corresponding capacitor are coupled in series between the output node of the i_th delay stage and a first reference voltage. An example of the first reference voltage is VSS as in FIG. 2B-2E, 4B-4E, or the like. Examples of the output nodes include, again, output nodes out(0)-out(N-1) correspondingly of driver devices drv(0)-drv(N-1) of FIG. 1A, or the like. Examples of the SCC_cell includes SCC_cells 206B(i,j), 206D(i,j), 208C(i,j), 208E(i,j), 406B(i,j), 406D(i,j), 408C(i,j) and 408E(i,j), or the like. Examples of the corresponding switch include switches P(i,j) of SCC_cells 206B(i,j), 208C(i,j), 406B(i,j) and 408C(i,j), P(i,7) of SCC_cells 206D(i,j), 208E(i,j), 406D(i,j) and 408E(i,j), or the like. Examples of the corresponding capacitors include the capacitor in each of SCC_cells 206B(i,j), 206D(i,j), 208C(i,j), 208E(i,j), 406B(i,j), 406D(i,j), 408C(i,j) and 408E(i,j), or the like. From block 610, flow proceeds to exit block 608. From block 608, flow proceeds to exit block 604. From block 604, flow proceeds to block 612.

[0149] At block 612, each SA_CDAC is adjusted separately relative to another SA_CDAC. Examples of adjusting each SA_CDAC separately relative to other SA_CDACs include the examples of Table 324 of FIG. 3, Table 548 of FIG. 5, or the like. In some embodiments, at least one but fewer than all of the SA_CDACs are a type of CDAC this is not separately adjustable. Within block 612, flow proceeds to block 614.

[0150] At block 614, for each delay stage, none or some or all of the SCC_cells are selectively coupled to the output node of the i_th delay stage. Examples of none or some or all of the SCC_cells being selectively coupled to the output node of the i_th delay stage include the examples of Table 324 of FIG. 3, Table 548 of FIG. 5, or the like. Within block 614, flow proceeds to block 616.

[0151] At block **616**, for the SCC_cells in each delay stage, none or some or all of the capacitors in the SCC_cells are selectively coupled to the output node of the i_th delay stage. Examples of none or some or all of capacitors of the SCC_cells being selectively coupled to the output node of the i_th delay stage include the examples of Table **324** of FIG. **3**, Table **548** of FIG. **5**, or the like.

[0152] In some embodiments, method **600** further includes controlling the delay chain to function as a controllable delay line including: receiving a clock signal at the input node of the first delay stage; and propagating the clock signal serially through the delay chain such that a signal on the output node of the last delay stage represents a delayed version of the clock signal. An example of the clock signal is clock signal CLK of FIG. **1D**, or the like. An example of the delayed version of the clock signal includes signal CLK_dlyd of FIG. **1D** which represents a delay version of signal CLK, or the like.

[0153] In some embodiments, method **600** further includes controlling the delay chain to function as a controllable oscillator including: feedback-coupling the output node of the last delay stage to the input node of the first delay stage. An example of the controllable oscillator is oscillator **114** of FIG. **1C**, or the like. An example of feedback-coupling the output node of the last delay stage to the input node of the first delay stage includes feedback coupling **116** of FIG. **1C**, or the like.

[0154] FIG. **7** is a flowchart (flow diagram) of a method **700** of manufacturing a system or device, in accordance with some embodiments.

[0155] Method **700** is implementable, for example, using EDA system **800** (FIG. **8**, discussed below) and an IC manufacturing system **900** (FIG. **9**, discussed below), in accordance with some embodiments. Examples of a system or device which can be manufactured according to method **700** include the systems or devices disclosed herein, or the like.

[0156] In FIG. **7**, the method of flowchart **700** includes blocks **702-704**. At block **702**, a layout diagram is generated which, among other things, includes one or more layout diagrams corresponding to one or more of the systems or devices disclosed herein, or the like. Block **702** is implementable, for example, using EDA system **800** (FIG. **8**, discussed below), in accordance with some embodiments. From block **702**, flow proceeds to block **704**.

