# **Discrete Time Fourier Transformation (DTFT)**

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### 1. DFT and DTFT

- DTFT is the Fourier transform. Useful for conceptual, but not Matlab friendly (infinitely-lority vector), We will derive DTFT as the limit of the DFT as the signal length  $N \to \infty$   $\omega = \frac{2\pi}{N} k$ · DTFT is the Fourier transform of choice for analyzing infinite-length signals and systems

$$\omega = rac{2\pi}{N} k$$

#### The Centered DFT

■ Both x[n] and X[k] can be interpreted as periodic with period N, so we will shift the intervals of interest in time and frequency to be centered around n,k=0

$$-\frac{N}{2} \le n, k \le \frac{N}{2} - 1$$

■ The modified forward and inverse DFT formulas are

$$X_u[k] \ = \ \sum_{n=-N/2}^{N/2-1} x[n] \, e^{-j\frac{2\pi}{N}kn}, \qquad -\frac{N}{2} \ \le \ k \ \le \ \frac{N}{2}-1$$

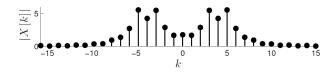
$$x[n] \; = \; \frac{1}{N} \sum_{k=-N/2}^{N/2-1} X_u[k] \, e^{j \frac{2\pi}{N} k n} \qquad - \frac{N}{2} \; \leq \; n \; \leq \; \frac{N}{2} - 1$$

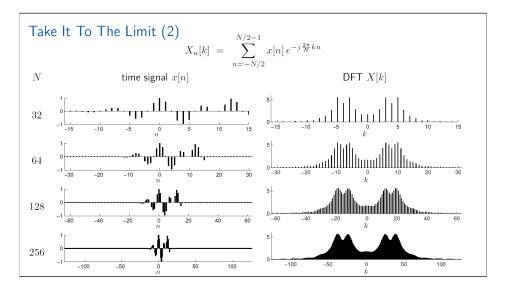
### Take It To The Limit (1)

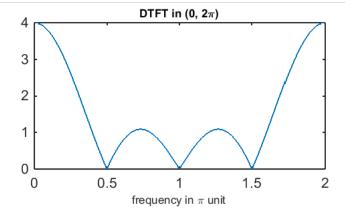
$$X_u[k] \; = \; \sum_{n=-N/2}^{N/2-1} x[n] \, e^{-j\frac{2\pi}{N}kn}, \qquad -\frac{N}{2} \; \le \; k \; \le \; \frac{N}{2}-1$$

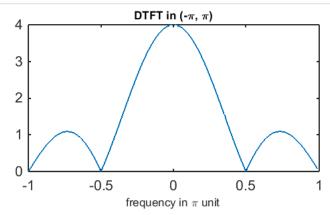
- lacktriangle Let the signal length N increase towards  $\infty$  and study what happens to  $X_u[k]$
- lacktriangle Key fact: No matter how large N grows, the frequencies of the DFT sinusoids remain in the interval

$$-\pi \leq \omega_k = \frac{2\pi}{N}k < \pi$$









Out[8]:

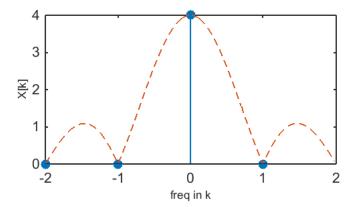
```
In [14]:
```

```
x = [1,1,1,1];
N = length(x);

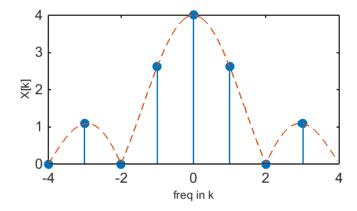
k = [0:N/2-1 -N/2:-1];
ks = fftshift(k);

X = dft(x,N);
Xs = fftshift(X);

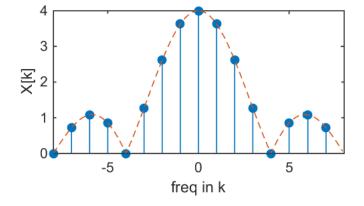
stem(ks,abs(Xs),'filled'),
xlabel('freq in k','fontsize',8), ylabel('X[k]','fontsize',8),
xlim([-N/2,N/2]), hold on
plot(ws*N/(2*pi),abs(Xdffts),'--'), hold off
```



Out[14]:



Out[15]:



Out[5]:

Out[2]: Undefined function or variable 'ws'.

