# **Control with Python**

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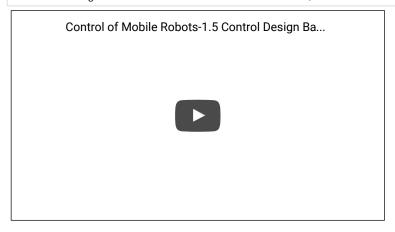
### 1. PID Control

• 1.5 Control Design Basics | Control of Mobile Robots

In [1]:

#### %%html

<iframe src="https://www.youtube.com/embed/DJuo9kLdr4M?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_a0qwjr"
width="560" height="315" frameborder="0" allowfullscreen></iframe>



For the given car model

$$\dot{x} = \frac{c}{m}u$$
 for the velocity of a car,  $x$ 

In a bloack diagram

$$u \longrightarrow \dot{x} = \frac{c}{m}u \longrightarrow x$$

in a Laplace transform

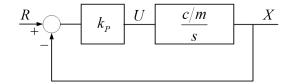
$$U \longrightarrow \begin{array}{|c|c|} \hline c/m \\ \hline s \\ \hline \end{array} \longrightarrow X$$

We want to achieve

## **Attempt 2: P Regulator**

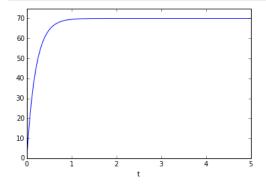
u = ke

- small error yeilds small control signals
- nice and smooth
- so-called proportional regulation (P regulator)



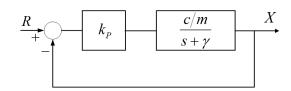
In [2]: import control
import numpy as np
import matplotlib.pyplot as plt
from control import \*

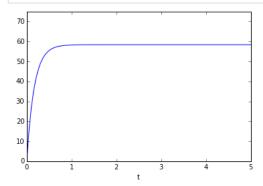
%matplotlib inline



### What if the true system is:

$$\dot{x} = \frac{c}{m}u - \gamma x$$



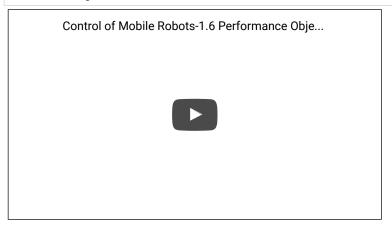


### Attempt 3

- 1.6 Performance Objectives | Control of Mobile Robots
- 1.7 PID Control | Control of Mobile Robots

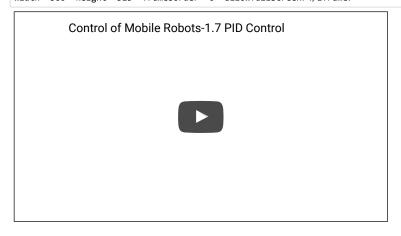
#### In [5]: %%html

<iframe src="https://www.youtube.com/embed/cQhqx65kLfM?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_a0qwjr"
width="560" height="315" frameborder="0" allowfullscreen></iframe>



In [6]:

<iframe src="https://www.youtube.com/embed/Mk1ygHj4zxw?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_a0qwjr"
width="560" height="315" frameborder="0" allowfullscreen></iframe>



$$u=ke+\gamma\frac{m}{c}x$$

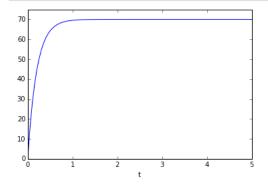
However, all of sudden we have to know all these physical parameters that we typically do not know - not robust !!!

### **Attempt 4: PI Regulators**

- Stability (BIBO)
- Tracking
- Robustness

$$u(t) = k_P e(t) + k_I \int_0^t e(\tau) d\tau$$

$$\frac{R}{s} + \frac{k_I}{s} - \frac{c/m}{s + \gamma}$$

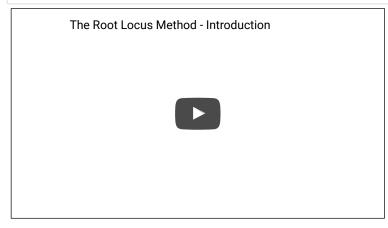


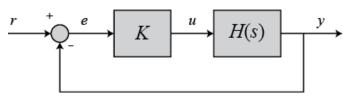
### 2. Root Locus

- The Root Locus Method by Brian Douglas
- from umich control (http://ctms.engin.umich.edu/CTMS/index.php?example=Introduction&section=ControlRootLocus)