[0157] At block **704**, based on the layout diagram, at least one of (A) one or more photolithographic exposures are made or (b) one or more photolithography masks are fabricated or (C) one or more components in a layer of a device, e.g., a device is fabricated. See discussion below of IC manufacturing system **900** in FIG. **9** below.

[0158] FIG. **8** is a block diagram of an electronic design automation (EDA) system **800** in accordance with some embodiments.

[0159] In some embodiments, EDA system **800** includes an automatic placement and routing (APR) system. In some embodiments, EDA system **800** is a general purpose computing device including a hardware processor **802** and a non-transitory, computer-readable storage medium **804**. Storage medium **804**, amongst other things, is encoded with, i.e., stores, computer program code **806**, i.e., a set of executable instructions. Execution of instructions **806** by hardware processor **802** represents (at least in part) an EDA tool which implements a portion or all of, e.g., methods of generating corresponding to the systems or devices disclosed herein, or the like, in accordance with one or more embodiments (hereinafter, the noted processes and/or methods).

[0160] Storage medium **804**, amongst other things, stores layout diagrams **811** such as the layout diagrams disclosed herein, other the like.

[0161] Processor **802** is electrically coupled to computer-readable storage medium **804** via a bus **808**. Processor **802** is further electrically coupled to an I/O interface **810** by a bus **808**. A network interface **812** is further electrically connected to processor **802** via bus **808**. Network interface **812** is connected to a network **814**, so that processor **802** and computer-readable storage medium **804** are capable of connecting to external elements via network **814**. Processor **802** is configured to execute computer program code **806** encoded in computer-readable storage medium **804** in order to cause EDA system **800** to be usable for performing a portion or all of the noted processes and/or

methods. In one or more embodiments, processor **802** is a central processing unit (CPU), a multi-processor, a distributed processing system, an application specific integrated circuit (ASIC), and/or a suitable processing unit.

[0162] In one or more embodiments, computer-readable storage medium **804** is an electronic, magnetic, optical, electromagnetic, infrared, and/or a semiconductor system (or apparatus or device). For example, computer-readable storage medium **804** includes a semiconductor or solid-state memory, a magnetic tape, a removable computer diskette, a random access memory (RAM), a read-only memory (ROM), a rigid magnetic disk, and/or an optical disk. In one or more embodiments using optical disks, computer-readable storage medium **804** includes a compact disk-read only memory (CD-ROM), a compact disk-read/write (CD-R/W), and/or a digital video disc (DVD).

[0163] In one or more embodiments, storage medium **804** stores computer program code **806** configured to cause EDA system **800** (where such execution represents (at least in part) the EDA tool) to be usable for performing a portion or all of the noted processes and/or methods. In one or more embodiments, storage medium **804** further stores information which facilitates performing a portion or all of the noted processes and/or methods. In one or more embodiments, storage medium **804** stores library **807** of standard cells including such standard cells as disclosed herein. In some embodiments, storage medium **804** stores one or more layout diagrams **811**.

[0164] EDA system **800** includes I/O interface **810**. I/O interface **810** is coupled to external circuitry. In one or more embodiments, I/O interface **810** includes a keyboard, keypad, mouse, trackball, trackpad, touchscreen, and/or cursor direction keys for communicating information and commands to processor **802**.

[0165] EDA system **800** further includes network interface **812** coupled to processor **802**. Network interface **812** allows EDA system **800** to communicate with network **814**, to which one or more other computer systems are connected. Network interface **812** includes wireless network interfaces such as BLUETOOTH, WIFI, WIMAX, GPRS, or WCDMA; or wired network interfaces such as ETHERNET, USB, or IEEE-1364. In one or more embodiments, a portion or all of noted processes and/or methods, is implemented in two or more EDA systems **800**.

[0166] EDA system **800** is configured to receive information through I/O interface **810**. The information received through I/O interface **810** includes one or more of instructions, data, design rules, libraries of standard cells, and/or other parameters for processing by processor **802**. The information is transferred to processor **802** via bus **808**. EDA system **800** is configured to receive information related to a user interface (UI) through I/O interface **810**. The information is stored in computer-readable medium **804** as UI **842**.