### Discrete Time Fourier Transform (Forward)

■ As  $N \to \infty$ , the forward DFT converges to a function of the **continuous frequency variable**  $\omega$  that we will call the **forward discrete time Fourier transform** (DTFT)

$$\sum_{n=-N/2}^{N/2-1} x[n] \, e^{-j\frac{2\pi}{N}kn} \, \longrightarrow \, \sum_{n=-\infty}^{\infty} x[n] \, e^{-j\omega n} \ = \ X(\omega), \qquad -\pi \le \omega < \pi$$

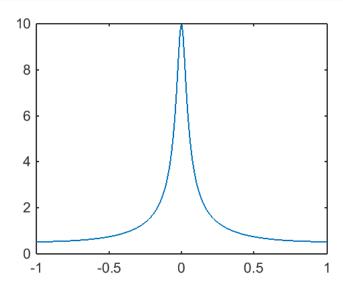
■ Recall: Inner product for infinite-length signals

$$\langle x, y \rangle = \sum_{n=-\infty}^{\infty} x[n] y[n]^*$$

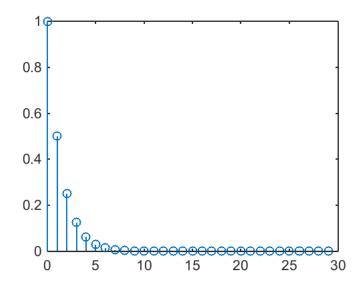
■ Analysis interpretation: The value of the DTFT  $X(\omega)$  at frequency  $\omega$  measures the similarity of the infinite-length signal x[n] to the infinite-length sinusoid  $e^{j\omega n}$ 

## 1.1. Analytic form of DTFT (exact transformation)

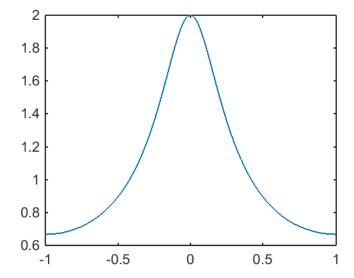
$$x[n] = (0.5)^n u[n] \qquad \leftrightarrow \qquad X(e^{j\omega}) = rac{e^{j\omega}}{e^{j\omega} - 0.5}$$



Out[4]:



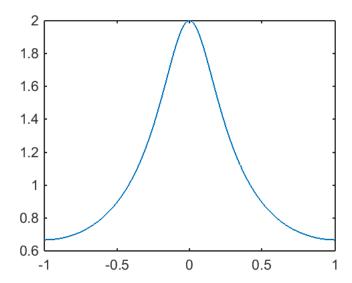
Out[6]:



Out[9]:

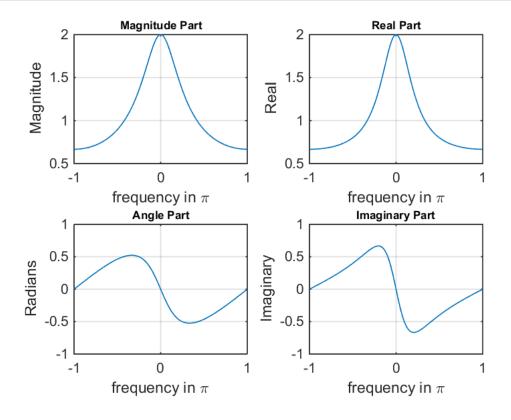
In [11]:

X = dtft(x,n,w);
plot(w/pi,abs(X))



