

# Modern Control Theory (ICE 3153)

Transfer Function to State Space

State space representation using phase variable &

Canonical Forms

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The phase variables are defined as those <u>particular state variables which</u> are obtained from one of the system variables and its derivatives.

Usually the <u>variable used is the system output</u> and the remaining state variables are then <u>derivatives of output</u>.

The state model using phase variable can be easily determined if the system model is already known in the differential equation or transfer function form.

#### <u>Advantages of Phase Variables</u>:

The state space model can be directly formed by inspection from the differential equations governing the system.

#### <u>Disadvantages of Phase Variables</u>:

The phase variables are not physical variables of the system and are therefore not available for measurement and control purposes.

#### **Case 1**:

Consider the differential equation

$$y^{(n)} + a_1 y^{(n-1)} + \dots + a_{n-1} y + a_n y = u$$

Choosing the output, y(t), and its (n - 1) derivatives as the state variables. This choice is called the *phase-variable choice* 

$$x_1 = y$$

$$x_2 = \dot{y}$$

$$\vdots$$

$$x_n = \begin{pmatrix} x_{n-1} \\ y \end{pmatrix}$$

The state equations are evaluated as:

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = x_3$$

$$\vdots$$

$$\dot{x}_{n-1} = x_n$$

$$\dot{x}_n = a_n x_1 - \dots - a_1 x_n + u$$

The phase-variable form of the state equations:

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}u$$

$$\begin{bmatrix} \dot{x_1} \\ \dot{x_2} \\ \vdots \\ \dot{x_{n-1}} \\ \dot{x_n} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & 1 \\ -a_n & -a_{n-1} & -a_{n-2} & \dots & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix} u$$

The solution to the differential equation is y(t), or  $x_1$  and the output equation is:

$$y = \begin{bmatrix} 1 & 0 & 0 & \cdots & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} \frac{Y(s)}{U(s)} = \frac{1}{s^n + a_1 s^{n-1} + \cdots + a_{n-1} s + a_n}$$

#### Question 1:

#### MCS, 12th Edition, Dorf and Bishop

Consider a fourth-order transfer function G(s) given by:

$$G(s) = \frac{Y(s)}{U(s)} = \frac{b_0}{s^4 + a_3 s^3 + a_2 s^2 + a_1 s + a_0}$$

Obtain the state space representation. Draw the signal flow graph and block diagram representation.

From Mason's signal-flow gain formula

$$G(s) = \frac{Y(s)}{U(s)} = \frac{\sum_{k} P_k \Delta_k}{\Delta}.$$

$$G(s) = \frac{\sum_{k} P_{k}}{1 - \sum_{q=1}^{N} L_{q}} = \frac{\text{Sum of the forward-path factors}}{1 - \text{sum of the feedback loop factors}}.$$

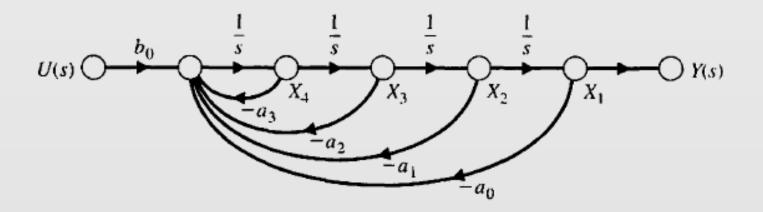
Where  $P_K = path$  gain of  $K^{th}$  forward path.

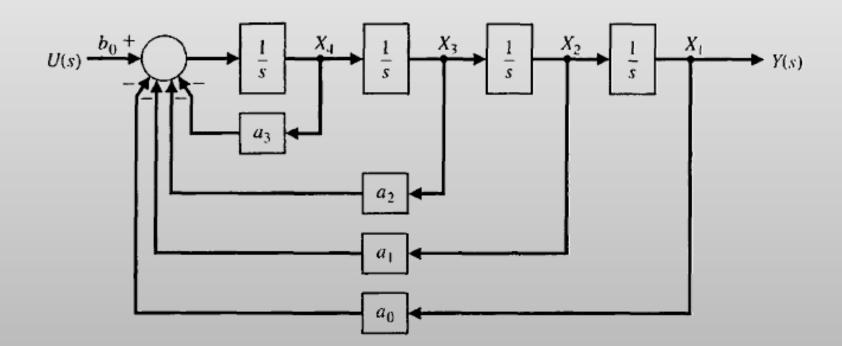
 $\Delta = 1 - (\text{sum of loop gain of all individual loops})$ 

+ (sum of gain products of all possible combinations of two non-touching loops) - ......