[0167] In some embodiments, a portion or all of the noted processes and/or methods is implemented as a standalone software application for execution by a processor. In some embodiments, a portion or all of the noted processes and/or methods is implemented as a software application that is a part of an additional software application. In some embodiments, a portion or all of the noted processes and/or methods is implemented as a plug-in to a software application. In some embodiments, at least one of the noted processes and/or methods is implemented as a software application that is a portion of an EDA tool. In some embodiments, a portion or all of the noted processes and/or methods is implemented as a software application that is used by EDA system **800**. In some embodiments, a layout which includes standard cells is generated using a tool such as VIRTUOSO® available from CADENCE DESIGN SYSTEMS, Inc., or another suitable layout generating tool.

[0168] In some embodiments, the processes are realized as functions of a program stored in a non-transitory computer readable recording medium. Examples of a non-transitory computer readable recording medium include, but are not limited to, external/removable and/or internal/built-in storage or memory unit, e.g., one or more of an optical disk, such as a DVD, a magnetic disk, such as a hard disk, a semiconductor memory, such as a ROM, a RAM, a memory card, and the like.

[0169] FIG. 9 is a block diagram of an integrated circuit (IC) manufacturing system **900**, and an IC manufacturing flow associated therewith, in accordance with some embodiments.

[0170] In some embodiments, based on the layout diagram generated by block **602** of FIG. 6, the IC manufacturing system **900** implements block **704** of FIG. 7 wherein at least one of (A) one or more semiconductor masks or (B) at least one component in a layer of an inchoate semiconductor integrated circuit is fabricated using manufacturing system **900**. In some embodiments, the IC manufacturing system **900** implements the flowcharts of FIGS. 7A-7B.

[0171] In FIG. 9, IC manufacturing system **900** includes entities, such as a design house **920**, a mask house **930**, and an IC manufacturer/fabricator ("fab") **950**, that interact with one another in the design, development, and manufacturing cycles and/or services related to manufacturing an IC device **960**. The entities in system **900** are connected by a communications network. In some embodiments, the communications network is a single network. In some embodiments, the communications network is a variety of different networks, such as an intranet and the Internet. The communications network includes wired and/or wireless communication channels. Each entity interacts with one or more of the other entities and supplies services to and/or receives services from one or more of the other entities. In some embodiments, two or more of design house **920**, mask house **930**, and IC fab **950** is owned by a single larger company. In some embodiments, two or more of design house **920**, mask house **930**, and IC fab **950** coexist in a common facility and use common resources.

[0172] Design house (or design team) **920** generates an IC design layout **922**. IC design layout **922** includes various geometrical patterns designed for an IC device **960**. The geometrical patterns correspond to patterns of metal, oxide, or semiconductor layers that make up the various components of IC device **960** to be fabricated. The various layers combine to form various IC features. For example, a portion of IC design layout **922** includes various IC features, such as an active region, gate terminal, source and drain, metal lines or vias of an interlayer interconnection, and openings for bonding pads, to be formed in a semiconductor substrate (such as a silicon wafer) and various material layers disposed on the semiconductor substrate. Source/drain region(s) may refer to a source or a drain, individually or collectively, dependent upon the context. Design house **920** implements a proper design procedure to form IC design layout **922**. The design procedure includes one or more of logic design, physical design or place and route. IC design layout **922** is presented in one or more data files having information of the geometrical patterns. For example, IC design layout **922** is expressed in a GDSII file format or DFII file format.

[0173] Mask house **930** includes data preparation **932** and mask fabrication **934**. Mask house **930** uses IC design layout **922** to manufacture one or more masks **935** to be used for fabricating the various layers of IC device **960** according to IC design layout **922**. Mask house **930** performs mask data preparation **932**, where IC design layout **922** is translated into a representative data file ("RDF"). Mask data preparation **932** supplies the RDF to mask fabrication **934**. Mask fabrication **934** includes a mask writer. A mask writer converts the RDF to an image on a substrate, such as a mask (reticle) or a semiconductor wafer. The design layout is manipulated by mask data preparation **932** to comply with particular characteristics of the mask writer and/or requirements of IC fab **950**. In FIG. 9, mask data preparation **932**, mask fabrication **934**, and mask **935** are illustrated as separate elements. In some embodiments, mask data preparation **932** and mask fabrication **934** are collectively referred to as mask data preparation.