### In [8]: %%html

<iframe src="https://www.youtube.com/embed/CRvVDoQJjYI?list=PLUMWjy5jgHK1NC52DXXrriwihVrYZKqjk"
width="560" height="315" frameborder="0" allowfullscreen></iframe>



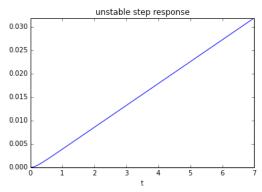


```
In [9]: sys = tf([1,7],[1,0])*tf(1,[1,5])*tf(1,[1,15])*tf(1,[1,20])

[y, tout] = step(sys)

plt.plot(tout, y)
plt.xlabel('t')
plt.axis('tight')
plt.title('unstable step response')

plt.show()
```



```
In [10]: print(pole(sys))
```

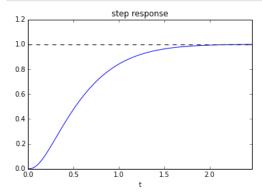
[-20. -15. -5. 0.]

```
In [11]:
             G = sys
              \# k = 10
              k = 800
              Gcl = feedback(k*G,1,-1)
              [y,tout] = step(Gcl)
             plt.plot(tout,y)
plt.plot(tout,np.ones(tout.shape),'k--')
              plt.axis('tight')
              plt.ylim(0.0,1.2)
              plt.show()
               1.0
               0.8
               0.6
               0.4
               0.2
              ما 0.0
0.0
                                                 1.5
                                                            2.0
In [12]:
              rlocus(G)
              plt.show()
              K = 350
In [34]:
              sys_cl = feedback(K*sys,1,-1) #negative feedback
              pzmap.pzmap(sys_cl)
              plt.show()
                                       Pole Zero Map
                  1.5
                  1.0
                  0.5
                  0.0
                 -0.5
                 -1.0
                 -1.5 L
                             -20
                                       -15
                                                -10
                                            Re
In [14]:
              sys_cl
```

Out[14]:

350 s + 2450 -----s^4 + 40 s^3 + 475 s^2 + 1850 s + 2450 In [15]:

```
[y,tout] = step(sys_cl)
plt.plot(tout,y)
plt.plot(tout,np.ones(tout.shape), 'k--')
plt.axis('tight')
plt.ylim(0,1.2)
plt.xlabel('t')
plt.title('step response')
plt.show()
```

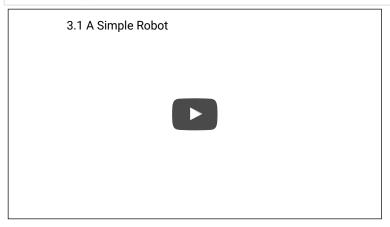


# 3. State Space Representation

- from 3.1 A Simple Robot | Control of Mobile Robots
- from 3.2 State Space Models | Control of Mobile Robots

In [16]:

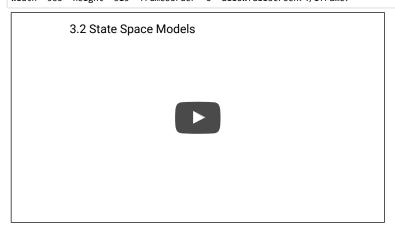
 $\label{local-com/embed/kQNUpNh6nBc?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl_a0qwjr" width="560" height="315" frameborder="0" allowfullscreen></iframe>$ 



In [17]:

%%html

<iframe src="https://www.youtube.com/embed/W6AUOyj5bFA?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_aOqwjr"</pre> width="560" height="315" frameborder="0" allowfullscreen></iframe>



Given a point mass on a line whose acceleration is directly controlled:

$$\ddot{p} = u$$

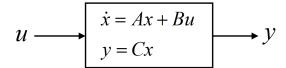
want to write this on a compact/general form

$$\dot{x}_1 = x_2 \ \dot{x}_2 = u$$

on a state space form

$$\dot{x} = egin{bmatrix} \dot{x}_1 \ \dot{x}_2 \end{bmatrix} = egin{bmatrix} 0 & 1 \ 0 & 0 \end{bmatrix} egin{bmatrix} x_1 \ x_2 \end{bmatrix} + egin{bmatrix} 0 \ 1 \end{bmatrix} u$$

$$y=p=x_1=\left[egin{array}{cc} 1 & 0 \end{array}
ight] \left[egin{array}{c} x_1 \ x_2 \end{array}
ight]$$



### 3.1. The car model

If we care about/can measure the velocity:

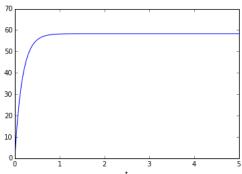
$$A = -\gamma, \qquad B = \frac{c}{m}, \qquad C = 1$$

