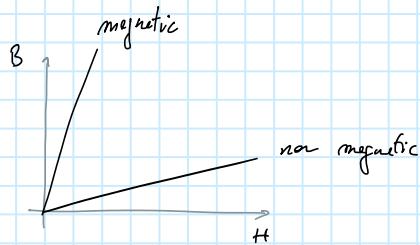


28/10 Bianchi

giovedì 3 dicembre 2020 12:57

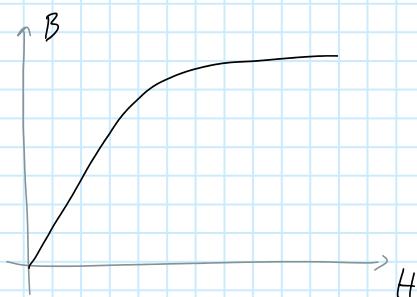
MAGNETIC MATERIALS:

- B : flux density (induzione) [T]
- H : magnetic field [A/m]



$$B = \mu_0 H \quad , \quad \mu: \text{permeability} [A/m]$$

$$\mu_0 = 4\pi \cdot 10^{-7} [A/m]$$



→ Saturation of the flux density
when the atoms are all oriented
they do not contribute anymore to
the permeability $\mu \rightarrow \mu_0$

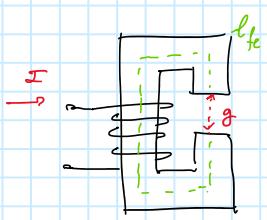
AMPERE'S LAW

$$\oint L H dL = I$$



The integral of the magnetic field in a closed line gives the current

$$\sum H dL = I$$



$$N \text{ turns} \rightarrow H_{fe} \cdot l_{fe} + H_f \cdot f = N \cdot I$$

$$\left\{ \begin{array}{l} B_{fe} = \mu_{fe} H_{fe} \\ B_f = \mu_0 H_f \end{array} \right.$$

$$\frac{B_{fe}}{\mu_{fe}} l_{fe} + \frac{B_f}{\mu_0} f = N I \quad \text{X}$$

GAUSS LAW

$$\oint B dS = 0 \quad \rightarrow \quad B_{fe} \cdot S_{fe} = B_f \cdot S_f$$

$$S_{fe} \approx S_f$$

$$\boxed{B_{fe} \approx B_d}$$

(*) $B \left(\frac{l_{fe}}{\mu_r} + \frac{g}{\mu_0} \right) = NI$

$$l_f \ll l_{fe}$$

$$\mu_0 \gg \mu_{fe}$$

$$\rightarrow \boxed{B \approx \frac{NI\mu_0}{l_f}}$$

LOW AIRGAP
TO INCREASE
MAGNETIC FLUX

$$\boxed{\Phi = B \cdot S \quad [Wb]}$$

FLUX

$$\rightarrow \boxed{\Lambda = N \cdot \Phi}$$

LINKAGE FLUX



Flux linked by the coil

RELUCTANCE

$$R = \frac{NI}{\Phi} = \frac{NI}{\Phi} \left[\frac{A}{Wb} \right]$$

$$I \rightarrow nIf \rightarrow H = \frac{NI}{\partial} \rightarrow B = \mu \cdot H \rightarrow \Phi = B \cdot S \rightarrow \Lambda = N\Phi$$

$$i(t) \rightarrow nIf = N \cdot iH \rightarrow h(t) \rightarrow b(t) \rightarrow \varphi(t) \rightarrow \lambda(t) \rightarrow e(t)$$

electromotive force $e(t)$
 [V]

FARADAY'S LAW

$$\boxed{e(t) = - \frac{d\lambda(t)}{dt}}$$

generator : minus sign
 motor : plus sign

INDUCTANCE

$$\boxed{L = \frac{\Lambda}{I} \quad \left[\frac{V \cdot s}{A} \right] = [H]}$$

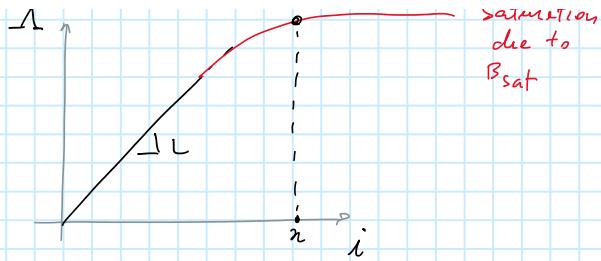
$$\Lambda = N\Phi = NBS = N\mu_0 H S$$

$$= N\mu_0 S \cdot \frac{NI}{\partial} = \boxed{\mu_0 \frac{N^2 S}{\partial} \cdot I}$$



Saturation
 due to
 β_{sat}

How to define the inductance



*saturation
due to
 B_{sat}*

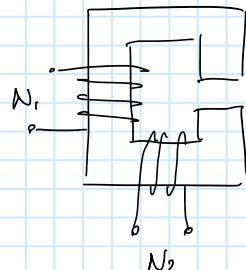
How to define the inductance
in a curved point:

$$L_{\text{APPARENT}} = \frac{\Delta(\infty)}{I(\infty)}$$

$$L_{\text{DIFFERENTIAL}} = \frac{\Delta L}{\Delta i} \quad (\text{INCREMENTAL})$$

MUTUAL INDUCTANCE

$$M = L_M = \frac{L_1}{I_B}$$



$$M_{12} = \frac{L_1}{I_1} = \frac{L_2}{I_2}$$

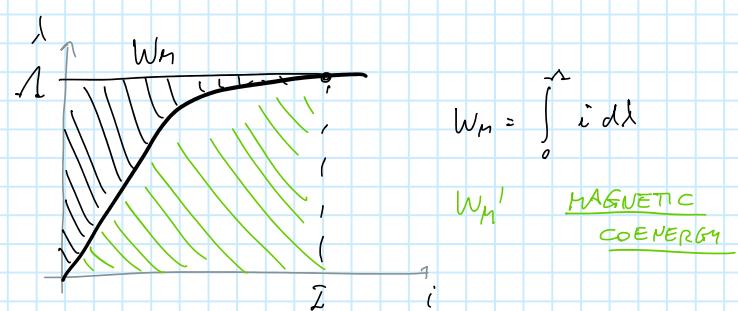
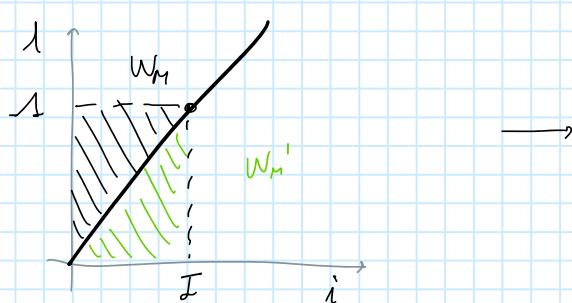
MAGNETIC ENERGY

$$\begin{aligned} W_M &= \frac{1}{2} B \cdot H \cdot \text{volume} [J] \\ &= \frac{N \cdot I}{S} \cdot \frac{1}{N \cdot S} \cdot S \cdot f \cdot \frac{1}{2} \\ &= \frac{1}{2} L \cdot I \end{aligned}$$

$$\text{volume} = S \cdot f$$

$$H = \frac{N \cdot I}{S}$$

$$B = \frac{F}{S} = \frac{1}{N \cdot S}$$



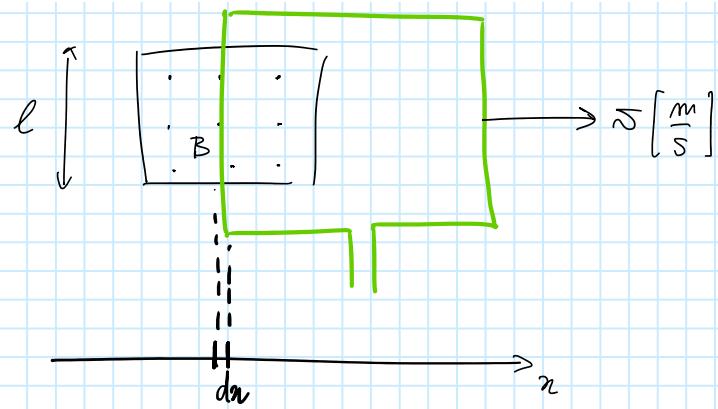
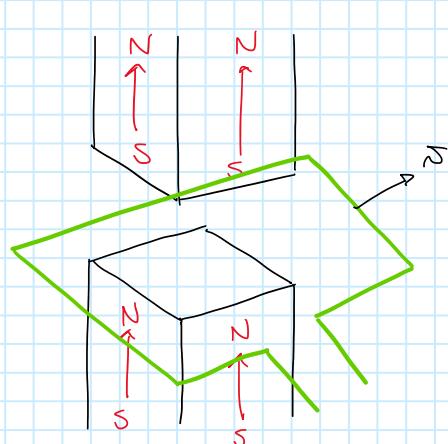
FORCE

$$F_x = \frac{dW_M}{dx} [N]$$

LORENTZ

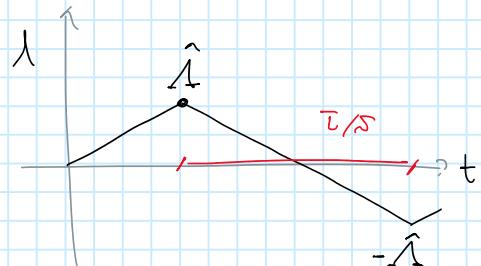
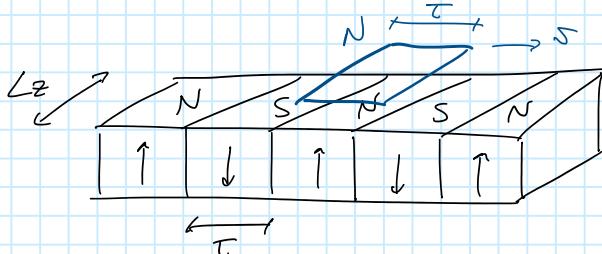
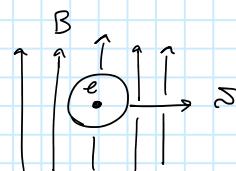
$$F = I \cdot B \cdot l$$

$$\begin{matrix} \uparrow & \uparrow & \uparrow & B \\ F & \leftarrow & \odot & I \\ \uparrow & \uparrow & \uparrow \end{matrix}$$



$$dt \rightarrow d\lambda = d\Phi = -B l dx$$

$$\boxed{e = -\frac{d\lambda}{dt} = N B l}$$

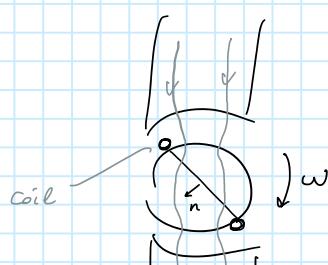
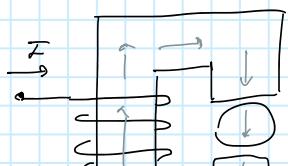


$$\hat{\lambda} = NBS = NBTL_z$$

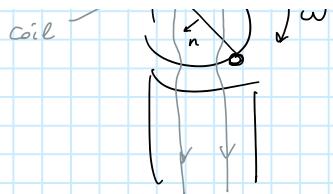
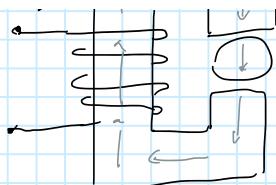
- $E_{tot} = \left| \frac{d\lambda}{dt} \right| = \frac{2\hat{\lambda}}{\tau/N} = \frac{2\hat{\lambda}}{\tau} = 2\hat{\lambda} \frac{N}{\tau}$

- $E_{induced} = \frac{E}{2N} = \hat{\lambda} B L_z$

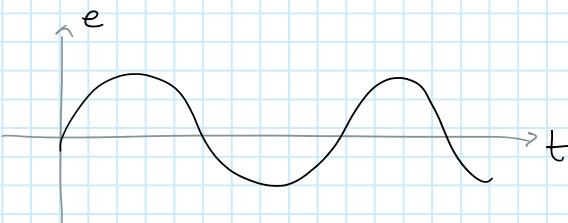
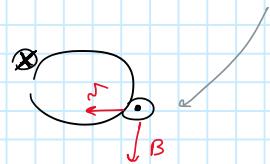
ROTATING COIL



- $\lambda(t)$ decreases by rotation
- $e(t) = -\frac{d\lambda}{dt}$ is positive

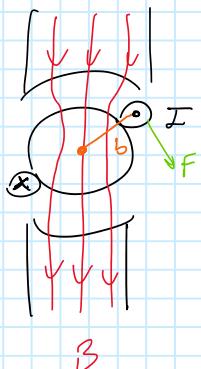


$$e(t) = -\frac{d\Phi}{dt} \text{ is positive}$$

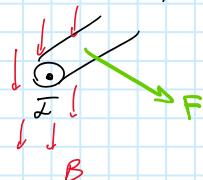


sinusoidal alternating voltage

$$\lambda(t) \rightarrow e(t) \rightarrow \text{if closed circuit} \rightarrow i(t)$$



$$(F = I B l)$$



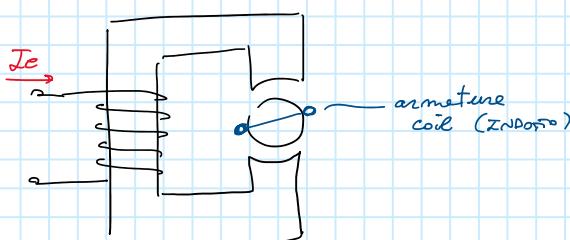
Torque = Force × lever arm

$$\begin{aligned} &2 \text{ forces on 2 conductors} \\ \Rightarrow &\boxed{T = 2F \cdot b} \end{aligned}$$

DC motor

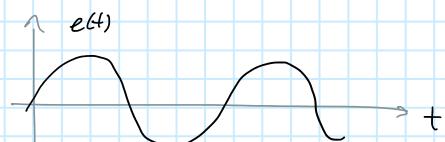
Ferromagnetic material → High $B = \mu H$

Airgap is minimum → just to avoid friction



If the rotor rotates at ω_m , emf e is induced in the coil:

$$e(t) \propto \lambda(t), \omega_m$$



$$\boxed{E = K \bar{\Phi} \omega_m}$$

MOTOR
EQUATIONS

a: constants

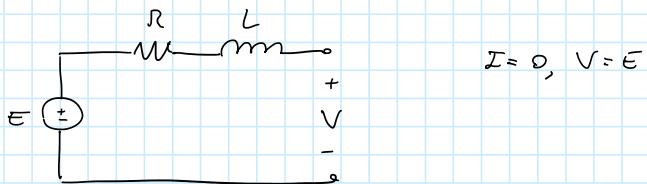
$$\overbrace{T_m = K \Phi I_a}^{\text{EQUATIONS}}$$

EQUATIONS
a: armature

30/10 Bianchi

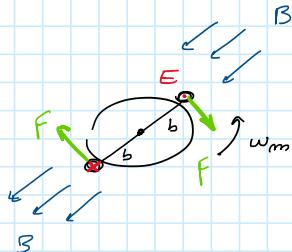
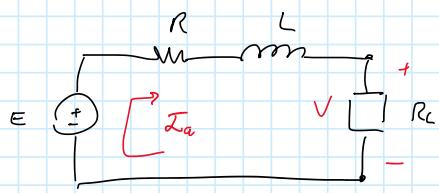
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$$E = K\Phi \omega_m$$



$$I = 0, V = E$$

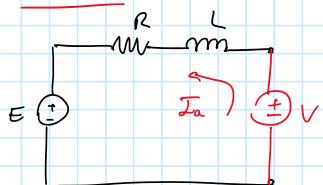
GENERATORE



Le corrente avrà la stessa direzione delle fem

Le forze di Lorentz si oppone al movimento rotatorio \rightarrow nasce una coppia elettromagnetica

MOTORE



$V > E \rightarrow$ forza la corrente dentro al motore vincendo la ferm

$$E = K\Phi \omega_m \rightarrow T_{em} = K\Phi I_a$$

BILANCIO POTENZE

$$P_e = P_m$$

$$E \cdot I_a = T_{em} \omega_m$$

$$(K \cdot \Phi \cdot \omega_m) \cdot I_a = (K \cdot \Phi \cdot I_a) \cdot \omega_m$$

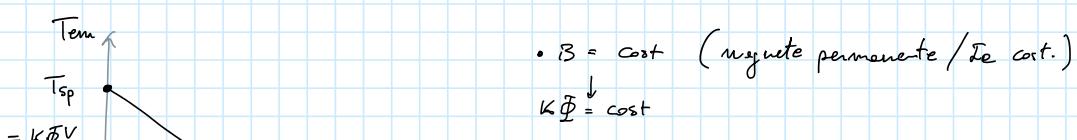
KIRKHOFF

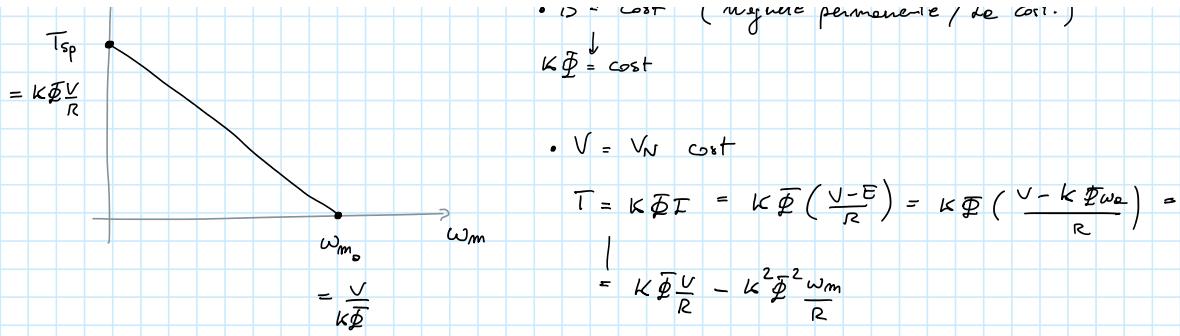
$$\begin{aligned} V &= E + L \frac{dI_a}{dt} + RI_a \\ &= E + RI_a \end{aligned}$$

EQUAZIONI MOTORE DC

$$\boxed{\begin{aligned} E &= K \cdot \Phi \cdot \omega_m \\ T &= K \cdot \Phi \cdot I_a \\ V &= E + R \cdot I_a \end{aligned}}$$

CARATTERISTICA MECCANICA (A REGIME)





EXERCISE 1

Pr excitation

Tor ?

$$R_a = 1.2$$

Power ?

$$V_a = 100V$$

$$\boxed{\begin{aligned} E &= K\bar{\Phi}\omega_m \\ T &= K\bar{\Phi}I_a \\ V &= E + R I_a \end{aligned}}$$

$$I_a = 1.5A$$

$$\bar{\omega} = 3000 \text{ rpm}$$

$$E = V - R I_a = 100 - 1.2 \cdot 1.5 = 85V$$

$$\omega_m = \frac{2\pi}{60} 3000 = 314 \text{ rad/s}$$

$$K\bar{\Phi} = \frac{E}{\omega_m} = 0.27 \text{ Vs}$$

$$\bullet \text{Torque} = K\bar{\Phi} \cdot I_a = 0.27 \cdot 1.5 = 4.06 \text{ Nm}$$

• Power :

$$- \text{input power} \quad V \cdot I_a = 150W$$

$$- \text{joule loss} \quad R \cdot I_a^2 = 1.2 \cdot 1.5^2 = 22.5W$$

$$- \text{electromech.} \quad \left. \begin{aligned} \text{Tor} \cdot \omega_m &= E \cdot I_a \\ 4.06 \cdot 314 &= 85 \cdot 1.5 \end{aligned} \right\} = 127.5W$$

EXERCISE 2

Pr DC motor

Torque ?

$$V = 200V$$

Tor Loss ?

$$I_a = 22.5A$$

iron + mech Losses ?

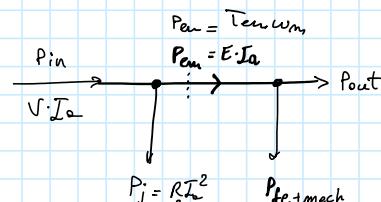
$$R_a = 0.5 \Omega$$

η ?

$$P_{out} = 4500W$$

$K\bar{\Phi}$?

$$\bar{\omega} = 6000 \text{ rpm}$$



$$\omega_m = \frac{2\pi}{60} \cdot 5 = 62.8 \text{ rad/s}$$

$$T_{sh} = \frac{P_{out}}{\omega_m} = 7.16 \text{ Nm}$$

$$P_{in} = V_a \cdot I_a = 4880 \text{ W}$$

$$\eta = \frac{P_{out}}{P_{in}} = 92.2 \%$$

$$P_J = R_a I_a^2 = 298 \text{ W}$$

$$P_{fe+mech} = \underbrace{P_{in} - P_J - P_{out}}_{P_{em}} = 4880 - 298 - 4500 = 82 \text{ W}$$

$$T_m = \frac{P_{em}}{\omega_m} = 7.3 \text{ Nm}$$

$$K\bar{\phi} = \frac{T_m}{I_a} = 0.299 \text{ Vs} \rightarrow E = K\bar{\phi} \omega_m = 187.8 \text{ V}$$

EXERCISE 3

PM DC motor

T_N ?

$$R_a = 0.2 \Omega$$

η_N ?

②

$$T_L = 10 \text{ Nm}$$

$$P_N = 4000 \text{ W}$$

$P_{fe+mech} \propto \omega_m^2$

$$\omega_L = 2000 \text{ rpm}$$

V_a ?

$$\omega_N = 3000 \text{ rpm}$$

$$V_N = 120 \text{ V}$$

$$I_N = 45 \text{ A}$$

$$\omega_N = 314 \text{ rad/s}$$

$$\textcircled{2} \rightarrow \omega_m = \frac{2\pi}{60} 2000 = 210 \text{ rad/s}$$

$$T_N = \frac{P_N}{\omega_N} = 12.74 \text{ Nm}$$

$$P_{fe} = P_{fe} \left(\frac{\omega_m}{\omega_N} \right)^2 = 42 \text{ W}$$

$$P_{inN} = V_a I_N = 4500 \text{ W}$$

$$T_{fe} = \frac{P_{fe}}{\omega_t} = 0.2 \text{ Nm}$$

$$\eta_N = 89.3 \%$$

$$T_{em} = T_L + T_{fe} = 10.2 \text{ Nm}$$

$$P_J = R_a I_a^2 = 405 \text{ W}$$

$$I_a = \frac{T_{em}}{K\bar{\phi}} = 35.2 \text{ A}$$

$$P_{fe} = P_{in} - P_J - P_{out} = 98 \text{ W}$$

$$V = E + R I_a = K\bar{\phi} \omega_m + R I_a = \\ = 0.29 \cdot 210 + 0.2 \cdot 35.2 = 66.9 \text{ V}$$

$$T_{emN} = \frac{P_{in} - P_J}{\omega_N} = 13 \text{ Nm}$$

$$K\bar{\phi} = \frac{T_{emN}}{I_N} = \frac{13}{45} = 0.29 \text{ Vs}$$

03/11 Bianchi

venerdì 4 dicembre 2020 18:41

EXERCISE

DC motor independent excitation

$$R_e = 1.5 \Omega$$

$$V_e = 75 V$$

$$\omega_{m0} = 600 \text{ rad/s} \quad (\text{at max flux})$$

- Requirement:

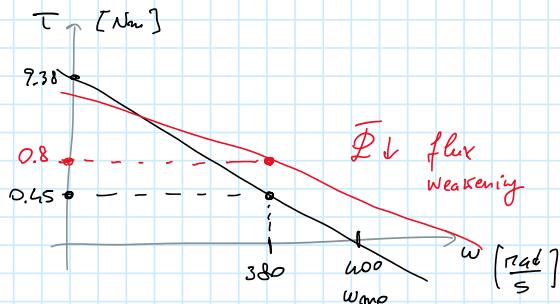
$$\tau_L = 0.8 Nm$$

$$\omega_m = 380 \text{ rad/s}$$

V_e ? I_e ? Power?

$$\omega_{m0} = \frac{V_e}{K\bar{\Phi}} \rightarrow K\bar{\Phi} = \frac{75}{600} = 0.125 V \cdot s \quad (\text{NO LOAD})$$

$$\tau_{sp} = K\bar{\Phi} \frac{V_e}{R_e} = 9.38 \text{ Nm} \quad (\text{START UP})$$



• at ω_m

$$E = K\bar{\Phi} \omega_m = 71.4 V$$

$$I_e = \frac{V - E}{R_e} = 2.37 A$$

$$\tau_{em} = \frac{K\bar{\Phi} \cdot I_e}{R_e} = 0.45 Nm < \tau_L \quad \textcircled{*}$$

$$\tau_L = \tau_{em} = K\bar{\Phi} \frac{V_e}{R_e} - (K\bar{\Phi})^2 \frac{\omega_m}{R_e}$$

$$K\bar{\Phi} = \underline{0.125} \text{ V} \quad 0.0125 \text{ T} \quad \text{low current} \quad \textcircled{*}$$

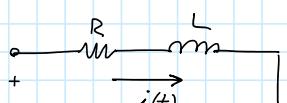
↓

- $I_e = \frac{0.8}{0.125} = 6.4 A$

- $V_e = E + R_e I_e =$

- $V_e I_e = E I_e + R_e I_e^2$

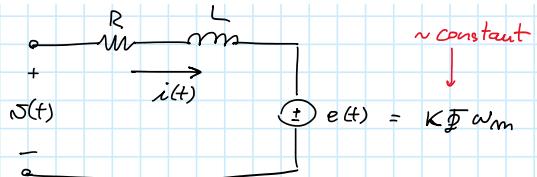
DYNAMICS



~ constant
↓

$$e \cdot l = K\bar{\Phi} \omega_m$$

$$\bullet \tau_{em} = K\bar{\Phi} i$$



$$e \cdot i = K\Phi \omega_m$$

$$\tau_m = K\Phi i$$

$$\mathcal{E} = e + Ri + L \frac{di}{dt}$$

$$\tau_m = \tau_L + B\omega_m + J \frac{d\omega}{dt}$$

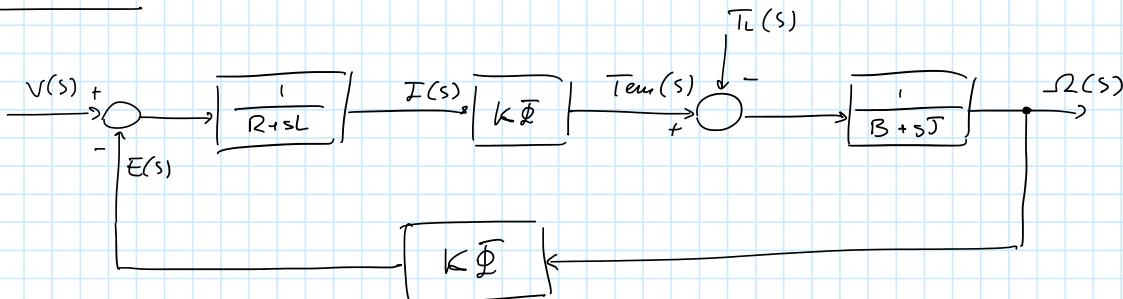
\downarrow Losses \downarrow Inertia

LAPLACE

$$\begin{cases} V(s) = E(s) + RI(s) + sL\mathcal{I}(s) \\ E(s) = K\Phi \Omega(s) \\ \tau_m(s) = K\Phi I(s) = \tau_L + B\Omega(s) + sJ\Omega(s) \end{cases}$$

$$\Omega(s) = \frac{\tau_m(s) - \tau_L(s)}{B + sJ}$$

BLOCK DIAGRAM



- Transfer function between Voltage and Speed :

$$\frac{\Omega_m(s)}{V(s)} = \frac{\frac{1}{R+sL} \cdot K\Phi \cdot \frac{1}{B+sJ}}{1 + \frac{1}{R+sL} \cdot (K\Phi)^2 \frac{1}{B+sJ}} = \frac{K\Phi}{(R+sL)(B+sJ) + (K\Phi)^2}$$

- B is negligible

$$\frac{\Omega_m(s)}{V(s)} = \frac{\frac{1}{K\Phi}}{s^2 \frac{LJ}{(K\Phi)^2} + s \frac{RJ}{(K\Phi)^2} + 1}$$

At steady state ($s=0$)

$$\Omega_{m0} = \frac{V}{K\Phi} \quad \checkmark$$

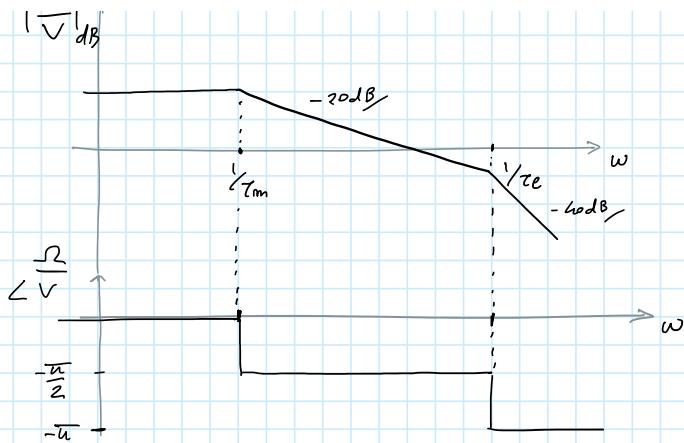
TIME CONSTANTS

$$\tau_m = \frac{RJ}{(K\Phi)^2}$$

$$\tau_e = \frac{L}{R}$$



if $\tau_m \gg \tau_e$



$\omega_m(s) \xrightarrow{t} ?$

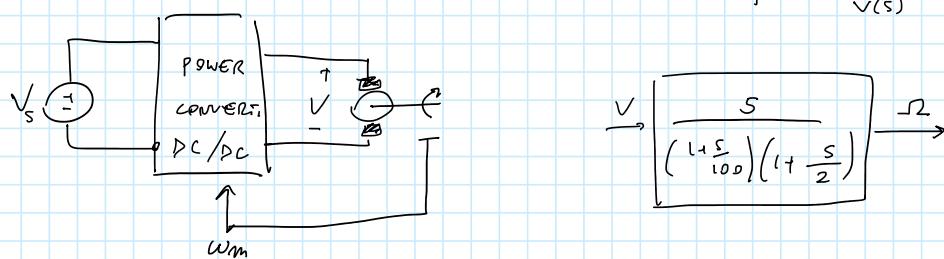
$$\begin{aligned}\omega_m(s) &= \frac{A}{s} + \frac{B}{s + \frac{1}{\tau_1}} + \frac{C}{s + \frac{1}{\tau_2}} \\ &= \frac{A(s + \frac{1}{\tau_1})(s + \frac{1}{\tau_2}) + B(s + \frac{1}{\tau_1})s + C(s + \frac{1}{\tau_2})s}{s(s + \frac{1}{\tau_1})(s + \frac{1}{\tau_2})}\end{aligned}$$

$$\left\{ \begin{array}{l} A = \lim_{s \rightarrow 0} (\omega_m(s) \cdot s) \\ B = \lim_{s \rightarrow -\frac{1}{\tau_1}} [\omega_m(s) \cdot (s + \frac{1}{\tau_1})] \\ C = \lim_{s \rightarrow -\frac{1}{\tau_2}} [\omega_m(s) \cdot (s + \frac{1}{\tau_2})] \end{array} \right.$$

$$\boxed{\omega_m(+)} = A + B e^{-\frac{t}{\tau_1}} + C e^{-\frac{t}{\tau_2}}$$

steady state after ~ 3 time constants

POWER CONVERTER



From before $\frac{\omega(s)}{V(s)}$: (numer. data)