[0174] In some embodiments, mask data preparation **932** includes optical proximity correction (OPC) which uses lithography enhancement techniques to compensate for image errors, such as those that can arise from diffraction, interference, other process effects and the like. OPC adjusts IC design layout **922**. In some embodiments, mask data preparation **932** includes further resolution enhancement techniques (RET), such as off-axis illumination, sub-resolution adjust features, phase-shifting masks, other suitable techniques, and the like or combinations thereof. In some embodiments, inverse lithography technology (ILT) is further used, which treats OPC as an inverse

imaging problem.

[0175] In some embodiments, mask data preparation **932** includes a mask rule checker (MRC) that checks the IC design layout that has undergone processes in OPC with a set of mask creation rules which contain certain geometric and/or connectivity restrictions to ensure sufficient margins, to account for variability in semiconductor manufacturing processes, and the like. In some embodiments, the MRC modifies the IC design layout to compensate for limitations during mask fabrication **934**, which may undo part of the modifications performed by OPC in order to meet mask creation rules.

[0176] In some embodiments, mask data preparation **932** includes lithography process checking (LPC) that simulates processing that will be implemented by IC fab **950** to fabricate IC device **960**. LPC simulates this processing based on IC design layout **922** to fabricate a simulated manufactured device, such as IC device **960**. The processing parameters in LPC simulation can include parameters associated with various processes of the IC manufacturing cycle, parameters associated with tools used for manufacturing the IC, and/or other aspects of the manufacturing process. LPC takes into account various factors, such as aerial image contrast, depth of focus (“DOF”), mask error enhancement factor (“MEEF”), other suitable factors, and the like or combinations thereof. In some embodiments, after a simulated manufactured device has been fabricated by LPC, if the simulated device is not close enough in shape to satisfy design rules, OPC and/or MRC are repeated to further refine IC design layout **922**.

[0177] The above description of mask data preparation **932** has been simplified for the purposes of clarity. In some embodiments, mask data preparation **932** includes additional features such as a logic operation (LOP) to modify the IC design layout according to manufacturing rules. Additionally, the processes applied to IC design layout **922** during data preparation **932** may be executed in a variety of different orders.

[0178] After mask data preparation **932** and during mask fabrication **934**, a mask **935** or a group of masks **935** are fabricated based on the modified IC design layout. In some embodiments, an electron-beam (e-beam) or a mechanism of multiple e-beams is used to form a pattern on a mask (photomask or reticle) based on the modified IC design layout. The masks are formed in various technologies. In some embodiments, the mask is formed using binary technology. In some embodiments, a mask pattern includes opaque regions and transparent regions. A radiation beam, such as an ultraviolet (UV) beam, used to expose the image sensitive material layer (e.g., photoresist) which has been coated on a wafer, is blocked by the opaque region and transmits through the transparent regions. In one example, a binary mask includes a transparent substrate (e.g., fused quartz) and an opaque material (e.g., chromium) coated in the opaque regions of the mask. In another example, the mask is formed using a phase shift technology. In the phase shift mask (PSM), various features in the pattern formed on the mask are configured to have proper phase difference to enhance the resolution and imaging quality. In various examples, the phase shift mask is an attenuated PSM or alternating PSM. The mask(s) generated by mask fabrication **934** is used in a variety of processes. For example, such a mask(s) is used in an ion implantation process to form various doped regions in the semiconductor wafer, in an etching process to form various etching regions in the semiconductor wafer, and/or in other suitable processes.

[0179] IC fab **950** is an IC fabrication business that includes one or more manufacturing facilities for the fabrication of a variety of different IC products. In some embodiments, IC fab **950** is a semiconductor foundry. For example, there may be a manufacturing facility for the front end fabrication of a plurality of IC products (front-end-of-line (FEOL) fabrication), while a second manufacturing facility may supply the back end fabrication for the interconnection and packaging of the IC products (back-end-of-line (BEOL) fabrication), and a third manufacturing facility may supply other services for the foundry business.