If we care about/can measure the position we have the same general equation with different matrices:

$$A = \left[ egin{array}{cc} 0 & 1 \\ 0 & -\gamma \end{array} 
ight], \qquad B = \left[ egin{array}{c} 0 \\ rac{c}{m} \end{array} 
ight], \qquad C = \left[ egin{array}{c} 1 & 0 \end{array} 
ight]$$

```
In [18]:
```

```
#system in ss
c = 1
m = 1
gamma = 1
A = -gamma
B = c/m
C = 1
D = 0
Gss = ss(A,B,C,D)
# P controller
k = 5
C = k
# close loop
Gcl = feedback(C*Gss,1,-1)
x0 = 0
t = np.linspace(0,5,100)
r = 70*np.ones(t.shape)
[y,tout,x] = lsim(Gcl,r,t,x0)
plt.plot(tout,y)
plt.xlabel('t')
plt.ylim(0,70)
plt.show()
```

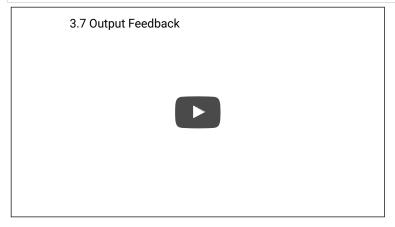


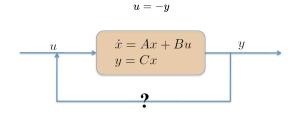
### 3.2. Back to the World's Simplest Robot (Output Feedback)

• from 3.7 Output Feedback | Control of Mobile Robots

In [19]:

%%html
<iframe src="https://www.youtube.com/embed/HmqOnsRH73w?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_aOqwjr"
width="560" height="315" frameborder="0" allowfullscreen></iframe>





```
In [20]:
```

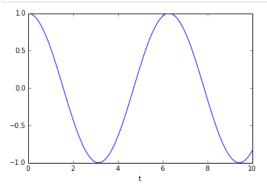
```
# to move towards the origin
# u = - y

A = np.array([[0, 1], [0,0]])
B = np.array([[0], [1]])
C = [1,0]
D = 0

G = ss(A,B,C,D)

K = 1
x0 = np.array([[1,0]]).T
t = np.linspace(0,10,100)
r = np.zeros(t.shape)

Gcl = feedback(G,K,-1)
[y,tout,x] = lsim(Gcl,r,t,x0)
plt.plot(tout,y)
plt.xlabel('t')
plt.show()
```



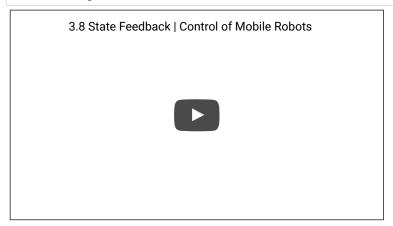
### 4. State Feedback

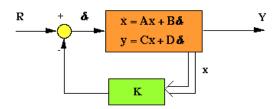
• from 3.8 State Feedback | Control of Mobile Robots

In [21]:

**%%html** 

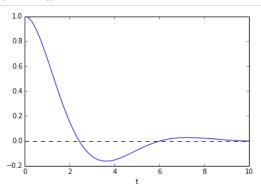
<iframe src="https://www.youtube.com/embed/y15IiJOYQps?list=PLciAw3uhNCiD3dkLTPJgHoMnsu8XgCt1m"
width="560" height="315" frameborder="0" allowfullscreen></iframe>





To move forwards origin,  $R=0\,$ 

```
A = np.array([[0,1],[0,0]])
In [22]:
              B = np.array([[0],[1]])
              C = [1,0]
              D = 0
              G = ss(A,B,C,D)
              k1 = 1
              k2 = 1
              K = [k1, k2]
              Gcl = ss(A-B*K,B,C,D)
              x0 = np.array([[1,0]]).T
t = np.linspace(0,10,100);
              r = np.zeros(t.shape)
              [y,tout,x] = lsim(Gcl,r,t,x0)
              plt.plot(tout,y)
              plt.plot(tout,np.zeros(tout.shape),'k--')
plt.xlabel('t')
              plt.show()
```



```
In [23]:
```

print(np.linalg.eigvals(Gcl.A))

[-0.5+0.8660254j -0.5-0.8660254j]

- It is clear that some eigenvalues are better than others. Some cause oscillations, some make the system respond too slowly, and so forth ...
- In the next module we will see how to select eigenvalues and how to pick control laws based on the output rather than the state.