 $\Delta_{K} = \Delta$  for that part of the graph which is not touching  $K^{th}$  forward path.

$$= \frac{b_0 s^{-4}}{1 + a_3 s^{-1} + a_2 s^{-2} + a_1 s^{-3} + a_0 s^{-4}}.$$

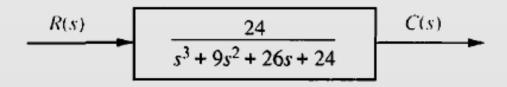




#### **Question 2:**

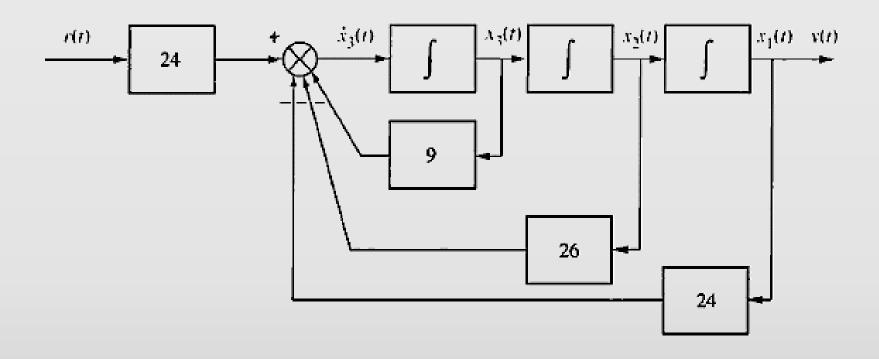
#### Ex 3.4, CSE – 6<sup>th</sup> Edition, Norman. S. Nise

Find the state-space representation in phase-variable form for the transfer function shown in figure.



$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -24 & -26 & -9 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 24 \end{bmatrix} r$$

$$y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$



#### **Case 2: Controllable Canonical Form**

Consider the differential equation

$$y + a_1 y + \cdots + a_{n-1} \dot{y} + a_n y = b_0 u + b_1 u + \cdots + b_{n-1} \dot{u} + b_n u$$

So the transfer function of system defined by

$$\frac{Y(s)}{U(s)} = \frac{b_0 s^n + b_1 s^{n-1} + \dots + b_{n-1} s + b_n}{s^n + a_1 s^{n-1} + \dots + a_{n-1} s + a_n}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_{n-1} \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \ddots & \ddots & \ddots & \vdots \\ 0 & 0 & 0 & \cdots & 1 \\ -a_n & -a_{n-1} & -a_{n-2} & \cdots & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} u$$

$$y = [b_n - a_n b_0 \mid b_{n-1} - a_{n-1} b_0 \mid \cdots \mid b_1 - a_1 b_0] \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + b_0 u$$

#### **Observable Canonical Form**

$$\begin{bmatrix} \dot{x_1} \\ \dot{x_2} \\ \vdots \\ \dot{x_{n-1}} \\ \dot{x_n} \end{bmatrix} = \begin{bmatrix} 0 & 0 & \dots & 0 & -a_n \\ 1 & 0 & \dots & 0 & -a_{n-1} \\ \vdots & \vdots & & \vdots & \vdots \\ 0 & 0 & \dots & 1 & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} + \begin{bmatrix} b_n - a_n b_0 \\ b_{n-1} - a_{n-1} b_0 \\ \vdots \\ b_1 - a_1 b_0 \end{bmatrix} u$$

$$y = \begin{bmatrix} 0 & 0 & \dots & 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} + b_0 u$$

$$A_{obs} = A_{cont}^{T}$$

$$B_{obs} = C_{cont}^{T}$$

$$C_{obs} = B_{cont}^{T}$$

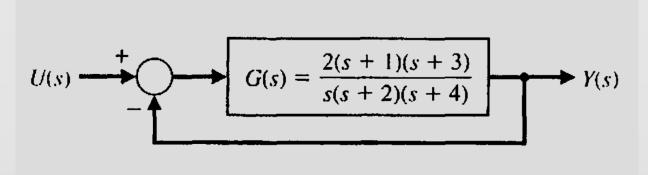
$$D_{obs} = D_{cont}$$

## Tutorial -2

#### **Question 1:**

#### Ex 3.2, MCS – 12<sup>th</sup> Edition, Dorf and Bishop

For the system shown in figure, derive the controllable canonical form and observable canonical form.