[0180] IC fab **950** uses mask (or masks) **935** fabricated by mask house **930** to fabricate IC device **960** using fabrication tools **952**. Thus, IC fab **950** at least indirectly uses IC design layout **922** to

fabricate IC device **960**. In some embodiments, a semiconductor wafer **953** is fabricated by IC fab **950** using mask (or masks) **935** to form IC device **960**. Semiconductor wafer **953** includes a silicon substrate or other proper substrate having material layers formed thereon. Semiconductor wafer further includes one or more of various doped regions, dielectric features, multilevel interconnects, and the like (formed at subsequent manufacturing steps).

[0181] In some embodiments, a delay chain includes: delay stages coupled in series, each delay stage including: a driver device coupled between an input node and an output node of the delay stage; and a capacitor digital-to-analog converter (CDAC) coupled to the output node of the delay stage; at least one of the CDACs being a separately adjustable (SA) type of CDAC (SA_CDAC) that is separately adjustable relative to at least another one of the CDACs; and a controller configured to adjust the at least one CDAC.

[0182] In some embodiments, the delay chain further includes: the delay stages include a first delay stage and a last delay stage; and the delay chain is configured as a controllable delay line such that: the first delay stage is configured to receive a clock signal at the input node thereof; and the last delay stage is configured to generate a delayed version of the clock signal at the output node thereof.

[0183] In some embodiments, the delay stages include a first delay stage and a last delay stage; and the delay chain is configured as a controllable oscillator such that: at the input node of the first delay stage, the first delay stage is configured to receive a signal fed back from the output node of the last delay stage.

[0184] In some embodiments, each of the CDACs is an SA_CDAC; and the controller is further configured to adjust correspondingly each of the SA_CDACs.

[0185] In some embodiments, the at least one SA_CDAC is a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells), each SCC_cell being separately controllable to be coupled to the output node of the corresponding delay stage relative to other ones of the SCC_cells, the one-dimensional array including rows and a single column, and each row including a single SCC_cell; and the controller is further configured to selectively couple none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage.

[0186] In some embodiments, for each of the at least one SA_CDAC, each SCC_cell includes: a switch and a capacitor coupled in series between the output node of the corresponding delay stage and a first reference voltage; and a decoder configured to receive corresponding addressing signals and generate a switch-state control signal to control the switch to be in an open state or a closed state.

[0187] In some embodiments, the decoders have corresponding row-number-parity-specific configurations such that the decoders in odd rows or even rows have a first configuration and conversely the decoders in even rows or odd rows have a second configuration different than the first configuration.

[0188] In some embodiments, each of the CDACs is an SA_CDAC; a count of the delay stages is M , where M is an integer and $2 \leq M$, such that a count of the one-dimensional arrays is M ; a count of the rows in each of the one-dimensional arrays including of SCC_cells is N , where N is an integer and $2 \leq N$; together the one-dimensional arrays including SCC_cells represent an $N \times M$ array of SCC_cells; and the controller is further configured to adjust correspondingly each of the SA_CDACs according to a thermometer encoding scheme for the $N \times M$ array; and for each of the SA_CDACs, the controller is further configured to generate the addressing signals according to the thermometer encoding scheme such that correspondingly resultant switch-state control signals control none or some or all of the capacitors in the corresponding one-dimensional arrays to be selectively coupled to the output node of the corresponding delay stage.

[0189] In some embodiments, the delay chain further includes: for each SCC_cell that is in a thermometer-code-ON state according to the thermometer encoding scheme, the controller is further configured to generate the addressing signals such that a correspondingly resultant switch-

state control signal controls the corresponding switch to be in the closed state according to a first duty cycle (DCY_1) of 100% such that $DCY_1=100\%$; and for a selected SCC_cell that otherwise would be in a thermometer-code-OFF state according to the thermometer encoding scheme, the controller is further configured to generate the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a second duty cycle (DCY_2) in a range $0\% < DCY_2 < 100\%$.