### 4.1. Pole Placement

- from 4.1 Stabilizing the Point Mass | Control of Mobile Robots
- from 4.2 Pole Placement | Control of Mobile Robots

#### In [24]:

#### %%html

viframe src="https://www.youtube.com/embed/S4WZTmEnbrY?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_a0qwjr"
width="560" height="315" frameborder="0" allowfullscreen></iframe>

4.1 Stabilizing the Point Mass



#### In [25]:

#### %h+m1

<iframe src="https://www.youtube.com/embed/5tWhOK8Klo0?list=PLp8ijpvp8iCvFDYdcXqqYU5Ibl\_a0qwjr"
width="560" height="315" frameborder="0" allowfullscreen></iframe>

4.2 Pole Placement

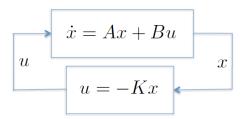


back to the point-mass, again

$$egin{aligned} u &= -Kx 
ightarrow \dot{x} = (A-BK)x \ A-BK &= egin{bmatrix} 0 & 1 \ 0 & 0 \end{bmatrix} - egin{bmatrix} 0 \ 1 \end{bmatrix} [\,k_1\,\,k_2\,] = egin{bmatrix} 0 & 1 \ -k_1 & -k_2 \end{bmatrix} \ egin{bmatrix} 0 & 1 \ -k_1 & -k_2 \end{bmatrix} = \lambda^2 + \lambda k_2 + k_1 \end{aligned}$$

Desired Eigenvalues: let's pick both eigenvalues at -1

$$(\lambda+1)(\lambda+1)=\lambda^2+2\lambda+1 \ k_1=2, k_2=1$$



Pick the control gains such that the eigenvalues (poles) of the closed loop system match the desired eigenvalues

#### Questions

- Is this always possible? (No)
- How should we pick the eigenvalues? (Mix of art and science)

Suppose

$$arphi = (\lambda + 1)^2 = \lambda^2 + \lambda (-3 + k_1 + k_2) + 2 - k_1 - k_2$$

Let's pick both eigenvalues at -1

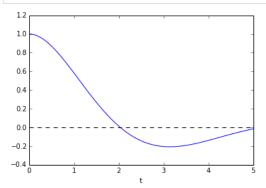
$$-3 + k_1 + k_2 = 2$$
 and  $2 - k_1 - k_2 = 1$ 

ightarrow no  $k_1$  and  $k_2$  exist

What's at play here is a lack of controllability, i.e., the effect of the input is not sufficiently rich to influence the system enough

In [27]: K = [2.6260, -0.6250]

$$\dot{x} = Ax + Bu = Ax - BKx = (A - BK)x$$



### 4.2. Controllability

- When can we place the eigenvalues using state feedback?
- When is B matrix (the actuator configuration) rich enough so that we can make the system do whatever we want it to do?
- The answer revolves around the concept of controllability

Given a discrete-time system

$$x_{k+1} = Ax_k + Bu_k$$

We would like to drive this system in n steps to a particular target state  $x^{st}$ 

$$x_1 = Ax_0 + Bu_0 = Bu_0$$
  
 $x_2 = Ax_1 + Bu_1 = ABu_0 + Bu_1$   
 $x_3 = Ax_2 + Bu_2 = A^2Bu_0 + ABu_1 + Bu_2$   
 $\vdots$   
 $x_n = A^{n-1}Bu_0 + \dots + Bu_{n-1}$ 

We want to solve

$$x^* = \left[egin{array}{cccc} B & AB & \cdots & A^{n-1}B \end{array}
ight] \left[egin{array}{c} u_{n-1} \ dots \ u_1 \ u_0 \end{array}
ight]$$

This is possible for any target state if and only if

$$\operatorname{rank}\left(\left[egin{array}{cccc} B & AB & \cdots & A^{n-1}B \end{array}
ight]
ight)=n$$

#### Example 1

$$\dot{x} = egin{bmatrix} 2 & 0 \ 1 & 1 \end{bmatrix} egin{bmatrix} x_1 \ x_2 \end{bmatrix} + egin{bmatrix} 1 \ 1 \end{bmatrix} u$$

[[1 2] [1 2]]

In [30]: print(np.linalg.matrix\_rank(G))

1

#### Example 2

$$\dot{x} = egin{bmatrix} 0 & 1 \ 0 & 0 \end{bmatrix} egin{bmatrix} x_1 \ x_2 \end{bmatrix} + egin{bmatrix} 0 \ 1 \end{bmatrix} u$$

In [32]: print(np.linalg.matrix\_rank(G))

2

[1 0]]