Power converter:

$$V \xrightarrow{[k_c]} V$$

Power converter:

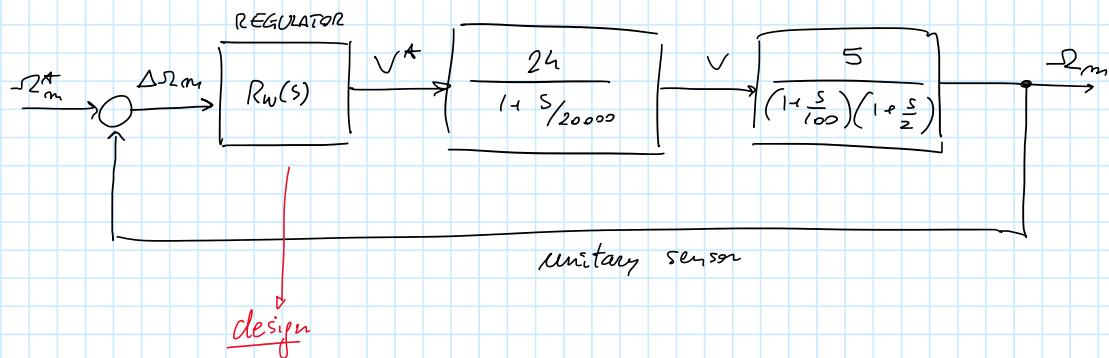
$$\frac{V^*}{\frac{Lc}{Kc} + Tc} \rightarrow V$$

$$Kc = 2L$$

$$V_N = 120V$$

$$V_{REF} = V^* = 5V$$

$$f_{sw} = 10 \text{ KHz} \rightarrow T_{sw} = 0.1 \text{ ms} \rightarrow T_c = \frac{T_{sw}}{2}$$



05/11 Bianchi

sabato 5 dicembre 2020 16:08

$$\varphi_m = 60^\circ \rightarrow \underline{\text{Proporzionalità}}$$

$$\angle G(j\omega) = -180 + \varphi_m = -120^\circ$$

$$\begin{aligned} &= -\arctan\left(\frac{\omega_A}{0.5}\right) - \arctan\left(\frac{\omega_A}{100}\right) - \cancel{\arctan\left(\frac{\omega_A}{20000}\right)} = -120^\circ \\ &\Rightarrow \omega_A = 60 \text{ rad/s} \end{aligned}$$

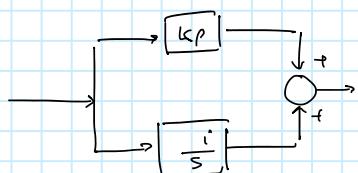
$$\bullet K_p: |G(j60)| = 1$$

$$\left| K_p \cdot \frac{120}{(1+j\frac{60}{0.5})(1+j\frac{60}{100})(1+j\frac{60}{20000})} \right| = 1$$

$$\Rightarrow K_p = \dots$$

$$e_{errore} = e(\infty) = \frac{1}{|G(j\infty)|+1} \neq 0$$

PI



$$P_W(s) = K_p + \frac{K_i}{s} = \frac{sK_p + K_i}{s} = K_I \frac{(1 + \frac{K_p}{K_i})}{s}$$

$$\frac{K_p}{K_i} = T_i \rightarrow P_W(s) = K_I \cdot \frac{(1 + sT_i)}{s}$$

EXERCISE

MOTOR

$$L_f = 0.1 \text{ Vs}$$

$$R = 2.5 \Omega$$

$$L = 10 \text{ mH}$$

$$B \approx 0$$

$$J = 10^{-3} \text{ kg m}^2$$

POWER

$$K_C = 8$$

$$f_{sw} = 2.5 \text{ kHz}$$

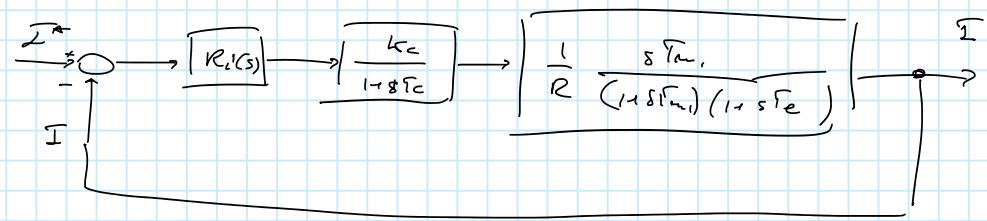
REQUIREMENT

speed control

$$- \omega_B \geq 60 \text{ rad/s}$$

$$- \varphi_m \geq 60^\circ$$

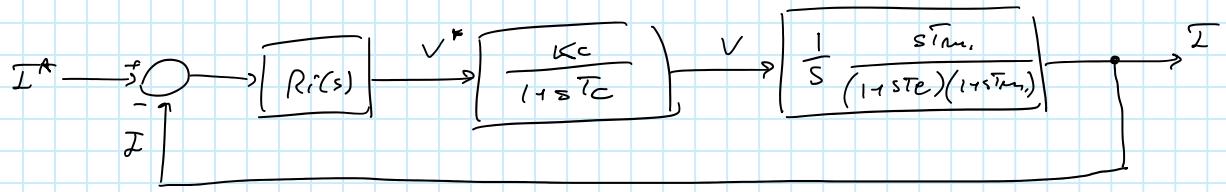
Current control :



$$T_c = \frac{T_{sw}}{2} = \frac{1}{2} \frac{1}{2.5 K_H} = 0.0002 \text{ s}$$

$$T_e = \frac{L}{R} = 0.004 \text{ s}$$

$$T_m = \frac{RJ}{(\bar{K}_f)^2} = 0.25 \text{ s}$$

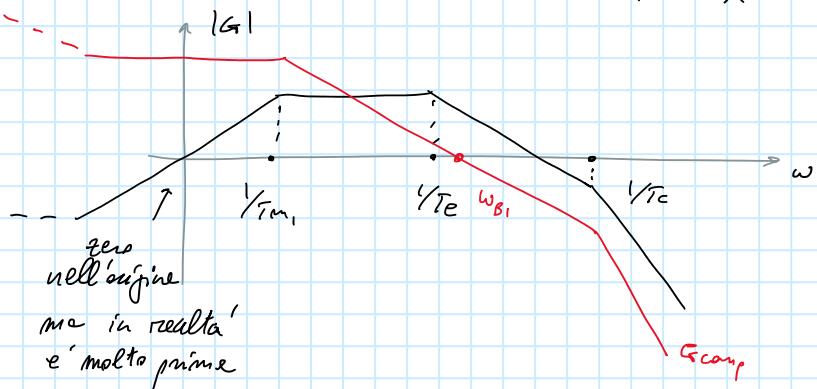


$$T_e = 0.04 \text{ s}$$

$$T_{m1} = 0.25 \text{ s}$$

$$T_c = 0.0002 \text{ s}$$

$$G(s) = R_i(s) \cdot \frac{K_c}{1+sT_c} \cdot \frac{1}{s} \frac{sT_{m1}}{(1+sT_e)(1+sT_{m1})}$$



Regolatore PI : $R_{p1}(s) = k_i \cdot \frac{1+sT_{p1}}{s}$

$$\bar{T}_{p1} = T_e$$

$$\omega_{B1} = \frac{1}{\bar{T}_{p1}} = 800 \text{ rad/s}$$

Impone : $|G(j\omega_{B1})| = 1$

$$\left| k_i \cdot \frac{1+j\omega_{B1}\bar{T}_{p1}}{j\omega_{B1}} \cdot \frac{K_c}{1+j\omega_{B1}T_c} \cdot \frac{1}{j\omega_{B1}T_{m1}} \frac{j\omega_{B1}T_{m1}}{(1+j\omega_{B1}T_e)(1+j\omega_{B1}T_{m1})} \right| = 1$$

$$\boxed{k_i = 157}$$

$$\boxed{k_p = \bar{T}_{p1} \cdot k_i = 0.628}$$

CON LUOGO DELLE RADICI :

$$G(s) = k_i \cdot \frac{1+s\bar{T}_{p1}}{s} \cdot \frac{K_c}{1+sT_c} \cdot \frac{1}{j\omega_{B1}} \frac{sT_{m1}}{(1+sT_e)(1+sT_{m1})}$$

$$\begin{aligned}
 G(s) &= K_i \cdot \frac{\frac{1+sT_m}{s}}{1+sT_c} \cdot \frac{\frac{nc}{1+sT_c}}{\frac{1}{R} \frac{sT_m}{(1+sT_c)(1+sT_m)}} \\
 &= \underbrace{\frac{K_i K_c T_m}{R}}_{K'} \frac{1}{(1+sT_m)(1+sT_c)} \\
 &= \frac{K'}{(1+sT_m)(1+sT_c)}
 \end{aligned}$$

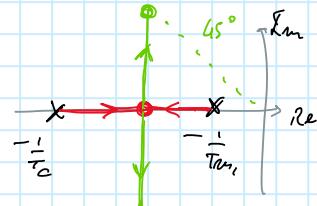
Closed Loop : $\frac{G_r}{1+G} = \frac{K'}{(1+sT_m)(1+sT_c) + K'}$

$$s^2(T_m T_c) + s(T_m + T_c) + (1 + K') = 0$$

$$s_{1,2} = \frac{-(T_m + T_c) \pm \sqrt{(T_m + T_c)^2 - 4c(T_m T_c)(1 + K')}}{2T_m T_c}$$

Vario K' :

$$\begin{aligned}
 K' = 0 &\rightarrow s_{1,2} = -\frac{1}{T_m}, -\frac{1}{T_c}
 \end{aligned}$$



$$K'_\Delta \mid \Delta = 0 \rightarrow (T_m + T_c)^2 - 4(T_m T_c)(1 + K'_\Delta) = 0$$

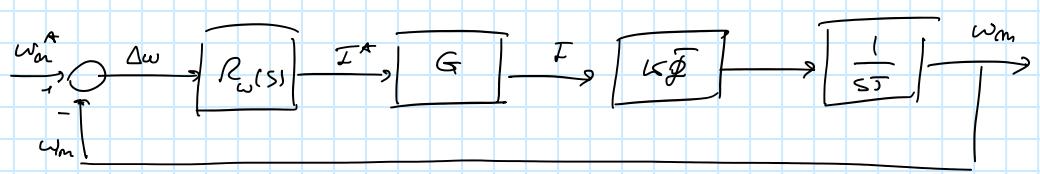
$$K'_\Delta = \frac{(T_m + T_c)^2}{4T_m T_c} - 1 \rightarrow s_{1,2} = -\frac{T_m + T_c}{2T_m T_c}$$

$$K' \mid \Delta < 0 \rightarrow K' > K'_\Delta \quad s_{1,2} = R_e \pm j I_m \quad \theta = \arctan \frac{I_m}{R_e} = 0.707 \quad \checkmark$$

con $R_e = \Sigma_m$

$$\begin{aligned}
 K' = 625 &= \frac{K_i K_c T_m}{R} \\
 \rightarrow K_i &= \frac{K'}{K_c T_m} \cdot R = 781
 \end{aligned}$$

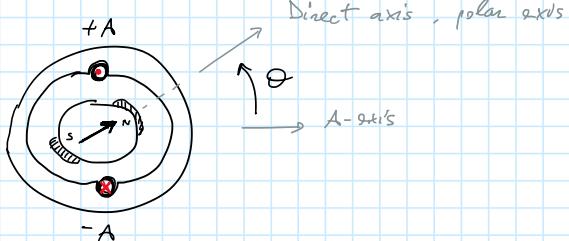
$$K_p = T_m, K_i = \underline{3.13}$$



$$\omega_m = \frac{k_p k_a}{B} \Delta \omega = \omega^* - \Delta \omega \rightarrow \Delta \omega = \frac{\omega^*}{1 + \frac{k_p k_a}{B}}$$

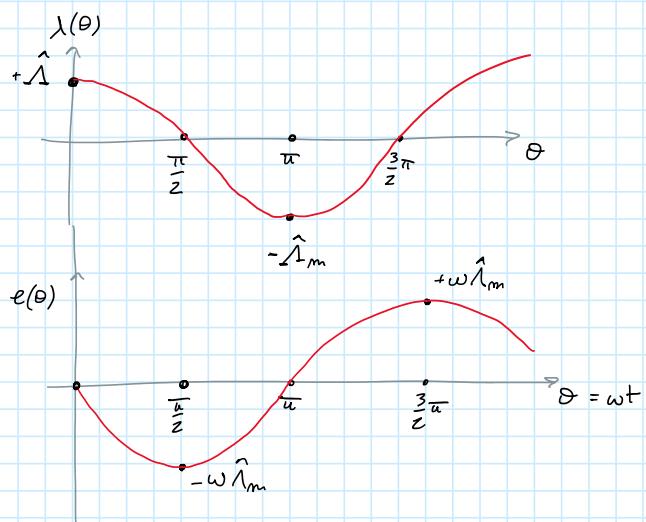
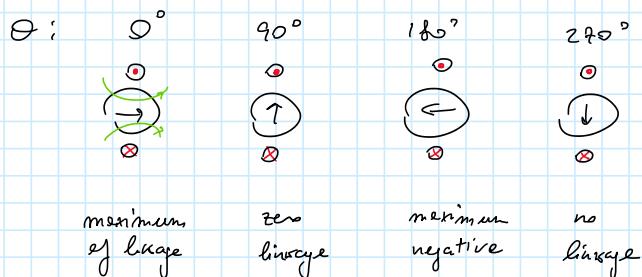
MOTORE SINCRONO A MAGNETE PERMANENTE SUPERFICIALE

[SPM MOTORS]



θ : angular position of the rotor axis

$\lambda(\theta)$ flux linkage is dependent to the angle



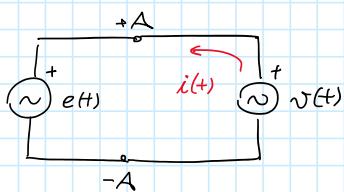
$$\lambda(\theta) = \hat{\lambda}_m \cos(\theta)$$

$$\omega = \frac{d\theta}{dt} \rightarrow \text{constant}$$

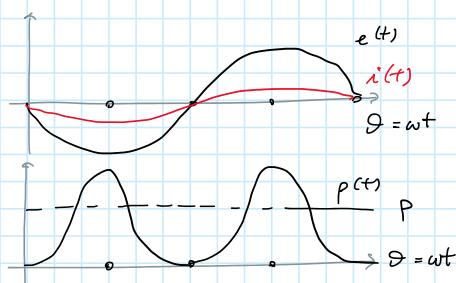
$$\begin{aligned} \theta &= \omega t \\ \Rightarrow e(t) &= \frac{d\lambda(t)}{dt} = \frac{d\lambda}{d\theta} \frac{d\theta}{dt} \\ &= -\omega \hat{\lambda}_m \sin(\theta) \end{aligned}$$

- $E = \omega \lambda_m$ in DC: $E = K \bar{\Phi} \omega_m$

EFFECT OF THE CURRENT



We want to impose a current which is in phase with the electromotive force $e(t)$





$$i(\theta) = I \sin \theta$$

$$p(\theta) = i(\theta) e(\theta) = \omega I_m \sin \theta \cdot I \sin \theta = \omega I_m I \cdot \sin^2 \theta \quad \text{INSTANTANEOUS POWER}$$

$$\cos(\alpha + \beta) = \cos \alpha \cos \beta - \sin \alpha \sin \beta$$

$$\sin \alpha \sin \beta = \frac{\cos(\alpha - \beta) - \cos(\alpha + \beta)}{2} \rightarrow \sin^2 \theta = \frac{1 - \cos(2\theta)}{2}$$

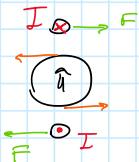
$$\Rightarrow p(\theta) = \omega I_m I \left(\frac{1 - \cos 2\theta}{2} \right) \quad P = \frac{\omega I_m I}{2} \quad \text{AVERAGE POWER}$$

$$T_{av} = \frac{P}{\omega_m} = \frac{1}{2} I_m \mathcal{E}_m \quad \text{AVERAGE TORQUE}$$

$$\theta = 0^\circ$$



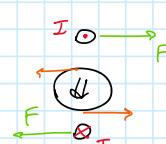
$$90^\circ$$



$$180^\circ$$



$$270^\circ$$



$$F = IBL$$

Force on the rotor

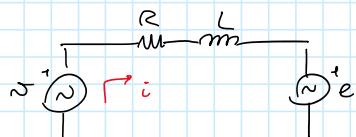
$$T = F \cdot b$$

b = radius of the rotor

$$T_{av} > 0$$

$$T_{av} > 0$$

VOLTAGE EQUATION



$$\mathcal{E} = e + R_i i + L \frac{di}{dt}$$

$$e = -\mathcal{E} \sin(\omega t)$$

$$i = -I \sin(\omega t)$$

$$a(t) = A \cos(\omega t + \alpha)$$

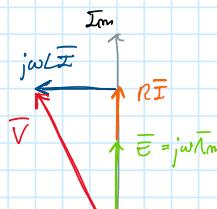
$$\bar{A} = |A| e^{j\alpha} \quad A_{RMS} = \frac{A_{peak}}{\sqrt{2}}$$

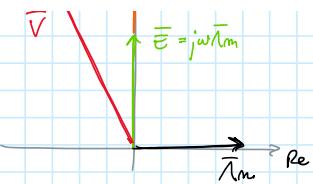
$$\frac{de}{dt} \rightarrow \bar{A} j\omega$$

$$\Rightarrow \mathcal{E} = e + R_i i + L \frac{di}{dt}$$

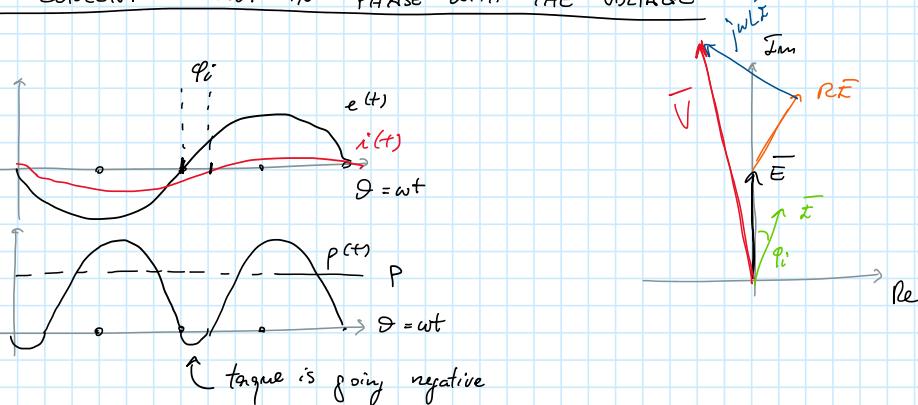
$$\boxed{\bar{V} = \bar{E} + R\bar{I} + j\omega L \bar{I}} \quad \text{VECTOR FORM}$$

$$e = \frac{dI_m}{dt} \rightarrow \bar{E} = j\omega \cdot \bar{I}_m$$



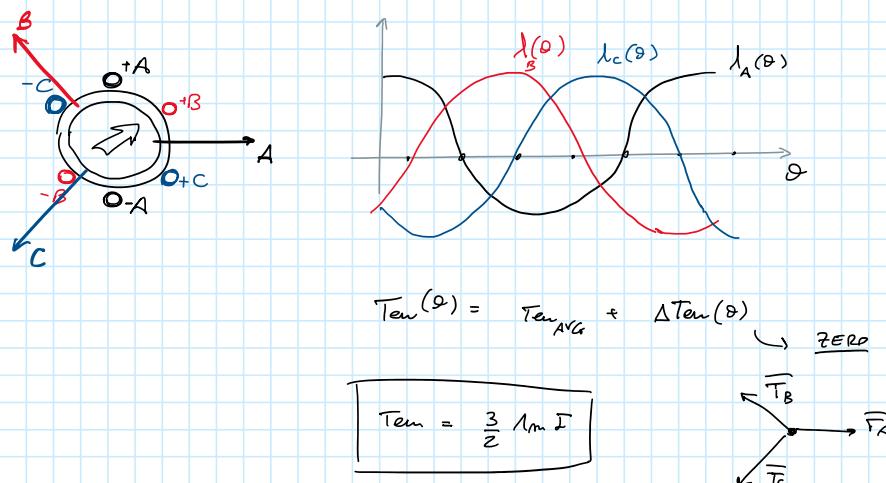


IF THE CURRENT IS NOT IN PHASE WITH THE VOLTAGE



Torque is not constant even if the current is in phase

How to make a constant torque?



EXERCISE

SPM motor, steady state operation

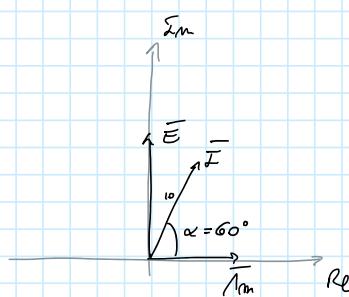
$$\lambda_m = 0.27 \text{ Vs}$$

$$R = 0.5 \Omega$$

$$L = 17.3 \text{ mH}$$

$$\omega = 200 \text{ rad/s}$$

$$\bar{I} = 10 \cdot e^{j60^\circ} \quad \left\{ \begin{array}{l} I = 10 \text{ A} \\ \alpha = 60^\circ \end{array} \right.$$



$$\bar{E} = j\omega \bar{L} I_m = j5A$$

$$\bar{I} = 5 + j8.66 A$$

$$\bar{V} = \bar{E} + R\bar{I} + j\omega L\bar{I}$$

$$= \bar{E} + (R\bar{I}_R - \omega L\bar{I}_L) + j(\bar{R}\bar{I}_I + \omega L\bar{I}_R)$$

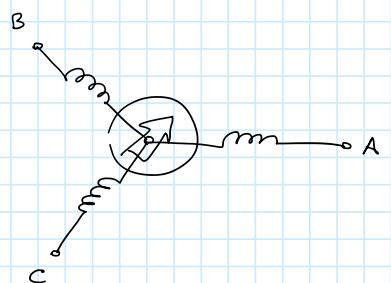
$$V_R = 0.5 \cdot 5 - 200 \cdot 0.0173 \cdot 8.66 = -24.5 V$$

$$V_I = E + 0.5 \cdot 8.66 + 200 \cdot 0.0173 \cdot 5 = 76.6 V$$

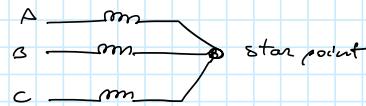
$$V = \sqrt{V_R^2 + V_I^2} = 80.45 V$$

$$\alpha_V = \arctan \frac{V_I}{V_R} = 110^\circ$$

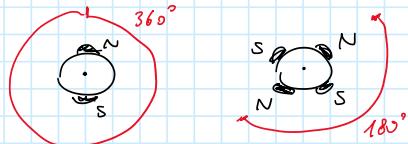
—



$$\sum i' = 0$$



Number of poles:



$$p = 1$$

$$\Theta = \Theta_m$$

$$p = 2$$

$$\Theta = 2\Theta_m$$

$$p = 3$$

$$\Theta = 3\Theta_m$$



p: number of pole pairs

The mechanical angle is always 360°

→ it is a turn of the rotor

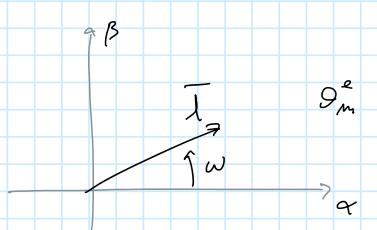
$\Theta = \Theta_m^e = p\Theta_m$
$\omega = \omega_m^e = p\omega_m$

SPACE VECTOR

$$\lambda_{m_A} = \lambda_m \cos \theta_m^e$$

$$\lambda_{m_B} = \lambda_m \cos \left(\theta_m^e - \frac{2\pi}{3} \right)$$

$$\lambda_{m_C} = \lambda_m \cos \left(\theta_m^e - \frac{4\pi}{3} \right)$$

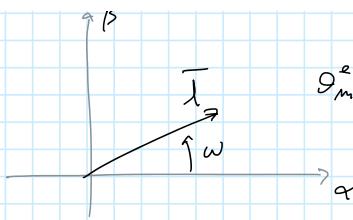


$$\theta_m^e = \omega t$$

$$\lambda_\alpha = \lambda_m \cos \theta_m^e$$

$$\lambda_\beta = \lambda_m \sin \theta_m^e$$

$$\lambda_{m_c} = \lambda_m \cos(\vartheta_m^e - \frac{1}{3}\pi)$$



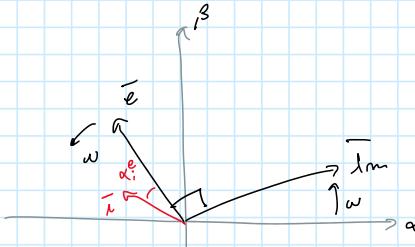
$$\vartheta_m^e = \omega t$$

$$\lambda_\alpha = \lambda_m \cos \vartheta_m^e$$

$$\lambda_\beta = \lambda_m \sin \vartheta_m^e$$

$$\begin{cases} e_A = \frac{d\lambda_\alpha}{dt} = -\omega \lambda_m \sin \vartheta_m^e \\ e_\beta = \frac{d\lambda_\beta}{dt} = -\omega \lambda_m \sin(\vartheta_m^e - \frac{2}{3}\pi) \\ e_c = \frac{d\lambda_c}{dt} = -\omega \lambda_m \sin(\vartheta_m^e - \frac{1}{3}\pi) \end{cases}$$

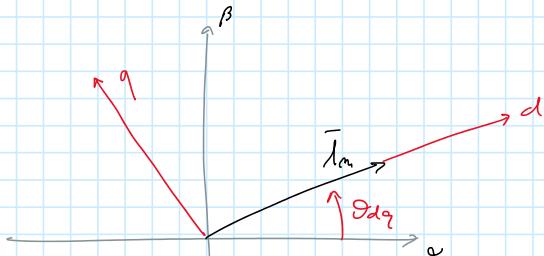
$$\bar{e}_{\alpha\beta} = \omega \lambda_m e^{j(\vartheta_m^e + \frac{\pi}{2})}$$



$$\begin{cases} i_A = I \cos(\vartheta_m^e + \alpha_i^e) \\ i_\beta = I \cos(\vartheta_m^e - \frac{2}{3}\pi, \alpha_i^e) \\ i_c = I \cos(\vartheta_m^e - \frac{1}{3}\pi, \alpha_i^e) \end{cases}$$

$$i_\beta = I e^{j(\vartheta_m^e + \alpha_i^e)}$$

• dq-reference system:



$$\bar{\lambda}_{\alpha\beta} \cdot e^{-j\vartheta_{dq}} = \bar{\lambda}_{dq} \quad [\vartheta_{dq} = \vartheta_m^e]$$

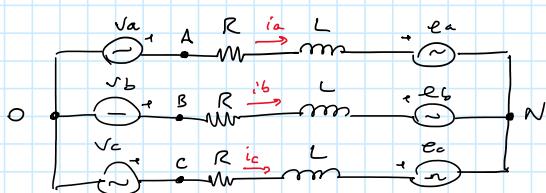
$$= \lambda_m e^{j\vartheta_m^e} e^{-j\vartheta_{dq}} = \lambda_m$$

$$\bullet \bar{e}_{dq} = \bar{e}_{\alpha\beta} e^{j(\vartheta_m^e + \frac{\pi}{2})} e^{-j\vartheta_{dq}} = \bar{e}_{\alpha\beta} e^{j\frac{\pi}{2}} = \omega \lambda_m e^{j\frac{\pi}{2}} = j\omega \lambda_m$$

$$\bullet \bar{\lambda}_{dq} = \lambda_m$$

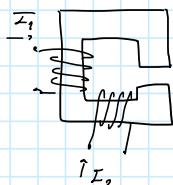
$$\bullet \bar{i}_{dq} = \bar{i}_{\alpha\beta} e^{j(\vartheta_m^e + \alpha_i^e)} e^{-j\vartheta_{dq}} = I e^{j\alpha_i^e} = I \cos \alpha_i^e - j I \sin \alpha_i^e$$

constant values in dq



$$\left\{ \begin{array}{l} \bar{\lambda}_{AN} = R_i a + L \frac{di_a}{dt} + e_a = \bar{\lambda}_a + \bar{\lambda}_{ON} \\ \bar{\lambda}_{BN} = R_i b + L \frac{di_b}{dt} + e_b = \bar{\lambda}_b + \bar{\lambda}_{ON} \\ \bar{\lambda}_{CN} = R_i c + L \frac{di_c}{dt} + e_c = \bar{\lambda}_c + \bar{\lambda}_{ON} \end{array} \right.$$

→ SPACE VECTOR : $R\bar{i} + L \frac{d\bar{i}}{dt} + \bar{e} = \bar{\lambda} + \phi$



$$\frac{\lambda_1}{I_1} = L_1 \quad (\text{self inductance})$$

$$\frac{\lambda_2}{I_1} = M_{12} \quad (\text{mutual inductance})$$

PHASE LINKAGE

$$\lambda_a = L_a i_a + M_{ab} i_b + M_{ac} i_c$$

- by geometry $M_{ab} = M_{ac} = -\frac{L_a}{2}$

$$\lambda_a = L_a i_a + M(i_b + i_c)$$

- $i_a + i_b + i_c = 0$

$$\lambda_a = i_a (L_a - M) \approx \frac{3}{2} L_a i_a$$

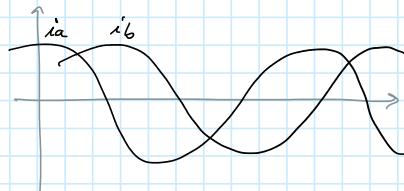
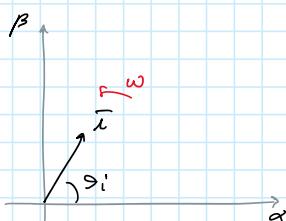
in $\alpha\beta$ reference : $\bar{\lambda} = \bar{e} \rightarrow R\bar{i} + L \frac{d\bar{i}}{dt}$