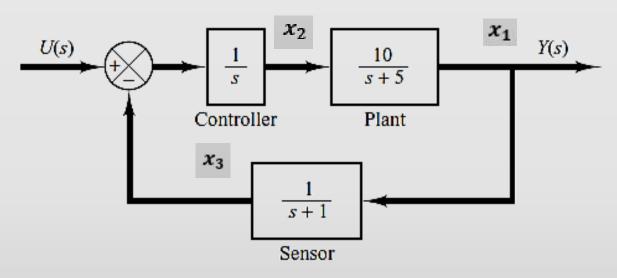


$$\frac{Y(s)}{U(s)} = \frac{2s^2 + 8s + 6}{s^3 + 8s^2 + 16s + 6}$$

#### **Question 2:**

#### A-3-8, MCE – 5<sup>th</sup> Edition, K. Ogata

Obtain a state-space model of the system shown in figure.



$$\frac{X_1(s)}{X_2(s)} = \frac{10}{s+5}$$

$$\frac{X_2(s)}{U(s) - X_3(s)} = \frac{1}{s}$$

$$\frac{X_2(s)}{X_1(s)} = \frac{1}{s}$$

$$\frac{X_2(s)}{X_2(s)} = \frac{1}{s}$$

$$\frac{X_2(s)}{X_1(s)} = \frac{1}{s}$$

$$\frac{X_2(s)}{X_1(s)} = \frac{1}{s}$$

$$\frac{X_2(s)}{X_2(s)} = -X_3(s) + U(s)$$

$$\frac{X_2(s)}{X_2(s)} = -X_3(s)$$

$$\frac{X_2(s)}{X_2(s)} = -X_3(s)$$

$$\frac{X_2(s)}{X_2(s)}$$

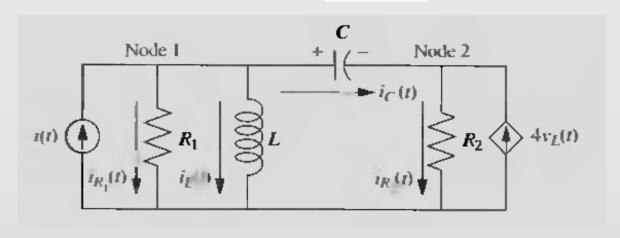
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -5 & 10 & 0 \\ 0 & 0 & -1 \\ 1 & 0 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

#### **Question 3:**

#### Ex 3.2, CSE – Norman S Nice

Find the state and output equations for the electrical network shown in the figure if the output variables are  $v_{R_2}$   $i_{R_2}$ 



$$L\frac{di_L}{dt} = v_L$$

$$C\frac{dv_C}{dt}=i_C$$

$$x_1 = i_L;$$

$$v_L = v_C + v_{R_2} = v_C + i_{R_2} R_2$$

$$i_{R_2} = i_C + 4v_L.$$

$$v_L = v_C + (i_C + 4v_L)R_2$$

$$v_L = \frac{1}{1 - 4R_2} (v_C + i_C R_2)$$

$$v_L = \frac{1}{\Delta} [R_2 i_L - v_C - R_2 i(t)]$$

$$i_C = i(t) - i_{R_1} - i_L$$

$$= i(t) - \frac{v_{R_1}}{R_1} - i_L$$

$$= i(t) - \frac{v_L}{R_1} - i_L$$

$$i_C = \frac{1}{\Delta} \left[ (1 - 4R_2)i_L + \frac{1}{R_1} v_C - (1 - 4R_2)i(t) \right]$$

$$\Delta = -\left[\left(1-4R_2\right) + \frac{R_2}{R_1}\right]$$

$$(1 - 4R_2)v_L - R_2i_C = v_C$$

$$-\frac{1}{R_1}v_L - i_C = i_L - i(t)$$

$$\begin{bmatrix} i_L \\ i_C \end{bmatrix} = \begin{bmatrix} R_2 & (L\Delta) & -1/(L\Delta) \\ (1 - 4R_2), & (C\Delta) & 1/(R_1C\Delta) \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix}$$

$$+ \begin{bmatrix} -R_2, & (L\Delta) \\ -(1 - 4R_2), & (C\Delta) \end{bmatrix} i(t)$$

$$v_{R_2} = -v_C + v_L$$
$$i_{R_2} = i_C + 4v_L$$

$$\begin{bmatrix} v_{R_2} \\ i_{R_2} \end{bmatrix} = \begin{bmatrix} R_{2/} \Delta & -(1+1/\Delta) \\ 1/\Delta & (1-4R_1) & (\Delta R_1) \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix} + \begin{bmatrix} -R_2/\Delta \\ -1/\Delta \end{bmatrix} i(t)$$

Direct, cascade, parallel realizations

### **Decomposition of Transfer Function**

- Process of going from TF to State Model is called decomposition
- Direct -> Phase variable form
- Parallel -> Canonical form
- Cascade