Out[11]:

```
%plot -s 800,600
N = 500;
w2 = [-1:1/N:1]*pi;
w = w2(1:end-1);
% closed form of DTFT
X = \exp(1j*w)./(\exp(1j*w) - 0.5*ones(size(w)));
magX = abs(X); angX = angle(X);
realX = real(X);
                   imagX = imag(X);
subplot(2,2,1); plot(w/pi,magX); grid
xlabel('frequency in \pi'); title('Magnitude Part', 'fontsize',8); ylabel('Magnitude')
subplot(2,2,3); plot(w/pi,angX); grid
xlabel('frequency in \pi'); title('Angle Part', 'fontsize',8); ylabel('Radians')
subplot(2,2,2); plot(w/pi,realX); grid
xlabel('frequency in \pi'); title('Real Part', 'fontsize',8); ylabel('Real')
subplot(2,2,4); plot(w/pi,imagX); grid
xlabel('frequency in \pi'); title('Imaginary Part', 'fontsize',8); ylabel('Imaginary')
```

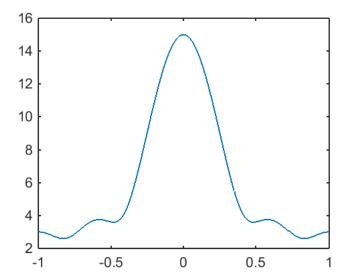


Out[7]:

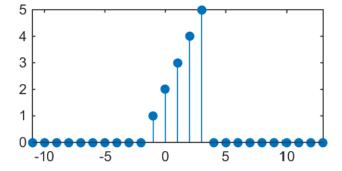
In [7]:

## 1.2. DTFT of a numerical computation (using a definition)

### Example 1



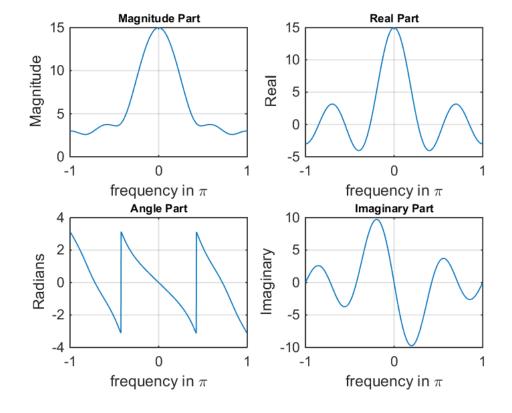
Out[13]:



Out[8]:

```
In [9]:
```

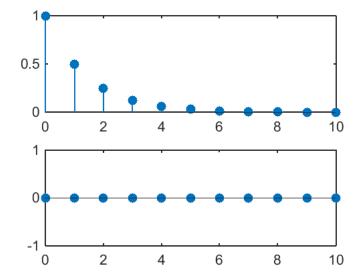
```
%plot -s 800,600
%% DTFT code
N = 500;
k = -N:N-1; w = (pi/N)*k;
X = x * (exp(-1j*pi/500)).^{(n'*k)};
                                   % DTFT using matrix-vector product
% plots
magX = abs(X); angX = angle(X);
realX = real(X);
                 imagX = imag(X);
subplot(2,2,1); plot(w/pi,magX); grid
xlabel('frequency in \pi'); title('Magnitude Part', 'fontsize',8); ylabel('Magnitude')
subplot(2,2,3); plot(w/pi,angX); grid
xlabel('frequency in \pi'); title('Angle Part', 'fontsize',8); ylabel('Radians')
subplot(2,2,2); plot(w/pi,realX); grid
xlabel('frequency in \pi'); title('Real Part','fontsize',8); ylabel('Real')
subplot(2,2,4); plot(w/pi,imagX); grid
xlabel('frequency in \pi'); title('Imaginary Part','fontsize',8); ylabel('Imaginary')
```



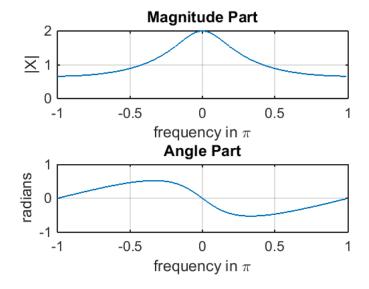
Out[9]:

Example 2

$$x[n] = \left(0.5e^{(j\pi/3)}
ight)^n$$



Out[20]:



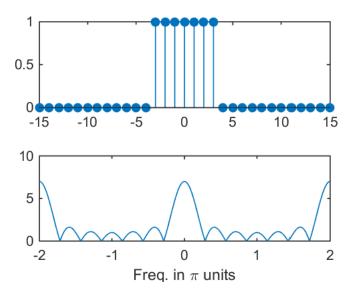
Out[21]:

# 2. Numerical DTFT Computation

```
function X = dtft(x,n,w)
% X = dtft(x, n, w)
%
% X = DTFT values computed at w frequencies
% x = finite duration sequence over n
% n = sample position vector
% w = frequency location vector
X = exp(-1j*(w'*n))*x';
end
```

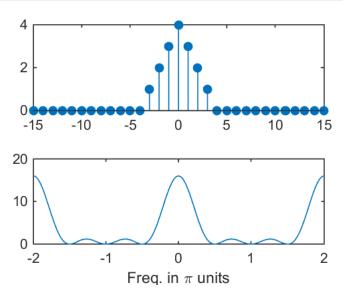
### DTFT of the unit pulse

$$p[n] \ = \ \begin{cases} 1 & -M \le n \le M \\ 0 & ext{otherwise} \end{cases}$$
  $P(\omega) \ = \sum_{n=-\infty}^{\infty} p[n] \, e^{-j\omega n} \ = \ \sum_{n=-M}^{M} e^{-j\omega n} \ = \ \frac{e^{j\omega M} - e^{-j\omega(M+1)}}{1 - e^{-j\omega}} \ = \ \frac{e^{-j\omega/2} \left(e^{j\omega \frac{2M+1}{2}} - e^{-j\omega \frac{2M+1}{2}}
ight)}{e^{-j\omega/2} \left(e^{j\omega/2} - e^{-j\omega/2}
ight)} \ = \ \frac{2j \, \sin\left(\omega \frac{2M+1}{2}
ight)}{2j \, \sin\left(\frac{\omega}{2}
ight)}$ 



Out[12]:

### **DTFT** of triangle



Out[13]:

**DTFT** of pulse

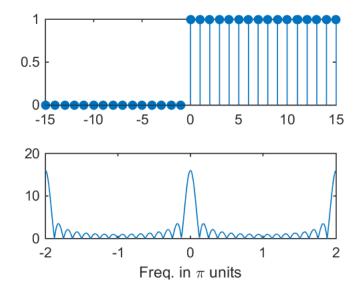
```
In [14]:
```

```
n = 0:15;
x = ones(1,length(n));
w = linspace(-2,2,2^10)*pi;

X = dtft(x,n,w);

xd = zeros(1,31);
nd = -15:15;
[y,ny] = sigadd(xd,nd,x,n);

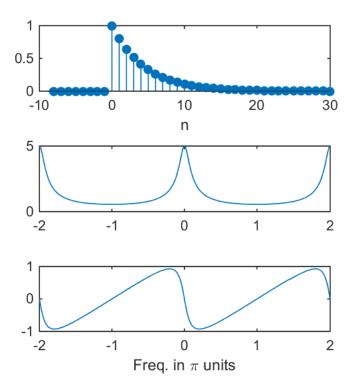
subplot(2,1,1), stem(ny,y,'filled')
subplot(2,1,2), plot(w/pi,abs(X)), xlabel('Freq. in \pi units')
```



Out[14]:

### DTFT of a one-sided exponential

$$h[n] = lpha^n u[n] \qquad \longleftrightarrow \qquad H(\omega) = rac{1}{1 - lpha e^{-j\omega}}$$



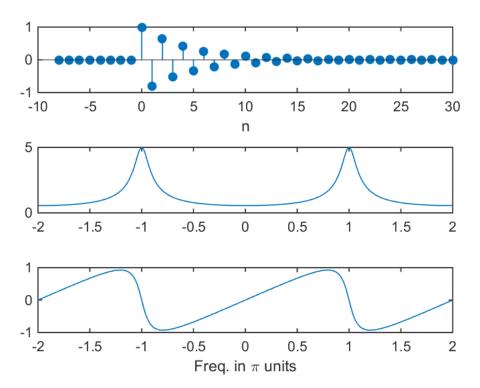
Out[15]:

## 3. Property:

**DTFT and Modulation** 

$$e^{-j\omega_0 n}x[n] \longleftrightarrow X(\omega-\omega_0) \ e^{-j\omega_0 n}x[n] = (-1)^nx[n] \quad ext{ when } \quad \omega_0 = rac{2\pi}{N}rac{N}{2} = \pi$$

```
In [16]:
            %plot -s 800,600
            N = 30;
            nd = -8:N;
            xd = zeros(size(nd));
            x = zeros(1,N);
            for i = 1:N
                x(i) = (-0.8)^{(i-1)};
            end
            n = 0:N-1;
            %w = linspace(-1,1,2^10)*pi;
            w = linspace(-2,2,2^10)*pi;
            X = dtft(x,n,w);
            subplot(3,1,1), stem(nd,sigadd(xd,nd,x,n),'filled'), xlabel('n')
            subplot(3,1,2), plot(w/pi,abs(X)),
            subplot(3,1,3), plot(w/pi,phase(X)), xlabel('Freq. in \pi units')
```



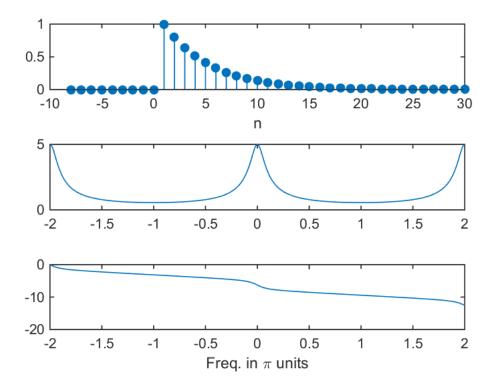
Out[16]:

### **DTFT and Time Shift**

$$x[n-m] \qquad \longleftrightarrow \qquad e^{-j\omega m}X(\omega)$$

- same amplitude
- phase changed (linearly  $-\angle \omega m$ )

```
In [17]:
             %plot -s 800,600
             N = 30;
nd = -8:N;
             xd = zeros(size(nd));
             x = zeros(1,N);
for i = 1:N
                 x(i) = 0.8^{(i-1)};
             end
             m = 1;
                              % m = 1 => one sample delay or shift
             n = 0+m:N-1+m;
             [y,ny] = sigadd(xd,nd,x,n);
             % w = linspace(-1,1,2^10)*pi;
             w = linspace(-2,2,2^10)*pi;
             X = dtft(x,n,w);
             subplot(3,1,1), stem(ny,y,'filled'), xlabel('n')
             subplot(3,1,2), plot(w/pi,abs(X)),
             subplot(3,1,3), plot(w/pi,phase(X)), xlabel('Freq. in \pi units')
```



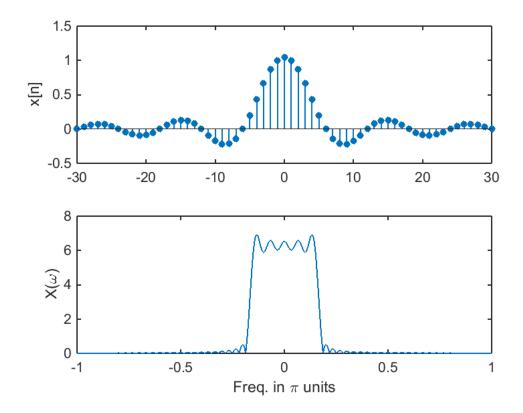
Out[17]:

### 4. Filters

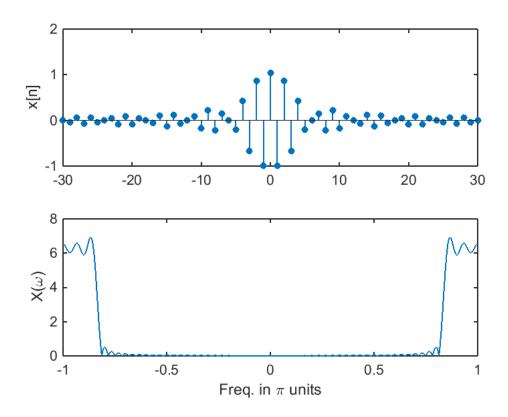
## Ideal lowpass filter and discrete-time sinc function