- $\bar{e} = \frac{d\bar{\lambda}_m}{dt}$

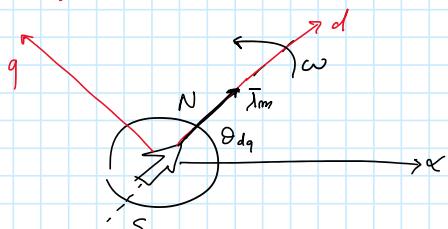
- $\frac{d}{dt}(L\bar{i}) = \frac{d(\bar{\lambda}_i)}{dt}$

$$\bar{\lambda} = R\bar{i} + \frac{d\bar{\lambda}_i}{dt}$$

$$\bar{\lambda} = \bar{\lambda}_m + \bar{\lambda}_i$$

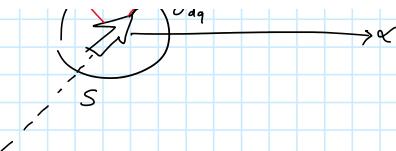


in dq reference :



Synchronous reference frame

$$\theta_{dq} = \theta_m$$



$$\alpha\beta: \bar{\lambda}_{m\alpha\beta} = \lambda_m e^{j\theta_m^e} = \lambda_m \cos \theta_m^e + j \lambda_m \sin \theta_m^e$$

$$dq: \bar{\lambda}_{m_{dq}} = \bar{\lambda}_{m\alpha\beta} e^{-j\theta_{dq}} = \lambda_m e^{j\theta_m^e} e^{-j\theta_{dq}} = \begin{cases} \lambda_{mdq} = \lambda_{m\alpha} \cos \theta_{dq} - \lambda_{m\beta} \sin \theta_{dq} = \lambda_m \\ \lambda_{mq} = 0 \end{cases}$$

$\boxed{\lambda_{mdq} = \lambda_m}$

EHF

$$\alpha\beta: \bar{e}_{\alpha\beta} = \frac{d \bar{\lambda}_{m\alpha\beta}}{dt} = \bar{e}_{dq} \cdot e^{+j\theta_{dq}} = \frac{d}{dt} (\bar{\lambda}_{m_{dq}} \cdot e^{+j\theta_{dq}})$$

$$dq: \bar{e}_{dq} = \bar{e}_{\alpha\beta} \cdot e^{-j\theta_{dq}}$$

$$\begin{aligned} \bar{e}_{dq} \cdot e^{j\theta_{dq}} &= \frac{d}{dt} (\bar{\lambda}_{m_{dq}}) \cdot e^{j\theta_{dq}} + \bar{\lambda}_{m_{dq}} \cdot \frac{d}{dt} (e^{j\theta_{dq}}) \\ &= 0 + \lambda_m \frac{d}{dt} e^{j\omega t} \\ &= j\omega \lambda_m e^{j\theta_{dq}} \end{aligned}$$

$$\theta_{dq} = \omega t = \omega_m t$$

$$\circ \lambda_{m_{dq}} = \lambda_m = \text{const.}$$

$\boxed{\bar{e}_{dq} = j\omega \lambda_m}$

CURRENT

$$\begin{aligned} \bar{i}_{\alpha\beta} &= I \cdot e^{j\theta_i} \\ &= I \cdot e^{j(\theta_m^e + \alpha_i)} \end{aligned}$$



$$\bar{i}_{dq} = \bar{i}_{\alpha\beta} e^{-j\theta_{dq}} = I \cdot e^{j(\theta_m^e + \alpha_i - \theta_{dq})} = I e^{j\alpha_i}$$

$\boxed{i_{dq} = I e^{j\alpha_i}}$

$$\begin{cases} i_d = I \cos \alpha_i \\ i_q' = I \sin \alpha_i \end{cases}$$

In steady state α_i is constant because ω_m is constant

CURRENT LINNACE

$$\bar{i}_{\alpha\beta} = L \cdot \bar{i}_{\alpha\beta}$$

$$\begin{aligned} \frac{d}{dt} \bar{i}_{\alpha\beta} &= \frac{d}{dt} (L \cdot \bar{i}_{\alpha\beta}) = L \frac{d \bar{i}_{\alpha\beta}}{dt} = L \frac{d}{dt} (\bar{i}_{dq} \cdot e^{+j\theta_{dq}}) = L \frac{d}{dt} (\bar{i}_{dq}) e^{+j\theta_{dq}} + L \bar{i}_{dq} \frac{d}{dt} (e^{+j\theta_{dq}}) = \\ &= L \frac{d}{dt} (\bar{i}_{dq}) e^{+j\theta_{dq}} + L \bar{i}_{dq} j\omega e^{+j\theta_{dq}} \end{aligned}$$

$$\theta_{dq} = \omega_m t$$

$$\begin{aligned}
 &= L \frac{d(\bar{i}_{dq})}{dt} e^{j\vartheta_{dq}} + L \bar{i}_{dq} j\omega e^{j\vartheta_{dq}} \\
 &= \frac{d}{dt} (\bar{\lambda}_{idq}) e^{j\vartheta_{dq}} + \underbrace{\bar{\lambda}_{idq} j\omega e^{j\vartheta_{dq}}}_{\text{motional EMF}}
 \end{aligned}$$

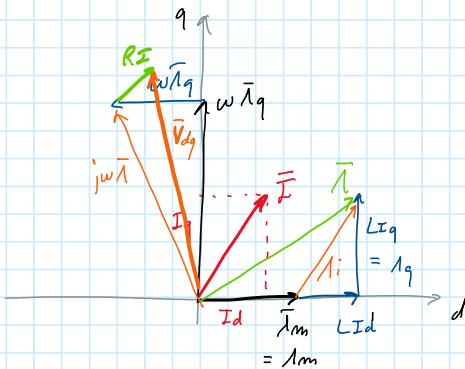
$$\vartheta_{dq} = \omega_m^e \cdot t$$

$$\begin{aligned}
 \bar{\eta}_{\alpha\beta} &= R \bar{i}_{\alpha\beta} + L \frac{d \bar{i}_{\alpha\beta}}{dt} + \bar{e}_{\alpha\beta} \quad \left(\bar{e}_{\alpha\beta} = j\omega \bar{L}_{m_{\alpha\beta}} \right) \\
 \bar{\eta}_{dq} &= R \bar{i}_{dq} + L \frac{d \bar{i}_{dq}}{dt} + j\omega \bar{L}_{idq} + \bar{e}_{dq} \quad \left(\bar{e}_{dq} = j\omega \bar{L}_{mdq} \right) \\
 &= R \bar{i}_{dq} + L \frac{d \bar{i}_{dq}}{dt} + j\omega \bar{L}_{dq} \\
 \left\{ \begin{array}{l} \bar{\eta}_d = R \bar{i}_d + L \frac{d \bar{i}_d}{dt} - \omega \bar{L}_q \\ \bar{\eta}_q = R \bar{i}_q + L \frac{d \bar{i}_q}{dt} + \omega \bar{L}_d \end{array} \right. & \left\{ \begin{array}{l} \bar{i}_d = i_{id} + i_{md} = 1_m + L_{id} \\ \bar{i}_q = i_{iq} + i_{mq} = 0 + L_{iq} \end{array} \right.
 \end{aligned}$$

In steady state all the terms are constant:

$$V_{dq} = V_d + V_q$$

$$\left\{ \begin{array}{l} V_d = R \bar{i}_d - \omega \bar{L}_q \\ V_q = R \bar{i}_q + \omega \bar{L}_d \end{array} \right. \quad \left\{ \begin{array}{l} \bar{i}_d = 1_m + L_{id} \\ \bar{i}_q = L_{iq} \end{array} \right.$$



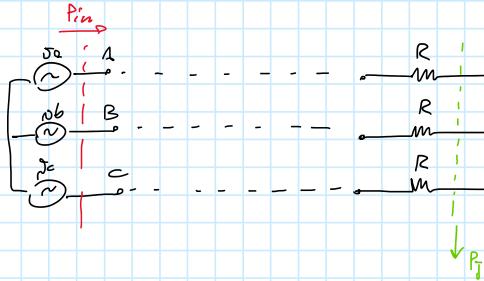
POWER BALANCE

$$\begin{aligned}
 \bar{i}_d \cdot i_d &= R \bar{i}_d^2 + i_d L \frac{d \bar{i}_d}{dt} - \omega L i_d i_q \\
 \bar{i}_q \cdot i_q &= R \bar{i}_q^2 + i_q L \frac{d \bar{i}_q}{dt} + \omega L i_d i_q + \omega L_m i_q
 \end{aligned}$$

$\sum (\dots) \cdot \frac{3}{2} \rightarrow$ 3 phase system
 \rightarrow peak to RMS ($\frac{\sqrt{2}}{2} \cdot \sqrt{2} \cdot i$)

$$\frac{3}{2} (\bar{i}_d \dot{i}_d + \bar{i}_q \dot{i}_q) = \frac{3}{2} R (i_d^2 + i_q^2) + \frac{d}{dt} \left[\frac{3}{2} \left(\frac{1}{2} L_i d^2 + \frac{1}{2} L_i q^2 \right) \right] + \frac{3}{2} \omega \left[-L_i d i_q + L_i q i_d + \lambda_m i_q \right]$$

Input power P_{in} Toule losses magnetic energy W_m Power converted P_{out} $P_{in} = P_{out}$
electromechanical power

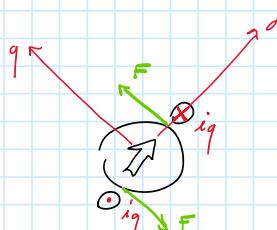


$$P_{in} = T_{em} \cdot \omega_m$$

$$\omega = p \cdot \omega_m$$

$$T_{em} = \frac{3}{2} p \lambda_m i_q$$

TORQUE



The torque depends on the pole numbers, λ_m and i_q

There is no need for i_d current as it only produces losses

EXERCISE

$$K_T = 1.05 \text{ Nm/A} \quad [\text{in RMS}]$$

$$T_{em} = K_T \cdot I_{rms} = K_T \cdot \frac{I}{\sqrt{2}} \quad \left. \begin{array}{l} \\ \\ \end{array} \right\} \quad \frac{3}{2} p \lambda_m = \frac{K_T}{\sqrt{2}} \quad \rightarrow \quad \boxed{\lambda_m = \frac{2 K_T}{3 \sqrt{2} \cdot p}}$$

- Speed $r_{pm} \rightarrow \text{rad/s}$ $\omega_m = r_{pm} \cdot \frac{2\pi}{60} \rightarrow \omega = p \cdot \omega_m (= \omega_m^e)$
- Phase to phase - Line to line voltage (tensione concatenata) $\rightarrow V_{pp_{RMS}} = 400V$

$$\sqrt{V_{pp_{RMS}}} \cdot \sqrt{2} = 322 \text{ V}$$

2.4

Synchronous SPN motor 3phase

only i_q current

data:

at no load:

with load: (current controlled)

$$2p = 2$$

$$U_{pp_{RMS}} = 38.48V$$

$$V_{pp_{RMS}} = 53.64V$$

data:

$$2p = 2$$

$$R = 2 \Omega$$

$$L = 25mH$$

$$\dot{\theta}_{\text{mech}} \approx 0$$

at no load:

$$U_{ppms} = 38.48V$$

$$\omega_{m_0} = 314.16 \frac{\text{rad}}{\text{s}}$$

with load: (current controlled)

$$\sqrt{U_{ppms}} = 53.64V$$

$$I_{\text{rms}} = 4A$$

determine: ω_m at load, torque at load

$$\sqrt{V_{NL}} = 38.48 \cdot \frac{\sqrt{2}}{\sqrt{3}} = 31.42 V$$

$$\sqrt{U_{load}} = 53.64 \cdot \sqrt{\frac{2}{3}} =$$

$$\omega = p \cdot \omega_{m_0} = 314.16 \frac{\text{rad}}{\text{s}}$$

$$I = \sqrt{2} I_{\text{rms}} = 5.66 A$$

$$\lambda_m = \frac{V}{\omega} = 0.1 V/s$$

$$I_q = I$$

$$T_{\text{em}} = \frac{3}{2} p \lambda_m I_q = \frac{3}{2} \cdot 0.1 \cdot 5.66 = 0.849 Nm$$

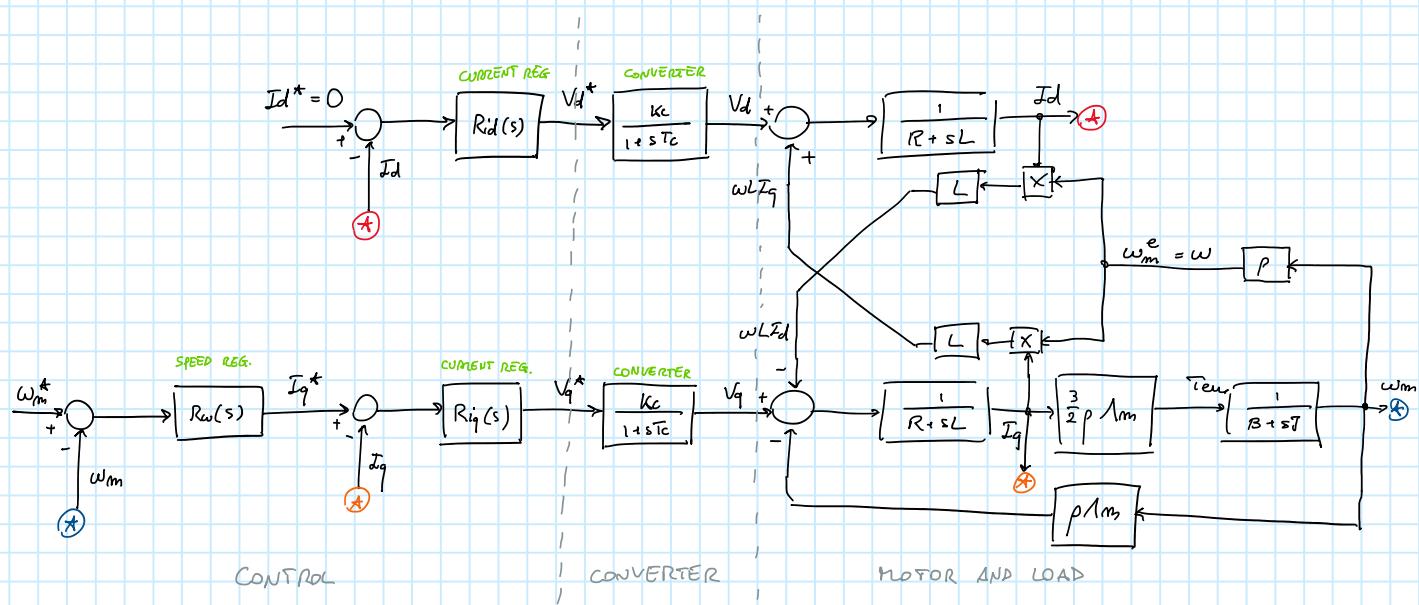
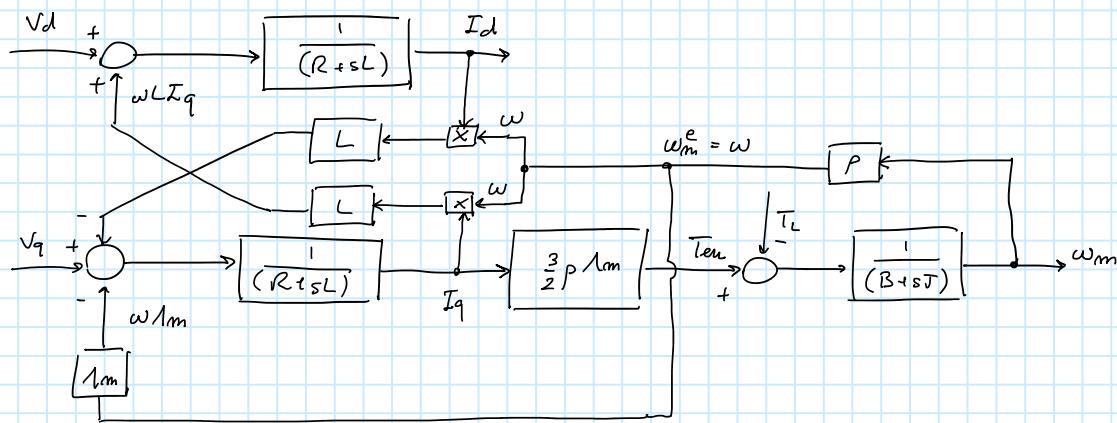
At steady state:

$$V_{dq} = \begin{cases} V_d = RId - \omega \lambda_q \\ V_q = RIq + \omega \lambda_d \end{cases} \rightarrow V^2 = V_d^2 + V_q^2 \\ = (RId - \omega \lambda_q)^2 + (RIq + \omega \lambda_d)^2$$

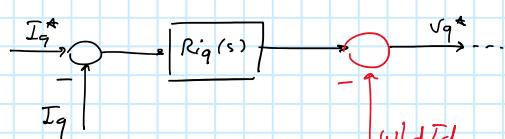
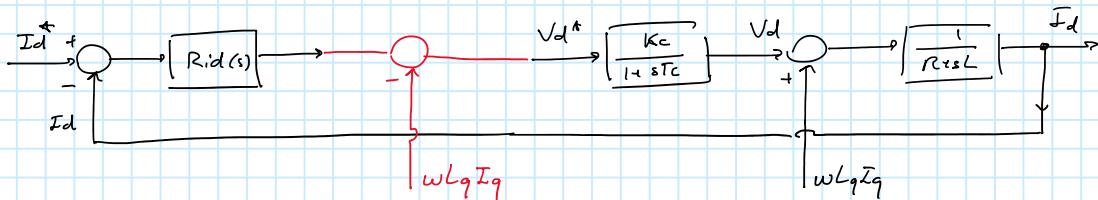
$$\lambda_{dq} = \begin{cases} \lambda_q = L_i q \\ L_d = \lambda_m + L_i d \end{cases}$$

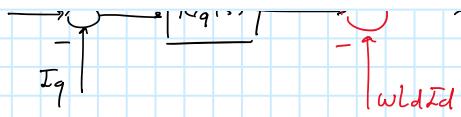
$\omega = 209 \frac{\text{rad}}{\text{s}}$

$\omega_m = \frac{\omega}{p} \rightarrow n = \frac{\omega_m \cdot 60}{2\pi} = \text{in rpm}$



CURRENT LOOP, decoupling of the two axis : dq current decoupling





EXERCISE

Design control of SPM motor:

Source:

$$V_{dc} = 400V$$

$$\lambda_m = 0.25Vs$$

$$R = 0.5\Omega$$

$$L = 10mH$$

$$2p = 8$$

$$J_m = 0.02 \text{ kg m}^2$$

inverter:

$$f_{sw} = 10 \text{ kHz}$$

$$k_C = 15$$

current loop:

$$B_w = 500 \text{ Hz}$$

$$\varphi_m \geq 60^\circ$$

load:

$$T_{load} \propto \omega_{load} ; \frac{262 \text{ Nm}}{600 \text{ rpm}} , J_c = 48 \text{ kg m}^2$$

$$k_{gear} = \frac{1}{10}$$

speed loop:

$$B_w = 32 \text{ Hz}$$

$$\varphi_m \geq 60^\circ$$

$$V_p = 400V \rightarrow V = \frac{V_p}{\sqrt{3}} = 230V$$

$$\bullet \text{ gearbox: } \omega_L = k_f \omega_m \quad \omega_m \rightarrow [k_f] \rightarrow \omega_L$$

$$T_m = k_f T_L \quad T_m \rightarrow \left[\frac{1}{k_f} \right] \rightarrow T_L$$

$$\bullet \text{ From data: } T_L \propto \omega_L = \frac{262}{100} \frac{\text{Nm}}{\text{r.p.m}}$$

$$T_L = B_L \omega_L + J_L \frac{d\omega_L}{dt} \rightarrow B_L = \frac{262}{100 \cdot \frac{2\pi}{60}} = 25 \text{ Nm s}$$

$$\begin{aligned} T_m &= k_f T_L = k_f J_L \frac{d\omega_L}{dt} + k_f B_L \omega_L = \\ &= k_f^2 J_L \frac{d\omega_m}{dt} + k_f^2 B_L \omega_m \end{aligned}$$

$$J_L' = 0.1^2 \cdot 48 = 0.48 \text{ kg m}^2$$

$$B = 0.1^2 \cdot 25 = 0.25 \text{ Nm s}$$

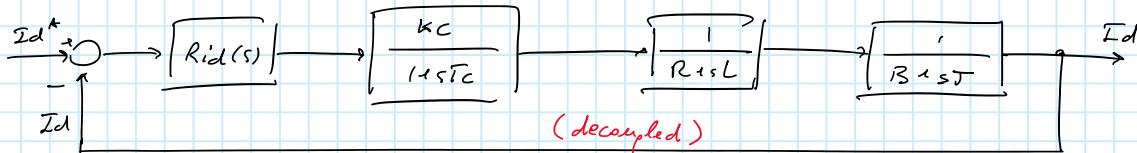
$$\bullet \text{ Total inertia: } J = J_m + J_L' = 0.02 + 0.48 = 0.5 \text{ kg m}^2$$

$$\bullet \tau_E = \frac{L}{R} = \frac{0.01}{0.5} = 0.2 \text{ s}$$

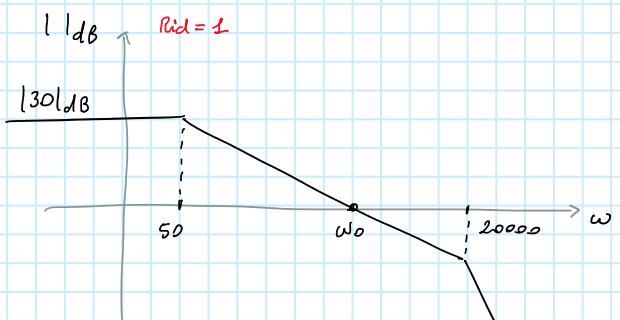
$$\bullet T_m = \frac{J}{B} = \frac{0.5}{0.25} = 2 \text{ s}$$

$$\bullet T_C = \frac{1}{2f_s} = \frac{1}{2 \cdot 10000} = 50 \text{ ms}$$

$$\begin{aligned} \left[\frac{1}{R+L} \right] &= \frac{1}{R} - \frac{1}{1+sT_e} = \frac{z}{1+s/50} \\ \left[\frac{1}{B+sT} \right] &= \frac{1}{B} - \frac{1}{1+sT_m} = \frac{4}{1+s/0.5} \quad (\text{NEGLECTED}) \\ \left[\frac{K_C}{1+sT_C} \right] &= \frac{15}{1+s/20000} \end{aligned}$$



$$\rightarrow \left[R_{id}(s) \right] \rightarrow \left[\frac{30}{(1+s/50)(1+s/20000)} \right]$$



$$\left| \frac{30}{(1+j\omega/50)(1+j\omega/20000)} \right| = 1$$

$$\approx \sqrt{\frac{30}{1 + \left(\frac{\omega_0}{50}\right)^2}} = 1 \rightarrow \omega_0 \approx 1500 \frac{\text{rad}}{\text{s}}$$

- With a PI regulator: compensate the first pole + pole in the origin $R_P := K_i \cdot \frac{1+sT_i}{s}$

$$f_{BW} = 500 \text{ Hz} \rightarrow \omega_{BW} = 2\pi \cdot 500 = 3140 \frac{\text{rad}}{\text{s}}$$

$$T_i = T_e = \frac{1}{50} = 0.02$$

$$K_i = ? \quad \left| K_i \cdot \frac{(1+sT_i)}{s} \cdot \frac{30}{(1+sT_e)(1+sT_m)} \right| = \frac{K_i}{s=j\omega_{BW}} \cdot \frac{30}{\sqrt{1 + \left(\frac{3140}{20000}\right)^2}} = 1$$

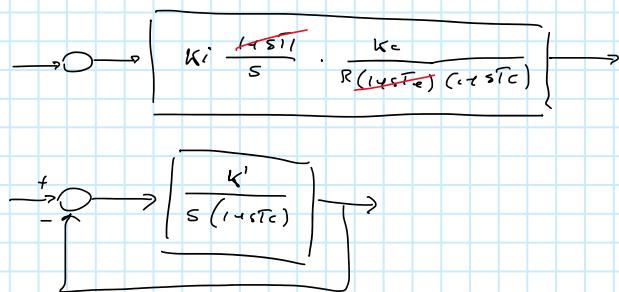
$$K_i \approx 106$$

$$K_p = K_i T_i = 2.16$$

$$\varphi_m = 180^\circ + \angle G(3140) = 180^\circ - 90^\circ - \arctan \frac{3140}{20000} = 81^\circ \quad (\checkmark)$$

Using root locus:

$$T_i = T_e$$



$$\frac{G}{1+G} = \frac{K'}{s(1+stc) + K'}$$

$$s(1+stc) + K' = 0$$

$$s^2 T_c + s + K' = 0$$

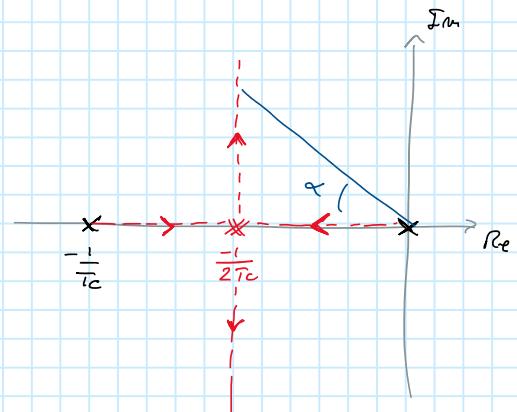
$$\delta = \frac{-1 \pm \sqrt{1 - 4T_c K'}}{2T_c}$$

- $K' = 0 \rightarrow s_{1,2} = \frac{-1 \pm 1}{2T_c} = \begin{cases} 0 \\ -\frac{1}{T_c} \end{cases}$

- $K'_A \rightarrow (-\sqrt{T_c} K'_A) = 0$

$$K'_A = \frac{1}{\sqrt{T_c}} \rightarrow s_{1,2} = -\frac{1}{2T_c}$$

- $K' : \Delta < 0 \rightarrow$ two imaginary solution



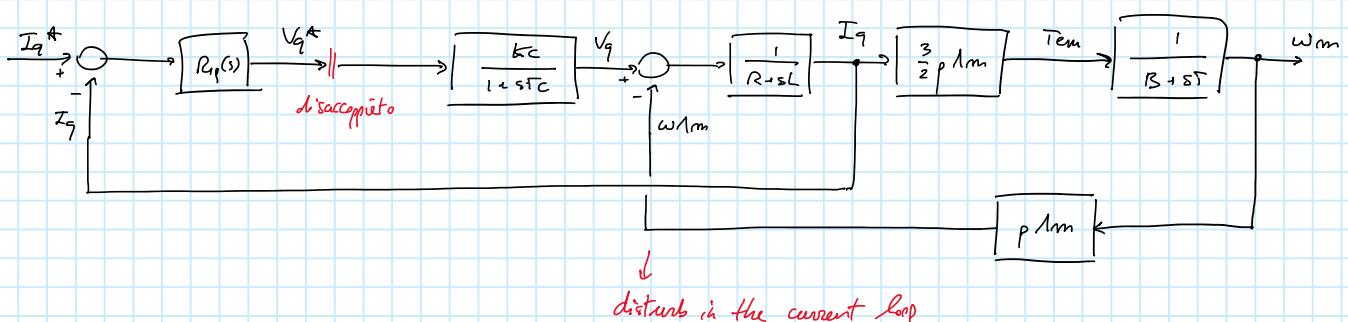
Good solution for $\xi \geq 0.707$

$$\xi = \cos \alpha \rightarrow \alpha = 45^\circ : \text{Re} = \text{Im}$$

$$\left| \frac{1}{2T_c} \right| = \left| \frac{1}{2T_c} \sqrt{1 - 4T_c K'} \right| \rightarrow K' = \frac{1}{2T_c} \rightarrow K_i = \frac{K' R}{K_c} = 333$$

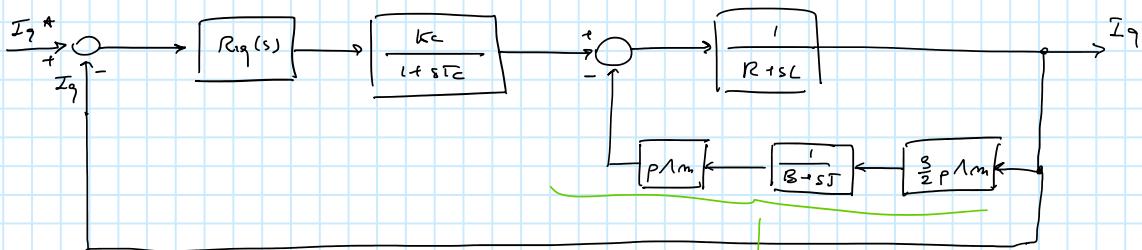
$$K_p = 6.67$$

I_q SIDE



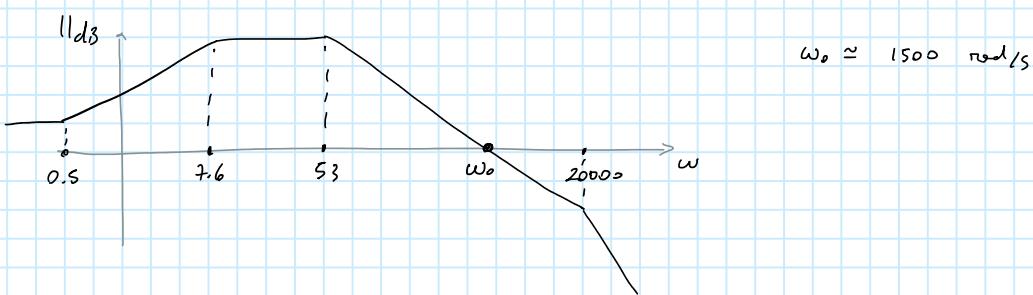
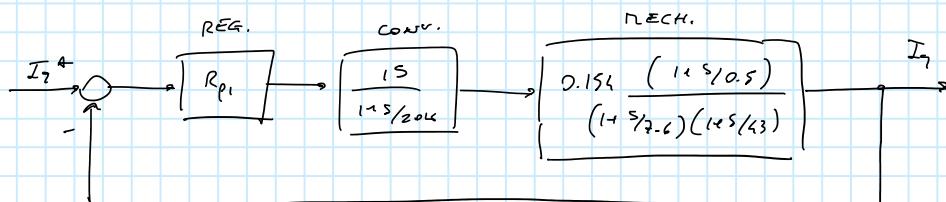
STRATEGIES