[0190] In some embodiments, a method (of operating a delay chain) includes: coupling delay stages in series as representing the delay chain, the coupling delay stages including, for each of the delay stages, the following: coupling a driver device between an input node and an output node of the delay stage; and coupling a capacitor digital-to-analog converter (CDAC) to the output node of the delay stage wherein for at least one of the delay stages, the corresponding CDAC being a separately adjustable (SA) type of CDAC (SA_CDAC) that is separately adjustable relative to the CDAC of at least another one of the CDACs; and adjusting at least one CDAC separately relative to at least another one of the delay stages; and adjusting at least one CDAC separately relative to at least another one of the CDACs.

[0191] In some embodiments, each of the CDACs is an SA_CDAC; and the method includes adjusting correspondingly each of the SA_CDACs.

[0192] In some embodiments, the at least one SA_CDAC is a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells), each SCC_cell being separately controllable to be coupled to the output node of the corresponding delay stage relative to other ones of the SCC_cells, the one-dimensional array including rows and a single column, and each row including a single SCC_cell; and the method further includes: selectively coupling none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage.

[0193] In some embodiments, for each of the at least one SA_CDAC, each SCC_cell includes a switch and a capacitor; and the method further includes, for each SCC_cell, the following: coupling the switch and the capacitor in series between the output node of the corresponding delay stage and a first reference voltage; generating a switch-state control signal to control the switch to be in an open state or a closed state based on corresponding addressing signals.

[0194] In some embodiments, the generating a switch-state control signal includes: before the generating a switch-state control signal, performing row-number-parity-specific manipulation of the corresponding addressing signals.

[0195] In some embodiments, the performing row-number-parity-specific manipulation includes: in even rows or odd rows, inverting at least one of the addressing signals before the generating a switch-state control signal; and, in odd rows or even rows, conversely not inverting any of the addressing signals before the generating a switch-state control signal.

[0196] In some embodiments, the generating a switch-state control signal includes: performing a logical AND-OR-INVERT (AOI) operation on the corresponding addressing signals.

[0197] In some embodiments, each of the CDACs is an SA_CDAC; a count of the delay stages is M, where M is an integer and $2 \leq M$, such that a count of the one-dimensional arrays is M; a count of the rows in each of the one-dimensional arrays including SCC_cells is N, where N is an integer and $2 \leq N$; together the one-dimensional arrays including SCC_cells represent an $N \times M$ array of SCC_cells; and the method further includes adjusting correspondingly each of the SA_CDACs according to a thermometer encoding scheme for the $N \times M$ array including: generating the addressing signals according to the thermometer encoding scheme such correspondingly resultant switch-state control signals control that none or some or all of the capacitors in the corresponding one-dimensional arrays are selectively coupled to the output node of the corresponding delay stage.

[0198] In some embodiments, for each SCC_cell that is in a thermometer-code-ON state according to the thermometer encoding scheme, the generating the addressing signals includes configuring the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a first duty cycle (DCY_1) of 100%

such that DCY_1=100%; and for a selected SCC_cell that otherwise would be in a thermometer-code-OFF state according to the thermometer encoding scheme, the generating the addressing signals includes configuring the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a second duty cycle (DCY_2) in a range $0\% < \text{DCY}_2 < 100\%$.

[0199] In some embodiments, a delay chain includes: delay stages coupled in series, each delay stage including a driver device coupled between an input node and an output node of the delay stage, and a capacitor digital-to-analog converter (CDAC) coupled to the output node of the delay stage; at least one of the CDACs being a separately adjustable (SA) type of CDAC (SA_CDAC) that is separately adjustable relative to at least another one of the CDACs; the at least one SA_CDAC being a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells), each SCC_cell being separately controllable to be coupled to the output node of the corresponding delay stage relative to other ones of the SCC_cells, the one-dimensional array including rows and a single column, and each row including a single SCC_cell; and a controller configured to adjust the at least one CDAC by selectively coupling none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage.

[0200] In some embodiments, for each of the at least one SA_CDAC, each SCC_cell includes: a switch and a capacitor coupled in series between the output node of the corresponding delay stage and a first reference voltage; and a decoder configured to receive corresponding addressing signals and generate a switch-state control signal to control the switch to be in an open state or a closed state.