Impulse Response of the Ideal Lowpass Filter

$$h[n] = 2\omega_c rac{sin(\omega_c n)}{\omega_c n} \qquad \longleftrightarrow \qquad H(\omega) = egin{cases} 1 & -\omega_c \leq \omega \leq \omega_c \ 0 & ext{otherwise} \end{cases} \ h[n] = \int_{-\pi}^{\pi} H(\omega) \, e^{j\omega n} rac{d\omega}{2\pi} = \int_{-\omega_c}^{\omega_c} e^{j\omega n} rac{d\omega}{2\pi} = rac{e^{j\omega_c n} - e^{-j\omega_c n}}{jn} = 2\omega_c rac{\sin(\omega_c n)}{\omega_c n} \ \sin(x) = rac{\sin(\pi x)}{\pi x} \end{cases}$$

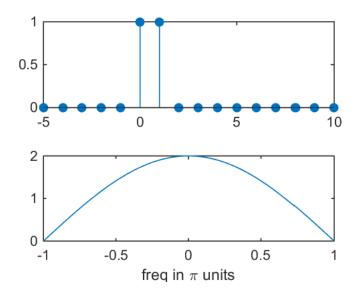


Out[1]:



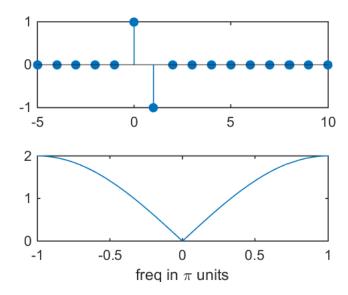
Out[2]:

**Linear Filters: Low-Pass** 



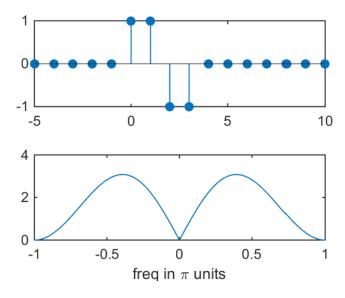
Out[20]:

Linear Filters: High-Pass



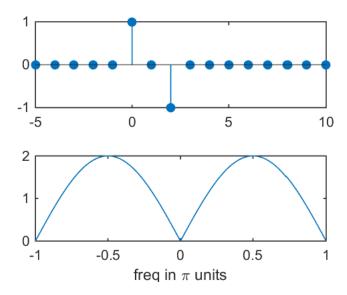
Out[21]:

**Linear Filters: Band-Pass** 



Out[22]:

**Linear Filters: Band-Stop** 



Out[23]:

## 5. High-density spectrum and high-resolution spectrum

$$x[n] = \cos(0.48\pi n) + \cos(0.52\pi n)$$

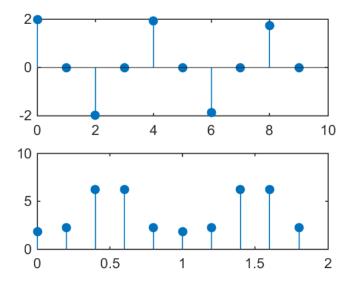
```
In [24]: N = 100;

n = 0:N-1;

x = cos(0.48*pi*n) + cos(0.52*pi*n);
```

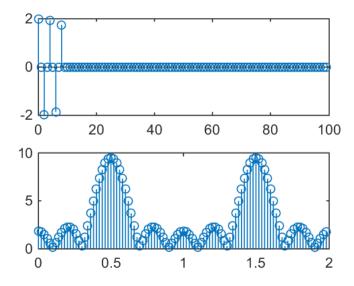
Out[24]:

use only 10-point DFT of x[n]



Out[25]:

pad 90 zeros to obtain a dense spectrum



Out[26]:

use 100 samples of x[n]