- 1) the speed has a dynamic slower than the current \rightarrow neglect the ω_{lm}
- 2) compensate ρ_{lm} in the control loop in V_q^*
- 3) introduce the mechanical feedback



$$\frac{G}{1+GH} = \frac{\frac{1}{R} \cdot \frac{1}{1+sTc}}{1 + \frac{1}{R} \cdot \frac{1}{1+sTc} \cdot \frac{3}{2} p1lm \frac{1}{B} \frac{1}{1+sTm} \cdot p1lm} =$$

$$= \frac{\frac{B}{\frac{3}{2}(p1lm)^2}}{\frac{RB}{\frac{3}{2}(p1lm)^2} \cdot \frac{1+sTm}{(1+sTc)(1+sTm)}} = 0.154 \cdot \frac{(1+s/0.5)}{(1+s/7.6)(1+s/43)}$$



PI regulator : $k_i \left(\frac{1+sT_i}{s} \right)$, $T_i = \frac{1}{7.6}$

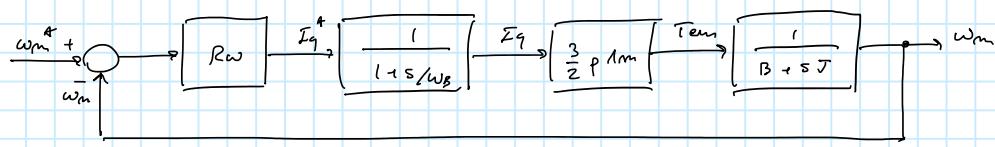
$$\left| k_i \left(\frac{1+sT_i}{s} \right) \cdot 2.31 \cdot \frac{1+s/0.5}{(1+s/7.6)(1+s/43)} \right| = 1$$

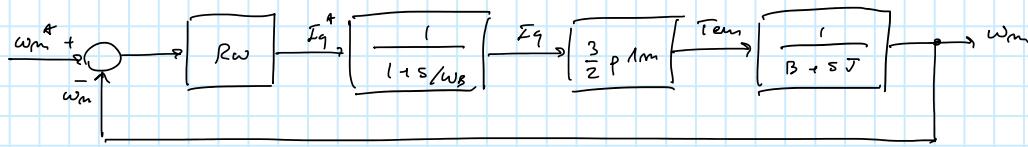
$s = j\omega_B = j3140$

$k_i = 16$, $k_p = k_i \cdot T_i = 2.1$

$$\varphi_m = 180 - 90 - \arctan\left(\frac{3140}{0.5}\right) - \arctan\left(\frac{3140}{48}\right) - \arctan\left(\frac{3140}{20000}\right) = 82^\circ \quad \checkmark$$

SPEED REGULATOR :



SPEED REGULATOR :

$$\begin{aligned}
 G_{ew}(s) &= R_w(s) \cdot \frac{3}{2} p 1/m \cdot \frac{1}{B} \frac{1}{(1+sT_m)(1+s/w_B)} \\
 &= \frac{6K_i}{s} \frac{1}{(1+s/3140)}
 \end{aligned}$$

• $w_B = 200 \text{ rad/s}$

$$|G_{ew}(200)| = 1 \rightarrow K_i = 33.4, K_p = K_i T_i = 66.8$$

$$\varphi_m = 180 - 90 - \arctan\left(\frac{200}{3140}\right) = 86^\circ$$

LIMITI DI FUNZIONAMENTO

$$\bar{I} = I_d + j I_q$$

$$\bar{I} = I_d + j I_q \quad \left\{ \begin{array}{l} I_d = I_m + L I_d \\ I_q = L I_d \end{array} \right.$$

$$\bar{V} = V_d + j V_q \quad \left\{ \begin{array}{l} V_d = R I_d - \omega I_q \\ V_q = R I_d + \omega I_d \end{array} \right.$$

At limit we can neglect $R I_d^0$
and consider only the speed: $V_d \approx -\omega I_q$
 $V_q \approx +\omega I_d$

$$T_{em} = \frac{3}{2} p 1/m I_q$$

$$\text{limits : } \begin{cases} I \leq I_N \\ V \leq V_N \end{cases}$$

electrical source \rightarrow power electronics \rightarrow motor

\Rightarrow max T_{em} and w_m ?

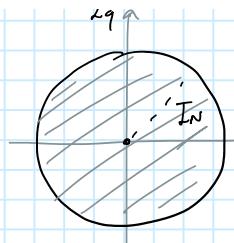
$$I_d^2 + I_q^2 \leq I_N^2$$



$$V_d^2 + V_q^2 \leq V_N^2$$

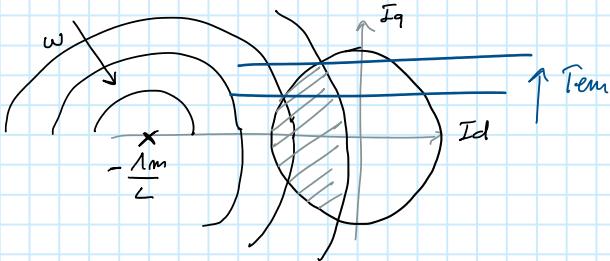
$$(\omega I_q)^2 + (\omega I_d)^2 \leq V_N^2$$

$$(L I_q)^2 + (I_m + L I_d)^2 \leq \left(\frac{V_N}{\omega}\right)^2$$



$$(\omega I_q) + (\omega I_d) = v_N \\ (L I_q)^2 + (I_m + L I_d)^2 = \left(\frac{v_N}{\omega}\right)^2$$

$$I_q^2 + \left(\frac{I_m}{L} + I_d\right)^2 = \left(\frac{v_N}{\omega L}\right)^2$$



Torque: $\frac{3}{2} \rho \lambda_m I_N$ → max torque @ ω_m very low (zero) → no v_N imperative
consider only I_N

$$I_q = \max I = I_N$$

$$T_{em, \text{max}} = \frac{3}{2} \rho \lambda_m I_N \quad \begin{cases} I_q = I_N \\ I_d = 0 \end{cases} \rightarrow \begin{cases} I_d = \lambda_m + L I_d = \lambda_m \\ I_q = L I_q = L I_N \end{cases}$$

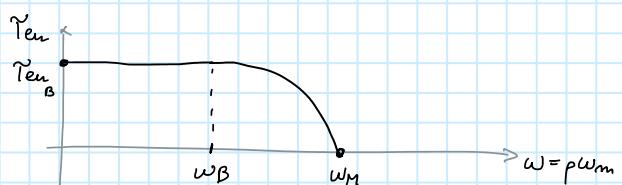
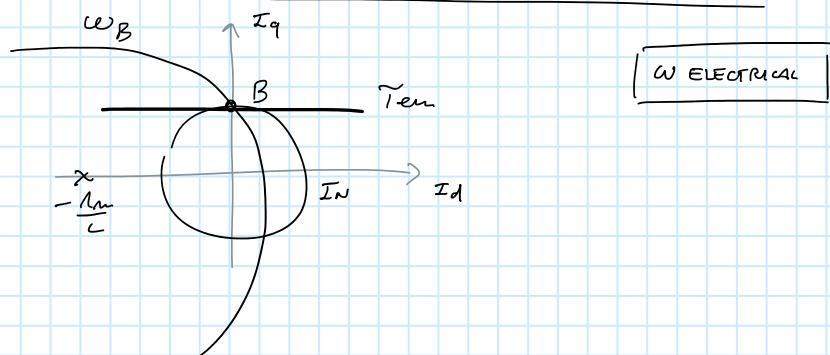
$$\begin{cases} V_d = -\omega L I_N \\ V_q = \omega \lambda_m \end{cases}$$

$$V^2 = \omega^2 (\lambda_m^2 + (L I_N)^2) = \omega^2 \lambda^2$$

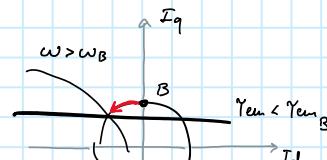
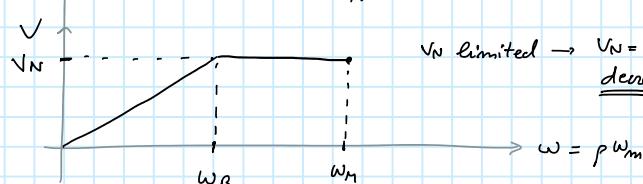
$$(V = v_N)$$

$$\omega = \frac{v_N}{\lambda} = \frac{v_N}{\sqrt{\lambda_m^2 + (L I_N)^2}}$$

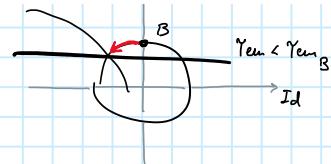
BASE SPEED ω_B



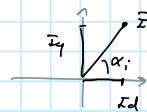
v_N limited → $v_N = \omega \lambda$
decrease λ



$$V^2 = \omega^2 (\lambda_d^2 + \lambda_q^2) = \omega^2 [(\lambda_m + L\lambda_d)^2 + (L\lambda_q)^2]$$



$$\lambda_N^2 = \lambda_d^2 + \lambda_q^2 \quad \begin{cases} \lambda_d = \lambda_N \cos \alpha_i \\ \lambda_q = \lambda_N \sin \alpha_i \end{cases}$$



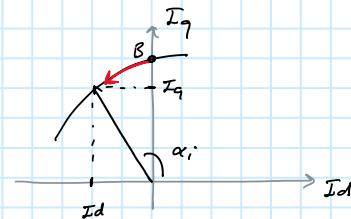
imposing $\omega_m^e = \omega$

$$V_N^2 = \omega_m^{e^2} \left[(\lambda_m + L\lambda_N \cos \alpha_i)^2 + (L\lambda_N \sin \alpha_i)^2 \right]$$

$$\frac{V_N^2}{\omega_m^{e^2}} = \lambda_m^2 + 2\lambda_m L\lambda_N \cos \alpha_i + (L\lambda_N \cos \alpha_i)^2 + (L\lambda_N \sin \alpha_i)^2 = \lambda_m^2 + 2\lambda_m L\lambda_N \cos \alpha_i + L^2 \lambda_N^2$$

FLUX WEAKENING

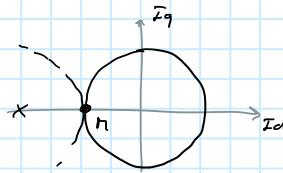
$$\cos \alpha_i = \frac{\left(\frac{V_N}{\omega_m^e} \right)^2 - \lambda_m^2 - L^2 \lambda_N^2}{2 \lambda_m L \lambda_N} < 0$$



Maximum speed:

$$\cos \alpha_i = 1$$

$$\begin{cases} I_q = 0 \\ \lambda_d = -\lambda_N \end{cases}$$



$$\omega_m = \frac{V_N}{\lambda_m - L\lambda_N}$$

MAX/MUM SPEED

EXERCISE

$$R \approx 0$$

$$V_{pp} = 250V$$

$$V = \frac{\sqrt{2}}{\sqrt{3}} \cdot 250 = 204V \text{ (phase peak)}$$

$$L = 0.15 \text{ mH}$$

$$I_{max} = 209A$$

$$I = 200 \cdot \sqrt{2} = 283A$$

$$z_p = 8$$

$$K_E = 0.6 \frac{V_{max,pp}}{rad/s} \rightarrow V = \sqrt{\frac{2}{3}} \cdot K_E \omega_m = \lambda_m \omega = \lambda_m \rho \omega_m$$

$$\lambda_m = \sqrt{\frac{2}{3}} \frac{K_E}{\rho} = 0.122 \text{ Vs}$$

maximum speed at no load :

$$\lambda_d = 0, \quad \lambda_d = \lambda_m, \quad V_d = 0$$

$$\lambda_q = 0, \quad \lambda_q = 0, \quad V_q = \omega \lambda_m$$

$$V_N = \omega_0 \lambda_m \rightarrow \omega_0 = \frac{V_N}{\lambda_m} = \frac{204}{0.122} = 1672 \frac{\text{rad}}{\text{s}}$$

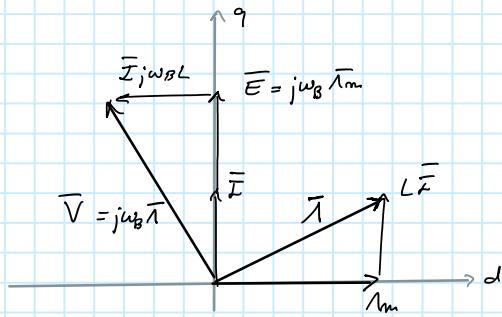
$$\omega_{m_0} = \frac{\omega_0}{s} = 418 \frac{\text{rad}}{\text{s}} \quad n_0 = \omega_{m_0} \cdot \frac{60}{2\pi} = 4000 \text{ rpm}$$

BASE POINT

$$\omega_B = \frac{V_N}{P} = \frac{V_N}{\sqrt{(I_m)^2 + (L_{IN})^2}} = \frac{204}{\sqrt{(0.122)^2 + (0.15e-3 \cdot 283)^2}} = 1580 \frac{\text{rad}}{\text{s}}$$

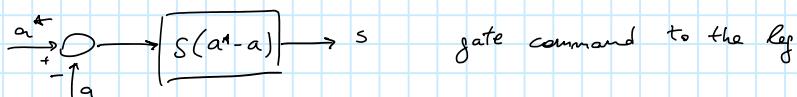
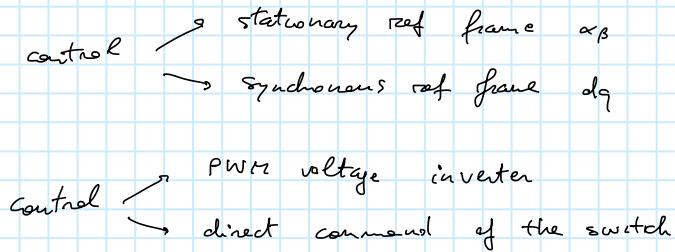
$$\omega_{mB} = \frac{\omega_B}{P} = 375 \frac{\text{rad}}{\text{s}} \quad n_B = 3780 \text{ rpm}$$

$$T_{em_B} = \frac{3}{2} P I_m \bar{I}_N = 207 \text{ Nm}$$

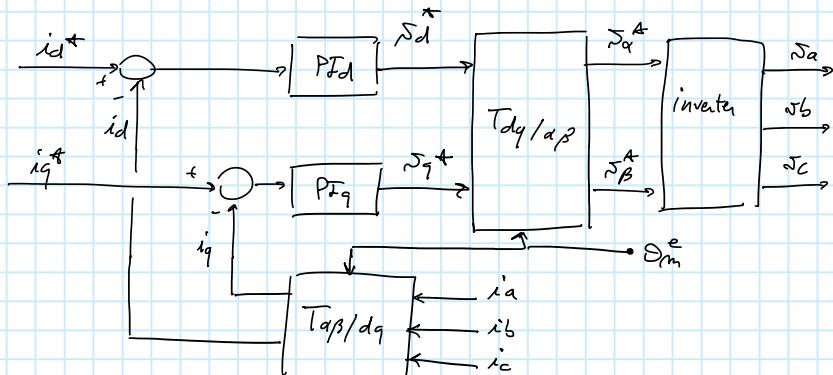
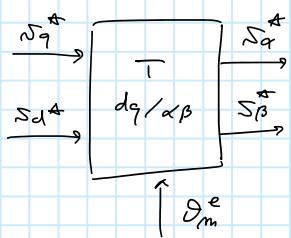
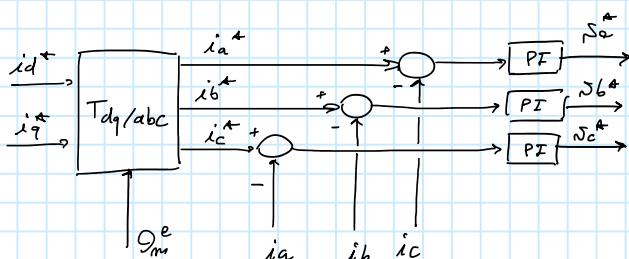


MAXIMUM SPEED

$$\omega_h = \frac{V_N}{I_{m_0} - L_{IN}} = 2570 \frac{\text{rad}}{\text{s}} \rightarrow n_h = \frac{2570}{2} \cdot \frac{60}{2\pi} = 6125 \text{ rpm}$$

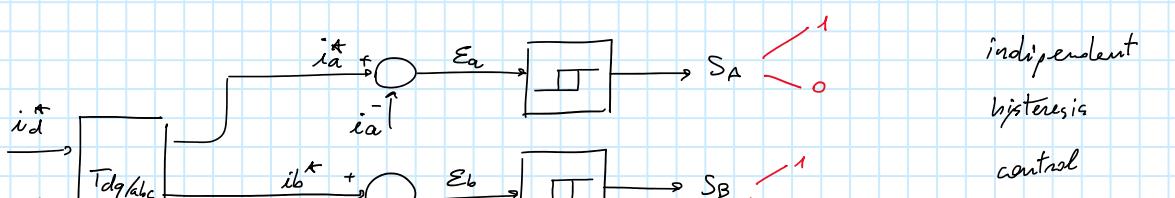
3 PHASE CURRENT CONTROL

- The motor works in $\alpha\beta$ quantities

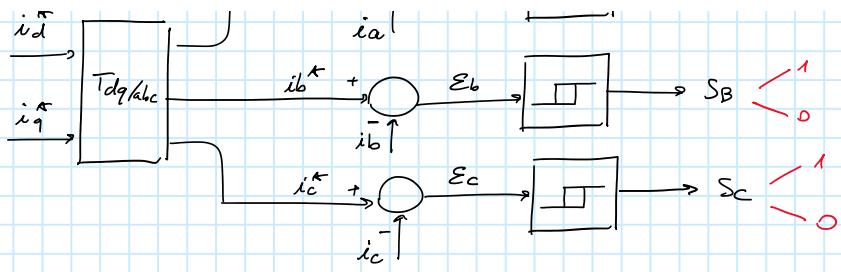
PI in abcDRAWBACK

the PI receive a sinusoidal error
 → at steady state the error is $\neq 0$

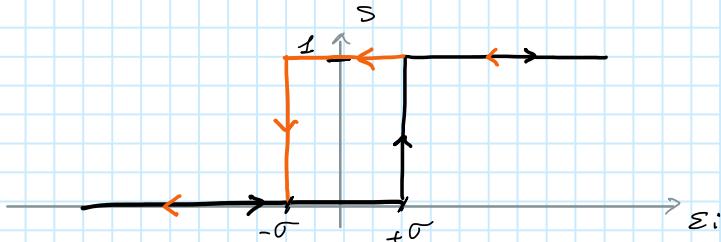
same if we use $T_{dq/d\beta}$ and use i_d^* i_q^*

DIRECT CONTROL Hysteresis

independent
hysteresis
control



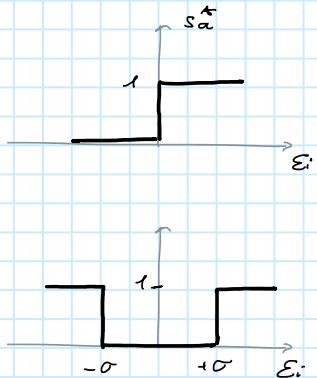
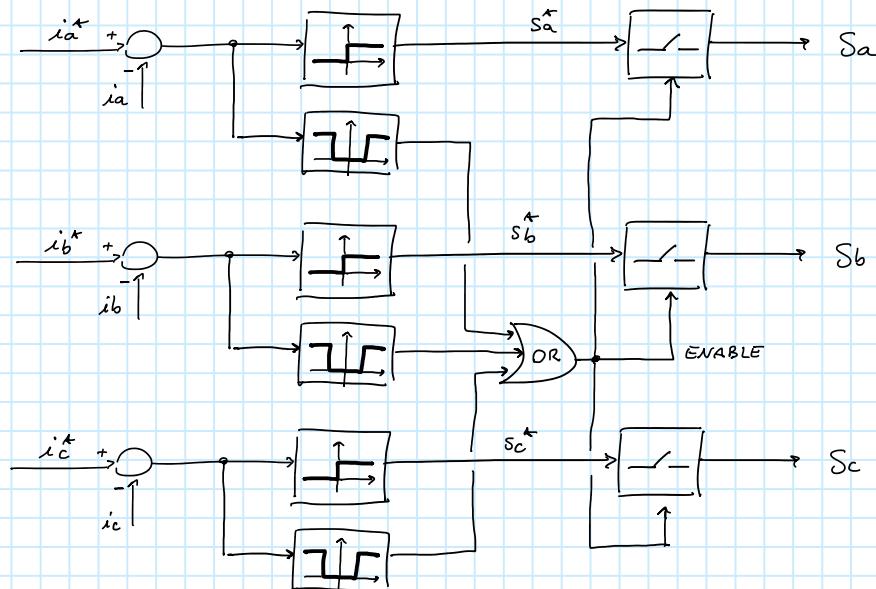
hysteresis
control
→ the error can be 20
because every phase
can give an error



- $e_i < 0 : i_i > i_i^* \rightarrow$ REDUCE
= open switch
- $e_i > 0 : i_i < i_i^* \rightarrow$ INCREASE
= closed switch

- $\varepsilon \leq 25^\circ$
- few depends on the current behaviour
 - high sw losses
 - distortion of voltage

DEPENDENT 3ph Hysteresis control



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SPM

$\bar{\lambda}_m$

$$\bar{e}_{\alpha\beta} = \frac{d\bar{\lambda}_m}{dt}$$

$$\bar{\lambda}_{\alpha\beta} = R\bar{i}_{\alpha\beta} + L \frac{di_{\alpha\beta}}{dt} \rightarrow \bar{e}_{\alpha\beta} = R\bar{i}_{\alpha\beta} + \frac{d\bar{\lambda}_{\alpha\beta}}{dt}$$

$$\bar{\lambda}_{\alpha\beta} = \bar{\lambda}_m + \bar{L}i_{\alpha\beta}$$

| in dq | At steady state all the quantities are constants

$$\bar{i}_{dq} = \bar{i}_{\alpha\beta} e^{-j\theta_m}$$

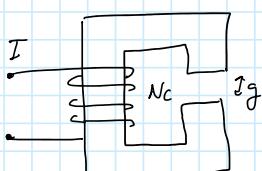
$$\begin{aligned} \bar{\lambda}_{dq} &= R\bar{i}_{dq} + L \frac{di_{dq}}{dt} + j\omega \bar{i}_{dq} \quad \rightarrow \quad \bar{\lambda}_d = R_i d + \frac{di_d}{dt} - \omega i_q \\ &\quad \rightarrow \quad \bar{\lambda}_q = R_i q + \frac{di_q}{dt} + \omega i_d \end{aligned}$$

$$T_{em} = \frac{3}{2} p \lambda_m i_q$$

IPM MOTORS : same equations in $\alpha\beta$ and in dq

$$\left\{ \begin{array}{l} \lambda_d = \lambda_m + L_i i_d \\ \lambda_q = L_q i_q \end{array} \right.$$

the inductances
are DIFFERENT



$$L = \frac{\lambda}{I} = \frac{N_c \phi}{I} \quad \phi \propto \frac{1}{g}$$

$L_d < L_q$ in the q axis the flux flows in the air gap
in the d axis the flux is inside the iron

$$\left\{ \begin{array}{l} \bar{\lambda}_d = R_i d + L_d \frac{di_d}{dt} - \omega L_q i_q \\ \bar{\lambda}_q = R_i q + L_q \frac{di_q}{dt} + \omega L_d i_d + \omega \lambda_m \end{array} \right.$$

$$\left\{ \begin{array}{l} \text{d}id = R_{id} i^2 + id L_d \frac{di}{dt} - \omega L_q i d i q \\ i q i q = R_{iq}^2 + i q L_q \frac{dq}{dt} + \omega L_m i q + \omega L_d i d i q \end{array} \right.$$

$\downarrow +, \times \frac{3}{2}$ to get POWER

$$Li \frac{di}{dt} = \frac{d}{dt} \left(\frac{1}{2} Li^2 \right)$$

$$\frac{3}{2} (R_{id} i^2 + R_{iq} i^2) = \frac{3}{2} R (i_d^2 + i_q^2) + \frac{3}{2} \left[\frac{d}{dt} \left(\frac{1}{2} L_d i d^2 + \frac{1}{2} L_q i_q^2 \right) \right] + \frac{3}{2} \omega (i_m i_q + L_d i_q i_d - L_q i_d i_q)$$

Pin Total losses Magnetic energy variation Pm

$$P_m = \frac{3}{2} \omega [i_m i_q + (L_d - L_q) i d i q]$$

SALIENCE $\xi = \frac{L_q}{L_d} = 2 : 8$

$$T_m = \frac{P_m}{\omega_m} = \frac{3}{2} \rho [\underbrace{i_m i_q}_{\text{PM component}} + \underbrace{(L_d - L_q) i d i q}_{\text{Reluctance Torque component}}]$$

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IPM MOTOR UNIT OPERATING CONDITION

$$\textcircled{1} \quad I_d^2 + I_q^2 = I_N^2$$

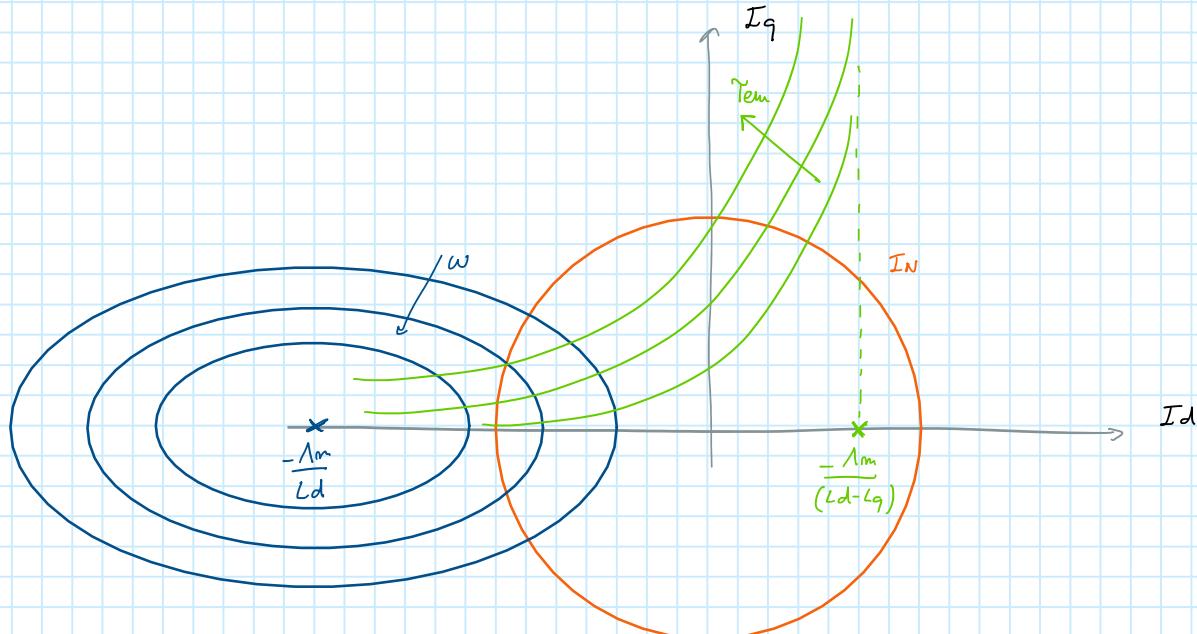
$$\textcircled{2} \quad V_d^2 + V_q^2 = V_N^2$$

$$\text{at high speed} \quad R \approx 0 \quad \rightarrow \quad (\omega I_q)^2 + (\omega I_d)^2 = V_N^2$$

$$(L_q I_q)^2 + (I_m + L_d I_d)^2 = \left(\frac{V_N}{\omega}\right)^2$$

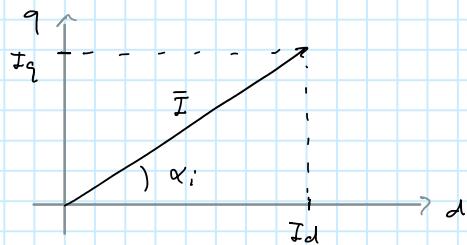
$$\textcircled{3} \quad T_{em} = \frac{3}{2} p \left[I_m I_q + (L_d - L_q) I_q I_d \right]$$

$$= \frac{3}{2} p \left[I_m + (L_d - L_q) I_d \right] \cdot I_q$$



MTPA:

Maximum Torque per Ampere, at zero speed the voltage is zero \Rightarrow no V_N limit



$$T_{em} = \frac{3}{2} P \left[I_m I_N \sin \alpha_i + (L_d - L_q) I_N^2 \sin \alpha_i \cos \alpha_i \right]$$

$$\frac{d T_{em}}{d \alpha_i} = 0 \rightarrow I_m I_N \cancel{\cos \alpha_i} + (L_d - L_q) I_N^2 (-\sin^2 \alpha_i + \cos^2 \alpha_i) = 0$$

$$I_m \cos \alpha_i + (L_d - L_q) I_N (2 \cos^2 \alpha_i - 1) = 0$$

$$[2 \cdot (L_d - L_q) I_N] \cos^2 \alpha_i + I_m \cos \alpha_i + [-(L_d - L_q) I_N] = 0$$

| $\overline{\text{TPA}}$ | $\rightarrow \cos \alpha_i = \dots$ solution in the II quadrant $\rightarrow \alpha_i \geq 90^\circ$
 $\Rightarrow \boxed{\cos \alpha_i \leq 0}$

- Particular case: synchronous reluctance motor ($I_m = 0$)

$$\cos \alpha_i = -\frac{\sqrt{2}}{2} = -\frac{\sqrt{2}}{2} \rightarrow \alpha_i = 90 + 45^\circ = 135^\circ$$