[0201] It will be readily seen by one of ordinary skill in the art that one or more of the disclosed embodiments fulfill one or more of the advantages set forth above. After reading the foregoing specification, one of ordinary skill will be able to affect various changes, substitutions of equivalents and various other embodiments as broadly disclosed herein. It is therefore intended that the protection granted hereon be limited only by the definition contained in the appended claims and equivalents thereof.

Claims

1. A delay chain comprising: delay stages coupled in series, each delay stage including: a driver device coupled between an input node and an output node of the delay stage; and a capacitor digital-to-analog converter (CDAC) coupled to the output node of the delay stage; at least one of the CDACs being a separately adjustable (SA) type of CDAC (SA_CDAC) that is separately adjustable relative to at least another one of the CDACs; and a controller configured to adjust the at least one CDAC.
2. The delay chain of claim 1, wherein: the delay stages include a first delay stage and a last delay stage; and the delay chain is configured as a controllable delay line such that: the first delay stage is configured to receive a clock signal at the input node thereof; and the last delay stage is configured to generate a delayed version of the clock signal at the output node thereof.
3. The delay chain of claim 1, wherein: the delay stages include a first delay stage and a last delay stage; and the delay chain is configured as a controllable oscillator such that: at the input node of the first delay stage, the first delay stage is configured to receive a signal fed back from the output node of the last delay stage.
4. The delay chain of claim 1, wherein: each of the CDACs is an SA_CDAC; and the controller is further configured to adjust correspondingly each of the SA_CDACs.
5. The delay chain of claim 1, wherein: the at least one SA_CDAC is a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells), each SCC_cell being separately controllable to be coupled to the output node of the corresponding delay stage relative to other ones of the SCC_cells, the one-dimensional array including rows and a single column, and each row

including a single SCC_cell; and the controller is further configured to selectively couple none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage.

6. The delay chain of claim 5, wherein: for each of the at least one SA_CDAC, each SCC_cell includes: a switch and a capacitor coupled in series between the output node of the corresponding delay stage and a first reference voltage; and a decoder configured to receive corresponding addressing signals and generate a switch-state control signal to control the switch to be in an open state or a closed state.

7. The delay chain of claim 6, wherein: the decoders have corresponding row-number-parity-specific configurations such that the decoders in odd rows or even rows have a first configuration and conversely the decoders in even rows or odd rows have a second configuration different than the first configuration.

8. The delay chain of claim 6, wherein: each of the CDACs is an SA_CDAC; a count of the delay stages is M, where M is an integer and $2 < M$, such that a count of the one-dimensional arrays is M; a count of the rows in each of the one-dimensional arrays including SCC_cells is N, where N is an integer and $2 \leq N$; together the one-dimensional arrays including SCC_cells represent an $N \times M$ array of SCC_cells; and the controller is further configured to adjust correspondingly each of the SA_CDACs according to a thermometer encoding scheme for the $N \times M$ array; and for each of the SA_CDACs, the controller is further configured to generate the addressing signals according to the thermometer encoding scheme such that correspondingly resultant switch-state control signals control none or some or all of the capacitors in the corresponding one-dimensional arrays to be selectively coupled to the output node of the corresponding delay stage.

9. The delay chain of claim 8, wherein: for each SCC_cell that is in a thermometer-code-ON state according to the thermometer encoding scheme, the controller is further configured to generate the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a first duty cycle (DCY_1) of 100% such that $DCY_1 = 100\%$; and for a selected SCC_cell that otherwise would be in a thermometer-code-OFF state according to the thermometer encoding scheme, the controller is further configured to generate the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a second duty cycle (DCY_2) in a range $0\% < DCY_2 < 100\%$.

10. A method of operating a delay chain, the method comprising: coupling delay stages in series as representing the delay chain, the coupling delay stages including, for each of the delay stages, the following: coupling a driver device between an input node and an output node of the delay stage; and coupling a capacitor digital-to-analog converter (CDAC) to the output node of the delay stage, wherein, for at least one of the delay stages, the corresponding CDAC being a separately adjustable (SA) type of CDAC(SA_CDAC) that is separately adjustable relative to the CDAC of at least another one of the delay stages; and adjusting at least one CDAC separately relative to at least another one of the CDACs.