:

$$I_d = I_N \cos \alpha_i_{\text{TPA}}$$

\rightarrow

$$V_d = -\omega L_q = -\omega L I_q$$

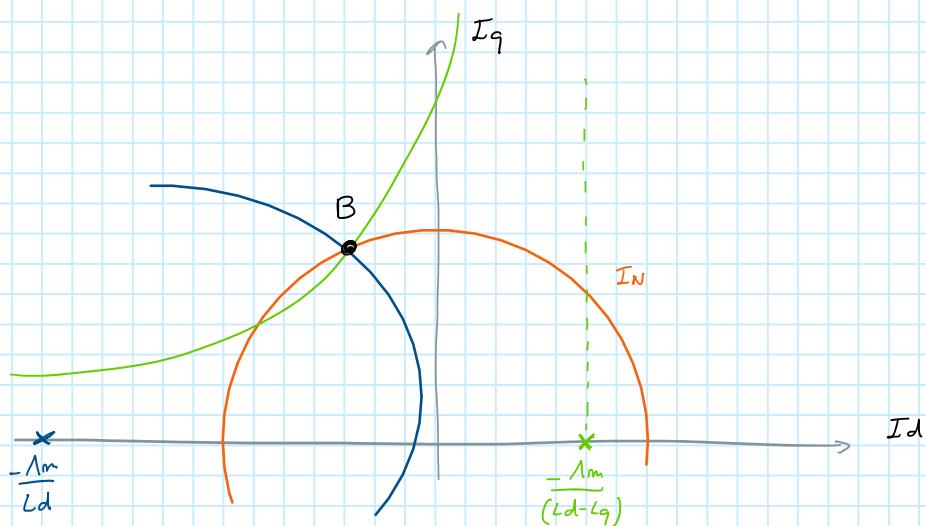
$$I_q = I_N \sin \alpha_i_{\text{TPA}}$$

$$V_d = \omega L I_d = \omega (I_m + L_d I_d)$$

$$V = \sqrt{V_d^2 + V_q^2} = \omega \sqrt{I_d^2 + I_q^2} = \omega \lambda$$

λ is fixed by the current

$$V \leq V_N \rightarrow \boxed{\omega_B \leq \frac{V_N}{\lambda}} \quad \underline{\text{speed in BASE POINT}}$$



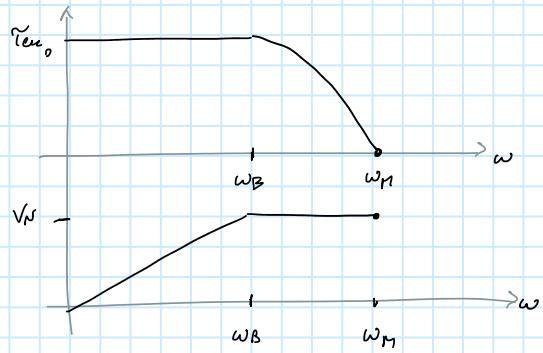
If $V = V_N$ is fixed, the ellipse collapses with increasing speed

\rightarrow Field weakening

→ Field weakening

$$V_N = \omega \lambda$$

$$\lambda = \begin{cases} \lambda_q = L_q I_q & \rightarrow \text{Lower } I_q \\ \lambda_d = \lambda_m + L_d I_d & \rightarrow \text{More negative } I_d \end{cases} \quad \left. \right\} \underline{\text{Torque DECREASES}}$$



Maximum speed :

$$\boxed{I_d = -I_N, \quad I_q = 0}$$
$$\boxed{\omega_M = \frac{V_N}{\lambda_m - L_d I_N}}$$

26/11 mtpa ipm

lunedì 14 dicembre 2020 15:56

EXERCISE

(PM motor)

$$\lambda_m = 0.4 V_s$$

$$L_d = 15.9 \text{ mH}$$

$$L_q = 21.2 \text{ mH}$$

$$2p = 4$$

$$R \approx 0$$

Determine

- $\tau_{em, max}$

- $\omega(\tau_{em, max}) = \omega_B$ (mtpa base point)

- ω_m maximum speed

- τ_{em} at $\omega_m = 2200 \text{ rpm}$

MTPA : only I_N limit

$$\cos \alpha_i^* = \frac{-\lambda_m + \sqrt{\lambda_m^2 + \delta(L_d - L_q)^2 I_N^2}}{\delta(L_d - L_q) I_N} = -0.236 \quad \Rightarrow \quad \alpha_i^* = 103.6^\circ$$

$$\begin{cases} I_d = I_N \cos \alpha_i^* = -4.77 \text{ A} \\ I_q = I_N \sin \alpha_i^* = 18.44 \text{ A} \end{cases}$$

$$\begin{cases} \lambda_d = \lambda_m + L_d I_d = 0.325 V_s \\ \lambda_q = L_q I_q = 0.412 V_s \end{cases}$$

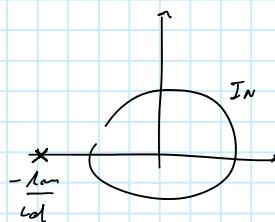
$$T_{em} = \frac{3}{2} p (\lambda_d I_q - \lambda_q I_d) = 24.78 \text{ Nm}$$

$$R \approx 0 \rightarrow V = \omega \lambda \quad \rightarrow \quad \omega_B = \frac{V_N}{\lambda_B} = \frac{210}{\sqrt{0.325^2 + 0.412^2}} = 600 \text{ rad/s}$$

$$\rightarrow \omega_{mB} = \frac{\omega_B}{p} \cdot \frac{60}{2\pi} = 1910 \text{ rpm}$$

maximum speed :

$$\frac{\lambda_m}{L_d} = \frac{0.4}{0.015 \text{ A}} = 25.16 \text{ A} > I_N \rightarrow$$



$$\begin{cases} I_d = -I_N \\ I_q = 0 \end{cases}$$

$$\begin{cases} \lambda_d = \lambda_m - LdI_N = \lambda_m \\ \lambda_q = 0 \end{cases} \rightarrow \omega_r = \frac{\sqrt{n}}{\lambda_m} = 2560 \frac{\text{rad}}{\text{s}}$$

$$\omega_{mM} = \frac{\omega_r}{P} \frac{60}{2\pi} = 12230 \text{ rpm}$$

\tilde{T}_{em} at 2200 rpm :

$$\omega = \omega_r \cdot P \cdot \frac{2\pi}{60} = 660 \frac{\text{rad}}{\text{s}} = \frac{\sqrt{n}}{\lambda} \rightarrow \lambda = \frac{\sqrt{n}}{\omega} = 0.456 \text{ Vs}$$

$$\lambda = \sqrt{\lambda_d^2 + \lambda_q^2} = \sqrt{(\lambda_m + LdI_N \cos \alpha_i)^2 + (LqI_N \sin \alpha_i)^2}$$

$$\rightarrow \cos \alpha_i = -0.455 \rightarrow \alpha_i = 117^\circ$$

$$\begin{cases} \bar{I}_d = -9.08 \text{ A} \\ \bar{I}_q = 17.82 \text{ A} \end{cases}$$

$$\begin{cases} \lambda_d = 0.256 \text{ Vs} \\ \lambda_q = 0.378 \text{ Vs} \end{cases} \rightarrow \tilde{T}_{em} = \frac{3}{2} P (\lambda_d \bar{I}_q + \lambda_q \bar{I}_d) = 24 \text{ Nm}$$

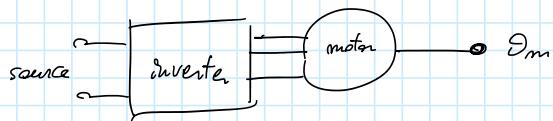
27/11 mtpa...

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We want to remove the position sensor



RECOGNIZE THE INITIAL POSITION

① SPM MOTOR:

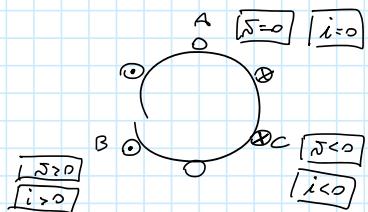
$$L = L_d = L_q \quad \text{isotropic motor}$$

$$\tau_{em} = \frac{3}{2} p N_m I_q$$

• Assumption: $T_L = 0$

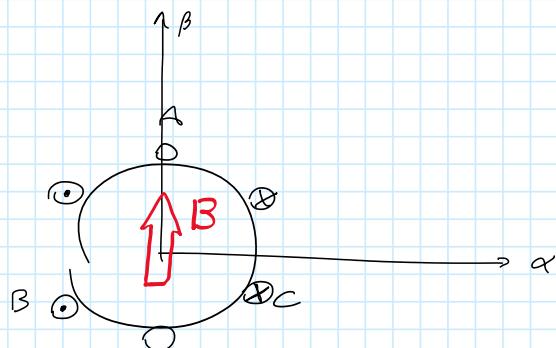
→ Voltages are imposed in the stator winding to have

$$\begin{cases} \bar{v}_\alpha = 0 \\ \bar{v}_\beta = v \end{cases}$$



$$\bar{v} = R_i + L \frac{di}{dt}, \text{ after transient}$$

$$\begin{cases} i_\alpha = 0 \\ i_\beta = \frac{v}{R} \end{cases}$$

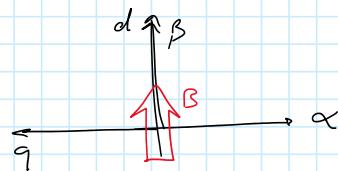


$$\text{if } \tilde{T}_L = 0, \tilde{\tau}_{em} = 0 \rightarrow \boxed{I_q = 0}$$

$$\begin{cases} I_q = \bar{i}_\alpha = 0 \\ I_d = \bar{i}_\beta \end{cases}$$

The dq axis is aligned with the αβ axis

$$\rightarrow \boxed{\theta_m^e = 90^\circ}$$



② IPM MOTOR:

$$L_d \neq L_q$$

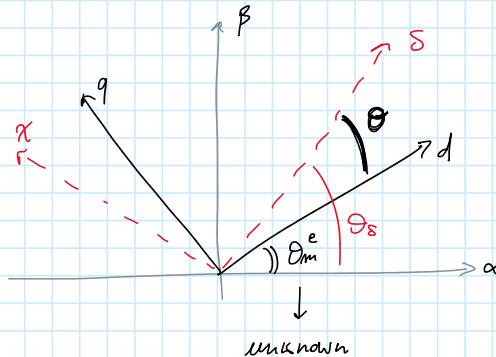
standstill motor: $\omega_m = 0$

motor equations:

$$\begin{cases} \dot{\bar{d}} = R_{cd} + L_d \frac{di_d}{dt} \\ \dot{\bar{q}} = " \end{cases}$$

$$\begin{cases} w_{dq} = 0 \\ \omega_{ld} = 0 \end{cases}$$

$$\bar{\Sigma}_{dq} = R \bar{i}_{dq} + [L_{dq}] d \frac{\bar{i}_{dq}}{dt}$$

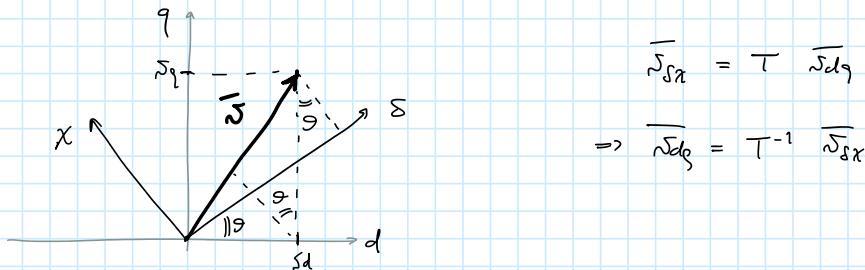


when $\theta = 0$ the δx is aligned with the dq reference frame
 $\theta = \theta_s - \theta_m$

- We inject a signal on the δx frame to recognize the position of dq , that gives θ_m

$$\bar{\Sigma}_{\delta x} = \bar{\Sigma}_{dq} \cdot e^{-j\theta} = (\bar{\Sigma}_d + j\bar{\Sigma}_q)(\cos\theta - j\sin\theta) = (\bar{\Sigma}_d \cos\theta + \bar{\Sigma}_q \sin\theta) + j(\bar{\Sigma}_q \cos\theta - \bar{\Sigma}_d \sin\theta)$$

$$\begin{bmatrix} \bar{\Sigma}_s \\ \bar{\Sigma}_x \end{bmatrix} = \begin{bmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{bmatrix} \cdot \begin{bmatrix} \bar{\Sigma}_d \\ \bar{\Sigma}_q \end{bmatrix} \quad \begin{bmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{bmatrix} = T$$



$$\Rightarrow T^{-1} \bar{\Sigma}_{\delta x} = R T^{-1} \bar{i}_{\delta x} + [L_{dq}] \cdot \frac{d}{dt} (T^{-1} \bar{i}_{\delta x})$$

$$\text{The motor is still} \rightarrow \theta = \text{constant} \Rightarrow \frac{d}{dt} T^{-1} = 0$$

$$\bar{\Sigma}_{\delta x} = R \bar{i}_{\delta x} + T [L_{dq}] T^{-1} \frac{d}{dt} \bar{i}_{\delta x}$$

$$[L_{\delta x}] = T [L_{dq}] T^{-1}$$

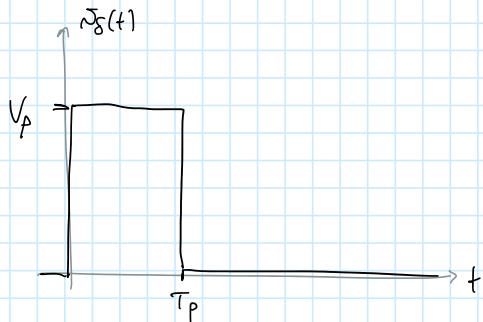
$$= \begin{bmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{bmatrix} \begin{bmatrix} L_d & 0 \\ 0 & L_q \end{bmatrix} \begin{bmatrix} \cos\theta & \sin\theta \\ \sin\theta & \cos\theta \end{bmatrix} = \begin{bmatrix} L_d \cos^2\theta + L_q \sin^2\theta \\ \frac{L_q - L_d}{2} \sin 2\theta \\ L_d \sin^2\theta + L_q \cos^2\theta \end{bmatrix}$$

numerical integration

• $[L_{\delta x}] = [L_{\delta x}(\theta)]$ is a function of the position θ

between δx axis

→ Apply a voltage pulse along the δ -axis:



$$T_p \text{ short } \rightarrow |R \bar{i}_{\delta x}| < \left| [L_{\delta x}] \frac{d \bar{i}_{\delta x}}{dt} \right|$$

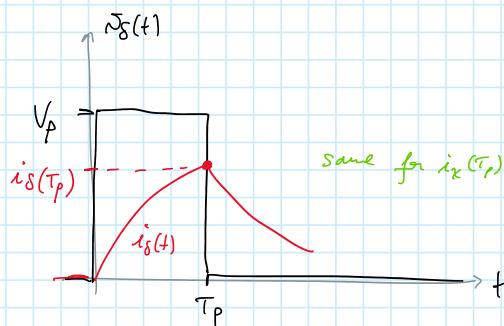
| the drop can be approximated to be
only inductive \sim no change in the current |

• EXTRACTION OF THE CURRENT

$$\begin{aligned} \frac{d \bar{i}_{\delta x}}{dt} &= [L_{\delta x}]^{-1} \cdot \bar{i}_{\delta x} \\ \bar{i}_{\delta x} &= [L_{\delta x}]^{-1} \cdot \int_0^t \bar{i}_{\delta x} dt \\ \bar{i}_{\delta x}(T_p) &= [L_{\delta x}]^{-1} \cdot \int_0^{T_p} \bar{i}_{\delta x} dt = [L_{\delta x}]^{-1} \cdot \begin{bmatrix} V_p \\ 0 \end{bmatrix} \cdot T_p \\ &= \begin{bmatrix} \frac{\cos^2 \theta}{L_d} + \frac{\sin^2 \theta}{L_q} & \frac{\sin 2\theta}{2} \frac{L_d - L_q}{L_d L_q} \\ \frac{\sin 2\theta}{2} \frac{L_d - L_q}{L_d L_q} & \frac{\cos^2 \theta}{L_q} + \frac{\sin^2 \theta}{L_d} \end{bmatrix} \cdot \begin{bmatrix} V_p \\ 0 \end{bmatrix} \cdot T_p \end{aligned}$$

$$\bar{i}_{\delta} = \begin{cases} V_d & \text{during } T_p \\ 0 & \text{elsewhere} \end{cases}, \quad \bar{i}_x = 0$$

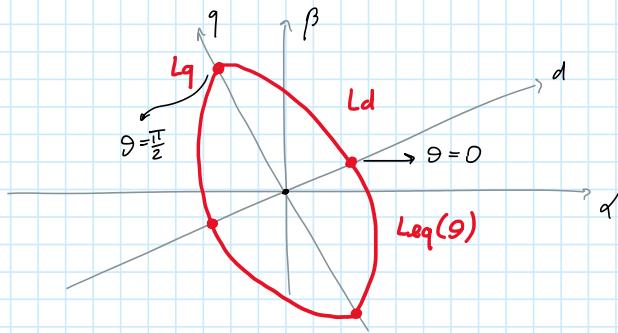
$$\begin{cases} i_{\delta}(T_p) = \left(\frac{\cos^2 \theta}{L_d} + \frac{\sin^2 \theta}{L_q} \right) V_p T_p \\ i_x(T_p) = \left(\frac{\sin 2\theta}{2} \frac{L_d - L_q}{L_d L_q} \right) V_p T_p \end{cases}$$



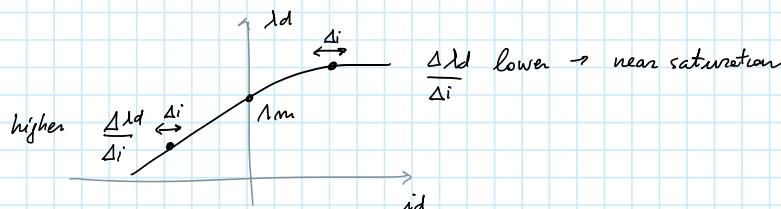
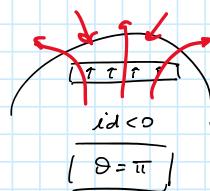
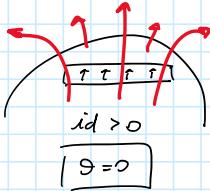
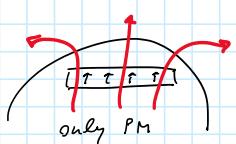
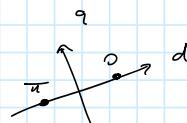
$$\left| \bar{i}_{\delta x}(T_p) \right| = \sqrt{i_{\delta}(T_p)^2 + i_x(T_p)^2} = V_p T_p \cdot \sqrt{\frac{\cos^2 \theta}{L_d^2} + \frac{\sin^2 \theta}{L_q^2}}$$

$$\boxed{\left| \bar{i}_{\delta x}(T_p) \right| = \frac{V_p T_p}{L_{eq}(\theta)}}$$

$$\boxed{L_{eq}(\theta) = \frac{L_d L_q}{\sqrt{L_d^2 \sin^2 \theta + L_q^2 \cos^2 \theta}}}$$



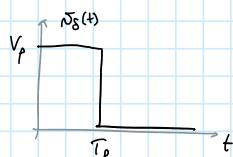
How to recognize $\theta = 0$ or $\theta = \pi$



EXERCISE

$$L_d = 100 \text{ mH}$$

$$L_q = 400 \text{ mH}$$



$$V_p = 60 \text{ V}$$

$$\theta_d = 30^\circ$$

$$T_p = 10 \text{ ms}$$

- $i_s, i_x (\tau_p)$?

- $|i_{sx}|$?

- $L_{eq}(\theta)$?

$$i_s(\tau_p) = \left(\frac{\cos^2 30^\circ}{0.100} + \frac{\sin^2 30^\circ}{0.400} \right) 10 \cdot 0.010 = 0.8125 \text{ A}$$

$$i_x(\tau_p) = \left(\frac{\sin 2 \cdot 30^\circ}{2} \cdot \frac{0.100 - 0.050}{0.1 \cdot 0.4} \right) 10 \cdot 0.010 = -0.325 \text{ A}$$

$$|i_{sx}(\tau_p)| = \sqrt{0.8125^2 + 0.325^2} = 0.875 \text{ A}$$

$$|\bar{x}_{\delta x}(\tau_p)| = \sqrt{0.8125^2 + 0.325^2} = 0.875 A$$

$$\zeta_{ep}(30^\circ) = \frac{V_p \tau_p}{|i_{\delta x}|} = \frac{10 \cdot 0.010}{0.875} = 114 \text{ rad/s}$$

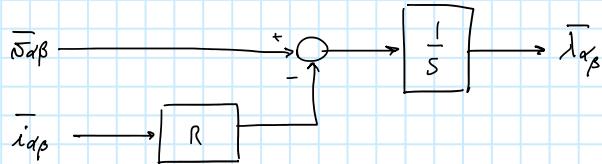
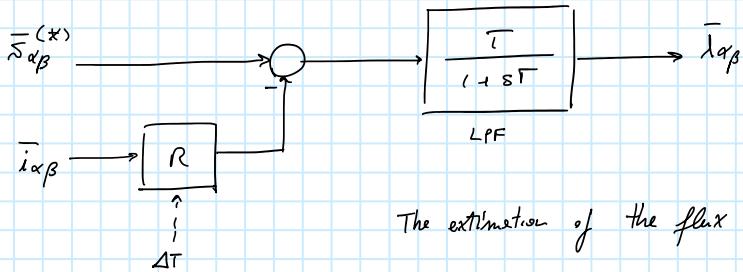
NO POSITION SENSOR

Recognize the position of the rotor from the flux linkage:

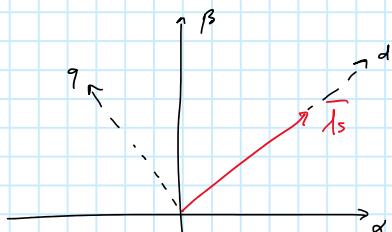
$$\bar{\lambda}_{\alpha\beta} = \int (\bar{\lambda}_{\alpha\beta} - R \bar{i}_{\alpha\beta}) dt$$

Performance of the computation:

- accuracy from hall sensor
- switching ripple / noise → usually reconstructed from Vdc + state inverter
- drift of the integral due to dc offset
- don't know the initial condition $\lambda(0) = \lambda_{m(0)}$, $i(0) = 0$
- parameter accuracy: $R_{120} = R_{20} (1 + \alpha \Delta T)$, $\alpha \Delta T = 40\%$

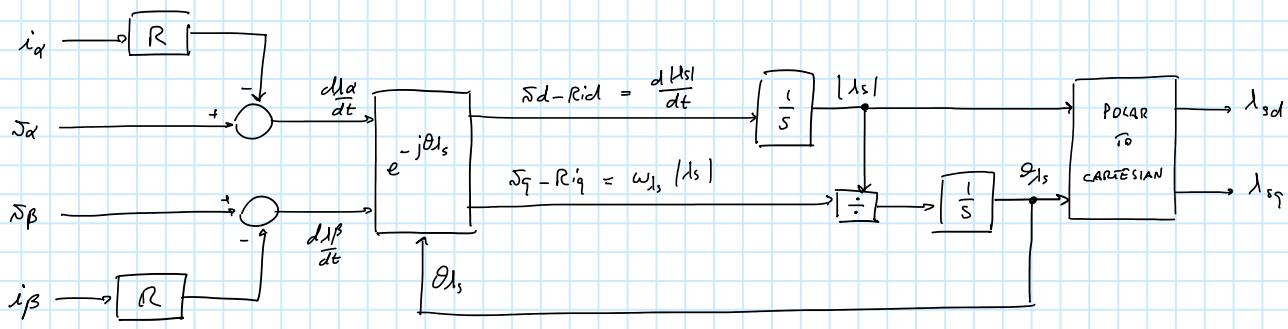
PRACTICAL IMPLEMENTATION

The estimation of the flux is accurate for $\omega > \frac{1}{T}$

STATOR LINKAGE

$$\bar{\lambda}_s = \begin{cases} \lambda_{sd} = |\lambda_s| \\ \lambda_{sq} = 0 \end{cases}$$

$$\bar{\lambda}_{dq} = R \bar{i}_{dq} + \frac{d \bar{\lambda}_{dq}}{dt} + j\omega \bar{\lambda}_{dq} \rightarrow \begin{cases} \bar{\lambda}_d = R \bar{i}_d + \frac{d \bar{\lambda}_d}{dt} \\ \bar{\lambda}_q = R \bar{i}_q + \omega \bar{\lambda}_d \end{cases}$$

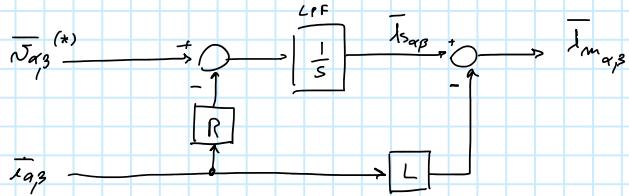


SPM motor has permanent magnet

$$\bar{\lambda}_s = \bar{\lambda}_m + L\bar{i}$$

Recognized by voltage and current integration

- $\bar{\lambda}_m = \lambda_m e^{j\vartheta_m^e} = \bar{\lambda}_s - L\bar{i}$



EXERCISE

steady state (sinusoidal)

$$\lambda_m = ?$$

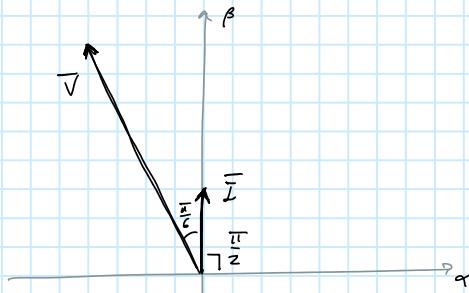
$$R = 3.05 \Omega$$

$$L = 17.6 \text{ mH}$$

$$\begin{aligned} \bar{V} &= 100 e^{j(\omega t + \frac{\pi}{6})} \\ \bar{i} &= 12 e^{j(\omega t + \frac{\pi}{2})} \end{aligned} \quad \left. \begin{array}{l} \text{ref.} \\ \alpha\beta \end{array} \right.$$

$$\omega = 100 \text{ rad/s}$$

$$\begin{aligned} \bar{V} &= 100 e^{j\frac{2\pi}{3}} = 100 e^{j\left(\frac{\pi}{2} + \frac{\pi}{6}\right)} \\ \bar{i} &= 12 e^{j\frac{\pi}{2}} \end{aligned}$$



$$j\omega\bar{\lambda}_s = \bar{V} - R\bar{i} = 100 e^{j\left(\frac{\pi}{2} + \frac{\pi}{6}\right)} - 36.6 e^{j\frac{\pi}{2}} = -50 + j50$$

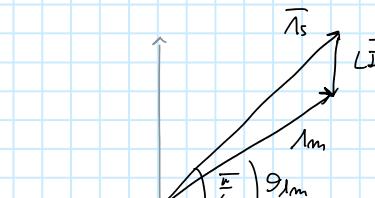
$$\bar{\lambda}_s = \frac{\bar{V} - R\bar{i}}{j\omega} = \frac{-50 + j50}{j100} = 0.5 + j0.5$$

$$\bar{\lambda}_m = \bar{\lambda}_s - L\bar{i} = 0.5 + j0.289$$

$$\vartheta_{\lambda_m} = \arctan \frac{0.289}{0.5} = \frac{\pi}{6}$$

$$|\lambda_m| = \sqrt{0.5^2 + 0.289^2} = 0.578 \sqrt{5}$$

$$\bar{\lambda}_m = (|\lambda_m| e^{j\vartheta_{\lambda_m}}) e^{j\omega t + \frac{\pi}{6}} = 0.578 \cdot e^{j\left(\omega t + \frac{\pi}{6}\right)}$$



$$\vartheta_{\lambda_m} = \vartheta_m^e$$

$$\overline{\lambda_m} = (\lambda_m / e)^j \theta_m = 0.571 \cdot e^{j(\omega t + \frac{\pi}{6})}$$

EXTRACTION OF THE SPEED FROM θ_s

$$\omega_m^e = \frac{d\theta_s}{dt}$$

$$\theta_s = \text{atan} \left(\frac{\lambda_{s\beta}}{\lambda_{s\alpha}} \right)$$

$$\overline{\lambda_{s\alpha\beta}} = \lambda_{s\alpha} + j \lambda_{s\beta}$$

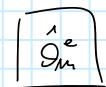
$$\frac{d}{dx} \text{atan}(x) = \frac{1}{1+x^2}$$

$$\frac{d}{dt} \text{atan}(f(t)) = \frac{1}{(1+f(t))^2} \cdot \frac{df(t)}{dt}$$

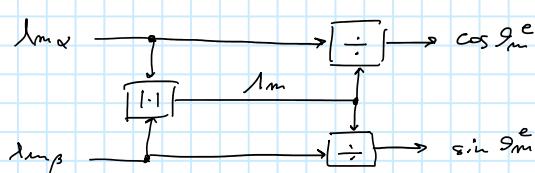
$$\omega_m^e = \frac{1}{1 + \left(\frac{\lambda_{s\beta}}{\lambda_{s\alpha}} \right)^2} \cdot \frac{d}{dt} \left(\frac{\lambda_{s\beta}}{\lambda_{s\alpha}} \right) = \frac{\cancel{\lambda_{s\alpha}^2}}{\lambda_{s\alpha}^2 + \lambda_{s\beta}^2} \cdot \frac{\lambda_{s\beta}' \lambda_{s\alpha} - \lambda_{s\alpha}' \lambda_{s\beta}}{\cancel{\lambda_{s\alpha}^2}} = \frac{\lambda_{s\alpha} \lambda_{s\beta}' - \lambda_{s\beta} \lambda_{s\alpha}'}{\lambda_{s\alpha}^2 + \lambda_{s\beta}^2}$$

- in digital : $\frac{d}{dt} \lambda_{s\alpha} = \frac{\lambda_{s\alpha}[k] - \lambda_{s\alpha}[k-1]}{Ts}$

$$\Rightarrow \omega_m^e = \frac{\lambda_{s\alpha}[k-1] \lambda_{s\beta}[k] - \lambda_{s\alpha}[k] \lambda_{s\beta}[k-1]}{Ts (\lambda_{s\alpha}^2[k] + \lambda_{s\beta}^2[k])}$$



MATH APPROACH :

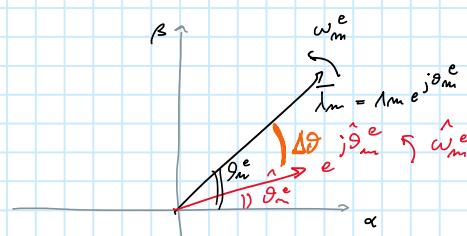


$$\theta_m^e = \text{atan} \left(\frac{\sin \theta_m^e}{\cos \theta_m^e} \right)$$

$$\omega_m^e = \frac{d\theta_m^e}{dt}$$

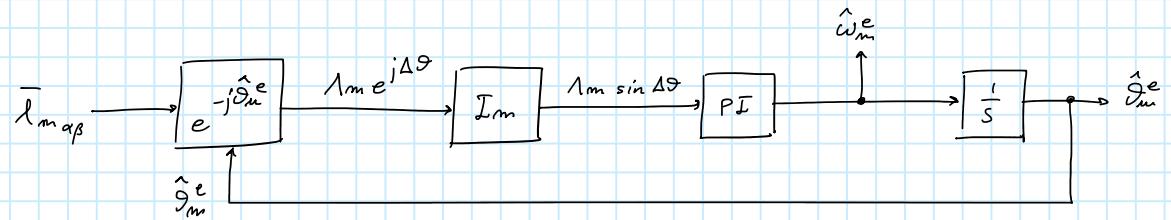
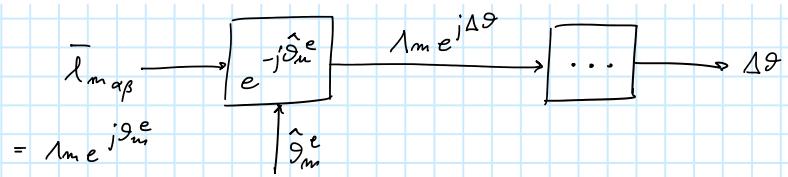
Implementation approach : PLL

PHASE-LOCKED-LOOP



If $\Delta\theta > 0$, $\hat{\omega}_m^e$ must increase

If $\Delta\theta < 0$, $\hat{\omega}_m^e$ must decrease



Linearization: $\sin \Delta\vartheta \approx \Delta\vartheta$

SYSTEM DESCRIPTION IN STATE-SPACE

$$\dot{x} = Ax + Bu$$

x: state

u: input

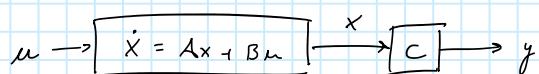
$$y = Cx$$

A: system matrix

y: output

B: input matrix

C: output matrix

EXAMPLE : DC motor

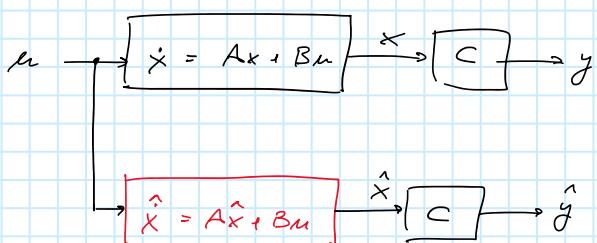
$$\begin{cases} \dot{\eta} = R_i + \frac{Ldi}{dt} + k\bar{\phi}w_m \\ T_{em} = k\bar{\phi}i = b\omega_m + J \frac{d\omega_m}{dt} \end{cases} \rightarrow \begin{cases} \frac{d\eta}{dt} = \frac{1}{L} (\dot{\eta} - R_i - k\bar{\phi}w_m) \\ \frac{d\omega_m}{dt} = \frac{1}{J} (k\bar{\phi}i - b\omega_m) \end{cases}$$

$$\begin{bmatrix} \frac{d\eta}{dt} \\ \frac{d\omega_m}{dt} \end{bmatrix} = \begin{bmatrix} -\frac{R}{L} & -\frac{k\bar{\phi}}{L} \\ \frac{k\bar{\phi}}{J} & -\frac{b}{J} \end{bmatrix} \cdot \begin{bmatrix} \dot{\eta} \\ i \end{bmatrix} + \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} \eta$$

\downarrow

$$\dot{x} = A \cdot x + B \cdot u$$

$$\begin{array}{c} w_m = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} i \\ \omega_m \end{bmatrix} \\ \downarrow \\ y = C \cdot x \end{array}$$

OBSERVER

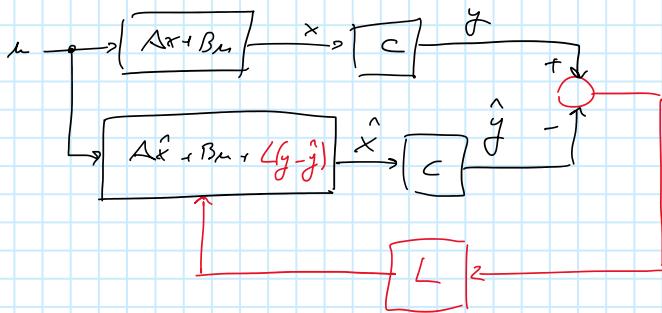
\hat{x} : estimated value
of state variables

open loop observer:
→ no check of the estimation accuracy

$$x_{\text{err}} = x - \hat{x}$$

$$\dot{x}_{\text{err}} = \dot{x} - \dot{\hat{x}} = [Ax + Bu] - [A\hat{x} + \cancel{Bu}] = Ax_{\text{err}}$$

CLOSED LOOP OBSERVER



$$\dot{\hat{x}} = A\hat{x} + Bu + L(y - \hat{y})$$

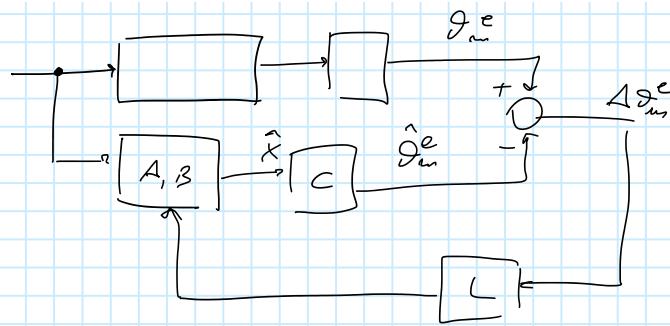
$$\begin{aligned}\dot{x}_{\text{err}} &= \dot{x} - \dot{\hat{x}} \\ &= A(x - \hat{x}) - L(Cx - C\hat{x}) \\ &= (A - LC)x_{\text{err}}\end{aligned}$$

$L \rightarrow$ have fast and stable convergence $\hat{x} \rightarrow x$

$$L = \begin{bmatrix} k_1 \\ k_2 \\ \vdots \\ k_n \end{bmatrix}$$

$$\det \left[sI - (A - LC) \right] = 0 \quad \Rightarrow \quad \begin{bmatrix} s - \beta_1 & & & \\ & s - \beta_2 & & \\ & & \ddots & \\ & & & s - \beta_n \end{bmatrix} = 0$$

ROTOR POSITION OBSERVER OF 2nd ORDER



$$x = \begin{bmatrix} \dot{\theta}_m^e \\ \hat{\omega}_m^e \end{bmatrix} \quad \left\{ \begin{array}{l} \dot{\theta}_m^e = \omega_m^e \\ \dot{\omega}_m^e = p \frac{\tilde{T}_{cm} - T_c}{J} = 0 \end{array} \right. \quad \begin{array}{l} \text{imposed} \rightarrow \text{constant speed, no acceleration} \end{array}$$

$$\begin{bmatrix} \dot{\hat{\theta}}_m^e \\ \dot{\hat{\omega}}_m^e \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{\theta}_m^e \\ \hat{\omega}_m^e \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \end{bmatrix} u + \begin{bmatrix} k_1 \\ k_2 \end{bmatrix} (\dot{\theta}_m^e - \hat{\dot{\theta}}_m^e)$$

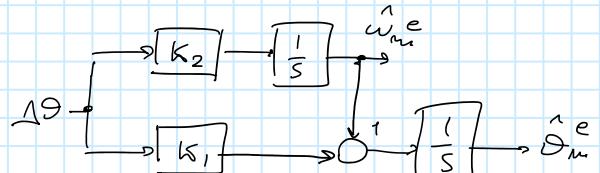
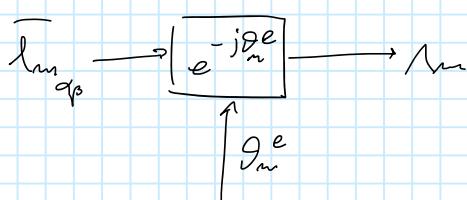
A X B L

$$y = \hat{\theta}_m^e = [1 \ 0] \begin{bmatrix} \hat{\theta}_m^e \\ \hat{\omega}_m^e \end{bmatrix}$$

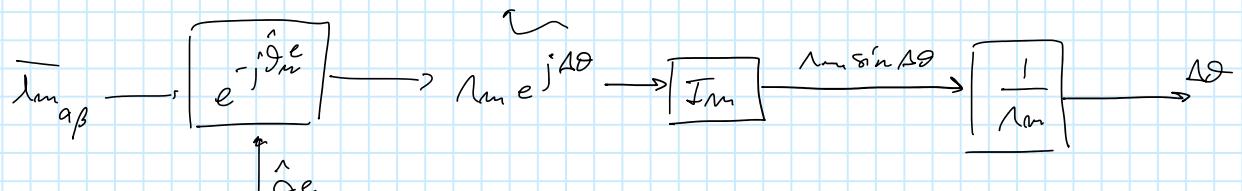
C X

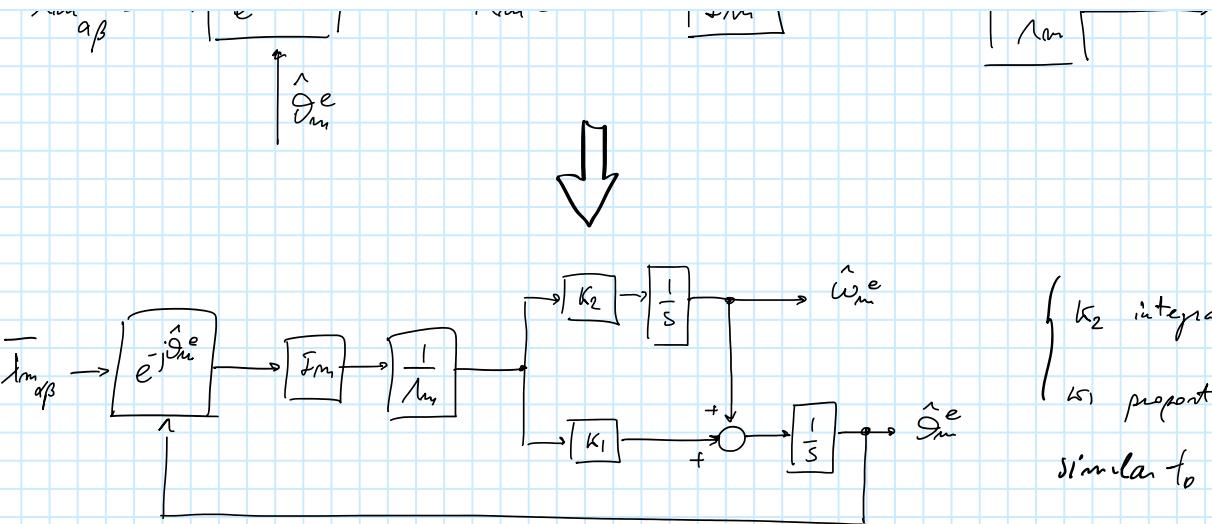
$$\left\{ \begin{array}{l} \dot{\hat{\theta}}_m^e = \hat{\omega}_m^e + k_1 (\dot{\theta}_m^e - \hat{\dot{\theta}}_m^e) \\ \dot{\hat{\omega}}_m^e = k_2 (\dot{\theta}_m^e - \hat{\dot{\theta}}_m^e) \end{array} \right.$$

$$\bar{I}_{m\alpha\beta} = \lambda_m e^{j\dot{\theta}_m^e}$$



$$\Delta \theta = \dot{\theta}_m^e - \hat{\dot{\theta}}_m^e$$





$\left\{ \begin{array}{l} K_2 \text{ integral gain} \\ K_1 \text{ proportional gain} \\ \text{similar to PLL} \end{array} \right.$

DETERMINE K_1, K_2

$$\dot{x}_{en} = Ax_{en} - LCx_{en} = (A - LC)x_{en}$$

$$\det [sI - (A - LC)] = 0$$

$$\det \left[\begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \left(\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} K_1 \\ K_2 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} \right) \right] = 0$$

$$\det \left[\begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} + \begin{bmatrix} K_1 & -1 \\ K_2 & 0 \end{bmatrix} \right] = 0$$

$$\det \begin{bmatrix} s + K_1 & -1 \\ K_2 & s \end{bmatrix} = 0$$

$$\Rightarrow s(s + K_1) + K_2 = 0 \quad \Rightarrow \boxed{s^2 + K_1 s + K_2 = 0}$$

FIXED, I WANT A DOUBLE POLE

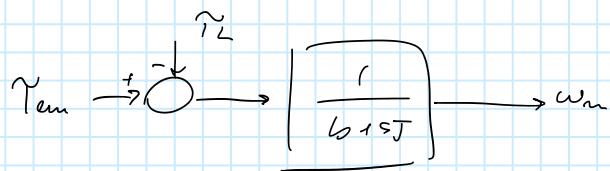
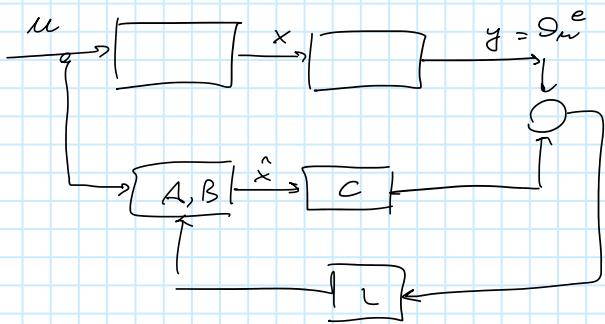
$$f = 10 \text{ Hz}, \omega = 60 \frac{\text{rad}}{\text{s}}$$

$$\beta_1 = \beta_2 = -60 \quad \rightarrow$$

$$(s + 60)^2 = \boxed{s^2 + 120s + 3600}$$

$$\boxed{\begin{array}{l} K_1 = 120 \\ K_2 = 3600 \end{array}}$$

3RD ORDER OBSERVER



$$\dot{\tilde{T}_e} = \frac{3}{2} p \lambda_m \hat{\lambda}_q$$

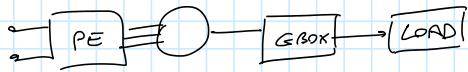
$$x = \begin{bmatrix} \dot{\theta}_m \\ \dot{w}_m \\ \tilde{T}_e \end{bmatrix}$$

$$\left\{ \begin{array}{l} \dot{\theta}_m = w_m \\ \dot{w}_m = \frac{\tilde{T}_e - T_L}{J} - \frac{b}{J} w_m \\ \dot{\tilde{T}}_e = 0 \end{array} \right. \rightarrow \begin{bmatrix} \dot{\theta}_m \\ \dot{w}_m \\ \dot{\tilde{T}}_e \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & -\frac{b}{J} & -\frac{1}{J} \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta_m \\ w_m \\ \tilde{T}_e \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{J} \\ 0 \end{bmatrix} \tilde{T}_e$$

$$\begin{bmatrix} \dot{\hat{\theta}}_m \\ \dot{\hat{w}}_m \\ \dot{\tilde{T}}_e \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & -\frac{b}{J} & -\frac{1}{J} \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{\theta}_m \\ \hat{w}_m \\ \hat{T}_e \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{J} \\ 0 \end{bmatrix} \tilde{T}_e + \begin{bmatrix} K_1 \\ K_2 \\ K_3 \end{bmatrix} (\theta_m^e - \hat{\theta}_m^e)$$

Gearbox

sabato 19 dicembre 2020 17:02



$$T_{em} = k_f \cdot I_a \quad (\Delta c)$$

$$P_J = R_a I_a^2$$

$$P_{em} = \frac{3}{2} \rho \lambda_m I \quad (\text{SPM})$$

$$P_J = \frac{3}{2} R_s I^2$$



$$K_{gear} < 1$$

$$\left\{ \begin{array}{l} Tr = \frac{Tm}{K_{gear}} \\ Wr = \omega_m \cdot K_{gear} \end{array} \right.$$

$$Wr = \omega_m \cdot K_{gear}$$

$$\eta_{gear} : \left\{ \begin{array}{l} Wr = K_{gear} \omega_m \\ Tr = \eta_{gear} \cdot \frac{Tm}{K_{gear}} \end{array} \right.$$

gli ingranaggi finiscono alla stessa velocità
la coppia è ridotta a causa delle perdite

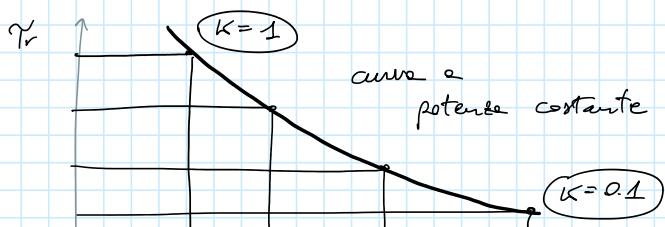
CARICO GENERICO (DINAMICO)

$$T_{carico} = Tr + Jr \frac{dw_r}{dt} \quad \longrightarrow \quad T_{motore} = Tm - Jm \frac{d\omega}{dt}$$

$$(Tm - Jm \frac{d\omega_m}{dt}) \omega_m = (Tr + Jr \frac{dw_r}{dt}) wr$$

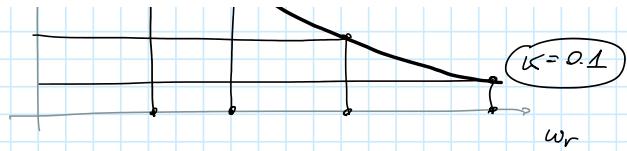
$$\boxed{Tm - Tr \cdot K_{gear} = \left(\frac{Jm}{K_{gear}} + K_{gear} Jr \right) \frac{dwr}{dt}}$$

CARICO STATICO



$$Tr wr = Tm \omega_m$$

per il motore la relazione



per il motore la relazione
e' inversa, per K crescenti w_m diminuisce

$$\eta \cdot P_m > P_r$$

$$w_r = w_m \frac{K_{gear} \min}{J_r}$$

$$\tilde{T}_m = \frac{T_m}{K_{gear} \max}$$

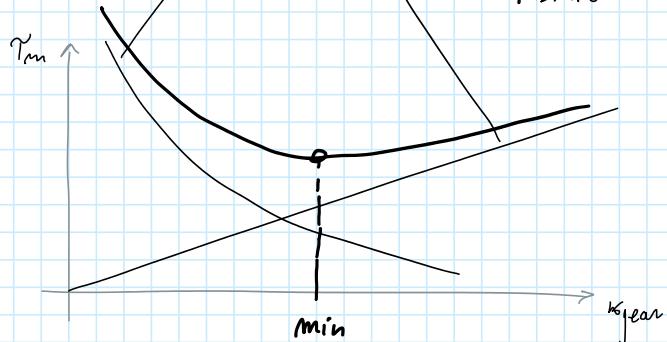
CARICO DINAMICO PURAMENTE INERZIALE

$$\tilde{T}_{carico} = \cancel{K} + J_r \frac{dw_r}{dt}$$

$$T_m = \left(\frac{J_m}{K_{gear}} + K_{gear} J_r \right) \frac{dw_r}{dt}$$

$$w_m = \frac{w_r}{K_{gear}}$$

FISIARO



$$\frac{d}{dK} \left(\frac{J_m}{K} + K J_r \right) = 0$$

$$K_{opt} = \sqrt{\frac{J_m}{J_r}}$$

$$T_{m,min} = T_m(K_{opt})$$

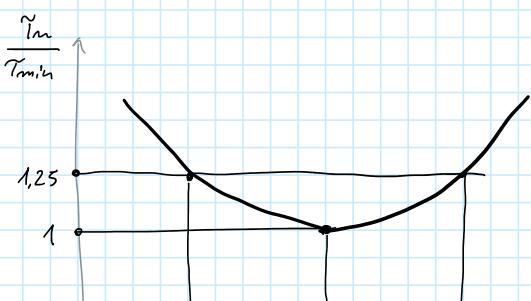
$$\frac{J_m}{K_{opt}} = J_r K_{opt} \rightarrow J_m = K_{opt}^2 J_r$$

$$T_{m,min} = 2 \sqrt{J_m J_r} \frac{dw_r}{dt}$$

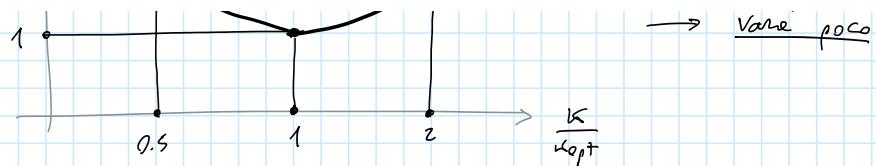
Fuerza del carico
repartida al motor
 $= J_r'$

RIDUZIONE NON OTTIMO:

$$\frac{\tilde{T}_m}{\tilde{T}_{m,min}} = \frac{1}{2} \left(\frac{K_{opt}}{K} + \frac{K}{K_{opt}} \right)$$



cambia del 25% "sbagliato"
del doppio il rapporto di K
 \rightarrow varie poco



COPPIA DI PICCO

$$K = \frac{\tilde{t}_{\text{pic}} \pm \sqrt{\tilde{t}_{\text{pic}}^2 - 4 J_m J_r \omega_{\text{max}}^2}}{2 \omega_{\text{max}}} = \begin{cases} K_{\min} \\ K_{\max} \end{cases}$$

COPPIA NOMINALE

$$K = \frac{\tilde{t}_N \pm \sqrt{\tilde{t}_N^2 - 4 J_m J_r \omega_{\text{rms}}^2}}{2 \omega_{\text{rms}}} = \begin{cases} K_{\min} \\ K_{\max} \end{cases}$$

$$\boxed{K_{\min} = \frac{\omega_{\text{max}}}{\omega_N}}$$

Gearbox es

domenica 20 dicembre 2020 11:52

SPM MOTORI

$$R \approx 0$$

$$I_N = 10 \text{ A}$$

$$\tau_r = 10^2 \text{ Nm}$$

$$L = 0.01 \text{ H}$$

$$I_p = 20 \text{ A}$$

$$\omega_r = 45 \text{ rad/s}$$

$$A_m = 0.2 \text{ Vs}$$

$$V_N = 190 \text{ V}$$

$$z_p = 8$$

$$\kappa_{year} ? \quad \eta_{year} = 90\%$$

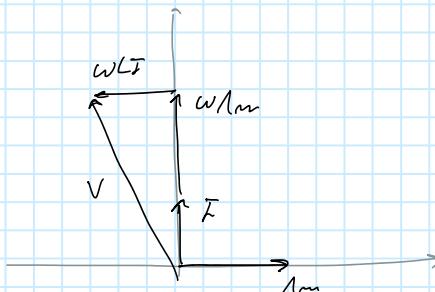
$$\tilde{\tau}_{em} = \frac{3}{2} p A_m I_q \rightarrow \tilde{\tau}_N = 12 \text{ Nm}, \quad \tilde{\tau}_p = 24 \text{ Nm}$$

$$V_N^2 = (\omega/m)^2 + (WL)^2$$

$$\omega = \frac{V_N}{\sqrt{A_m^2 + (LF)^2}}$$

- $I_N = 10, \omega_N = 45 \text{ rad/s} \rightarrow \omega_{m_N} = 112 \text{ rad/s}$

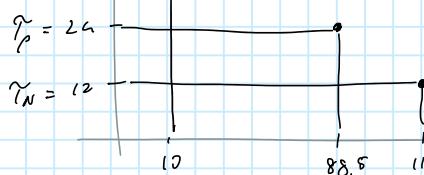
- $I_p = 20, \omega_p = 354 \text{ rad/s} \rightarrow \omega_{m_p} = 88.5 \text{ rad/s}$



$$\frac{\gamma_r}{\eta} = \frac{100}{0.9} = 111$$

$$P_N = 112 \cdot 12 = 1344 \text{ W}$$

$$P_r = 111 \cdot 10 = 1110 \text{ W}$$

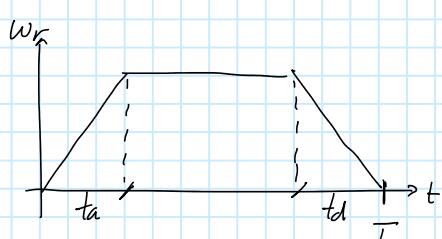


$$\frac{\omega_r}{\omega_N} = \frac{10}{112} = 0.0893$$

$$\frac{\tilde{\tau}_N}{\tilde{\tau}_r} = \frac{12}{111} = 0.108$$

$$0.0893 \leq \kappa_{year} \leq 0.108$$

CALCO PURAMENTE INERZIALE



$$\tau_r = 8 \cdot 10^{-3} \kappa_{year}^2$$

$$\dot{\omega}_a = 500 \text{ rad/s}^2$$

$$\dot{\omega}_d = -600 \text{ rad/s}^2 \quad T = 2 \text{ s}$$

$$t_a = 400 \text{ ms}$$

	1	2	3
\tilde{T}_P	5	2.5	1
\tilde{T}_N	2	1.3	0.5
ω_N	1000	1000	1600
J_m	$2e^{-3}$	$0.5e^{-3}$	$0.08e^{-3}$

Quale motore c' è il più adeguato?

$$\omega_{\max} = \omega_{\text{ATA}} = 200 \text{ rad/s}$$

$$t_0 = 0.5 \text{ s}$$

$$\dot{\omega}_{\max} = \dot{\omega}_{\text{ATA}}$$

$$\dot{\omega}_{\text{rms}} = \sqrt{\frac{1}{2}(500^2 \cdot 0.4 + 400^2 \cdot 0.5)} = \\ = 800 \text{ rad/s}^2$$

$$\tilde{T}_A = \dot{\omega}_A \tilde{J}_r = 0.008 \cdot 800 = 4 \text{ Nm}$$

$$\tilde{T}_D = \tilde{J}_r \dot{\omega}_D = 3.2 \text{ Nm}$$

$$\tilde{T}_{\max} = \tilde{T}_A, \quad \tilde{T}_{\text{rms}} = \sqrt{\frac{1}{2} (4^2 \cdot 0.4 + 3.2^2 \cdot 0.5)} = 2.4 \text{ Nm} = \tilde{J}_r \dot{\omega}_{\text{rms}}$$

$$\textcircled{1} \quad K_{opt} = \sqrt{\frac{J_m}{J_r}} = \sqrt{\frac{2}{8}} = 0.5$$

$$\tilde{T}_{\min} \geq 2 \sqrt{\tilde{J}_r J_m} \dot{\omega}_{\max} = 4 \text{ Nm} \quad \checkmark$$

$$\tilde{T}_N \geq 2 \sqrt{\tilde{J}_r J_m} \dot{\omega}_{\text{rms}} = 2.4 \text{ Nm} \quad \times$$

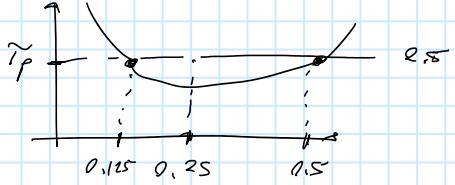
$$\textcircled{2} \quad K_{opt} = \sqrt{\frac{2.5}{8}} = 0.25$$

$$\tilde{T}_{\min} \geq 2 \sqrt{\tilde{J}_r J_m} \dot{\omega}_{\max} = 2 \text{ Nm} \quad \checkmark$$

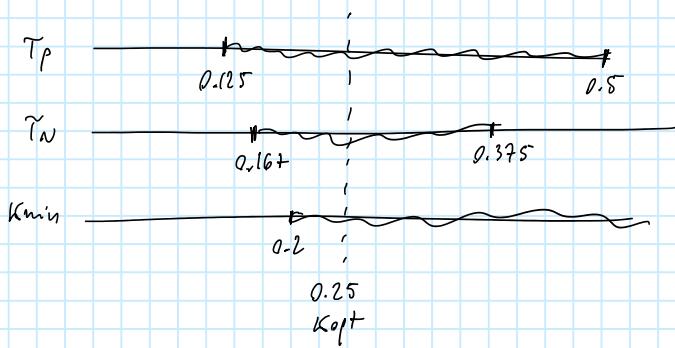
$$\tilde{T}_N \geq 2 \sqrt{\tilde{J}_r J_m} \dot{\omega}_{\text{rms}} = 1.2 \text{ Nm} \quad \checkmark$$

$$\bullet \quad \tilde{T}_P : \quad K_{1,2} = \frac{1}{2} \frac{2.5 \pm \sqrt{2.5^2 - 4 \cdot 500^2 \cdot 8 \cdot 0.5 \cdot 10^{-6}}}{500 \cdot 0.008} = \begin{cases} 0.125 \\ 0.5 \end{cases}$$

$$\bullet \quad \tilde{T}_N : \quad K_{1,2} = \frac{1}{2} \frac{1.3 \pm \sqrt{1.3^2 - 4 \cdot 700^2 \cdot 8 \cdot 0.5 \cdot 10^{-6}}}{300 \cdot 0.008} = \begin{cases} 0.167 \\ 0.375 \end{cases}$$



$$\bullet \quad \omega_N = 1000 \text{ rad/s} \quad \rightarrow \quad K_{\min} = \frac{\omega_{\max}}{\omega_N} = \frac{200}{1000} = 0.2$$



$$0.2 \leq K_{\text{year}} \leq 0.375 \quad \text{per il motore ②}$$

$$\boxed{0.2 \leq K_{gear} \leq 0.375} \quad \text{per il motore ②}$$

$$③ K_{opt} = 0.1$$

$$T_p = 9.8 \text{ Nm} \quad \checkmark$$

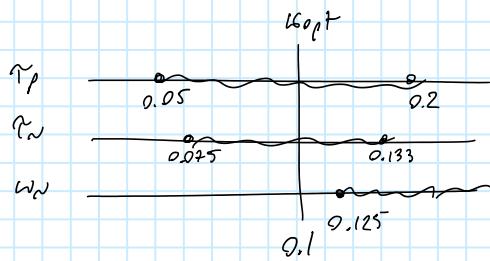
$$T_N = 0.48 \text{ Nm} \quad \checkmark$$