11. The method of claim 10, wherein: each of the CDACs is an SA_CDAC; and the method further comprises: adjusting correspondingly each of the SA_CDACs.

12. The method of claim 10, wherein: the at least one SA_CDAC is a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells), each SCC_cell being separately controllable to be coupled to the output node of the corresponding delay stage relative to other ones of the SCC_cells, the one-dimensional array including rows and a single column, and each row including a single SCC_cell; and the method further comprises: selectively coupling none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage.

13. The method of claim 12, wherein: for each of the at least one SA_CDAC, each SCC_cell includes a switch and a capacitor; and the method further comprises, for each SCC_cell, the

following: coupling the switch and the capacitor in series between the output node of the corresponding delay stage and a first reference voltage; and generating a switch-state control signal to control the switch to be in an open state or a closed state based on corresponding addressing signals.

14. The method of claim 13, wherein: the generating a switch-state control signal includes: before the generating a switch-state control signal, performing row-number-parity-specific manipulation of the corresponding addressing signals.

15. The method of claim 14, wherein: the performing row-number-parity-specific manipulation includes: in even rows or odd rows, inverting at least one of the addressing signals before the generating a switch-state control signal; and in odd rows or even rows, conversely not inverting any of the addressing signals before the generating a switch-state control signal.

16. The method of claim 13, wherein: the generating a switch-state control signal includes: performing a logical AND-OR-INVERT (AOI) operation on the corresponding addressing signals.

17. The method of claim 13, wherein: each of the CDACs is an SA_CDAC; a count of the delay stages is M , where M is an integer and $2 \leq M$, such that a count of the one-dimensional arrays is M ; a count of the rows in each of the one-dimensional arrays including SCC_cells is N , where N is an integer and $2 \leq N$; together the one-dimensional arrays including SCC_cells represent an $N \times M$ array of SCC_cells; and the method further comprises: adjusting correspondingly each of the SA_CDACs according to a thermometer encoding scheme for the $N \times M$ array including: generating the addressing signals according to the thermometer encoding scheme such that correspondingly resultant switch-state control signals control that none or some or all of the capacitors in the corresponding one-dimensional arrays are selectively coupled to the output node of the corresponding delay stage.

18. The method of claim 17, wherein: for each SCC_cell that is in a thermometer-code-ON state according to the thermometer encoding scheme, the generating the addressing signals includes: configuring the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a first duty cycle (DCY_1) of 100% such that $DCY_1 = 100\%$; and for a selected SCC_cell that otherwise would be in a thermometer-code-OFF state according to the thermometer encoding scheme, the generating the addressing signals includes: configuring the addressing signals such that a correspondingly resultant switch-state control signal controls the corresponding switch to be in the closed state according to a second duty cycle (DCY_2) in a range $0\% < DCY_2 < 100\%$.

19. A delay chain comprising: delay stages coupled in series, each delay stage including: a driver device coupled between an input node and an output node of the delay stage; and a capacitor digital-to-analog converter (CDAC) coupled to the output node of the delay stage; at least one of the CDACs being a separately adjustable (SA) type of CDAC(SA_CDAC) that is separately adjustable relative to at least another one of the CDACs; the at least one SA_CDAC being a one-dimensional array of separately controllable capacitance (SCC) cells (SCC_cells), each SCC_cell being separately controllable to be coupled to the output node of the corresponding delay stage relative to other ones of the SCC_cells, the one-dimensional array including rows and a single column, and each row including a single SCC_cell; and a controller configured to adjust the at least one CDAC by selectively coupling none or some or all of the SCC_cells in the at least one SA_CDAC to the output node of the corresponding delay stage.

20. The delay chain of claim 19, wherein: for each of the at least one SA_CDAC, each SCC_cell includes: a switch and a capacitor coupled in series between the output node of the corresponding delay stage and a first reference voltage; and a decoder configured to receive corresponding addressing signals and generate a switch-state control signal to control the switch to be in an open state or a closed state.