$\omega_N = 2000 \quad X$, però OK fuori dal K_{opt}

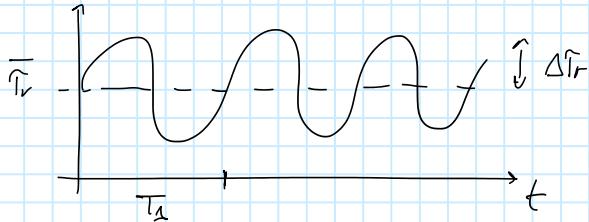
$$T_p : K_{1,2} \leq \frac{0.05}{0.2}$$

$$T_N : K_{1,2} = \frac{0.095}{0.133}$$

$$\omega_N \text{ e } K_{min} = \frac{\omega_{max}}{\omega_N} = 0.125$$



$$\boxed{0.125 \leq K_{gear} \leq 0.133} \quad \text{motore ③}$$

$\tilde{T}_r(t)$ 

$\tilde{T}_r = 50 \text{ Nm}$

$\Delta T_r = 60 \text{ Nm}$

$T_A = 2.5$

$J_r = 0.138 \text{ kg m}^2$

$K_{year} = \frac{1}{4}$

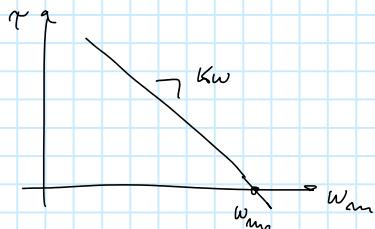
 $\Delta \omega$ motor

$K_w = 0.4$

$\omega_{mo} = 200 \text{ rad/s}$

$J_m = 7 \cdot 10^{-3} \text{ kg m}^2$

$\tilde{\omega}_m = K_w(\omega_{mo} - \omega_m)$



$\tilde{T}_r' = K_{year} \tilde{T}_r = 12.5 \text{ Nm}$ coppia imposta al motore

$\Delta \tilde{T}_r' = K_{year} \Delta \tilde{T}_r = 10 \text{ Nm}$

$J_r' = K_{year}^2 J_r = 0.0086 \text{ kg m}^2$

$J = J_m + J_r' = 13.625 \cdot 10^{-3} \text{ kg m}^2$

$f_A = 0.5 \text{ Hz}$

$\tilde{T}_r'(t) = \tilde{T}_r' + \Delta \tilde{T}_r' \sin(2\pi f_A t) = 12.5 + 10 \sin(\pi t)$

$$\begin{cases} \tilde{\omega}_m = K_w (\omega_{mo} - \omega_m) \\ \tilde{\omega}_m = \tilde{T}_r'(t) + J \dot{\omega}_m \end{cases} \rightarrow J \ddot{\omega}_m + \tilde{T}_r' + \Delta \tilde{T}_r' \sin(\pi t) = K_w (\omega_{mo} - \omega_m)$$

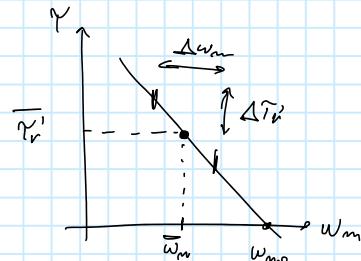
The speed is oscillating

$\omega_m = \bar{\omega}_m + \Delta \omega_m \Rightarrow \dot{\omega}_m = \Delta \dot{\omega}_m$

$$\underline{J \Delta \dot{\omega}_m + \tilde{T}_r' + \Delta \tilde{T}_r' \sin(\pi t)} = \underline{K_w (\omega_{mo} - \bar{\omega}_m)} - K_w \Delta \omega_m$$

constant part

$\Delta \dot{\omega}_m + \frac{K_w}{J} \Delta \omega_m = - \frac{\Delta \tilde{T}_r'}{J} \sin(\pi t)$



$\Rightarrow \Delta \omega_m(t) = A \sin(2\pi f_A t) + B \cos(2\pi f_A t) + \cancel{D e^{-\frac{t}{J}}} \rightarrow \text{we focus on steady state}$

$\Delta \dot{\omega}(t) = A 2\pi f_A \cos(2\pi f_A t) - B 2\pi f_A \sin(2\pi f_A t)$

$\Rightarrow [A 2\pi f_A \cos(2\pi f_A t) + B 2\pi f_A \sin(2\pi f_A t)] + \frac{K_w}{J} [A \sin(2\pi f_A t) + B \cos(2\pi f_A t)] = - \frac{\Delta \tilde{T}_r'}{J} \sin(2\pi f_A t)$

$$\left. \begin{array}{l} \bar{J}\bar{\omega}_A = -k_w B \\ -\bar{J}\bar{\omega}_A + A\bar{\tau}'_r = -k_w A \end{array} \right\} \rightarrow \Delta\omega_m = \sqrt{A^2 + B^2}$$

$$\boxed{\Delta\omega_m = \frac{\Delta\tau'_r}{\sqrt{\bar{J}^2 R_A^2 + k_w^2}}} = \underline{\underline{24,81 \text{ rad/s}}}$$

- In this case $\bar{J}\bar{\omega}_A \approx 0$

\Rightarrow the oscillation is almost only dependent on the motor characteristics

The requirement is to reduce the oscillations to 1.5% of the average speed:

$$\Delta\omega_m \leq 1.5\% \bar{\omega}_m$$

$$\bar{\tau}'_r = k_w(\omega_{mo} - \bar{\omega}_m) \rightarrow \bar{\omega}_m = \frac{-\bar{\tau}'_r}{k_w} + \omega_{mo} = 168,75 \text{ rad/s}$$

$$\Rightarrow \Delta\omega_m \leq 2.53 \text{ rad/s} \quad \underline{\text{REQUIREMENT}}$$

The only method to reduce the oscillations is to increase \bar{J} by adding a flywheel

$$\Delta\omega_m = \frac{\Delta\tau'_r}{\sqrt{\bar{J}^2 R_A^2 + k_w^2}} \rightarrow \bar{J}^2 = \frac{\left(\frac{\Delta\tau'_r}{\Delta\omega_m}\right)^2 - k_w^2}{R_A^2} = 1.56t$$

$$\bar{J} = 1.25 \text{ kgm}^2 = J_m + \frac{k_g^2}{k_g} (J_r + J_{fw})$$

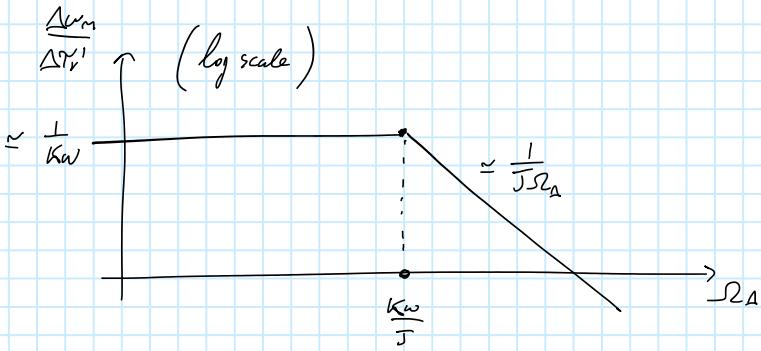
\uparrow added from before

$$J_{fw} = \frac{\bar{J} - J_m}{\frac{k_g^2}{k_g}} - J_r = 12,75 \text{ kgm}^2$$

We cannot add the flywheel on the motor shaft \rightarrow the flywheel should be on the shaft that presents load variations, which is the load itself

$$\frac{\Delta\omega_m}{\Delta\tau'_r} = \frac{1}{\sqrt{\bar{J}^2 R_A^2 + k_w^2}} = \begin{cases} \approx \frac{1}{k} & \text{if } R_A \ll \frac{k_w}{\bar{J}} \\ \approx \frac{1}{\bar{J} R_A} & \text{if } R_A \gg \frac{k_w}{\bar{J}} \end{cases} \quad \begin{aligned} &\text{if } R_A \ll \frac{k_w}{\bar{J}} & (1) \\ &\text{if } R_A \gg \frac{k_w}{\bar{J}} & (2) \end{aligned}$$

- ① The motor limits the oscillations of the speed
 ② The inertia of the system limits the oscillations



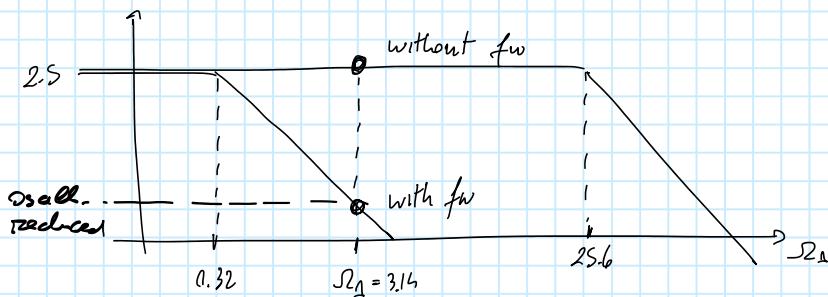
in the exercise:

$$\omega_{\Delta} = 3.14 \frac{\text{rad}}{\text{s}}$$

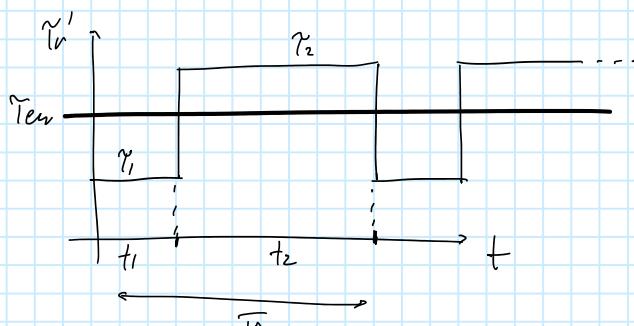
$$\frac{1}{Kw} = 2.5$$

$$\frac{Kw}{J} = \sqrt{25.6} \quad (\text{w/o fw}) \text{ rad/s}$$

$$\sqrt{J} = \sqrt{0.32} \quad (\text{w/ fw}) \text{ rad/s}$$



The inertia has the same effect of a low pass filter



$$\tau_r \text{ in } t_1$$

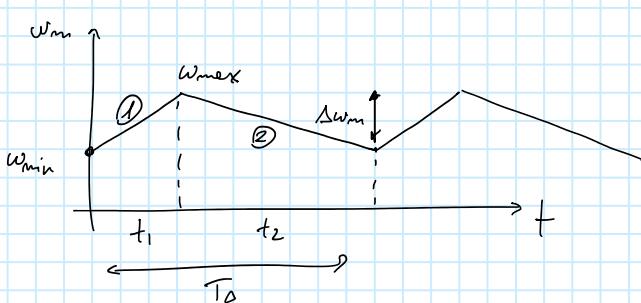
$$\tau_r \text{ in } t_2$$

$$t_1 + t_2 = T_{\Delta}$$

τ_{en} not changing with load

$\tau_{\text{en}} \rightarrow$ constant (simplification)

In reality, the motor helps in reducing the oscillations by changing its torque



$$\tau_{\text{en}} = \tau_r' + J \dot{\omega}_{\text{m}}$$

$$\tau_r' = \tau_r \kappa_{\text{gear}}$$

$$J = J_{\text{m}} + J_{\text{r}}' + J_{\text{fw}}$$

$$\dot{\omega}_{\text{m}} = \frac{\tau_{\text{en}} - \tau_r'}{J} \rightarrow \omega_{\text{m}} = \frac{\tau_{\text{en}} - \tau_r'}{J} \int dt$$

$$\textcircled{1}: \omega_{\text{min}} + \frac{\tau_{\text{en}} - \tau_r'}{J} \cdot t_1 = \omega_{\text{max}}$$

$$\textcircled{2}: \omega_{\text{max}} + \frac{\tau_{\text{en}} - \tau_r'}{J} \cdot t_2 = \omega_{\text{min}}$$

$$\Delta \omega = \omega_{\text{max}} - \omega_{\text{min}} = \frac{\tau_{\text{en}} - \tau_r'}{J} = \frac{-\tau_{\text{en}} + \tau_r'}{J} + t_2$$

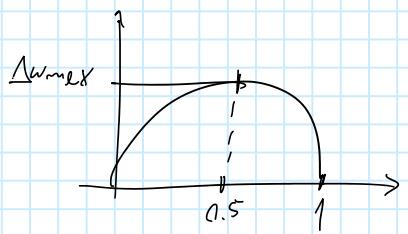
$$\Rightarrow \underbrace{(\tilde{\tau}_{\text{em}} - \tilde{\tau}_1) t_1 = (-\tilde{\tau}_{\text{em}} + \tilde{\tau}_2) t_2}_{\tilde{\tau}_{\text{em}} = \frac{\tilde{\tau}_1 t_1 + \tilde{\tau}_2 t_2}{t_1 + t_2}} \quad \text{torque average required from the motor}$$

$$\Delta \omega_m = \frac{\tilde{\tau}_{\text{em}} - \tilde{\tau}_1}{J} t_1 = \frac{\tilde{\tau}_1 t_1 + \tilde{\tau}_2 t_2 - \tilde{\tau}_1 T_A}{J} \cdot \frac{t_1}{T_A} =$$

$$\boxed{\Delta \omega_m = \frac{\tilde{\tau}_2 - \tilde{\tau}_1}{f_A J} \cdot \frac{t_1 + t_2}{T_A^2}}$$

$$\begin{cases} \frac{t_1}{T_A} = D \\ \frac{t_2}{T_A} = 1-D \end{cases} \rightarrow D(1-D) \quad (\text{Duty cycle})$$

similar to the oscillation of the current in a buck converter



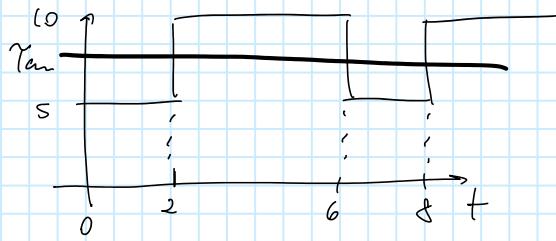
$$\Delta \omega_{\text{max}} = \frac{\tilde{\tau}_2 - \tilde{\tau}_1}{4 f_A J} \quad \text{for } D = 0.5$$

$$\boxed{\Delta \omega_m = \frac{\tilde{\tau}_2 - \tilde{\tau}_1}{J} \cdot \frac{D(1-D)}{f_A}}$$

$$J = \frac{\Delta \tilde{\tau}'_r}{\Delta \omega_m} = \frac{D(1-D)}{f_A}$$

$$J_{\text{max}} = \frac{\Delta \tilde{\tau}'_r}{\Delta \omega_m} \cdot \frac{1}{4 f_A}$$

Exercise



$$\Delta \omega_m \leq 11.1 \text{ rad/s}$$

$$\tilde{\tau}_1 = 5 \text{ Nm}, \quad t_1 = 2s$$

$$\tilde{\tau}_2 = 10 \text{ Nm}, \quad t_2 = 4s$$

$$\frac{1}{T_A} = 6s \quad \rightarrow \quad D = \frac{t_1}{T_A} = \frac{2}{6} = \frac{1}{3}$$

$$f_A = \frac{1}{6} \text{ Hz}$$

$$\tilde{\tau}_{\text{em}} = D \tilde{\tau}_1 + (1-D) \tilde{\tau}_2 = \frac{1}{3} \cdot 5 + \frac{2}{3} \cdot 10 = \frac{25}{3} = 8.33 \text{ Nm}$$

$$\Delta \tilde{\tau}'_r = |5 - 10| = 5 \text{ Nm}$$

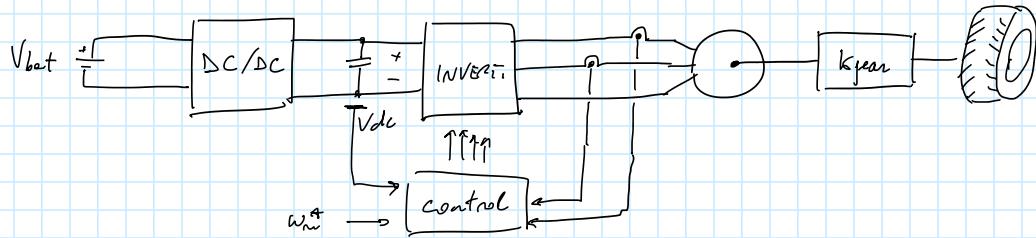
$$J = \frac{\Delta \tilde{\tau}'_r}{\Delta \omega_m} \cdot \frac{D(1-D)}{f_A} = \frac{5}{11.1} \cdot \frac{\frac{1}{3} \left(\frac{2}{3} \right)}{\frac{1}{6}} = 0.6 \text{ kgm}^2$$

$$J_{\text{max}} = \frac{\Delta \tilde{\tau}'_r}{\Delta \omega_m} \cdot \frac{1}{4 f_A} = 0.68 \text{ kgm}^2 \rightarrow \underline{\text{worst case}}$$

Esempio

lunedì 21 dicembre 2020 15:03

Electric car : battery, PE, motor, gearbox



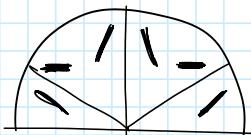
$$V_{bat} = 96V$$

$$V_{dc} = 400V$$

$$\text{max speed} = 150 \text{ km/h} \rightarrow 42 \text{ m/s}$$

$$\text{wheel diameter} = 0.6m \rightarrow \omega_{\text{wheel}} = \frac{\pi}{r} = \frac{42}{0.2} = 210 \text{ rad/s}$$

MOTOR (IPM)



$$V_{shape} = 2p = 8 \quad (\text{as in picture})$$

$$I_m = 21V_s$$

$$\left. \begin{aligned} L_d &= 0.32 \text{ mH} \\ L_q &= 0.64 \text{ mH} \end{aligned} \right\} \zeta = 2$$

$$I_N = 150 \text{ A}$$

$$I_p = 200 \text{ A}$$

$$\omega_{m_{\text{min}}} = 2000 \text{ rad/s}$$

REQUIREMENTS

① high slope

$$\tau_L = 600 \text{ Nm}$$

$$\omega_L = 30 \text{ rad/s}$$

② high speed operation

$$\tau_L = 120 \text{ Nm}$$

$$\omega_L = 200 \text{ rad/s}$$

DC/DC is a boost converter

$$\begin{aligned} V_N &= \frac{V_{dc}}{\sqrt{3}} \\ &= 83.0 \text{ V} \end{aligned}$$

V_{dc} = phase to phase peak voltage

Nominal current (nTPA)

$$\cos \alpha_i = \frac{-I_m \pm \sqrt{I_m^2 + 8(L_d - L_q)^2 I_N^2}}{4(L_d - L_q) I_N} = -0.357$$

$$I_{dN} = -53.6 \text{ A}$$

$$\lambda_d = \lambda_m + L_d I_{dN} = 0.0828 \text{ Vs}$$

$$I_{qN} = 140.1 \text{ A}$$

$$\lambda_q = L_q I_{qN} = 0.0897 \text{ Vs}$$

$$\left. \begin{array}{l} \lambda = 0.122 \text{ Vs} \\ \end{array} \right\}$$

• $T_{em} = \frac{3}{2} p [\lambda_d I_q - \lambda_q I_d] = 98.45 \text{ Nm}$

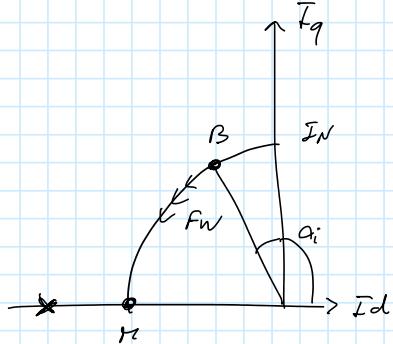
• $\omega_B = \sqrt{\frac{V_N}{\lambda}} = \frac{230}{0.122} = 1884 \text{ rad/s} \rightarrow \omega_{B_m} = \frac{\omega_B}{p} = 671 \text{ rad/s}$

• $P_N = T_{em} \omega_{B_m} = 98.45 \times 671 = 66370 \text{ W}$

Above nominal speed

$$\lambda_m = 0.1$$

$$L_d I_N = 0.00032 \cdot 150 = 0.048$$



$\lambda_m > L_d I_N$ outside the circle

• maximum speed: $\omega_m = \frac{V_N}{\lambda_m - L_d I_N} = 1313 \text{ rad/s} \quad \omega_{mp} = \frac{\omega_m}{p} = 1106 \text{ rad/s}$

Same considerations for the peak values

$$\cos \alpha_i' = -0.693$$

$$\rightarrow T_{emp} = 230.7 \text{ Nm}$$

$$\alpha_i' = 119.5^\circ$$

$$\omega_{Bp} = 1313 \text{ rad/s} \rightarrow \omega_{mp} = 328 \text{ rad/s}$$

$$I_d = -148 \text{ A}$$

$$P_{peak} = T_{emp} \omega_{mp} = 75750 \text{ W}$$

$$I_q = 261 \text{ A}$$

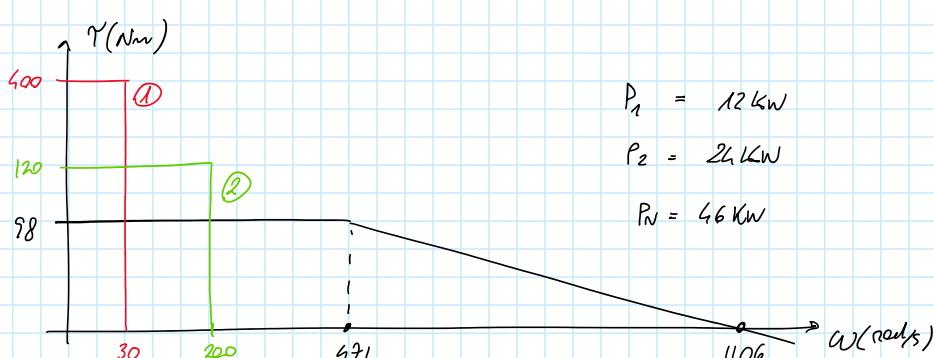
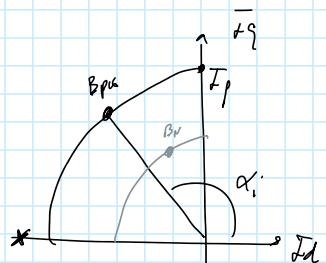
$$\lambda_d = 0.0526 \text{ Vs}$$

$$\lambda_m > L_d I_N$$

$$\lambda_q = 0.167 \text{ Vs}$$

outside as before

$$\omega_m = \dots$$



$$P_1 = 12 \text{ kW}$$

$$\eta_{peak} = 80\%$$

$$P_2 = 24 \text{ kW}$$

$$P_N = 66 \text{ kW}$$

gearbox ①

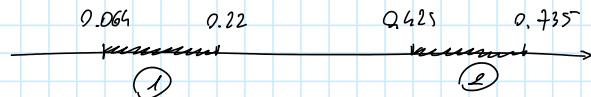
$$k'_{year} = \frac{20}{471} = 0.064$$

$$k''_{year} = \frac{289}{400} = 0.722$$

②

$$k'_{year} = \frac{200}{471} = 0.425$$

$$k''_{year} = \frac{269}{120} = 0.735$$



the second load could operate in flux weakening:

$$k_{year} = 0.2 \rightarrow T_{em(2)} = k_{year} \frac{\gamma_i}{\gamma_{year}} = 0.2 \frac{120}{0.9} = 26.7 \text{ Nm}$$

$$\omega_m(2) = \frac{\omega_i}{k_{year}} = \frac{200}{0.2} = 1000 \text{ rad/s}$$

$$T_{em} = \frac{3}{2} p \left[I_m I_N \sin \alpha_i + (L_d - L_q) I_N^2 \sin \alpha_i \cos \alpha_i \right] = 20 \sin \alpha_i - 43.2 \sin \alpha_i \cos \alpha_i \approx 26.7 \text{ Nm}$$

$$\Rightarrow \text{iteration} \rightarrow \boxed{\alpha_i = 168^\circ} \quad \boxed{T_{em} = 27.5 \text{ Nm}}$$

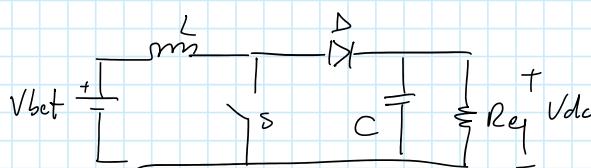
$$Id = \quad \omega =$$

$$I_q = \quad \omega_m = 1013 \text{ rad/s} \quad \boxed{1013}$$

$$1d =$$

$$1q =$$

BOOST CONVERTER



$$V_o = V_{dc}$$

$$P_o = P_{peak} = 75 \text{ kW}$$

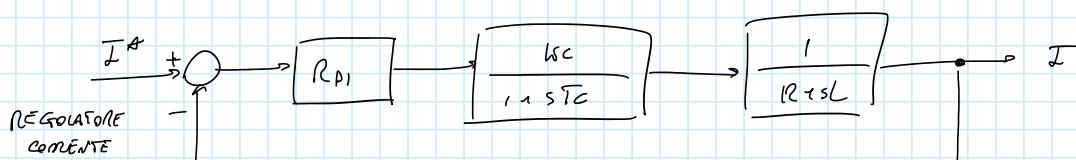
$$I_o = \frac{P_o}{V_o} = 180 \text{ A}$$

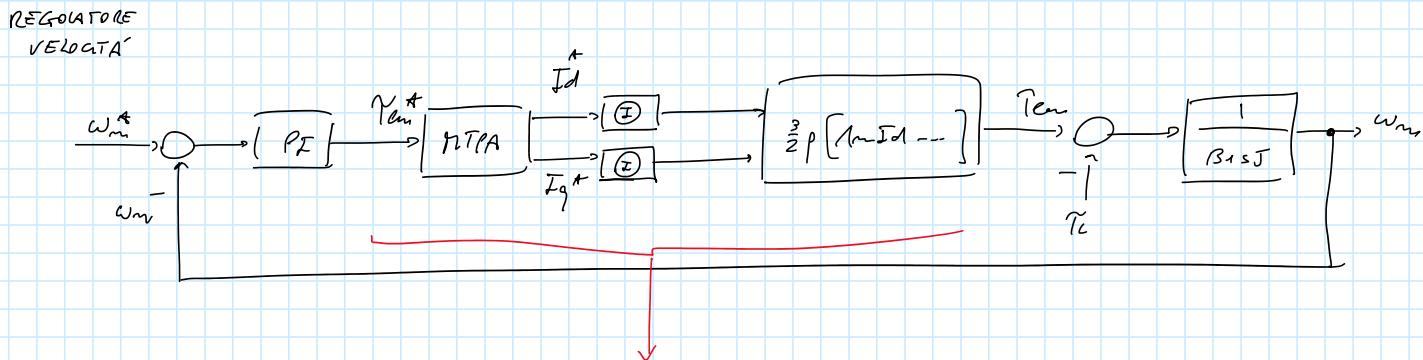
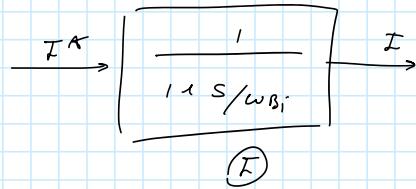
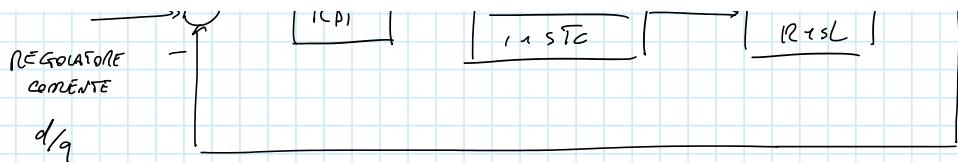
$$R_q = \frac{V_o}{I_o} = 2.11 \Omega$$

$$M = \frac{V_{dc}}{V_{betz}}$$

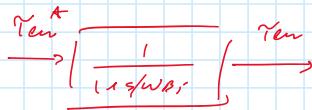
$$\eta = \frac{1}{1 - \Delta}$$

Control loop Regulation



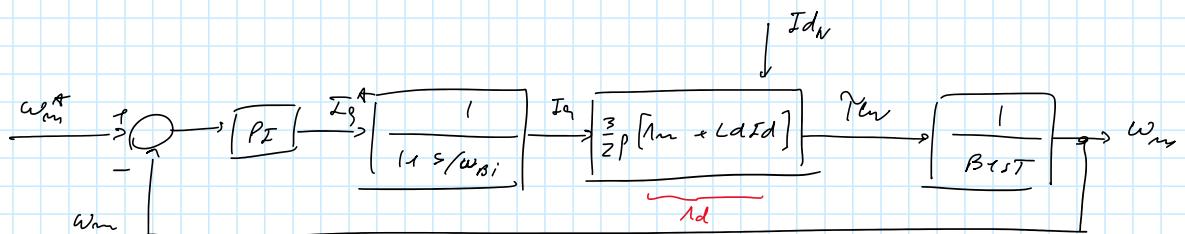


Soluzione 1 :



unico blocco con le dinamiche delle correnti

Soluzione 2 :



Davide Fogagnolo 1233994, Laurea Magistrale, progetto Arduino (drone ?) durante la prima sessione.

martedì 22 dicembre 2020 08:18

$$\underline{\text{MOTOR}} : L_d \neq L_q \Rightarrow \underline{\text{IPM MOTOR}}$$

$$2p = 8$$

$$J_m = 0.035 \text{ kgm}^2$$

$$V_{dc} = 200V = V_{ph-ph, ph}$$

$$\lambda_m = 0.085 \text{ Vs}$$

$$b = 10^{-3} \text{ Nms}$$

$$V_N = \frac{V_{dc}}{\sqrt{3}} = 115.5 \text{ V}$$

$$L_d = 5 \mu\text{H}$$

$$\omega_{m,lim} = 1500 \text{ rad/s}$$

$$L_q = 20 \mu\text{H}$$

$$I_N = 15 \text{ A}$$

$$R \approx 0.5 \Omega$$

$$I_p = 30 \text{ A}$$

C) AC MOTOR

$$\cos \alpha_i = \frac{\lambda_m \mp \sqrt{\lambda_m^2 + 8(L_d - L_q)^2 I_N^2}}{4(L_d - L_q) I_N} = -0.6189$$

NOMINAL

$$\alpha_i = 128.24^\circ$$

$$\begin{cases} I_{dN} = I_N \cos \alpha_i = -9.28 \text{ A} \\ I_{qN} = I_N \sin \alpha_i = 11.78 \text{ A} \end{cases}$$

$$\begin{cases} \lambda_{dN} = \lambda_m + L_d I_{dN} = 0.0386 \text{ Vs} \\ \lambda_{qN} = L_q I_{qN} = 0.2356 \text{ Vs} \end{cases}$$

$$\lambda_N = \sqrt{\lambda_{dN}^2 + \lambda_{qN}^2} = 0.2388 \text{ Vs}$$

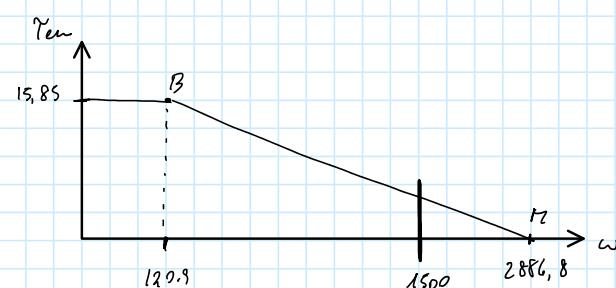
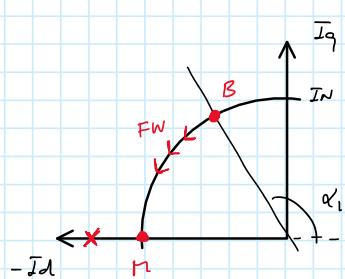
$$T_{emN} = \frac{3}{2} p \left[\lambda_d I_{qN} - \lambda_q I_{dN} \right] = 15.85 \text{ Nm}$$

$$\omega_{BN} = \frac{V_N}{\lambda_N} = 483.6 \text{ rad/s} \rightarrow \omega_{mBN} = \frac{\omega_{BN}}{p} = 120.9 \text{ rad/s}$$

$$P_{mN} = T_{emN} \cdot \omega_{mBN} = 1916 \text{ W}$$

- Flux weakening: $\lambda_m > L_d I_N$
 $0.085 > 0.075$ ✓ outside the circle

$$\omega_{mN} = \frac{V_N}{\lambda_m - L_d I_N} = 11547 \frac{\text{rad}}{\text{s}} \rightarrow \omega_{m_{mN}} = \frac{\omega_{mN}}{p} = 2886.8 \text{ rad/s}$$



PEAK

$$\cos \alpha_i = -0.6615 \rightarrow \alpha_i = 131,4^\circ$$

$$\begin{cases} I_{dp} = -19.84 \text{ A} \\ I_{qp} = 22.5 \text{ A} \end{cases} \quad \begin{cases} \lambda_{dp} = -0.0142 \text{ Vs} \\ \lambda_{qp} = 0.49 \text{ Vs} \end{cases} \rightarrow \lambda_p = 0.4502 \text{ Vs}$$

$$T_{em,p} = 51,65 \text{ Nm}$$

$$\omega_{Bp} = 256.7 \text{ rad/s} \rightarrow \omega_{m_{Bp}} = 64 \text{ rad/s}$$

$$P_{em,p} = T_{em,p} \cdot \omega_{m_{Bp}} = 3312 \text{ W}$$

• Flux weakening: $\lambda_m > Ld \bar{I}_p$
 $0.085 > 0.15 \quad \text{X} \quad \text{inside the circle}$

$$\xi = \frac{Lq}{Ld} = 4$$

$$\boxed{\text{MTPV}} \rightarrow |I_{dp}| = \xi |I_{qp}|$$

$$\tan \alpha_F = -\frac{1}{\xi} = -0.25 \rightarrow \alpha_F = 166^\circ$$

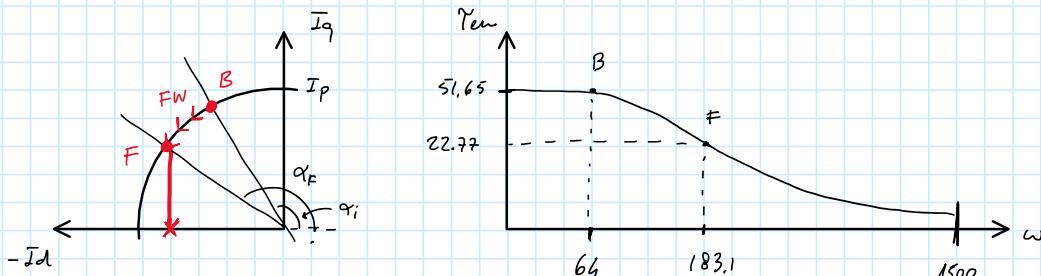
$$\begin{cases} I_{dp} = I_p \cdot \cos \alpha_F = -29,10 \text{ A} \\ I_{qp} = I_p \cdot \sin \alpha_F = 7,27 \text{ A} \end{cases} \quad \begin{cases} \lambda_{dp} = \lambda_m + Ld I_{dp} = -0.06 \text{ Vs} \\ \lambda_{qp} = Lq I_{qp} = 0.14 \sqrt{5} \text{ Vs} \end{cases}$$

$$\lambda_F = 0.1576 \text{ Vs}$$

$$T_{em,F} = 22.77 \text{ Nm}$$

$$\omega_F = \frac{V_n}{\lambda_F} = 732.6 \text{ rad/s} \rightarrow \omega_{m_F} = 183.1 \text{ rad/s}$$

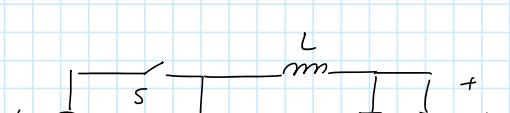
$$P_{em,F} = T_{em,F} \cdot \omega_{m_F} = 6170 \text{ W}$$



A DC-DC converter

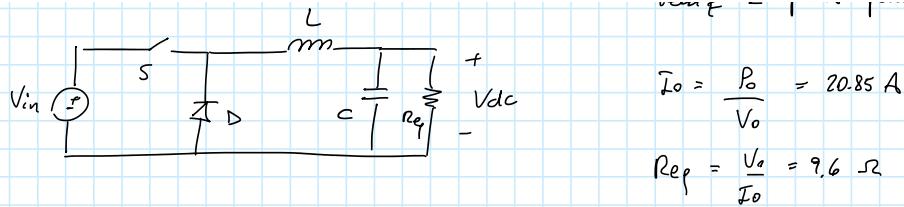
$$V_{in} > V_{out} \rightarrow \underline{\text{buck}}$$

$$\sqrt{f} = V_{in} \quad V_o = V_{dc}$$



$$P_{em,F} = \text{peak power}$$

$$I_o = \frac{P_o}{V_o} = 20.85 \text{ A}$$



$$\eta = D = \frac{V_o}{V_f} = 0.625$$

$$\Delta i_c = 2\% I_L = 2 \times I_o = 417 \text{ mA} \Rightarrow L = \frac{V_f}{f_s \Delta i_c} D(1-D) = 30 \text{ mH}$$

$$\Delta v_{out} = 1\% V_o = 2 \text{ V} \Rightarrow C = \frac{\Delta i_c}{\delta f_s \Delta v_o} = 4.34 \mu\text{F}$$

$$I_{sw} = D I_L = 13.03 \text{ A} \quad I_{sw,pu} = I_o + \frac{\Delta i_c}{2} = 21.06 \text{ A}$$

$$I_o = (1-D) I_L = 7.8187 \text{ A} \quad I_{o,pu} = I_{sw,pu}$$

Maximum reverse voltage: $V_f = 320 \text{ V}$

D STATIC LOAD

$$\tau_i = 500 \text{ Nm} \quad \omega_{mN} = 120.9 \text{ rad/s}, \quad \tau_{em,N} = 15.85 \text{ Nm}$$

$$\omega_r = 2.5 \text{ rad/s} \quad \omega_{mp} = 64 \text{ rad/s}, \quad \tau_{em,p} = 51.65 \text{ Nm}$$

$$\kappa_{gear} = 80 \%$$

$$\tilde{\tau}_r = \frac{\tau_i}{\kappa_{gear}} = 555.55 \text{ Nm}$$

$$k_{min} = \frac{\omega_r}{\omega_N} = 0.02067$$

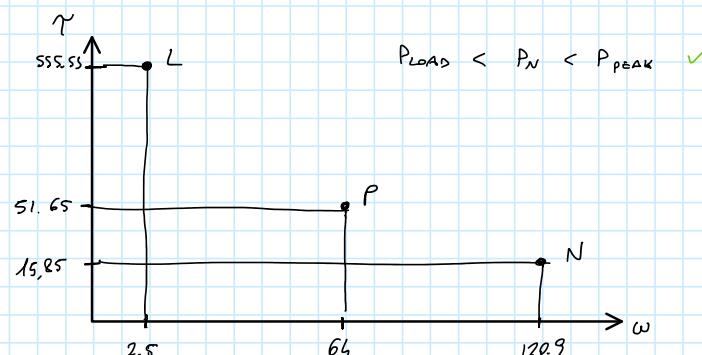
$$k_{max} = \frac{\tilde{\tau}_r}{\tau_r} = 0.0285$$

$$k_{min} \leq \kappa_{gear} \leq k_{max}$$

$$\bullet \quad \kappa_{gear} = \frac{1}{40} = 0.025$$

$$\Rightarrow \omega_m = \frac{\omega_r}{\kappa_{gear}} = 100 \text{ rad/s}$$

$$\Rightarrow \tau_m = \kappa_{gear} \tau_r = 13.89 \text{ Nm}$$



$$\tau_{em} = \frac{3}{2} p [1_{mN} I_N \sin \alpha_i + (L_d - L_g) I_N^2 \sin \alpha_i \cos \alpha_i]$$

$$= \frac{3}{2} \cdot 4 [0.085 \cdot 15 \sin \alpha_i + (0.005 - 0.02) 15^2 \sin \alpha_i \cos \alpha_i]$$

$$= 7.65 \sin \alpha_i - 20.25 \sin \alpha_i \cos \alpha_i = \boxed{13.89 \text{ Nm}}$$

α_i	130°	120°	110°	111°
γ_{ew}	15.83	15.33	13.69	13.91

$$\left\{ \begin{array}{l} I_d = T_N \cos \alpha_i = -5.37 \text{ A} \\ I_q = T_N \sin \alpha_i = 14 \text{ A} \end{array} \right. \rightarrow I = 15 e^{j111^\circ}$$

$$\left\{ \begin{array}{l} \lambda_d = \lambda_m + L_d I_d = 0.05815 \text{ Vs} \\ \lambda_q = L_q I_q = 0.280 \text{ Vs} \end{array} \right. \rightarrow \lambda = 0.286 \text{ Vs}$$

$$V = \omega_m \lambda = 28.6 \text{ V}$$

(E) DYNAMIC LOAD

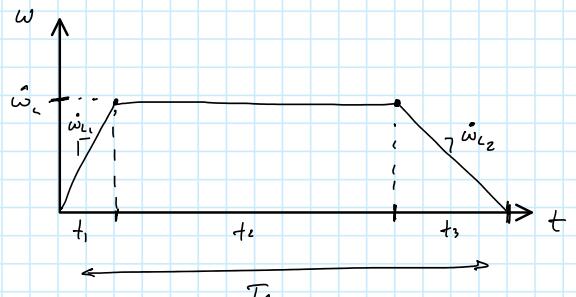
$$J_L = 0.45 \text{ kgm}^2$$

$$\dot{\omega}_{L_1} = 150 \text{ rad/s}^2, \quad t_1 = 1s$$

$$\dot{\omega}_L = \text{const}, \quad t_2 = 9s$$

$$\dot{\omega}_{L_2} = ?, \quad t_3 = 25s$$

$$T_A = \lambda \cdot 9 + 2 = 12s$$



$$\hat{\omega}_L = \dot{\omega}_{L_1} \cdot t_1 = 150 \text{ rad/s}$$

$$\hat{\omega}_L - t_2 \dot{\omega}_{L_2} = 0$$

$$\hookrightarrow \dot{\omega}_{L_2} = \frac{\hat{\omega}_L}{t_2} = 75 \text{ rad/s}^2$$

$$\bullet k_{opt} = \sqrt{\frac{J_m}{J_L}} = 0.2789$$

$$\dot{\omega}_L \text{ rms} = \sqrt{\frac{1}{T_A} (\dot{\omega}_{L_1}^2 t_1 + \dot{\omega}_{L_2}^2 t_2)} = 53 \text{ rad/s}$$

$$\dot{\omega}_L \text{ max} = \dot{\omega}_{L_1} = 150 \text{ rad/s}^2$$

$$\left(T_p = 51.65 \text{ Nm}, \quad \gamma_N = 15.85 \text{ Nmm} \right)$$

$$\bullet k_{p_{12}} = \frac{T_p \pm \sqrt{T_p^2 - 4 J_m J_L \dot{\omega}_{L \text{ max}}}}{2 \dot{\omega}_{L \text{ max}}} = \begin{cases} 0.344 \\ 0.0003 \end{cases}$$

$$\bullet k_{N_{12}} = \frac{\gamma_N \pm \sqrt{\gamma_N^2 - 4 J_m J_L \dot{\omega}_{L \text{ rms}}}}{2 \dot{\omega}_{L \text{ rms}}} = \begin{cases} 0.298 \\ \end{cases}$$

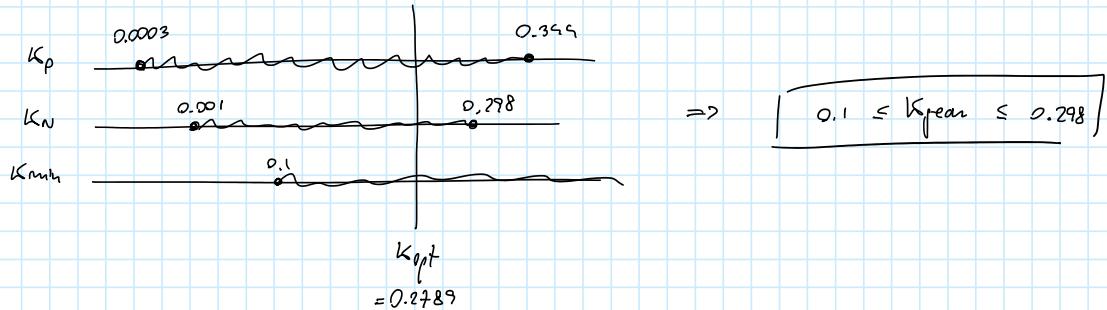
$$\bullet K_{N12} = \frac{\tilde{T}_N \pm \sqrt{\tilde{T}_N^2 - 4J_L J_M \tilde{\omega}_{rms}}}{2 \cdot \tilde{\omega}_{rms}} = \begin{cases} 0.298 \\ 0.001 \end{cases}$$

$$\bullet K_{min} = \frac{\omega_{Lmax}}{\omega_{max}} = \frac{150}{1500} = 0.1$$

$$\omega_{max} \geq \frac{\omega_{Lmax}}{K_{opt}} \quad \checkmark$$

$$\tilde{T}_P \geq 2\sqrt{J_L J_M} \tilde{\omega}_{rms} \quad \checkmark$$

$$\tilde{T}_N \geq 2\sqrt{J_L J_M} \tilde{\omega}_{rms} \quad \checkmark$$



$$\tilde{T}_{Lrms} = J_L \cdot \tilde{\omega}_{Lrms} = 23.85 \text{ Nm}$$

$$\rightarrow T_{em,rms} = \frac{\tilde{T}_{Lrms} \cdot K_{year}}{\eta_{year}}$$

I can use the optimal K_{year} since it is inside the range:

$$\bullet T_{em,rms}(K_{opt}) = \frac{\tilde{T}_{Lrms}}{\eta_{year}} K_{opt} = 7.39 \text{ Nm}$$

F Design of the regulators

CURRENT LOOP

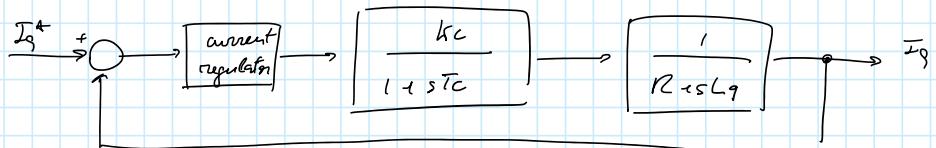
$$K_C = 20$$

$$T_C = 0.5 \text{ ms}$$

$$K_{f/b} = 1$$

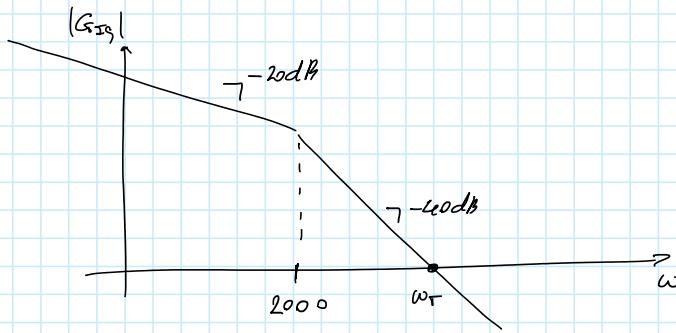
$$\omega_B = 250 \text{ rad/s}$$

$$\varphi_m = 60^\circ$$



- Since $R \approx 0$: $\frac{1}{R+sL} \approx \frac{1}{sL}$ → No need for a PI regulator → only proportional. There is already a pole in the origin for zero static error of the loop

$$\frac{K_C}{1+sT_C} \cdot \frac{1}{sLq} = \frac{20}{1 + s \cdot 0,0005} \cdot \frac{1}{s \cdot 0,02} = \frac{1000}{s(1 + \frac{s}{2000})}$$



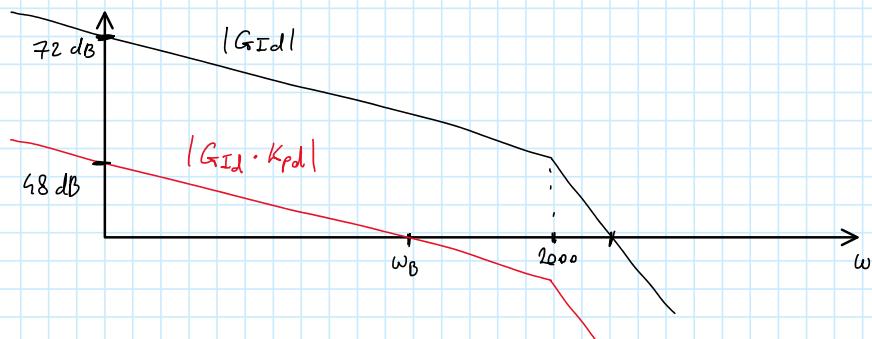
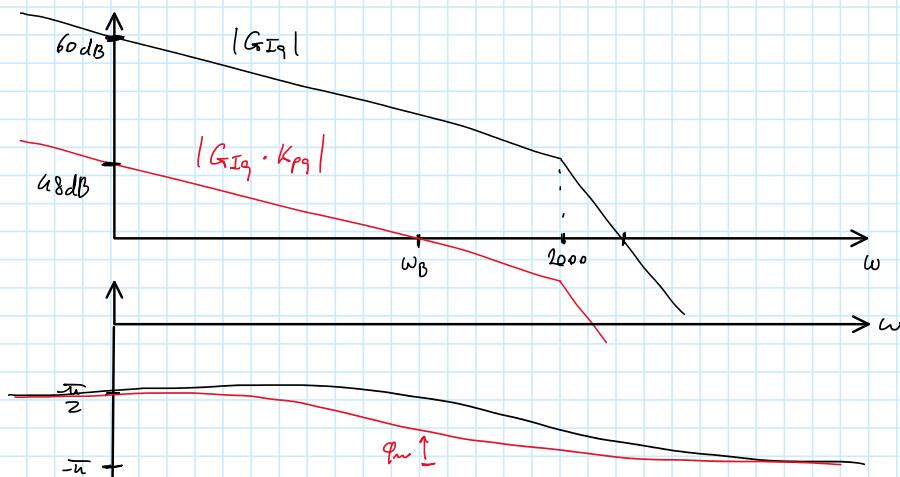
$$\varphi_m = 180 + \angle G_{Iq} = 180 - 90 - \text{and } \frac{250}{2000} = 82.875^\circ > 60^\circ \quad \checkmark$$

- $|G_{Iq}(\omega_B)| = 1$

$$\Rightarrow \left| K_{Pq} \cdot \frac{K_C}{1+sT_C} \cdot \frac{1}{sLq} \right| = 1 \quad \xrightarrow[s=j\omega_B]{} \boxed{K_{Pq} = 0.2519}$$

- for Id:

$$\left| K_{Id} \cdot \frac{K_C}{1+sT_C} \cdot \frac{1}{sLd} \right| = 1 \quad \xrightarrow{} \boxed{K_{Id} = 0.063}$$

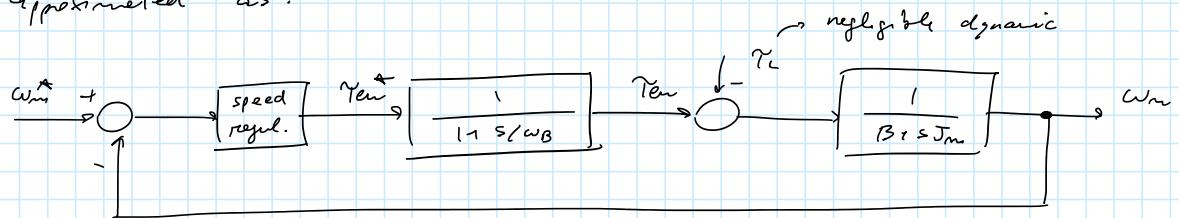


SPEED LOOP CONTROL

The whole current loop can be approximated as

$$I_{q/d}^* \xrightarrow{\left[\frac{1}{1 + s/\omega_B} \right]} I_{q/d}$$

Since the torque has the same dynamic of the current, the speed loop can be approximated as:



In this case we need a PI regulator to have a pole in the origin -

$$\begin{aligned} G_{w_m} &= \frac{1}{1 + \frac{s}{\omega_B}} \cdot \frac{1}{B + sJ_m} = \frac{\frac{1}{B}}{\left(1 + \frac{s}{\omega_B}\right)\left(1 + s \frac{J_m}{B}\right)} = \\ &= \frac{1000}{\left(1 + \frac{s}{250}\right)\left(1 + s \cdot 35\right)} \end{aligned}$$

$$\varphi_m = 180 - \angle G_{w_m} = 85.5^\circ \quad \checkmark \quad \rightarrow \text{no need to compensate but just a pole in the origin}$$

- PI regulator : compensate the first pole

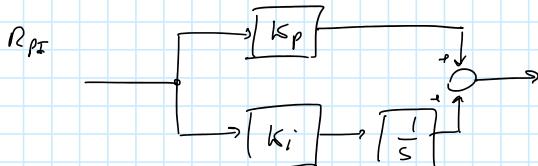
$$G_{p1} = K_i \cdot \frac{1 + sT_i}{s}$$

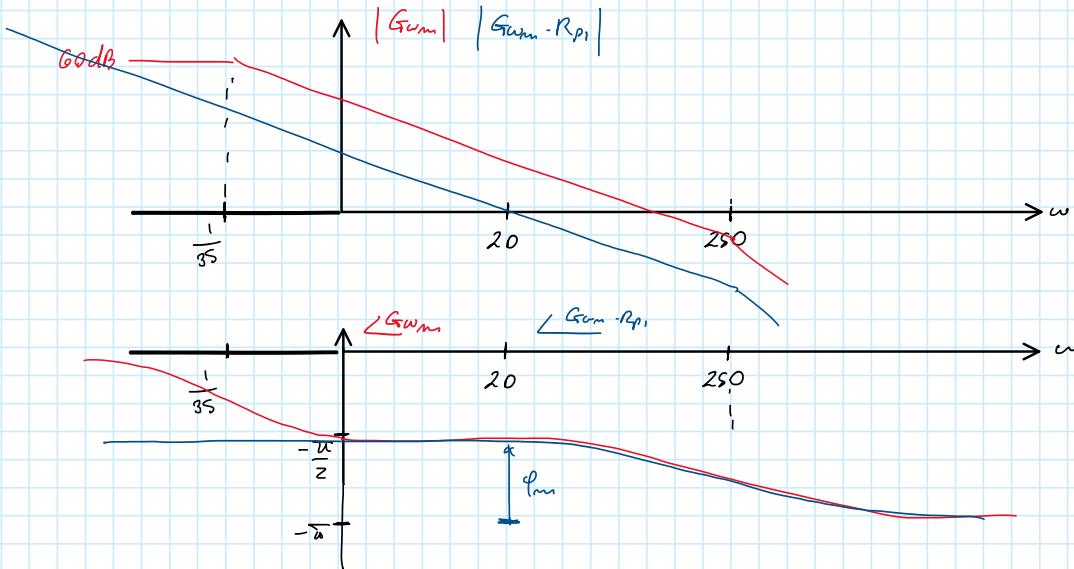
$$\boxed{T_i = \frac{J}{B} = 3s}$$

$$\varphi_m' = 180 - \angle G_{w_m} \cdot G_{p1} = 180 - 90 - \arctan\left(\frac{20}{250}\right) = 85.5^\circ \quad \checkmark$$

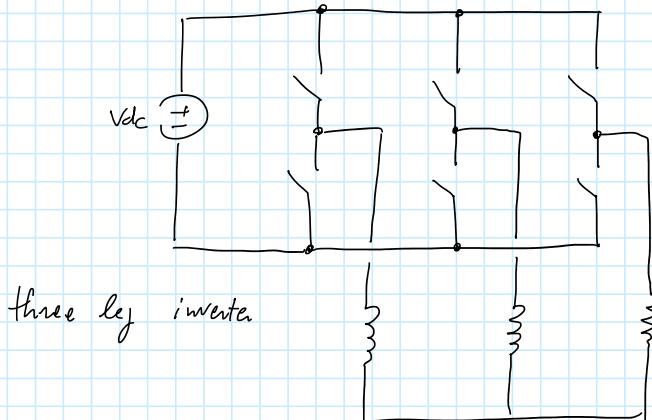
$$K_i : \quad \left| K_i \cdot \frac{1 + sT_i}{s} \cdot \frac{1}{\left(1 + \frac{s}{\omega_B}\right)\left(1 + s \frac{J_m}{B}\right)} \right| = 1$$

$$\boxed{\begin{aligned} K_i &= 0.02 \\ K_p &= K_i \cdot T_i = 0.02 \cdot 3s = 0.7 \end{aligned}}$$





③ Inverter design

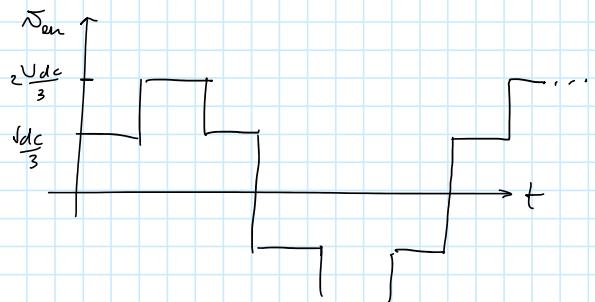
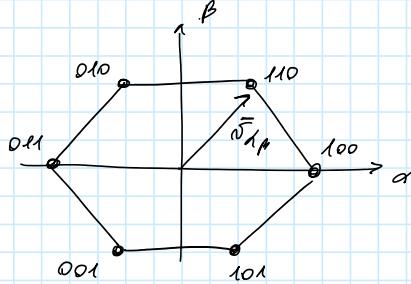


$$\begin{aligned} \text{peak current} &= I_{\text{dc, branch}} + A_{\text{ic, branch}} \\ &= 21.06 \text{ A} \end{aligned}$$

$$\text{maximum voltage} = V_{\text{dc}} = 200 \text{ V}$$

$$\bar{\alpha}_{abc} = \frac{2}{3} \left[\alpha_a + \alpha_b e^{j\frac{2}{3}\pi} + \alpha_c e^{-j\frac{2}{3}\pi} \right]$$

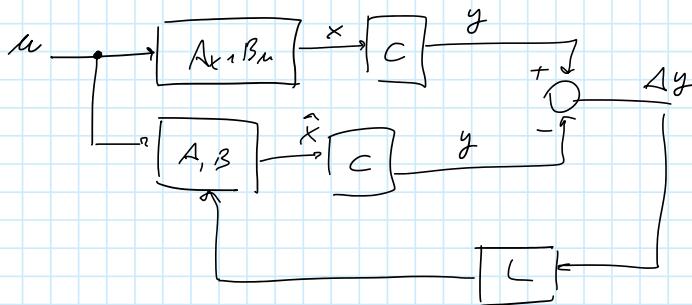
$$\begin{bmatrix} \bar{\alpha}_a \\ \bar{\alpha}_b \\ \bar{\alpha}_c \end{bmatrix} = \begin{bmatrix} \frac{2}{3} & -\frac{1}{3} & -\frac{1}{3} \\ 0 & \frac{1}{\sqrt{3}} & -\frac{1}{\sqrt{3}} \\ \frac{1}{3} & \frac{1}{3} & \frac{1}{3} \end{bmatrix} \begin{bmatrix} \alpha_a \\ \alpha_b \\ \alpha_c \end{bmatrix}$$



every state has a binary code

(G) sensorless control

I can implement a second order observer to get both position and speed. This technique requires the motor to run at a speed sufficiently high to avoid errors in the measurements.



$$x = \begin{bmatrix} \dot{\theta}_m^e \\ \dot{\omega}_m^e \end{bmatrix} \quad \left\{ \begin{array}{l} \dot{\theta}_m^e = \omega_m^e \\ \dot{\omega}_m^e = p \frac{T_{em} - T_b}{J} = 0 \end{array} \right. \rightarrow \text{assume constant speed}$$

$$\begin{bmatrix} \dot{\hat{\theta}}_m^e \\ \dot{\hat{\omega}}_m^e \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{\theta}_m^e \\ \hat{\omega}_m^e \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \end{bmatrix} u + \begin{bmatrix} k_1 \\ k_2 \end{bmatrix} (\theta_m^e - \hat{\theta}_m^e)$$

$$\dot{\hat{x}} = A \hat{x} + B u + L \Delta \theta$$

$$y = \hat{\theta}_m^e = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} \hat{\theta}_m^e \\ \hat{\omega}_m^e \end{bmatrix}$$

$$\left\{ \begin{array}{l} \dot{x} = A_x + B_m \\ \dot{\hat{x}} = A \hat{x} + B_m - LC(x - \hat{x}) \end{array} \right.$$

$$\dot{x}_{em} = Ax_{em} - LCx_{em} = (A - LC)x_{em}$$

$$\det [sI - (A - LC)] = 0$$

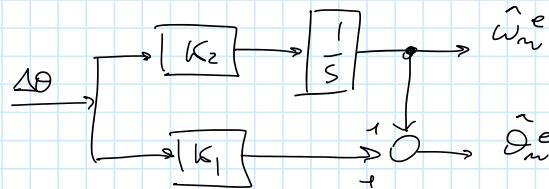
$$\Rightarrow s^2 + k_1 s + k_2 = 0$$

From specification ω_B speed = 20 rad/s

I want two coincident poles :

$$(s + 20)^2 = s^2 + 40s + 400$$

$K_1 = 40$
$K_2 = 400$



- $\bar{J}_{m\alpha\beta} = J_m e^{j\delta_m^e}$

